



Master in Computer Vision *Barcelona*

Module: Introduction to Human and Computer Vision

Lecture 5: Space-frequency representation (I)

Lecturer: Javier Ruiz Hidalgo

Me

- Javier Ruiz Hidalgo
 - Email: j.ruiz@upc.edu
 - Office: UPC, Campus Nord, D5-008
- Teaching experience
 - Basic signal processing
 - Image processing & computer vision
 - M1, M4 and M6 (Computer Vision Master)
- Research experience
 - Master on hierarchical image representations by UEA (UK)
 - PhD on video coding by UPC (Spain)
 - Interests in image & video coding, 3D analysis and super-resolution

Outline

- Introduction
 - Image model definition
 - Convolution
 - Correlation
- 2D Fourier Analysis
 - Definition, properties and basic transforms
- 2D Discrete Fourier Analysis
 - Definition, properties and basic transforms

Outline

- **Introduction**
 - Image model definition
 - Convolution
 - Correlation
- 2D Fourier Analysis
 - Definition, properties and basic transforms
- 2D Discrete Fourier Analysis
 - Definition, properties and basic transforms

Image model definition

- In the **space/frequency image model**, the image is understood as a **linear combination** of simpler functions; typically:
 - Impulse functions, as in the canonical representation.
 - Complex exponentials, as in the Fourier representation.
- One of the most natural operations are **Linear and Space-Invariant** operations. They linearly combine the pixel values in a given neighborhood of the pixel being processed (**impulse response, convolution mask, kernel**).
- Linear Space-Invariant (**LSI**) operators can be defined:
 - In the original (space) domain, through a **convolution**.
 - In the transformed (frequency) domain, through a **product**.

Canonical representation

- Unit Impulse

$$\delta[m, n] = \begin{cases} 1 & m = n = 0 \\ 0 & \text{other} \end{cases}$$

$$\delta[m, n] = \delta[m] \cdot \delta[n]$$

separable

- Linear decomposition

- Every sequence can be expressed as a linear combination of displaced unit impulses

$$x[m, n] = \sum_{m'=-\infty}^{\infty} \sum_{n'=-\infty}^{\infty} x[m', n'] \delta[m - m', n - n']$$

The diagram shows a blue rectangular matrix representing a signal $x[m, n]$. The top-left element is highlighted with red numbers 3, 2, 1. To its right is an equals sign. Following the equals sign are three terms: a scalar 3 followed by a blue dashed rectangular matrix where the top row has values 1, 0, 0, ..., and all other rows are zero; a scalar 2 followed by a blue dashed rectangular matrix where the first column has values 0, 1, 0, ..., and all other columns are zero; and a scalar 1 followed by a vertical ellipsis. This illustrates that the original signal is a linear combination of three displaced unit impulse signals.

$$\begin{matrix} 3 & 2 & 1 \\ 6 & 4 & 7 \\ \vdots & & \end{matrix} = 3 \cdot \begin{matrix} 1 & 0 & 0 & \dots \\ 0 & 0 & 0 & \vdots \end{matrix} + 2 \cdot \begin{matrix} 0 & 1 & 0 & \dots \\ 0 & 0 & 0 & \vdots \end{matrix} + 1 \cdot \dots$$

2D impulse response

- On a **linear** system:

$$\begin{aligned}y[m,n] &= T\{x[m,n]\} = T\left\{\sum_{m'} \sum_{n'} x[m',n'] \cdot \delta[m - m', n - n']\right\} = \\&= \sum_{m'} \sum_{n'} x[m',n'] \cdot T\{\delta[m - m', n - n']\}\end{aligned}$$

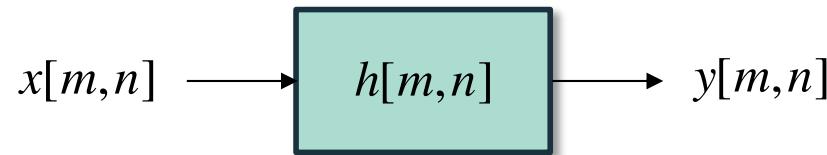
- If the system is **space-invariant**:

$$h[m,n] = T\{\delta[m,n]\} \quad \text{Impulse response}$$

$$T\{\delta[m-m', n-n']\} = h[m-m', n-n']$$

Image convolution (1)

- A Linear Space-Invariant operator can be defined by its **impulse response** $h[m, n]$:



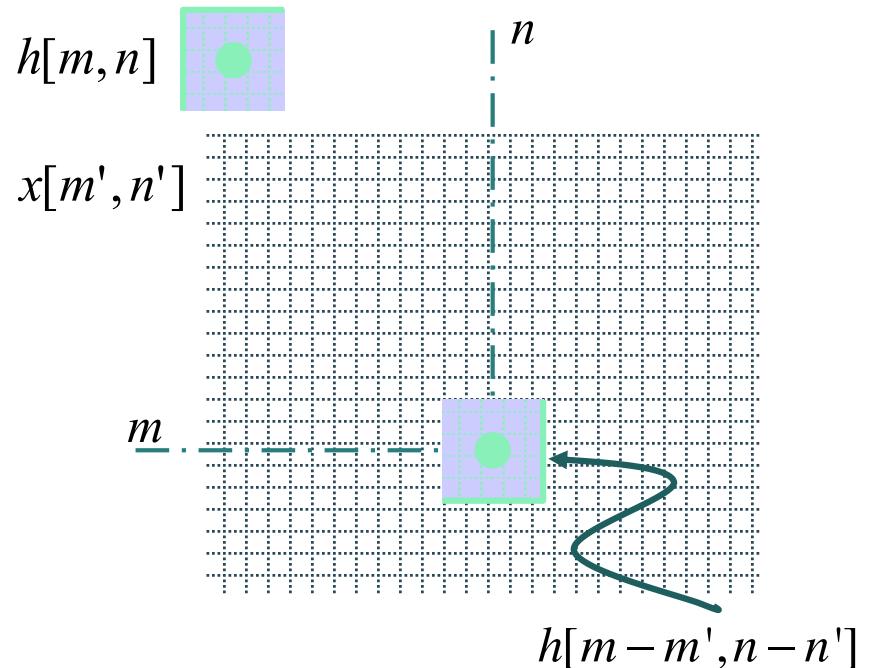
and the output is computed by the **convolution** equation:

$$y[m, n] = x[m, n] * h[m, n] = \sum_{m'} \sum_{n'} x[m', n'] h[m - m', n - n']$$

Image convolution (2)

- For each pixel position (m, n) :
 1. The **reflected version** of $h[m',n']$ is shifted at position $(h[m-m',n-n'])$
 2. the output $y[m,n]$ is the sum of the pixel values included in the neighborhood defined by the filter and weighted by the co-located filter values ($h[m,n]$ is the **filter impulse response**)

$$y[m,n] = \sum_{m'} \sum_{n'} x[m',n'] h[m - m', n - n']$$



- A **similar interpretation** where the input signal is shifted is also possible

$$y[m,n] = \sum_{m'} \sum_{n'} x[m - m',n - n'] h[m',n']$$

Image convolution (3)

- The result of the convolution of an image of size $M \times N$ and a filter with impulse response of size $H_1 \times H_2$ is an image of size $(M+H_1-1) \times (N+H_2-1)$
- Nevertheless, in almost all practical cases, the output image support is **finally restricted** to the original image size.

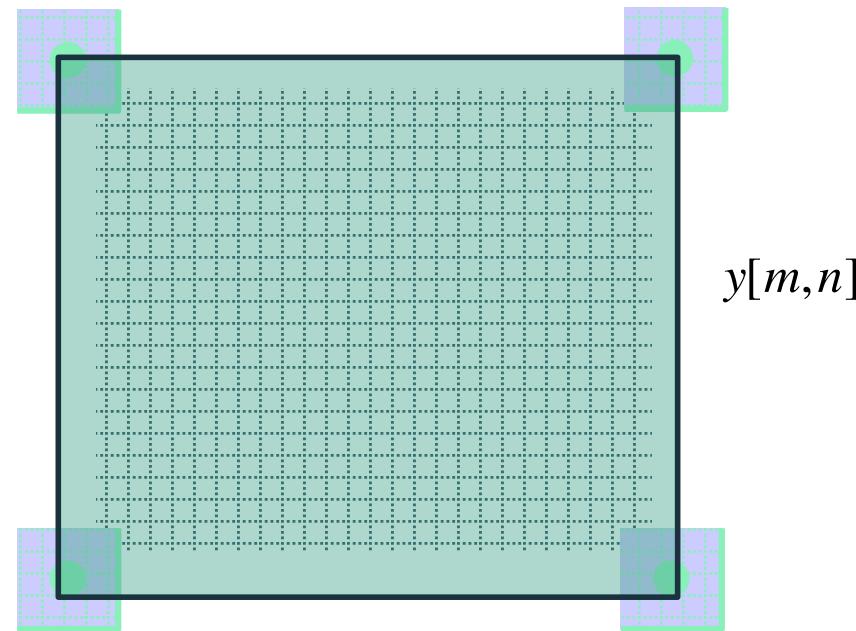
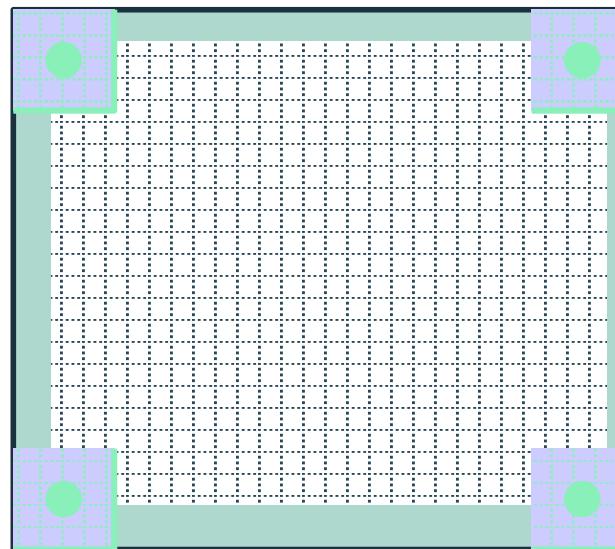


Image padding

- The computation of the output values close to the image border requires the use of pixel values outside the input image support.
- These values are commonly introduced by:
 - **Zero padding**: Including a frame of zeros in the signal to be processed (may introduce discontinuities in the signal).
 - **Mirroring**: Including pixel values that are a symmetric mirroring of the existing ones.
 - Implement a **space variant** filter to only use known pixel values.



Separability

- In some cases, the 2D impulse response can be defined as the product of two 1D impulse responses with different spatial indexes (**separability**)

$$h[m,n] = h_1[m] \cdot h_2[n]$$

- Separability allows for **faster implementations**

$$\begin{aligned} y[m,n] &= \sum_{m'} \sum_{n'} x[m - m', n - n'] h[m', n'] = \sum_{m'} \sum_{n'} x[m - m', n - n'] h_1[m'] h_2[n'] = \\ &= \sum_{n'} h_2[n'] \sum_{m'} x[m - m', n - n'] h_1[m'] = x[m,n] *_{rows}^{1D} h_1[m] *_{columns}^{1D} h_2[n] \end{aligned}$$

Outline

- **Introduction**
 - Image model definition
 - Convolution
 - **Correlation**
- 2D Fourier Analysis
 - Definition, properties and basic transforms
- 2D Discrete Fourier Analysis
 - Definition, properties and basic transforms

Overview of correlation between 1D signals

- Distance between $x[n]$ and $y[n-m]$:

$$\begin{aligned} D[m] &= \sum_{n=-\infty}^{\infty} |x[n] - y[n-m]|^2 = \\ &= \underbrace{\sum_{n=-\infty}^{\infty} |x[n]|^2}_{E_x} + \underbrace{\sum_{n=-\infty}^{\infty} |y[n-m]|^2}_{E_y} - \underbrace{\sum_{n=-\infty}^{\infty} x[n]y^*[n-m]}_{x[m]*y^*[-m]} - \underbrace{\sum_{n=-\infty}^{\infty} x^*[n]y[n-m]}_{x^*[m]*y[-m]} \end{aligned}$$

- Cross-correlation between $x[n]$ and $y[n]$

$$r_{xy}[m] \equiv \sum_{n=-\infty}^{\infty} x[n]y^*[n-m] = x[m]*y^*[-m]$$

therefore, $D[m] = E_x + E_y - (r_{xy}[m] + r_{xy}^*[m])$

if E_x and E_y are constant:

minimum
distance

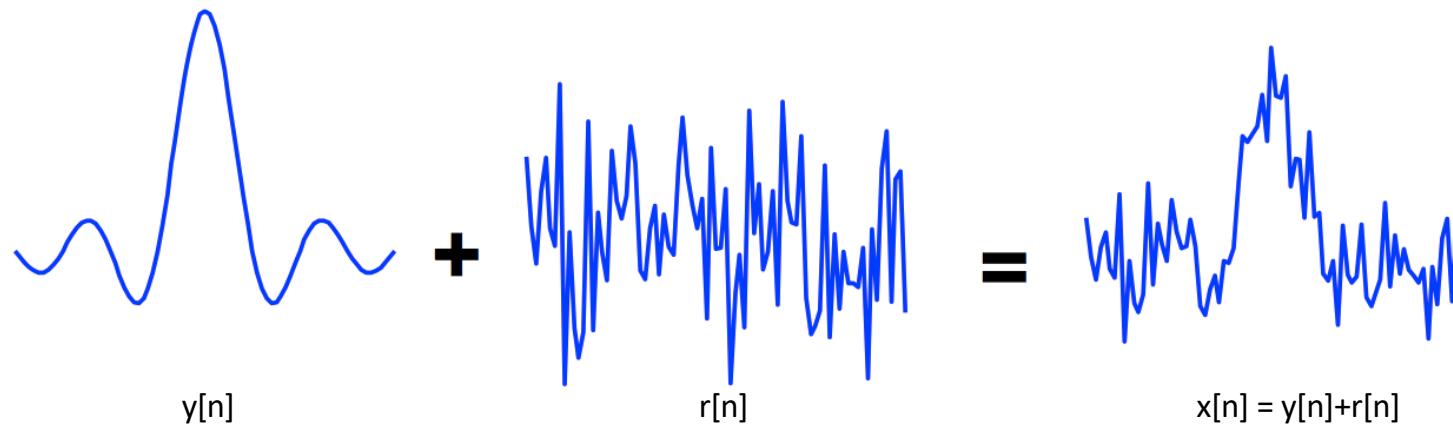
↔
maximum
correlation

Matched filtering

- We can interpret the correlation as a filter with impulse response $h[n]=y^*[-n]$ (**matched filter**)

$$r_{xy}[m] \equiv x[m] * y^*[-m] = x[m] * h[m]$$

- Used in binary detection systems:
 - $y[n]$ is a known signal
 - $x[n]$ is the test signal (contaminated with Gaussian noise $x[n] = y[n] + r[n]$)
 - The matched filter maximizes the SNR at $m=0$ ($r_{xy}[0]$)



Correlation of images

- Cross-correlation:

$$r_{xy}[m,n] = x[m,n] * y^*[-m,-n] = \sum_{m'=-\infty}^{\infty} \sum_{n'=\infty}^{\infty} x[m',n'] y^*[m'-m, n'-n]$$

- Auto-correlation:

$$r_{xx}[m,n] = x[m,n] * x^*[-m,-n] = \sum_{m'=-\infty}^{\infty} \sum_{n'=\infty}^{\infty} x[m',n'] x^*[m'-m, n'-n]$$

- Properties:

