



Lecture 4

Electrical Design

AC Resistance & Inductance

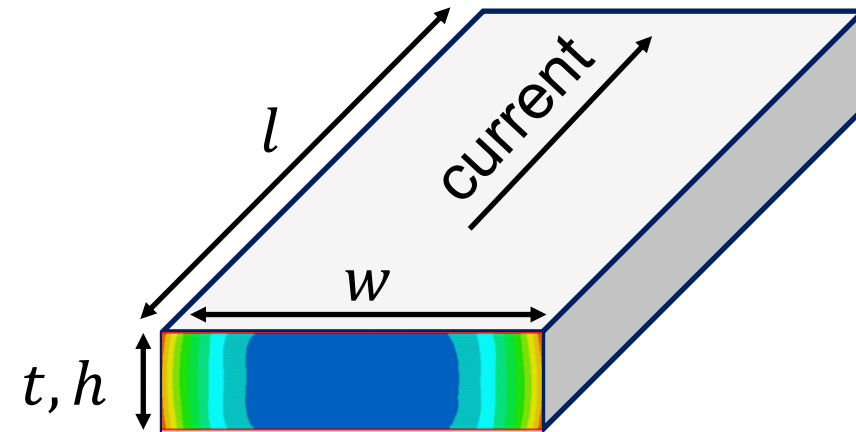
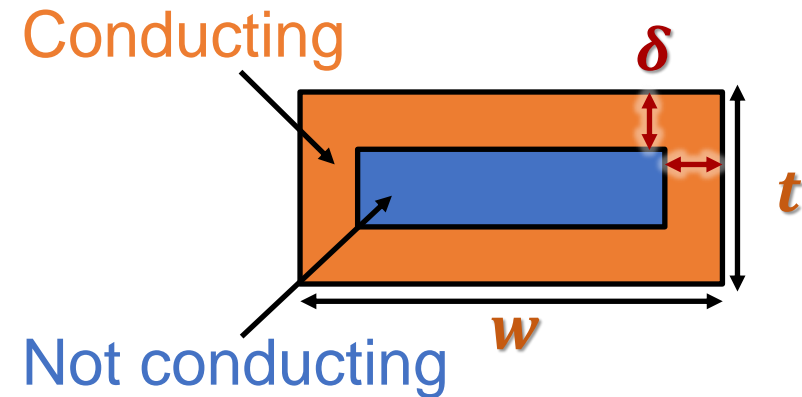
January 31, 2025

Reminders and Announcements

- Office hours: Monday, 3:30pm-4:30pm
- Quiz #1 answers and grades on Canvas
- Homework #1 will be assigned tomorrow (tentative)
- Download and install ANSYS Electronics Desktop and LTspice and check that you can open them before Feb. 3rd

AC Resistance of Rectangular Conductors (First-Order Approximation)

- $R_{AC} = \rho l / A_{eff}$
- When $2\delta \ll w$ and t
 - $A_{eff} = wt - (w - 2\delta)(t - 2\delta)$
 - $A_{eff} = 2w\delta + 2t\delta - 4\delta^2$
 - $A_{eff} = 2\delta(w + t - 2\delta)$
- $\delta = \sqrt{(\rho / (\pi f \mu))}$
- When $2\delta \geq w$ or t
 - $A_{eff} = wt$



Example: AC Resistance of PCB Trace

- 200-mm-long, 0.1-mm-wide PCB trace with 1 oz* copper
- $\rho_{copper} = 1.7 \times 10^{-8} \Omega \cdot m$
- 1 oz copper = 1.37 mils = 0.00137 in = 0.034798 mm
- What is the AC resistance of the copper trace at 150 MHz?

$$R_{AC} = \rho l / A_{eff}$$

$$\delta = \sqrt{(\rho / (\pi f \mu))}$$

$$= \sqrt{(1.7 \cdot 10^{-8} \Omega \cdot m / (\pi (150 \text{ MHz}) (4\pi \cdot 10^{-7} \text{ H/m})))} = 5.36 \cdot 10^{-6} \text{ m}$$

Check $2\delta \ll w$ and t : $10.7 \mu\text{m} \ll 100 \mu\text{m}$ and $35 \mu\text{m}$

$$A_{eff} = 2\delta(w + t - 2\delta)$$

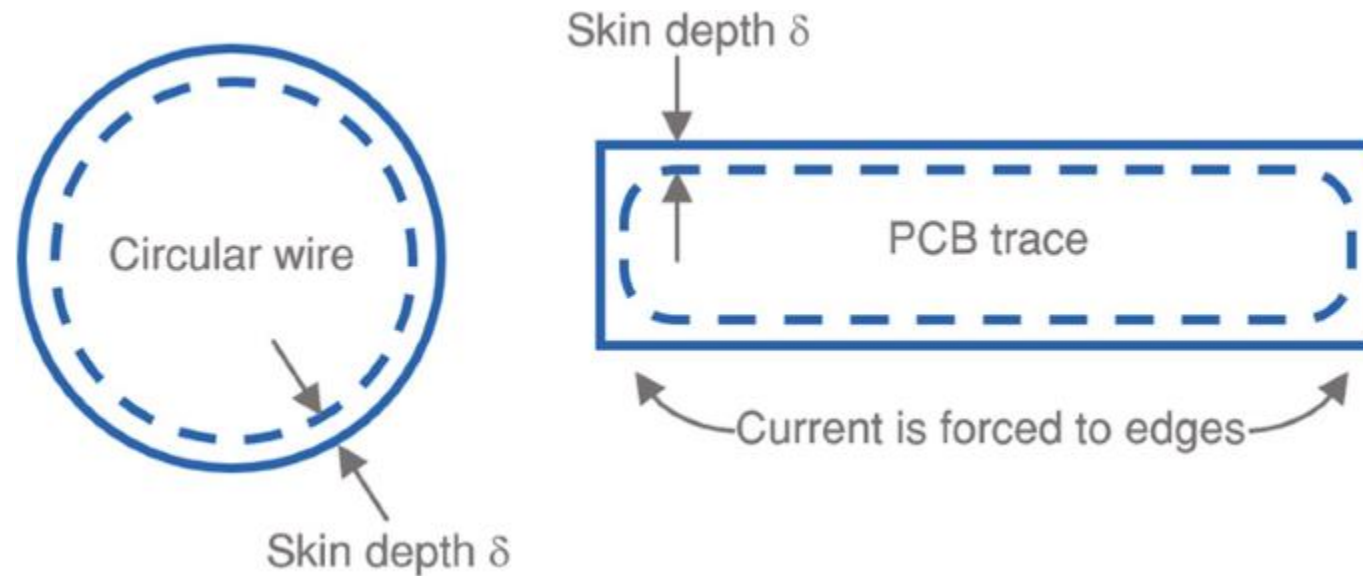
$$\begin{aligned} &= 2(5.36 \cdot 10^{-6} \text{ m})(1 \cdot 10^{-4} \text{ m} + 3.5 \cdot 10^{-5} \text{ m} - 2(5.36 \cdot 10^{-6} \text{ m})) \\ &= 1.33 \cdot 10^{-9} \text{ m}^2 \end{aligned}$$

$$R_{AC} = (1.7 \times 10^{-8} \Omega \cdot m) \cdot (0.2 \text{ m}) / (1.33 \times 10^{-9} \text{ m}^2) = 2.56 \Omega$$

→ 2.6x higher than the DC resistance

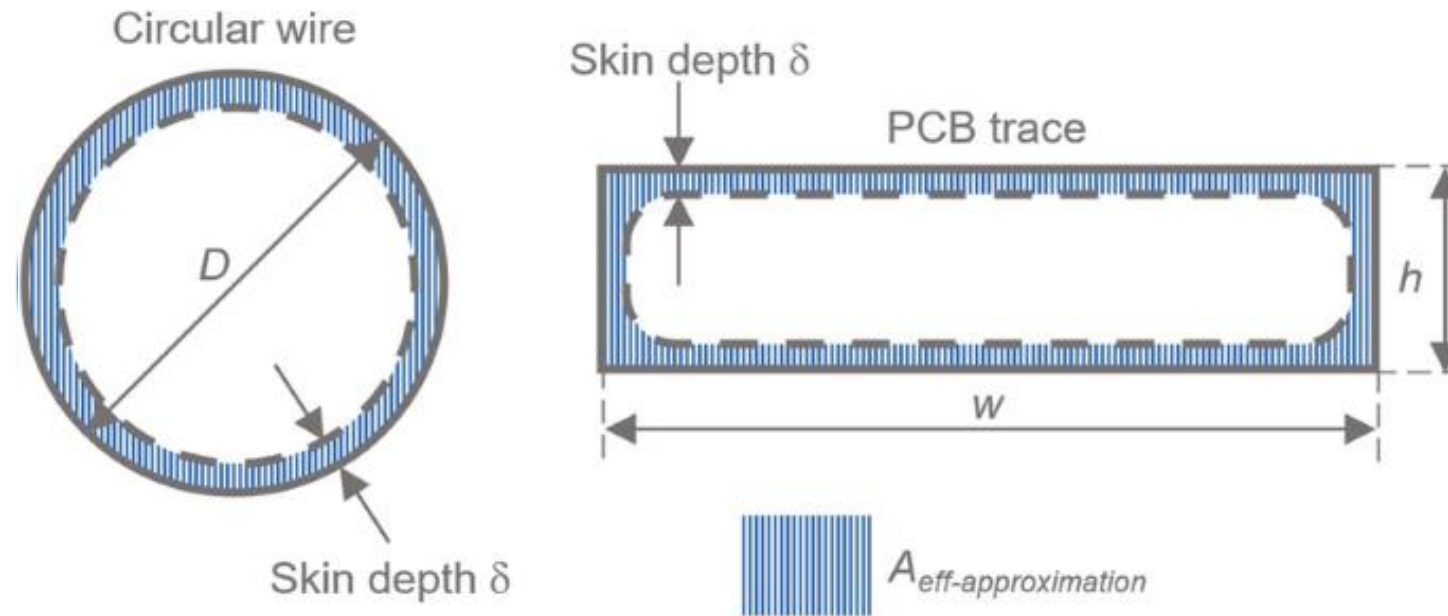
AC Resistance: Circular vs Rectangular Cross-Sections

