

= Mechanical filter =

A mechanical filter is a signal processing filter usually used in place of an electronic filter at radio frequencies . Its purpose is the same as that of a normal electronic filter : to pass a range of signal frequencies , but to block others . The filter acts on mechanical vibrations which are the analogue of the electrical signal . At the input and output of the filter , transducers convert the electrical signal into , and then back from , these mechanical vibrations .

The components of a mechanical filter are all directly analogous to the various elements found in electrical circuits . The mechanical elements obey mathematical functions which are identical to their corresponding electrical elements . This makes it possible to apply electrical network analysis and filter design methods to mechanical filters . Electrical theory has developed a large library of mathematical forms that produce useful filter frequency responses and the mechanical filter designer is able to make direct use of these . It is only necessary to set the mechanical components to appropriate values to produce a filter with an identical response to the electrical counterpart .

Steel and nickel ? iron alloys are common materials for mechanical filter components ; nickel is sometimes used for the input and output couplings . Resonators in the filter made from these materials need to be machined to precisely adjust their resonance frequency before final assembly .

While the meaning of mechanical filter in this article is one that is used in an electromechanical role , it is possible to use a mechanical design to filter mechanical vibrations or sound waves ( which are also essentially mechanical ) directly . For example , filtering of audio frequency response in the design of loudspeaker cabinets can be achieved with mechanical components . In the electrical application , in addition to mechanical components which correspond to their electrical counterparts , transducers are needed to convert between the mechanical and electrical domains . A representative selection of the wide variety of component forms and topologies for mechanical filters are presented in this article .

The theory of mechanical filters was first applied to improving the mechanical parts of phonographs in the 1920s . By the 1950s mechanical filters were being manufactured as self @-@ contained components for applications in radio transmitters and high @-@ end receivers . The high " quality factor " ,  $Q$  , that mechanical resonators can attain , far higher than that of an all @-@ electrical LC circuit , made possible the construction of mechanical filters with excellent selectivity . Good selectivity , being important in radio receivers , made such filters highly attractive . Contemporary researchers are working on microelectromechanical filters , the mechanical devices corresponding to electronic integrated circuits .

= = Elements = =

The elements of a passive linear electrical network consist of inductors , capacitors and resistors which have the properties of inductance , elastance ( inverse capacitance ) and resistance , respectively . The mechanical counterparts of these properties are , respectively , mass , stiffness and damping . In most electronic filter designs , only inductor and capacitor elements are used in the body of the filter ( although the filter may be terminated with resistors at the input and output ) . Resistances are not present in a theoretical filter composed of ideal components and only arise in practical designs as unwanted parasitic elements . Likewise , a mechanical filter would ideally consist only of components with the properties of mass and stiffness , but in reality some damping is present as well .

The mechanical counterparts of voltage and electric current in this type of analysis are , respectively , force (  $F$  ) and velocity (  $v$  ) and represent the signal waveforms . From this , a mechanical impedance can be defined in terms of the imaginary angular frequency ,  $j\omega$  , which entirely follows the electrical analogy .

The scheme presented in the table is known as the impedance analogy . Circuit diagrams produced using this analogy match the electrical impedance of the mechanical system seen by the electrical circuit , making it intuitive from an electrical engineering standpoint . There is also the mobility analogy , in which force corresponds to current and velocity corresponds to voltage . This has

equally valid results but requires using the reciprocals of the electrical counterparts listed above . Hence ,  $M \propto C$  ,  $S \propto 1 / L$  ,  $D \propto G$  where  $G$  is electrical conductance , the inverse of resistance . Equivalent circuits produced by this scheme are similar , but are the dual impedance forms whereby series elements become parallel , capacitors become inductors , and so on . Circuit diagrams using the mobility analogy more closely match the mechanical arrangement of the circuit , making it more intuitive from a mechanical engineering standpoint . In addition to their application to electromechanical systems , these analogies are widely used to aid analysis in acoustics .

Any mechanical component will unavoidably possess both mass and stiffness . This translates in electrical terms to an LC circuit , that is , a circuit consisting of an inductor and a capacitor , hence mechanical components are resonators and are often used as such . It is still possible to represent inductors and capacitors as individual lumped elements in a mechanical implementation by minimising ( but never quite eliminating ) the unwanted property . Capacitors may be made of thin , long rods , that is , the mass is minimised and the compliance is maximised . Inductors , on the other hand , may be made of short , wide pieces which maximise the mass in comparison to the compliance of the piece .

