

Solutions

$$(1) (a) H(z) = \frac{2}{1 - e^{-0.2} z^{-1}} - \frac{1}{1 - e^{-0.4} z^{-1}}$$

$$T_d = 2,$$

$$h[n] = 2h_c[2n], \quad h_c(t) \text{ is real}$$

Poles of the discrete time system are at $z = e^{-0.2}, e^{-0.4}$

$$\text{So, } s_1 T_d = -0.2 \quad s_2 T_d = -0.4$$

$$\Rightarrow s_1 = -0.1 \quad \Rightarrow s_2 = -0.2$$

$$\text{And } T_d A_1 = 2, \quad T_d A_2 = -1$$

$$\Rightarrow A_1 = 1 \quad \Rightarrow A_2 = -0.5$$

The discrete-time system can be expressed as

$$H(z) = \sum_{k=1}^2 \frac{T_d A_k}{1 - e^{s_k T_d} z^{-1}}$$

The corresponding s-domain poles are at $s = -0.2, -0.1$. The continuous-time system

has a transfer function

$$H_c(s) = \sum_{k=1}^2 \frac{A_k}{s - s_k} = \frac{1}{s + 0.1} - \frac{0.5}{s + 0.2} = \frac{0.5(s + 0.3)}{(s + 0.1)(s + 0.2)}$$

(P2)

(b) Using Bilinear Transformation, we have

$$s = \frac{2}{T_d} \left(\frac{1-z^{-1}}{1+z^{-1}} \right) = \frac{1-z^{-1}}{1+z^{-1}} \quad (\because T_d = 2)$$

$$\Rightarrow s = \frac{z-1}{z+1} \Rightarrow z = \frac{1+s}{1-s}$$

$$\text{So, } H_c(s) = H(z) \Big|_{z=\frac{1+s}{1-s}} = \frac{2}{1 - e^{-0.2} \left(\frac{1-s}{1+s} \right)} - \frac{1}{1 - e^{-0.4} \left(\frac{1-s}{1+s} \right)}$$

$$= \frac{2(1+s)}{(1+s) - e^{-0.2}(1-s)} - \frac{(1+s)}{(1+s) - e^{-0.4}(1-s)}$$

$$= \frac{2(1+s)}{(1-e^{-0.2}) + s(1+e^{-0.2})} - \frac{(1+s)}{(1-e^{-0.4}) + s(1+e^{-0.4})}$$

$$= \left(\frac{2}{1+e^{-0.2}} \right) \frac{(1+s)}{s + \left(\frac{1-e^{-0.2}}{1+e^{-0.2}} \right)} - \left(\frac{1}{1+e^{-0.4}} \right) \frac{(1+s)}{s + \left(\frac{1-e^{-0.4}}{1+e^{-0.4}} \right)}$$

$$= \frac{1.1(1+s)}{(s+0.1)} - \frac{0.598(1+s)}{(s+0.197)}$$

$$= \frac{(1+s)(0.58+0.156s)}{(s+0.1)(s+0.197)}$$

(2)

$$(a) \quad H_c(s) = \frac{s+a}{(s+a)^2 + b^2} = \frac{s+a}{(s+a+bj)(s+a-bj)}$$

$$= \frac{1/2}{s+a+bj} + \frac{1/2}{s+a-bj}$$

The system has two poles at $-(a+bj)$, $-(a-bj)$

Taking inverse Laplace transform,

$$h_c(t) = \frac{1}{2} e^{-(a+bj)t} u(t) + \frac{1}{2} e^{-(a-bj)t} u(t)$$

$$\Rightarrow h_1[n] = T h_c(nT) = \frac{T}{2} \left[e^{-(a+bj)nT} + e^{-(a-bj)nT} \right] u[n]$$

Taking Z-transform, we get,

$$H_1(z) = \frac{T}{2} \left[\frac{1}{1 - e^{-(a+bj)T} z^{-1}} + \frac{1}{1 - e^{-(a-bj)T} z^{-1}} \right]$$