- Circles have less perimeter for a given cross-sectional area, so the skin effect is more pronounced
- Flat, wide conductors with rectangular cross-sections are preferred for high-frequency applications



AC Resistance: Circular vs Rectangular Cross-Sections

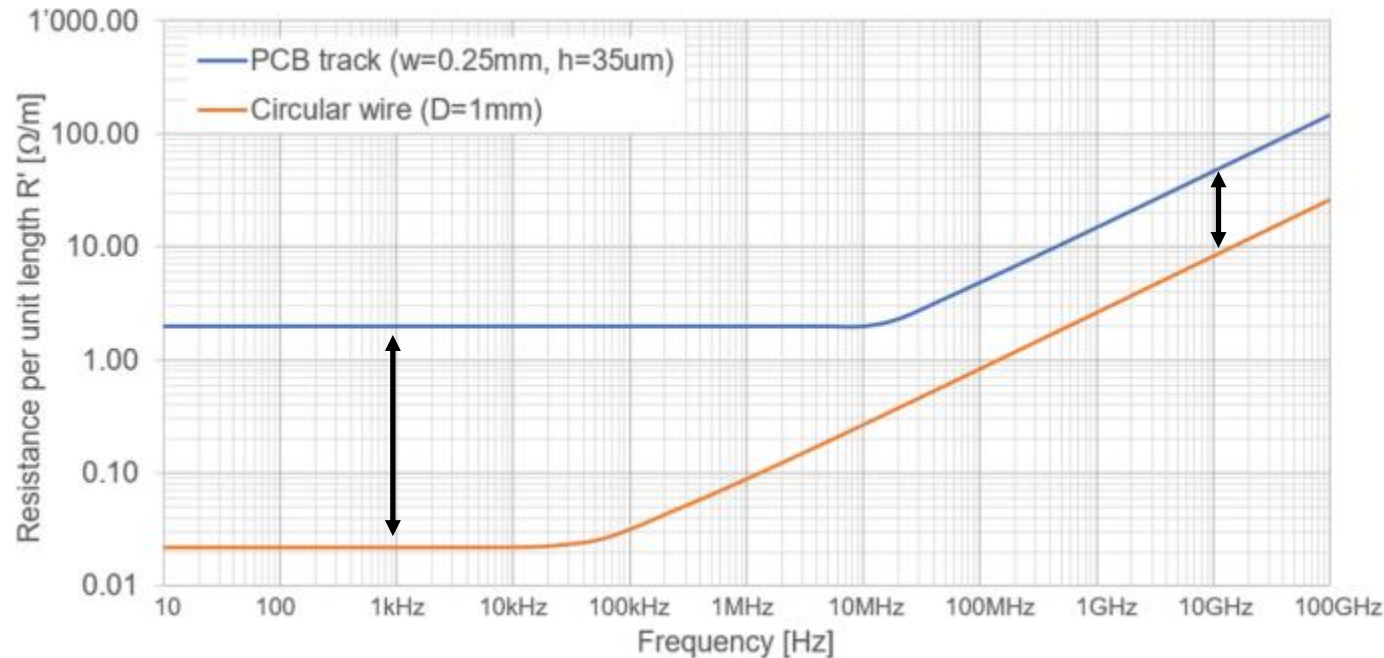
- Circles have less perimeter for a given cross-sectional area, so the skin effect is more pronounced
- Flat, wide conductors with rectangular cross-sections are preferred for high-frequency applications



Keller, R.B. (2023). Skin Effect. In: Design for Electromagnetic Compatibility--In a Nutshell. Springer, Cham. https://doi.org/10.1007/978-3-031-14186-7_10

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Resistance Reduction Methods

DC Resistance

$$R_1 = \rho_0 l / A \cdot [1 + \alpha(T_1 - T_0)]$$

- Material
 - High electrical conductivity (low electrical resistivity)
- Geometry
 - Increase cross-sectional area
 - Decrease length
 - Parallel conductors
- Application
 - Decrease temperature
 - Decrease current

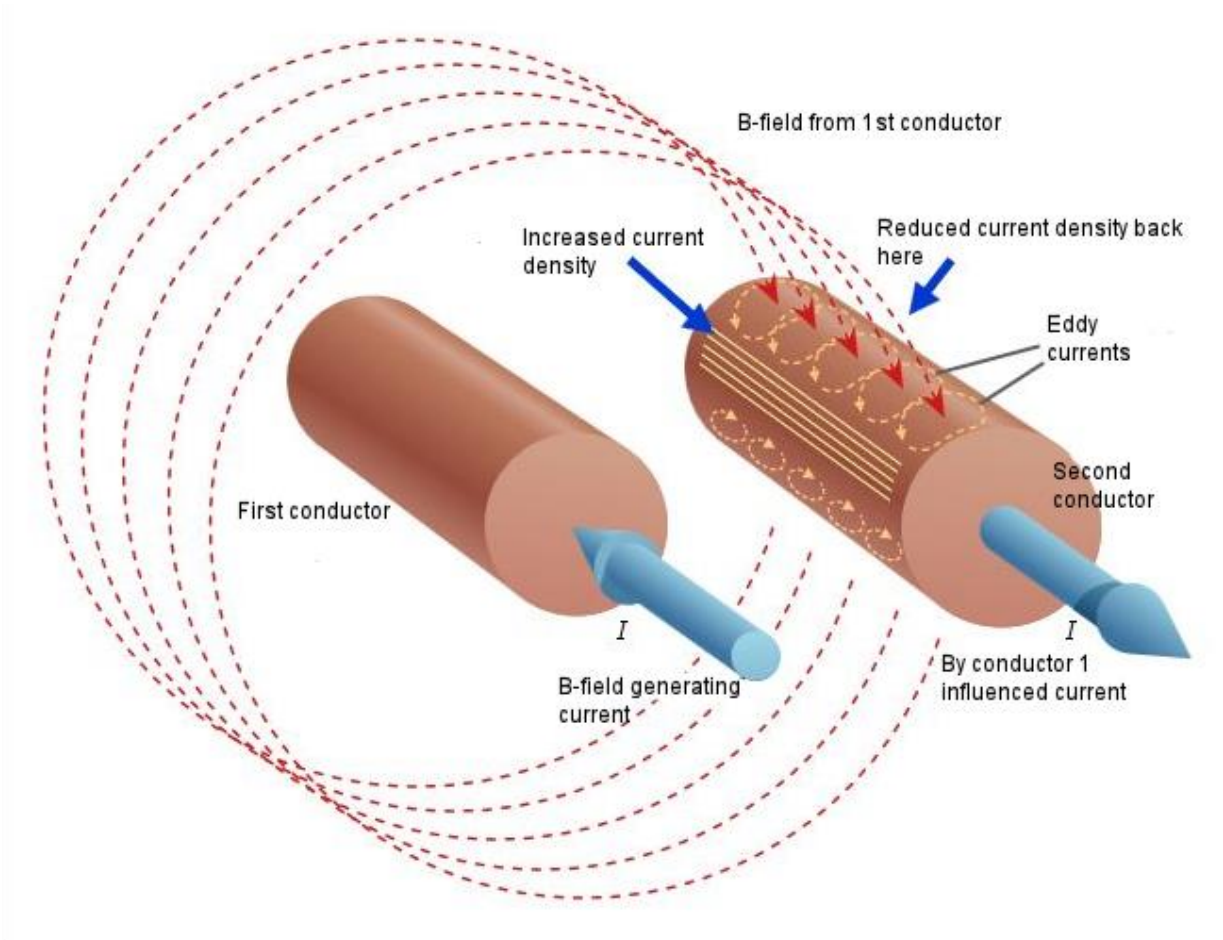
AC Resistance

$$\delta = \sqrt{(\rho / (\pi f \mu))}$$

- Material
 - Conductors with low permeability and low resistivity
- Geometry
 - Increase circumference/perimeter
 - Use wide flat conductor
 - Parallel smaller conductors
- Application
 - Decrease frequency

Skin and Proximity Effects

- Skin and proximity effects result from eddy currents
- When the magnetic field is generated by the conductor itself, this phenomena is called “skin effect”
- If the magnetic field is generated by an adjacent conductor, the phenomena is called “proximity effect”

