

Due Saturday April 2nd, at 12:00pm CST.

**Your name:**

**Your UTD ID:**

**Important points:**

- This is a take-home, open lecture/notes exam. You are not allowed to look at your homework assignments.
- You are not allowed to talk to anybody about this exam until you submit it.
- There are 5 problems. The last one is MATLAB-based and will probably take a bit longer to complete. Use your time wisely!
- Think before you start solving the problems! If you are spending too much time on any of the problems, you're probably on the wrong track.
- You **cannot** use MATLAB except in the last question.
- I will not be responding to emails during the exam time. Explain your solutions clearly and provide all necessary details including the listings of your MATLAB scripts.
- Your answers should be combined into a single PDF file and submitted via eLearning. To submit, click on the midterm assignment in the **Exams** folder and use the ASSIGNMENT SUBMISSION feature to browse your computer and upload your file. Note that I will not be accepting submissions via email after the deadline. **Don't wait until the last minute!**

**Good luck!**

1. [15 points] Consider the feedback system shown in Figure 1 where  $G(s)$  is the nominal plant model and  $\Delta(s)$  is stable. Assume that  $G(s)$ ,  $K(s)$ , and  $\Delta(s)$  are all SISO.
  - (a) The dashed box represents an uncertain model  $\hat{G}(s)$  that depends on both  $G(s)$  and  $\Delta(s)$ . What is the set of models  $\mathcal{A}$  corresponding to this block diagram?
  - (b) What can you conclude about the classical gain margins if the feedback system is stable for all  $\|\Delta\|_\infty < 0.5$ ?
  - (c) Find a necessary and sufficient condition for  $K(s)$  to stabilize all  $\hat{G}(s) \in \mathcal{A}$ . Briefly describe a proof that your condition is sufficient for  $K(s)$  to achieve robust stability. You do not need to prove necessity.

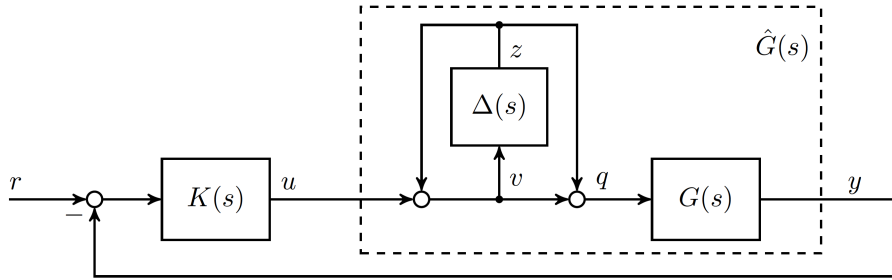


Fig. 1: Feedback system

2. [20 points] Consider a string of two vehicles as shown in Figure 2. The leading (first) vehicle is tracking a reference  $r_1$  and the following (second) vehicle is tracking a reference  $r_2 = x_1 - \delta$  to maintain a distance of  $\delta$  with the first vehicle. The second vehicle uses a radar device to measure the distance to the first vehicle and compute the error  $e_2 = r_2 - x_2$ . Let each vehicle be modeled by the transfer function  $G(s)$  and assume that both vehicles use the same control law  $K(s)$ . Figure 3 shows the feedback diagram for the two vehicle string.

