Control Systems

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1 Feedback Circuits

Abstract—The objective of this manual is to introduce control system design at an elementary level.

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1 FEEDBACK CIRCUITS

1.0.1. Consider a Feedback Current Amplifier formed by cascading an Inverting Opamp μ with a MOSFET (NMOS). The output current is the Drain Current of the NMOS. Assume that Opamp has an input resistance R_{id} , an Open Circuit Voltage Gain μ , and an output resistance r_{o1}

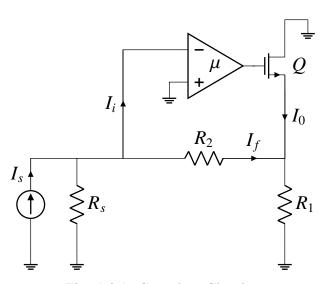


Fig. 1.0.1: Complete Circuit

1.0.2. Identify the type of Feedback Circuit and draw its corresponding Block Diagram Representation

> Solution: The Feedback Circuit is a Shunt-Series Feedback Current Amplifier.

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It is represented in a Feedback Block Diagram as follows

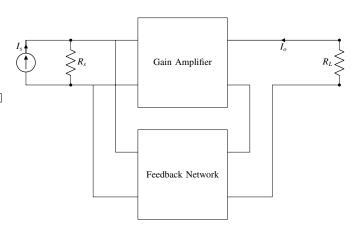


Fig. 1.0.2: Shunt Series Amplifier Block Diagram

1.0.3. Represent the Control System using a block diagram

Solution:

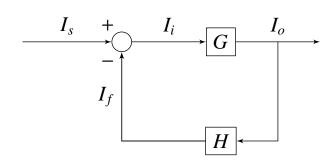


Fig. 1.0.3: Block Diagram

1.0.4. Represent the given circuit using a Small Signal Equivalent Model.

> Solution: To draw Small Signal Equivalent model,

- Replace MOSFET with Current Source and Resistance in parallel with it.
- Replace Opamp with Voltage Source of the input Voltage multiplied by Gain
- 1.0.5. Describe the resistances involved in the circuit **Solution:**
- with the corresponding Circuit.

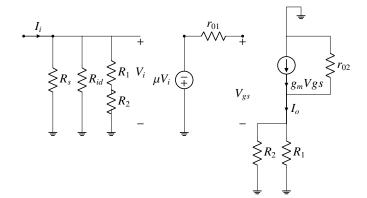


Fig. 1.0.4: Small Signal Model

Resistance	Description
R_{in}	Total Input Resistance
R_{out}	Total Ouput Resistance
R_{id}	Input resistance of Opamp
r_{o1}	Output resistance of Opamp
r_{o2}	Output resistance of MOSFET
R_i	Input resistance of Open Loop
R_o	Output resistance of Open Loop
R_{if}	Input resistance of Feedback
R_{of}	Output resistance of Feedback
R_s	Resistance of Current Source

TABLE 1.0.5

Solution:

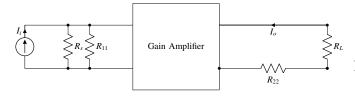


Fig. 1.0.6: Gain Block

1.0.7. Calculate R_{11} and use it to find the Gain G. **Solution:** R_{11} is the resistance in parallel with the Resistance of the Source and the Input Impedance of the Opamp

$$R_{11} = R_1 + R_2 \tag{1.0.7.1}$$

$$R_i = R_s ||R_{id}|| (R_1 + R_2)$$
 (1.0.7.2)

$$V_i = I_i R_i \tag{1.0.7.3}$$

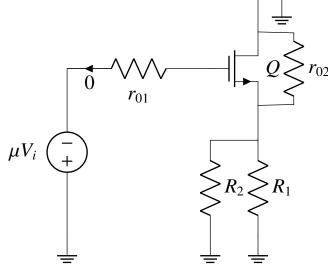


Fig. 1.0.6: Gain Circuit

$$I_o = -\mu V_i \frac{1}{1/g_m + (R_1 || R_2 || r_{o2})} \frac{r_{o2}}{r_{o2} + (R_1 || R_2)}$$
(1.0.7.4)

$$G = \frac{I_o}{I_i} = -\mu \frac{R_i}{1/g_m + (R_1||R_2||r_{o2})} \frac{r_{o2}}{r_{o2} + (R_1||R_2)}$$
(1.0.7.5)

We use the approximation

$$1/g_m \ll (R_1 || R_2 || r_{o2}) \tag{1.0.7.6}$$

This is because the $\frac{1}{g_m}$ is in order of few Ω s but, R_1 , R_2 and r_{o2} are in order of $k\Omega$ s

$$G = -\mu \frac{R_i}{R_1 || R_2} \tag{1.0.7.7}$$

1.0.8. Draw the Block Diagram for the Feedback Block, with the corresponding Circuit. **Solution:**



Fig. 1.0.8: Feedback Block

(1.0.7.2) 1.0.9. Calculate R_{22} and use it to find the Feedback Gain H. **Solution:** From Feedback Circuit, R_{22} is resistance obtained by shorting Feedback Network