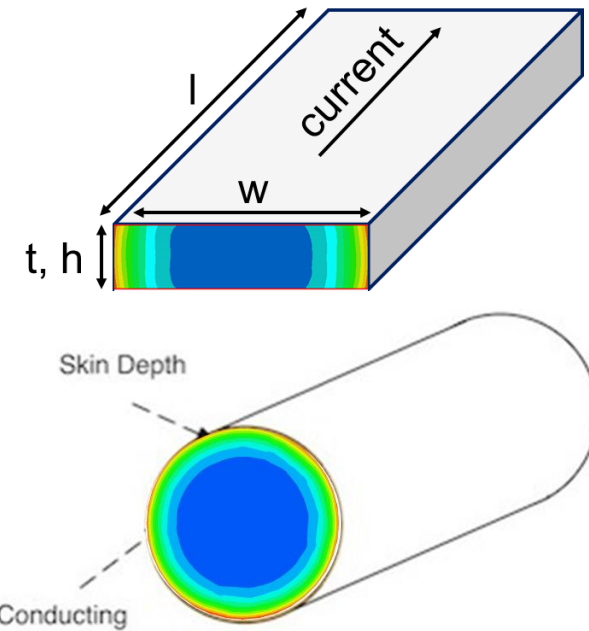


Proximity Effect

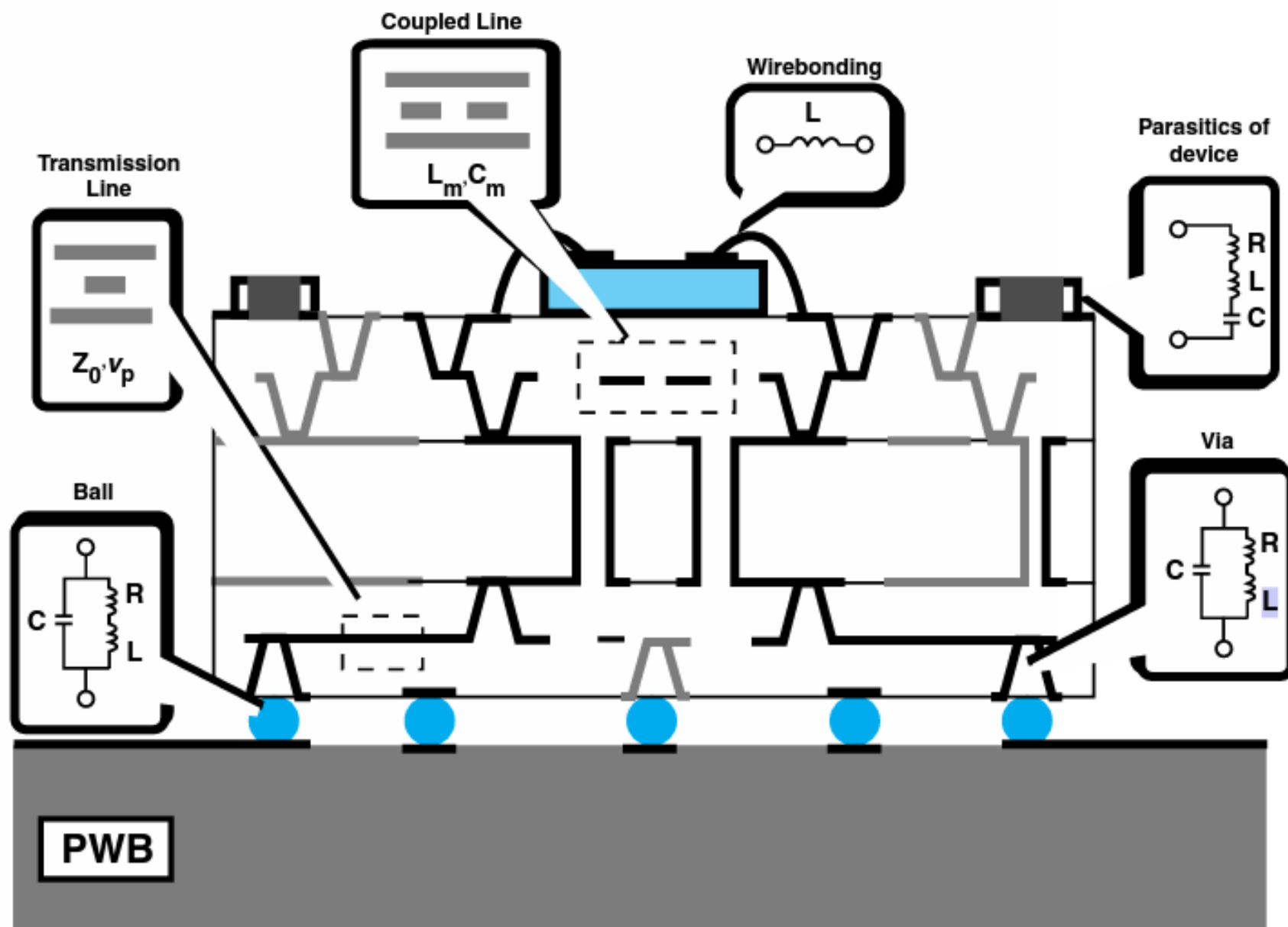
- Alternating current (AC) creates an alternating magnetic field around the conductor
- The alternating magnetic field induces eddy currents in adjacent conductors, changing the distribution of the current
- The result is current becoming concentrated in areas of the conductor closest to the first conductor
- Proximity effect significantly increases AC resistance of adjacent conductors
- The effect increases with frequency

Summary: Resistance

- DC resistance at room temperature: $R = \rho l / A$
- Temperature correction: $R = R_0 \cdot [1 + \alpha (T - T_0)]$
- AC resistance
 - Skin effect: $\delta = \sqrt{(\rho / (\pi f \mu))}$
 - AC current \rightarrow alternating magnetic field \rightarrow opposing EMF \rightarrow opposes current flow in center of wire \rightarrow majority of current flows through outer surface
 - Related to circumference, ρ , and μ of conductor, and frequency
 - Effective area for circular conductors: $A_{eff} = \pi(d\delta - \delta^2)$, when $\delta \ll r$
 - Effective area for rectangular conductors: $A_{eff} = 2\delta(w + t - 2\delta)$, when $2\delta \ll w$ and t

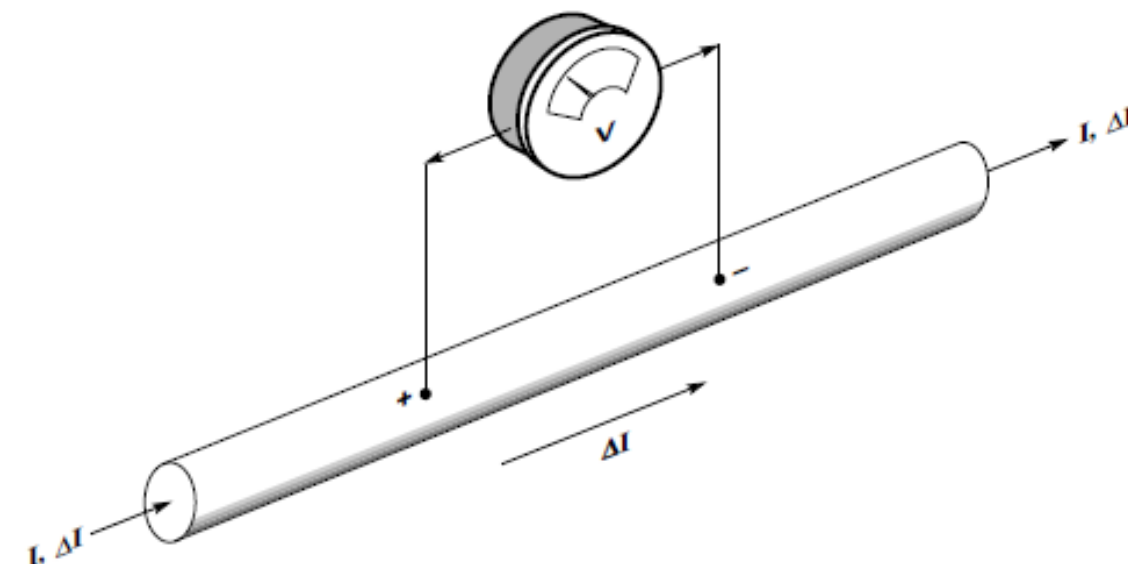


Package Parasitics



Inductance

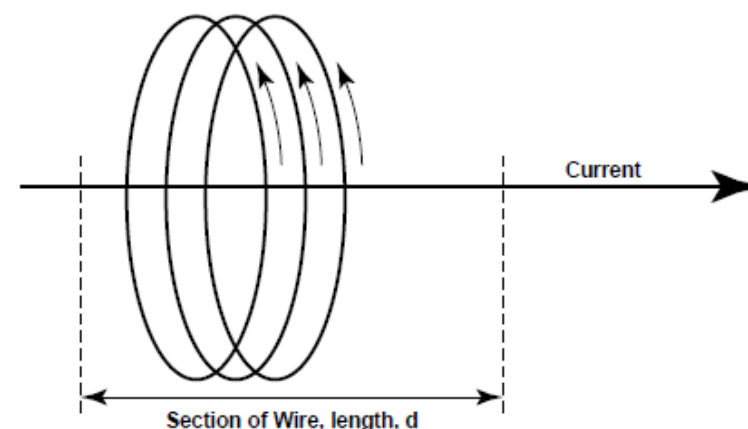
- Any current-carrying conductor can have parasitic inductance:
 - Wire / wire bond
 - Lead / terminals
 - PCB trace
- $V = L di/dt$, oscillations, coupling
- Types:
 - Self/partial inductance
 - Mutual inductance
 - Loop/effective/total inductance



