

TSDT14 Signal Theory

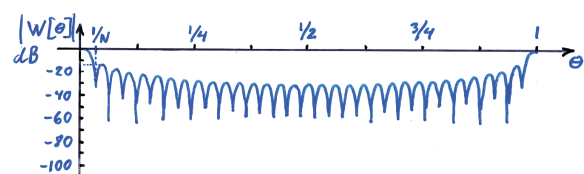
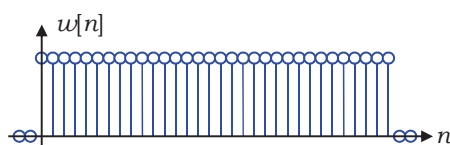
Lecture 4 Estimation

Mikael Olofsson
Department of EE (ISY)
Div. of Communication Systems

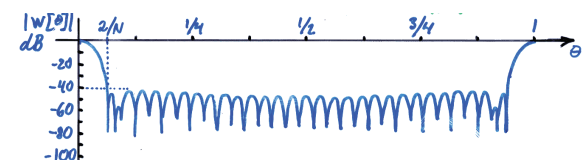
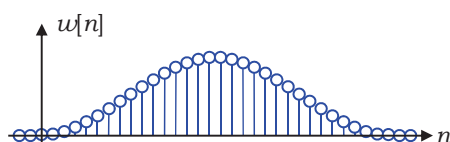


Examples of Windows, $N=32$

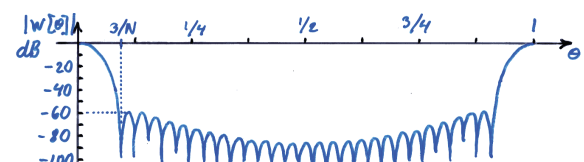
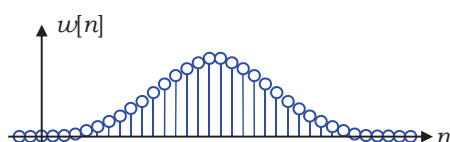
Rectangular window: $w[n] = 1, \quad n \in \{0, 1, \dots, N-1\}$



Hamming window: $w[n] = 0.54 - 0.46\cos(\frac{2\pi n}{N-1}), \quad n \in \{0, 1, \dots, N-1\}$

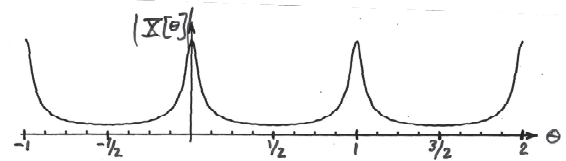
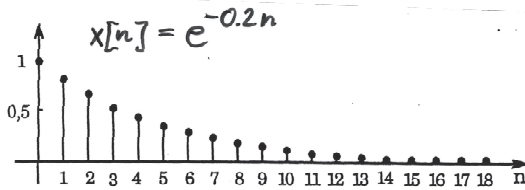


Blackman window: $w[n] = 0.42 - 0.5\cos(\frac{2\pi n}{N-1}) + 0.08\cos(\frac{4\pi n}{N-1}), \quad n \in \{0, 1, \dots, N-1\}$

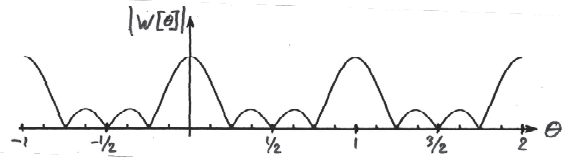
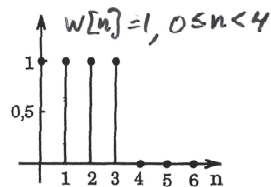


