

# Solution

Worcester Polytechnic Institute — Department of Electrical and Computer Engineering  
ECE2311 — Continuous-Time Signal and System Analysis — Term B'17

## Homework 11: Due Tuesday, 12 December 2017 (3:00 P.M.)

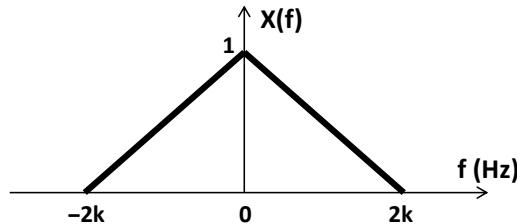
Write your name and ECE box at the top of each page.

### General Reminders on Homework Assignments:

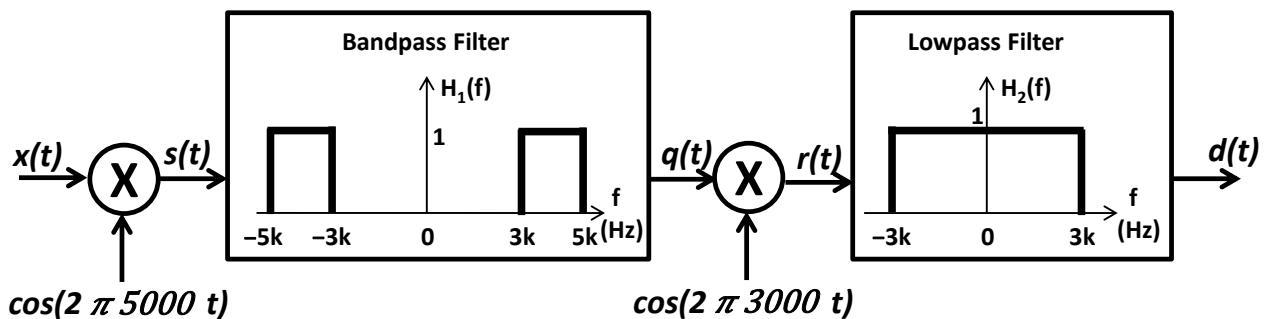
- Always complete the reading assignments *before* attempting the homework problems.
- Show all of your work. Use written English, where applicable, to provide a log or your steps in solving a problem. (For numerical homework problems, the writing can be brief.)
- A solution that requires physical units is *incorrect* unless the units are listed as part of the result.
- Get in the habit of underlining, circling or boxing your result.
- Always write neatly. Communication skills are essential in engineering and science. “If you didn’t write it, you didn’t do it!”

### 1) Sinusoidal Amplitude Modulation:

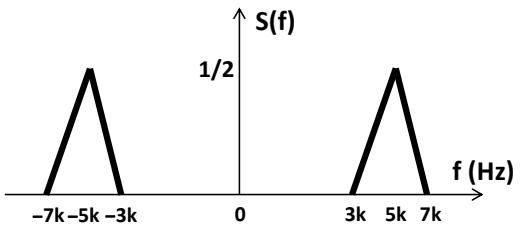
- a) Consider the signal  $x(t)$  with magnitude spectrum shown below:



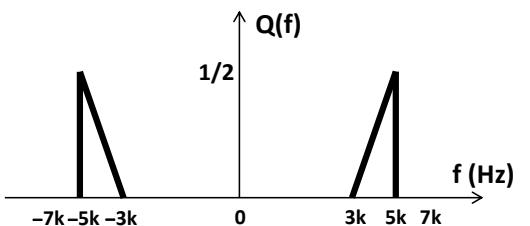
Let  $x(t)$  be processed through the system shown below to produce output signal  $d(t)$ . Draw the magnitude spectra of signals  $s(t)$ ,  $q(t)$ ,  $r(t)$  and  $d(t)$ .



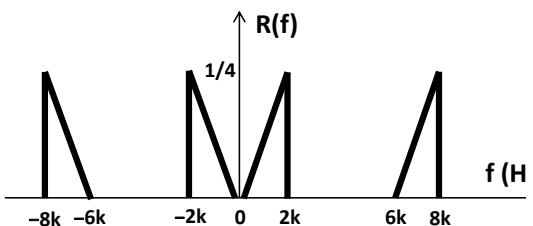
The first mixer creates two copies of the original magnitude spectrum, shifted by  $\pm 5000$  Hz and reduced in magnitude by half. The resulting  $S(f)$  is shown to the right.