Source: ICE, "Roadmaps of Packaging Technology"

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Figure 7-21. Voltage Induced in a Circuit Element Due to a Change in Current



Source: ICE, "Roadmaps of Packaging Technology"

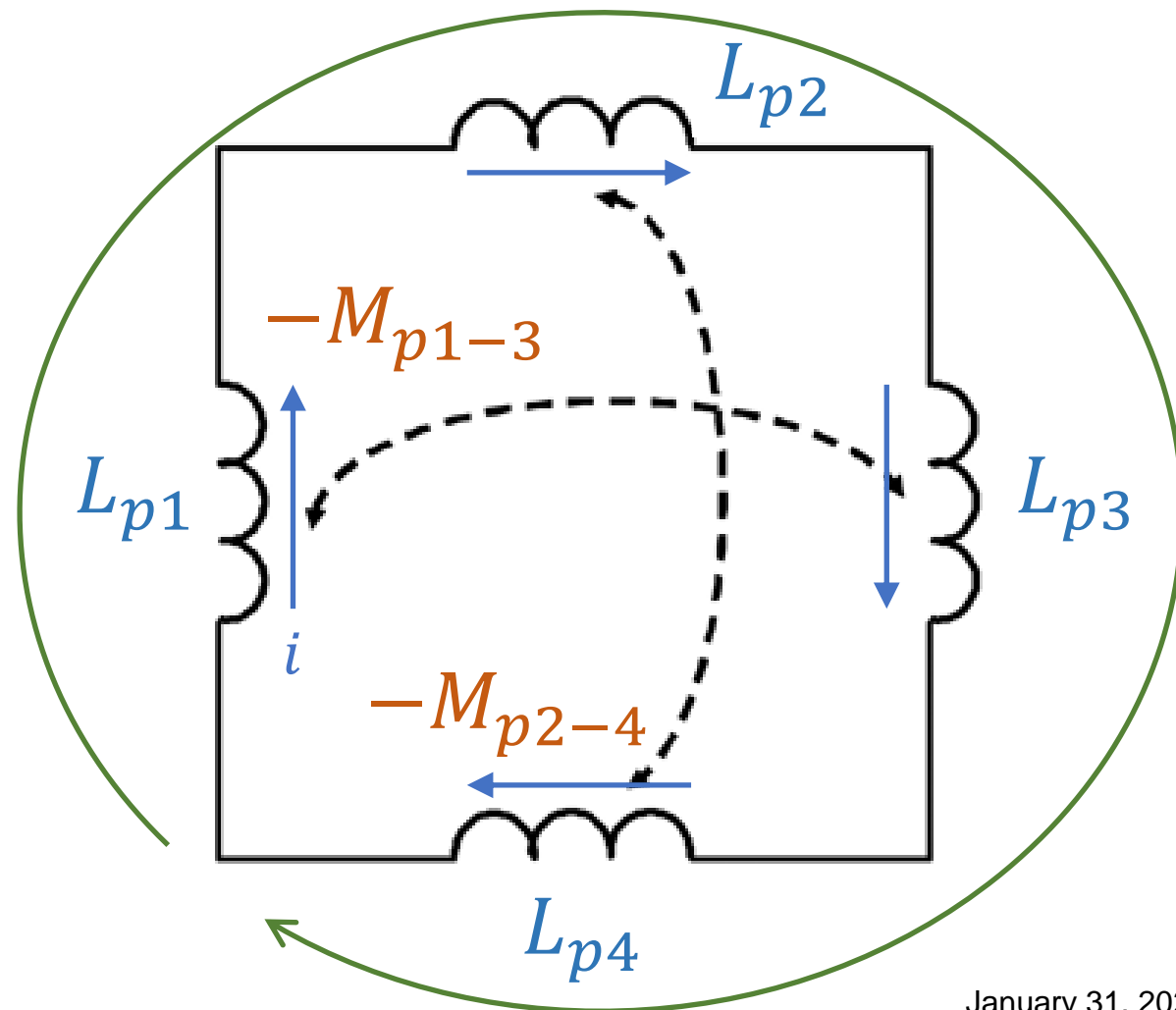
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Figure 7-22. Magnetic Field Lines About a Current Carrying Wire

Types of Inductance

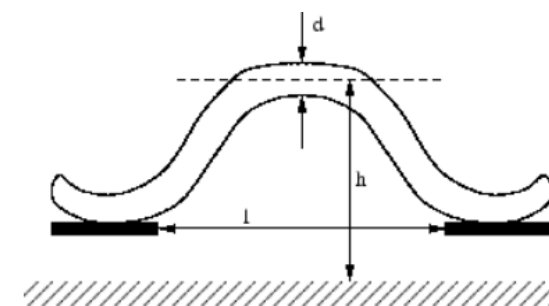
- Self/partial inductance, L_p
 - Its own magnetic field opposes any change in current
- Mutual inductance, M_p
 - Magnetic field lines from one conductor pass through another
- Loop/total/effective inductance, L_{total}
 - Inductance of the overall loop

$$L_{total} = L_{p1} + L_{p2} + L_{p3} + L_{p4} - 2M_{p1-3} - 2M_{p2-4}$$



Self Inductance: Wire (Round Conductor)

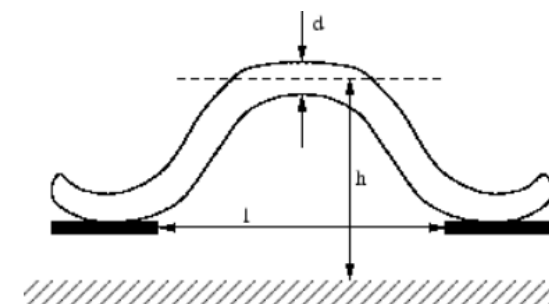
- $$L_{self} = \frac{\mu}{2\pi} l \left[\ln \left(\frac{l}{r} + \sqrt{1 + \frac{l^2}{r^2}} \right) - \sqrt{1 + \frac{r^2}{l^2}} + \frac{r}{l} + \frac{1}{4} \right]$$
 - l = length in meters
 - r = radius in meters
 - $\mu = \mu_0 \mu_r \approx (4\pi \times 10^{-7} \text{ H/m})(1) \approx (4\pi \times 10^{-7} \text{ H/m})$ (for non-ferromagnetic conductors)
 - For straight wires (does not consider curvature)



Self Inductance: Wire (Round Conductor)

When $l \gg r$:

- $L_{self} = 0.002l \left[\ln \left(\frac{2l}{r} \right) - \frac{3}{4} \right]$
(μH) (cm)
 - l = length in centimeters
 - r = radius in centimeters
 - For straight wires (does not consider curvature)



- Rule of thumb for inductance of a wire per unit length:
 - Self inductance ~ 25 nH/in (~ 1 nH/mm = 10 nH/cm = 0.01 $\mu\text{H}/\text{cm}$)

