

## COM303: Digital Signal Processing

Lecture 6: DFS and DTFT

# Overview

- ▶ periodicity in the DFT
- ▶ the DFS
- ▶ the DTFT

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# DFT formulas

Analysis formula:

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi}{N}nk}, \quad k = 0, 1, \dots, N-1$$

$N$ -point signal in the *frequency domain*

Synthesis formula:

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j\frac{2\pi}{N}nk}, \quad n = 0, 1, \dots, N-1$$

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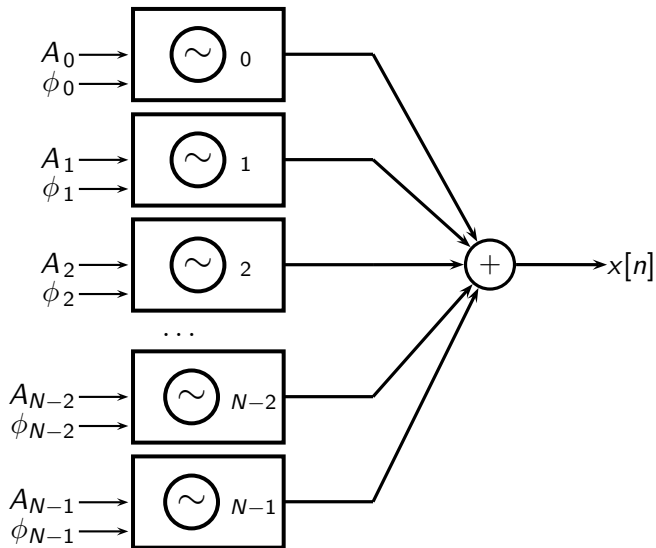
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$N$ -point signal in the *“time” domain*

## DFT synthesis formula



## Running the machine too long...

$$x[n + N] = x[n]$$

output signal is  $N$ -periodic!



## Inherent periodicities in the DFT

the synthesis formula:

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# Discrete Fourier Series (DFS)

DFS = DFT with periodicity explicit

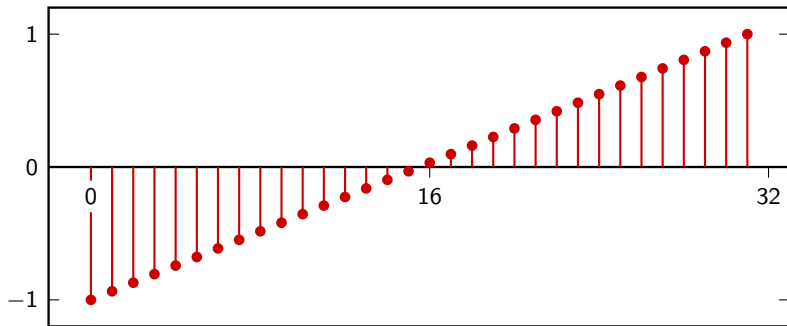
- ▶ the DFS maps an  $N$ -periodic signal onto an  $N$ -periodic sequence of Fourier coefficients
- ▶ the inverse DFS maps an  $N$ -periodic sequence of Fourier coefficients a set onto an  $N$ -periodic signal
- ▶ the DFS of an  $N$ -periodic signal is mathematically equivalent to the DFT of one period

the Fourier transform for periodic signals

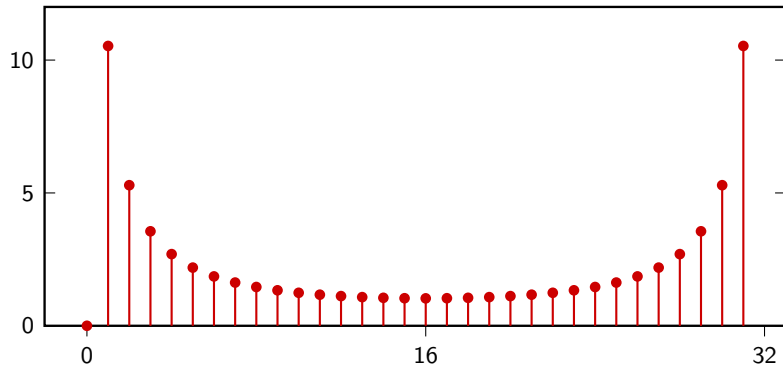
## Periodic sequences: a bridge to infinite-length signals

- ▶  $N$ -periodic sequence:  $N$  degrees of freedom
- ▶ DFS: only  $N$  Fourier coefficients capture all the information

## Example: 32-tap sawtooth wave

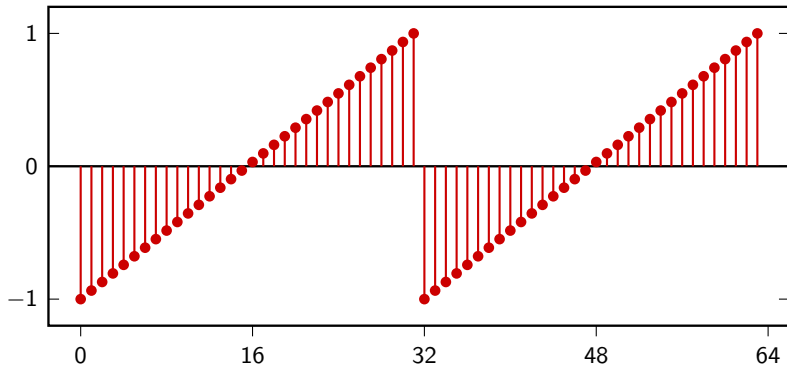


## Example: DFT of 32-tap sawtooth wave

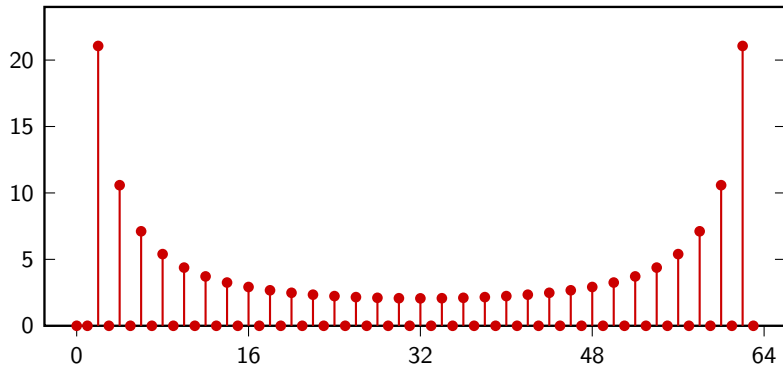




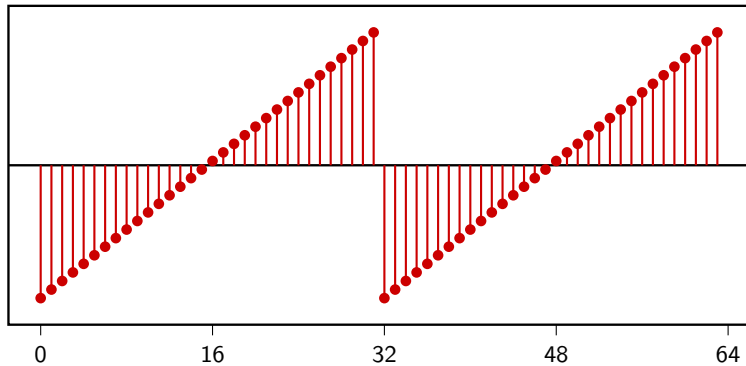
What if we take the DFT of two periods?



