


# Lecture 13: Lyapunov Theory Beyond Stability

goals:

- discuss the use of underlying Lyapunov ideas beyond stability
  - reachable sets
  - input-output properties
  - control design

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## Systems with Inputs

- Consider  $\dot{x} = f(x, u)$ ,  $x(t) \in \mathbb{R}^n$ ,  $u(t) \in \mathbb{R}^m$   


- For a function  $V: \mathbb{R}^n \rightarrow \mathbb{R}$  we have

$$\dot{V} = \nabla V(x)^T \dot{x} = \nabla V(x)^T f(x, u)$$

- $\dot{V}$  depends on both state and input
- Could we simultaneously search for a Lyapunov function and a feedback controller  $u = \pi(x)$  that stabilizes an equilibrium point?
  - related to Control Lyapunov functions (CLF)

- In general, joint search for Lyapunov function coefficients and feedback controller coefficients (gains) is non-convex; however, convex optimization techniques can be used effectively (more later)
  - alternating optimization
  - fixing a CLF and computing control online
  - in some cases, exact reformulation by change of variables

## Reachable Sets

- Suppose  $u(t) \in \mathcal{U} \quad \forall t$
- Define reachable set

$$R_\tau = \{ x(\tau) \mid \dot{x} = f(x, u), x(0) = 0, u(t) \in \mathcal{U}, \tau > 0 \}$$

i.e. the set of states that can be hit by a trajectory at time ~~at~~  $T$  with some admissible input signal

- If  $u$  is a control input,  $R_\tau$  contains states we can reach (so big  $R_\tau$  is good)
- If  $u$  is a disturbance/noise/adversarial input,  $R_\tau$  describes its worst-case effect on the state (so big  $R_\tau$  is bad)

## Lyapunov bound on reachable set

- Suppose  $\exists V: \mathbb{R}^n \rightarrow \mathbb{R}^{\text{PD}}$  and  $a > 0$  such that

$$\cancel{V(x)=b} \quad V(x)=b \text{ and } u \in \mathcal{U} \Rightarrow \dot{V}(x,u) \leq -a$$

and define  $C = \{x \mid V(x) \leq b\}$ .

Then  $R_T \subseteq C$  (i.e.  $C$  outer approximates  $R_T$ )

- Idea: On the boundary of  $C$ , every trajectory goes into  $C$ , no matter what input is applied
- $C$  is a robust invariant set: every trajectory starting in  $C$  stays in  $C$  for any admissible input
- Can use SOS and SOS S-Procedure to automatically search for polynomial  $V$  when  $f$  is polynomial and  $\mathcal{U}$  is a basic semialgebraic set:  $\mathcal{U} = \{u \in \mathbb{R}^m \mid g_i(u) \leq 0\}$   $i=1, \dots, P$

find coefficients of polynomials  $V, \tau, \lambda_i$

Subject to  $V(0) = 0$

$V(x) - \varepsilon \phi(x) \in \text{SOS}$  some  $\varepsilon > 0$ ,  $\phi$  PD

$$\nabla V(x)^T f(x,u) + a \leq \sum_{i=1}^P \lambda_i(u) g_i(u) + \tau(x)(V(x)-b) \quad \forall x, \forall u$$

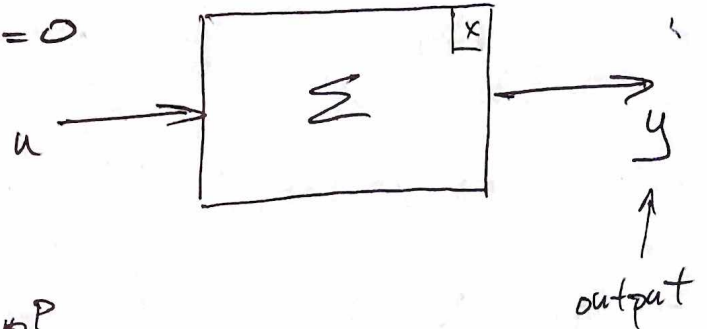
$\lambda_i(u) \in \text{SOS}$

# Lyapunov Method for bounding system $H_\infty$ norm

- Consider the system

$$\dot{x} = f(x, u), \quad x(0) = 0$$

$$y = g(x, u)$$



$$x(t) \in \mathbb{R}^n, \quad u(t) \in \mathbb{R}^m, \quad y(t) \in \mathbb{R}^p$$

- Define system  $H_\infty$  norm (aka induced  $L_2$  gain)

$$\|\Sigma\|_\infty = \sup_{\|u\| \neq 0} \frac{\|y\|_2}{\|u\|_2}$$

$$\text{where } \|\cdot\|_2^2 = \int_0^\infty \|u(t)\|_2^2 dt$$

• signal norm

• system norm

- Suppose  $\gamma \geq 0$  and  $\exists V: \mathbb{R}^n \rightarrow \mathbb{R}$  such that

• ~~the following~~  $V \geq 0, \quad V(0) = 0$

$$\dot{V}(x, u) \leq \gamma^2 u^T u - y^T y = \gamma^2 u^T u - g(x, u)^T g(x, u) \quad \forall x, u$$

Then  $\|\Sigma\|_\infty \leq \gamma$

- $V$  certifies bound on  $H_\infty$  system norm



Proof: Suppose we've found such a  $V$ . For any  $T > 0$

$$\begin{aligned} V(x(T)) - V(x(0)) &= \int_0^T \dot{V}(x(t), u(t)) dt \\ &\leq \int_0^T (\gamma^2 u(t)^T u(t) - y(t)^T y(t)) dt \quad \forall u, y \end{aligned}$$

