

# Report worksheet 2

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## Appendix A Fourier transform of a continuous periodic signal

We prove in Lemma 1 that the Fourier transform of a continuous periodic signal is a sum of scaled delta functions at multiples of the frequency of this signal.

**Definition 1** (continuous signal). *A continuous signal  $x(t)$  is periodic if and only if there exists a period  $T > 0$  such that*

$$x(t) = x(t + T), \forall t \in \mathbb{R} \quad (1)$$

**Lemma 1** (Fourier transform of a periodic signal). *If  $x(t)$  is periodic, with period  $T$ , then*

$$\mathcal{FT}\{x(t)\}(j\Omega) = 2\pi \sum_{k=-\infty}^{\infty} X^S[k] \delta\left(\Omega - \frac{2\pi k}{T}\right) \quad (2)$$

with  $X^S[k]$  the Fourier series coefficient at frequency  $k$  (Eq. 4).

*Proof.* Because  $x(t)$  is a periodic signal, it admits a Fourier series representation (Porat, 1997, Section 2.3)

$$x(t) = \sum_{k=-\infty}^{\infty} X^S[k] \exp\left(\frac{j2\pi kt}{T}\right) \quad (3)$$

with

$$X^S[k] = \frac{1}{T} \int_{-T/2}^{T/2} x(t) \exp\left(-\frac{j2\pi kt}{T}\right) dt \quad (4)$$

By the linearity of the Fourier transform (Porat, 1997, Eq. 2.4), from Eq. 3, we have

$$\mathcal{FT}\{x(t)\}(j\Omega) = \sum_{k=-\infty}^{\infty} X^S[k] \mathcal{FT}\left\{\exp\left(\frac{j2\pi kt}{T}\right)\right\}(j\Omega) \quad (5)$$

We next compute the Fourier transform of the exponential in the right hand side of Eq. 5

$$\mathcal{FT} \left\{ \exp \left( \frac{j2\pi kt}{T} \right) \right\} (j\Omega) = \mathcal{FT} \left\{ 1 \exp \left( \frac{j2\pi kt}{T} \right) \right\} (j\Omega) \quad (6)$$

$$= \mathcal{FT} \{1\} \left( j \left( \Omega - \frac{2\pi k}{T} \right) \right) \quad (7)$$

$$= 2\pi \delta \left( \Omega - \frac{2\pi k}{T} \right) \quad (8)$$

Notes:

1. Eq. 7 follows from Eq. 6 by the the frequency shift property of the Fourier transform<sup>1</sup>.
2. Eq. 8 follows from Eq. 7 by the Fourier transform of the DC function (Lemma 2).

Replacing Eq. 8 into Eq. 5 yields Eq. 2.

□

**Lemma 2** (Fourier transform of the DC function).

$$\mathcal{FT}\{1\}(j\Omega) = 2\pi \delta(\Omega) \quad (9)$$

*Proof.* We start by computing the Fourier transform of the delta function.

$$\mathcal{FT}\{\delta(t)\}(j\Omega) = \int_{-\infty}^{\infty} \delta(t) \exp(-j\Omega t) dt = \exp(-j\Omega t)|_{t=0} = 1 \quad (10)$$

Then by the duality property of the Fourier transform (Lemma 3) we have

$$\mathcal{FT}\{1\}(j\Omega) = 2\pi \delta(-\Omega) = 2\pi \delta(\Omega) \quad (11)$$

Notes:

1. The last equality in Eq. 11 holds because the delta function is even.

□

**Lemma 3** (Duality of the Fourier transform). *Let  $x(t)$  be a signal and  $X(j\Omega)$  be its Fourier transform, then*

$$\mathcal{FT}\{X(jt)\}(j\Omega) = 2\pi x(-\Omega) \quad (12)$$

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<sup>1</sup> $y(t) = e^{j\Omega_0 t} x(t) \leftrightarrow Y(j\Omega) = X(j(\Omega - \Omega_0))$ , Porat (1997, Section 2.1)

*Proof.* If  $x(t)$  is a signal, with real or complex values, and  $X(j\Omega)$  is its Fourier transform, then they are related by the following equations (Porat, 1997, Section 2.1)

$$X(j\Omega) = \mathcal{FT}\{x(t)\}(j\Omega) = \int_{-\infty}^{\infty} x(t) \exp(-j\Omega t) dt \quad (13)$$

$$x(t) = \mathcal{IFT}\{X(j\Omega)\}(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\Omega) \exp(j\Omega t) d\Omega \quad (14)$$

Then

$$\mathcal{FT}\{X(jt)\}(j\Omega) = \int_{-\infty}^{\infty} X(jt) \exp(-j\Omega t) dt \quad (15)$$

$$= 2\pi \left( \frac{1}{2\pi} \int_{-\infty}^{\infty} X(jt) \exp(jt(-\Omega)) dt \right) \quad (16)$$

$$= 2\pi x(-\Omega) \quad (17)$$

Notes:

1. in Eq. 15 we applied the Fourier transform (Eq. 13) to the complex signal  $X(jt)$
2. in Eq. 17 we used the inverse Fourier transform (Eq. 14) with the change of variables  $\Omega$  in Eq. 14 to  $t$  in Eq. 16 and  $t$  in Eq. 14 to  $-\Omega$  in Eq. 16.

□

## References

Porat, B. (1997). *A course in digital signal processing*. John Wiley & Sons, Inc.