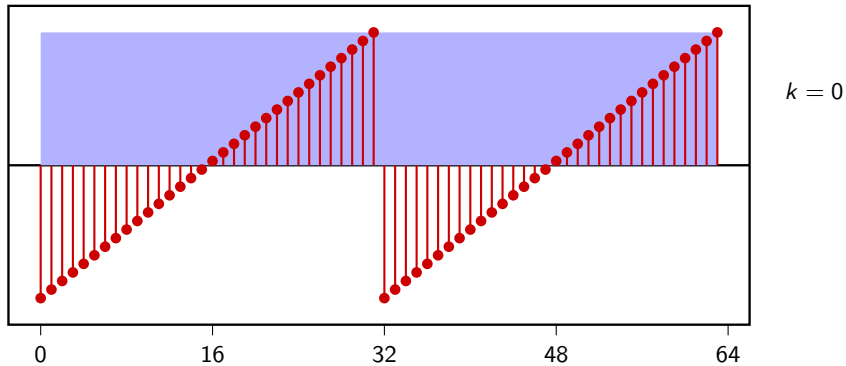
## Example: 64-point DFT of two periods



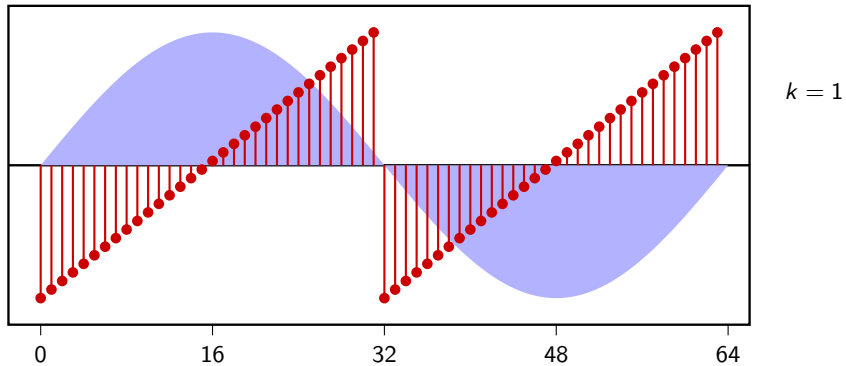
## DFT of two periods: intuition



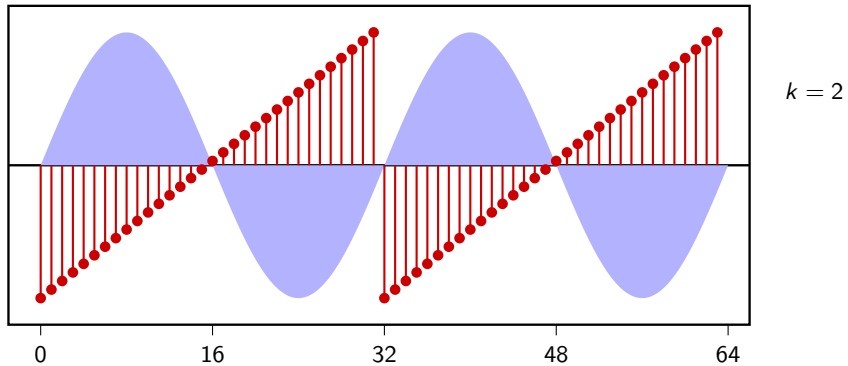
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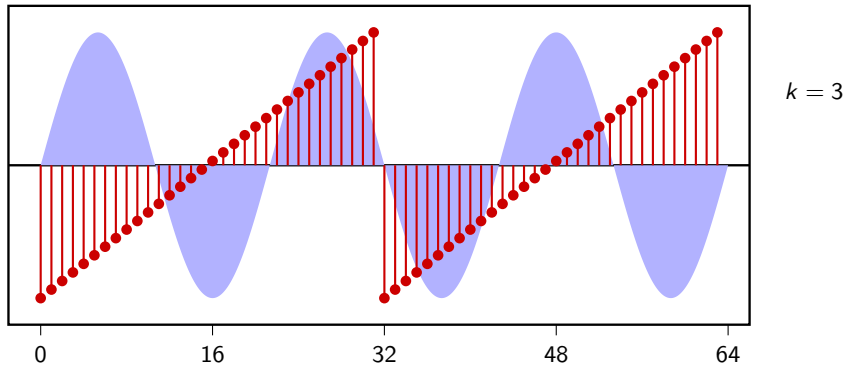
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## DFT of two periods: intuition



## DFT of $L$ periods

ingredients:

- ▶ finite-length signal  $x[n]$ ,  $n = 0, 1, \dots, N - 1$
- ▶  $N$ -periodic signal:  $\tilde{x}[n] = x[n \bmod N]$
- ▶ obviously  $\tilde{x}[n] = \tilde{x}[n + pN]$  for all  $p \in \mathbb{Z}$



## DFT of $L$ periods

$$\begin{aligned}X_L[k] &= \sum_{n=0}^{LN-1} \tilde{x}[n] e^{-j \frac{2\pi}{LN} nk} \quad k = 0, 1, 2, \dots, LN - 1 \\&= \sum_{p=0}^{L-1} \sum_{n=0}^{N-1} \tilde{x}[n + pN] e^{-j \frac{2\pi}{LN} (n+pN)k} \\&= \sum_{p=0}^{L-1} \sum_{n=0}^{N-1} \tilde{x}[n] e^{-j \frac{2\pi}{LN} nk} e^{-j \frac{2\pi}{L} pk} \\&= \left( \sum_{p=0}^{L-1} e^{-j \frac{2\pi}{L} pk} \right) \sum_{n=0}^{N-1} \tilde{x}[n] e^{-j \frac{2\pi}{LN} nk}\end{aligned}$$

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We've seen this before

$$\sum_{p=0}^{L-1} e^{-j\frac{2\pi}{L}pk} = \begin{cases} L & \text{if } k \text{ multiple of } L \\ 0 & \text{otherwise} \end{cases}$$

(remember the orthogonality proof for the DFT basis)

## DFT of $L$ periods

if  $k$  is a multiple of  $L$  then  $k/L$  is an integer, so:

$$\sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi}{N}n\frac{k}{L}} = X[k/L]$$

## DFT of $L$ periods

$$X_L[k] = \begin{cases} L X[k/L] & \text{if } k = 0, L, 2L, 3L, \dots \\ 0 & \text{otherwise} \end{cases}$$

- ▶ again, all the spectral information for a periodic signal is contained in the DFT coefficients of a single period
- ▶ to stress the periodicity of the underlying signal, we use the term DFS



# DFT and DFS

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- ▶ to stress the periodicity of the underlying signal, we use the term DFS

## Finite-length time shifts revisited

The DFS helps us understand how to define time shifts for finite-length signals.

For an  $N$ -periodic sequence  $\tilde{x}[n]$ :

- ▶  $\tilde{x}[n - M]$  is well-defined for all  $M \in \mathbb{N}$
- ▶  $\text{DFS} \{ \tilde{x}[n - M] \} = e^{-j\frac{2\pi}{N}Mk} \tilde{X}[k]$  (easy derivation)
- ▶  $\text{IDFS} \{ \tilde{X}[k] \} = \tilde{x}[n - M]$

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 a delay in time becomes a *linear phase* factor in frequency

# Finite-length time shifts revisited

For an  $N$ -point signal  $x[n]$  :

►  $x[n - M]$  is *not* well-defined

► what is IDFT  $\left\{ e^{-j\frac{2\pi}{N}Mk} X[k] \right\}$  ?

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## Finite-length time shifts revisited

$$\begin{aligned}\text{IDFT} \left\{ e^{-j\frac{2\pi}{N}Mk} X[k] \right\} [n] &= \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{-j\frac{2\pi}{N}Mk} e^{j\frac{2\pi}{N}nk} \\ &= \frac{1}{N} \sum_{k=0}^{N-1} \left( \sum_{m=0}^{N-1} x[m] e^{-j\frac{2\pi}{N}mk} \right) e^{-j\frac{2\pi}{N}Mk} e^{j\frac{2\pi}{N}nk} \\ &= \frac{1}{N} \sum_{m=0}^{N-1} x[m] \sum_{k=0}^{N-1} e^{j\frac{2\pi}{N}(n-M-m)k}\end{aligned}$$

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$$\sum_{k=0}^{N-1} e^{-j\frac{2\pi}{N}nk} = \begin{cases} N & \text{if } n \text{ multiple of } N \\ 0 & \text{otherwise} \end{cases}$$

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$$\sum_{k=0}^{N-1} e^{j\frac{2\pi}{N}((n-M)-m)k} = \begin{cases} N & \text{for } ((n-M) - m) \text{ multiple of } N \\ 0 & \text{otherwise} \end{cases}$$

- ▶  $n, M$  are fixed
- ▶  $m$  goes from 0 to  $N - 1$
- ▶ is there always a value for  $m$  that makes  $((n - M) - m)$  a multiple of  $N$ ?

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## Modulo operator

given  $C \in \mathbb{N}$ , find  $m$  such that  $0 \leq m < N$  and  $C - m$  is a multiple of  $N$

any integer  $C$  can be written as  $C = pN + (C \bmod N)$ ,  $p \in \mathbb{N}$ :

- ▶  $0 \leq (C \bmod N) < N$
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shifts for finite-length signals are “naturally” circular

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# The situation so far

Fourier representation for signal classes:

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the Fourier transform for infinite-length signals

# DFT of increasingly long signals

► Start with the DFT. What happens when  $N \rightarrow \infty$  ?