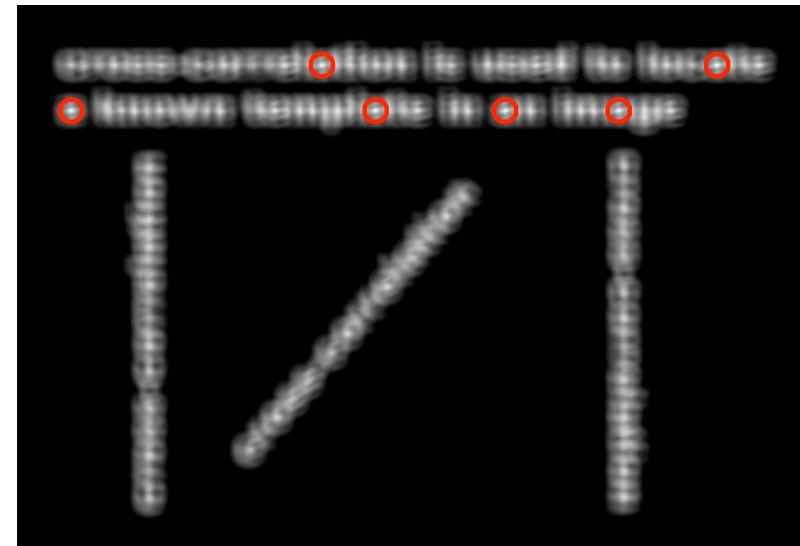
- Energy: $E_x = r_x[0,0] = \sum_{n=-\infty}^{\infty} x[m,n]x^*[m-0, n-0] = \sum_{m=-\infty}^{\infty} \sum_{n=-\infty}^{\infty} |x[m,n]|^2$
- Maximum: $|r_x[m,n]| \leq r_x[0,0] = E_x$
- Symmetry: $r_x[m,n] = r_x^*[-m,-n]$

Application: Template matching (1)



a
 $t[m,n]$

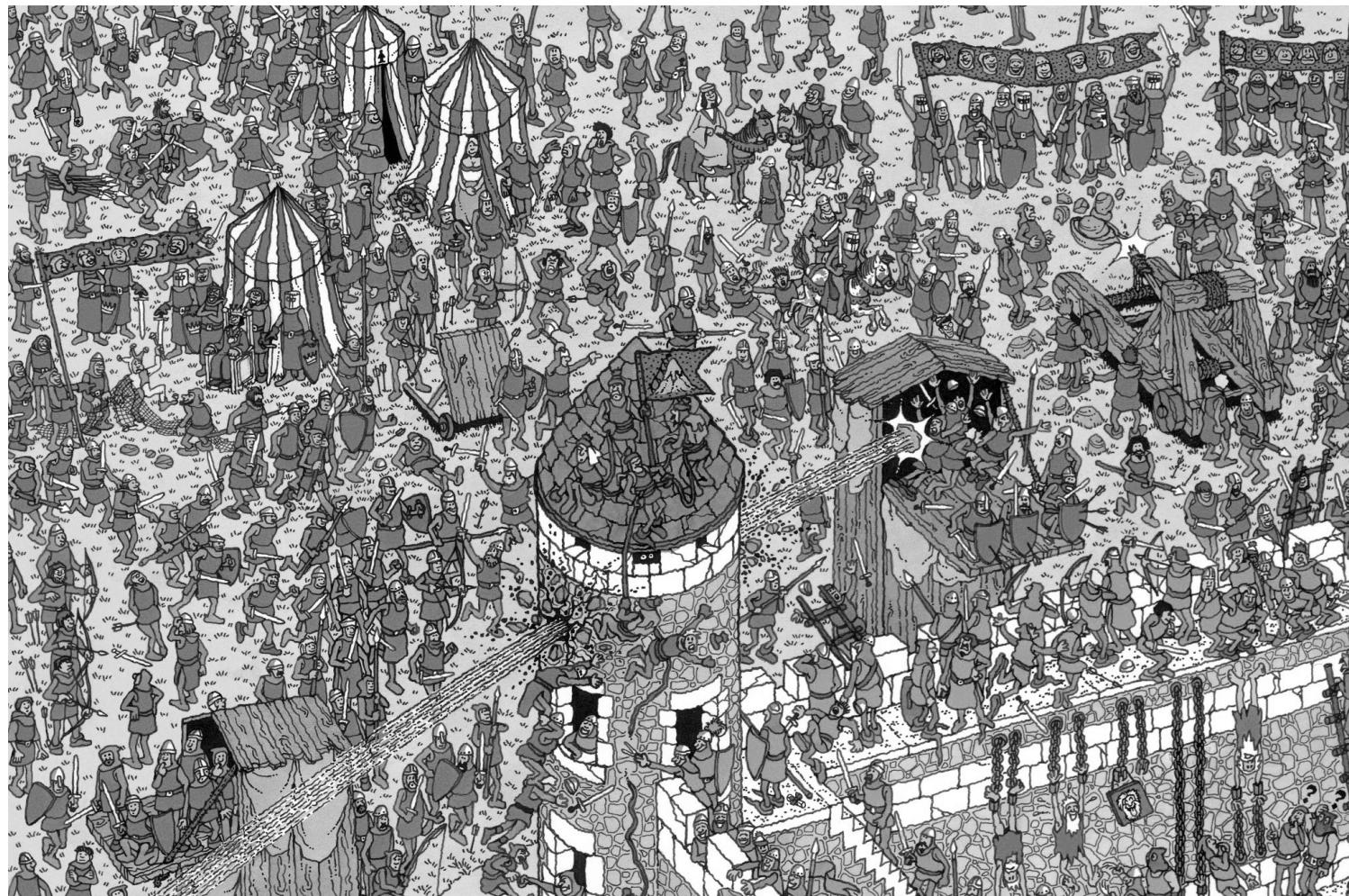
$x[m,n]$



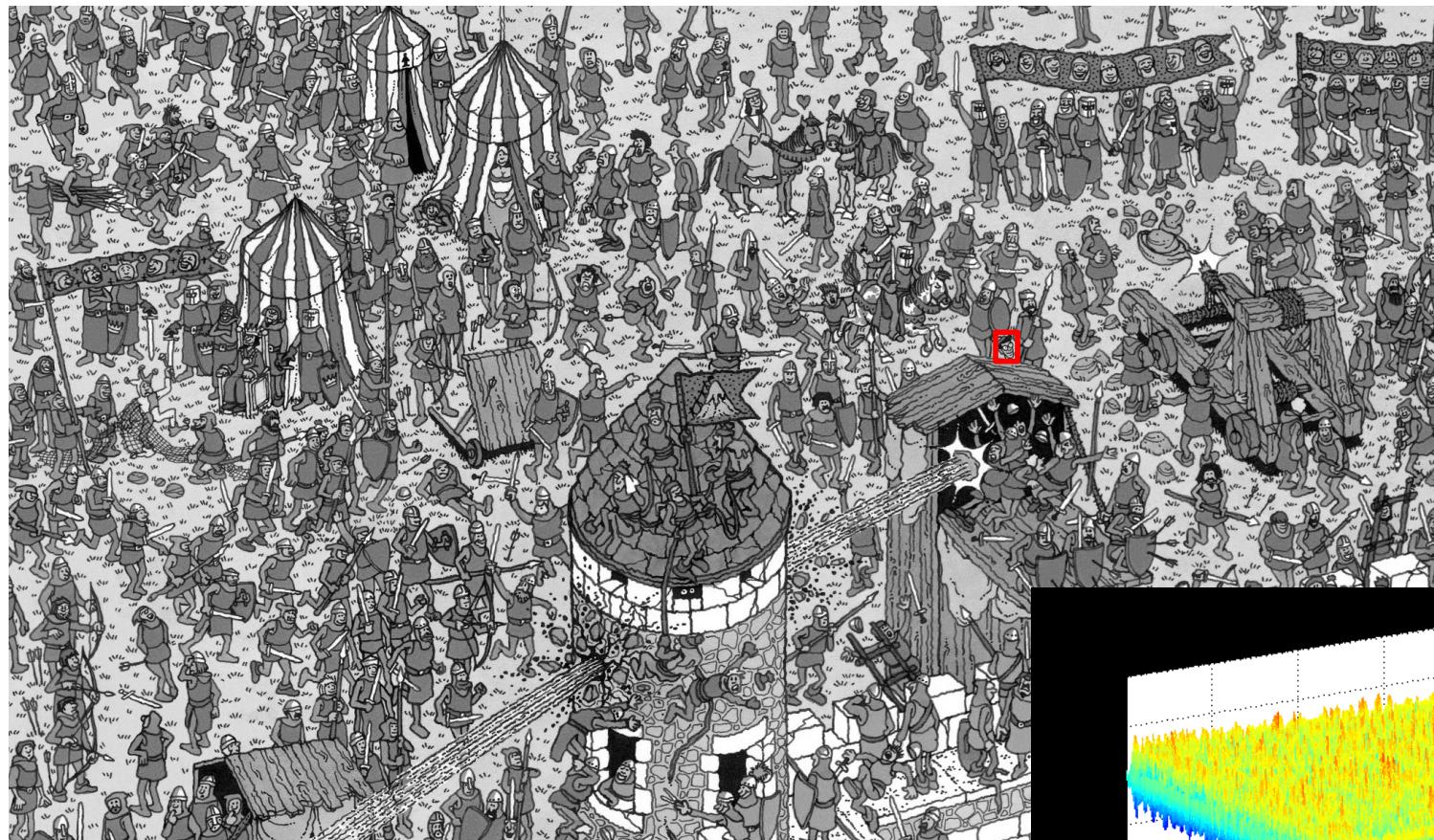
$r_{xt}[m,n]$

- Locate known objects or regions in image $x[m,n]$ could be performed using cross-correlation with a **template** $t[m,n]$
- If there is a match the correlation will be maximum at the position where $t[m,n]$ appears in $x[m,n]$
- There is a need of pre-processing (scale, align, etc.)

Application: Template matching (2)



Application: Template matching (3)

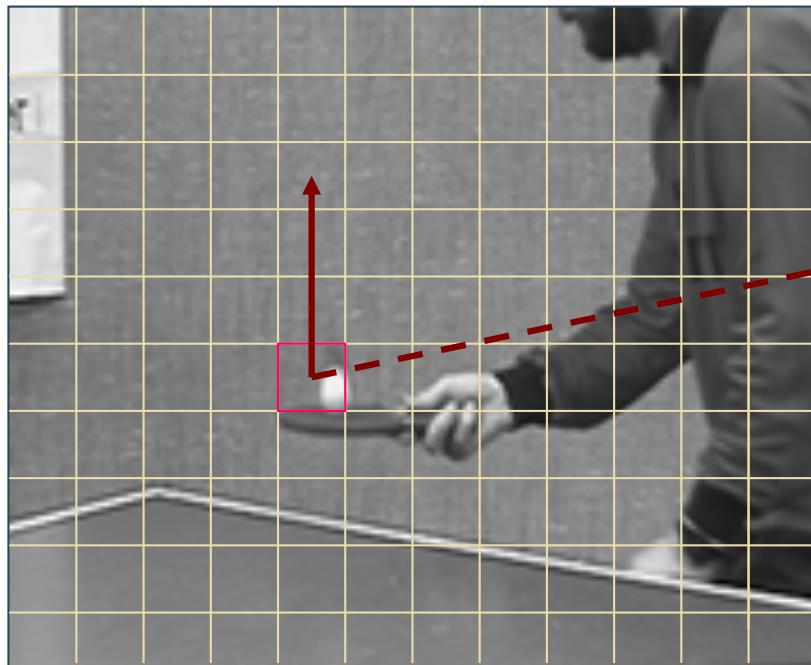


$$r_{xy}[m, n] = \frac{\sum_{m'=-\infty}^{\infty} \sum_{n'=-\infty}^{\infty} (x[m', n'] - \bar{x}_{m,n})(t[m' - m, n' - n] - \bar{t})}{\sqrt{E_{x[m',n']-\bar{x}_{m,n}} E_{t[m',n']-\bar{t}}}}$$

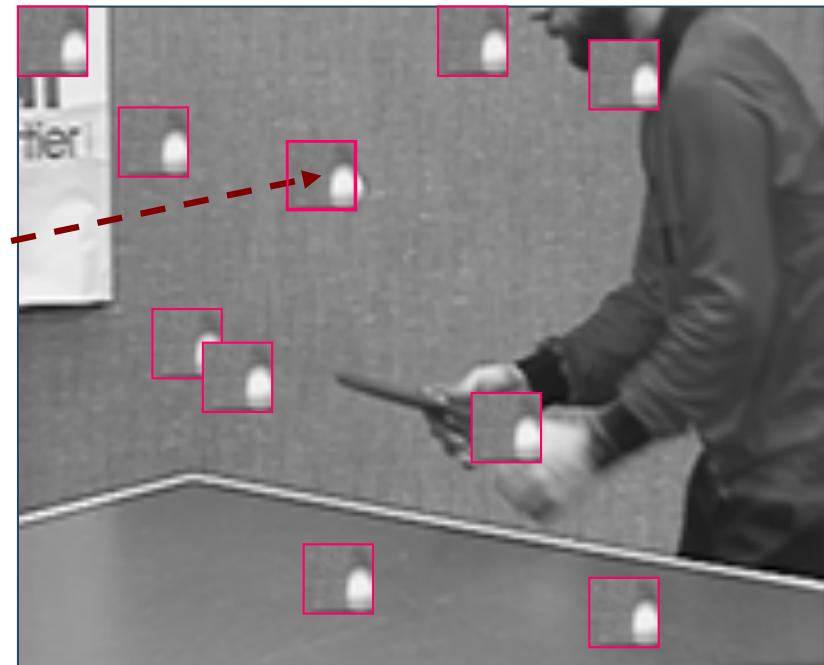
Application: Motion estimation (1)

- Motion estimation between two images (reference and actual)
- Block-Matching
 - Actual image is divided into blocks
 - Motion is estimated independently for each block
 - Translational motion
 - Luminance changes are assumed to be caused by motion (no illumination changes, noise, etc.)

