

Description

The 8V97003 is a high-performance mmWave wideband Synthesizer / Phase Lock Loop (PLL) that generates output frequencies up to 18GHz from an integrated Voltage Controlled Oscillator (VCO) offering an octave of frequency tuning range. The device also offers a high-performance 32-bit fractional feedback divider and an output divider to allow users to fully benefit from the wideband characteristics of the VCO.

The device's figure of merit (FOM) of -236dBc/Hz and the excellent VCO performance allow for very low phase noise and RMS phase jitter.

The 8V97003 offers a very low output-to-output phase skew drift of $< 10^\circ$ across all operating conditions and frequencies, reducing radio path recalibration occurrences in beamforming applications, such as 5G radio card massive MIMO systems.

The output drivers have programmable output power settings and can deliver high single-ended output power up of +12dBm at 8GHz, and +4dBm at 18GHz, when using inductively loaded output terminations (double termination). When the outputs are resistively loaded, the output drivers can deliver a single-ended output power of +9.5dBm at 8GHz, and up to -2.5dBm at 18GHz. The output power can be further increased when using differential outputs and measuring the output power differentially.

The 8V97003 relies on a single 3.3V power supply and offers low noise integrated LDOs for excellent power supply noise immunity.

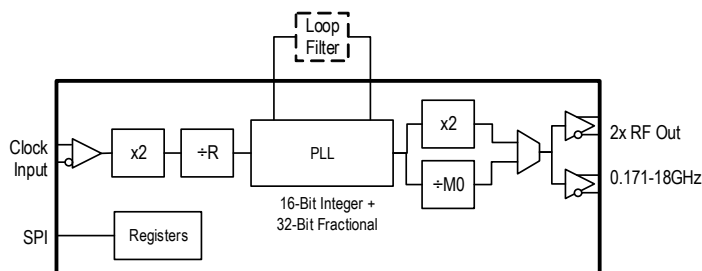
Typical Applications

- 5G millimeter wave wireless infrastructure
- Massive MIMO
- Phase Array Antennas and beam forming
- Wireless backhaul
- Point-to-point and point-to-multipoint microwave links
- Satellites / VSATs
- Test equipment/instrumentation
- Clock generation
- High-speed RF converters sampling clocks
- Radar

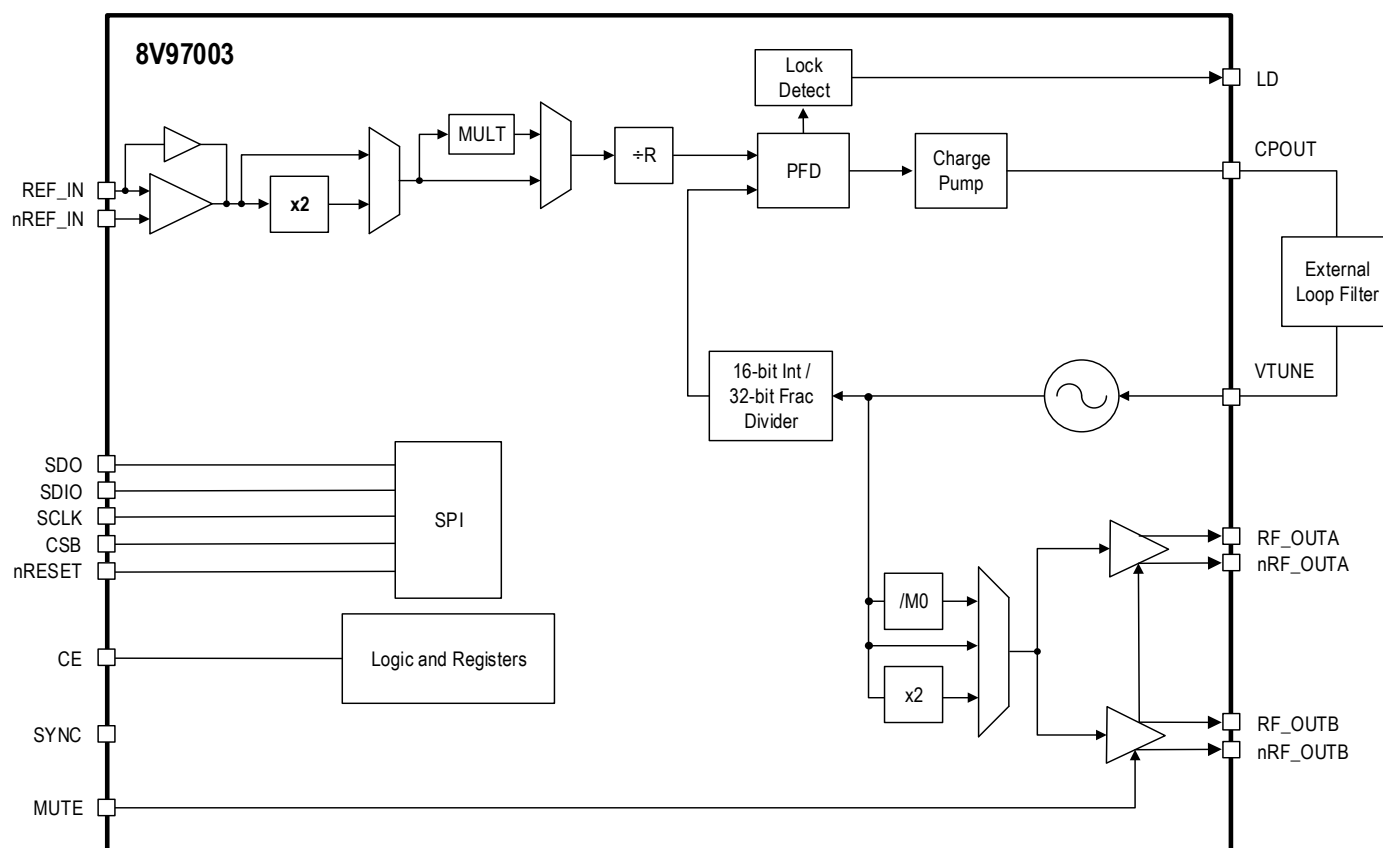
Features

- Output frequency range: 171.875MHz to 18GHz
- Ultra-low phase noise VCO
 - -60.6dBc integrated phase jitter (35fs rms jitter) from 20kHz to 100MHz at 6GHz
- Figure of Merit: -236dBc/Hz
- Input reference frequency:
 - 10MHz to 1.6GHz (LVPECL, LVDS)
 - 10MHz to 250MHz (LVCMOS)
- Fractional-N synthesizer and integer-N synthesizer
- 32-bit of fractional and modulus resolution
- Phase frequency detector (PFD) operation up to 500MHz (Integer mode) or 250MHz (Fractional mode)
- Programmable RF output power levels
- RF output power < -80 dBm when in MUTE
- Programmable input multiplier (MULT) to increase PFD frequency when using a low input frequency
- -40°C to $+95^\circ\text{C}$ ambient temperature range; and up to $+105^\circ\text{C}$ board temperature
- 3.3V single power supply operation
- 7×7 mm 48-VFQFPN package
- SPI interface is compatible with 1.8V logic and tolerant to 3.3V
- Supported in the Timing Commander™ design tool