Using Windows

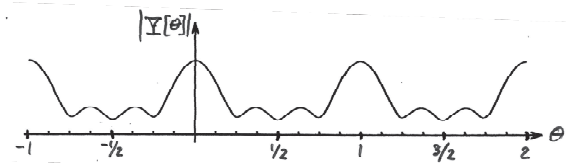
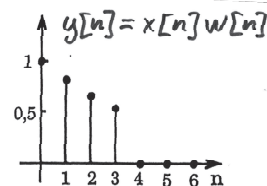
Signal:



Rectangular window:



Result:



Ergodicity again

A WSS process:

$$m_X = E\{X(t)\}$$

Time average of one realization:

$$m_T = \frac{1}{2T} \int_{-T}^T x(t) dt$$

Time average of the process:

$$M_T = \frac{1}{2T} \int_{-T}^T X(t) dt$$

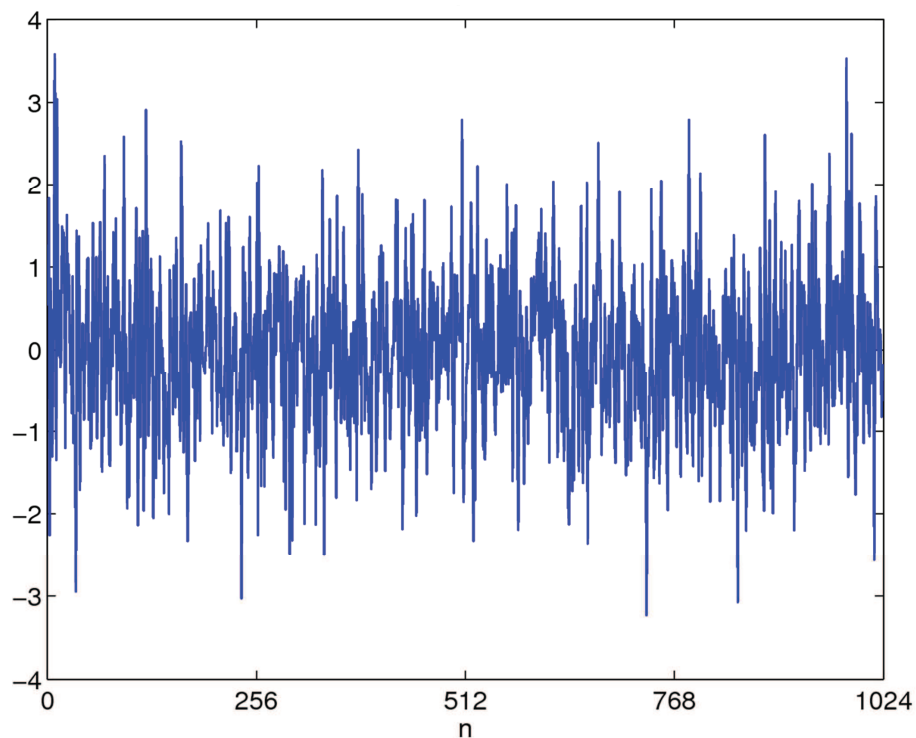
$$E\{M_T\} = E\left\{\frac{1}{2T} \int_{-T}^T X(t) dt\right\} = \frac{1}{2T} \int_{-T}^T E\{X(t)\} dt = \frac{1}{2T} \int_{-T}^T m_X dt = m_X$$

Definition: If $\lim_{T \rightarrow \infty} E\{(M_T - m_X)^2\} = 0$ then $X(t)$ is said to be ergodic with respect to the mean, and we write