►  $\frac{2\pi}{N}k$  becomes denser in  $[0, 2\pi]$ ...

► In the limit  $\frac{2\pi}{N}k \rightarrow \omega$ :

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# Discrete-Time Fourier Transform (DTFT)

Formal definition:

►  $x[n] \in \ell_2(\mathbb{Z})$

► define the *function* of  $\omega \in \mathbb{R}$

$$F(\omega) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

► inversion (when  $F(\omega)$  exists):

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} F(\omega) e^{j\omega n} d\omega, \quad n \in \mathbb{Z}$$

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# DTFT periodicity and notation

- ▶  $e^{j\omega n} = e^{j(\omega+2k\pi)n} \quad \forall k \in \mathbb{N}$

- ▶  $F(\omega)$  is  $2\pi$ -periodic

- ▶ to stress periodicity (and for other reasons) we will write

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

- ▶ by convention,  $X(e^{j\omega})$  is represented over  $[-\pi, \pi]$

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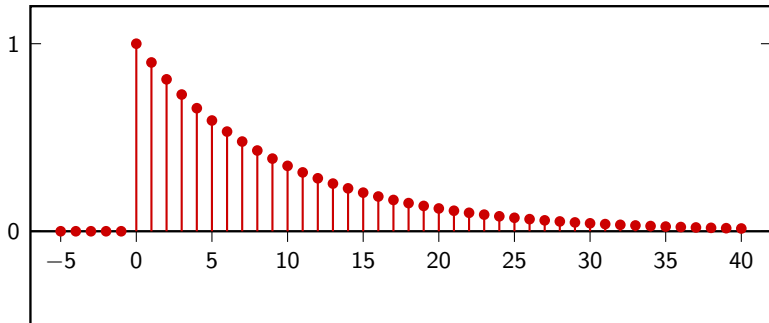
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- ▶ to stress periodicity (and for other reasons) we will write

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

- ▶ by convention,  $X(e^{j\omega})$  is represented over  $[-\pi, \pi]$

$$x[n] = \alpha^n u[n], \quad |\alpha| < 1$$



DTFT of  $x[n] = \alpha^n u[n]$ ,  $|\alpha| < 1$

$$\begin{aligned} X(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} \\ &= \sum_{n=0}^{\infty} \alpha^n e^{-j\omega n} \\ &= \sum_{n=0}^{\infty} (\alpha e^{-j\omega})^n \\ &= \frac{1}{1 - \alpha e^{-j\omega}} \end{aligned}$$

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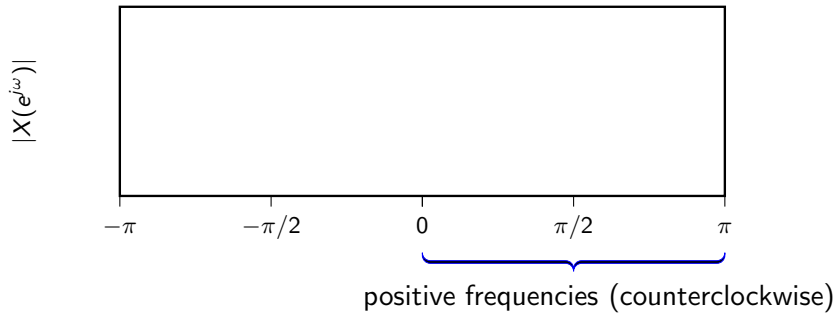
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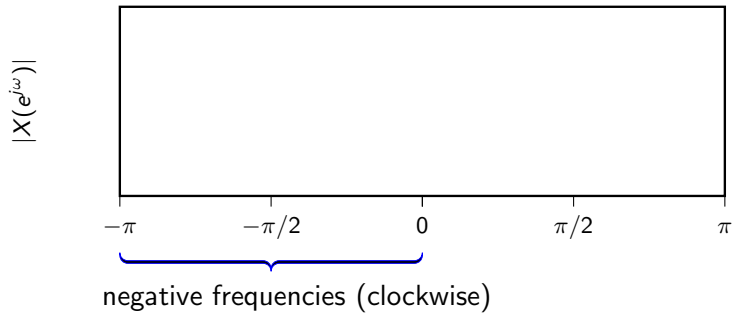
$$|X(e^{j\omega})|^2 = \frac{1}{1 + \alpha^2 - 2\alpha \cos \omega}$$

## Plotting the DTFT

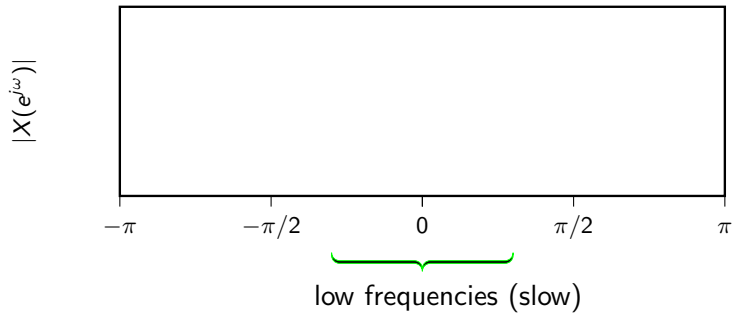




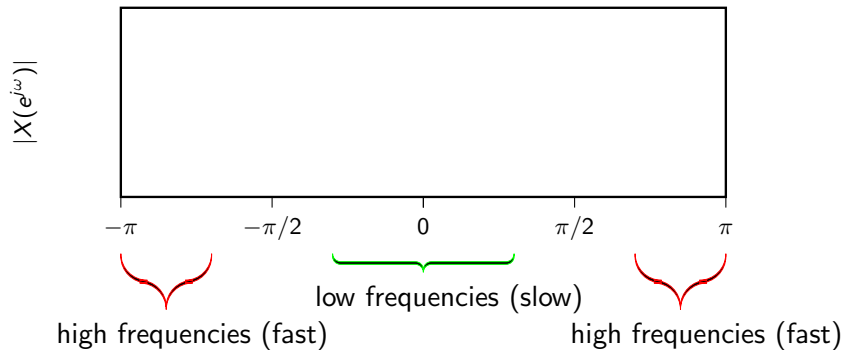
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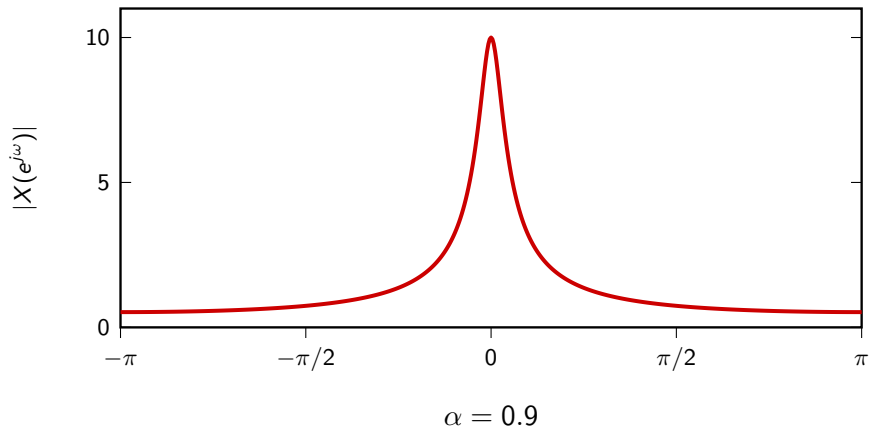
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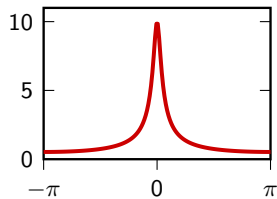
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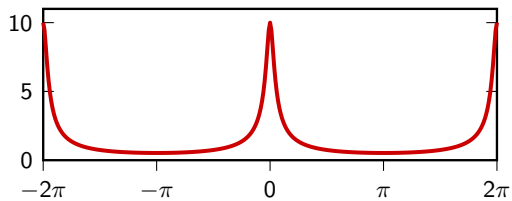
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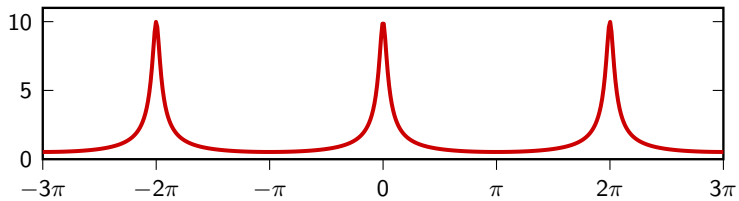
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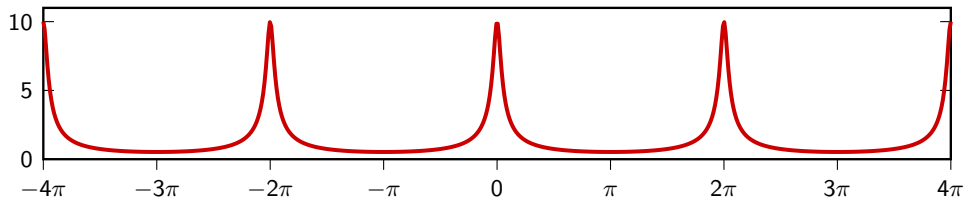
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## DTFT intuition and properties