Simplified Block Diagram



Block Diagram



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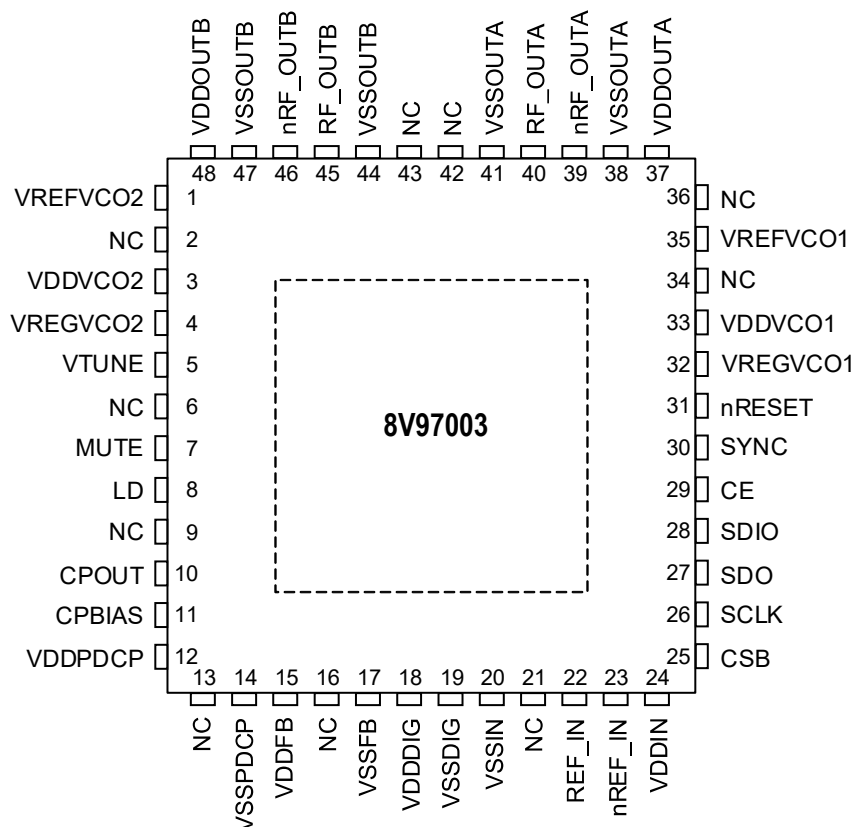
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Pin Assignments

Figure 1. Pin Assignments for 7 × 7 mm 48-VFQFPN Package – Top View



Pin Descriptions

Table 1. Pin Descriptions

| Pin Number | Name | Type | Pull-up/ Pull-Down | Description |
|------------|----------|--------|-----------------------|--|
| 1 | VREFVCO2 | Analog | | Reference node for VCO regulator. Connect 22μF capacitor from this pin to GND. |
| 2 | NC | Unused | | Do not connect. |
| 3 | VDDVCO2 | Power | | VDD power supply for VCO. |
| 4 | VREGVCO2 | Analog | | Regulator for VCO. Connect 22μF capacitor from this pin to GND. |
| 5 | VTUNE | Analog | | VCO Tuning Voltage. |
| 6 | NC | Unused | | Do not connect. |
| 7 | MUTE | Input | PD | Outputs disable / High-Impedance. 1.8V LVCMOS logic levels (3.3V tolerant). |
| 8 | LD | Output | | Lock Detector (CMOS). |
| 9 | NC | Unused | | Do not connect. |
| 10 | CPOUT | Analog | | Charge Pump Output. |

Table 1. Pin Descriptions (Cont.)

| Pin Number | Name | Type | Pull-up/ Pull-Down | Description |
|------------|----------|--------------|-----------------------|--|
| 11 | CPBIAS | Analog | | Bias node for charge pump. Connect 22μF capacitor from this pin to GND. |
| 12 | VDDPDCP | Power | | VDD power supply for phase detector and charge pump. |
| 13 | NC | Unused | | Do not connect. |
| 14 | VSSPDCP | Ground | | VSS power supply ground for phase detector and charge pump. |
| 15 | VDDFB | Power | | VDD power supply for feedback divider. |
| 16 | NC | Unused | | Do not connect. |
| 17 | VSSFB | Ground | | VSS power supply ground for feedback divider. |
| 18 | VDDDIG | Power | | VDD power supply for Digital, SPI and SDM. |
| 19 | VSSDIG | Ground | | VSS power supply ground for Digital, SPI and SDM. |
| 20 | VSSIN | Ground | | VSS power supply ground for reference input path. |
| 21 | NC | Unused | | Do not connect. |
| 22 | REF_IN | Input | PD | Differential reference clock input+ (LVDS, LVPECL, CMOS). |
| 23 | nREF_IN | Input | PD/PU | Differential reference clock input– (LVDS, LVPECL). |
| 24 | VDDIN | Power | | VDD power supply for reference input path. |
| 25 | CSB | Input | PD | SPI Chip Select Bar. 1.8V LVCMOS logic levels (3.3V tolerant). |
| 26 | SCLK | Input | | SPI Clock Input. 1.8V LVCMOS logic levels (3.3V tolerant). |
| 27 | SDO | Output | | SPI Data Output. |
| 28 | SDIO | Input/Output | PU | SPI Data Input/ Output. 1.8V LVCMOS logic levels (3.3V tolerant). |
| 29 | CE | Input | PU | Chip Enable. 1.8V LVCMOS logic levels (3.3V tolerant). CE = 0: Power-down mode CE = 1: Normal operation |
| 30 | SYNC | Input | PD | SYNC pin can be used to implement a deterministic delay between the reference input rising edge and the output signal rising edge. If not used, this pin can either be tied to ground, or left floating since it has an internal pulldown. 3.3V LVCMOS input. |
| 31 | nRESET | Input | PU | Chip Reset. 1.8V LVCMOS logic levels (3.3V tolerant). |
| 32 | VREGVCO1 | Analog | | Regulator for VCO. Connect 22μF capacitor from this pin to GND. |
| 33 | VDDVCO1 | Power | | VDD power supply for VCO. |
| 34 | NC | Unused | | Do not connect. |
| 35 | VREFVCO1 | Analog | | Reference node for VCO regulator. Connect 22μF capacitor from this pin to GND. |
| 36 | NC | Unused | | Do not connect. |
| 37 | VDDOUTA | Power | | VDD output power supply for output pair A. |
| 38 | VSSOUTA | Ground | | VSS power supply ground for output pair A. |
| 39 | nRF_OUTA | Output | | Negative side of output pair A (CML – Open Collector). The output power level is programmable. |