Application: Motion estimation (2)



**Image to estimate
(actual)**



reference

Application: Motion estimation (3)

- Possible searching algorithm → **Correlation**
 - A search window is used to reduce the computational complexity (of doing a correlation between the block and the entire image for all blocks)
- The correlation maximum locates the most similar block
 - There is a need to **normalize the energy** of the blocks



Image to estimate



reference

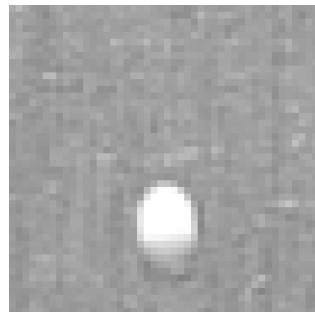
Application: Motion estimation (4)

b[m,n]



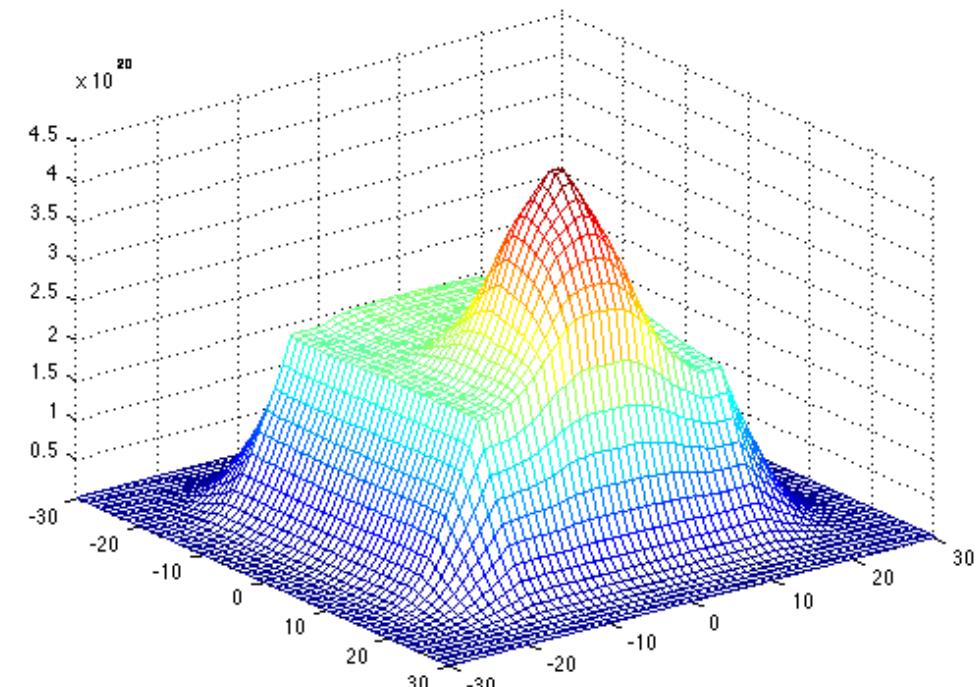
block (16x16)

a[m,n]



Window search (46x46)

$$r_{ab}[m,n] = a[m,n] * b^*[-m,-n]$$



Correlation (61x61)

Outline

- Introduction
 - Image model definition
 - Convolution
 - Correlation
- 2D Fourier Analysis
 - Definition, properties and basic transforms
- 2D Discrete Fourier Analysis
 - Definition, properties and basic transforms

Space / Frequency image processing tools

- If LSI operators involve large impulse responses, they can be **efficiently implemented** thanks to the use of fast transforms:
 1. The input image is (fast) transformed (FFT).
 2. The operation is performed in the transformed domain by a product.
 3. The output image is obtained through an inverse FFT.
- Space/Frequency image processing tools allow (among others):
 - Convolution operations
 - Linear filter design
 - Analysis of sampling
 - Multi-resolution analysis:
 - Decomposition of the input image into multiple images containing the information at different 2D frequency bands

2D Fourier transform

- Definition:
 - The Fourier Transform (**FT**) of an image $x[m,n]$ is defined as:

$$X(F_x, F_y) = \sum_{m=-\infty}^{\infty} \sum_{n=-\infty}^{\infty} x[m, n] e^{-j(m2\pi F_x + n2\pi F_y)} \quad -\frac{1}{2} \leq \{F_x, F_y\} < \frac{1}{2}$$

- Whereas the Inverse Fourier Transform (**IFT**) is defined as:

$$x[m, n] = \int_{F_x=-\frac{1}{2}}^{\frac{1}{2}} \int_{F_y=-\frac{1}{2}}^{\frac{1}{2}} X(F_x, F_y) e^{j(m2\pi F_x + n2\pi F_y)} dF_x dF_y$$

F_x and F_y correspond to the horizontal and vertical frequencies respectively

FT properties (1)

- Usually samples in the spatial domain $x[m,n]$ are real (or integer) whereas **samples in the frequency domain are complex**:

$$X(F_x, F_y) = X_R(F_x, F_y) + jX_I(F_x, F_y) = |X(F_x, F_y)| e^{j\theta_X(F_x, F_y)}$$

- It is a periodic signal (for all integer k):

$$X(F_x, F_y) = X(F_x \pm k, F_y) = X(F_x, F_y \pm k)$$

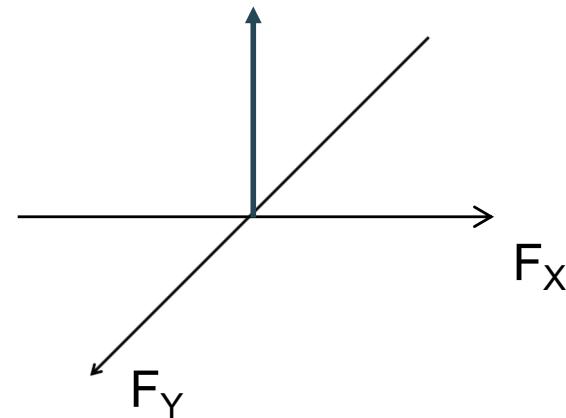
FT properties (2)

- Uniqueness
- Linearity
- Separability (compute it by rows and columns)
- Invariant against rotation
- Symmetry
- Convolution
 - Spatial convolution implies frequency product

$$x[m,n] * y[m,n] \Leftrightarrow X(F_x, F_y) \cdot Y(F_x, F_y)$$

FT examples of infinite images (1)

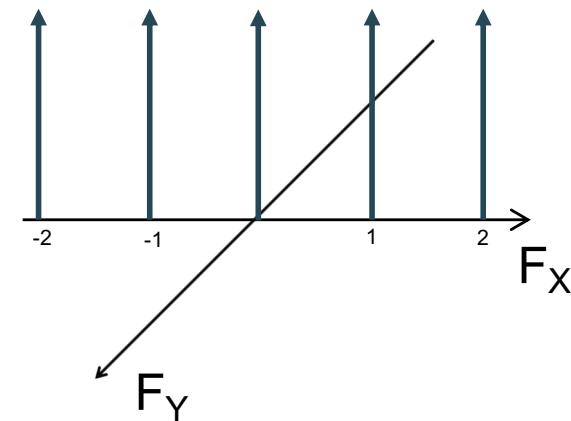
$$x[m, n] = 1 \iff \sum_{k=-\infty}^{\infty} \sum_{l=-\infty}^{\infty} \delta(F_x - k, F_y - l)$$



$$\delta(F_x, F_y)$$

FT examples of infinite images (1)

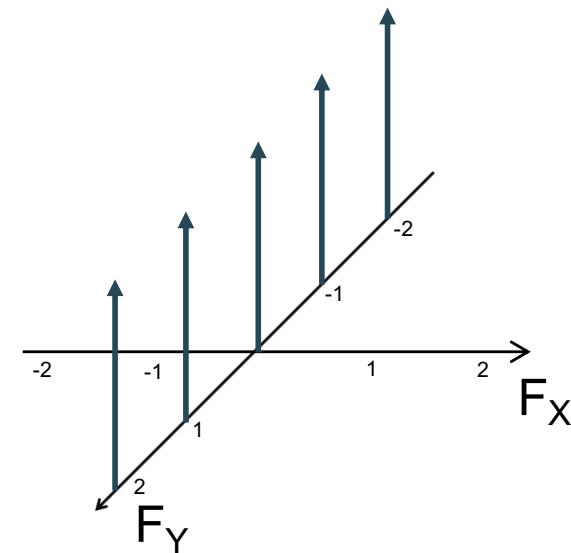
$$x[m, n] = 1 \iff \sum_{k=-\infty}^{\infty} \sum_{l=-\infty}^{\infty} \delta(F_x - k, F_y - l)$$



$$\sum_{k=-\infty}^{\infty} \delta(F_x - k, F_y)$$

FT examples of infinite images (1)

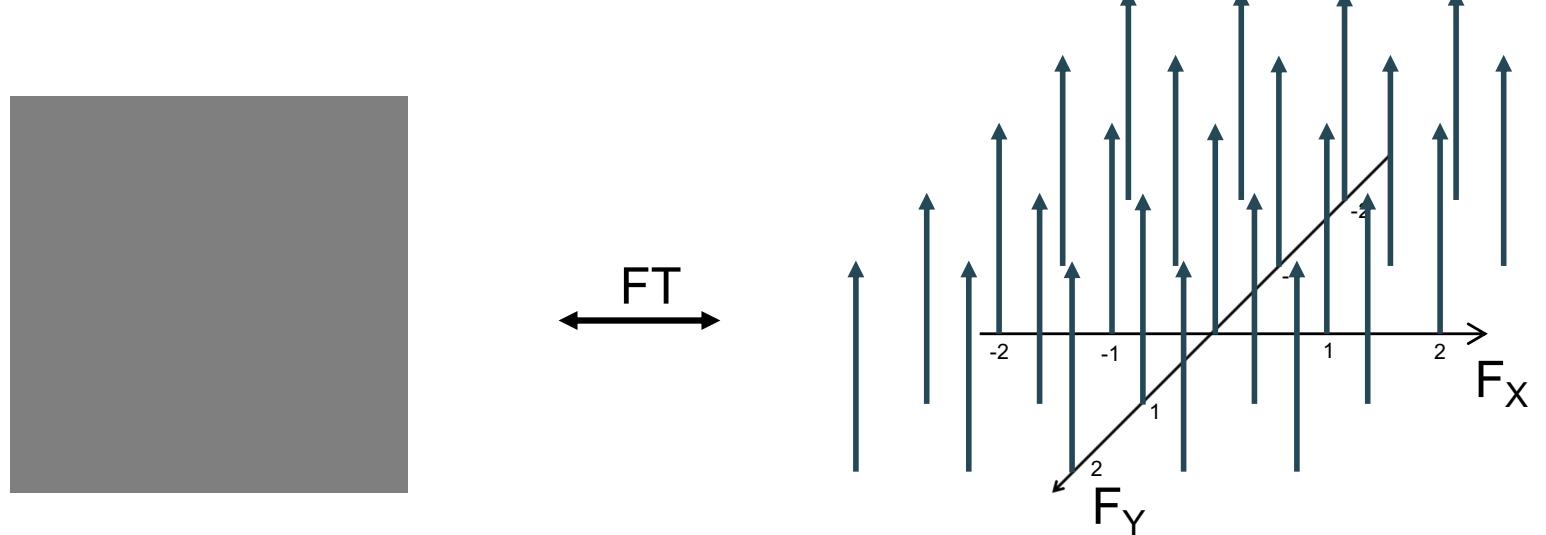
$$x[m, n] = 1 \iff \sum_{k=-\infty}^{\infty} \sum_{l=-\infty}^{\infty} \delta(F_x - k, F_y - l)$$



$$\sum_{l=-\infty}^{\infty} \delta(F_x, F_y - l)$$

FT examples of infinite images (1)

$$x[m, n] = 1 \iff \sum_{k=-\infty}^{\infty} \sum_{l=-\infty}^{\infty} \delta(F_x - k, F_y - l)$$

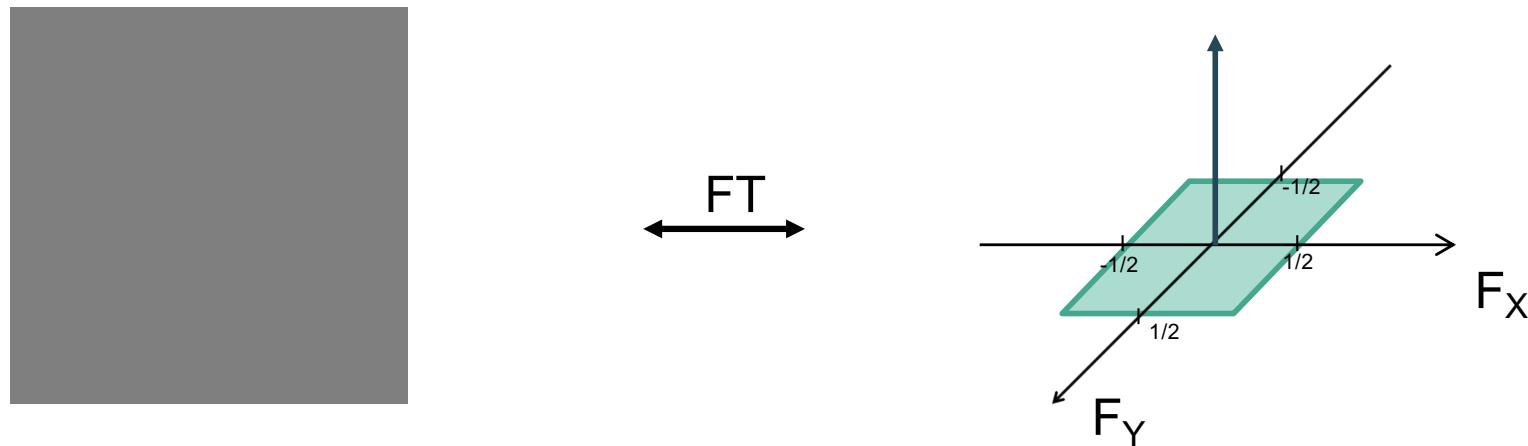


$$x[m, n] = 1$$

$$\sum_{k=-\infty}^{\infty} \sum_{l=-\infty}^{\infty} \delta(F_x - k, F_y - l)$$

FT examples of infinite images (1)

$$x[m, n] = 1 \iff \delta(F_x, F_y) - \frac{1}{2} \leq \{F_x, F_y\} < \frac{1}{2}$$



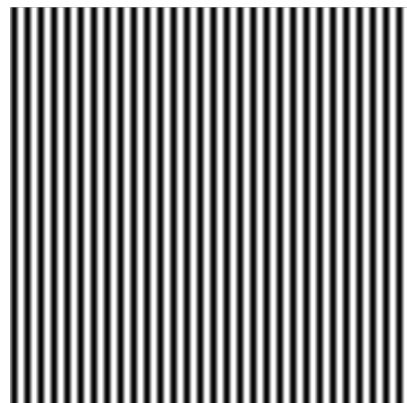
$$x[m,n] = 1$$

$$\delta(F_x, F_y) - \frac{1}{2} \leq \{F_x, F_y\} < \frac{1}{2}$$

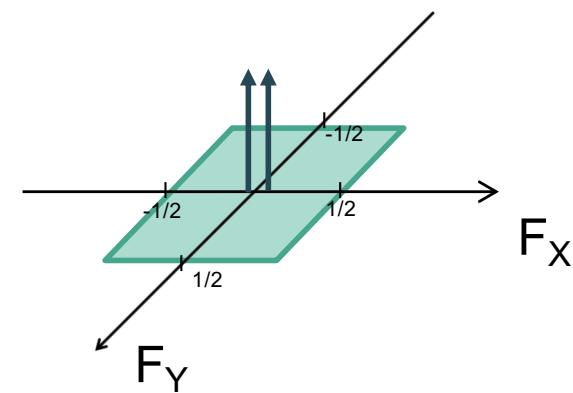
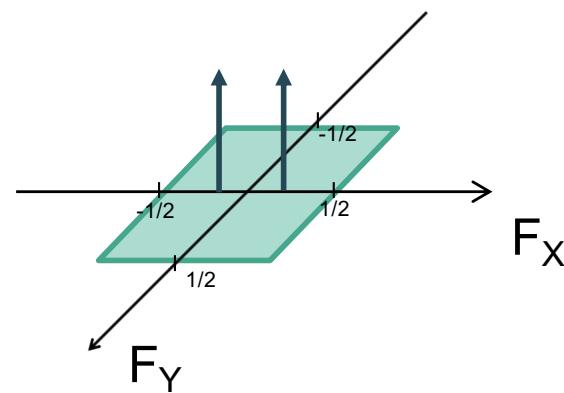
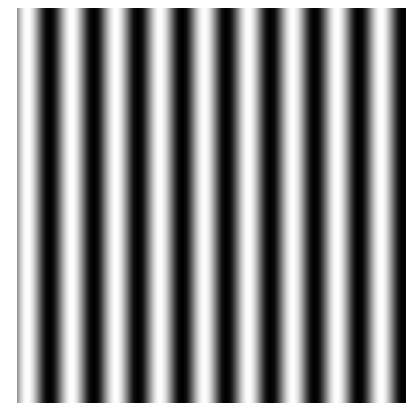
FT examples of infinite images (2)

$$x[m, n] = \cos(2\pi F_1 m) \iff \frac{1}{2}\delta(F_x - F_1, F_y) + \frac{1}{2}\delta(F_x + F_1, F_y) \quad -\frac{1}{2} \leq \{F_x, F_y\} < \frac{1}{2}$$