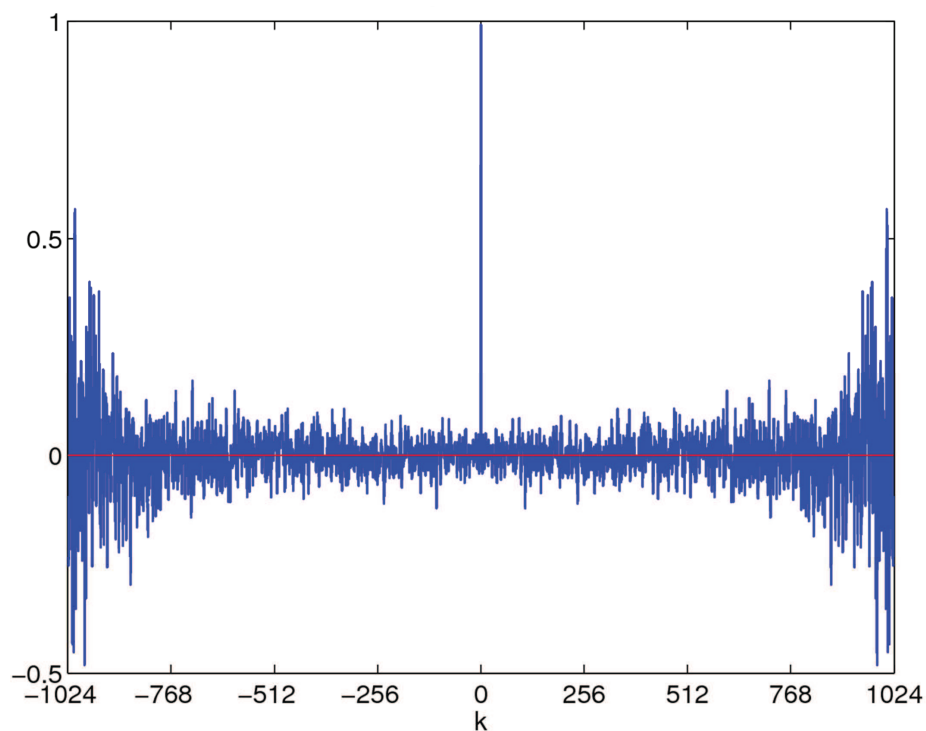
$$m_X = \text{l.i.m.}_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T X(t) dt \quad (\text{limes in mean})$$

Interpretation: The time average of a realization is very close to the ensemble mean with a probability that is very close to 1 ($\rightarrow 1, T \rightarrow \infty$).

