$\begin{array}{c} {\rm EE413} \\ {\rm Lab~005} \end{array}$ the Operational Amplifier

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Abstract

"This lab is meant to show the practical use of the operational amplifier in analog circuit design. Several common circuit configurations will be discussed."

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1 Circuit prototyping setup

The circuit was build on a solderless breadboard, using through-hole parts. A classic 741 op amp was used with a \pm 15V power supply.

2 Inverting DC Amplifier

2.1 Theory

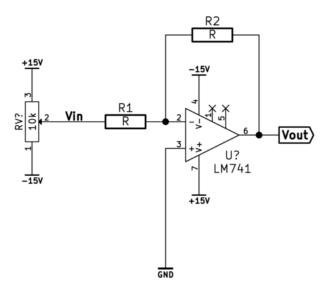


Figure 1: Inverting DC amplifier

The basic topology for an inverting amplifier is shown in Figure 1. Gain A_v , can be expressed as a ratio of the feedback impedance to the input impedance. Op amp action makes the negative input appear as a "virtual earth" summing node. The voltage drops across the resistors scale linearly with their value, and since the op amp compensates to ensure equality in the summing junction, the net effect is an amplified and inverted output.

$$A_v = \frac{R_2}{R_1} \tag{1}$$

The circuit gain for ideal components is therefore; For $R_2=100k\Omega$:

$$A_v = \frac{V_{out}}{V_{in}} = -\frac{R_2}{R_1}$$

$$= \frac{100k\Omega}{10k\Omega} = 10\times$$
(2)

$$=\frac{100k\Omega}{10k\Omega} = 10\times\tag{3}$$

$$= 20 \times \log \frac{10}{1} = 20dB \tag{4}$$

For $R_2 = 10k\Omega$:

$$A_v = \frac{V_{out}}{V_{in}} = -\frac{R_2}{R_1} \tag{5}$$

$$=\frac{10k\Omega}{10k\Omega} = 10\times\tag{6}$$

$$= 20 \times \log \frac{1}{1} = 0dB \tag{7}$$

In both cases, the signal phase is inverted 180°.

2.2 Measurements

Measured values for the test setup.

$\overline{U_{in} (V)}$	U_{out} (V)	$Av(\times)$
-0.103	+1.087	-10.35
-1.008	+10.236	-10.15
+1.004	-10.104	-10.06

Table 1: $R_2 = 100k\Omega$

$$\frac{U_{in} \text{ (V)} \quad U_{out} \text{ (V)} \quad Av \text{ (×)}}{-0.1003 \quad +0.112 \quad -1.116}$$
$$-1.000 \quad +1.038 \quad -1.038$$
$$+1.005 \quad -1.027 \quad -1.022$$

Table 2: $R_2 = 10k\Omega$

For $R_2 = 100k\Omega$, the actual measured in circuit values of R_2 and R_1 were $119k\Omega$ and $11.7k\Omega$, respectively. Calculated circuit gain for non-ideal, real components;

$$A_v = \frac{V_{out}}{V_{in}} = -\frac{R_2}{R_1}$$

$$= \frac{119k\Omega}{11.7k\Omega} = 10.17 \times$$
(9)

$$=\frac{119k\Omega}{11.7k\Omega} = 10.17\times\tag{9}$$

$$= 20 \times \log \frac{\frac{119k\Omega}{11.7k\Omega}}{1} = 20.15dB \tag{10}$$

For $R_2=10k\Omega,$ the actual measured in circuit values of R_2 and R_1 were $12.17k\Omega$ and $11.7k\Omega$, respectively. Calculated circuit gain for non-ideal, real components;

$$A_{v} = \frac{V_{out}}{V_{in}} = -\frac{R_{2}}{R_{1}}$$

$$= \frac{12.17k\Omega}{11.7k\Omega} = 1.04 \times$$

$$= 20 \times \log \frac{\frac{12.17k\Omega}{11.7k\Omega}}{1} = 0.34dB$$
(11)

$$= \frac{12.17k\Omega}{11.7k\Omega} = 1.04 \times \tag{12}$$

$$= 20 \times \log \frac{\frac{12.17k\Omega}{11.7k\Omega}}{1} = 0.34dB \tag{13}$$

Inverting AC Amplifier 3

Theory 3.1

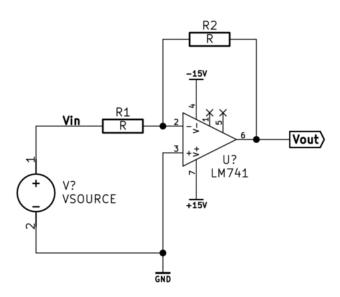


Figure 2: Inverting AC amplifier

The basic topology for an inverting AC amplifier is shown in Figure 2.