Table 1. Pin Descriptions (Cont.)

| Pin Number | Name | Type | Pull-up/ Pull-Down | Description |
|------------|----------|--------|-----------------------|--|
| 40 | RF_OUTA | Output | | Positive side of output pair A (CML – Open Collector). The output power level is programmable. |
| 41 | VSSOUTA | Ground | | VSS power supply ground for output pair A. |
| 42 | NC | Analog | | Do not connect. |
| 43 | NC | Analog | | Do not connect. |
| 44 | VSSOUTB | Ground | | VSS power supply ground for output pair B. |
| 45 | RF_OUTB | Output | | Positive side of output pair B (CML – open collector). The output power level is programmable. |
| 46 | nRF_OUTB | Output | | Negative side of output pair B (CML – open collector). The output power level is programmable. |
| 47 | VSSOUTB | Ground | | VSS power supply ground for output pair B. |
| 48 | VDDOUTB | Power | | VDD output power supply output pair B. |
| EP | EPAD | Ground | | Must be connected to ground. |

Table 2. Pin Characteristics

| Symbol | Parameter | Test Condition | Minimum | Typical | Maximum | Units |
|----------------|--------------------------|-----------------|---------|---------|---------|------------|
| C_{IN} | Input Capacitance | REF_IN, nREF_IN | | 1 | | pF |
| R_{OUT} | LVC MOS Output Impedance | LD | | 15 | | Ω |
| R_{PULLUP} | Input Pullup Resistor | | | 50 | | k Ω |
| $R_{PULLDOWN}$ | Input Pulldown Resistor | | | 50 | | k Ω |

Table 3. Supply Pins and Associated Current Return Paths

| Power Supply | | Associated Ground | |
|--------------|----------|-------------------|----------|
| Pin Number | Pin Name | Pin Number | Pin Name |
| 3 | VDDVCO2 | EP | EPAD |
| 12 | VDDPDCP | 14 | VSSPDCP |
| 15 | VDDFB | 17 | VSSFB |
| 18 | VDDDIG | 19 | VSSDIG |
| 24 | VDDIN | 20 | VSSIN |
| 33 | VDDVCO1 | EP | EPAD |
| 37 | VDDOUTA | 38; 41 | VSSOUTA |
| 48 | VDDOUTB | 44; 47 | VSSOUTB |

Absolute Maximum Ratings

The absolute maximum ratings are stress ratings only. Stresses greater than those listed below can cause permanent damage to the device. Functional operation of the 8V97003 at absolute maximum ratings is not implied. Exposure to absolute maximum rating conditions may affect device reliability.

Table 4. Absolute Maximum Ratings^[a]

| Symbol | Parameter | | Rating | Units |
|---------------------|---|--------------------|----------------------------------|-------|
| VDDx ^[b] | Supply Voltage | | 3.63 | V |
| V _I | Input Voltage (REF_IN, nREF_IN, MUTE, SDIO, VTUNE, SCLK, CSB, CE, nRESET, and SYNC) | | -0.5 to VDDx ^[a] +0.5 | V |
| V _o | Output Voltage (RF_OUTA, nRF_OUTA, RF_OUTB, nRF_OUTB, LD, CPOUT, SDIO, SDO) | | -0.5 to VDDx ^[a] +0.5 | V |
| I ₀ | Output Current | Continuous Current | 60 | mA |
| | | Surge Current | 90 | mA |
| I ₀ | Output Current (SDO, SDIO, LD) | Continuous Current | 40 | mA |
| | | Surge Current | 65 | mA |
| T _J | Maximum Junction Temperature | | 150 | °C |
| T _S | Storage Temperature | | -65 to 150 | °C |
| — | ESD – Human Body Model | | 2000 | V |
| — | ESD – Charged Device Model | | 750 | V |

[a] Over operating ambient temperature range (unless otherwise indicated).

[b] VDDx denotes, VDDVCO2, VDDPDCP, VDDFB, VDDDIG, VDDIN, VDDVCO1, VDDOUTA, VDDOUTB.

Recommended Operating Conditions

Table 5. Recommended Operating Conditions^{[a][b]}

| Symbol | Parameter | Minimum | Typical | Maximum | Units |
|----------------|----------------------------------|---------|---------|---------|-------|
| T _A | Ambient Air Temperature | -40 | | +95 | °C |
| T _B | Board Temperature ^[c] | | | +105 | °C |
| T _J | Junction Temperature | | | +125 | °C |

[a] It is the user's responsibility to ensure that device junction temperature remains below the maximum allowed.

[b] All conditions in this table must be met to guarantee device functionality.

[c] Measured at solder connection to printed circuit board on exposed pad.

Thermal Characteristics and Reliability Information

Table 6. Thermal Characteristics

| Symbol | Parameter | Value | Units |
|---------------|-----------------------------|-------|-------|
| θ_{JB} | Theta JB. Junction to board | 0.76 | °C/W |
| θ_{JC} | Theta JC. Junction to case | 10.33 | °C/W |

Table 7. Thermal Resistance θ_{JA} for 48-VFQFPN, Forced Convection

| θ_{JA} by Velocity | | | | |
|---|------|-------|------|------|
| Air Flow | 0 | 1 | 2 | m/s |
| Multi-Layer PCB, JEDEC Standard Test Boards | 22.3 | 18.84 | 17.3 | °C/W |

DC Electrical Characteristics

Table 8. Power Supply DC Characteristics, VDDx = 3.3V ±5%, T_A = -40°C to +95°C (Not Exceeding Max. Board or Junction Temp.)^{[a][b][c]}

| Symbol | Parameter | Conditions | Minimum | Typical | Maximum | Units |
|---------------------|-------------------------------------|--|---------|---------|---------|-------|
| VDDx ^[d] | Power Supply Voltage | | 3.135 | 3.3 | 3.465 | V |
| IDDx ^[e] | Power Supply Current ^[f] | RF_OUTA, nRF_OUTA – Active RF_OUTB, nRF_OUTB – Muted | | 516 | 590 | mA |
| | | RF_OUTA, nRF_OUTA – Active RF_OUTB, nRF_OUTB – Active | | 566 | 650 | mA |
| | | RF_OUTA, nRF_OUTA – Muted RF_OUTB, nRF_OUTB – Muted | | 464 | 538 | mA |
| I _{VCO} | VCO Supply Current | VCO Frequency = 6GHz | | 225 | 271 | mA |
| — | Power Down Current ^[g] | VCO Frequency = 6GHz, CE = Low, VCO_EN = 1 | | 250 | 302 | mA |
| | | CE = Low, VCO_EN = 0 ^[h] | | 71 | 90 | mA |

[a] RF outputs terminated to 50Ω to VDDOUT[A:B].

