

ECE4904 Lecture 11

MOS Capacitor Review (16.1)

Energy Band Diagram (16.2.1)

Flat Band potential ϕ_F

Surface potential ϕ_S

Accumulation/Depletion/Inversion (16.2.2)

“Delta-Depletion” solution

Threshold Voltage (16.3)

MOSFET Quantitative (17.2)

Surface Mobility

“On” Resistance (Triode region)

Square Law (Saturation region)

Handout package:

Ch. 16, 17 figures

HW 5 due Thu 12/6

Critical Quantities:

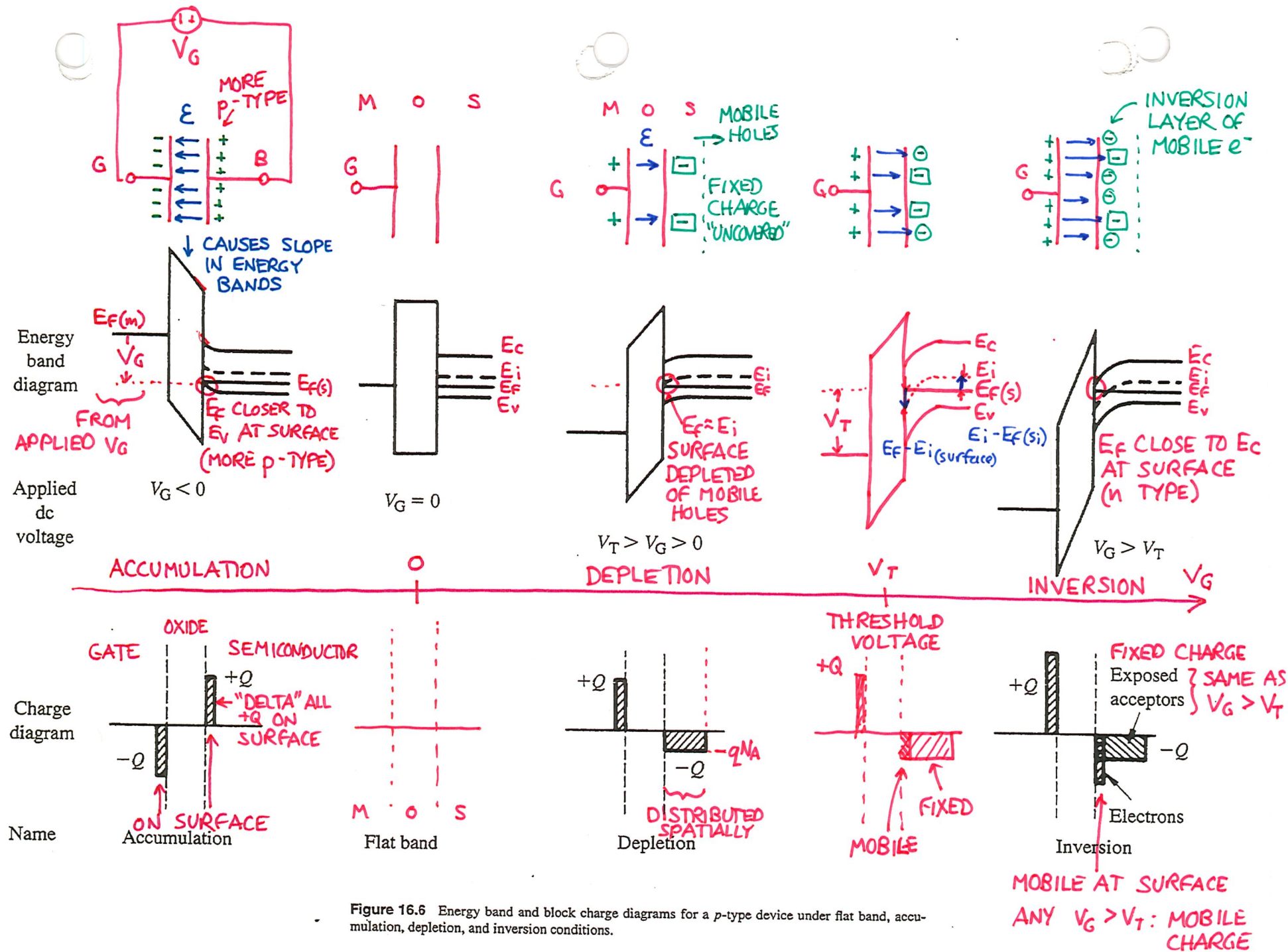
MOS structure:

ϕ_F Flat Band potential

x_O Oxide thickness

Change with applied voltage:

ϕ_S Surface potential

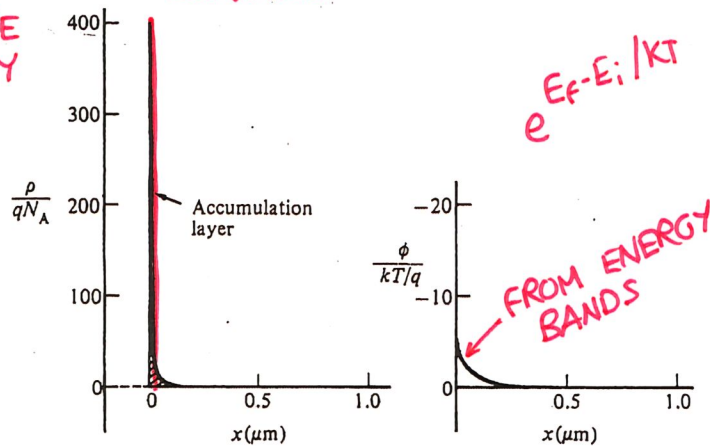


MOBILE AT SURFACE
ANY $V_G > V_T$: MOBILE CHARGE

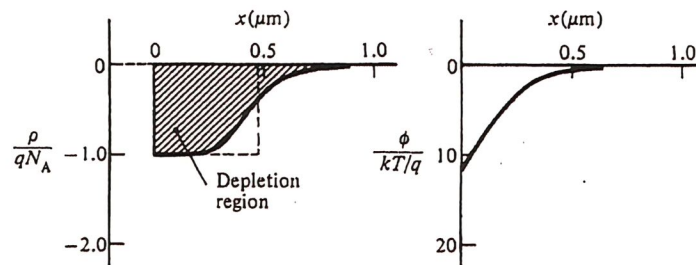
"DELTA DEPLETION APPROXIMATION"

MODEL ALL
CHARGE AT
SURFACE

CHARGE
DENSITY

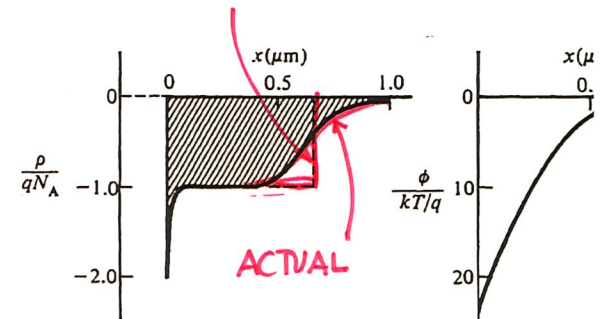


(a) Accumulation ($\phi_s = -6kT/q$)

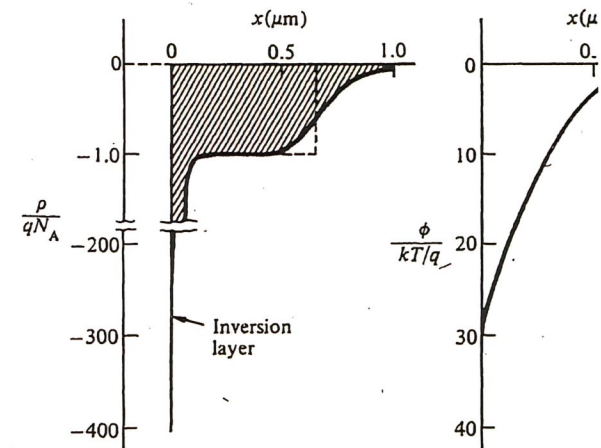


(b) Middle of depletion ($\phi_s = \phi_F = 12kT/q$)

DEPLETION APPROXIMATION



(c) Onset of inversion ($\phi_s = 2\phi_F = 24kT/q$)



