# CHAPTER 3: TIME-DOMAIN ANALYSIS OF DISCRETE-TIME SYSTEMS

linear, time-invariant, discrete-time (LTID)

### 3.1 Introduction

discrete-time signal: sequence of numbers

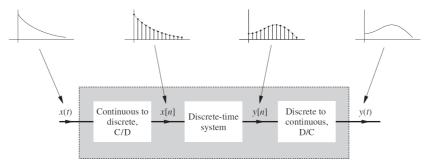


Figure 3.2 Processing a continuous-time signal by means of a discrete-time system.

continuous-time exponential  $x(t) = e^{-t}$  sampled every T = 0.1 seconds results in a discrete-time signal x(nT):

$$x(nT) = e^{-nT} = e^{-0.1n}$$

#### 3.1-1 Size of a Discrete-Time Signal

Size measured by energy:

$$E_x = \sum_{n=-\infty}^{\infty} \left|x[n]
ight|^2 \qquad \qquad (3.1.1\text{-}1)$$

if  $E_x$  is finite, signal is **energy signal**. Else, measured by signal power:

$$P_x = \lim_{N o \infty} rac{1}{2N+1} \sum_{-N}^N |x[n]|^2 \eqno(3.1.1-2)$$

\*2N+1 samples in interval from -N to N

# 3.2 Signal Operations

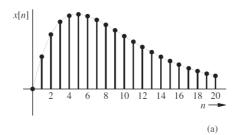
Shifting (by M units)

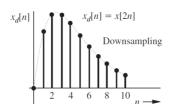
$$x_s[n] = x[n-M]$$

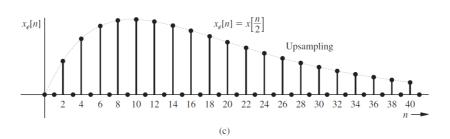
**Time Reversal** 

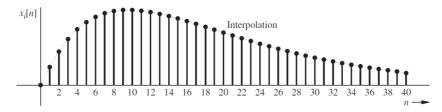
$$x_r[n] = x[-n]$$

Sampling Rate Alteration: Downsampling, Upsampling, Interpolation









Downsampling: Compression by factor M

$$x_d[n] = x[Mn], M \in \mathbb{N}^+$$

Interpolated signal

$$x_e[n] = egin{cases} x[n/L] & n = 0, \pm L, \pm 2L, \dots \ 0 & ext{otherwise} \end{cases}$$

Upsampling: L times that of x[n]: general sequence:

$$x_e[n] = x[0], \underbrace{0, 0, \dots, 0, 0}_{L-1 \ zeros}, x[1], \underbrace{0, 0, \dots, 0, 0}_{L-1 \ zeros}, x[2], \underbrace{0, 0, \dots, 0, 0}_{L-1 \ zeros}, \dots$$

# 3.3 Discrete-Time Signal Models

## 3.3-1 Discrete-Time Impulse Function $\delta[n]$

Unit impulse sequence: Kronecker delta

$$\delta[n] = \begin{cases} 1 & n = 0 \\ 0 & n \neq 0 \end{cases}$$
 (3.3.1-1)

## **3.3-2 Discrete-Time Unit Step Function** u[n]

$$u[n] = egin{cases} 1 & ext{for } n \geq 0 \ 0 & ext{for } n < 0 \end{cases}$$

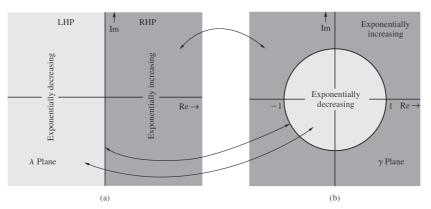
#### 3.3-3 Discrete-Time Exponential $\gamma^n$

continuous-time exponential  $e^{\lambda t}$  can be alternatively expressed as:

$$e^{\lambda t} = \gamma^t$$
  $(\gamma = e^{\lambda} \text{ or } \lambda = \ln \gamma)$ 

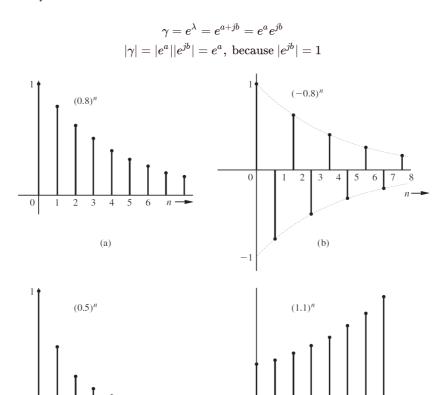
discrete:

$$\gamma^n = e^{\lambda n} \qquad (\gamma = e^{\lambda} ext{ or } \lambda = \ln \gamma)$$