[b] Output power set to 0101 (see RF_OUTx_pwr[3:0] in [Table 55](#)).

[c] Over [Recommended Operating Conditions](#) (unless otherwise indicated).

[d] VDDx denotes, VDDVCO2, VDDPDCP, VDDFB, VDDDIG, VDDIN, VDDVCO1, VDDOUTA, VDDOUTB.

[e] IDDX denotes IDDVCO2, IDDPDCP, IDDFB, IDDDIG, IDDIN, IDDVCO1, IDDOUTA, IDDOUTB.

[f] Input Frequency = 122.88MHz, Input Doubler Enabled, Output Frequency = 6GHz; PLL is in Fractional mode.

[g] VCO_EN is located in register 0x28, bit position 0. CE: Chip Enable, pin 29.

[h] Power Down Current with VCO_EN = 0 and CE = Low is independent of the VCO frequency.

Table 9. Typical Current by Power Domain^{[a][b]}

| Pin Name | Pin Number | Condition | Typical | Unit |
|------------------------------------|------------|--|---------|------|
| VDDVCO2, VDDVCO1 ^[c] | 3, 33 | 8GHz VCO frequency | 270 | mA |
| VDDPDCP | 12 | I _{CP} = 8mA | 65 | mA |
| VDDFB | 15 | | 49 | mA |
| VDDDIG | 18 | PFD frequency: 245.76MHz DSM setting: 3rd Order (Fractional mode) | 34 | mA |
| | | PFD frequency: 245.76MHz DSM OFF (Integer Mode) | 23 | mA |
| VDDIN | 24 | Input doubler OFF, MULT OFF, Input divider OFF | 36 | mA |
| VDDOUTA | 37 | RF_OUT Disabled | 10 | mA |
| | | RF_OUTA Enabled; RF Output Power Setting: 0001 (Minimum setting) | 43 | mA |
| | | RF_OUTA Enabled; RF Output Power Setting: 1100 (Maximum setting) | 82 | mA |
| VDDOUTB | 48 | RF_OUTB Disabled | 54 | mA |
| | | RF_OUTB Enabled; RF Output Power Setting: 0001 (Minimum setting) | 88 | mA |
| | | RF_OUTB Enabled; RF Output Power Setting: 1100 (Maximum setting) | 126 | mA |
| | | RF_OUTB Enabled; RF Output Power Setting: 1100; Output doubler ON | 137 | mA |
| | | RF_OUTB Enabled; RF Output Power Setting: 0010; Output Divider: Divide by 2 | 89 | mA |

[a] Operating conditions are: REF_IN = 122.88MHz; RF_OUTA = RF_OUTB = 8GHz.

[b] Over [Recommended Operating Conditions](#) (unless otherwise indicated).

[c] Total current from VDDVCO1 and VDDVCO2 (externally connected).

Table 10. LVC MOS DC Characteristics, $V_{DDx}^{[a]} = 3.3V \pm 5\%$, $T_A = -40^\circ C$ to $+95^\circ C$ (Not Exceeding Max. Board or Junction Temp.)^[b]

| Symbol | Parameter | | Test Conditions | Minimum | Typical | Maximum | Units |
|----------|---------------------|-----------------------------------|-----------------------------------|---------|---------|-----------------|---------|
| V_{IH} | Input High Voltage | nRESET, CE, MUTE, CSB, SCLK, SDIO | | 1.2 | | $V_{DDx}^{[a]}$ | V |
| | | SYNC | | 2 | | $V_{DDx}^{[a]}$ | V |
| V_{IL} | Input Low Voltage | nRESET, CE, MUTE, CSB, SCLK, SDIO | | | | 0.65 | V |
| | | SYNC | | | | 0.8 | |
| I_{IH} | Input High Current | nRESET, CE, SDIO | $V_{DDx} = V_{IN} = 3.465V$ | | | 5 | μA |
| | | CSB, MUTE, SYNC | $V_{DDx} = V_{IN} = 3.465V$ | | | 150 | μA |
| I_{IL} | Input Low Current | nRESET, CE, SDIO | $V_{DDx} = 3.465V, V_{IN} = 0V$ | -150 | | | μA |
| | | CSB, MUTE, SYNC | $V_{DDx} = 3.465V, V_{IN} = 0V$ | -5 | | | μA |
| V_{OH} | Output High Voltage | SDO, SDIO ^[c] , LD | $V_{DDx} = 3.465V, I_{OH} = -2mA$ | 2.4 | | | V |
| V_{OL} | Output Low Voltage | SDO, SDIO ^[c] , LD | $V_{DDx} = 3.465V, I_{OL} = 2mA$ | | | 0.4 | V |

[a] V_{DDx} denotes, V_{DDVCO2} , V_{DDPDCP} , V_{DDFB} , V_{DDDIG} , V_{DDIN} , V_{DDVCO1} , V_{DDOUTA} , V_{DDOUTB} .

[b] Over [Recommended Operating Conditions](#) (unless otherwise indicated).

[c] SDIO as output.

AC Electrical Characteristics

Table 11. AC Characteristics, $V_{DDx}^{[a]} = 3.3V \pm 5\%$, $T_A = -40^\circ C$ to $+95^\circ C$ (Not Exceeding Maximum Board or Junction Temp.)^[b]