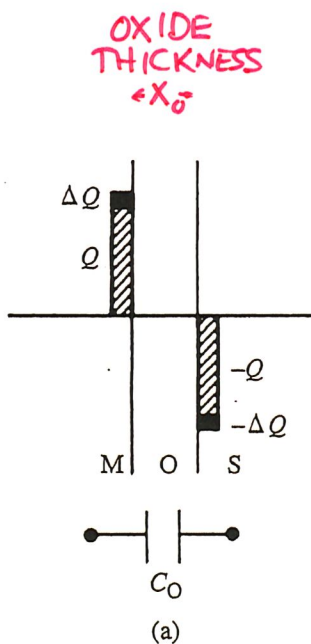
(d) Deep into inversion ($\phi_s = 2\phi_F + 6kT/q$)

Figure 16.8 Exact solution for the charge density and potential inside the semiconductor component of an MOS-C assuming $\phi_F = 12kT/q$ and $T = 300$ K ($kT/q = 0.0259$ V). (a) Accumulation ($\phi_s = -6kT/q$), (b) middle of depletion ($\phi_s = \phi_F = 12kT/q$), (c) onset of inversion ($\phi_s = 2\phi_F = 24kT/q$), and (d) heavily inverted ($\phi_s = 2\phi_F + 6kT/q = 30kT/q$). The ρ -diagrams were drawn on a linear scale and the $+\phi$ axes oriented downward to enhance the correlation with the diagrams sketched in Fig. 16.6. The dashed lines on the part (b) through (d) ρ -plots outline the depletion approximation version of the charge distribution.

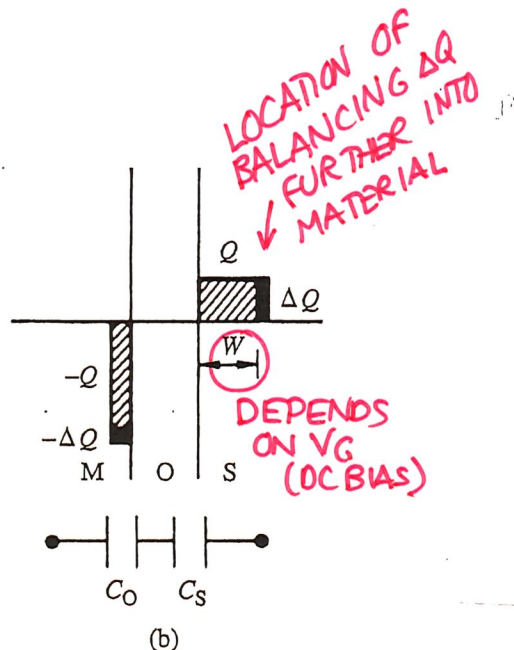
Figure 16.8 Continued.

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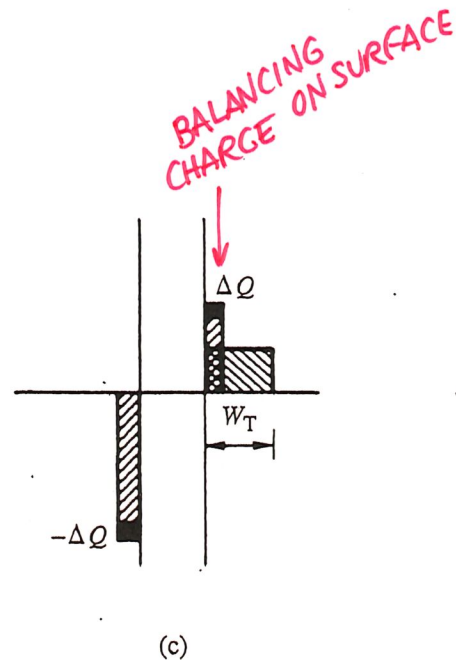
n SUBSTRATE