## Overview:

- ▶ DTFT Existence
- ▶ Properties
- ▶ DTFT as basis expansion

# Discrete-Time Fourier Transform

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

- ▶ when does it exist?
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## Existence easy for absolutely summable sequences

$$\begin{aligned} |X(e^{j\omega})| &= \left| \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} \right| \\ &\leq \sum_{n=-\infty}^{\infty} |x[n] e^{-j\omega n}| \\ &= \sum_{n=-\infty}^{\infty} |x[n]| \\ &< \infty \end{aligned}$$

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## Inversion easy for absolutely summable sequences

$$\begin{aligned}\frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega &= \frac{1}{2\pi} \int_{-\pi}^{\pi} \left( \sum_{k=-\infty}^{\infty} x[k] e^{-j\omega k} \right) e^{j\omega n} d\omega \\ &= \sum_{k=-\infty}^{\infty} x[k] \int_{-\pi}^{\pi} \frac{e^{j\omega(n-k)}}{2\pi} d\omega \\ &= x[n]\end{aligned}$$

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$$\int_{-\pi}^{\pi} \frac{e^{j\omega m}}{2\pi} d\omega = \frac{1}{2\pi} \int_{-\pi}^{\pi} d\omega = 1 \quad \text{for } m = 0$$

$$= \frac{1}{2\pi} \frac{1}{jm} e^{j\omega m} \Big|_{-\pi}^{\pi}$$

$$= \frac{1}{2\pi} \frac{1}{jm} (e^{j\pi m} - e^{-j\pi m}) = 0 \quad \text{otherwise}$$

the DTFT as the limit of the DFS

# Synopsis

- ▶  $x[n]$  absolutely summable  $\Rightarrow X(e^{j\omega})$  exists formally
- ▶  $x[n]$  absolutely summable  $\Rightarrow$  we can *periodize* it into  $\tilde{x}_N[n]$
- ▶ natural Fourier representation for  $\tilde{x}_N[n]$  is DFS
- ▶ DFS of  $\tilde{x}_N[n]$  turns out to be  $X(e^{j\omega})$  at  $\omega = (2\pi/N)k$
- ▶ as  $N$  grows to infinity  $\tilde{x}_N[n]$  becomes  $x[n]$
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## Some intuition

With  $x[n]$  absolutely summable we can build arbitrarily “periodized” sequences:

$$\tilde{x}_N[n] = \sum_{p=-\infty}^{\infty} x[n + pN]$$

clearly  $\tilde{x}_N[n] = \tilde{x}_N[n + N]$

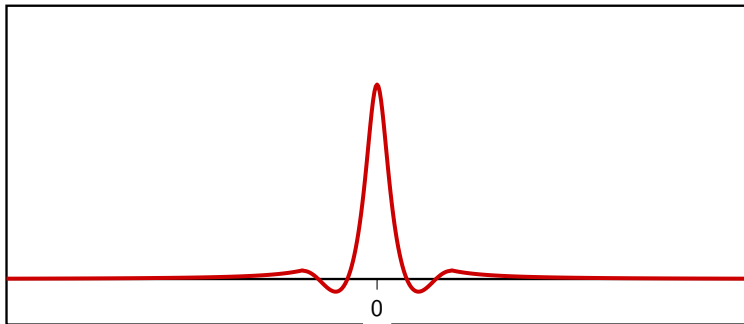
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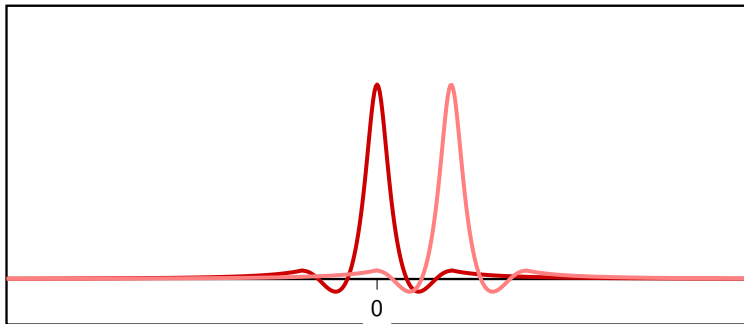
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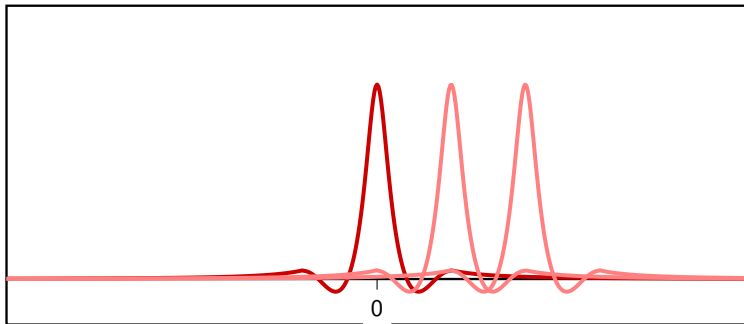
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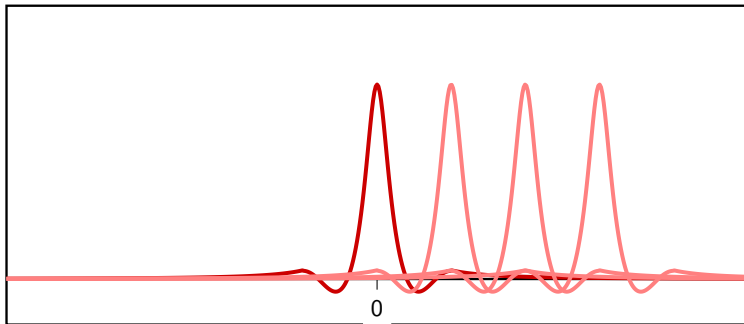


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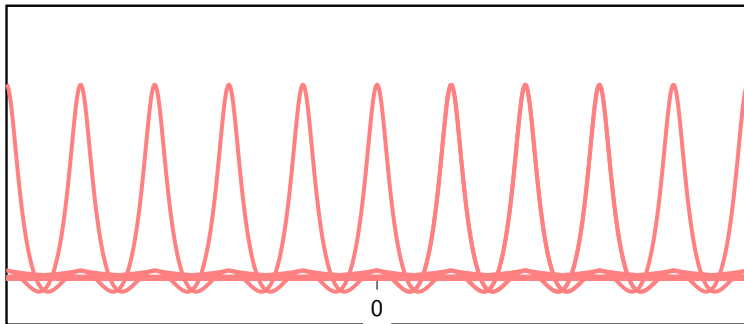