Mechanical parts act as a transmission line for mechanical vibrations . If the wavelength is short in comparison to the part then a lumped element model as described above is no longer adequate and a distributed element model must be used instead . The mechanical distributed elements are entirely analogous to electrical distributed elements and the mechanical filter designer can use the methods of electrical distributed element filter design .

= = History = =

= = Harmonic telegraph = = =

Mechanical filter design was developed by applying the discoveries made in electrical filter theory to mechanics . However , a very early example ( 1870s ) of acoustic filtering was the " harmonic telegraph " , which arose precisely because electrical resonance was poorly understood but mechanical resonance ( in particular , acoustic resonance ) was very familiar to engineers . This situation was not to last for long ; electrical resonance had been known to science for some time before this , and it was not long before engineers started to produce all @-@ electric designs for filters . In its time , though , the harmonic telegraph was of some importance . The idea was to combine several telegraph signals on one telegraph line by what would now be called frequency division multiplexing thus saving enormously on line installation costs . The key of each operator activated a vibrating electromechanical reed which converted this vibration into an electrical signal . Filtering at the receiving operator was achieved by a similar reed tuned to precisely the same frequency , which would only vibrate and produce a sound from transmissions by the operator with the identical tuning .

Versions of the harmonic telegraph were developed by Elisha Gray , Alexander Graham Bell , Ernest Mercadier and others . Its ability to act as a sound transducer to and from the electrical domain was to inspire the invention of the telephone .

= = Mechanical equivalent circuits = = =

Once the basics of electrical network analysis began to be established , it was not long before the ideas of complex impedance and filter design theories were carried over into mechanics by analogy . Kennelly , who was also responsible for introducing complex impedance , and Webster were the first to extend the concept of impedance into mechanical systems in 1920 . Mechanical admittance and the associated mobility analogy came much later and are due to Firestone in 1932 .

It was not enough to just develop a mechanical analogy . This could be applied to problems that were entirely in the mechanical domain , but for mechanical filters with an electrical application it is necessary to include the transducer in the analogy as well . Poincaré in 1907 was the first to

describe a transducer as a pair of linear algebraic equations relating electrical variables ( voltage and current ) to mechanical variables ( force and velocity ) . These equations can be expressed as a matrix relationship in much the same way as the  $z$  @-@ parameters of a two @-@ port network in electrical theory , to which this is entirely analogous :

<formula>

where  $V$  and  $I$  represent the voltage and current respectively on the electrical side of the transducer

Wegel , in 1921 , was the first to express these equations in terms of mechanical impedance as well as electrical impedance . The element <formula> is the open circuit mechanical impedance , that is , the impedance presented by the mechanical side of the transducer when no current is entering the electrical side . The element <formula> , conversely , is the clamped electrical impedance , that is , the impedance presented to the electrical side when the mechanical side is clamped and prevented from moving ( velocity is zero ) . The remaining two elements , <formula> and <formula> , describe the transducer forward and reverse transfer functions respectively . Once these ideas were in place , engineers were able to extend electrical theory into the mechanical domain and analyse an electromechanical system as a unified whole .

= = = Sound reproduction = = =

An early application of these new theoretical tools was in phonographic sound reproduction . A recurring problem with early phonograph designs was that mechanical resonances in the pickup and sound transmission mechanism caused excessively large peaks and troughs in the frequency response , resulting in poor sound quality . In 1923 , Harrison of the Western Electric Company filed a patent for a phonograph in which the mechanical design was entirely represented as an electrical circuit . The horn of the phonograph is represented as a transmission line , and is a resistive load for the rest of the circuit , while all the mechanical and acoustic parts ? from the pickup needle through to the horn ? are translated into lumped components according to the impedance analogy . The circuit arrived at is a ladder topology of series resonant circuits coupled by shunt capacitors . This can be viewed as a bandpass filter circuit . Harrison designed the component values of this filter to have a specific passband corresponding to the desired audio passband ( in this case 100 Hz to 6 kHz ) and a flat response . Translating these electrical element values back into mechanical quantities provided specifications for the mechanical components in terms of mass and stiffness , which in turn could be translated into physical dimensions for their manufacture . The resulting phonograph has a flat frequency response in its passband and is free of the resonances previously experienced . Shortly after this , Harrison filed another patent using the same methodology on telephone transmit and receive transducers .