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|---------------------|---|---|----------------|---------|-----------------|-------|
| REF_IN | Input Reference Frequency | Reference doubler disabled | 10 | | 1600 | MHz |
| | | Reference doubler enabled | 10 | | 250 | MHz |
| V _{PP} | REF_IN Input Sensitivity | Biased at $V_{DDx}/2^{[c]}$ | 0.4 | | $V_{DDx}^{[a]}$ | V |
| f _{VCO} | VCO Frequency | Fundamental VCO mode | 5,500 | | 11,000 | MHz |
| f _{RF_OUT} | Output Frequency | Output doubler disabled | 171.875 | | 11,000 | MHz |
| | | Output doubler enabled | 11,000 | | 18,000 | MHz |
| f _{PFD} | PFD Frequency | Fractional mode | | | 250 | MHz |
| | | Integer mode | | | 500 | MHz |
| K _{VCO} | VCO Sensitivity ^[d] | VCO Frequency = 5.625GHz | | 100 | | MHz/V |
| | | VCO Frequency = 6.23GHz | | 120 | | |
| | | VCO Frequency = 6.975GHz | | 140 | | |
| | | VCO Frequency = 7.8GHz | | 160 | | |
| | | VCO Frequency = 8.65GHz | | 210 | | |
| | | VCO Frequency = 9.575GHz | | 165 | | |
| | | VCO Frequency = 10.3GHz | | 155 | | |
| | | VCO Frequency = 10.9GHz | | 170 | | |
| t _{LOCK} | PLL Lock Time ^[e] | Time from Low to High of CSB until Low to High of LD | | 1 | | ms |
| | Frequency Lock Time | One frequency to another frequency locking time when VCO calibration is bypassed. | | 70 | | μs |
| - | Single-Ended RF Output Power ^[f] | Muted | | < -80 | | dBm |
| | | RF_OUTn_pwr = 1100 (max) 50-ohm Resistive Loading | RF_OUT = 8GHz | 9.5 | | dBm |
| | | | RF_OUT = 18GHz | -2.5 | | dBm |
| | | RF_OUTn_pwr = 0110 50-ohm Resistive Loading | RF_OUT = 8GHz | 5 | | dBm |
| | | | RF_OUT = 18GHz | -6.5 | | dBm |
| | | RF_OUTn_pwr = 1100 (max) 1nH Inductive Loading | RF_OUT = 8GHz | 12 | | dBm |
| | | RF_OUTn_pwr = 1100 (max) 0.6nH Inductive Loading | RF_OUT = 18GHz | 4 | | dBm |
| - | RF Output Power Variation Across Temperature ^[g] | RF_OUT at 11GHz | | ±2 | | dBm |
| | | RF_OUT from 11GHz to 18GHz | | ±4 | | dBm |
| - | Min/Max VCO Tuning Voltage | | | 0.3/2.2 | | V |

Table 11. AC Characteristics, $V_{DDx}^{[a]} = 3.3V \pm 5\%$, $T_A = -40^\circ C$ to $+95^\circ C$ (Not Exceeding Maximum Board or Junction Temp.)^[b] (Cont.)

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|--------|-----------------------------|---|---------|---------|---------|-------|
| - | Output-to-Output Skew | | | | 5 | ps |
| - | Output-to-Output Skew Drift | Any value of f_{RF_OUT} . Measurement taken from $-40^\circ C$ to $+95^\circ C$. | | 2 | 10 | ° |
| - | Input-to-Output Skew Drift | Any value of f_{RF_OUT} . Measurement taken from $-40^\circ C$ to $+95^\circ C$. | | 9 | | ps |

[a] V_{DDx} denotes V_{DDVCO2} , V_{DDPDCP} , V_{DDFB} , V_{DDDIG} , V_{DDIN} , V_{DDVCO1} , V_{DDOUTA} , V_{DDOUTB} .

[b] Over [Recommended Operating Conditions](#) (unless otherwise indicated).

[c] AC-coupling the reference signal ensures $V_{DDx} / 2$ biasing.

[d] The value depends on VCO frequency.

[e] Band Select/Calibration Resolution = $4x$ (BandSelAcc[1:0]=10), $f_{PFD}/BndSelDiv \sim 100kHz$.

[f] Single-Ended RF Output Power values are based on after de-embedding the trace and cable losses while other output is connected with same length of cable and terminated to 50ohm resistor. For test setup, see [Figure 9](#). For output terminations, see [Figure 26](#).

[g] Output Power setting = 0110b.

Table 12. RF_OUT[A:B] Phase Noise Char., $V_{DDx}^{[a]} = 3.3V \pm 5\%$, $T_A = -40^\circ C$ to $+95^\circ C$ (Not Exceeding Max. Board or Junction Temp.)^[b]

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|---|---|--|---------|---------|---------|--------|
| $t_{jit}(\emptyset)$ | Integrated Phase Noise/Jitter | $f_{RF_OUT} = 6GHz$ Integration Range: 20kHz – 100MHz PFD = 245.76MHz (Fractional Mode) | | -55 | -48.8 | dBc |
| | | | | 66 | | fs RMS |
| | | $f_{RF_OUT} = 6GHz$ Integration Range: 20kHz – 100MHz PFD = 500MHz (Integer Mode) | | -60.6 | -57.4 | dBc |
| | | | | 35 | 50 | fs RMS |
| | | $f_{RF_OUT} = 8GHz$ Integration Range: 20kHz – 100MHz PFD = 500MHz (Integer Mode) | | -57.6 | -55 | dBc |
| | | | | 36.8 | 50 | fs RMS |
| $\Phi_N(10k)$ $\Phi_N(100k)$ $\Phi_N(1M)$ $\Phi_N(10M)$ $\Phi_N(60M)$ | RF Output Phase Noise Performance at 6GHz (Open Loop) | $f_{RF_OUT} = 11GHz$ Integration Range: 20 kHz – 100MHz PFD = 500MHz (Integer Mode) | | -53.3 | -47 | dBc |
| | | | | 44 | 82 | fs RMS |
| | | $f_{RF_OUT} = 18GHz$ Integration Range: 20 kHz – 100MHz PFD = 250kHz (Integer Mode) | | -50.2 | -46.4 | dBc |
| | | | | 38.5 | 58 | fs RMS |
| | | 10kHz offset from carrier | | -74.9 | -72.8 | dBc/Hz |
| | | 100kHz offset from carrier | | -104.8 | -102.7 | dBc/Hz |
| | | 1MHz offset from carrier | | -133.2 | -130.9 | dBc/Hz |
| | | 10MHz offset from carrier | | -153.8 | -152.6 | dBc/Hz |
| | | 60MHz offset from carrier | | -157.8 | -156.8 | dBc/Hz |

Table 12. RF_OUT[A:B] Phase Noise Char., VDDx^[a] = 3.3V ±5%, T_A = -40°C to +95°C (Not Exceeding Max. Board or Junction Temp.)^[b] (Cont.)