Example: Self Inductance of a Wire Bond

- $L_{self} = 0.002l \left[\ln \left(\frac{2l}{r} \right) - \frac{3}{4} \right]$

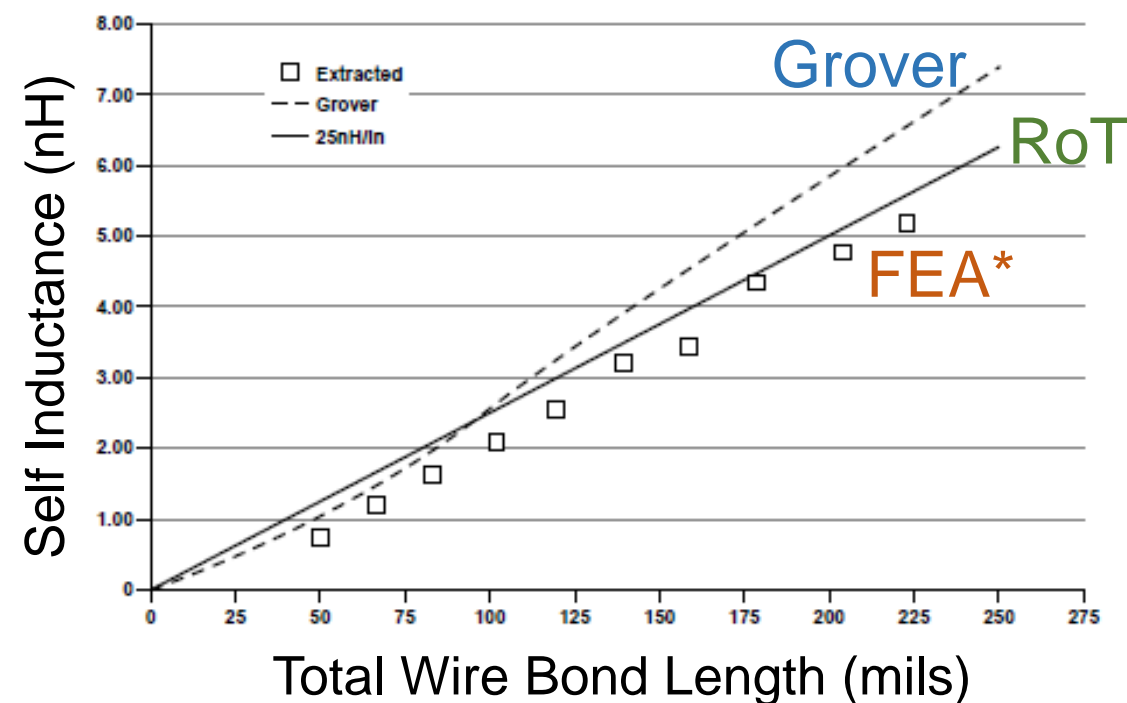
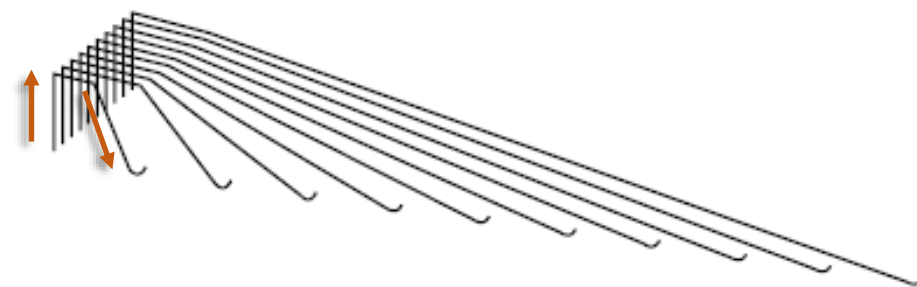
- $l = 10 \text{ mm} = 1 \text{ cm}$

- $r = 5 \text{ mils} = 0.0127 \text{ cm}$

- $L_{self} = 0.002(1\text{cm}) \left[\ln \left(\frac{2(1\text{cm})}{(0.0127\text{cm})} \right) - \frac{3}{4} \right]$
= 8.6 nH

- Rule of thumb:

- **1 nH/mm x 10 mm = 10 nH**



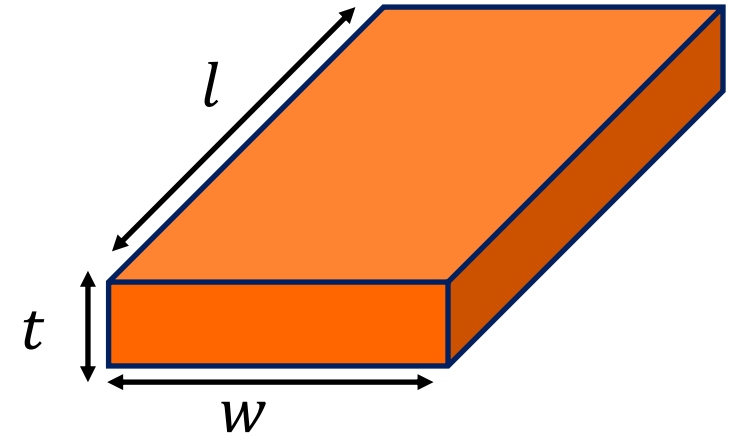
Source: Ansoft Corp./ICE, "Roadmaps of Packaging Technology"

Self Inductance: Rectangular Conductor

- In case of DC, low frequency, or a thin rectangular conductor, Grover gives the following self inductance formula:
- $$L_{self} = 0.002l \left(\ln \left(\frac{2l}{w+t} \right) + 0.50049 + \frac{w+t}{3l} \right)$$

(μH) (cm)

 - w = width in centimeters
 - l = length in centimeters
 - t = thickness in centimeters



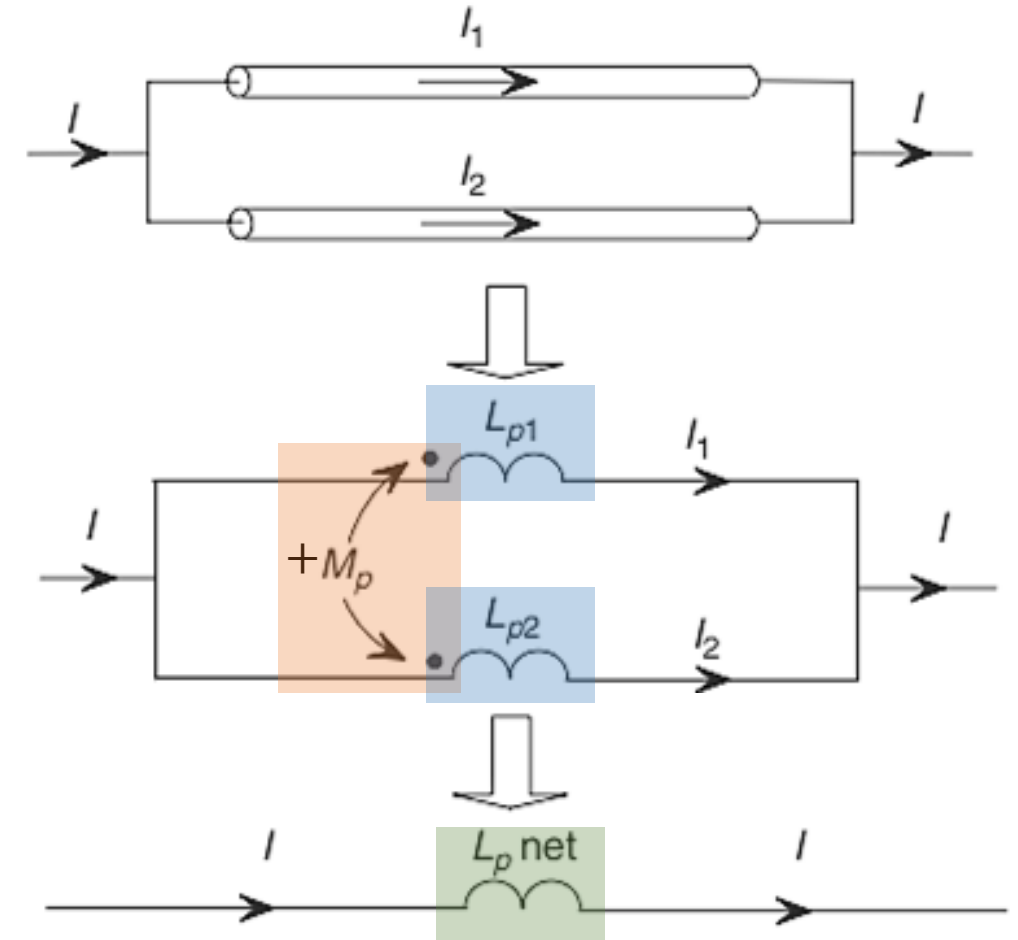
Example: Inductance of Parallel Wire Bonds

- If we have two identical wire bonds in parallel (electrically and physically), would $L_{eq} = L_p/2$?
 - Only if the wires are far apart!
 - If the wires are close together, there will be mutual inductance, M

$$L_{parallel} = \frac{L_{p1}L_{p2} - M^2}{L_{p1} + L_{p2} - 2M}$$

If $L_{p1} = L_{p2} = L_p$,

$$L_{parallel} = \frac{L_p + M}{2}$$



Mutual Inductance

- M = mutual inductance
 - Measure of shared field lines per amp of current in one conductor
- Crosstalk
 - Δi in one conductor induces V in the other
- Can increase total L for conductors carrying current in the same direction
- Return path cancellation could reduce total L