Harrison used Campbell 's image filter theory , which was the most advanced filter theory available at the time . In this theory , filter design is viewed essentially as an impedance matching problem . More advanced filter theory was brought to bear on this problem by Norton in 1929 at Bell Labs . Norton followed the same general approach though he later described to Darlington the filter he designed as being " maximally flat " . Norton 's mechanical design predates the paper by Butterworth who is usually credited as the first to describe the electronic maximally flat filter . The equations Norton gives for his filter correspond to a singly terminated Butterworth filter , that is , one driven by an ideal voltage source with no impedance , whereas the form more usually given in texts is for the doubly terminated filter with resistors at both ends , making it hard to recognise the design for what it is . Another unusual feature of Norton 's filter design arises from the series capacitor , which represents the stiffness of the diaphragm . This is the only series capacitor in Norton 's representation , and without it , the filter could be analysed as a low @-@ pass prototype . Norton moves the capacitor out of the body of the filter to the input at the expense of introducing a transformer into the equivalent circuit ( Norton 's figure 4 ) . Norton has used here the " turning round the L " impedance transform to achieve this .

The definitive description of the subject from this period is Maxfield and Harrison 's 1926 paper . There , they describe not only how mechanical bandpass filters can be applied to sound

reproduction systems , but also apply the same principles to recording systems and describe a much improved disc cutting head .

== Volume production ==

The first volume production of mechanical filters was undertaken by Collins Radio Company starting in the 1950s . These were originally designed for telephone frequency @-@ division multiplex applications where there is commercial advantage in using high quality filters . Precision and steepness of the transition band leads to a reduced width of guard band , which in turn leads to the ability to squeeze more telephone channels into the same cable . This same feature is useful in radio transmitters for much the same reason . Mechanical filters quickly also found popularity in VHF / UHF radio intermediate frequency ( IF ) stages of the high end radio sets ( military , marine , amateur radio and the like ) manufactured by Collins . They were favoured in the radio application because they could achieve much higher Q @-@ factors than the equivalent LC filter . High Q allows filters to be designed which have high selectivity , important for distinguishing adjacent radio channels in receivers . They also had an advantage in stability over both LC filters and monolithic crystal filters . The most popular design for radio applications was torsional resonators because radio IF typically lies in the 100 to 500 kHz band .

== Transducers ==

Both magnetostrictive and piezoelectric transducers are used in mechanical filters . Piezoelectric transducers are favoured in recent designs since the piezoelectric material can also be used as one of the resonators of the filter , thus reducing the number of components and thereby saving space . They also avoid the susceptibility to extraneous magnetic fields of the magnetostrictive type of transducer .

== Magnetostrictive ==

A magnetostrictive material is one which changes shape when a magnetic field is applied . In reverse , it produces a magnetic field when distorted . The magnetostrictive transducer requires a coil of conducting wire around the magnetostrictive material . The coil either induces a magnetic field in the transducer and sets it in motion or else picks up an induced current from the motion of the transducer at the filter output . It is also usually necessary to have a small magnet to bias the magnetostrictive material into its operating range . It is possible to dispense with the magnets if the biasing is taken care of on the electronic side by providing a d.c. current superimposed on the signal , but this approach would detract from the generality of the filter design .

The usual magnetostrictive materials used for the transducer are either ferrite or compressed powdered iron . Mechanical filter designs often have the resonators coupled with steel or nickel @-@ iron wires , but on some designs , especially older ones , nickel wire may be used for the input and output rods . This is because it is possible to wind the transducer coil directly on to a nickel coupling wire since nickel is slightly magnetostrictive . However , it is not strongly so and coupling to the electrical circuit is weak . This scheme also has the disadvantage of eddy currents , a problem that is avoided if ferrites are used instead of nickel .

The coil of the transducer adds some inductance on the electrical side of the filter . It is common practice to add a capacitor in parallel with the coil so that an additional resonator is formed which can be incorporated into the filter design . While this will not improve performance to the extent that an additional mechanical resonator would , there is some benefit and the coil has to be there in any case .