**Figure 3.8** The  $\lambda$  plane, the  $\gamma$  plane, and their mapping.

e.g. signal  $e^{\lambda n}$  where  $\lambda$  lies on left half-plane (  $\lambda=a+jb, a<0$  ), exponential decay



**Figure 3.9** Discrete-time exponentials  $\gamma^n$ .

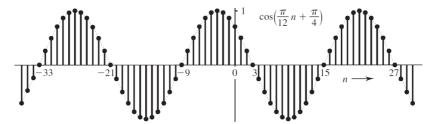
(c)

## 3.3-4 Discrete-Time Sinusoid $cos(\Omega n + \theta)$

General discrete-time sinusoid:

$$C\cos(\Omega n + heta) = C\cos(2\pi F n + heta)$$
 where  $F = \Omega/2\pi$ 

(d)



**Figure 3.11** A discrete-time sinusoid  $\cos(\frac{\pi}{12}n + \frac{\pi}{4})$ .

Sampled Continuous-Time Sinusoid Yields a Discrete-Time Sinusoid

A continuous-time sinusoid,  $\cos wt$ , sampled every T seconds yields a discrete-time sinusoid. Sample signal x[n]:

$$x[n] = \cos \omega n T = \cos \Omega n$$
 where  $\Omega = \omega T$ 

# 3.3-5 Discrete-Time Complex Exponential $e^{j\Omega n}$

$$e^{j\Omega n} = (\cos\Omega n + j\sin\Omega n)$$
  
 $e^{-j\Omega n} = (\cos\Omega n - j\sin\Omega n)$ 

For r=1 and  $\theta=n\Omega$ ,

$$e^{j\Omega n} = re^{j\theta}$$

# 3.4 Examples of Discrete-Time Systems

#### **Savings Account**

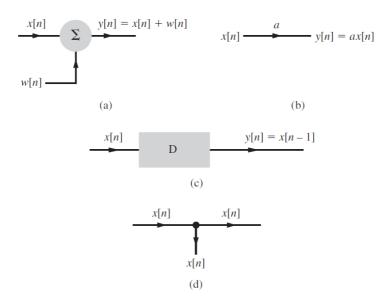


Figure 3.13 Schematic representations of basic operations on sequences.

x[n] = deposit made at the nth discrete instant

y[n] = account balance at the  $n^{th}$  instant computed immediately after receipt of the  $n^{th}$  deposit x[n]

r = interest per dollar per period T

$$y[n] = y[n-1] + ry[n-1] + x[n]$$
  
=  $(1+r)y[n-1] + x[n]$ 

Delayed form:

$$y[n] - ay[n-1] = x[n], \ a = 1 + r$$
 (3.4-1)

Advanced form:

$$y[n+1] - ay[n] = x[n+1]$$
 (3.4-2)

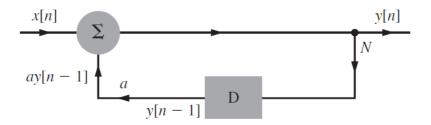


Figure 3.14 Realization of the savings account system.

- addition
- · scalar multiplication
- delay
- pickoff node (node N): provides multiple copies of a signal at input

#### **Sales Estimate**

y[n]: new books sold by publisher

x[n]: students enrolled in  $n^{th}$  semester

book life: 3 semesters

1/4 students resell texts

$$y[n] + \frac{1}{4}y[n-1] + \frac{1}{16}y[n-2] = x[n]$$
 (3.4-3)

$$y[n+2] + \frac{1}{4}y[n+1] + \frac{1}{16}y[n] = x[n+2]$$
 (3.4-4)

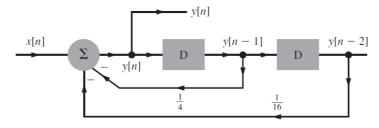


Figure 3.15 Realization of the system representing sales estimate in Ex. 3.7.

