



	<h1>Sampling</h1>
	<ul style="list-style-type: none"> ■ Signal Types ■ Why sampling ■ Sampling Theory ■ Nyquist-shannon theorem ■ Periodic Sampling ■ Aliassing ■ Anti-aliassing ■ Practical implications

	<h2>Harry Nyquist and Claude Elwood Shannon</h2>
	<div>  <p>Harry Nyquist (1889-1976)</p> </div> <div>  <p>Claude Elwood Shannon (1916 – 2001)</p> </div>

Signal Types

- Analog signals: continuous in time and amplitude
 - Example: voltage, current, temperature etc.
- Digital signals: discrete both in time and amplitude
- Discrete-time signal: discrete in time, continuous in amplitude

Why Sampling?

Analogue signals, in general, are continuous in time.

- For continuous signals to be processed by digital computers, the following two steps are needed:
 - (1) Convert the analogue signal to a digital signal by sampling.
 - (2) Process the digital signal.

Nyquist–Shannon sampling theorem

■ Sampling Theorem:

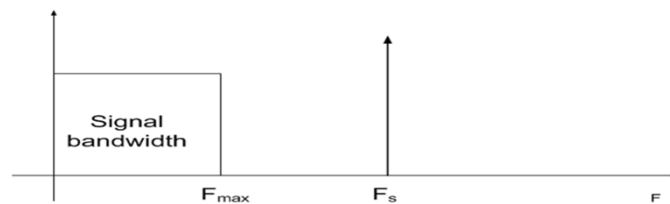
"When sampling a band-limited signal (e.g., converting from an analog signal to digital), the sampling frequency must be greater than twice the input signal bandwidth in order to be able to reconstruct the original perfectly from the sampled version"

$$\text{i.e., } F_s > 2 \cdot F_{\max}$$

Reference:

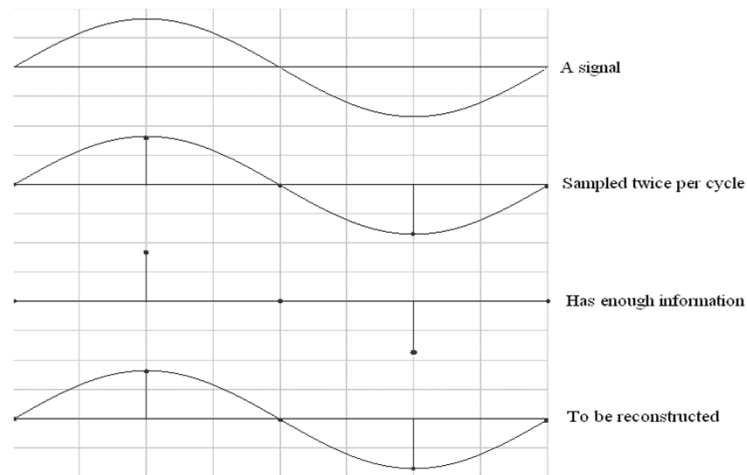
Claude E. Shannon in 1949 ("*Communication in the presence of noise*")

Sampling theorem Nyquist-shannon theorem



- The frequency $2 \cdot F_{\max}$ is called the Nyquist sampling rate.
- And F_{\max} is sometimes called the Nyquist frequency.
- The sampling theorem is considered to have been articulated by Nyquist in 1928 and mathematically proven by Shannon in 1949.
- The term "Nyquist Sampling Theorem" and "Shannon Sampling Theorem" are in fact the same sampling theorem.

Sampling theorem Nyquist-shannon theorem (Cnt)



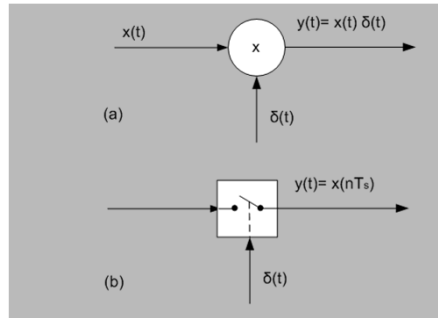
Periodic sampling

- Sampling is a continuous to discrete-time conversion
- Most common sampling is periodic

$$x[n] = x_c(nT) \quad -\infty < n < \infty$$

- T is the sampling period in second.
- $F_s = 1/T$ is the sampling frequency in Hz.
- Sampling frequency in radian-per-second $\Omega_s = 2\pi F_s$ rad/sec.
- Use $[]$ for discrete-time and $()$ for continuous time signals.