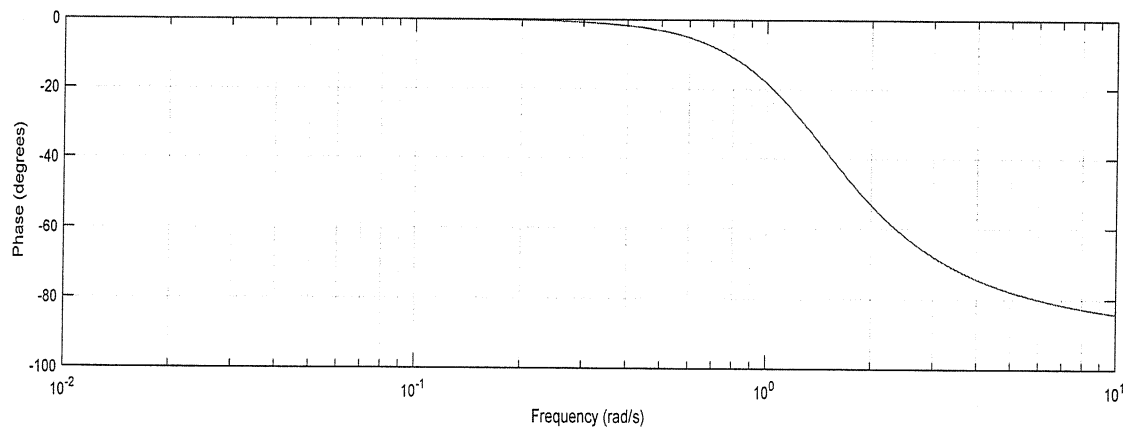
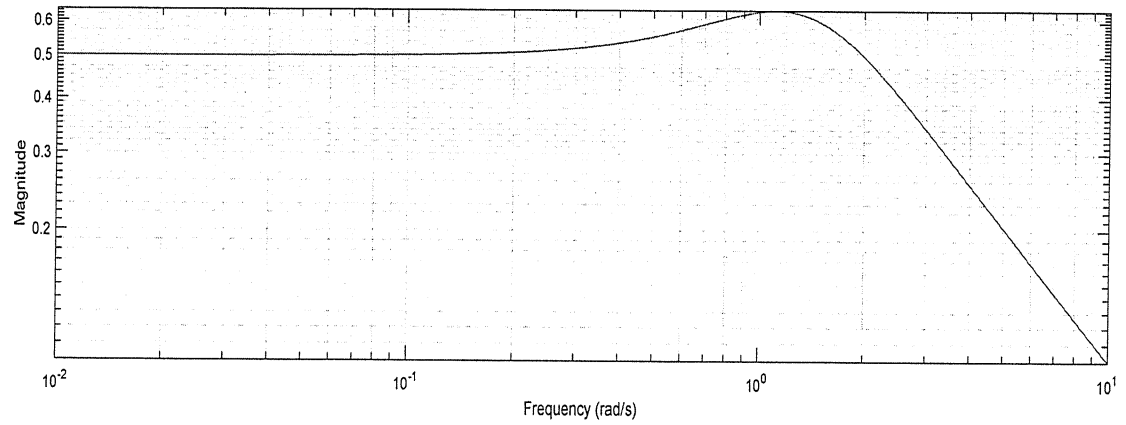
$$(b) \quad H_1(z) = \frac{T}{2} \left[\frac{2 - (e^{bjT} + e^{-bjT}) e^{-aT} z^{-1}}{1 - (e^{bjT} + e^{-bjT}) e^{-aT} z^{-1} + e^{-2aT} z^{-2}} \right]$$

$$= T \left[\frac{1 - \cos bT e^{-aT} z^{-1}}{1 - 2 \cos bT e^{-aT} z^{-1} + e^{-2aT} z^{-2}} \right]$$

Frequency response of the Continuous Time filter:

$$H_c(s) = \frac{s + a}{(s + a)^2 + b^2}$$

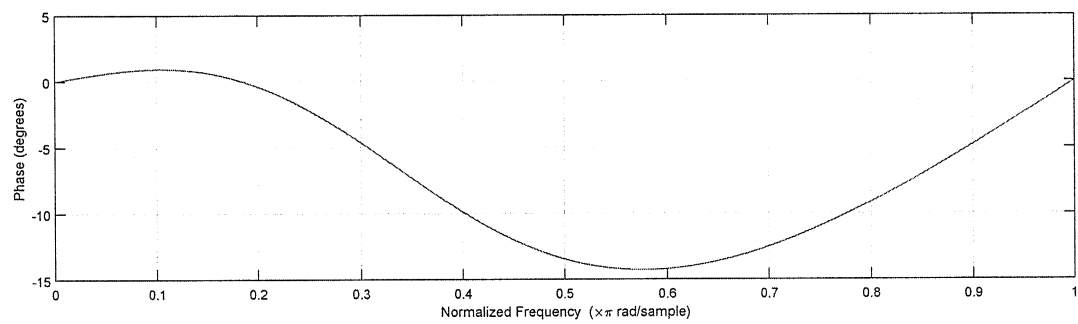
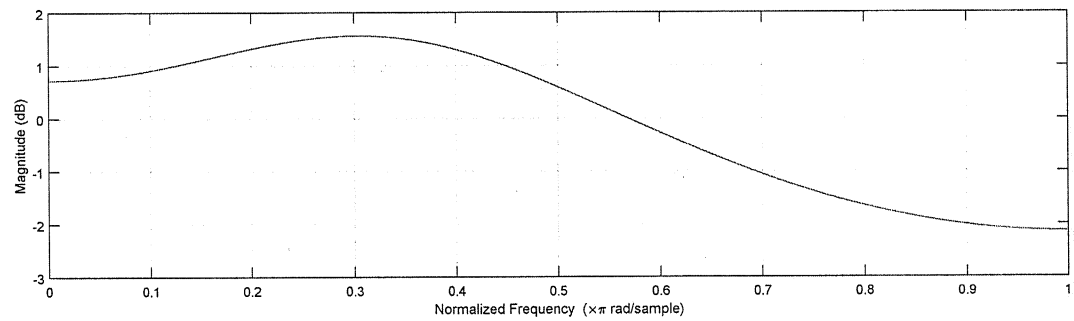
with $a=1$, $b=1$

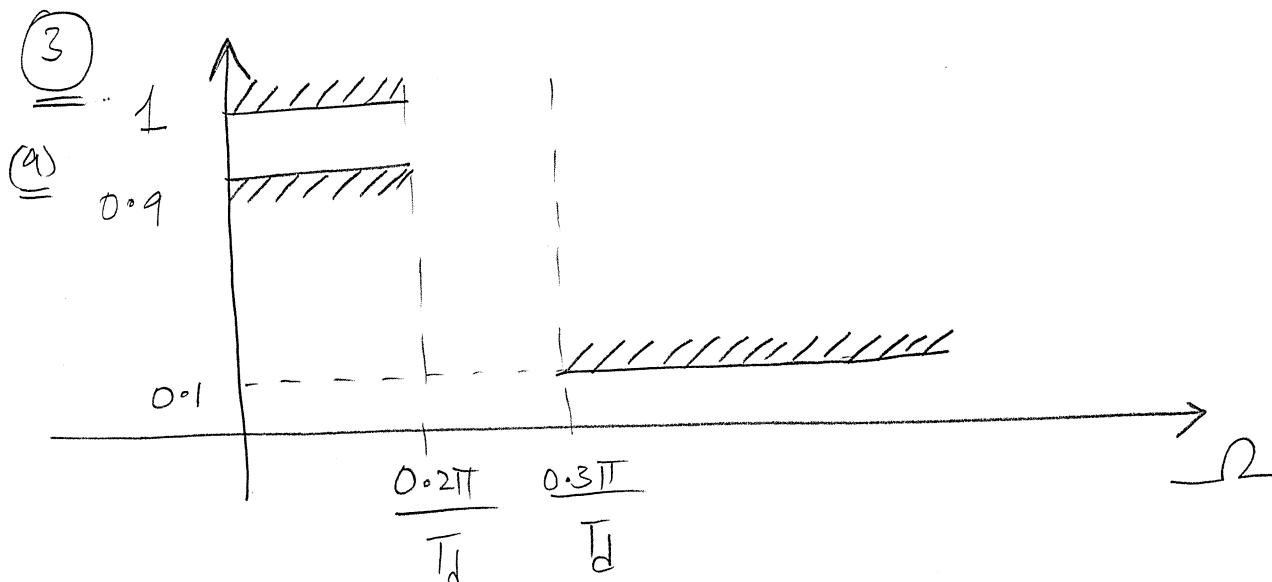


Frequency response of the Discrete Time filter:

With T=1

$$H(z) = \frac{1 - 0.198Z^{-1}}{1 - 0.397Z^{-1} + 0.135Z^{-2}}$$





(b) $\delta_p = 0.1$, $\delta_s = 0.1$, $\frac{\Omega_s}{\Omega_p} = \frac{\omega_s}{\omega_p} = \frac{0.3\pi}{0.2\pi} = 1.5$

$$N \geq \left[\log\left(\frac{1}{\delta_s^2} - 1\right) - \log\left(\left(\frac{1}{1-\delta_p}\right)^2 - 1\right) \right]$$

$$2 \log\left(\frac{\Omega_s}{\Omega_p}\right)$$

$N \geq 7.4546$, we should choose $N=8$

∴ we have N , we need to find Ω_c .