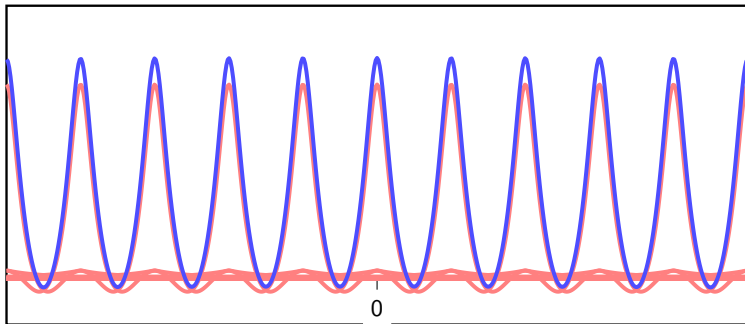
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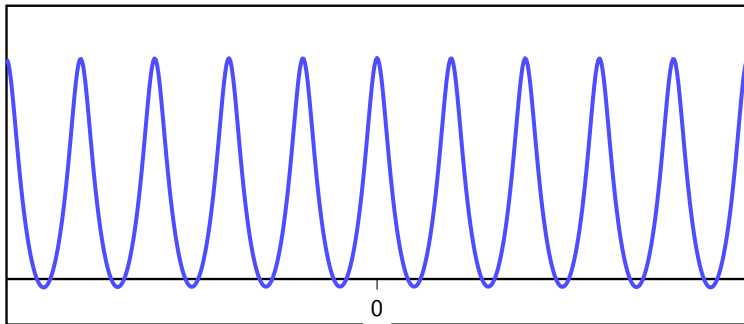
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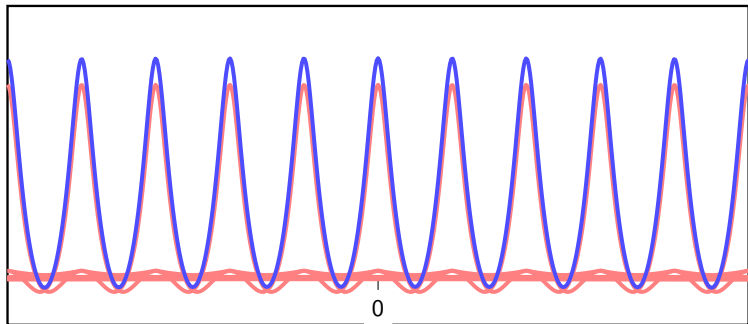


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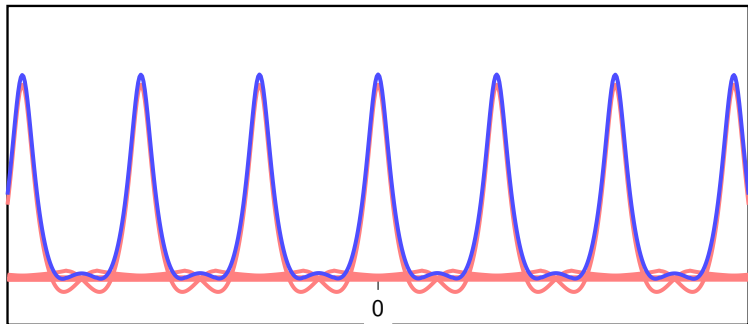
Let  $N$  grow large...



$N = 10$

# Periodization

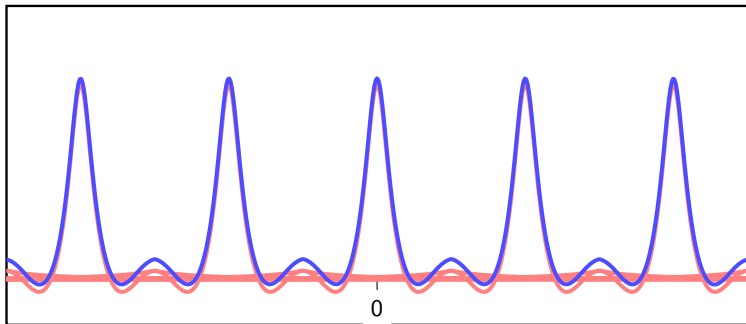
Let  $N$  grow large...



$N = 16$

# Periodization

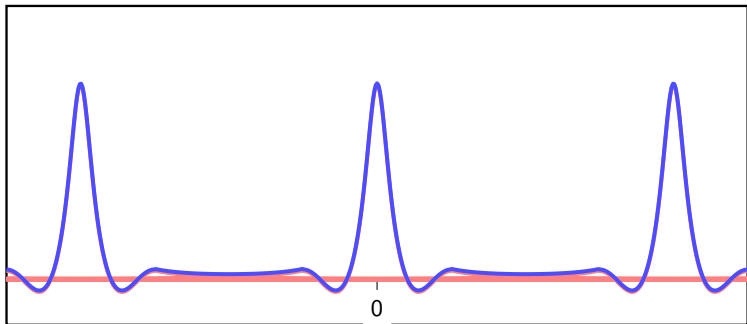
Let  $N$  grow large...



$N = 20$

# Periodization

Let  $N$  grow large...

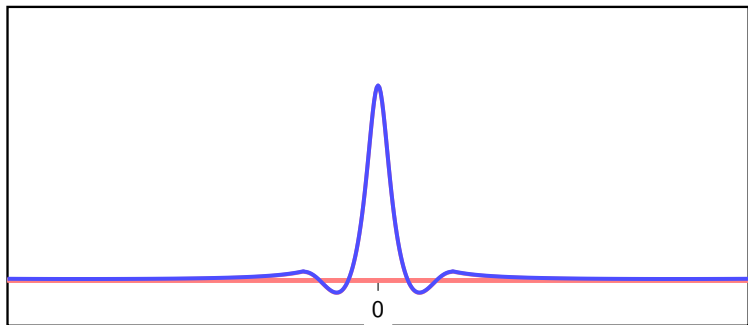


$N = 40$



# Periodization

... as  $N$  grows,  $\tilde{x}_N[n] \rightarrow x[n]$



## From DFS to DTFT

Natural spectral representation for  $\tilde{x}_N[n]$  is the DFS:

$$\begin{aligned}\tilde{X}[k] &= \sum_{n=0}^{N-1} \tilde{x}_N[n] e^{-j\frac{2\pi}{N}nk} \\ &= \sum_{n=0}^{N-1} \sum_{p=-\infty}^{\infty} x[n + pN] e^{-j\frac{2\pi}{N}nk} \\ &= \sum_{p=-\infty}^{\infty} \sum_{n=0}^{N-1} x[n + pN] e^{-j\frac{2\pi}{N}(n+pN)k}\end{aligned}$$

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## From double sum to single sum

we can always write for all  $N \in \mathbb{N}^+$

$$\sum_{m=-\infty}^{\infty} y[m] = \sum_{p=-\infty}^{\infty} \sum_{n=0}^{N-1} y[n + pN]$$

## Example (N=4)

		$n$			
		0	1	2	3
$p$					
	...				
	-1				
	0				
	1				
	2				
	...				

$$m = n + 4p$$

## Example (N=4)

		$n$			
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## Example (N=4)

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$p$					
	...				
	-1				
	0	0	1	2	3
	1	4			
	2				
	...				

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## Example (N=4)

		$n$			
		0	1	2	3
$p$					
	...				
	-1				
	0	0	1	2	3
	1	4	5		
	2				
	...				

$$m = n + 4p$$

## Example (N=4)

		$n$			
		0	1	2	3
$p$					
	...				
	-1				
	0	0	1	2	3
	1	4	5	6	
	2				
	...				

$$m = n + 4p$$

## Example (N=4)

		<i>n</i>			
		0	1	2	3
<i>p</i>					
	...				
	-1				
	0	0	1	2	3
	1	4	5	6	7
	2				
	...				

$$m = n + 4p$$

## Example (N=4)

		$m$			
		0	1	2	3
$p$					
	...				
	-1	-4	-3	-2	-1
	0	0	1	2	3
	1	4	5	6	7
	2	8	9	10	11
	...				