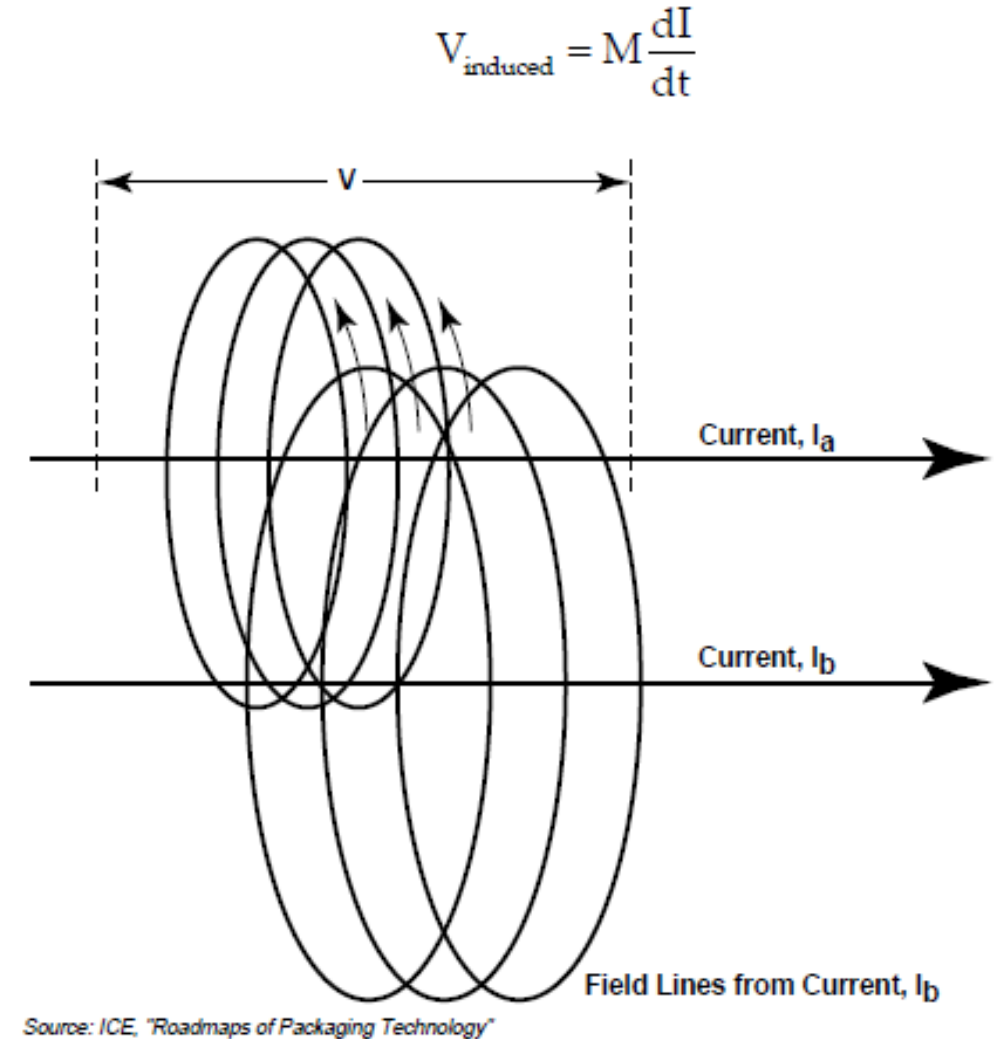


Figure 7-24. Origin of Mutual Inductance

Mutual Inductance

- $$M = 0.002l \left[\ln \left(\frac{l}{s} + \sqrt{1 + \left(\frac{l}{s} \right)^2} \right) - \sqrt{1 + \left(\frac{s}{l} \right)^2} + \frac{s}{l} \right]$$
 - l = length in centimeters
 - s = conductor spacing in centimeters
- When $s \ll l$:
$$M = \frac{\mu l}{2\pi} \left[\ln \left(\frac{2l}{s} \right) - 1 \right]$$

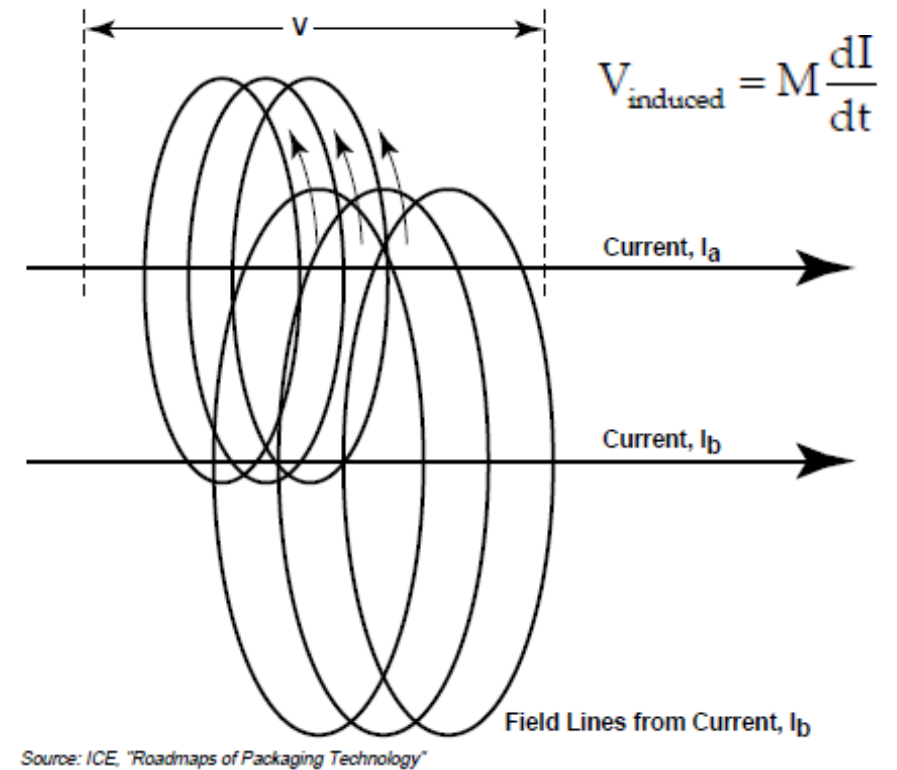


Figure 7-24. Origin of Mutual Inductance

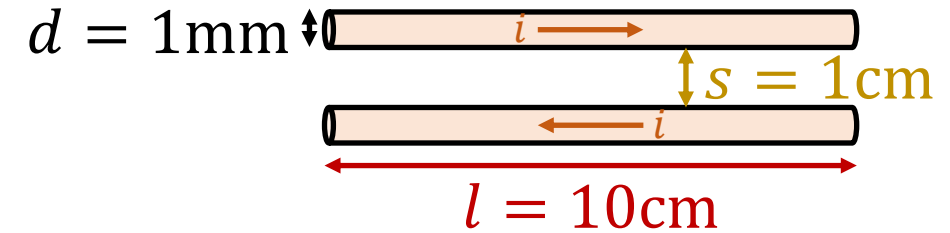
Mutual Inductance

- $$M = 0.002l \left[\ln \left(\frac{l}{s} + \sqrt{1 + \left(\frac{l}{s} \right)^2} \right) - \sqrt{1 + \left(\frac{s}{l} \right)^2} + \frac{s}{l} \right]$$
 - l = length in centimeters
 - s = conductor spacing in centimeters
- $$V = L_1 \frac{dI}{dt} - M_{12} \frac{dI}{dt} + IR + L_2 \frac{dI}{dt} - M_{12} \frac{dI}{dt}$$
- $$V = (L_1 + L_2 - 2M_{12}) \frac{dI}{dt} + IR$$



Example: Inductance of Adjacent Wires

- Find the equivalent inductance of two adjacent wires carrying current in opposite directions with 1-mm diameter, 10-cm length, and spaced 10 mm apart.



- For $r \ll l$ ($0.05\text{cm} \ll 10\text{cm}$), the self inductance of each wire is:

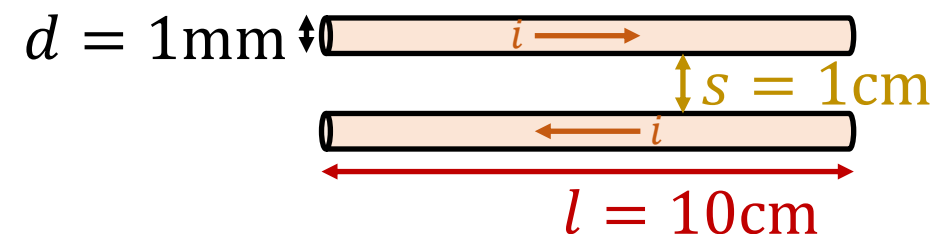
$$L(\mu\text{H}) = 0.002(10\text{cm}) \left[\ln \frac{(2)(10\text{cm})}{0.05\text{cm}} - \frac{3}{4} \right] = 0.02[5.99 - 0.75] = \mathbf{0.105 \mu\text{H}}$$

- For $s \ll l$ ($1\text{cm} \ll 10\text{cm}$), the mutual inductance is:

$$M = 0.002(10\text{cm}) \left[\ln \left(\frac{2(10\text{cm})}{(1\text{cm})} \right) - 1 \right] = \mathbf{0.040 \mu\text{H}}$$

Example: Inductance of Adjacent Wires

- Find the equivalent inductance of two adjacent wires carrying current in opposite directions with 1-mm diameter, 10-cm length, and spaced 10 mm apart.