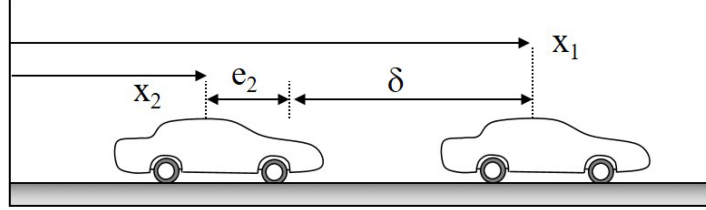


Fig. 2: String of two vehicles

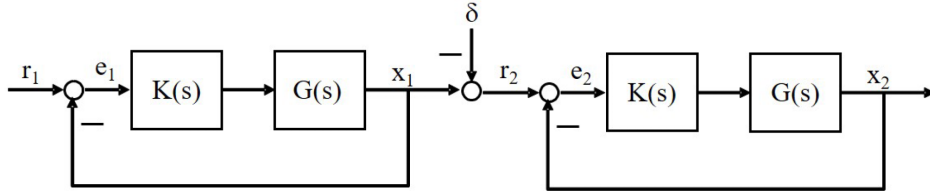


Fig. 3: Feedback diagram for two-vehicle system

- Compute the transfer function from: (i)  $r_1$  to  $e_1$ , (ii)  $r_1$  to  $x_1$ , (iii)  $r_2$  to  $e_2$ , and (iv)  $r_2$  to  $x_2$ . Your answers should be expressed in terms of the sensitivity  $S(s) = \frac{1}{1+G(s)K(s)}$  and complementary sensitivity  $T(s) = \frac{G(s)K(s)}{1+G(s)K(s)}$ .
- What is the transfer function from the first vehicle's reference  $r_1$  to the tracking error for the second vehicle  $e_2$ ? Note: By linearity you may assume  $\delta = 0$  in this calculation.
- The goal is for the first vehicle to track the reference command  $r_1$ . In addition, the second vehicle should achieve a much smaller tracking error than the first vehicle, i.e., we would like  $|e_2|$  to be much smaller than  $|e_1|$ . Use your results from the previous parts to express these objectives in terms of  $S(s)$  and  $T(s)$ . You may assume that the reference command  $r_1$  mainly consists of low frequency content.
- Is it possible to achieve the two goals in part (c)? If yes, describe how you would design  $K(s)$  to achieve these goals. If no, then describe a constraint that prevents you from achieving both goals.

3. [15 points] Let  $A \in \mathbb{C}^{n \times m}$  where  $n > m$ . The singular value decomposition (SVD) of  $A$  is given by

$$A = U \Sigma V^* = \begin{bmatrix} U_1 & U_2 \end{bmatrix} \begin{bmatrix} \hat{\Sigma} \\ 0_{n-m, m} \end{bmatrix} V^*$$

where  $U$  and  $V$  are unitary matrices and  $\hat{\Sigma} = \text{diag}(\sigma_1, \dots, \sigma_m)$ . Assume  $A$  is full rank and hence  $\sigma_i > 0$  for  $i = 1, \dots, m$ .

- (a) What is the SVD of  $(A^*A)^{-1}A^*$ ?  
 (b) Show that for any vectors  $x \in \mathbb{C}^m$  and  $b \in \mathbb{C}^n$ ,

$$\|Ax - b\|_2^2 = \|\hat{\Sigma} V^* x - U_1^* b\|_2^2 + \|U_2^* b\|_2^2. \quad (1)$$

- (c) Use Equation (1) to specify the vector  $x$  that solves

$$\min_x \|Ax - b\|_2$$

How is the solution related to the matrix in part (a)?

4. [20 points] Consider the following set of models:

$$\mathcal{I} := \left\{ \frac{G(s)}{1 + w(s)\Delta(s)} \mid \Delta \text{ stable, } \|\Delta\|_\infty < 1 \right\}$$

$G(s)$  is SISO and  $w(s)$  is a stable, proper rational transfer function.

- Draw a block diagram showing the structure of any  $G_p \in \mathcal{I}$ . Your diagram should only include blocks for  $G$ ,  $w$ , and  $\Delta$ .
- Consider the feedback diagram show in Figure 4. Assume the controller  $K$  stabilizes the nominal system  $G$ . Find a necessary and sufficient condition for  $K$  to stabilize all  $G_p \in \mathcal{I}$ . Provide a proof of sufficiency, i.e., prove that  $K$  achieves robust stability if your condition is satisfied. You do not need to prove necessity.