$$n = 4p + m$$

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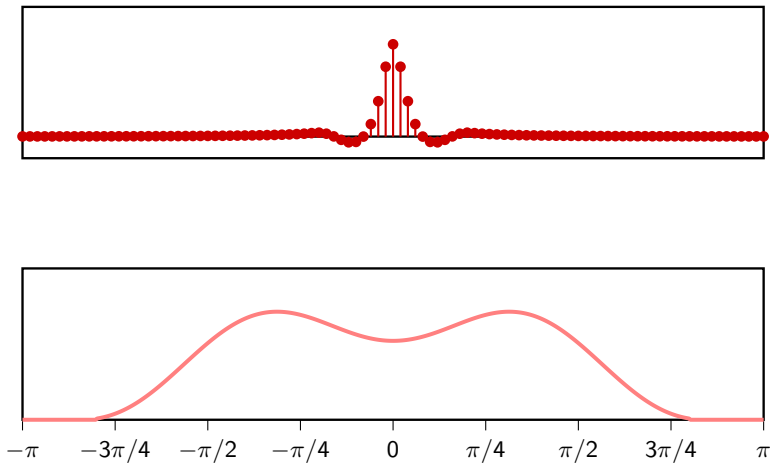
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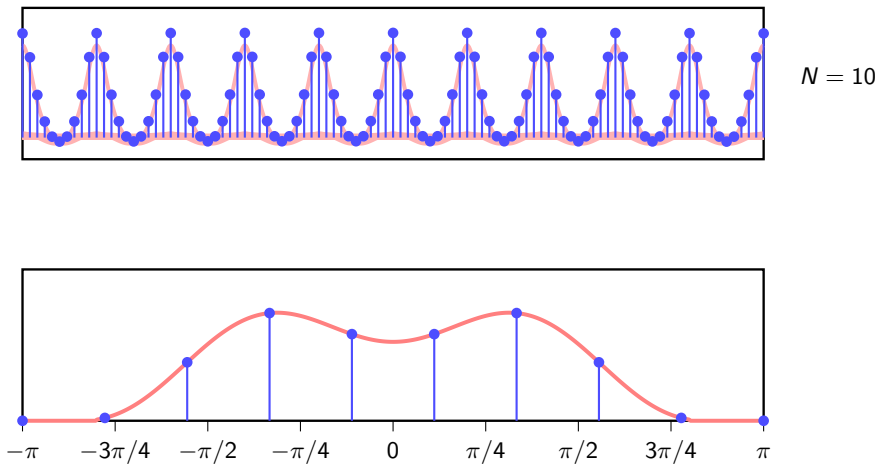
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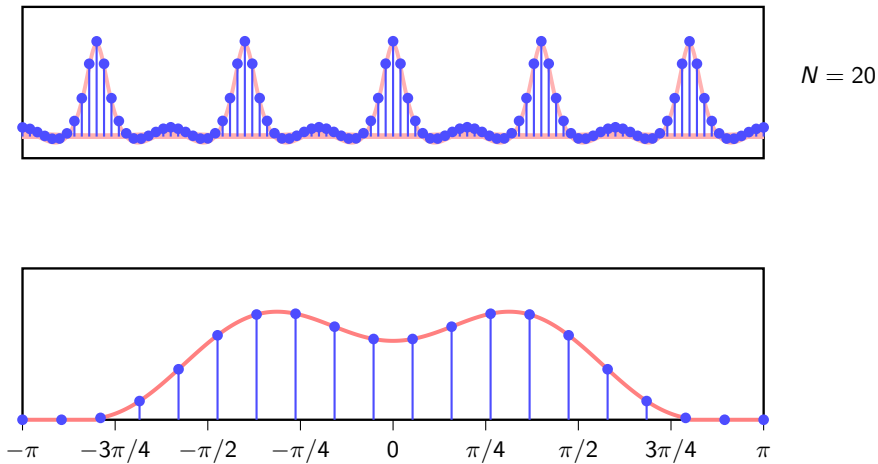
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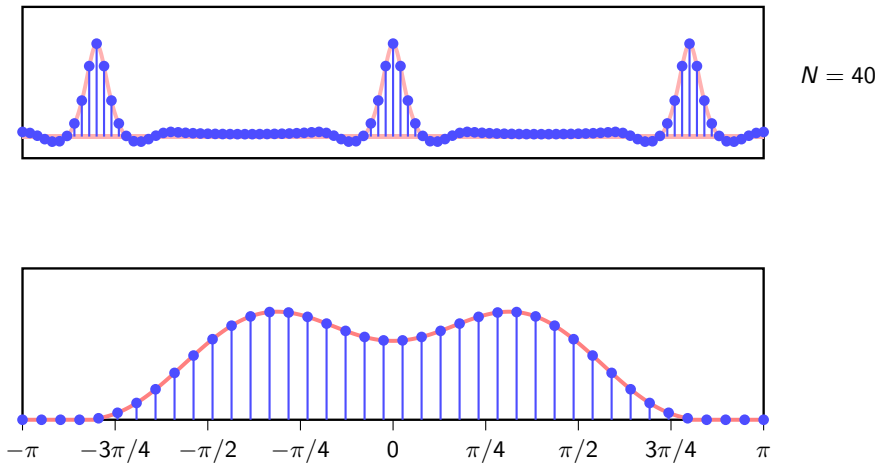
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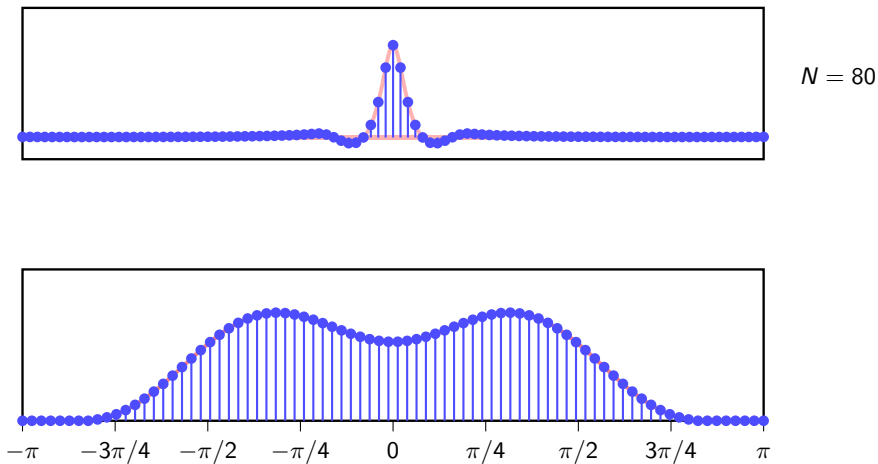
## From DFS to DTFT



# From DFS to DTFT



# From DFS to DTFT



# From DFS to DTFT

- ▶ we're comfortable with DFS: change of basis, energy conservation, etc.
- ▶ as  $N$  grows,  $\tilde{x}_N[n] \rightarrow x[n]$  and the spectral representation “becomes” the DTFT
- ▶ we can retain the “change of basis” paradigm for the DTFT



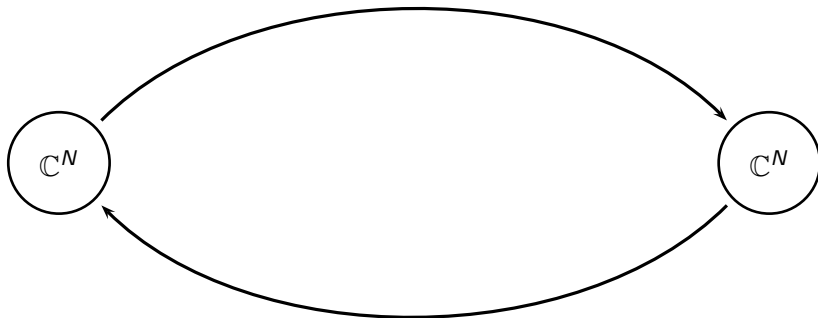
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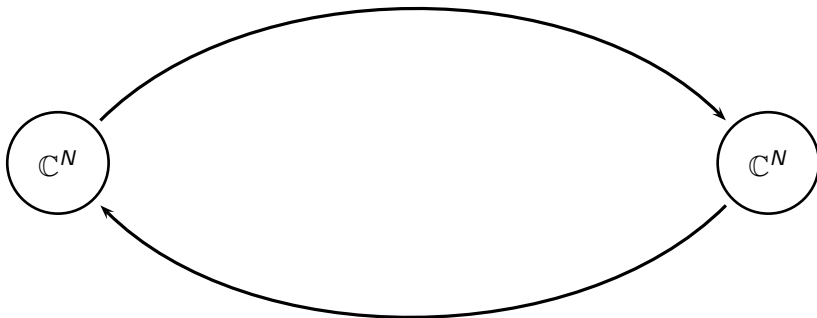
- ▶ we're comfortable with DFS: change of basis, energy conservation, etc.
- ▶ as  $N$  grows,  $\tilde{x}_N[n] \rightarrow x[n]$  and the spectral representation “becomes” the DTFT
- ▶ we can retain the “change of basis” paradigm for the DTFT