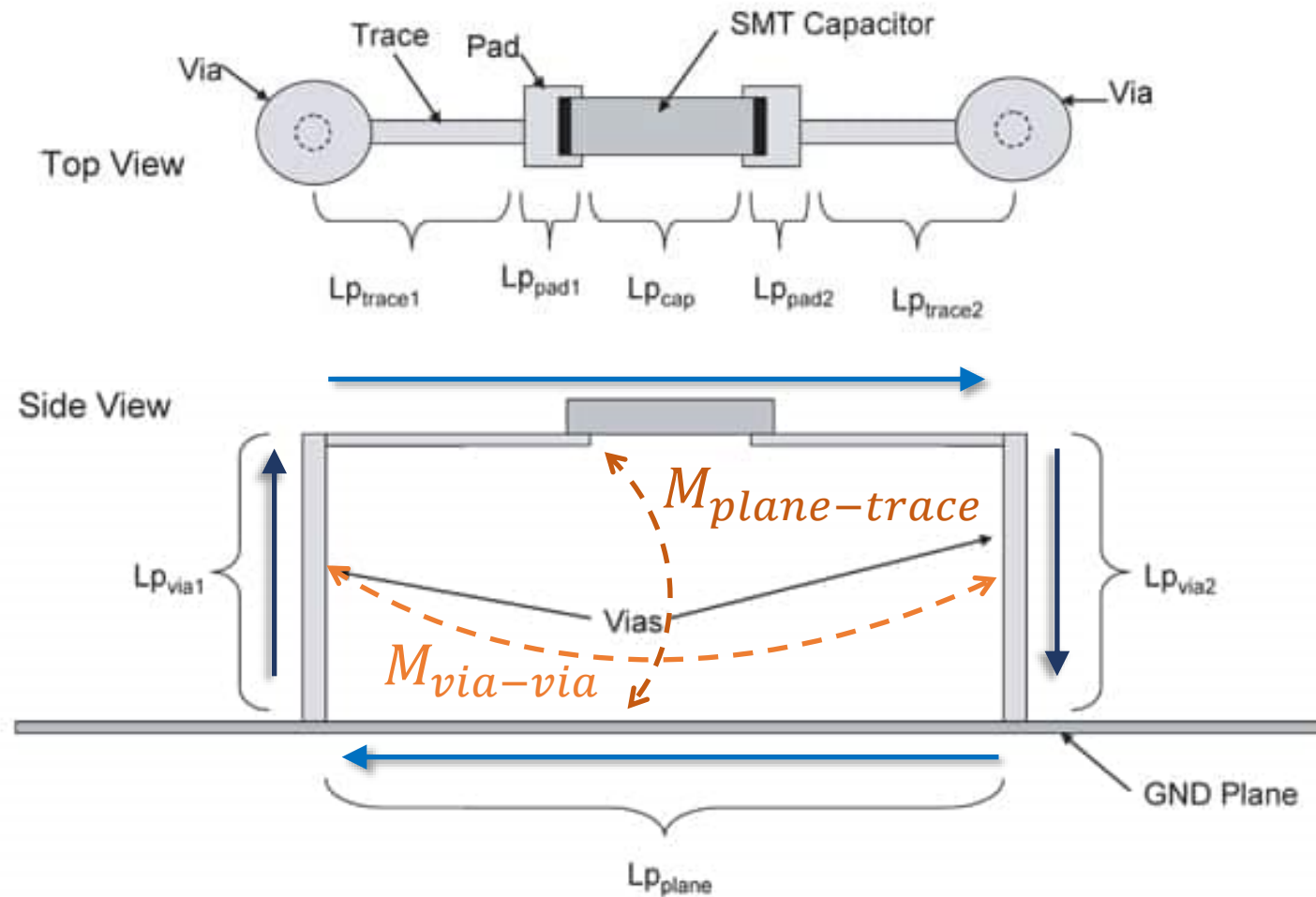
- Total inductance for the return circuit:

$$\begin{aligned}
 L_{eq} &= L_1 + L_2 - 2M_{12} \\
 &= 2(0.105 \mu\text{H}) - 2(0.042 \mu\text{H}) \\
 &= 0.126 \mu\text{H}
 \end{aligned}$$

If s is very large such that M_{12} is negligible, then $L_{tot} = 2(0.105 \mu\text{H})$

→ 40% reduction in L_{tot} due to M_{12}

Example: Capacitor Mounted to PCB



L_p = partial inductance
 M = mutual inductance of parallel components

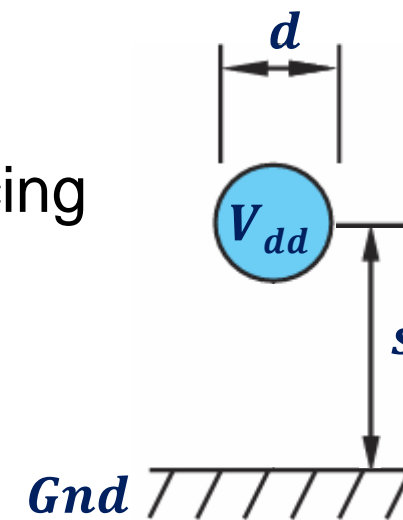
$$\begin{aligned}
 L_{total} = & L_{p_{trace1}} + L_{p_{pad1}} + L_{p_{cap}} + L_{p_{pad2}} + L_{p_{trace2}} + L_{p_{via2}} \\
 & + L_{p_{plane}} + L_{p_{via1}} - 2M_{via-via} - 2M_{plane-trace}
 \end{aligned}$$

Effective Inductances for Different Structures

*Note: Tummala textbook 2nd Ed. shows 4π in the denominator, which is incorrect.

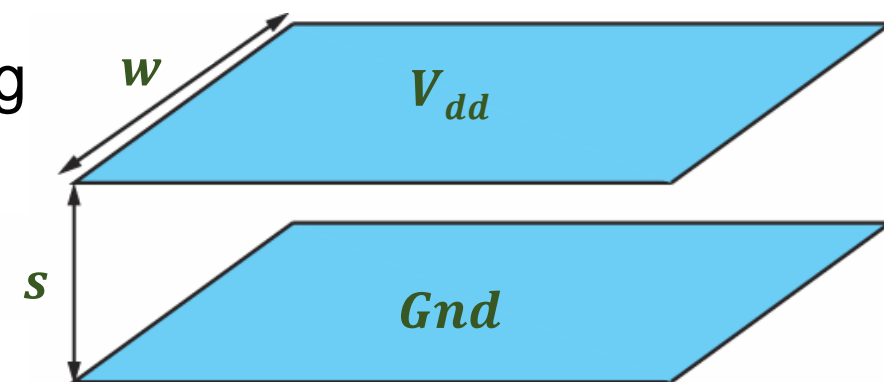
• Wire above a ground plane

- $L_{eff} = \frac{\mu l}{2\pi^*} \cosh^{-1} \left(\frac{2s}{d} \right)$, where l = length, d = diameter, s = spacing
- Package examples: TAB, QFP w/ ground plane
- Typical inductance range: 1 – 10 nH
- Assumes $s \ll l$, and $d \ll s$

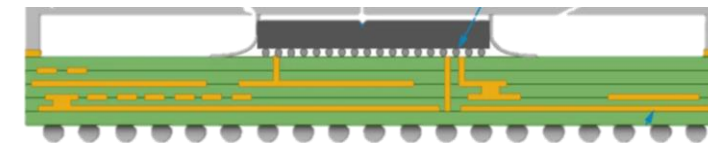


• Parallel planes

- $L_{eff} = \frac{\mu l s}{w}$, where l = length, w = width, s = spacing
- Package examples: PGA, BGA
- Typical inductance range: 0.25 – 1 nH
- Assumes $s \ll l$

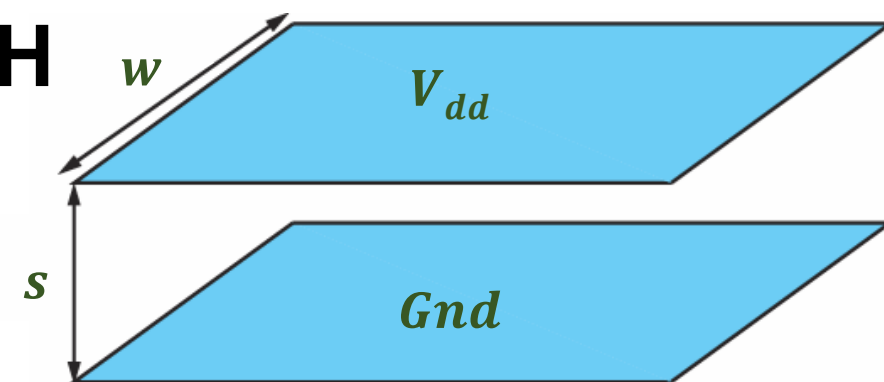


Example: Parallel Planes



A multi-layer ball grid array (BGA) package has plane layers used to supply both the V_{dd} and the ground (GND). Find the effective inductance for a pair of planes with dimensions of 1 cm by 1 cm, and a spacing of 6 mils.