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|------------------------|--|--|---------|---------|---------|--------|
| $\Phi_N(10k)$ | RF Output Phase Noise Performance at 8GHz (Open Loop) | 10kHz offset from carrier | | -71.5 | -69.3 | dBc/Hz |
| $\Phi_N(100k)$ | | 100kHz offset from carrier | | -101.9 | -100.6 | dBc/Hz |
| $\Phi_N(1M)$ | | 1MHz offset from carrier | | -130.6 | -128.4 | dBc/Hz |
| $\Phi_N(10M)$ | | 10MHz offset from carrier | | -152.1 | -149.5 | dBc/Hz |
| $\Phi_N(60M)$ | | 60MHz offset from carrier | | -157.7 | -155.7 | dBc/Hz |
| $\Phi_N(10k)$ | RF Output Phase Noise Performance at 11GHz (Open Loop) | 10kHz offset from carrier | | -67.1 | -65.2 | dBc/Hz |
| $\Phi_N(100k)$ | | 100kHz offset from carrier | | -98 | -93.3 | dBc/Hz |
| $\Phi_N(1M)$ | | 1MHz offset from carrier | | -125.1 | -114.2 | dBc/Hz |
| $\Phi_N(10M)$ | | 10MHz offset from carrier | | -144.7 | -133.2 | dBc/Hz |
| $\Phi_N(60M)$ | | 60MHz offset from carrier | | -150.8 | -142.2 | dBc/Hz |
| $\Phi_N(10k)$ | RF Output Phase Noise Performance at 18GHz (Open Loop) | 10kHz offset from carrier | | -64.6 | -62.7 | dBc/Hz |
| $\Phi_N(100k)$ | | 100kHz offset from carrier | | -95.5 | -93.2 | dBc/Hz |
| $\Phi_N(1M)$ | | 1MHz offset from carrier | | -123.4 | -119.9 | dBc/Hz |
| $\Phi_N(10M)$ | | 10MHz offset from carrier | | -144.1 | -140.0 | dBc/Hz |
| $\Phi_N(60M)$ | | 60MHz offset from carrier | | -148.6 | -142.1 | dBc/Hz |
| — | Spurious Signals due to PFD Frequency | f _{PFD} = 245.76MHz; RF_OUT = 7.86432GHz; Integer Mode | | -83.2 | -77.5 | dB |
| | | f _{PFD} = 245.76MHz; RF_OUT = 8GHz; Fractional Mode | | -74.1 | -69.9 | dB |
| $\Phi_N(\text{SYNTH})$ | Normalized Phase Noise Floor | | | -236 | | dBc/Hz |
| $\Phi_N(1/f)$ | Normalized 1/f Noise ^[c] | 10kHz Offset; f _{RF_OUT} = 8GHz | | -132 | | dBc/Hz |
| H2 | VCO second harmonic | f _{VCO} = 8GHz | | -36 | | dB |
| H3 | VCO third harmonic | f _{VCO} = 8GHz | | -33 | | |

[a] VDDx denotes VDDVCO2, VDDPDCP, VDDFB, VDDDIG, VDDIN, VDDVCO1, VDDOUTA, VDDOUTB.

[b] Over recommended operating conditions (unless otherwise indicated).

[c] $N(1/f) = \Phi_N(\text{RF_OUT}) - 10 \log(10\text{kHz} \div f) - 20 \log(f_{\text{RF_OUT}} \div 1\text{GHz})$ where $\Phi_N(1/f)$ is the 1/f noise contribution at a RF_OUT frequency (f_{RF_OUT}) and at a frequency offset f.

Typical Performance Characteristics

Figure 2. Phase Noise at 6GHz (Fractional Mode)

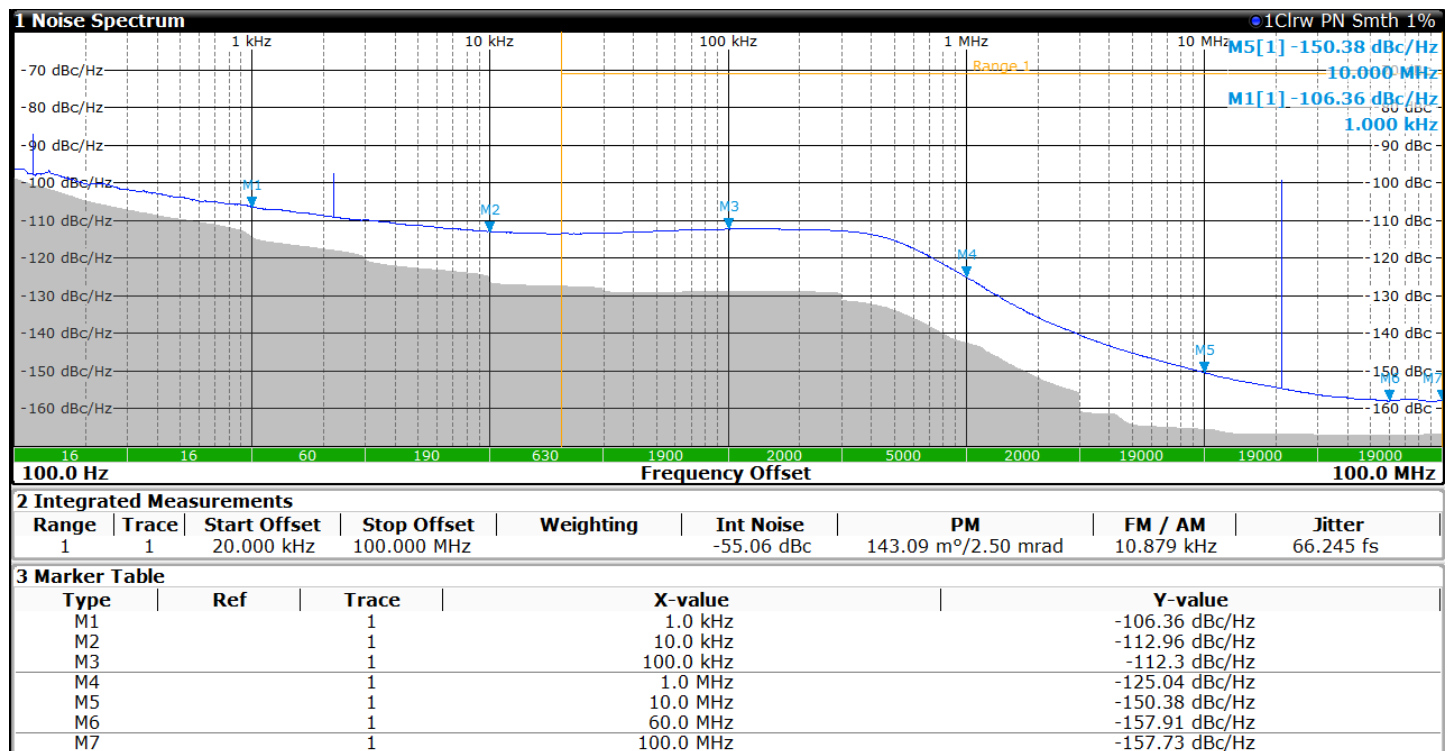


Figure 3. Phase Noise at 6GHz (Integer Mode)