$F_1 = 0.1$



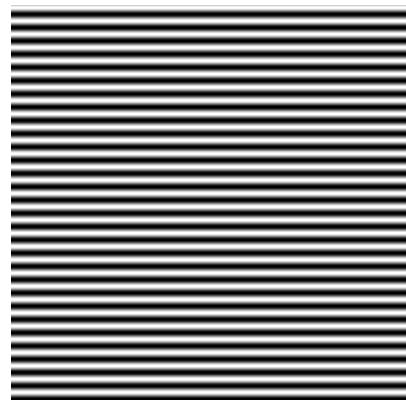
$F_1 = 0.03$



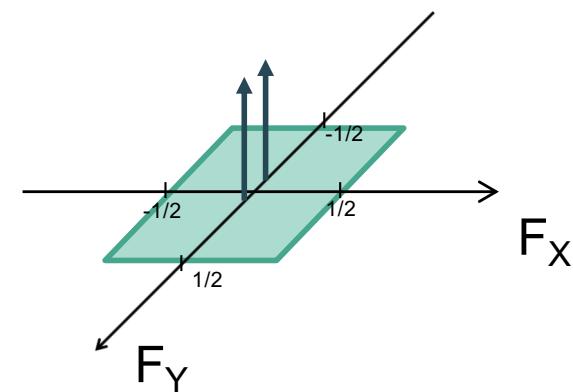
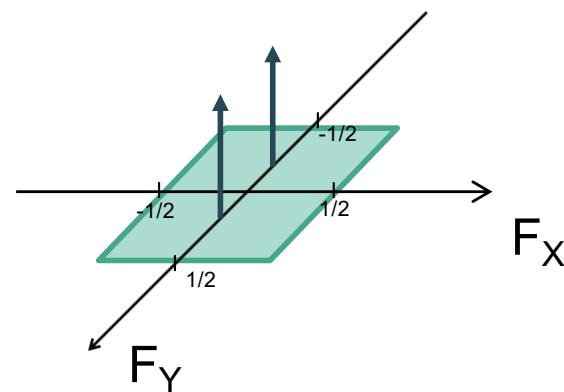
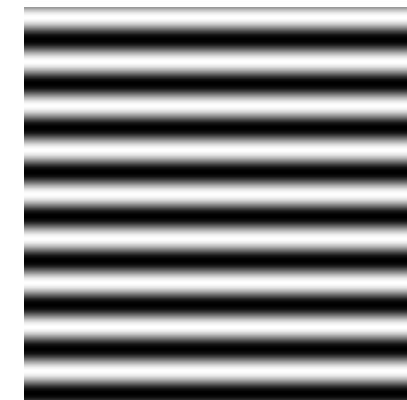
FT examples of infinite images (2)

$$x[m, n] = \cos(2\pi F_2 n) \iff \frac{1}{2}\delta(F_x, F_y - F_2) + \frac{1}{2}\delta(F_x, F_y + F_2) \quad -\frac{1}{2} \leq \{F_x, F_y\} < \frac{1}{2}$$

$F_2 = 0.1$



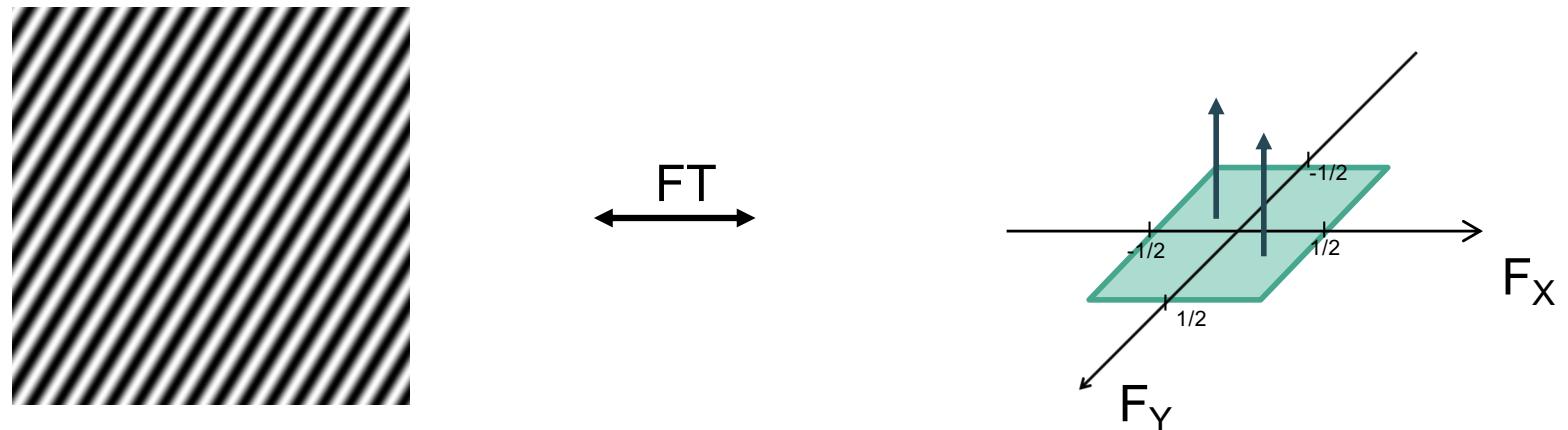
$F_2 = 0.03$



FT examples of infinite images (2)

$$x[m, n] = \cos(2\pi F_1 m + 2\pi F_2 n) \xleftrightarrow{FT} \frac{1}{2}\delta(F_x - F_1, F_y - F_2) + \frac{1}{2}\delta(F_x + F_1, F_y + F_2) \quad -\frac{1}{2} \leq \{F_x, F_y\} < \frac{1}{2}$$

$$F_1=0.05, F_2= 0.03$$



FT examples of infinite images (3)

- FT of a rectangular pulse MxN (8x60)

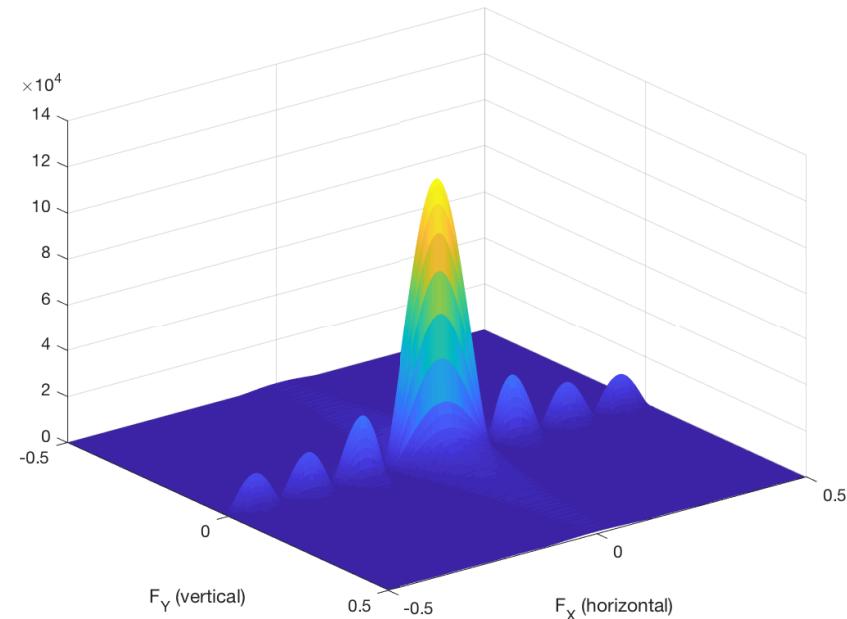
$$x[m, n] = \Pi_{M \times N}[m, n] = \Pi_M[m] \cdot \Pi_N[n] \xleftrightarrow{FT} X(F_x, F_y) = X_M(F_x) \cdot F_N(F_y)$$

separable

$$|X_M(F_x)| = \frac{1}{M} \frac{\sin(M\pi F_x)}{\sin(\pi F_x)}$$



$x[m, n]$



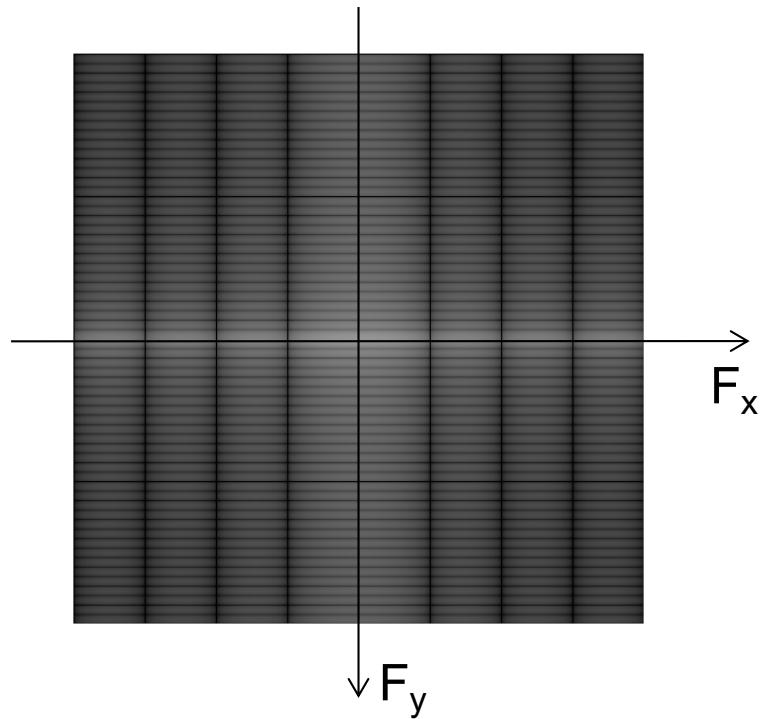
TF magnitude (modulus)

FT examples of infinite images (3)

- FT of a rectangular pulse (8x60)



image



TF magnitude (modulus)

FT examples of infinite images (4)

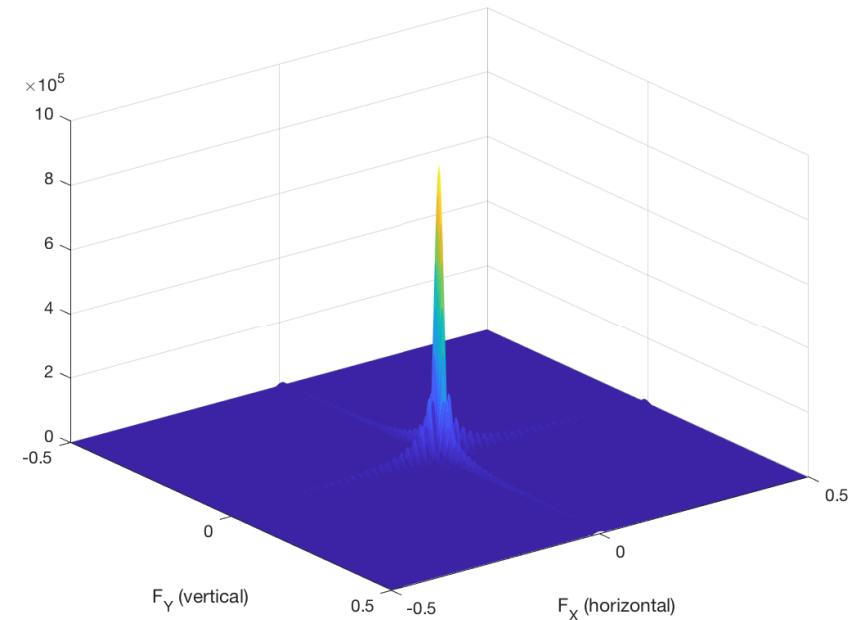
- FT of a rectangular pulse MxN (60x60)

$$x[m, n] = \Pi_{M \times N}[m, n] = \Pi_M[m] \cdot \Pi_N[n] \xleftrightarrow{FT} X(F_x, F_y) = X_M(F_x) \cdot F_N(F_y)$$

$$|X_M(F_x)| = \frac{1}{M} \frac{\sin(M\pi F_x)}{\sin(\pi F_x)}$$



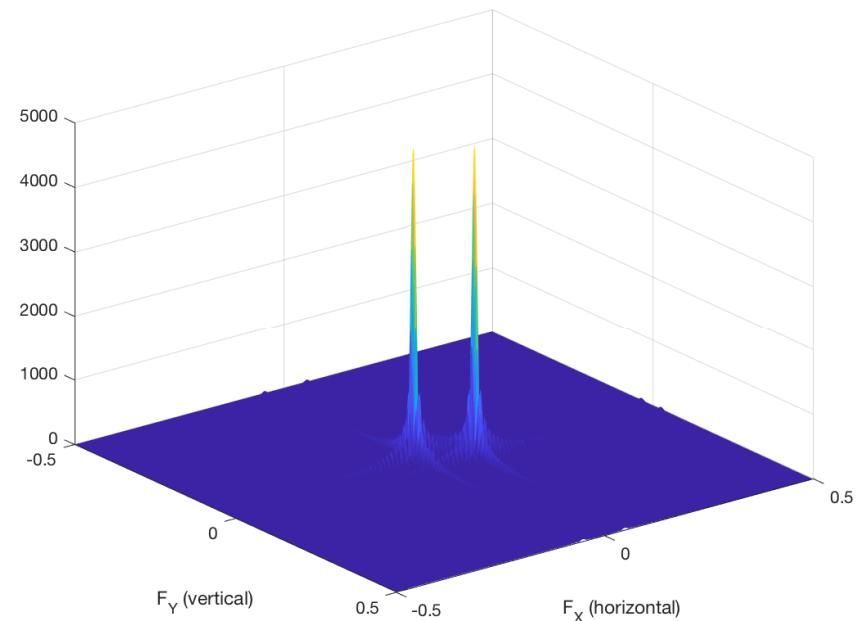
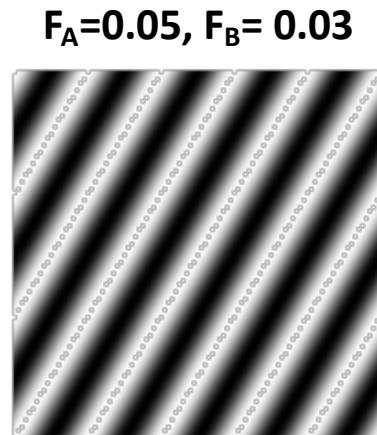
$x[m, n]$



TF magnitude (modulus)

FT example of finite images

- Windowing effect (100x100)



TF magnitude (modulus)

Outline

- Introduction
 - Image model definition
 - Convolution
 - Correlation
- 2D Fourier Analysis
 - Definition, properties and basic transforms
- **2D Discrete Fourier Analysis**
 - **Definition, properties and basic transforms**

2D Discrete Fourier analysis

- The **Discrete Fourier Transform (DFT)** of an image is defined as:

$$X[k,l] = \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} x[m,n] e^{-j2\pi \left(\frac{k}{M}m + \frac{l}{N}n \right)} \quad 0 \leq \begin{Bmatrix} k \\ l \end{Bmatrix} < \begin{Bmatrix} M \\ N \end{Bmatrix}$$

whereas the **Inverse Discrete Fourier Transform (IDFT)** is defined as:

$$x[m,n] = \frac{1}{MN} \sum_{k=0}^{M-1} \sum_{l=0}^{N-1} X[k,l] e^{j2\pi \left(\frac{k}{M}m + \frac{l}{N}n \right)} \quad 0 \leq \begin{Bmatrix} k \\ l \end{Bmatrix} < \begin{Bmatrix} M \\ N \end{Bmatrix}$$

- Therefore, the DFT **transforms a 2D signal of $M \times N$ samples** in the original (space) domain into a set of **$M \times N$ samples in the transformed** (frequency) domain:
 - Usually, samples in the spatial domain are real (or integer) whereas **samples in the frequency domain are complex**.

