

Numerical Methods

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Teil I Computing with Matrices

1 Computational Effort

We use *Asymptotic Computation Complexity* (O-bounds) to judge our algorithms. But note that in many cases this approach is imprecise because on today's computers, *memory* bandwidth and latency is a key bottleneck.

Lem. (Matrix Product) The matrix product of an $m \times n$ and $n \times k$ matrix has complexity $O(mnk)$

Tricks to reduce complexity:

- Exploit associativity of operations
- Exploit hidden summations

2 Machine Arithmetic

Machine Numbers have a finite precision, hence any involved computation suffers from **Roundoff**.

Def. (Cancellation) Subtraction of almost equal numbers and accordingly an extreme *amplification of relative errors*.

Tricks to avoid cancellation:

- Trigonometric Identities
- Case Distinction
- Taylor Approximations
- Computing Diff. Quot. through Approx.

Def. (Stability) An Algorithm F is **numerically stable** if for all $x \in X$ its result $F(x)$ (possibly affected by roundoff) is the exact result for 'slightly perturbed' data.

Teil II Linear Systems

1 Theory

Def. The Condition Number of a matrix is

$$\text{cond}(\mathbf{A}) := \|\mathbf{A}\| \|\mathbf{A}^{-1}\|$$

and is a measure for stability (≈ 1 for stability). Note that it is infinite for singular matrices. Intuitively $\text{cond}(\mathbf{A})$ catches the maximum stretching factor divided by the minimum stretching factor.

Lem. If $m = n$ and \mathbf{A} is regular/invertible, then its 2-norm condition number is $\text{cond}_2(\mathbf{A}) = \sigma_1/\sigma_n$

2 LU Decomposition

There are two methods to choose pivots in the elimination process with a Permutation Matrix:

- Partial Pivoting: Max of column
- Full Pivoting: Max of remaining matrix

Complexities:

- Gauss elim. $\mathcal{O}(n^3)$, back substitution $\mathcal{O}(n^2)$
- LU: decomp. $\mathcal{O}(n^3)$, solve $\mathcal{O}(n^2)$
- Inverse: compute $\mathcal{O}(n^3)$, solve $\mathcal{O}(n^2)$

3 Exploiting Structure when Solving Linear Systems

Method: (Low Rank Modification) Use the *Sherman-Morrison-Woodbury* Formula to compute the inverse of slightly modified (rank-1-modification) matrix.

$$(\mathbf{A} + \mathbf{U}\mathbf{V}^H)^{-1} = \mathbf{A}^{-1} - \mathbf{A}^{-1}\mathbf{U}(\mathbf{I} + \mathbf{V}^H\mathbf{A}^{-1}\mathbf{U})^{-1}\mathbf{V}^H\mathbf{A}^{-1}$$

4 Matrix Storage Formats

Dense Matrices are either stored in Column-Major Ordering or Row-Major Ordering. By default Eigen employs Column-Major Ordering.

Sparse Matrices can be stored in

- COO/Triplet: Vector of triplets of (row, col, value)
- CCS: Compressed Column Storage
- CRS: Compressed Row Storage

where CRS consists of the arrays:

- val: All non-zero values in Row-Major Order

- colIdx: Column index of each entry in val
- row_ptr: Pointers to first elements of row

Note that $\text{size}(\text{row_ptr}) = m + 1$ (Sentinel).

Teil III Linear Least Squares

Critical comparison of Methods:

- Normal Equations: $\mathcal{O}(n^2m + n^3)$
→ blows up instability
- Extended Normal Equations:
→ same conditioning as \mathbf{A}
→ sparsity preserved
- Householder QR: $\mathcal{O}(n^2m)$
→ always stable
→ loss of sparsity
- SVD:
→ no full rank requirement for \mathbf{A}

1 Normal Equation Methods

$$\mathbf{A}^T \mathbf{A} \mathbf{x} = \mathbf{A}^T \mathbf{b}$$

Warning: Normal Equations are vulnerable to instability since $\text{cond}_2(\mathbf{A}^T \mathbf{A}) = \text{cond}_2(\mathbf{A})^2$