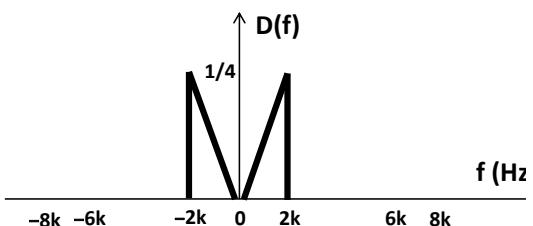
The bandpass filter removes the frequencies above 5k Hz in the modulated signal, producing  $Q(f)$  shown to the right.



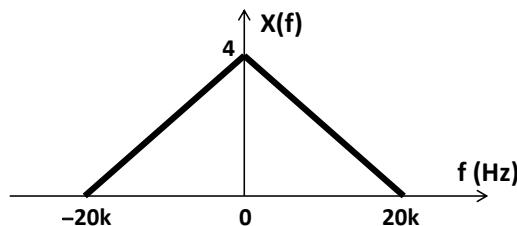
The second mixer shifts BOTH the positive and negative frequencies each by 3k Hz and reduces the magnitude in half. The resulting  $R(f)$  is shown to the right.



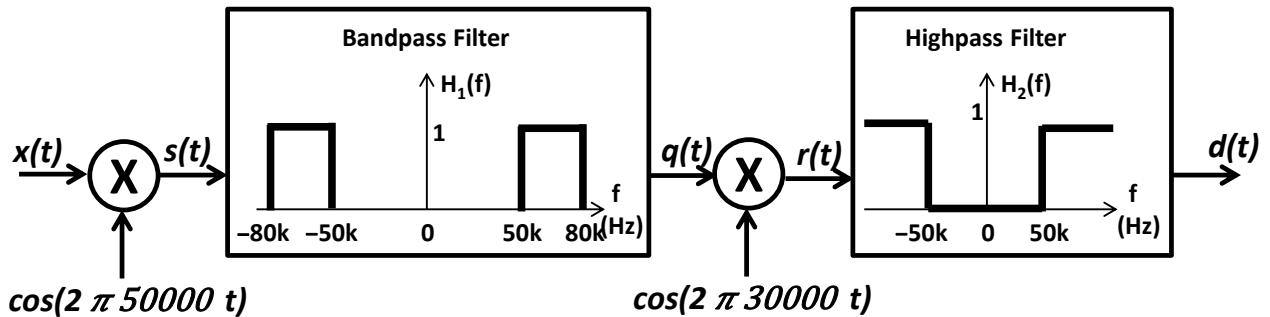
Finally, the lowpass filter eliminates the higher frequency portion, producing the final output  $D(f)$ .



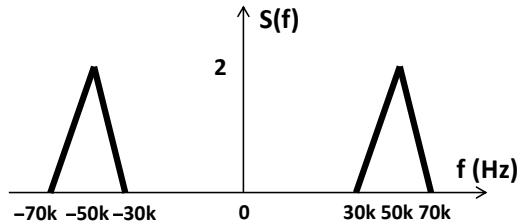
- b) Consider the signal  $x(t)$  with magnitude spectrum shown below:



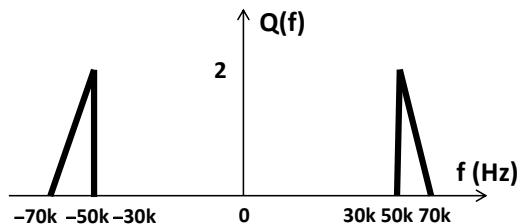
Let  $x(t)$  be processed through the system shown below to produce output signal  $d(t)$ . Draw the magnitude spectra of signals  $s(t)$ ,  $q(t)$ ,  $r(t)$  and  $d(t)$ .



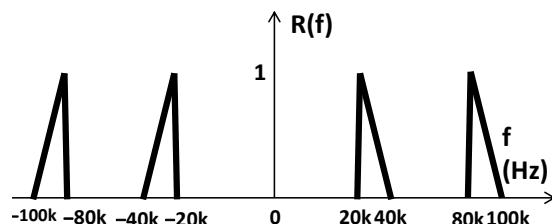
The first mixer creates two copies of the original magnitude spectrum, shifted by  $\pm 50$  Hz and reduced in magnitude by half. The resulting  $S(f)$  is shown to the right.



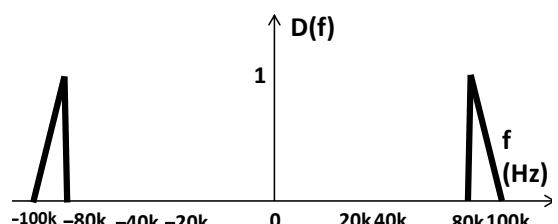
The bandpass filter removes the frequencies below 50k Hz in the modulated signal, producing  $Q(f)$  shown to the right.