$$\Omega_p \left(\frac{1}{\left(\frac{1}{1-\delta_p}\right)^2 - 1} \right)^{\frac{1}{2N}} \leq \Omega_c \leq \Omega_s \left(\frac{1}{\frac{1}{\delta_s^2} - 1} \right)^{\frac{1}{2N}}$$

$$0.2\pi \left(\frac{1}{\left(\frac{10}{9}\right)^2 - 1} \right)^{\frac{1}{16}} \leq \Omega_c \leq 0.3\pi \left(\frac{1}{100 - 1} \right)^{\frac{1}{16}}$$

$$0.219\pi \leq \Omega_c \leq 0.225\pi$$

let us choose $\Omega_c = 0.7$

(C) So, $H_c(s) H_c(-s) = \left[\frac{1}{1 + \left(\frac{s}{j\Omega_c} \right)^{2N}} \right]$

With $N = 8$, $\Omega_c = 0.7$

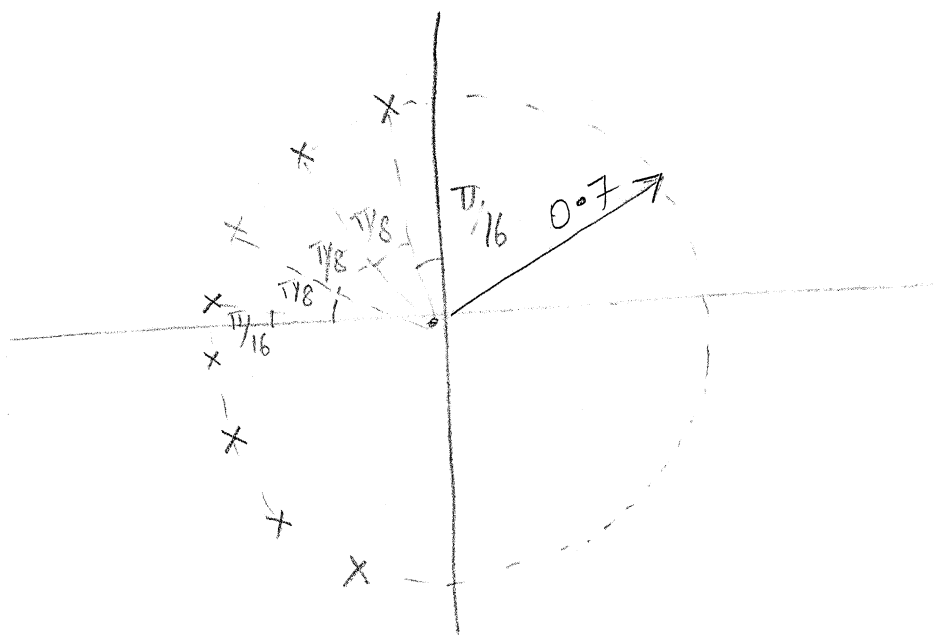
We would have 4 pole-pairs in the left half of the s -plane with the coordinates:

Pole pair 1: $0.7 e^{\pm j \frac{9\pi}{16}} = -0.1366 \pm j 0.6865$

Pole pair 2: $0.7 e^{\pm j \frac{11\pi}{16}} = -0.3889 \pm j 0.5820$

Pole pair 3: $0.7 e^{\pm j \frac{13\pi}{16}} = -0.582 \pm j 0.3889$

Pole pair 4: $0.7 e^{\pm j \frac{15\pi}{16}} = -0.6865 \pm j 0.1366$



```
%%HW-7 Problem 3%%
```

```
clear all;
```

```
clc;
```

```
[N,Wn] = buttord(0.2*pi,0.3*pi,0.915,20,'s');
```

```
[z,p,k]=butter(N,Wn,'s');
```

```
[b,a]=zp2tf(z,p,k);
```

```
p
```

```
Hs= tf(b,a)
```

```
[numd,dend]=impinvar(b,a,1);
```

```
Hs=tf(numd,dend,1)
```

```
p =
```

```
-0.6936 + 0.1380i
```

```
-0.6936 - 0.1380i
```

```
-0.5880 + 0.3929i
```

```
-0.5880 - 0.3929i
```

```
-0.3929 + 0.5880i
```

```
-0.3929 - 0.5880i
```

```
-0.1380 + 0.6936i
```

```
-0.1380 - 0.6936i
```

```
Hs =
```

```
0.06257
```

```
s^8 + 3.625 s^7 + 6.57 s^6 + 7.727 s^5 + 6.426 s^4 + 3.865 s^3 + 1.643 s^2 + 0.4535 s + 0.06257
```

Continuous-time transfer function.