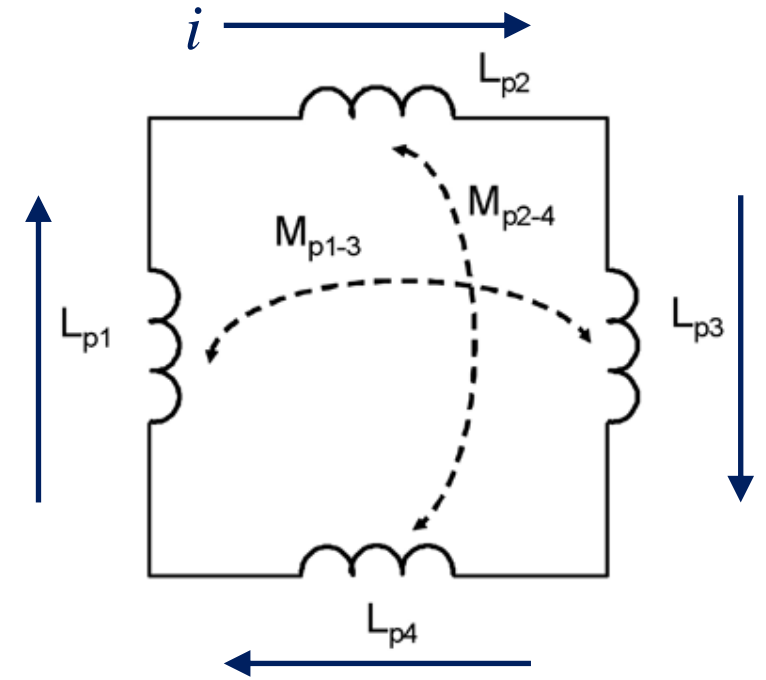
- $L_{eff} = \frac{\mu l s}{w}$, where $l = w = 1 \text{ cm} = 0.01 \text{ m}$, $s = 6 \text{ mils} = 1.5\text{e-}4 \text{ m}$
- $\mu = \mu_0 \mu_r \approx 4\pi \times 10^{-7} \text{ H/m} = 1.26\text{e-}6 \text{ H/m}$
- $L_{eff} = (1.26\text{e-}6 \text{ H/m})(1.5\text{e-}4 \text{ m}) = \mathbf{0.19 \text{ nH}}$



$$V = L \, dI/dt$$

Summary: Inductance

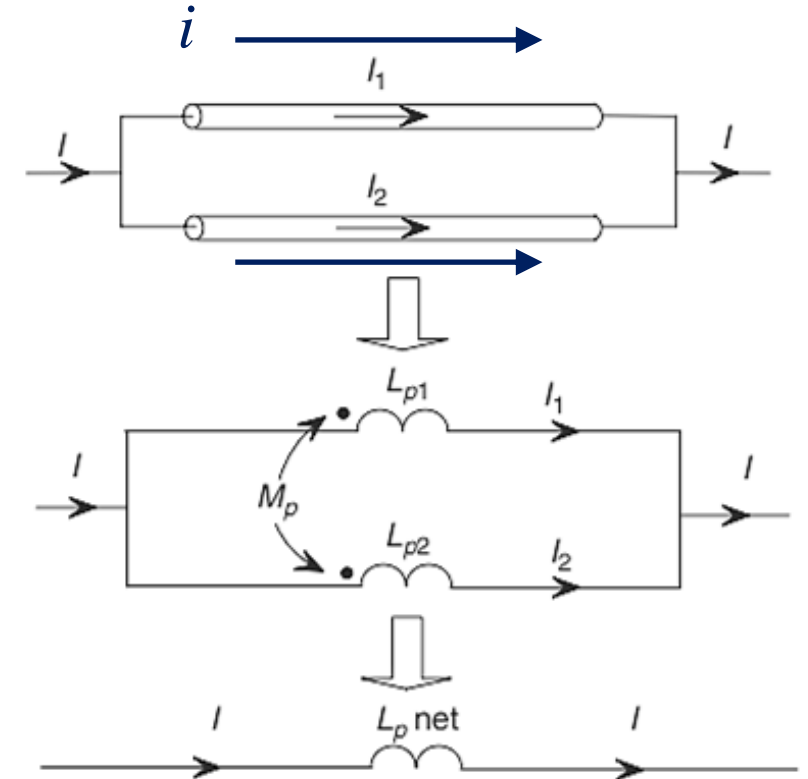
- Inductive delay: $\tau = L/R$
- Self inductance: L_p
 - Example: L_{p1} , L_{p2} , L_{p3} , L_{p4}
- Mutual inductance: M
 - Example: $-M_{p1-3}$, $-M_{p2-4}$
 - Subtractive: current flowing in opposite directions (this example)
- Loop inductance: L_{total}
- Example: $L_{total} = L_{p1} + L_{p2} + L_{p3} + L_{p4} - 2M_{p1-3} - 2M_{p2-4}$



$$V = L \, dI/dt$$

Summary: Inductance

- Example 2: Additive
 - Two parallel wires with current in the same direction
 - Self inductance: L_{p1} , L_{p2}
 - Mutual inductance: M
 - Positive
 - If $s \ll l$: $M = \frac{\mu l}{2\pi} \left[\ln \left(\frac{2l}{s} \right) - 1 \right]$
 - Total inductance if $L_{p1} = L_{p2}$:
 - $L_{parallel} = (L_p + M) / 2$



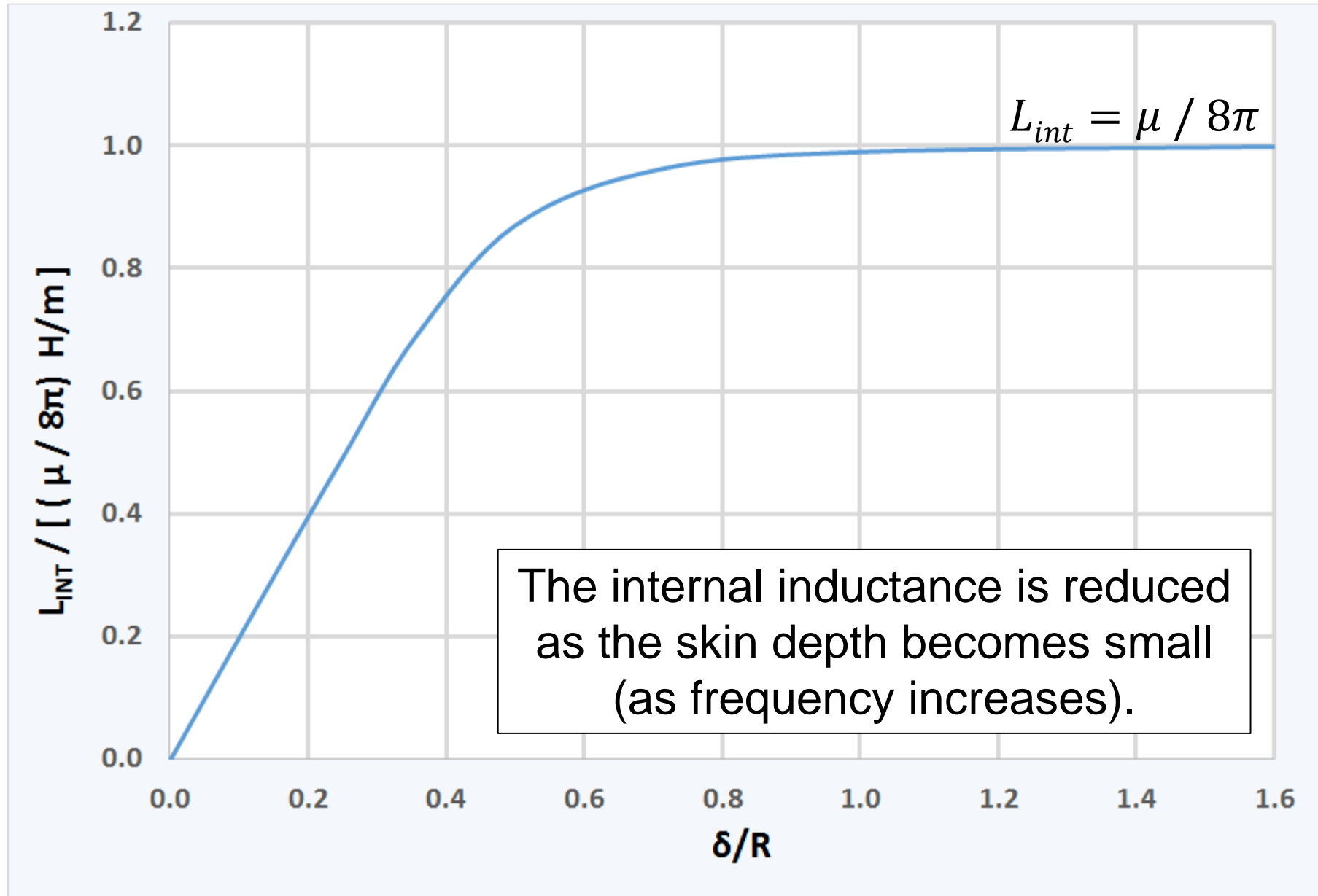
References on Inductances

- F. W. Grover, “Inductance calculations: working formulas and tables,” https://nvlpubs.nist.gov/nistpubs/bulletin/08/nbsbulletinv8n1p1_A2b.pdf
- Xiaoning Qi, “High frequency characterization and modeling of on-chip interconnects and RF IC wire bonds,” <http://www-tcad.stanford.edu/tcad/pubs/theses/qi.pdf>
- Clayton R. Paul, “Partial and internal inductance,” <https://ieeexplore.ieee.org/document/6507331>
- Eric Bogatin, “Roadmaps of packaging technology,” Chapter 7: Electrical Performance, ISBN: 1-877750-61-1, 1997.

Skin and Proximity Effects

- Skin effect reduces the *internal* wire inductance at high frequencies
- Internal inductance is due to the internal magnetic flux of a conductor
- Internal inductance is typically much less than the external inductance, which is due to external magnetic flux
- Proximity effect reduces the wire inductance by redistributing the currents to form a smaller current loop
- The skin and proximity effect eddy currents superimpose to form the total eddy current distribution

Internal Inductance vs. Skin Depth/Radius



Next Class

- Electrical Design (Chapter 2)
 - Inductance
 - Capacitance
 - Intro to Finite Element Analysis (FEA) and ANSYS Q3D