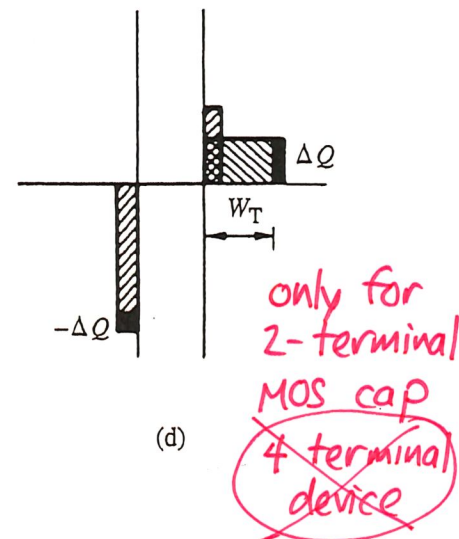
ACCUMULATION



DEPLETION



INVERSION



$$C = \frac{1}{\frac{1}{C_o} + \frac{1}{C_s}}$$

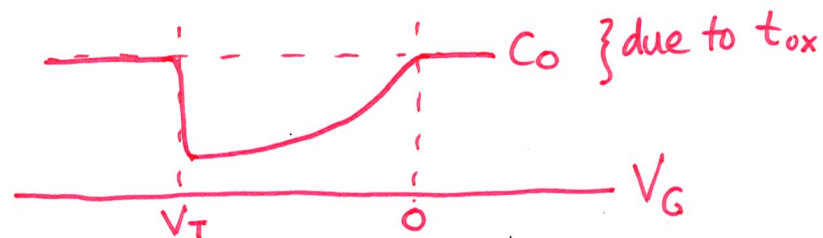
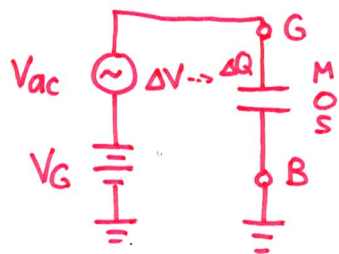


Figure 16.12 a.c. charge fluctuations inside an n-type MOS-capacitor under d.c. biasing conditions corresponding to (a) accumulation, (b) depletion, (c) inversion when $\omega \rightarrow 0$, and (d) inversion when $\omega \rightarrow \infty$. Equivalent circuit models appropriate for accumulation and depletion biasing are also shown beneath the block charge diagrams in parts (a) and (b), respectively.



$$Q = CV$$

$$\Delta Q = C \Delta V$$

$$C_g = \frac{\Delta Q}{\Delta V}$$

ECE4904 Threshold Voltage

Flat band voltage:

TELLS YOU IF
SUBSTRATE
IS p- OR n-TYPE

p-type bulk (NMOS)

$$\phi_F = \frac{kT}{q} \ln\left(\frac{N_A}{n_i}\right)$$

n-type bulk (PMOS)

$$\phi_F = -\frac{kT}{q} \ln\left(\frac{N_D}{n_i}\right)$$

— E_F
- - - E_i

Threshold voltage:

VOLTAGE DROP IN OXIDE: GATE TO SURFACE

$$V_T = 2\phi_F + \frac{K_S}{K_O} x_O \sqrt{\frac{4qN_{SUB}}{K_S\epsilon_0}} \phi_F$$

INVERT Si AT
OXIDE SURFACE

Substitute general expression for flat band voltage:

$$\phi_F = \frac{kT}{q} \ln\left(\frac{N_{SUB}}{n_i}\right)$$

Threshold voltage depends primarily on oxide thickness x_O , substrate doping N_{SUB} :

$$V_T = \frac{2kT}{q} \ln\left(\frac{N_{SUB}}{n_i}\right) + \frac{K_S}{K_O} x_O \sqrt{\frac{4kT N_{SUB}}{K_S\epsilon_0} \ln\left(\frac{N_{SUB}}{n_i}\right)}$$

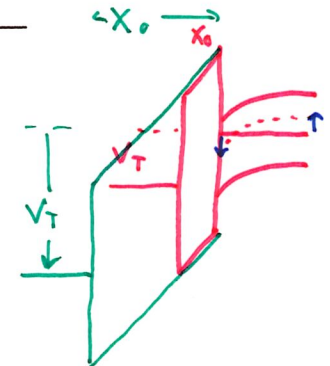
$N_{SUB} \uparrow \Rightarrow |V_T| \uparrow$

NEED TO DEplete
HIGHER CONCENTRATION
OF DOPANT ATOMS

OXIDE
THICKNESS

$x_O \uparrow \Rightarrow |V_T| \uparrow$

NEED MORE V_G TO GET
NEEDED ϕ_S TO INVERT CHANNEL



Z, L DETERMINED
BY GATE GEOMETRY
(MASK USED IN
FABRICATION [Ch. 4])