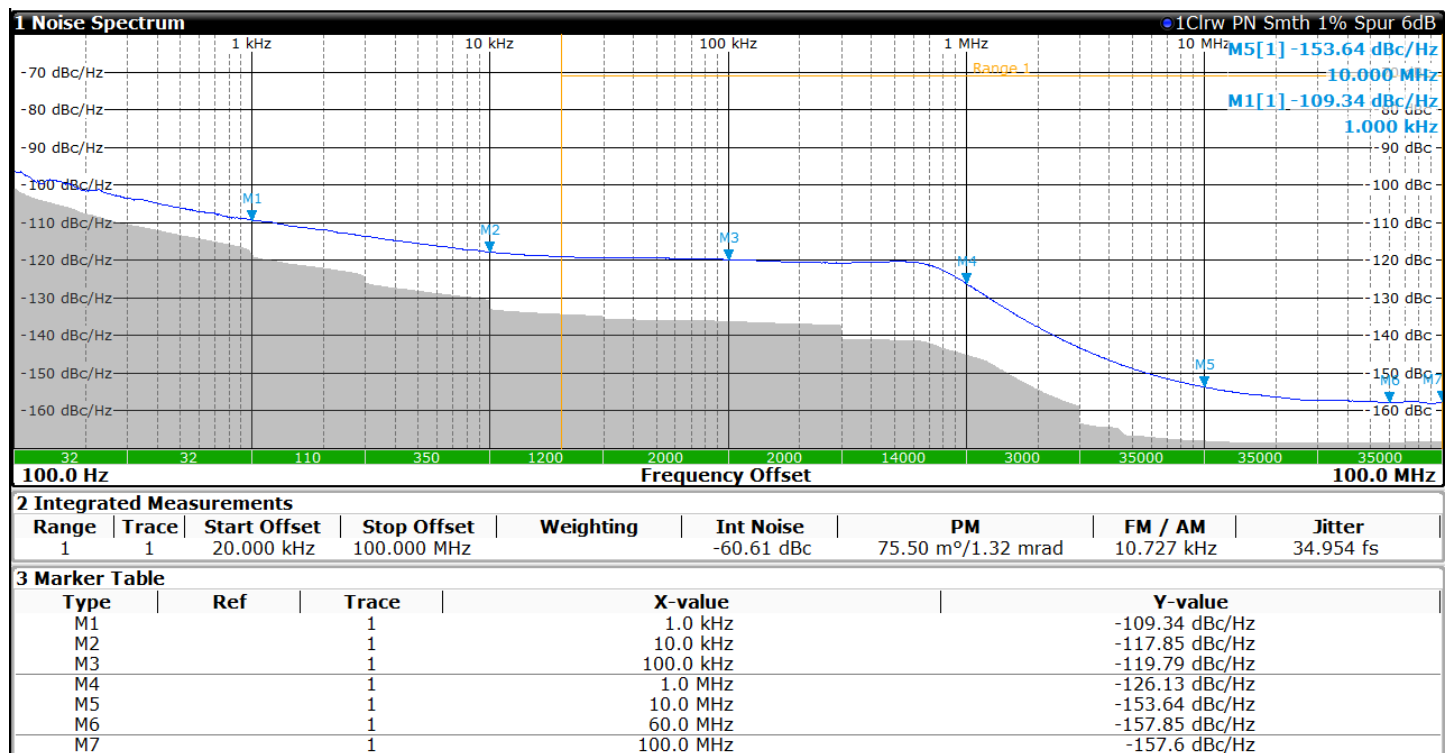


Figure 4. Phase Noise at 8GHz (Integer Mode)

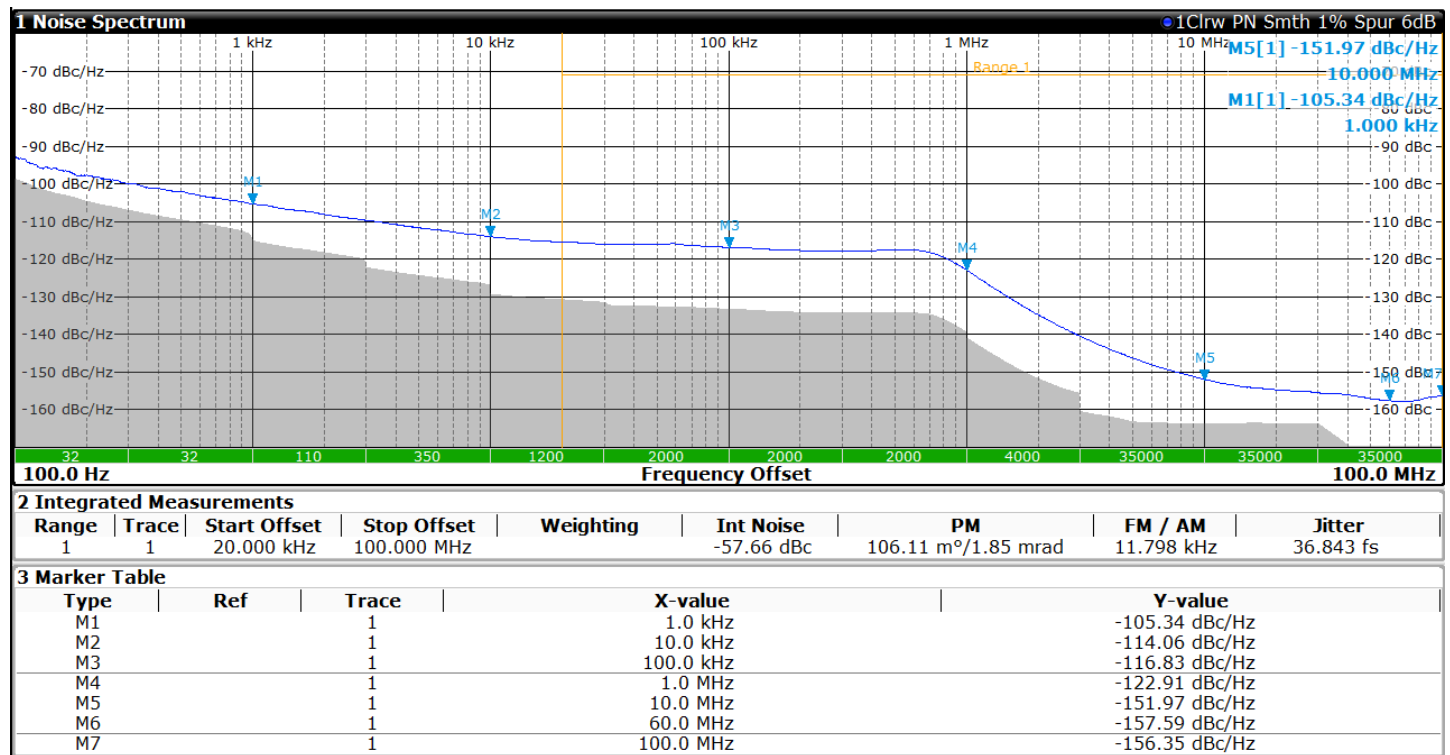


Figure 5. Phase Noise at 11GHz (Integer Mode)