2 Singular Value Decomposition

$$\mathbf{A} = \mathbf{U}\mathbf{\Sigma}\mathbf{V}^H = \sum_{i=1}^r \sigma_i \mathbf{u}_i \mathbf{v}_i^H, \quad \mathbf{A} \in \mathbb{K}^{m \times n},$$

$p := \min\{m, n\}$, $r := \text{rank}(\mathbf{A})$,
 $\mathbf{\Sigma} = \text{diag}(\sigma_1, \dots, \sigma_p)$ $\sigma_1 > \sigma_2 > \dots > \sigma_r > \sigma_{r+1} = \dots = \sigma_p = 0$

Full:

- $\mathbf{U} \in \mathbb{K}^{m \times m}$ $[\mathcal{R}(\mathbf{A}) | \mathcal{N}(\mathbf{A})]$ (unitary)
- $\mathbf{\Sigma} \in \mathbb{K}^{m \times n}$ (generalized diagonal)
- $\mathbf{V} \in \mathbb{K}^{n \times n}$ $[\mathcal{R}(\mathbf{A}^H) | \mathcal{N}(\mathbf{A}^H)]$ (unitary)

Economical:

- $\mathbf{U} \in \mathbb{K}^{m \times p}$ $[\mathcal{R}(\mathbf{A})]$ (orthogonal columns)
- $\mathbf{\Sigma} \in \mathbb{K}^{p \times p}$ (diagonal)
- $\mathbf{V} \in \mathbb{K}^{n \times p}$ $[\mathcal{R}(\mathbf{A}^H)]$ (orthogonal columns)

Numerical Rank $r := \max_{j \in \{1, \dots, p\}} \left(\frac{\sigma_j}{\sigma_1} \geq \text{TOL} \right)$

Cost of Eco SVD $\mathcal{O}(\min\{m, n\}^2 \max\{m, n\})$

Lem. (LSQ by SVD)

$$\mathbf{x}^* = \arg \min_{\mathbf{x} \in \mathbb{R}^n} \|\mathbf{A}\mathbf{x} - \mathbf{b}\|_2 = \mathbf{V}_1 \mathbf{\Sigma}_r^{-1} \mathbf{U}_1^T \mathbf{b}$$

If the lsq has multiple solutions, \mathbf{x}^* has minimal norm.

Thm. (Low Rank Approximation) Let \mathbf{A}_k be the matrix resulting from only keeping the first k singular values of \mathbf{A} . \mathbf{A}_k is the best rank- k approximation of \mathbf{A} .

Method: (Principal Component Analysis)

Given a dataset in the form of a matrix $\mathbf{A} = \mathbf{U}\mathbf{\Sigma}\mathbf{V}^H$, such that every column represents a time series.

- Columns of \mathbf{U} capture dominant trends.
- Columns of \mathbf{V} capture how strong the trends are in a particular column of \mathbf{A}
- Diagonal Entries of $\mathbf{\Sigma}$ tell us how dominant the trend is overall.

Method: (Proper orthogonal decomp.)

Closely related is the POD problem, where we seek k -dimensional subspace U_k of \mathbb{R}^m , for which the sum of squared distances of the data points $\mathbf{A} = [\mathbf{a}_1, \dots, \mathbf{a}_n]$ to U_k is minimal.

Using the *Low Rank Approximation* Theorem we can prove that

$$U_k = \mathcal{R}(\mathbf{U}_{:,1:k})$$

3 QR Decomposition

$$\mathbf{A} = \mathbf{Q}\mathbf{R}, \quad \mathbf{A} \in \mathbb{K}^{m \times n} \quad \mathbf{Q} \in \mathbb{K}^{m \times m}$$

Idea: LLQ Problems are easier to solve for triangular matrices, hence we manage to decompose \mathbf{A} into $\mathbf{Q}\mathbf{R}$ we have such a System since orthogonal matrices are norm-invariant.

Method: (Householder Reflections) The following Householder matrix performs a reflection

$$\mathbf{H}(\mathbf{v}) := \mathbf{I} - 2\mathbf{v}\mathbf{v}^T$$

at the hyperplane with normal unit vector \mathbf{v} (Intuitively $\mathbf{H}(\mathbf{v})$ subtracts the projection of \mathbf{x} on \mathbf{v} twice).