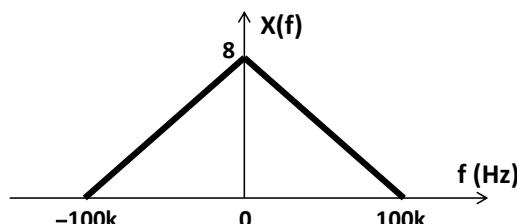
The second mixer shifts BOTH the positive and negative frequencies each by 30k Hz and reduces the power in half. The resulting  $R(f)$  is shown to the right.



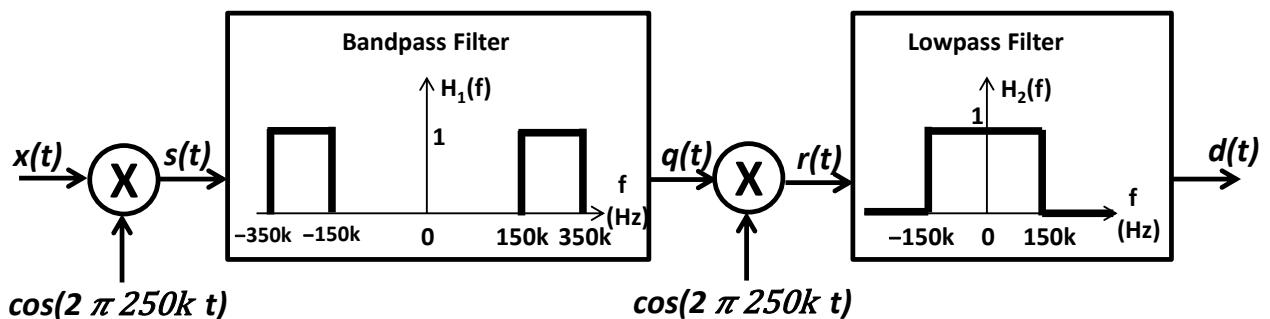
Finally, the highpass filter eliminates the lower frequency portion, producing the final output  $D(f)$ .



- c) Consider the signal  $x(t)$  with magnitude spectrum shown below:



Let  $x(t)$  be processed through the system shown below to produce output signal  $d(t)$ . Draw the magnitude spectra of signals  $s(t)$ ,  $q(t)$ ,  $r(t)$  and  $d(t)$ .

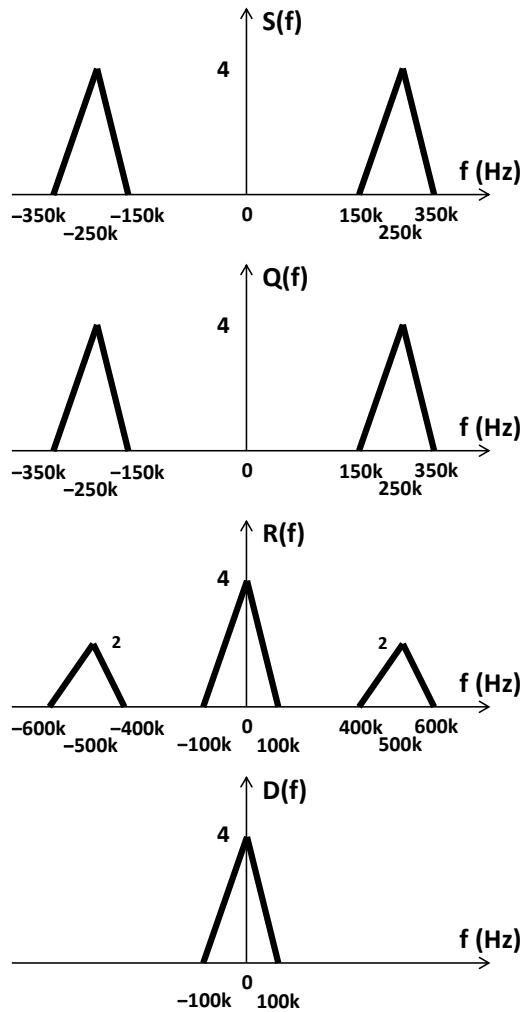


The first mixer creates two copies of the original magnitude spectrum, shifted by  $\pm 250$  kHz (thus, centered at  $\pm 250$  kHz) and reduced in magnitude by half. The resulting  $S(f)$  is shown to the right.

*The bandpass filter* extends exactly over the bandwidth of the input signal (and has a pass band gain of 1), thus does not alter the signal. The magnitude spectrum of  $Q(f)$  is shown to the right.