DFT relations with FT

$$X[k,l] = X(F_x, F_y) * V(F_x, F_y) \Big|_{F_x=\frac{k}{M}, F_y=\frac{l}{N}}$$

where $V(F_x, F_y)$ represents the FT of the implicit rectangular window in the DFT of $M \times N$ samples

If $M \times N$ is greater than or equal to the size of the image then:

$$X[k,l] = X(F_x, F_y) \Big|_{F_x=\frac{k}{M}, F_y=\frac{l}{N}}$$

DFT properties (1)

- **Properties of DFT with respect to TF:**
 - Similar properties but, in the case of a DFT of length L, indexes should remain in the interval [0,L-1]. This leads to **circular** convolution, displacement or time-reversal. These can be expressed with the help of:

$$\tilde{t}[m,n] = \sum_{r=-\infty}^{\infty} \sum_{s=-\infty}^{\infty} t[m+rM, n+sN], \text{ for } \begin{cases} 0 \leq m \leq M-1 \\ 0 \leq n \leq N-1 \end{cases}$$

- **Convolution:**
 - Spatial convolution implies frequency product.
- **Convolution and windowing:**
 - Windowing in the spatial domain implies frequency convolution.

$$\tilde{t}[m,n] \Leftrightarrow X[k,l] \cdot Y[k,l]$$

if $t[m,n] = x[m,n] * y[m,n]$

$$x[m,n] \cdot y[m,n] \Leftrightarrow \frac{1}{MN} \tilde{T}[k,l]$$

if $T[k,l] = X[k,l] * Y[k,l]$

DFT properties (2)

- **Spatial shift or Translation:**
 - A spatial shift only affects the phase of the transformed signal.

$$\tilde{x}[m - m', n - n'] \Leftrightarrow X[k, l] e^{-j2\pi\left(\frac{km'}{M} + \frac{ln'}{N}\right)}$$

example

- **Frequency shift or Modulation:**
 - Multiplication of an image by a complex exponential implies a frequency shift.

$$x[m, n] e^{j2\pi\left(\frac{k'm}{M} + \frac{l'n}{N}\right)} \Leftrightarrow \tilde{X}[k - k', l - l']$$

DFT properties (3)

- **Separability**
 - 1D algorithms can be used

$$DFT_{image}^{2D}[\cdot] = DFT_{rows}^{1D} \left[DFT_{columns}^{1D}[\cdot] \right]$$

$$\begin{aligned} X[k,l] &= \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} x[m,n] e^{-j2\pi \left(\frac{mk}{M} + \frac{nl}{N} \right)} = \sum_{m=0}^{M-1} e^{-j2\pi \frac{mk}{M}} \sum_{n=0}^{N-1} x[m,n] e^{-j2\pi \frac{nl}{N}} = \\ &= \sum_{m=0}^{M-1} e^{-j2\pi \frac{mk}{M}} V[m,l] \quad 0 \leq k \leq M-1 \quad 0 \leq l \leq N-1 \end{aligned}$$

DFT^{1D}_{rows} → DFT^{1D}_{columns}

- If the image is also a function of two 1D functions:

$$\begin{aligned} X[k,l] &= DFT_n^{1D} \left[DFT_m^{1D} [x[m,n]] \right] = DFT_n^{1D} \left[DFT_m^{1D} [x_1[m]x_2[n]] \right] = \\ &= DFT_n^{1D} [x_2[n]] DFT_m^{1D} [x_1[m]] = X_1[k] \cdot X_2[l] \end{aligned}$$

DFT properties (4)

- **Rotation:**

- A spatial rotation corresponds to the same frequency rotation

$$m = r \cos \theta \quad n = r \sin \theta$$

$$k = \omega \cos \phi \quad l = \omega \sin \phi$$

$$x[r, \theta + \theta_0] \leftrightarrow X[\omega, \phi + \theta_0]$$

- **Conjugate symmetry:**

- If the **image is real**, its Fourier transform has the Hermitian symmetry
- Useful when implementing filters

$$X[k, l] = "X^*[-k, -l]"$$

$$\left\{ \begin{array}{l} X[0,0] = X^*[0,0] \\ X[0,l] = X^*[0,N-l], \text{ for } 1 \leq l \leq N-1 \\ X[k,0] = X^*[M-k,0], \text{ for } 1 \leq k \leq M-1 \\ X[k,l] = X^*[M-k,N-l], \text{ for } \begin{array}{l} 1 \leq k \leq M-1 \\ 1 \leq l \leq N-1 \end{array} \end{array} \right.$$

Other DFT issues (1)

- **Centered representation:**

- Since images are generally positive signals, the highest value of the magnitude of their transform is at $(k, l) = (0, 0)$. Given the symmetries of the DFT, this makes that the four corners of the transformed image contain the highest values of the magnitude. In order **to help visualizing**, the transformed image is represented centered at $(M/2, N/2)$.
- This can be seen as an example of the **modulation property**

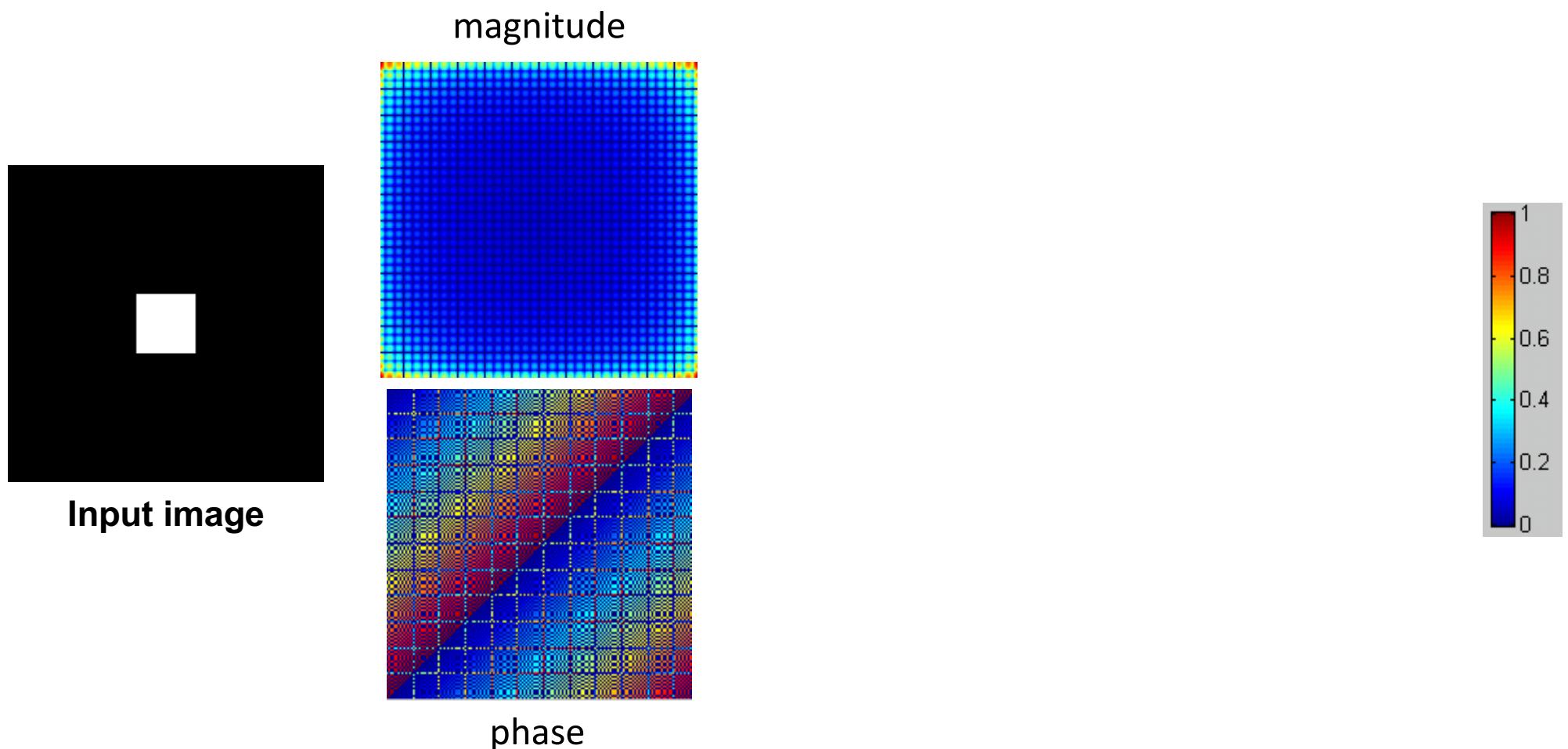
$$x[m,n]e^{j2\pi\left(\frac{k'm}{M}+\frac{l'n}{N}\right)} \leftrightarrow \tilde{X}[k-k',l-l']$$

$$\text{For } k' = \frac{M}{2}, l' = \frac{N}{2}, \quad x[m,n]e^{j2\pi\left(\frac{m}{2}+\frac{n}{2}\right)} \leftrightarrow \tilde{X}\left[k - \frac{M}{2}, l - \frac{N}{2}\right]$$

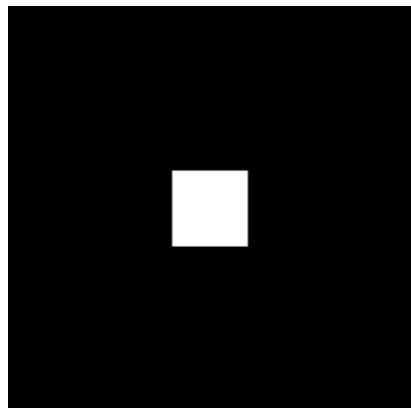
$$x[m,n](-1)^{m+n} \leftrightarrow \tilde{X}\left[k - \frac{M}{2}, l - \frac{N}{2}\right]$$

Other DFT issues (2)

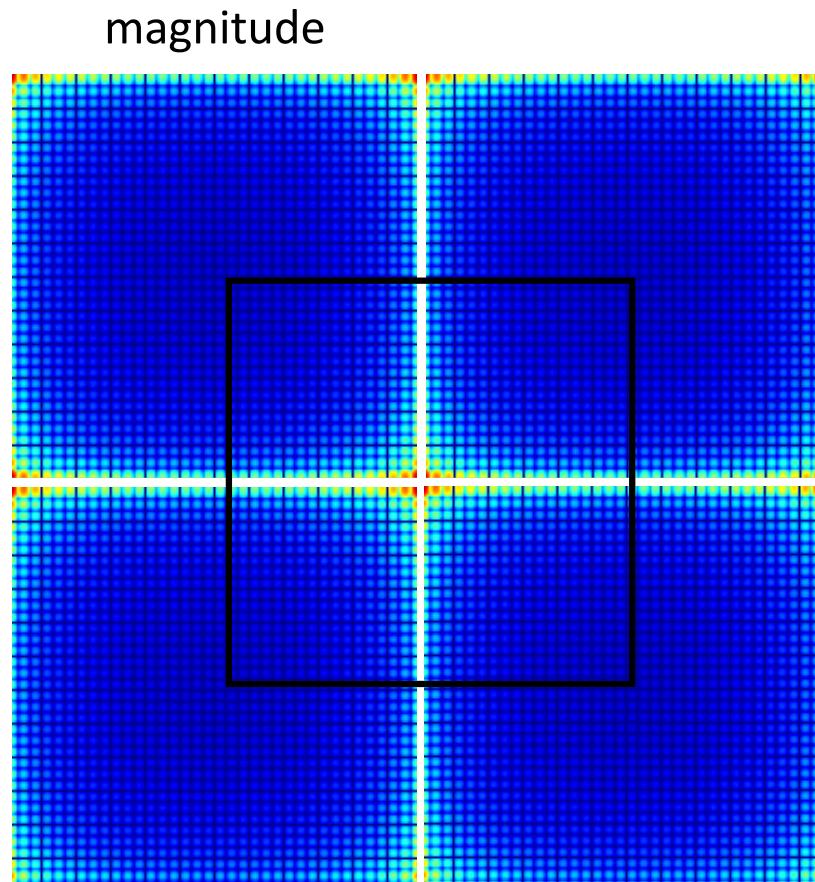
Non-centered representation



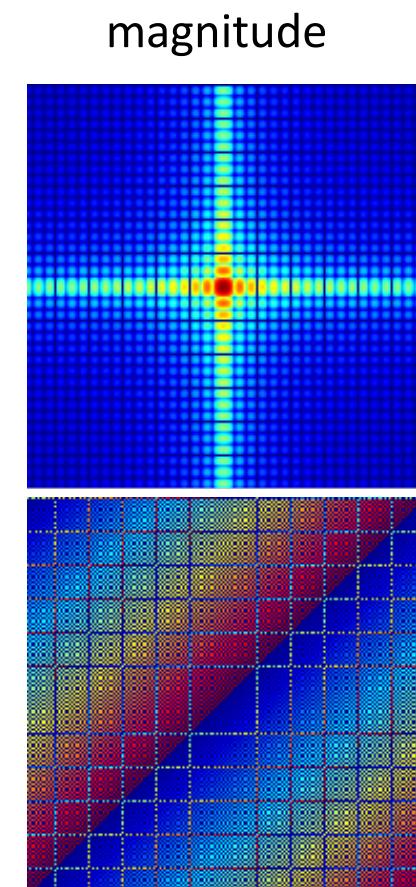
Other DFT issues (2)



Input image



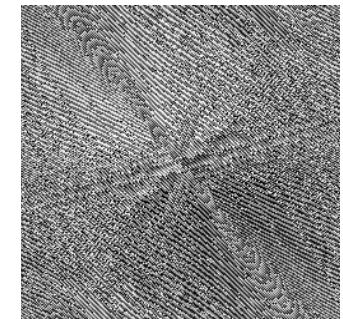
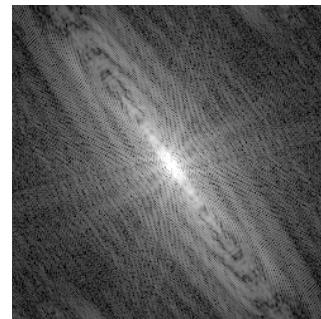
Centered representation



Other DFT issues (3)