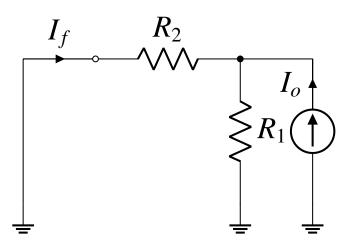


Fig. 1.0.8: Feedback Circuit

$$\implies \frac{1}{R_{if}} = \frac{1}{R_i} + \frac{\mu}{R_2} \tag{1.0.12.2}$$

$$\implies R_{if} = R_i \| \frac{R_2}{\mu} \tag{1.0.12.3}$$

Substituting the value of R_i ,

$$R_{if} = R_s ||R_{id}|| (R_1 + R_2) || \frac{R_2}{\mu}$$
 (1.0.12.4)

$$R_{if} = R_s || R_{in} \tag{1.0.12.5}$$

$$\implies R_{in} = R_{id} ||(R_1 + R_2)|| \frac{R_2}{\mu}$$
 (1.0.12.6)

$$R_{in} \simeq \frac{R_2}{\mu} \tag{1.0.12.7}$$

$$R_{22} = R_1 || R_2$$
 (1.0.9.1)^{1.0.13}. Give expressions for R_o , R_{of} , R_{out} Solution:

 I_f is the current in R_2

$$\implies H = \frac{I_f}{I_o} = -\frac{R_1}{R_1 + R_2}$$
 (1.0.9.2)

1.0.10. Calculate Loop Gain GH

Solution:

$$GH = \mu \frac{R_i}{\frac{1}{g_m} + (R_1 || R_2 || r_{o2})} \frac{r_{o2}}{r_{o2} + (R_1 || R_2)} \frac{R_1}{R_1 + R_2}$$
(1.0.10.1)

$$\implies GH \simeq \mu \frac{R_i}{R_1 || R_2} \frac{R_1}{R_1 + R_2} = \mu \frac{R_i}{R_2}$$
(1.0.10.2)

$$R_o = r_{o2} + (R_1||R_2) + (g_m r_{o2})(R_1||R_2)$$
(1.0.13.1)

$$\implies R_o \simeq g_m r_{o2} (R_1 || R_2)$$
 (1.0.13.2)

$$R_{of} = R_o(1 + GH) \simeq GHR_o$$
 (1.0.13.3)

$$R_{of} \simeq \mu(\frac{R_i}{R_2})(g_m r_{o2})(R_1 || R_2)$$
 (1.0.13.4)

$$R_{out} = R_{of} = \mu \frac{R_i}{R_1 + R_2} (g_m r_{o2}) R_1$$
 (1.0.13.5)

1.0.11. If loop gain is large, find approximate expres-1.0.14. Given the following values sion for closed loop gain T

Solution: Given,

$$GH \gg 1 \tag{1.0.11.1}$$

$$T = \frac{G}{1 + GH} \simeq \frac{1}{H} \tag{1.0.11.2}$$

$$T \simeq \frac{1}{H} = -\left(1 + \frac{R_2}{R_1}\right)$$
 (1.0.11.3)

1.0.12. Give expressions for R_{if} , R_{in}

Parameter	Value
μ	1000
R_s	∞
R_{id}	∞
r_{o1}	$1k\Omega$
R_1	$10k\Omega$
R_2	$90k\Omega$
g_m	5mA/V
r_{o2}	$20k\Omega$

TABLE 1.0.14

$$R_{if} = R_i/(1 + GH)$$
 (1.0.12.1)

Find numerical value of R_i and use it to find

the value of G

Solution: Using the given numerical values on the previously obtained equations, we obtain:

$$R_i = \infty ||\infty|| (10 + 90) = 100k\Omega \qquad (1.0.14.1)$$

$$G = -1000 \frac{100}{10||90} = -11.11 \times 10^{3} \quad (1.0.14.2)$$

1.0.15. Check the validity of the approximation that we use to neglect $1/g_m$

Solution:

$$1/g_m = 0.2k\Omega \ll (10||90||20)k\Omega = 6.2k\Omega$$
(1.0.15.1)

Hence, we can see that our approximation is valid

1.0.16. Find the value of feedback gain H and open loop gain GH

Solution:

$$H = -\frac{R_1}{R_1 + R_2} = -\frac{10}{10 + 90} = -0.1 \ (1.0.16.1)$$

$$GH = 1111 \gg 1$$
 (1.0.16.2)

1.0.17. Find the approximate value of closed loop gain T

Solution:

$$T \simeq \frac{1}{H} = -\frac{1}{0.1} = -10$$
 (1.0.17.1)

1.0.18. Find the values of R_{in} and R_{out}

Solution:

$$R_{in} = \frac{R_2}{\mu} = \frac{90k\Omega}{1000} = 90\Omega \tag{1.0.18.1}$$

$$R_o = g_m r_{o2}(R_1 || R_2) = 5 \times 20(10 || 90) = 900k\Omega$$
(1.0.18.2)

$$R_{out} = (1 + GH)R_o = 1112 \times 900 \simeq 1000M\Omega$$
(1.0.18.3)

1.0.19. Verify the above calculations using a Python code.

Solution:

codes/ee18btech11021/ee18btech11021_calc.

Parameter	Value
R_i	$100k\Omega$
$1/g_m$	200Ω
G	-1.11×10^4
Н	-0.1
GH	1111
T	-10
R_{in}	90Ω
R_o	$900k\Omega$
Rout	$1000M\Omega$

TABLE 1.0.18