#### **Digital Differentiator**

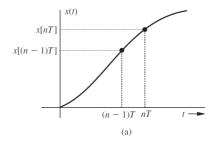
- used in audio system with input signal bandwidth below 20kHz
- output y(t) is derivative of input x(t)

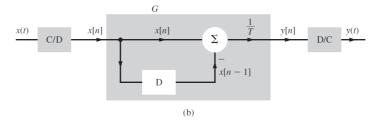
x[n], y[n]: samples of signals x(t) and y(t), T seconds apart.

$$\begin{aligned} x[n] &= x(nT) \\ y[n] &= y(nT) \end{aligned} \tag{3.4-5}$$

@t = nT:

$$egin{aligned} y(t) &= rac{dx(t)}{dt} \ y(nT) &= rac{dx(t)}{dt} igg|_{t=nT} \ &= \lim_{T o 0} rac{1}{T} [x(nT) - x[(n-1)T]] \end{aligned}$$





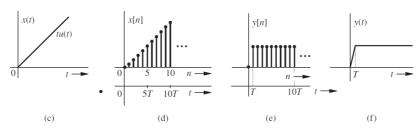


Figure 3.16 Digital differentiator and its realization.

backward difference system:

$$y[n] = \lim_{T o 0}rac{1}{T}[x[n]-x[n-1]]$$

but

- in practice,  $T \neq 0$
- T is sufficiently small

non-recursive form:

$$y[n] = \frac{1}{T}[x[n] - x[n-1]] \tag{3.4-6}$$

$$y[n] = \frac{1}{T}[x[n+1] - x[n]] \tag{3.4-7}$$

**Digital Integrator** 

$$egin{aligned} y(t) &= \int_{-\infty}^t x( au) d au \ y(nT) &= \lim_{T o 0} \sum_{k=-\infty}^n x(kT) T \ y[n] &= \lim_{T o 0} T \sum_{k=-\infty}^n x[k] \end{aligned}$$

accumulator system: non-recursive form

$$y[n] = T \sum_{k=-\infty}^{n} x[k]$$
 (3.4-8)

recursive form:

$$y[n] - y[n-1] = Tx[n]$$
 (3.4-9)

#### **Recursive & Non-Recursive forms of Difference Equation**

Recursive:

- 3.4-1
- 3.4-3
- 3.4-9

Non-Recursive:

- 3.4-6
- 3.4-8

## **Kinship of Difference Equations to Differential Equations**

• differential equation can be approximated by a difference equation of the same order

1st ODE:

Assuming non-zero, but very small T:

$$y[n] = \alpha y[n-1] = \beta x[n]$$
 (3.4-10)

where

$$lpha = rac{-1}{1+cT} \ eta = rac{T}{1+cT}$$

advance form:

$$y[n+1] + \alpha y[n] = \beta x[n+1] \tag{3.4-11}$$

## **Order of Difference Equation**

difference equations:

- 3.1
- 3.3
- 3.7
- 3.9
- 3.11

#### 1st ODEs:

- 3.1
- 3.7
- 3.9
- 3.11

#### Analog, Digital, Continuous-Time & Discrete-Time Systems

- digital filters => discrete-time systems
- analog filters => continuous-time system

#### **Advantages of Digital Signal Processing**

- 1. **Precision & Stability**: can tolerate considerable variation, hence less sensitive in component parameter
- Easily Duplicated & Fully Integrated: doesn't require factory adjustments, complex systems placed on VLSI (very-large-scaleintegrated) circuits
- 3. Flexible: easily alterable characteristics
- 4. Greater Variety of Filters
- 5. Easy & Cheap **Storage**: can also search/select information from
- 6. Low Error Rates & High Fidelity: privacy & sophistication
- 7. Simultaneously serve a number of inputs: time-shared
- 8. Reliable Reproduction without Deterioration

### 3.4-1 Classification of Discrete-Time Systems

#### **Linearity & Time Invariance**

- linearity => continuous-time systems
- time/shift-invariant => discrete-time systems
  - systems whose parameter doesn't change with time

#### **Causal & Noncausal Systems**

• **causal** if output at any instant n=k depends only on value of input x[n] for  $n \leq k$ . Output depends only on past & present values of input.

## **Invertible & Noninvertible Systems**

 invertible if ∃ inverse system S<sub>i</sub> s.t. the cascade of S and S<sub>i</sub> results in an identity system (output is identical to input)

#### Stable & Unstable Systems

- internal, OR
- external: bound input results in bounded output

#### **Memoryless Systems & Systems with Memory**

ullet memoryless if it responds at an instant n depending on input at the same time instant

## **DT System Properties**

- 1. **linear**: requires homogeneity & additivity
- 2. **time-invariant**: shift in input => shift in output
- 3. causal: doesn't depend on future values
- 4. **invertible**: every input generates unique output
- 5.  $\boldsymbol{BIBO\text{-stable}}:$  bound input results in bounded output
- 6. **memoryless**: depend on strength of current input