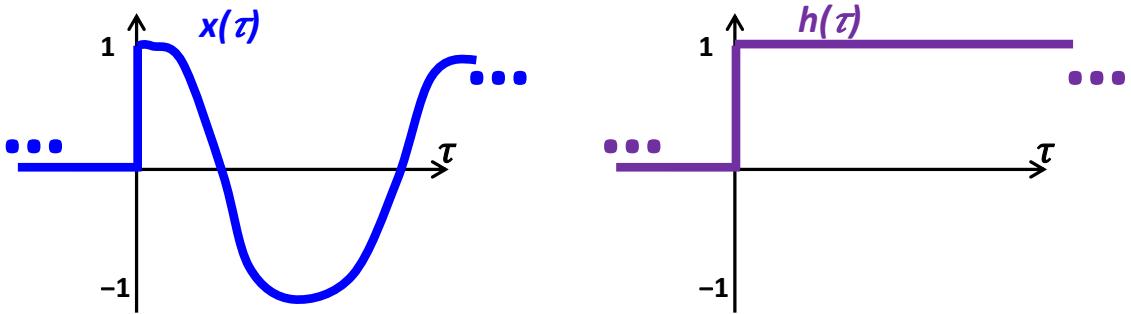
The second mixer shifts BOTH the positive and negative frequencies each by 250 kHz (so, each has one term re-centered at DC) and reduces the magnitude of each in half. Since the overlapping magnitudes at DC sum, the resulting  $R(f)$  is shown to the right.

Finally, the lowpass filter eliminates the upper frequency portion, producing the final output  $D(f)$ .

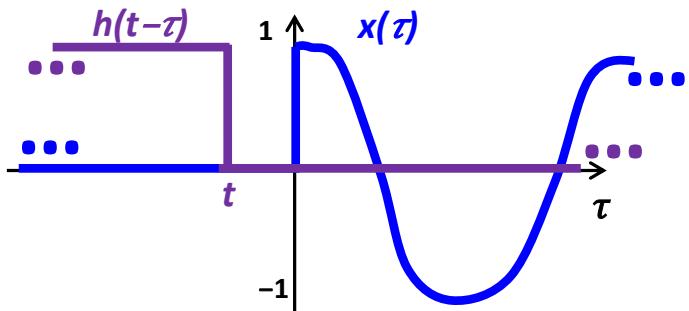


## 2) Convolution:

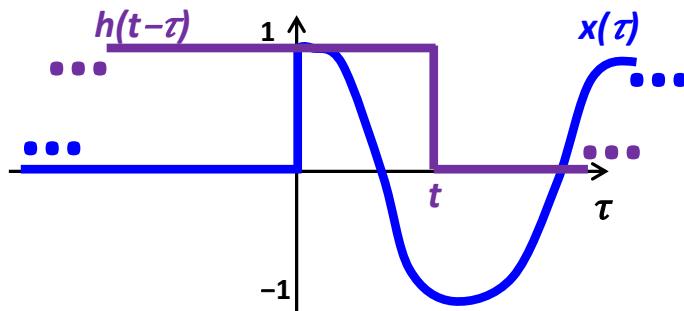
- Use direct integration of the convolution integral/graphical convolution to determine the convolution  $y(t) = x(t) * h(t)$  if  $x(t) = \cos(t) \mu(t)$  and  $h(t) = \mu(t)$ .



- For  $t < 0$ : No overlap. Convolution equals zero in this range.



- For  $t \geq 0$ : Partial overlap from the left.

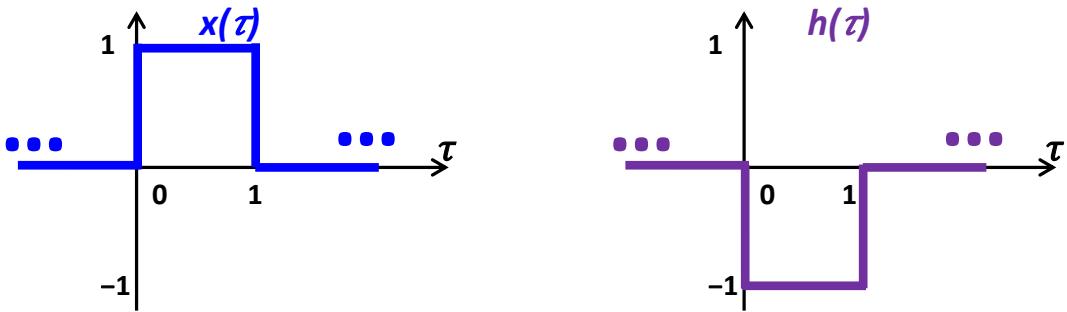


$$y(t) = \int_{\tau=0}^t \cos(\tau) \cdot 1 \, d\tau = \sin(\tau) \Big|_{\tau=0}^t = \sin(t) - \sin(0) = \sin(t)$$