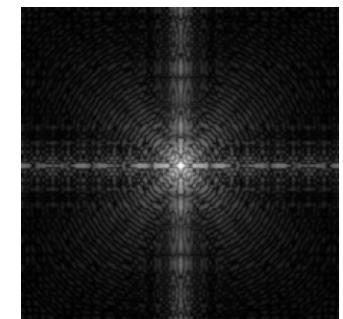
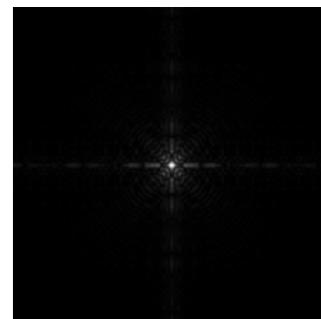
- **Magnitude and phase**

$$X[k,l] = |X[k,l]| e^{-j\varphi(k,l)}$$

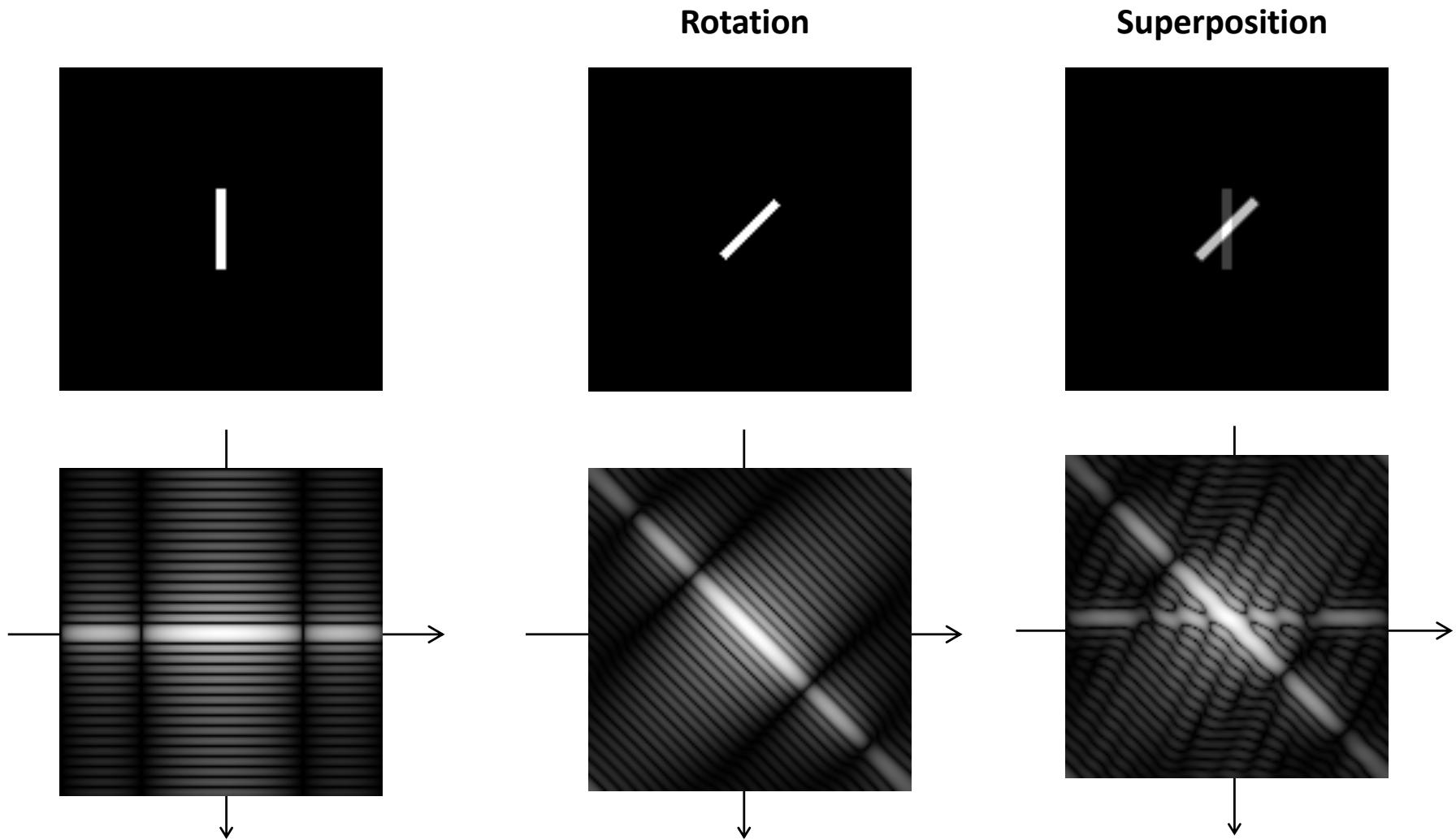


- **Logarithmic representation:**
 - To highlight the lowest values

$$\log(1 + |X[k,l]|)$$



Basic DFT transforms (1)

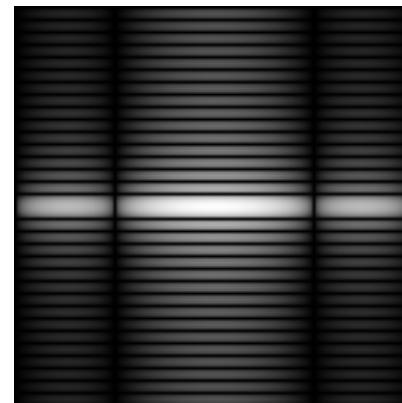


Basic DFT transforms (2)

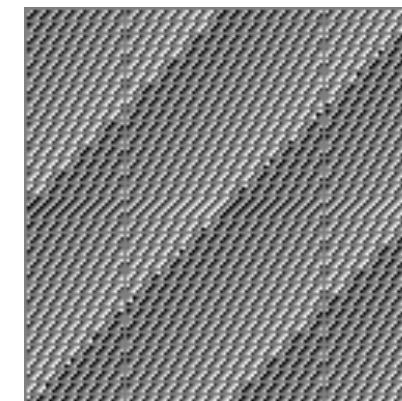
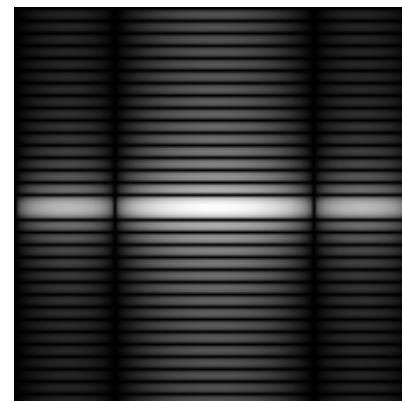
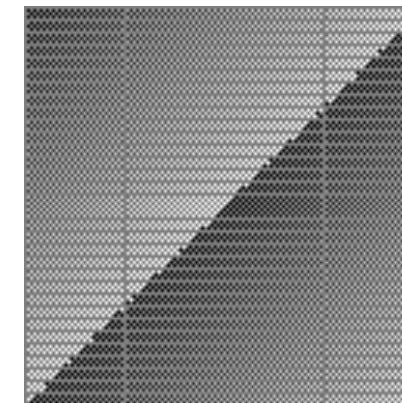
image



DFT magnitude



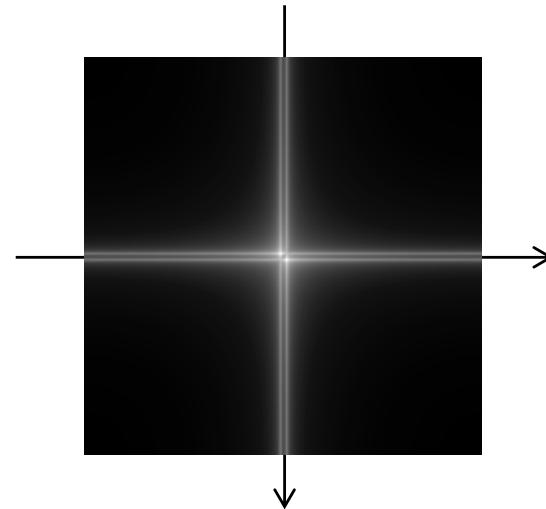
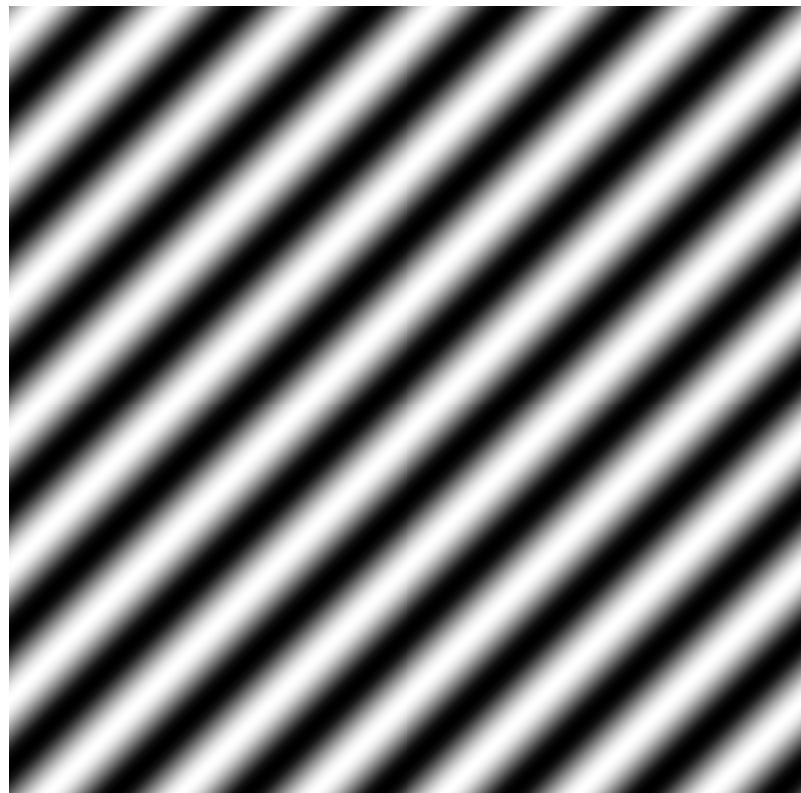
DFT phase



DFT windowing effect (1)

$$I[m, n] = \sin(2\pi F_1 m + 2\pi F_2 n)$$

$$F_1 = 0.008, F_2 = 0.008$$

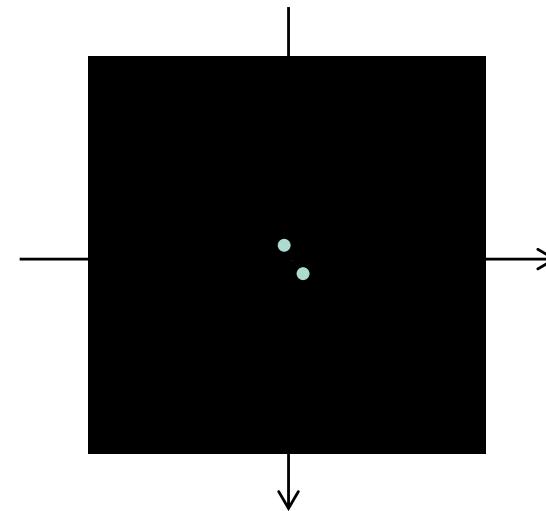
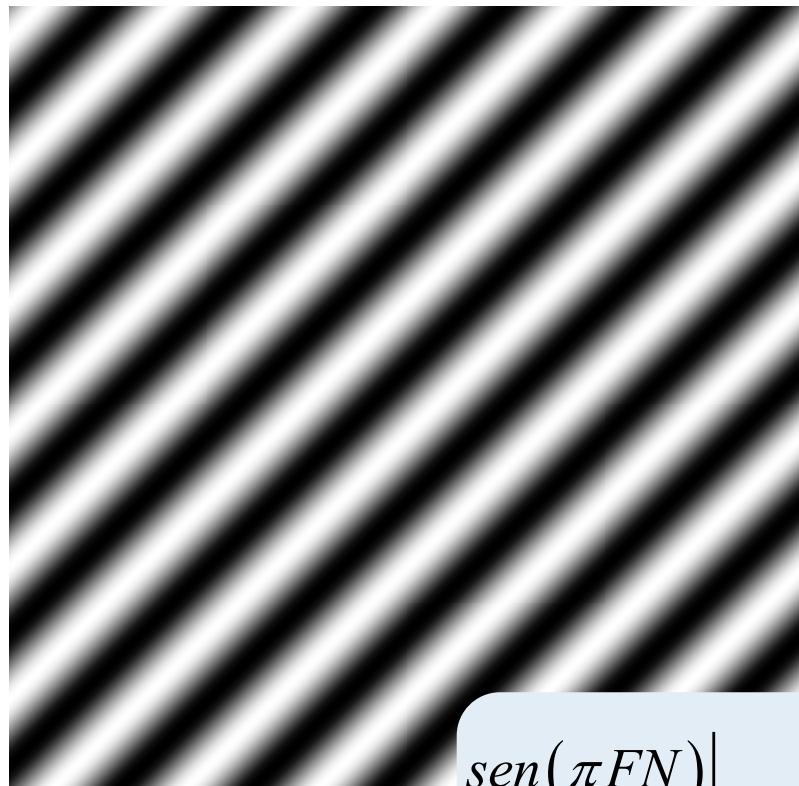


$$|F_1| \neq k / M, |F_2| \neq l / N$$

DFT windowing effect (2)

$$I[m, n] = \sin(2\pi F_1 m + 2\pi F_2 n)$$

$$F_1 = 0.01, F_2 = 0.01$$



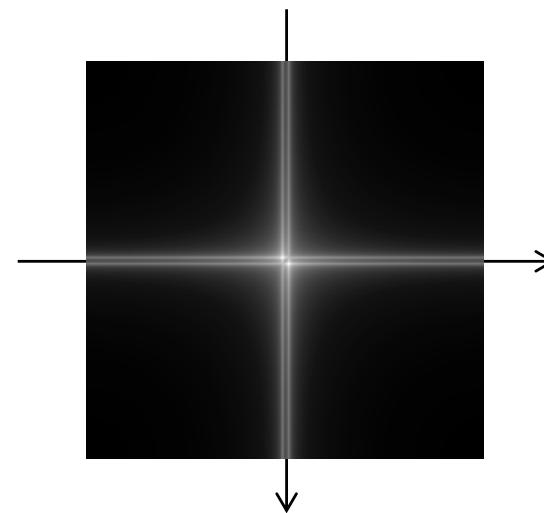
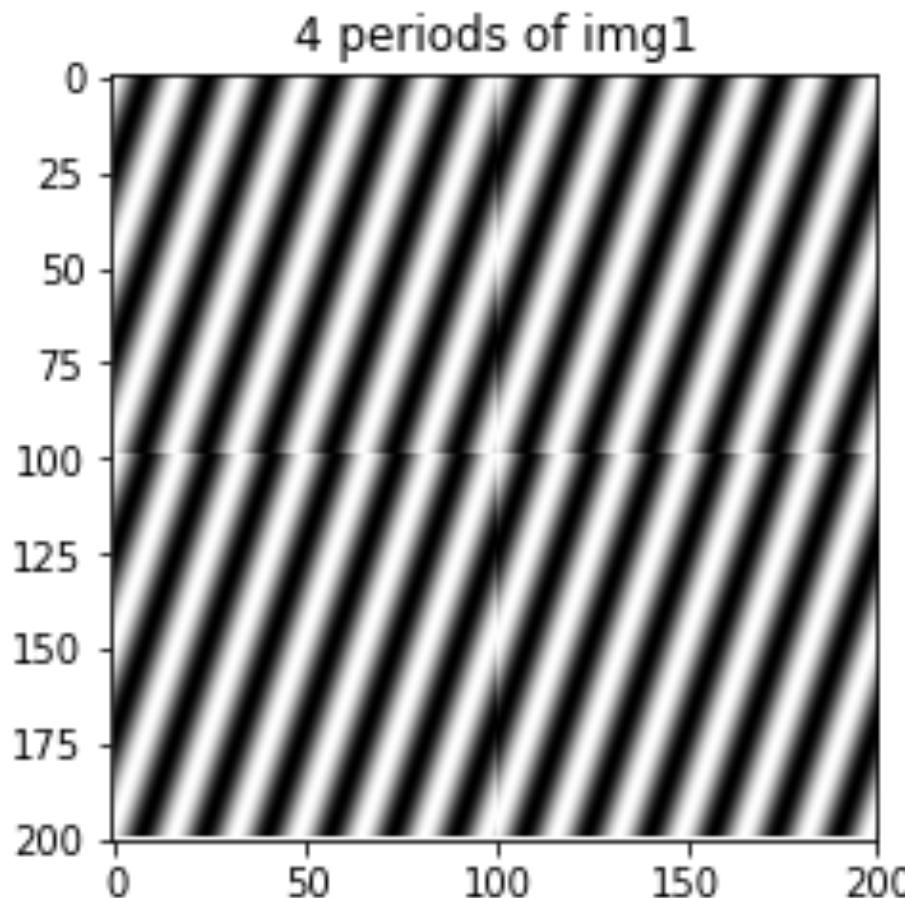
$$\left. \frac{\sin(\pi FN)}{\sin(\pi F)} \right|_{F=\frac{k}{N}} = N \delta[k]$$