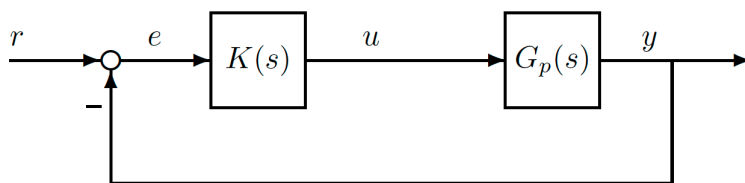


Fig. 4: Feedback loop

5. [30 points] Figure 5 shows the block diagram for a plane autopilot system with state-space representation

$$\begin{bmatrix} \dot{x} \\ \dot{x}_c \end{bmatrix} = \begin{bmatrix} A - BD_c C & BC_c \\ -B_c C & A_c \end{bmatrix} \begin{bmatrix} x \\ x_c \end{bmatrix} + \begin{bmatrix} BD_c \\ B_c \end{bmatrix} r := A_{cl} \begin{bmatrix} x \\ x_c \end{bmatrix} + B_{cl} r.$$

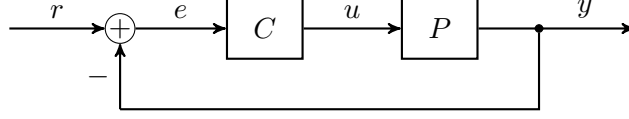


Fig. 5: Autopilot feedback system

The controller state matrices are given by

$$A_c := \begin{bmatrix} 0 & 0 \\ K_q a_q & 0 \end{bmatrix}, \quad B_c := \begin{bmatrix} K_a a_z & 0 \\ K_a K_q a_q & K_q a_q \end{bmatrix}, \quad C_c := [K_q \quad 1], \quad D_c := [K_a K_q \quad K_q]$$

where  $K_a = -0.0015$ ,  $K_q = -0.32$ ,  $a_z = 2$ , and  $a_q = 6$ . The plant state matrices are given by

$$A := \begin{bmatrix} Z_\alpha & 1 & Z_\delta & 0 \\ M_\alpha & 0 & M_\delta & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & -\omega^2 & -2\zeta\omega \end{bmatrix}, \quad B := \begin{bmatrix} 0 \\ 0 \\ 0 \\ \omega^2 \end{bmatrix}, \quad C := \begin{bmatrix} VZ_\alpha & 0 & VZ_\delta & 0 \\ 0 & 1 & 0 & 0 \end{bmatrix}, \quad D := \begin{bmatrix} 0 \\ 0 \end{bmatrix}$$

where  $V = 886.78$ ,  $\zeta = 0.6$ , and  $\omega = 113$ . The aerodynamic coefficients have significant uncertainty which we model as:

$$\begin{aligned} Z_\alpha &= -1.3046(1 + k\delta_1), & Z_\delta &= -0.2142(1 + k\delta_2) \\ M_\alpha &= 47.7109(1 + k\delta_3), & M_\delta &= -104.8346(1 + k\delta_4). \end{aligned}$$

Each  $\delta_i$  is norm-bounded,  $|\delta_i| \leq 1$ , and  $k$  measures the size of the allowable uncertainty. We want to answer the following question: What is the largest value of  $k$  for which we can guarantee that the closed loop system is stable for all possible values of  $|\delta_i| \leq 1$ ? This value, denoted  $k_{\max}$  is called the stability margin of the system.

- Show that the nominal system is stable by computing the closed loop eigenvalues with all  $\delta_i = 0$ .
- Estimate the stability margin with a simple random search. Choose  $k_{\text{try}} = 10$  and randomly set each  $\delta_i$  to have a value in  $-1 \leq \delta_i \leq 1$ . Check the eigenvalues of the closed loop system for this perturbation. If the system is unstable, then store the perturbation and reduce  $k_{\text{try}}$ , e.g.  $k_{\text{try}} = 0.99 k_{\text{try}}$ . Repeat this for 10000 random samples. Let  $\bar{k}$  denote the smallest value of  $k_{\text{try}}$  for which you found a destabilizing perturbation. Record your  $\bar{k}$  and the  $\delta_i$  that caused instability. Explain why  $k_{\max} \leq \bar{k}$ .
- Construct the uncertain closed loop system in MATLAB and plot its frequency response using the `ufrd` command. Next, find the bounds on the stability margin using the `robstab` command and compare the resulting bounds with the  $\bar{k}$  you computed in part (b). Provide an interpretation for these bounds based on the structured singular value.
- Verify that the uncertainty returned by the `robstab` command causes the closed-loop to become unstable.