We can reduce a matrix \mathbf{A} to \mathbf{R} using n successive transformations

$$\mathbf{H}(\mathbf{v}_n) \dots \mathbf{H}(\mathbf{v}_1) \mathbf{A} = \mathbf{R}$$

, where $\mathbf{H}(\mathbf{v}_i)$ reflects the lower part of the i 'th column of the current \mathbf{A} on \mathbf{e}_i . E.g

$$\mathbf{v}_1 = \mathbf{a}_1 \pm \|\mathbf{a}_1\| \mathbf{e}_1$$

Method: (Givens Rotations) We can selectively eliminate entries with Givens rotations. The following matrix rotates everything in the hyperplane defined by $\mathbf{e}_1, \mathbf{e}_k$:

$$\mathbf{G}(1, k, \theta) = \begin{bmatrix} c & \cdots & s & \cdots & 0 \\ \vdots & \ddots & \vdots & \ddots & \vdots \\ -s & \cdots & c & \cdots & 0 \\ \vdots & & \vdots & \ddots & \vdots \\ 0 & \cdots & 0 & \cdots & 1 \end{bmatrix} \begin{bmatrix} a_1 \\ \vdots \\ a_k \\ \vdots \\ a_n \end{bmatrix}$$

Note that $c = \cos(\theta)$, $s = \sin(\theta)$. Two eliminate a_k we solve the equations.

$$c^2 + s^2 = 1 \quad \text{and} \quad -sa_1 + ca_k = 0$$

As with householder reflections, there are always two possibilities, one of whom is cancellation free.

Lem. (Modifications Techniques) Computing the QR decomposition of a slightly modified matrix (rank-1-modification, adding a row, adding a column) can be done efficiently in $\mathcal{O}(mn + n^2)$.

4 Constrained Least Squares

Problem: Find $x \in \mathbb{R}^n$ such that

$$\|\mathbf{Ax} - \mathbf{b}\| \rightarrow \min \quad \text{and} \quad \mathbf{Cx} = \mathbf{d}.$$

Method: (Lagrangian Multipliers) We introduce the multiplier $\mathbf{m} \in \mathbb{R}^p$ to solve

$$\mathbf{x}^* = \arg \min_{\mathbf{x} \in \mathbb{R}^n} \sup_{\mathbf{m} \in \mathbb{R}^p} \|\mathbf{Ax} - \mathbf{b}\|_2 + \mathbf{m}^T (\mathbf{Cx} - \mathbf{d})$$

and we notice that for any finite solution $(\mathbf{Cx} - \mathbf{d}) = 0$. By realising that that for the solution all partial derivatives must be zero we can obtain the *augmented normal equations*.

Method: (By SVD) We have

$$\mathbf{x} \in \mathbf{x}_0 + \mathcal{N}(\mathbf{C}) \quad \mathbf{x}_0 = \text{particular solution}$$

Hence since the SVD gives a basis of $\mathcal{N}(\mathbf{C})$ we write

$$\mathbf{x} = \mathbf{x}_0 + \mathbf{V}_2 \mathbf{y}$$

for some \mathbf{y} which leads to the standard lsq:

$$\|\mathbf{AV}_2 \mathbf{y} - (\mathbf{b} - \mathbf{Ax}_0)\| \rightarrow \min$$

Teil IV

Filtering Algorithms

1 Filters & Convolution

Def. (Filter) A function $F : l^\infty(\mathbb{Z}) \rightarrow l^\infty(\mathbb{Z})$ where $l^\infty(\mathbb{Z})$ is the space of bounded infinite sequences

$$l^\infty(\mathbb{Z}) = \left\{ (x_j)_{j \in \mathbb{Z}} : \sup |x_j| < \infty \right\}$$

Def. (LT-FIR) A Filter that is:

- Linear
- Time-Invariant: Shifting the input in time leads to the same output shifted in time.
- Finite: $\exists M \forall j : x_j = 0 \text{ if } |j| > M \implies \exists N \forall j : y_j = 0 \text{ if } |j| > N$
- Causal: The output does not start before the input.