$$|F_1| = k / M, |F_2| = l / N$$

DFT windowing effect (3)

$$I[m, n] = \sin(2\pi F_1 m + 2\pi F_2 n)$$

$$F_1 = 0.008, F_2 = 0.008$$

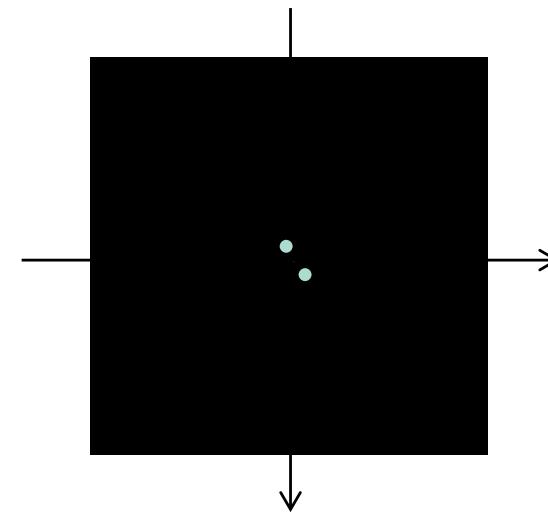
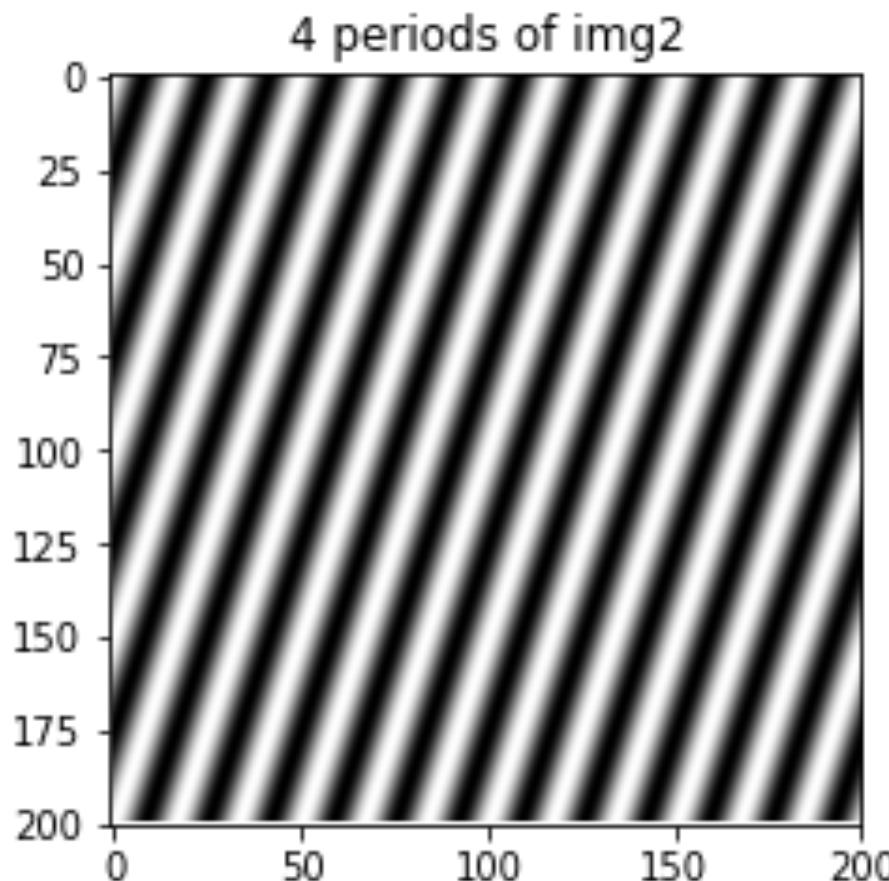


$$|F_1| \neq k / M, |F_2| \neq l / N$$

DFT windowing effect (4)

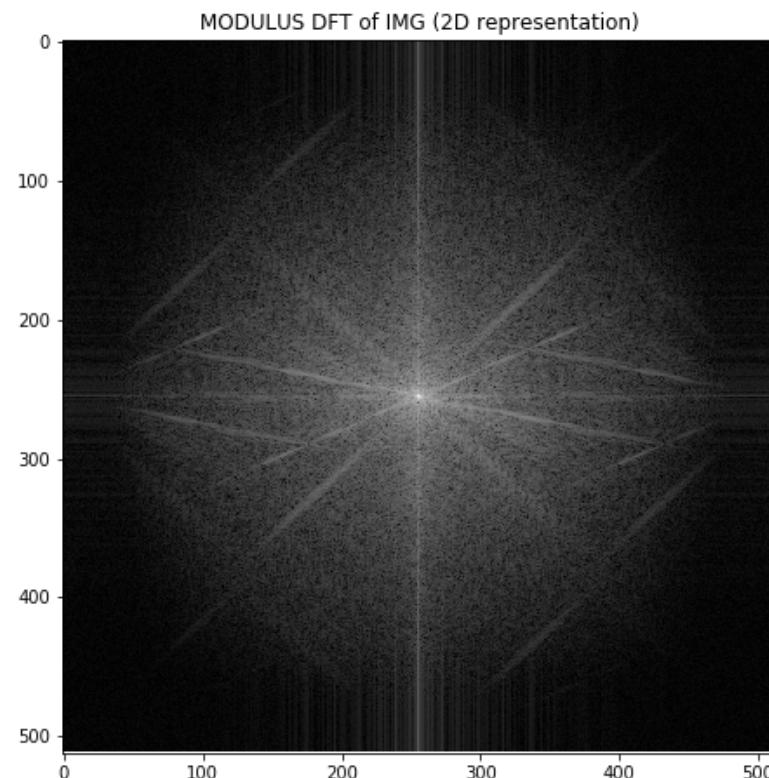
$$I[m, n] = \sin(2\pi F_1 m + 2\pi F_2 n)$$

$$F_1 = 0.01, F_2 = 0.01$$

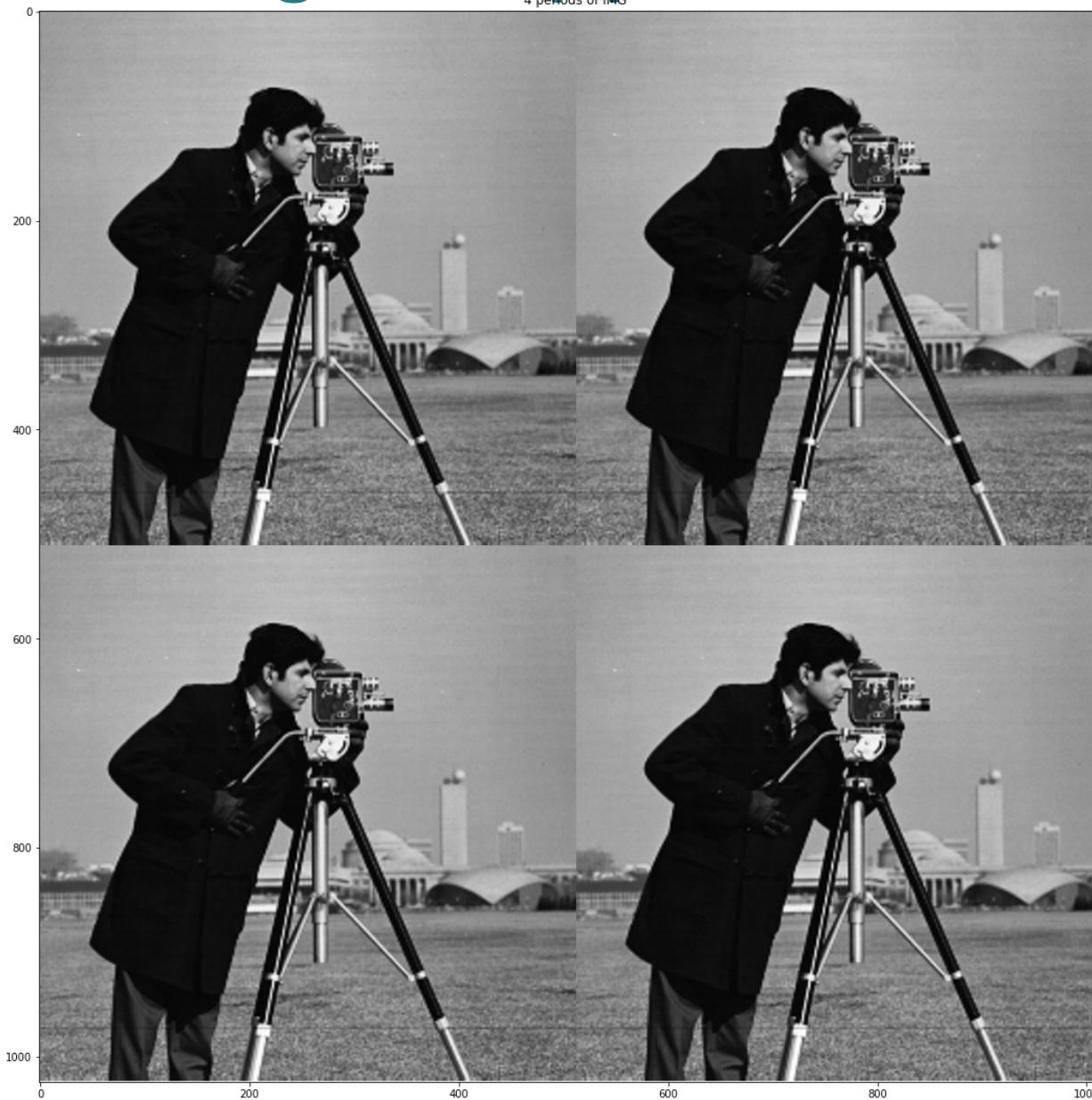


$$|F_1| = k / M, |F_2| = l / N$$

DFT windowing effect (5)

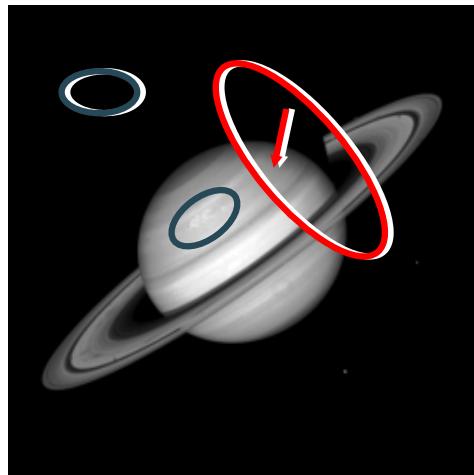


DFT windowing effect (6)

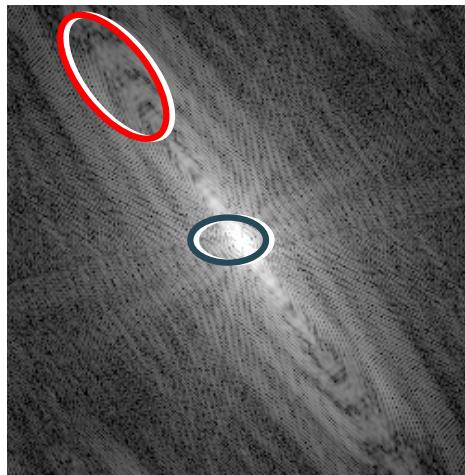


DFT transform examples (1)

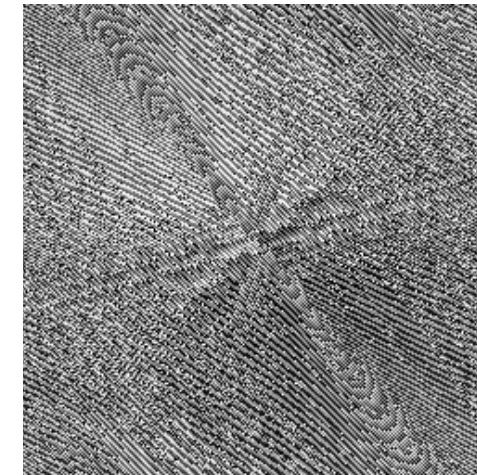
High frequency
Low frequency



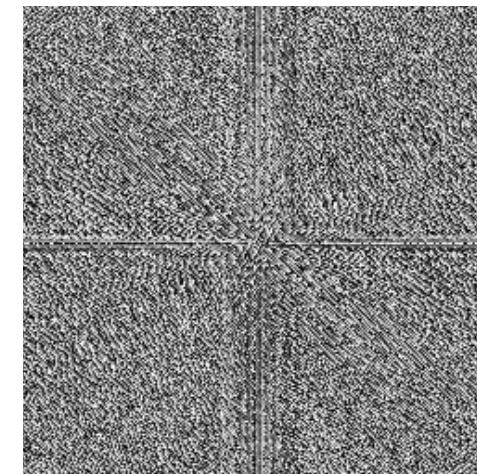
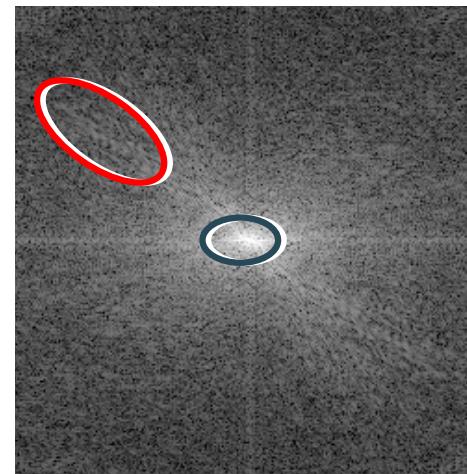
Signal



DFT magnitude



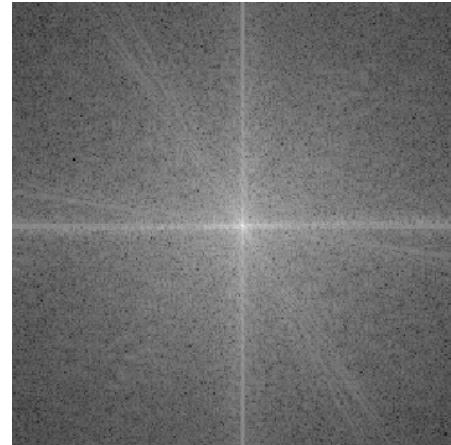
DFT phase



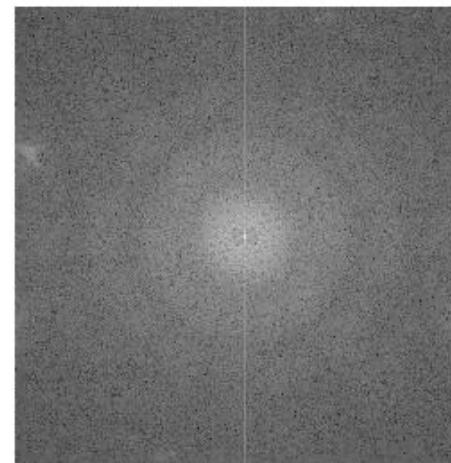
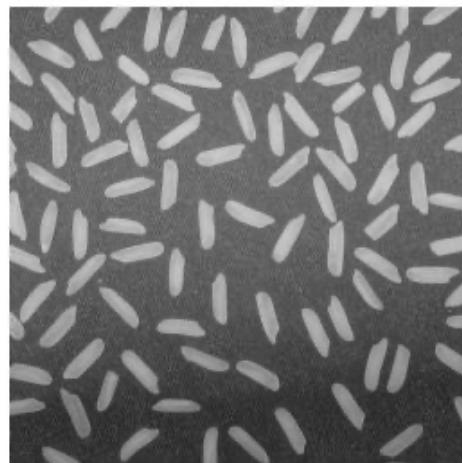
DFT transform examples (2)