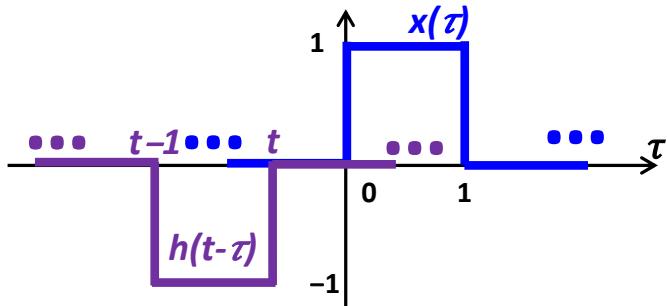
- Total Solution:**

$$y(t) = x(t) * h(t) = \begin{cases} 0, & t < 0 \\ \sin(t), & t \geq 0 \end{cases} = \sin(t)\mu(t)$$

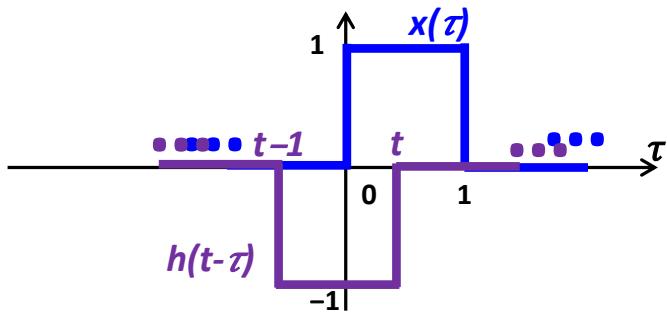
- b) Use direct integration of the convolution integral/graphical convolution to determine the convolution  $y(t) = x(t) * h(t)$  if  $x(t) = \mu(t) - \mu(t-1)$  and  $h(t) = -\mu(t) + \mu(t-1)$ .



- For  $t < 0$ : No overlap. Convolution equals zero in this range.

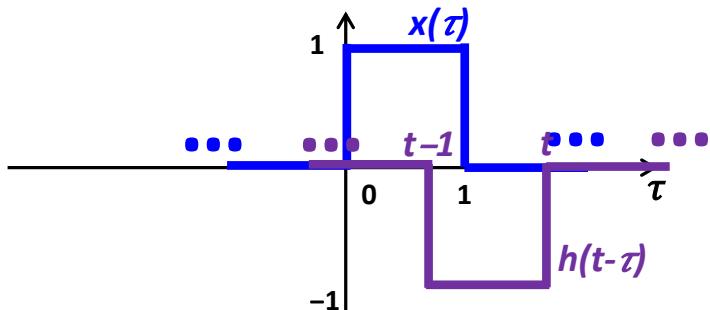


- For  $0 \leq t \leq 1$ : Partial overlap from the left.



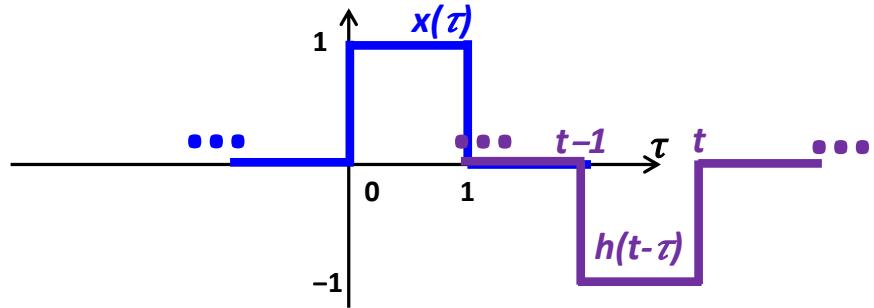
$$y(t) = \int_{\tau=0}^t 1 \cdot (-1) d\tau = -\tau \Big|_{\tau=0}^t = -t - 0 = -t$$

- For  $1 \leq t \leq 2$ : Partial overlap from the right.



$$y(t) = \int_{\tau=t-1}^1 1 \cdot (-1) d\tau = -\tau \Big|_{\tau=t-1}^1 = -1 - (t-1) = t - 2$$

- For  $t > 2$ : No overlap. Convolution equals zero in this range.



- Total Solution:

$$y(t) = x(t) * h(t) = \begin{cases} 0, & t < 0 \\ -t, & 0 \leq t \leq 1 \\ t-2, & 1 < t \leq 2 \\ 0, & t > 2 \end{cases}$$