How can we sample?



Principle of sampling, (a) Multiplication and (b) Switching

SAMPLING

- Sampling is treated as a multiplication of the input analog signal with a periodic Delta, Dirac, or impulse function.
- So let's have a look in details at the impulse function.

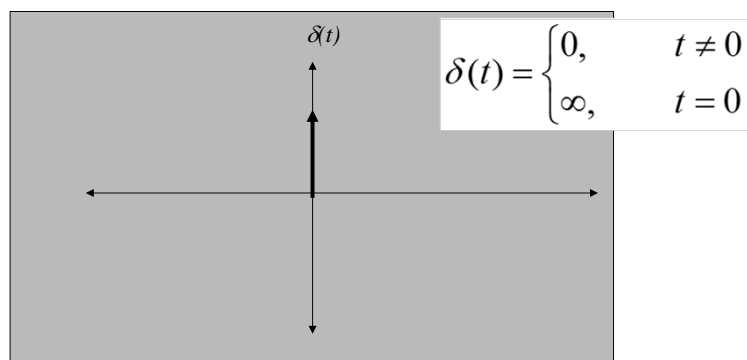
IMPULSE FUNCTION

- The impulse function is represented by $\delta(t)$ and is defined for every value of t with the following expression:

$$\delta(t) = \begin{cases} 0, & t \neq 0 \\ \infty, & t = 0 \end{cases}$$

IMPULSE FUNCTION

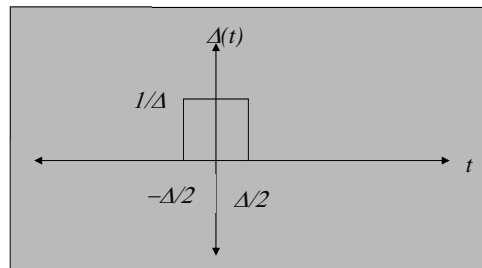
- The graphical representation of the delta function, $\delta(t)$ is given below



IMPULSE FUNCTION

- To obtain some insight into the δ function, we can consider the function $\Delta(t)$ as shown below.

$$\Delta(t) = \begin{cases} \frac{1}{\Delta}, & -\frac{\Delta}{2} \leq t \leq \frac{\Delta}{2} \\ 0, & \text{otherwise} \end{cases}$$



- The function has a height of $1/\Delta$ and a width of Δ , resulting in an area of 1.

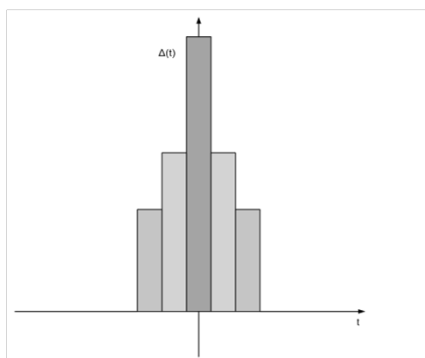
$$\int_{-\infty}^{\infty} \Delta(t) dt = \int_{-\frac{\Delta}{2}}^{\frac{\Delta}{2}} \frac{1}{\Delta} dt = \frac{1}{\Delta} \left[\frac{\Delta}{2} - \left(-\frac{\Delta}{2} \right) \right] = 1$$

IMPULSE FUNCTION

- The δ function is obtained from $\Delta(t)$ by the limiting process

$$\delta(t) = \lim_{\Delta \rightarrow 0} \Delta(t)$$

- The width of function goes to zero as its amplitude goes to infinity.
- The area under the curve to remain constant at unity.