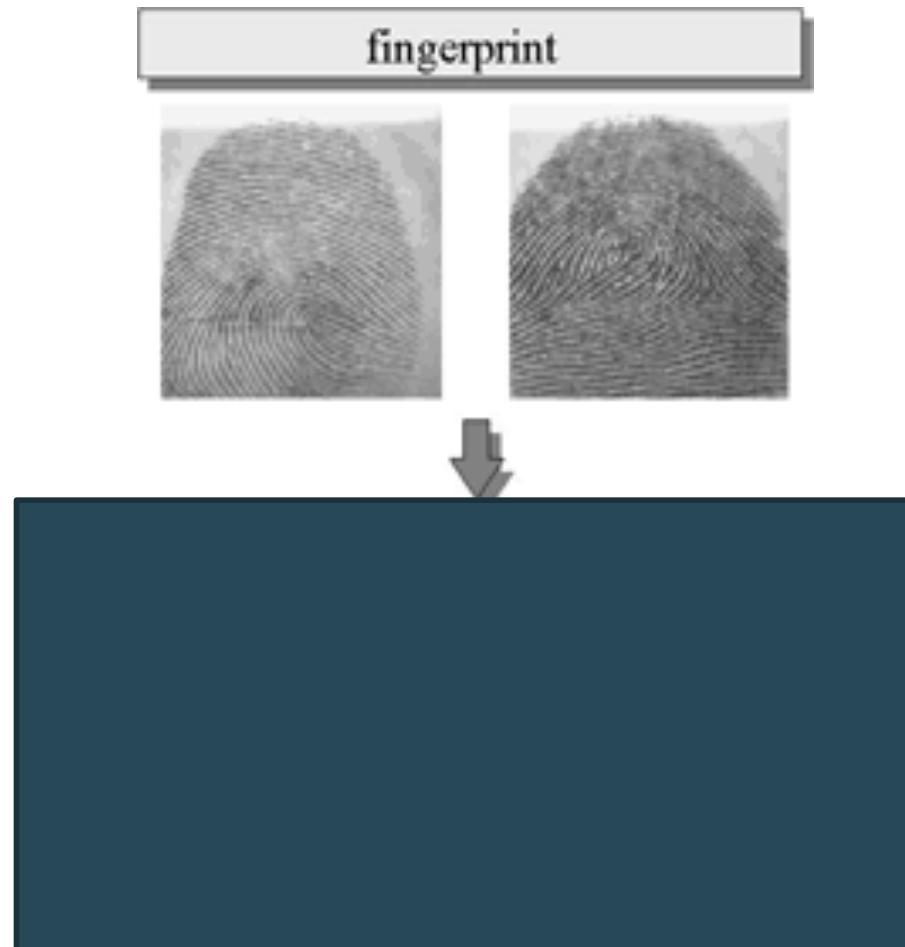
signal



DFT magnitude



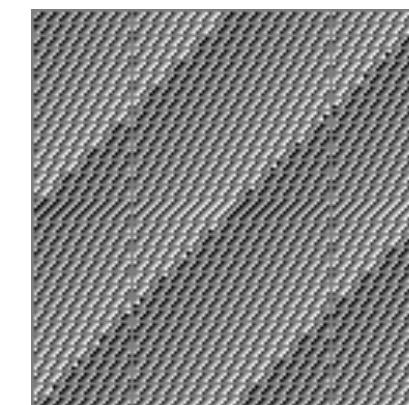
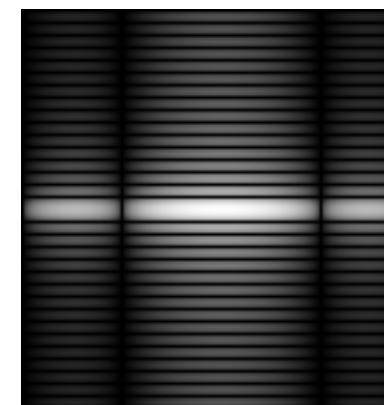
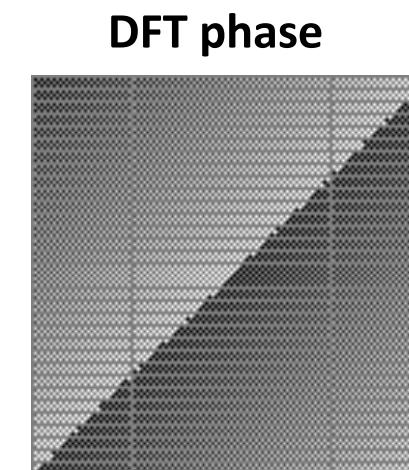
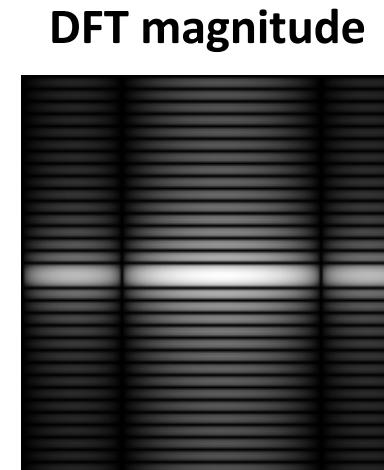
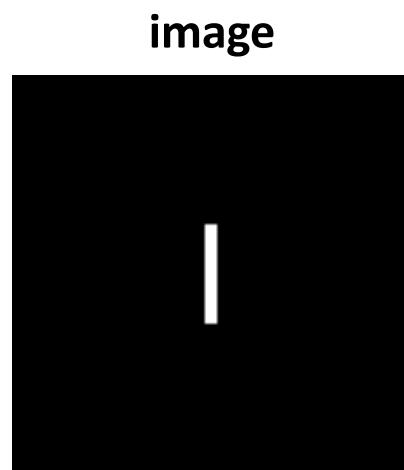
DFT transform examples (3)



Ej. Gonzalez - Woods

Relevance of the phase in images (1)

- Associated to position of objects



image

DFT magnitude

DFT phase

Relevance of the phase in images (2)

- Necessity of zero phase in processing

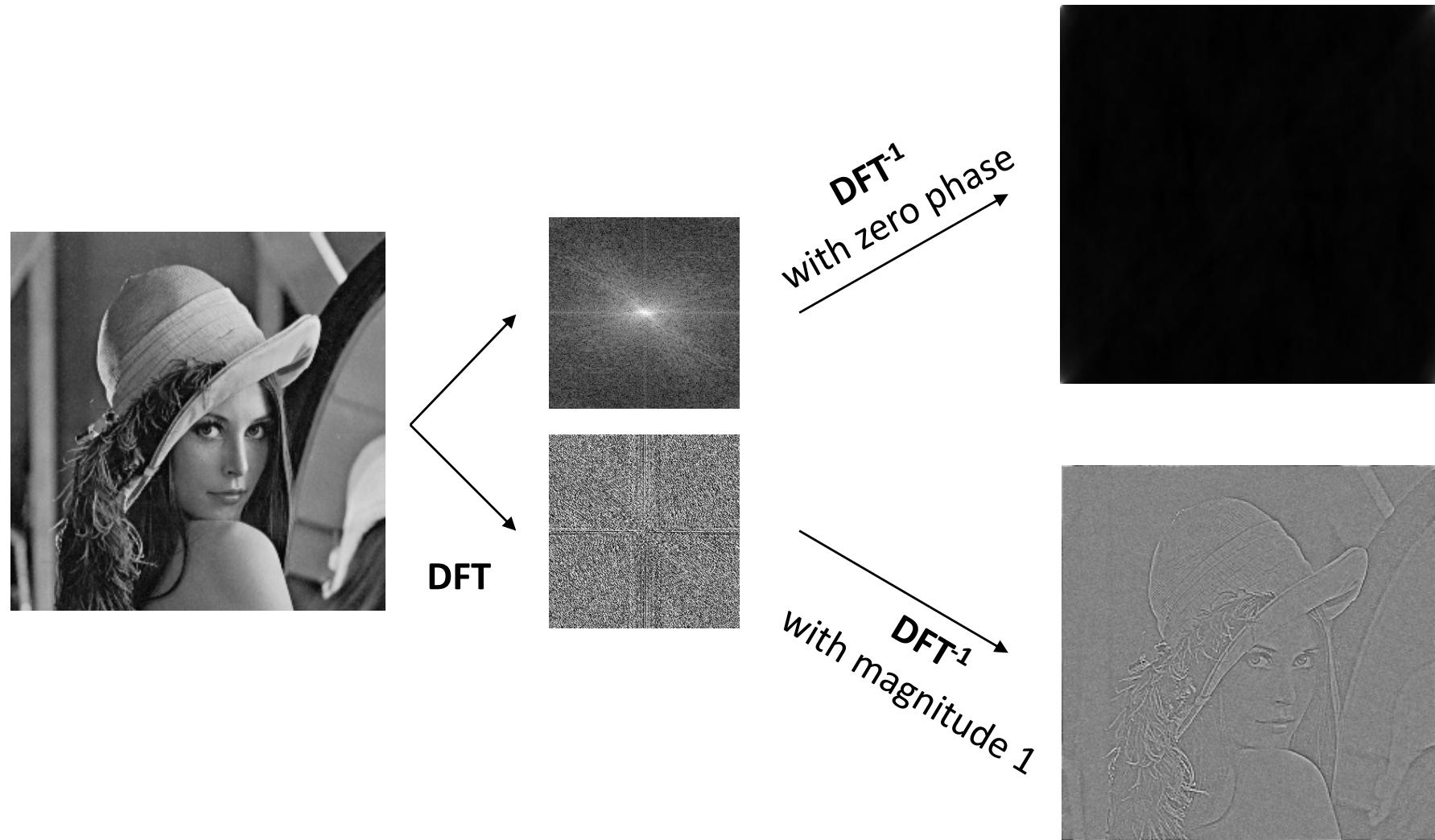


$x[m,n]$

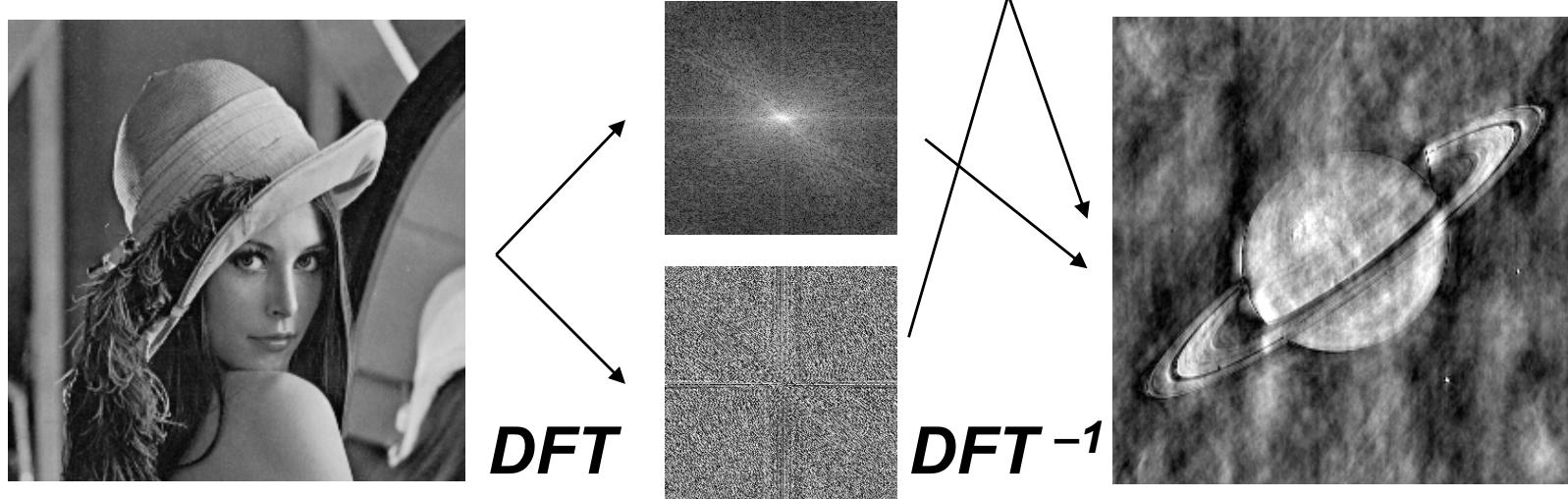
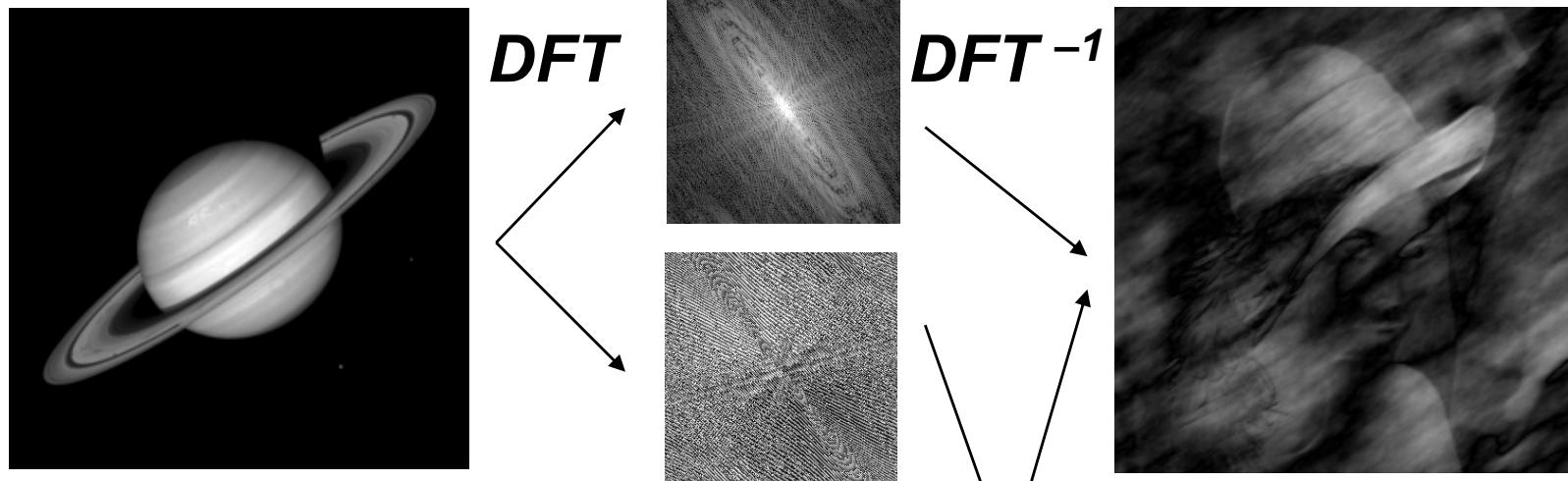
$$DFT^{-1} \{ DFT\{x[m,n]\} e^{-j2\pi \frac{30}{M}m} e^{-j2\pi \frac{20}{M}n} \}$$



Relevance of the phase in images (3)



Relevance of the phase in images (4)



Summary

- Introduction
 - Image model definition
 - Convolution
 - Correlation
- 2D Fourier Analysis
 - Definition, properties and basic transforms
- 2D Discrete Fourier Analysis
 - Definition, properties and basic transforms