## Review: DFT

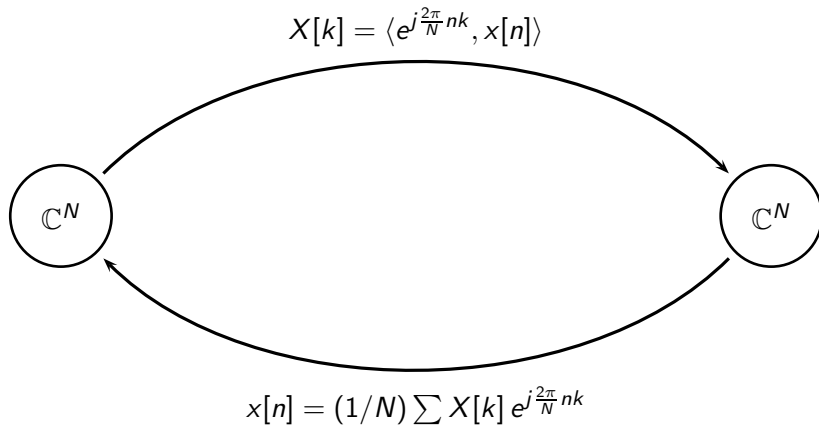


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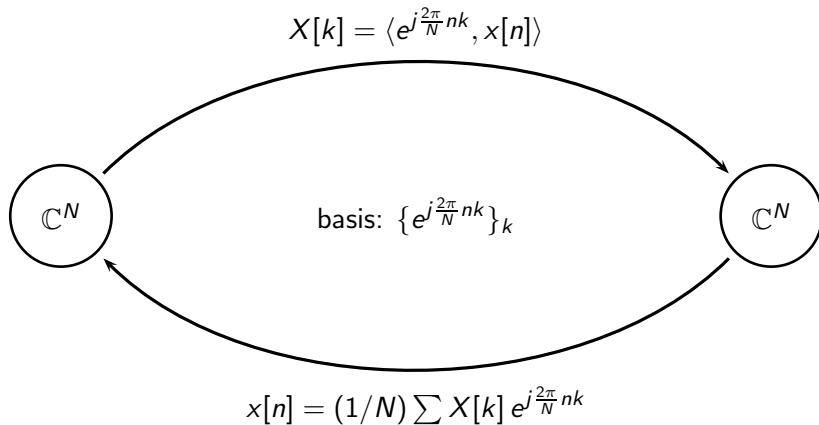
$$X[k] = \langle e^{j\frac{2\pi}{N}nk}, x[n] \rangle$$



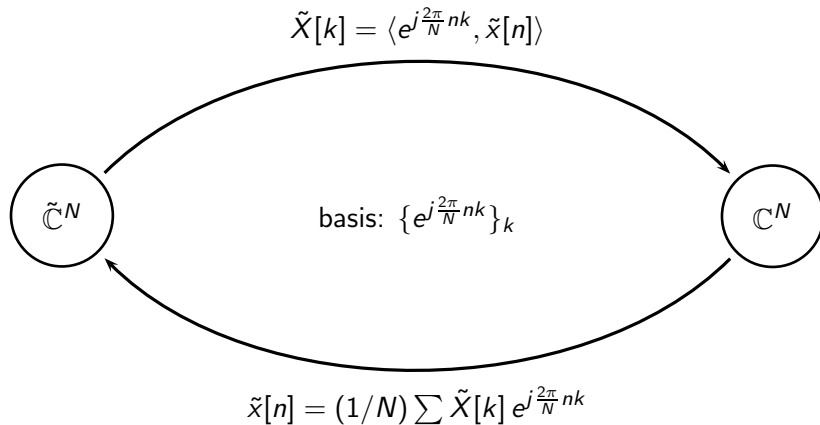
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## Review: DFS



## What about the DTFT?

- ▶ formally DTFT is an inner product in  $\mathbb{C}^\infty$ :

$$\sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} = \langle e^{j\omega n}, x[n] \rangle$$

- ▶ “basis” is an infinite, uncountable basis:  $\{e^{j\omega n}\}_{\omega \in \mathbb{R}}$
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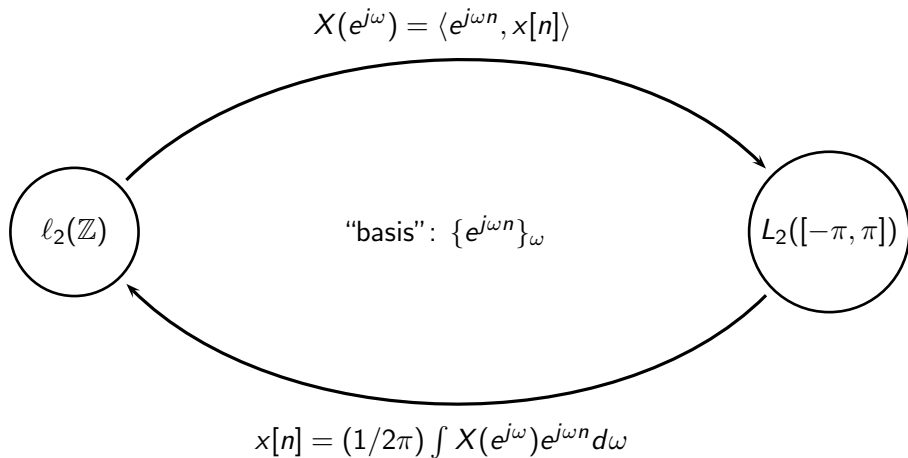
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# DTFT properties

- ▶ linearity

$$\text{DTFT}\{\alpha x[n] + \beta y[n]\} = \alpha X(e^{j\omega}) + \beta Y(e^{j\omega})$$

- ▶ time shift

$$\text{DTFT}\{x[n - M]\} = e^{-j\omega M} X(e^{j\omega})$$

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## Some particular cases:

- ▶ if  $x[n]$  is symmetric, the DTFT is symmetric:

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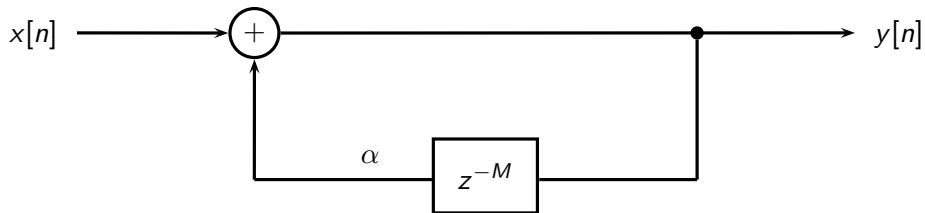
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The DTFT of the Karplus-Strong output

## Karplus-Strong revisited



$$y[n] = \alpha y[n - M] + x[n]$$

## Karplus-Strong revisited

- ▶ choose a signal  $\bar{x}[n]$  that is nonzero only for  $0 \leq n < M$

- ▶ generated signal is infinite-length but not periodic:

$$y[n] = \bar{x}[0], \bar{x}[1], \dots, \bar{x}[M-1], \alpha\bar{x}[0], \alpha\bar{x}[1], \dots, \alpha\bar{x}[M-1], \alpha^2\bar{x}[0], \alpha^2\bar{x}[1], \dots$$

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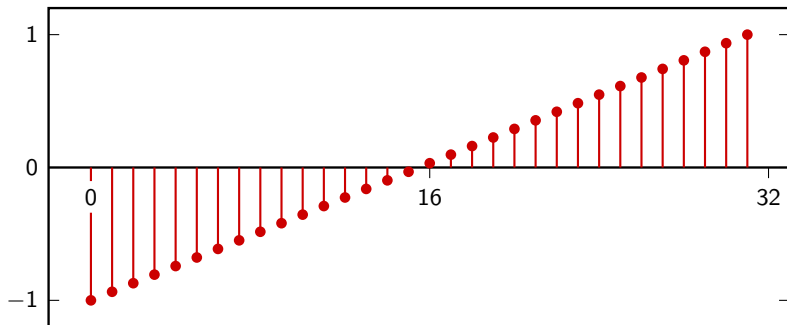
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## KS revisited: 32-tap sawtooth wave