Hz =

$$\frac{1.23e-13 z^8 + 7.798e-06 z^7 + 0.0005763 z^6 + 0.003549 z^5 + 0.004546 z^4 + 0.00144 z^3 + 9.449e-05 z^2 + 5.148e-07 z}{z^8 - 4.479 z^7 + 9.26 z^6 - 11.39 z^5 + 9.038 z^4 - 4.718 z^3 + 1.576 z^2 - 0.307 z + 0.02665}$$

Sample time: 1 seconds

Discrete-time transfer function.

```
%%HW-7 Problem-4 second way%%
clear all;
clc;

[N,Wn] = cheblord(0.2*pi,0.3*pi,1,15,'s')
[b,a]=cheby1(N,Wn,0.2*pi,'s')
[num,den]=bilinear(b,a,1)
tf(num,den,1)
```

N =

4

Wn =

0.6283

b =

0 0 0 0 0.0494

(4) n:

$$\frac{1}{1 + \epsilon^2 v_n^2 \left(\frac{\Omega_s}{\Omega_p} \right)} \leq \delta_s^2$$

Here $(1 - \delta_p) = 10^{-1/20} = 0.8913$
 $\delta_s = 10^{-15/20} = 0.1778$
 $\Omega_c = \Omega_p = 0.2\pi$

$$1 + \epsilon^2 v_n^2 \left(\frac{\Omega_s}{\Omega_p} \right) \geq \left(\frac{1}{\delta_s^2} \right)$$

$$\epsilon^2 v_n^2 \left(\frac{\Omega_s}{\Omega_p} \right) \geq \left(\frac{1}{\delta_s^2} - 1 \right)$$

$$v_n^2 \left(\frac{\Omega_s}{\Omega_p} \right) \geq \frac{1}{\epsilon^2} \left(\frac{1}{\delta_s^2} - 1 \right)$$

$$v_n \left(\frac{\Omega_s}{\Omega_p} \right) \geq \sqrt{\frac{\left(\frac{1}{\delta_s^2} - 1 \right)}{\epsilon^2}}$$

$$\epsilon = \sqrt{\frac{1}{(1 - \delta_p)^2} - 1} \quad \text{--- (1)}$$

$$\cosh \left(n \cosh^{-1} \left(\frac{\Omega_s}{\Omega_p} \right) \right) \geq \sqrt{\frac{\left(\frac{1}{\delta_s^2} - 1 \right)}{\epsilon^2}}$$

$$n \cosh^{-1} \left(\frac{\Omega_s}{\Omega_p} \right) \geq \cosh^{-1} \left(\sqrt{\frac{\frac{1}{\delta_s^2} - 1}{\epsilon^2}} \right)$$

$$n \geq \frac{\cosh^{-1} \left(\sqrt{\frac{\frac{1}{\delta_s^2} - 1}{\epsilon}} \right)}{\cosh^{-1} \left(\frac{\Omega_s}{\Omega_p} \right)} \quad \text{--- (2)}$$

Using (1), we estimate $\epsilon = \sqrt{\frac{1}{(1 - 0.89)^2} - 1} = 0.51$

$$\text{and } \Omega_c = \Omega_p = 0.2\pi$$

$$\Omega_s = 0.3\pi$$

$$\therefore n \geq \cosh^{-1} \left(\frac{\sqrt{\left(\frac{1}{0.1778}\right)^2 - 1}}{0.51} \right) = 3.2$$

$$\underline{\underline{\text{So, } n=4}}$$

a =

1.0000 0.7011 0.6406 0.2294 0.0531

num =

0.0020 0.0080 0.0120 0.0080 0.0020

den =

1.0000 -3.0016 3.6946 -2.1670 0.5084

ans =

$0.002 z^4 + 0.008002 z^3 + 0.012 z^2 + 0.008002 z + 0.002$

$z^4 - 3.002 z^3 + 3.695 z^2 - 2.167 z + 0.5084$

Sample time: 1 seconds

Discrete-time transfer function.