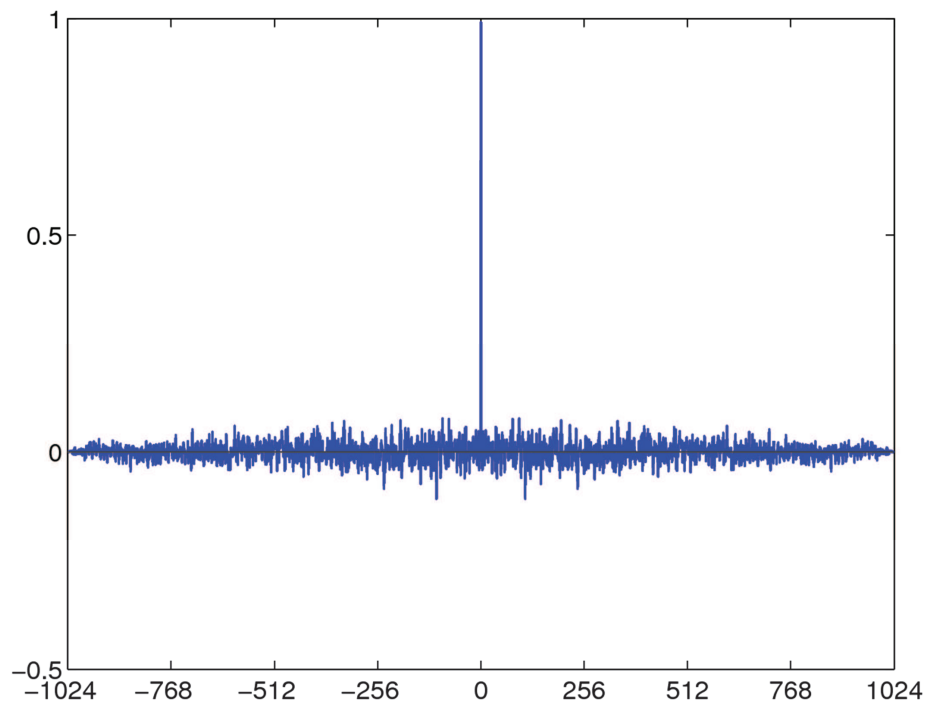
A Signal



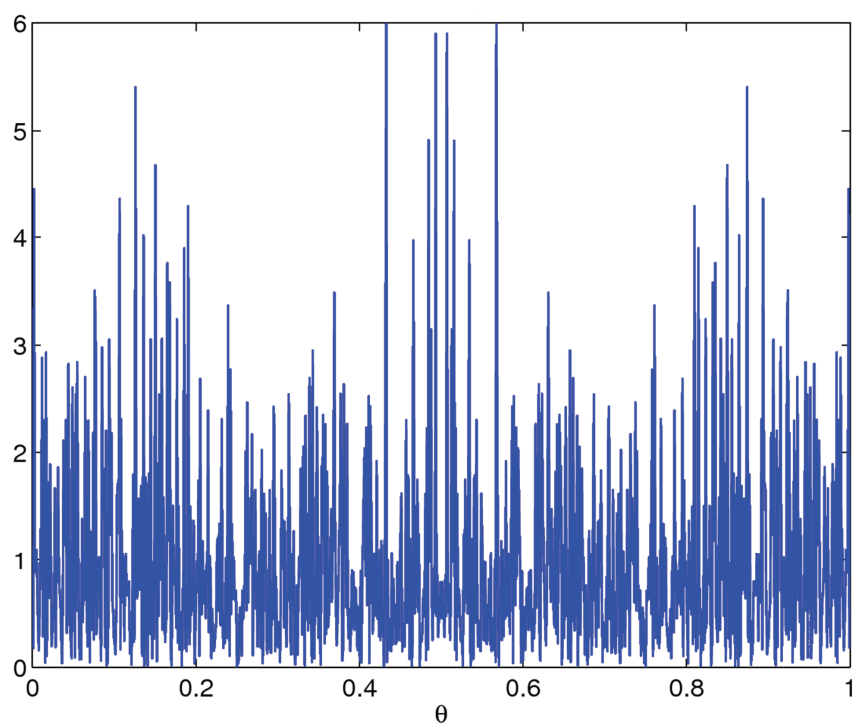
Blackman-Tukey's Estimate of the ACF



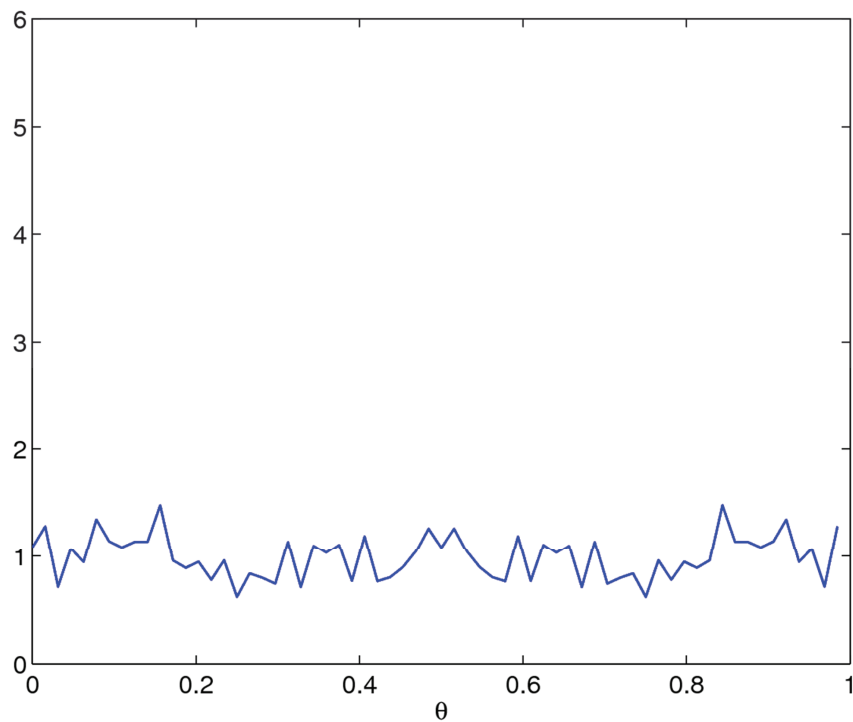
Bartlett's Estimate of the ACF



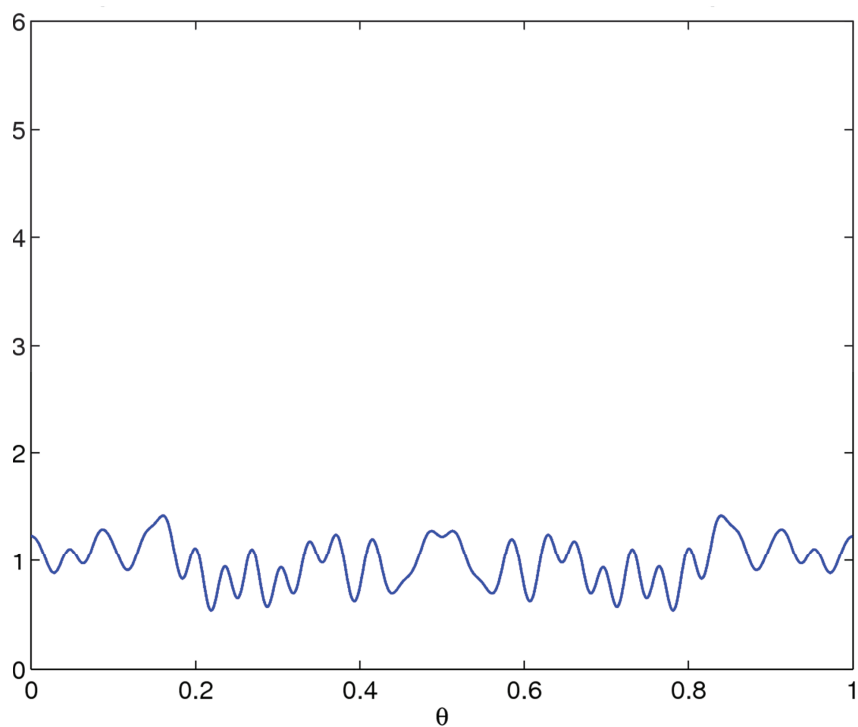
Periodogram



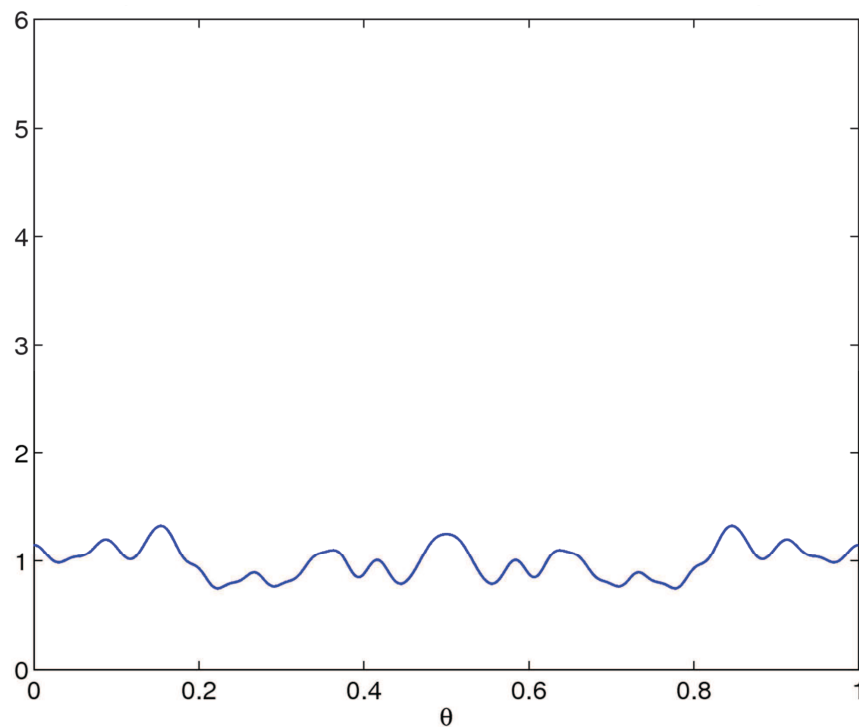
Averaged Periodogram



Estimated PSD using Rectangular Window