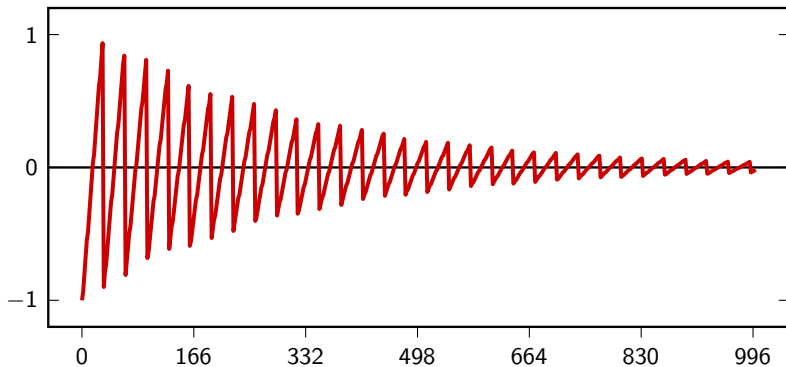
$$x[n] = 2n/(M - 1) - 1, \quad n = 0, 1, \dots, M - 1$$





## KS revisited: decay $\alpha = 0.9$

$$y[n] = \alpha^{\lfloor n/M \rfloor} \bar{x}[n \bmod M] u[n]$$



## DTFT of KS signal

$$Y(e^{j\omega}) = \sum_{n=-\infty}^{\infty} y[n] e^{-j\omega n}$$

Same trick we used before:

$$\sum_{m=-\infty}^{\infty} y[m] = \sum_{p=-\infty}^{\infty} \sum_{n=0}^{N-1} y[n + pN]$$

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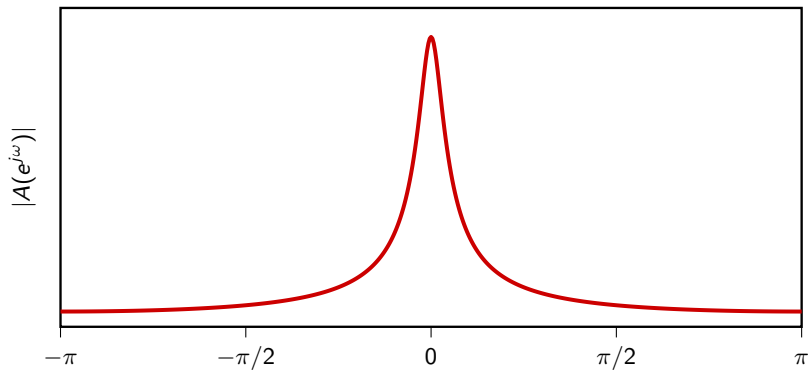
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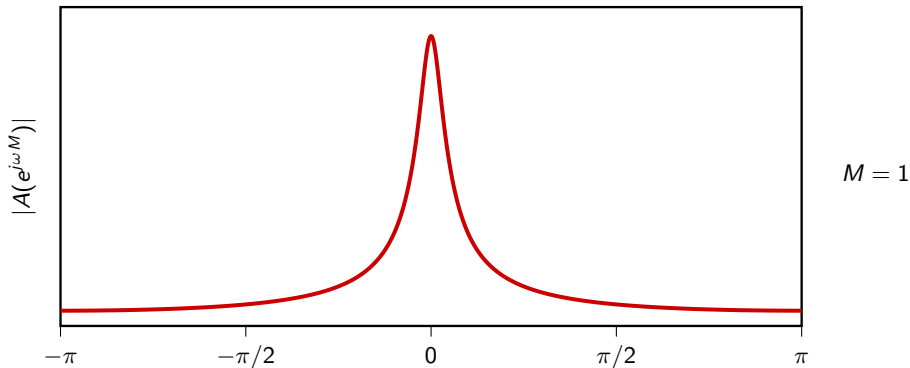
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$$A(e^{j\omega}) = \text{DTFT} \{ \alpha^n u[n] \} = \frac{1}{1 - \alpha e^{-j\omega}}$$



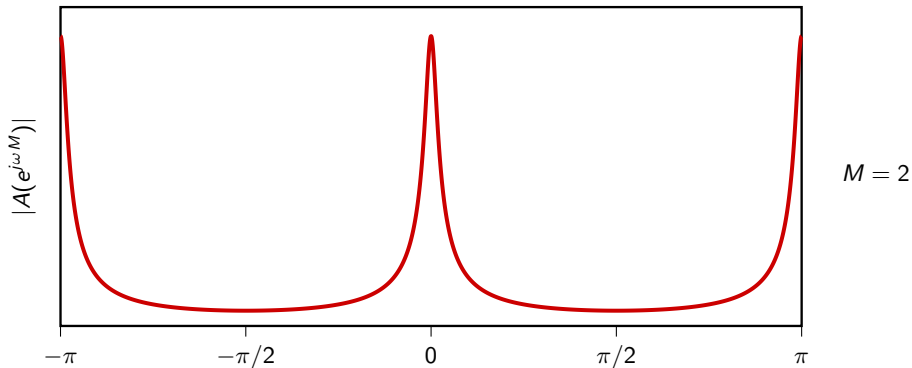
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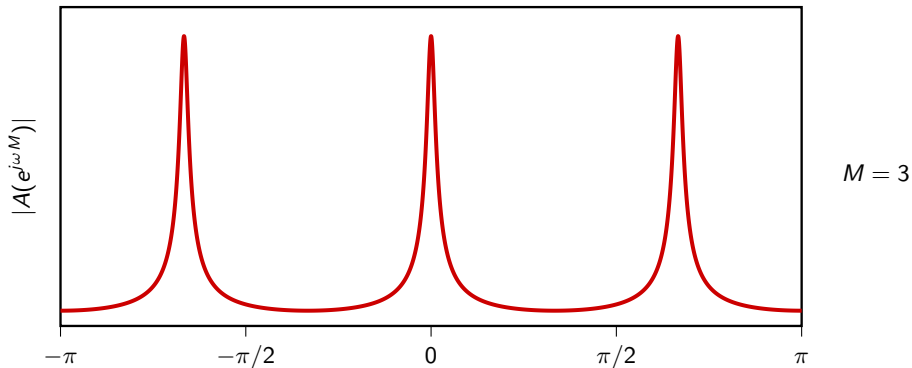
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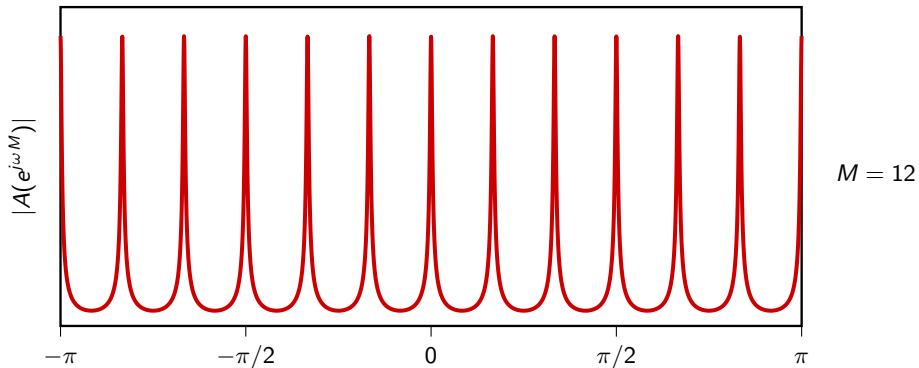
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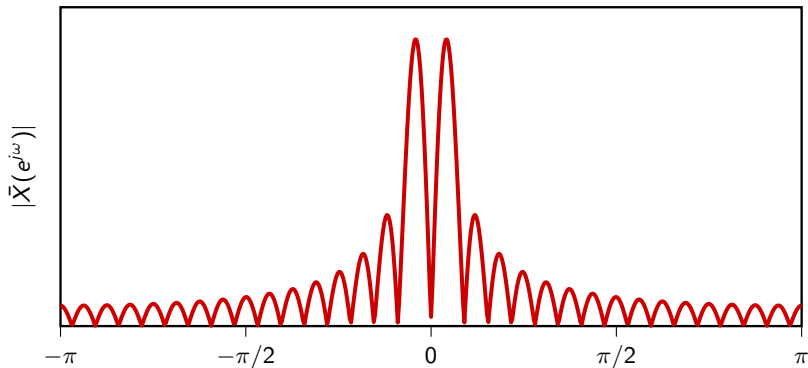
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Second term is left as an exercise

$$\bar{X}(e^{j\omega}) = e^{-j\omega} \left( \frac{M+1}{M-1} \right) \frac{1 - e^{-j(M-1)\omega}}{(1 - e^{-j\omega})^2} - \frac{1 - e^{-j(M+1)\omega}}{(1 - e^{-j\omega})^2}$$



## DTFT of KS with decay

$$Y(e^{j\omega}) = A(e^{j\omega M})\bar{X}(e^{j\omega})$$