== Piezoelectric ==

A piezoelectric material is one which changes shape when an electric field is applied . In reverse , it

produces an electric field when it is distorted . A piezoelectric transducer , in essence , is made simply by plating electrodes on to the piezoelectric material . Early piezoelectric materials used in transducers such as barium titanate had poor temperature stability . This precluded the transducer from functioning as one of the resonators ; it had to be a separate component . This problem was solved with the introduction of lead zirconate titanate ( abbreviated PZT ) which is stable enough to be used as a resonator . Another common piezoelectric material is quartz , which has also been used in mechanical filters . However , ceramic materials such as PZT are preferred for their greater electromechanical coupling coefficient .

One type of piezoelectric transducer is the Langevin type , named after a transducer used by Paul Langevin in early sonar research . This is good for longitudinal modes of vibration . It can also be used on resonators with other modes of vibration if the motion can be mechanically converted into a longitudinal motion . The transducer consists of a layer of piezoelectric material sandwiched transversally into a coupling rod or resonator .

Another kind of piezoelectric transducer has the piezoelectric material sandwiched in longitudinally , usually into the resonator itself . This kind is good for torsional vibration modes and is called a torsional transducer .

= = Resonators = =

It is possible to achieve an extremely high Q with mechanical resonators . Mechanical resonators typically have a Q of 10 @,@ 000 or so , and 25 @,@ 000 can be achieved in torsional resonators using a particular nickel @-@ iron alloy . This is an unreasonably high figure to achieve with LC circuits , whose Q is limited by the resistance of the inductor coils .

Early designs in the 1940s and 1950s started by using steel as a resonator material . This has given way to nickel @-@ iron alloys , primarily to maximise the Q since this is often the primary appeal of mechanical filters rather than price . Some of the metals that have been used for mechanical filter resonators and their Q are shown in the table .

Piezoelectric crystals are also sometimes used in mechanical filter designs . This is especially true for resonators that are also acting as transducers for inputs and outputs .

One advantage that mechanical filters have over LC electrical filters is that they can be made very stable . The resonance frequency can be made so stable that it varies only 1 @.@ 5 parts per billion ( ppb ) from the specified value over the operating temperature range ( ? 25 to 85 ° C ) , and its average drift with time can be as low as 4 ppb per day . This stability with temperature is another reason for using nickel @-@ iron as the resonator material . Variations with temperature in the resonance frequency ( and other features of the frequency function ) are directly related to variations in the Young 's modulus , which is a measure of stiffness of the material . Materials are therefore sought that have a small temperature coefficient of Young 's modulus . In general , Young 's modulus has a negative temperature coefficient ( materials become less stiff with increasing temperature ) but additions of small amounts of certain other elements in the alloy can produce a material with a temperature coefficient that changes sign from negative through zero to positive with temperature . Such a material will have a zero coefficient of temperature with resonance frequency around a particular temperature . It is possible to adjust the point of zero temperature coefficient to a desired position by heat treatment of the alloy .

= = = Resonator modes = = =

It is usually possible for a mechanical part to vibrate in a number of different modes , however the design will be based on a particular vibrational mode and the designer will take steps to try to restrict the resonance to this mode . As well as the straightforward longitudinal mode some others which are used include flexural mode , torsional mode , radial mode and drumhead mode .

Modes are numbered according to the number of half @-@ wavelengths in the vibration . Some modes exhibit vibrations in more than one direction ( such as drumhead mode which has two ) and consequently the mode number consists of more than one number . When the vibration is in one of

the higher modes , there will be multiple nodes on the resonator where there is no motion . For some types of resonator , this can provide a convenient place to make a mechanical attachment for structural support . Wires attached at nodes will have no effect on the vibration of the resonator or the overall filter response . In figure 5 , some possible anchor points are shown as wires attached at the nodes . The modes shown are ( 5a ) the second longitudinal mode fixed at one end , ( 5b ) the first torsional mode , ( 5c ) the second torsional mode , ( 5d ) the second flexural mode , ( 5e ) first radial expansion mode and ( 5f ) first radially symmetric drumhead mode .