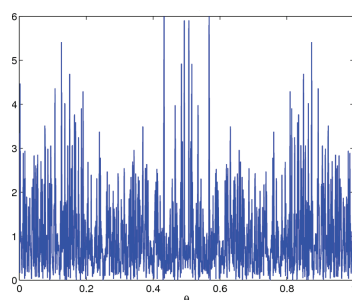


Estimated PSD using Hamming Window

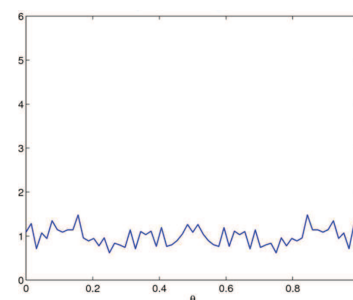


Smoothing – Overview

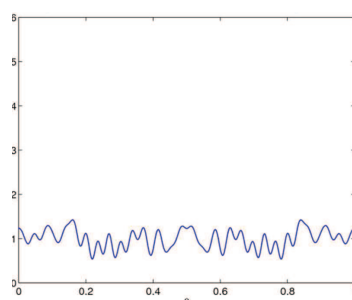
Raw periodogram



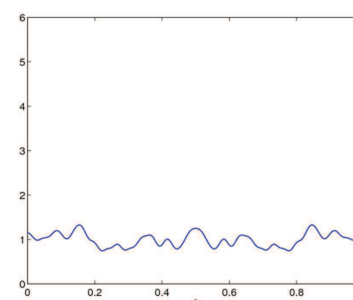
Averaged periodograms



Smoothing - Rectangular



Smoothing - Hamming



DFT – Signal Analysis

Time-discrete signal with limited duration:

$$x[n] = 0 \text{ for } n \notin \{0, 1, \dots, N-1\}$$

Fourier transform: $X[\theta] = \sum_n x[n] e^{-j2\pi\theta n} = \sum_{n=0}^{N-1} x[n] e^{-j2\pi\theta n}$
cont. w. period 1.

DFT of length L : $X_L[k] = X[k/L] = \sum_{n=0}^{N-1} x[n] e^{-j2\pi \frac{k}{L} n}$ for $k \in \{0, 1, \dots, L-1\}$