Since  $V(x(0)) = V(0) = 0$  and  $V(x(T)) \geq 0$

$$\int_0^T y(t)^T y(t) dt \leq \gamma^2 \int_0^T u(t)^T u(t) dt \quad \forall u, y$$

Taking the limit as  $T \rightarrow \infty$  gives

$$\|y\|_2^2 \leq \gamma^2 \|u\|_2^2 \Rightarrow \|y\|_2 \leq \|u\|_2 \equiv \gamma \quad \forall u, y$$

- Once again, can automatically search for polynomial  $V$  when  $f$  and  $g$  are polynomial using SOS programming

find coefficients of polynomial  $V$   
subject to  $V(x) \in \text{SOS}$

$$\gamma^2 u^T u - g(x, u)^T g(x, u) - \nabla V(x)^T f(x, u) \in \text{SOS}$$

## Bounded Real Lemma

- Let's consider a quadratic Lyapunov function

$$V(x) = x^T P x \quad \text{and linear system}$$

$$\dot{x} = Ax + Bu \quad x(0) = 0$$

$$y = Cx$$

- The  $\dot{V}$  condition becomes

$$\dot{V}(x, u) - \gamma^2 u^T u + y^T y \leq 0$$

$$(Ax + Bu)^T P x + x^T P (Ax + Bu) - \gamma^2 u^T u + x^T C^T C x \leq 0$$

$$\begin{bmatrix} x \\ u \end{bmatrix}^T \begin{bmatrix} ATP + PA + C^T C & PB \\ B^T P & -\gamma^2 I \end{bmatrix} \begin{bmatrix} x \\ u \end{bmatrix} \leq 0 \quad \forall x, u$$

$$\Leftrightarrow \begin{bmatrix} ATP + PA + C^T C & PB \\ B^T P & -\gamma^2 I \end{bmatrix} \preceq 0 \quad \text{LMI in } P!$$

( $\exists P \succ 0$ )

- Turns out for linear systems this condition is both necessary + sufficient for  $\|z\|_\infty \leq \gamma$

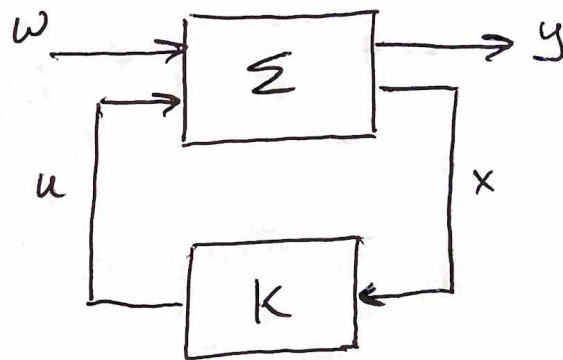
• called Bounded Real Lemma

# $H_\infty$ Control Design via SDP

- Consider the system

$$\dot{x} = Ax + Bu + Fw$$

$$y = Cx + Du$$



- Goal: Design a full state feedback controller  $u(t) = Kx(t)$  that minimizes the  $H_\infty$  system norm from disturbance input  $w$  to performance output  $y$  (in closed-loop)
  - i.e. minimize the worst-case effect of  $w$  on  $y$
  - closely related to robust control of uncertain systems
  - also closely related to the dynamic game problem

$$\min_u \max_w \int_0^\infty \left( y(t)^T y(t) - \gamma^2 w(t)^T w(t) \right) dt$$

subject to

$$\begin{aligned} \dot{x}(t) &= Ax(t) + Bu(t) + Fw(t) \\ y(t) &= Cx(t) + Du(t) \end{aligned}$$

- Let's use the Bounded Real LMI for the closed-loop with  $u = Kx$

$$\Rightarrow \begin{aligned} \dot{x} &= (A+BK)x + Fw \\ y &= (C+DK)x \end{aligned}$$

$$\begin{bmatrix} (A+BK)^T P + P(A+BK) + (C+DK)^T (C+DK) & PF \\ F^T P & -\gamma^2 I \end{bmatrix} \preceq 0$$

- not an LMI in  $P$  and  $K$ , but can reformulate via Schur complement + variable substitutions
- Assume  $P \succ 0$  and let  $Q = P^{-1}$ , then multiply LHS on left and right by  $\begin{bmatrix} Q & 0 \\ 0 & I \end{bmatrix}$

$$\begin{bmatrix} Q(A+BK)^T + (A+BK)Q + Q(C+DK)^T (C+DK)Q & F \\ F^T & -\gamma^2 I \end{bmatrix} \preceq 0$$

- Now take the Schur complement of upper left block and define  $L = KQ$

$$\begin{bmatrix} QA^T + L^T B^T + A Q + B L & F & Q C^T + L^T D^T \\ F & -\gamma^2 I & 0 \\ C Q + D L & 0 & -I \end{bmatrix} \preceq 0$$

- LMI in variables  $Q$  and  $L$ !
- problem data:  $A, B, F, C, D, \gamma$



- If feasible, can recover controller gain by

$$K = LQ^{-1}$$

which gives closed-loop  $H_{\infty}$  norm of  $\gamma$

- Can find optimal  $H_{\infty}$  controller (one that gives smallest  $\gamma$ ) by defining variable  $\eta = \gamma^2$  and solving the SDP

$$\begin{array}{ll} \text{minimize} & \eta \\ & \eta, Q, L \end{array}$$

$$\text{subject to} \quad \begin{bmatrix} QA^T + L^TB^T + A^TQ + BL & F & QE^T + L^TD^T \\ F & -\eta I & 0 \\ CQ + DL & 0 & -I \end{bmatrix} \leq 0$$

$$Q \succ 0$$