PROCESS
PARAMETER
OXIDE
THICKNESS

MOSFETS—THE ESSENTIALS

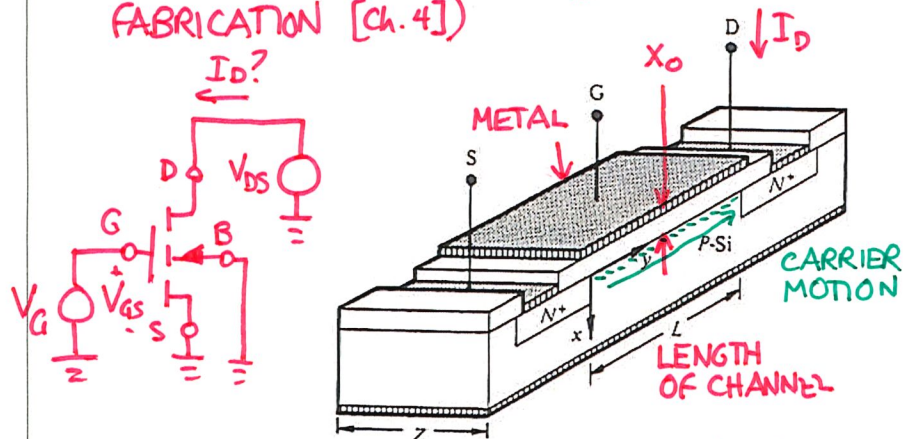


Figure 17.6 Device structure, dimensions, and coordinate orientations assumed in the quantitative analysis.

START: DC CURRENT $I_D = \frac{\Delta Q}{\Delta t}$ ALL MOBILE CHARGE IN CHANNEL TIME S \rightarrow D

$$\Delta t \text{ FROM VELOCITY } v = \frac{L}{\Delta t} = \underbrace{\mu E}_{\text{MOBILITY}} \underbrace{E = \frac{V_{DS}}{L}}_{\text{HORIZONTAL E FIELD}} \Rightarrow \Delta t = \frac{L^2}{\mu V_{DS}}$$

$$I_D = \frac{\Delta Q}{\Delta t} = \frac{C_0 Z L (V_G - V_T) \mu V_{DS}}{L^2} \text{ DRAIN VOLTAGE}$$

DRAIN
CURRENT

"ON"
RESISTANCE

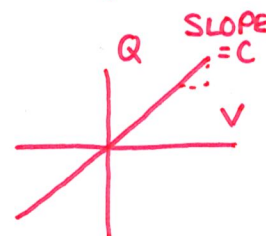
$$\frac{V_{DS}}{I_D} = \frac{1}{\mu C_0 \frac{Z}{L} (V_{GS} - V_T)} \quad r_{ds(on)} \rightarrow 0$$

HIGHER μ \rightarrow $\frac{K_0 \epsilon_0}{x_0}$ SIZE V_{GS} HIGHER
 $\mu_n > \mu_p$ HIGH K_0 x_0 THIN OXIDE WIDE SHORT
DIELEC

FOR ΔQ : WHEN WE APPLY V_G TO GATE
619 HOW MUCH MOBILE CHARGE IN CHANNEL?

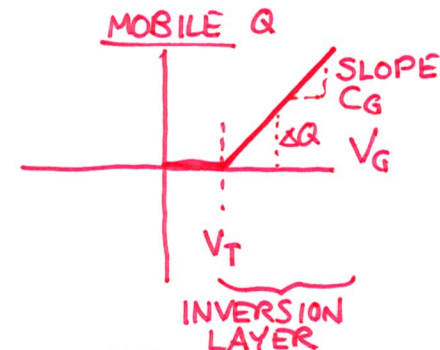
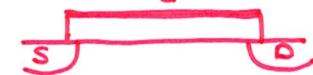
IDEAL CAP

$$\begin{array}{c} +Q \\ | \\ -Q \end{array} Q = CV$$



$$C = \epsilon \frac{A}{d}$$

MOS CAPACITOR
 V_G



GATE OXIDE CAPACITANCE

$$C_G = \epsilon \frac{A}{d} = K_0 \epsilon_0 \frac{ZL}{x_0}$$

$$C_G = \frac{K_0 \epsilon_0}{x_0} \underbrace{ZL}_{\text{AREA OF GATE}} \underbrace{C_0}_{\text{GATE OXIDE CAPACITANCE/UNIT AREA}}$$

GATE OXIDE CAPACITANCE/UNIT AREA
MOBILE CHARGE IN CHANNEL:

$$\Delta Q = C_0 Z L (V_G - V_T)$$

ONLY $V_G > V_T$ MAKES
MOBILE e^- IN CHANNEL

$V_G < V_T$ CUTOFF $I_D = 0$
 $V_G > V_T$