Such a filter is uniquely characterized by its **Impulse response**:

$$F(\delta_{j,0}) = \dots, 0, h_0, h_1, \dots, h_{n-1}, 0, \dots$$

For inputs $\mathbf{x} \in \mathbb{R}^m$ we get outputs $\mathbf{y} \in \mathbb{R}^{m+n-1}$

$$\begin{bmatrix} y_0 \\ y_1 \\ y_2 \\ y_3 \\ y_4 \\ y_5 \\ y_6 \end{bmatrix} = \begin{bmatrix} h_0 & 0 & 0 & 0 \\ h_1 & h_0 & 0 & 0 \\ h_2 & h_1 & h_0 & 0 \\ h_3 & h_2 & h_1 & h_0 \\ 0 & h_3 & h_2 & h_1 \\ 0 & 0 & h_3 & h_2 \\ 0 & 0 & 0 & h_3 \end{bmatrix} \begin{bmatrix} x_0 \\ x_1 \\ x_2 \\ x_4 \end{bmatrix}$$

Def. (Discrete Convolution) A Filter where given $\mathbf{x} = [x_0, \dots, x_{m-1}]^T \in \mathbb{K}^m$, $\mathbf{h} = [h_0, \dots, h_{n-1}]^T \in \mathbb{K}^n$ their DCONV is the vector $\mathbf{y} \in \mathbb{K}^{m+n-1}$ with components.

$$y_k = \sum_{j=0}^{m-1} h_{k-j} x_j$$

Def. (Periodic Convolution) Given two n -periodic signals $(u_k)_{k \in \mathbb{Z}}, (x_k)_{k \in \mathbb{Z}}$ PCONV yields the n -periodic signal:

$$(y_k) := (u_k) *_{\mathbf{n}} (x_k), y_k := \sum_{j=0}^{n-1} u_{k-j} x_j$$

PCONV can be represented by a matrix

$$\mathbf{C} = [c_{ij}]_{i,j=1}^n, c_{ij} = u_{j-i}$$

. Such a matrix is called **circulant**.

2 Discrete Fourier Transform

The Fourier Matrix for inputs $\mathbf{y} \in \mathbb{C}^n$ is given by

$$\mathbf{F}_n = \left[\omega_n^{lj} \right]_{l,j=0}^{n-1} \in \mathbb{C}^{n,n} \quad \omega_n = \exp \left(\frac{-2\pi i}{n} \right)$$

and $\text{DFT}_n(\mathbf{y}) := \mathbf{F}_n \mathbf{y} = \mathbf{c}$. Properties include

$$\mathbf{F}_n^{-1} = \frac{1}{n} \mathbf{F}_n^H = \frac{1}{n} \bar{\mathbf{F}}_n$$

Method: (Frequency Filtering) The columns of \mathbf{F}_n are trigonometric basis vectors, where

$$\mathbf{v}_k = \left[\cos \left(j \frac{2\pi}{n} \cdot k \right) + i \sin \left(\frac{2\pi j k}{n} \right) \right]_{j=0}^{n-1}$$

k is the 'frequency' and $j2\pi/n$ is the sample point. Hence the DFT_n is a basis transformation

$$B_E \leftrightarrow B_{\text{trig}}$$

Denosing and low/high filters can be implemented by manipulating the signal in the frequency domain.

Lem. (Diagonalizing circulant matrices)

For any circulant matrix $\mathbf{C} \in \mathbb{C}^{n,n}$ we have

$$\mathbf{C} = \mathbf{F}_n^{-1} \text{diag}(d_1, \dots, d_n) \mathbf{F}_n,$$

$$[d_0, \dots, d_{n-1}]^T = \mathbf{F}_n [u_0, \dots, u_{n-1}]^T$$

In other words, the columns of \mathbf{F}_n are eigenvectors of \mathbf{C} .

Thm. (Convolution Theorem) Periodic convolution in the time-domain equals to multiplication in the frequency-domain.

$$(\mathbf{u}) *_{\mathbf{n}} (\mathbf{x}) = \mathbf{F}_n^{-1} \left[(\mathbf{F}_n \mathbf{u})_j (\mathbf{F}_n \mathbf{x})_j \right]_{j=1}^n$$

Method: (2D DFT)

3 Fast Fourier Transform

Method: (Toeplitz Matrix Technique)