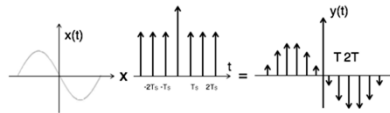


	<h2 style="text-align: center;">IMPULSE FUNCTION</h2>
	<ul style="list-style-type: none"> As Δ decreases in value, the function becomes narrower and higher. In the limit, as $\Delta \rightarrow 0$, $\Delta t \rightarrow \delta(t)$. The integral of $\Delta(t)$ is unity for any value of Δ. It follows that the integral of the δ function is also unity provided that the origin is included in the limits of integration. Otherwise, it is zero. $\int_a^b \delta(t) dt = \begin{cases} 1, & a \leq 0 \leq b \\ 0, & \text{otherwise} \end{cases}$

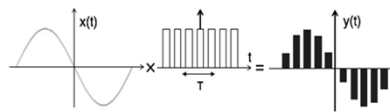
	<h2 style="text-align: center;">IMPULSE FUNCTION</h2>
	<ul style="list-style-type: none"> If we multiply the δ function by a constant, k, and then integrate the product over all time, we get: $\int_{-\infty}^{\infty} k\delta(t) dt = k \int_{-\infty}^{\infty} \delta(t) dt = k$ <ul style="list-style-type: none"> Multiplication of $\delta(t)$ by a constant is equivalent to choosing $\Delta(t)$ with a height of k/Δ. This results in an area equal to k.

Sampling Functions

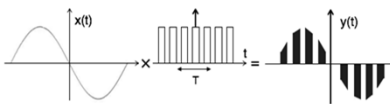
1. ideal impulse



2. flat-topped pulses



3. rectangular pulses



TIME-SHIFTED FUNCTIONS

- If the argument of the δ function is replaced with $t - \tau$, we see that the impulse still occurs when the argument takes on the value of zero (i.e., when $t - \tau = 0$ or $t = \tau$).

$$\delta(t - \tau) = \begin{cases} 0, & t - \tau \neq 0 \\ \infty, & t - \tau = 0 \end{cases}$$

TIME-SIFTING PROPERTY

- If we integrate the product of the δ function and any time function, we can observe the sifting property of δ function

$$\int_{-\infty}^{\infty} f(t)\delta(t)dt = \int_{-\infty}^{\infty} f(0)\delta(t)dt = f(0)\int_{-\infty}^{\infty} \delta(t)dt$$
$$= f(0).$$

TIME-SIFTING PROPERTY

- So all the values of $f(t)$ are sifted out except the value at the origin, $f(0)$.

$$\int_{-\infty}^{\infty} f(t)\delta(t)dt = f(0)$$

TIME-SIFTING PROPERTY

- If we use the time-shifted δ function, we obtain:

$$f(\tau) = \int_{-\infty}^{\infty} f(t) \delta(t - \tau) dt$$

SAMPLING

- We can describe our sampling function, $s(t)$, as the sum of all the individual impulse functions:

$$s(t) = \delta(t - \infty) + \dots + \delta(t - 2T_s) + \delta(t - T_s) + \delta(t) + \delta(t + T_s) + \delta(t + 2T_s) + \dots + \delta(t + \infty)$$

- Shorter version of the previous expression is:

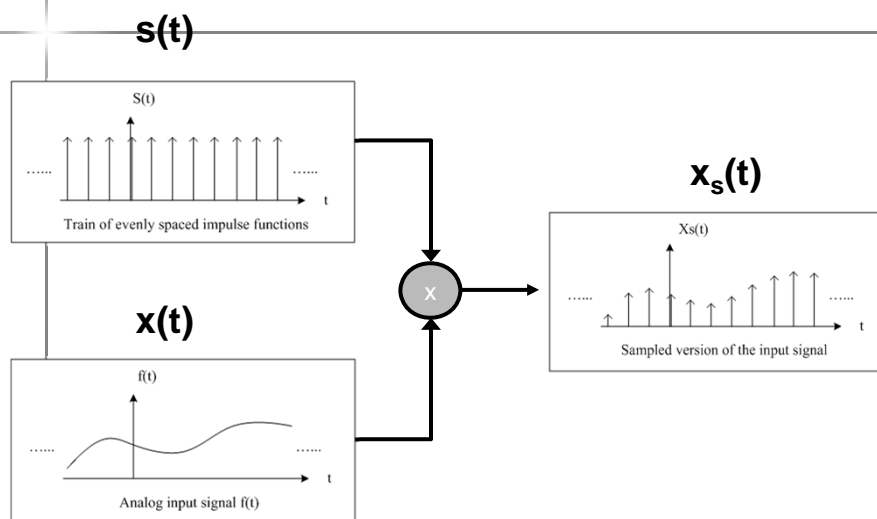
$$s(t) = \sum_{k=-\infty}^{k=\infty} \delta(t - kT_s)$$