= = Circuit designs = =

There are a great many combinations of resonators and transducers that can be used to construct a mechanical filter . A selection of some of these is shown in the diagrams . Figure 6 shows a filter using disc flexural resonators and magnetostrictive transducers . The transducer drives the centre of the first resonator , causing it to vibrate . The edges of the disc move in antiphase to the centre when the driving signal is at , or close to , resonance , and the signal is transmitted through the connecting rods to the next resonator . When the driving signal is not close to resonance , there is little movement at the edges , and the filter rejects ( does not pass ) the signal . Figure 7 shows a similar idea involving longitudinal resonators connected together in a chain by connecting rods . In this diagram , the filter is driven by piezoelectric transducers . It could equally well have used magnetostrictive transducers . Figure 8 shows a filter using torsional resonators . In this diagram , the input has a torsional piezoelectric transducer and the output has a magnetostrictive transducer . This would be quite unusual in a real design , as both input and output usually have the same type of transducer . The magnetostrictive transducer is only shown here to demonstrate how longitudinal vibrations may be converted to torsional vibrations and vice versa . Figure 9 shows a filter using drumhead mode resonators . The edges of the discs are fixed to the casing of the filter ( not shown in the diagram ) so the vibration of the disc is in the same modes as the membrane of a drum . Collins calls this type of filter a disc wire filter .

The various types of resonator are all particularly suited to different frequency bands . Overall , mechanical filters with lumped elements of all kinds can cover frequencies from about 5 to 700 kHz although mechanical filters down as low as a few kilohertz ( kHz ) are rare . The lower part of this range , below 100 kHz , is best covered with bar flexural resonators . The upper part is better done with torsional resonators . Drumhead disc resonators are in the middle , covering the range from around 100 to 300 kHz .

The frequency response behaviour of all mechanical filters can be expressed as an equivalent electrical circuit using the impedance analogy described above . An example of this is shown in figure 8b which is the equivalent circuit of the mechanical filter of figure 8a . Elements on the electrical side , such as the inductance of the magnetostrictive transducer , are omitted but would be taken into account in a complete design . The series resonant circuits on the circuit diagram represent the torsional resonators , and the shunt capacitors represent the coupling wires . The component values of the electrical equivalent circuit can be adjusted , more or less at will , by modifying the dimensions of the mechanical components . In this way , all the theoretical tools of electrical analysis and filter design can be brought to bear on the mechanical design . Any filter realisable in electrical theory can , in principle , also be realised as a mechanical filter . In particular , the popular finite element approximations to an ideal filter response of the Butterworth and Chebyshev filters can both readily be realised . As with the electrical counterpart , the more elements that are used , the closer the approximation approaches the ideal , however , for practical reasons the number of resonators does not normally exceed eight .

= = = Semi @-@ lumped designs = = =

Frequencies of the order of megahertz ( MHz ) are above the usual range for mechanical filters . The components start to become very small , or alternatively the components are large compared to the signal wavelength . The lumped element model described above starts to break down and the

components must be considered as distributed elements . The frequency at which the transition from lumped to distributed models takes place is much lower for mechanical filters than it is for their electrical counterparts . This is because mechanical vibrations travel at the speed of sound for the material the component is composed of . For solid components , this is many times (  $\times 15$  for nickel @-@ iron ) the speed of sound in air (  $343 \text{ m / s}$  ) but still considerably less than the speed of electromagnetic waves ( approx .  $3 \times 10^8 \text{ m / s}$  in vacuum ) . Consequently , mechanical wavelengths are much shorter than electrical wavelengths for the same frequency . Advantage can be taken of these effects by deliberately designing components to be distributed elements , and the components and methods used in electrical distributed element filters can be brought to bear . The equivalents of stubs and impedance transformers are both achievable . Designs which use a mixture of lumped and distributed elements are referred to as semi @-@ lumped .

An example of such a design is shown in figure 10a . The resonators are disc flexural resonators similar to those shown in figure 6 , except that these are energised from an edge , leading to vibration in the fundamental flexural mode with a node in the centre , whereas the figure 6 design is energised in the centre leading to vibration in the second flexural mode at resonance . The resonators are mechanically attached to the housing by pivots at right angles to the coupling wires . The pivots are to ensure free turning of the resonator and minimise losses . The resonators are treated as lumped elements ; however , the coupling wires are made exactly one half @-@ wavelength (  $\lambda / 2$  ) long and are equivalent to a  $\lambda / 2$  open circuit stub in the electrical equivalent circuit . For a narrow @-@ band filter , a stub of this sort has the approximate equivalent circuit of a parallel shunt tuned circuit as shown in figure 10b . Consequently , the connecting wires are being used in this design to add additional resonators into the circuit and will have a better response than one with just the lumped resonators and short couplings . For even higher frequencies , microelectromechanical methods can be used as described below .