$$A_v = 1 + \frac{R_2}{R_1} \tag{14}$$

The circuit gain for ideal components is; For $R_2 = 100k\Omega$:

$$A_v = \frac{V_{out}}{V_{in}} = -\frac{R_2}{R_1} \tag{15}$$

$$= \frac{100k\Omega}{10k\Omega} = 10\times\tag{16}$$

$$= 20 \times \log \frac{10}{1} = 20dB \tag{17}$$

For $R_2 = 10k\Omega$:

$$A_v = \frac{V_{out}}{V_{in}} = -\frac{R_2}{R_1} \tag{18}$$

$$=\frac{10k\Omega}{10k\Omega} = 10\times\tag{19}$$

$$=20 \times \log \frac{1}{1} = 0dB \tag{20}$$

In both cases, the signal phase is inverted 180° .

3.2 Measurements

Oscilloscope photos in figure 3 and figure 4 show the amplifier input on channel one the amplifier output on channel two. Channel two volts/div is set to compensate for the high impedance 10:1 probe setting.

3.3 Oscilloscope shots

4 Non-inverting DC Amplifier

4.1 Theory

The basic topology for an non-inverting DC amplifier is shown in Figure 5. Gain A_v , is set by the attenuation-factor of the circuit in the feedback-loop. A fraction of the output is fed back, causing the op amp to compensate and in effect amplify.

$$A_v = 1 + \frac{R_2}{R_1} \tag{21}$$

4.2 Measurements

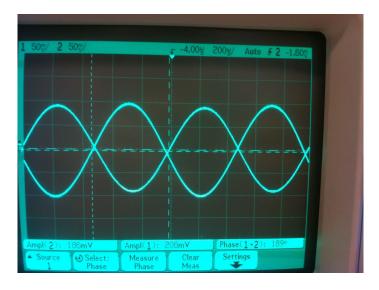


Figure 3: Inverting AC amplifier - 20 dB gain

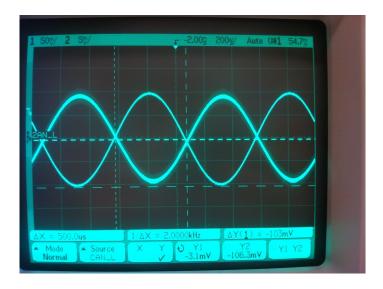


Figure 4: Inverting AC amplifier - unity gain

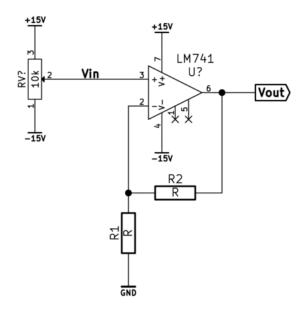


Figure 5: Non-inverting DC amplifier

U_{in} (V)	U_{out} (V)	$Av(\times)$
+0.1007	+0.2164 2	.15
+1.002	$+2.048\ 2$.04
-1.005	-2.03 2	.019

Table 3: $R_2 = 10k\Omega$

$\overline{U_{in}}$ (V)	U_{out} (V)	Av(x)
+0.1009	+1.178	11.67
+1.1013	+11.3	11.15
-1.004	-11.09	11.05

Table 4: $R_2 = 100k\Omega$

5 Non-inverting AC Amplifier

5.1 Measurements

 $\begin{array}{ll} \text{Input signal amplitude} = \\ \text{Output signal amplitude} = \\ \text{Measured amplification} = \\ \text{Measured phase} = \\ \text{Theoretical amplification} = \text{Theoretical phase} = \\ \end{array}$

6 Active full wave rectifier

Active rectifier does not suffer from the "deadzone" when the signal is too small to turn on the rectifying diede. The on amp components for the diede forward

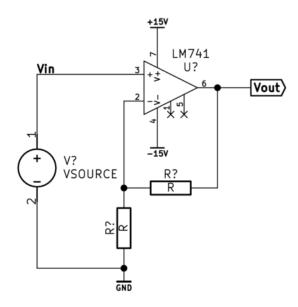


Figure 6: Non-inverting AC amplifier

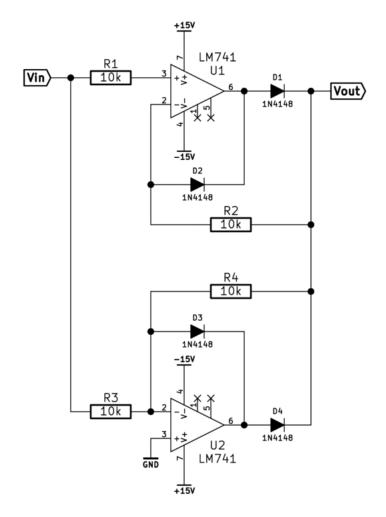


Figure 7: Active full wave rectifier