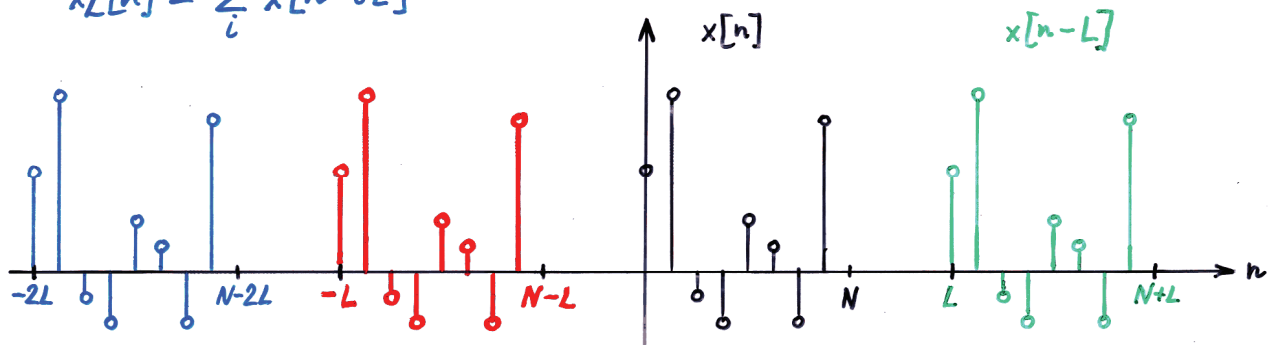
IDFT (inverse): $x_L[n] = \frac{1}{L} \sum_{k=0}^{L-1} X_L[k] e^{j2\pi \frac{k}{L} n} \Rightarrow x_L[n+L] = x_L[n]$
since $e^{j2\pi L n/L} = e^{j2\pi n} = 1$

Relation to $x[n]$:
$$x_L[n] = \frac{1}{L} \sum_{k=0}^{L-1} \sum_{m=0}^{N-1} x[m] e^{-j2\pi \frac{k}{L} m} e^{j2\pi \frac{k}{L} n}$$

$$= \sum_{m=0}^{N-1} x[m] \cdot \underbrace{\frac{1}{L} \sum_{k=0}^{L-1} e^{-j2\pi(m-n)\frac{k}{L}}}_{= \begin{cases} 1, & m-n=0 \bmod L \\ 0, & \text{elsewhere} \end{cases}} = \sum_{i=-\infty}^{\infty} x[n-iL]$$

DFT – Avoiding Aliasing

$$x_L[n] = \sum_i x[n-iL]$$



If $L < N$, then we get overlap and aliasing in the time domain.

Therefore: Demand $L \geq N$.

Note:
$$X_L[k] = \sum_{n=0}^{N-1} x[n] e^{-j2\pi \frac{k}{L} n} = \sum_{n=0}^{L-1} x_L[n] e^{-j2\pi \frac{k}{L} n}$$

DFT – Periodic Convolution

We are used to: $y[n] = (x * h)[n] \Leftrightarrow Y[\theta] = X[\theta] \cdot H[\theta]$

But we have: $y[n] = x[n] \cdot h[n] \Leftrightarrow Y[\theta] = \int_0^1 X[\phi] H[\theta - \phi] d\phi$

With DFT: $Y_L[k] = X_L[k] \cdot H_L[k] \Leftrightarrow$

$$\begin{aligned} y_L[n] &= \text{IDFT}\{X_L[k] H_L[k]\} = \frac{1}{L} \sum_{k=0}^{L-1} X_L[k] H_L[k] e^{j2\pi \frac{k}{L} n} \\ &= \frac{1}{L} \sum_{k=0}^{L-1} X_L[k] \cdot \sum_{m=0}^{L-1} h_L[m] \cdot e^{-j2\pi \frac{k}{L} m} \cdot e^{j2\pi \frac{k}{L} n} \\ &= \sum_{m=0}^{L-1} h_L[m] \cdot \underbrace{\frac{1}{L} \sum_{k=0}^{L-1} X_L[k] e^{j2\pi \frac{k}{L} (n-m)}}_{x_L[n-m]} \\ &= \sum_{m=0}^{L-1} h_L[m] x_L[n-m] \end{aligned}$$

And also: $y_L[n] = x_L[n] \cdot h_L[n] \Leftrightarrow Y_L[k] = \frac{1}{L} \sum_{m=0}^{L-1} X_L[m] H_L[k-m]$

Mikael Olofsson
ISY/CommSys

www.liu.se