== Bridging wires ==

Bridging wires are rods that couple together resonators that are not adjacent . They can be used to produce poles of attenuation in the stopband . This has the benefit of increasing the stopband rejection . When the pole is placed near the passband edge , it also has the benefit of increasing roll @-@ off and narrowing the transition band . The typical effects of some of these on filter frequency response are shown in figure 11 . Bridging across a single resonator ( figure 11b ) can produce a pole of attenuation in the high stopband . Bridging across two resonators ( figure 11c ) can produce a pole of attenuation in both the high and the low stopband . Using multiple bridges ( figure 11d ) will result in multiple poles of attenuation . In this way , the attenuation of the stopbands can be deepened over a broad frequency range .

The method of coupling between non @-@ adjacent resonators is not limited to mechanical filters . It can be applied to other filter formats and the general term for this class is cross @-@ coupled filter . For instance , channels can be cut between cavity resonators , mutual inductance can be used with discrete component filters , and feedback paths can be used with active analogue or digital filters . Nor was the method first discovered in the field of mechanical filters ; the earliest description is in a 1948 patent for filters using microwave cavity resonators . However , mechanical filter designers were the first ( 1960s ) to develop practical filters of this kind and the method became a particular feature of mechanical filters .

== Microelectromechanical filters ==

A new technology emerging in mechanical filtering is microelectromechanical systems ( MEMS ) . MEMS are very small micromachines with component sizes measured in micrometres (  $\mu\text{m}$  ) , but not as small as nanomachines . These systems are mostly fabricated from silicon ( Si ) , silicon nitride (  $\text{Si}_3\text{N}_4$  ) , or polymers . A common component used for radio frequency filtering ( and MEMS applications generally ) , is the cantilever resonator . Cantilevers are simple mechanical components to manufacture by much the same methods used by the semiconductor industry ; masking ,

photolithography and etching , with a final undercutting etch to separate the cantilever from the substrate . The technology has great promise since cantilevers can be produced in large numbers on a single substrate ? much as large numbers of transistors are currently contained on a single silicon chip .

The resonator shown in figure 12 is around 120 ?m in length . Experimental complete filters with an operating frequency of 30 GHz have been produced using cantilever varactors as the resonator elements . The size of this filter is around 4 × 3 @. @ 5 mm . Cantilever resonators are typically applied at frequencies below 200 MHz , but other structures , such as micro @- @ machined cavities , can be used in the microwave bands . Extremely high Q resonators can be made with this technology ; flexural mode resonators with a Q in excess of 80 @, @ 000 at 8 MHz are reported .

= = Adjustment = =

The precision applications in which mechanical filters are used require that the resonators are accurately adjusted to the specified resonance frequency . This is known as trimming and usually involves a mechanical machining process . In most filter designs , this can be difficult to do once the resonators have been assembled into the complete filter so the resonators are trimmed before assembly . Trimming is done in at least two stages ; coarse and fine , with each stage bringing the resonance frequency closer to the specified value . Most trimming methods involve removing material from the resonator which will increase the resonance frequency . The target frequency for a coarse trimming stage consequently needs to be set below the final frequency since the tolerances of the process could otherwise result in a frequency higher than the following fine trimming stage could adjust for .

The coarsest method of trimming is grinding of the main resonating surface of the resonator ; this process has an accuracy of around  $\pm 800$  ppm . Better control can be achieved by grinding the edge of the resonator instead of the main surface . This has a less dramatic effect and consequently better accuracy . Processes that can be used for fine trimming , in order of increasing accuracy , are sandblasting , drilling , and laser ablation . Laser trimming is capable of achieving an accuracy of  $\pm 40$  ppm .

Trimming by hand , rather than machine , was used on some early production components but would now normally only be encountered during product development . Methods available include sanding and filing . It is also possible to add material to the resonator by hand , thus reducing the resonance frequency . One such method is to add solder , but this is not suitable for production use since the solder will tend to reduce the high Q of the resonator .

In the case of MEMS filters , it is not possible to trim the resonators outside of the filter because of the integrated nature of the device construction . However , trimming is still a requirement in many MEMS applications . Laser ablation can be used for this but material deposition methods are available as well as material removal . These methods include laser or ion @- @ beam induced deposition .