SAMPLING

- If we multiply these by our analog input signal, $x(t)$, we obtain a train of pulses whose amplitudes are equal to the amplitude of $x(t)$ at that moment in time.
- Mathematically, the output sampled waveform, $x_s(t)$, is just the multiplication of $s(t)$ with the input analog signal $x(t)$:

$$x_s(t) = \sum_{k=-\infty}^{k=\infty} x(t) \delta(t - kT_s)$$

SAMPLING



SAMPLING

- Using the representation of the sampled version of CT signal $x(t)$

$$x_s(t) = \sum_{k=-\infty}^{k=\infty} x(t) \delta(t - kT_s)$$

- $x(t) = x(kT_s)$ at the moment $t = kT_s$, therefore:

$$x_s(t) = \sum_{k=-\infty}^{k=\infty} x(kT_s) \delta(t - kT_s)$$

SAMPLING

- By assuming the $kT_s = k$ we can write:

$$x_s(t) = \sum_{k=-\infty}^{k=\infty} x(k) \delta(t - k)$$

- Actually, the above equation is not correct, t represents time and k is just an integer
- We have already mentioned that the sampled signal $x_s(t)$ exists only as an ordered sequence of numbers – samples
- It is common to use index n instead of time t to indicate this fact.

SAMPLING

- The equation for the sampled signal therefore turns into:

$$x_s(t) = x(n) = \sum_{k=-\infty}^{k=\infty} x(k)\delta(n-k)$$

↖ No need for index s here, as x(n)
already assumes sampled signal

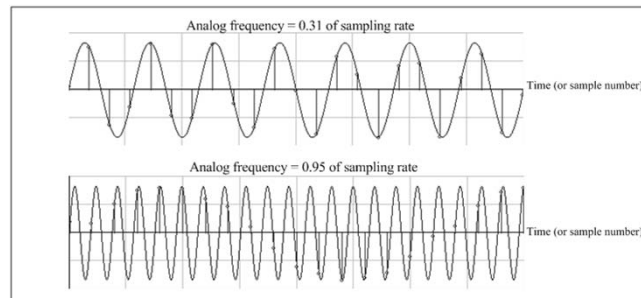
SAMPLING

- If we use index n instead of (continuous) time t in the previous equation, the Impulse function $\delta(t)$ mutates into it's discrete version $\delta(n)$.
- This is usually called unit impulse and is defined as:

$$\delta(n) = \begin{cases} 1, & n = 0 \\ 0, & \textit{otherwise} \end{cases}$$

2 Aliasing

- if a signal contains frequencies above the Nyquist frequency.
 - Loss of information
 - Introduces wrong information
 - Loss of phase information (phase shift)



Mathematical Explanation of aliasing

A continuous-time sinusoidal signal can be express as:

$$x(t) = \sin(2\pi ft + \phi)$$

Sampling $x(t)$ ($t = nT$) will result in discrete-time sequence $x[n]$ as:

$$x[n] = \sin(2\pi fTn + \phi)$$

Exploiting the periodicity of the sine wave, we can write:

$$x[n] = \sin(2\pi fTn \pm 2\pi m + \phi) = \sin\left(2\pi T\left(f \pm \frac{m}{Tn}\right)n + \phi\right)$$

Replacing m/n by k will lead to:

$$x[n] = \sin[2\pi T(f \pm kF_s)n + \phi]$$

These equation shows that we cannot distinguish between a signal with a frequency f and a signal with a frequency $(f \pm kF_s)$

	<h2>Can aliasing be avoided?</h2>
	<ul style="list-style-type: none">■ The answer is yes and no!■ Yes, by using a Low-pass analog filter (anti-aliasing filter) before sampling.■ No, in practical situation you can only reduce aliasing because of the characteristic of the anti-aliasing filter (no filter is perfect !).

	<p>End of Chapter</p>