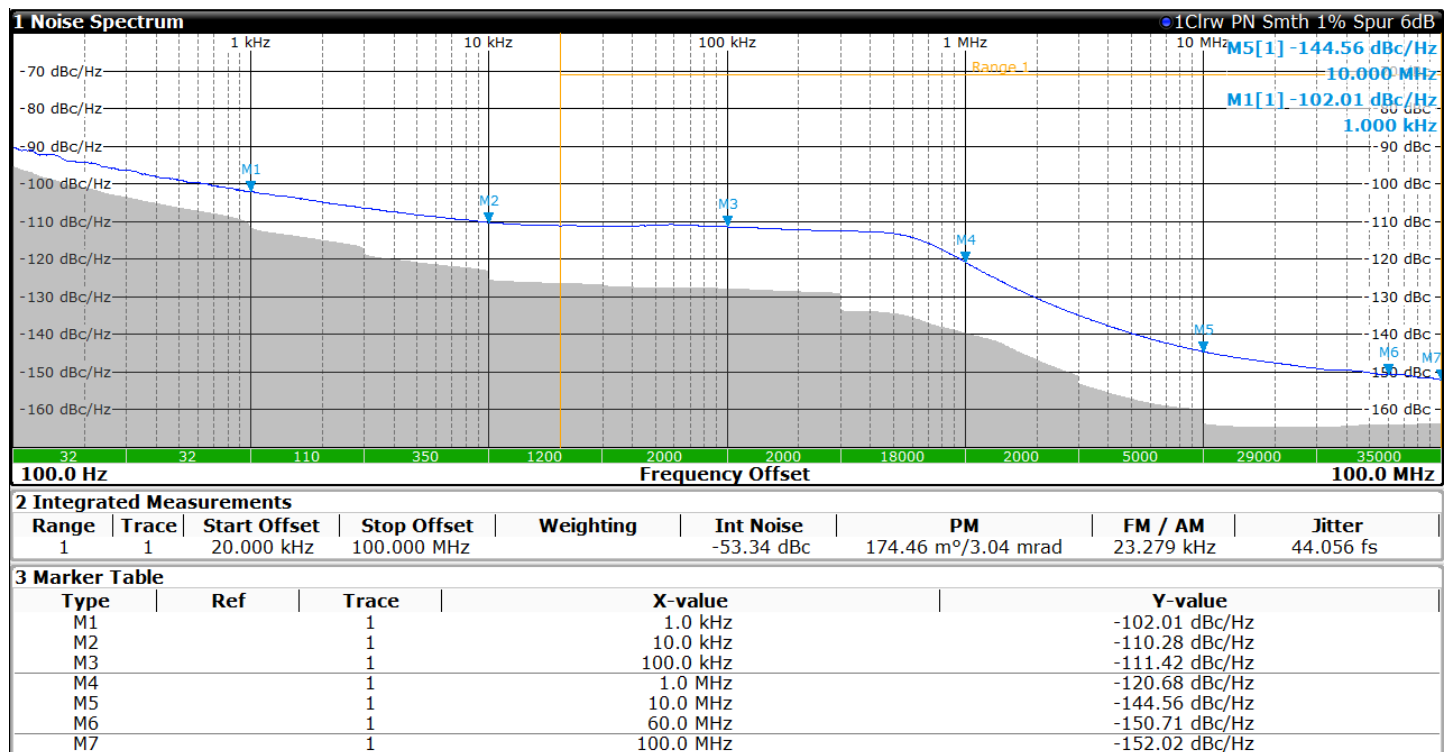


Figure 6. Phase Noise at 18GHz (Integer Mode)

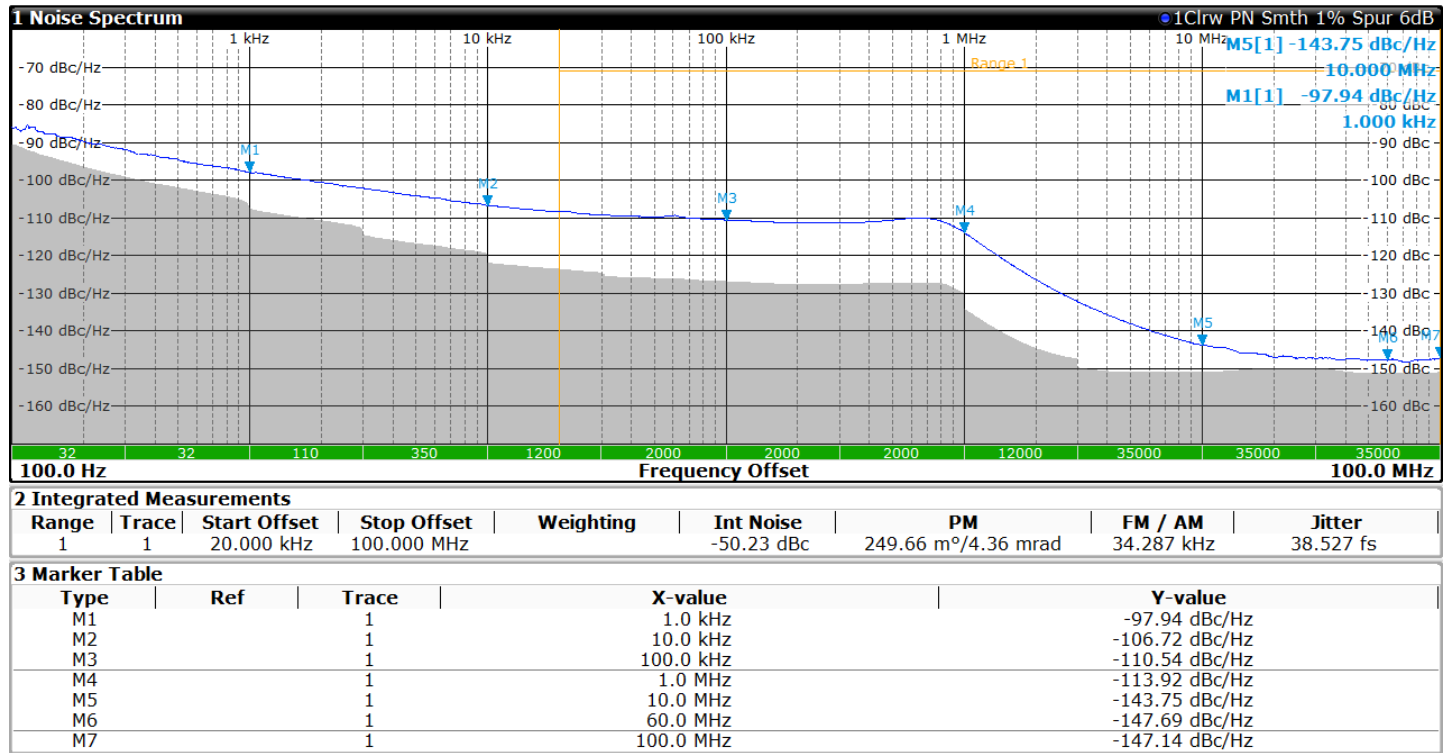


Figure 7. Phase Noise at 8GHz (Open Loop)

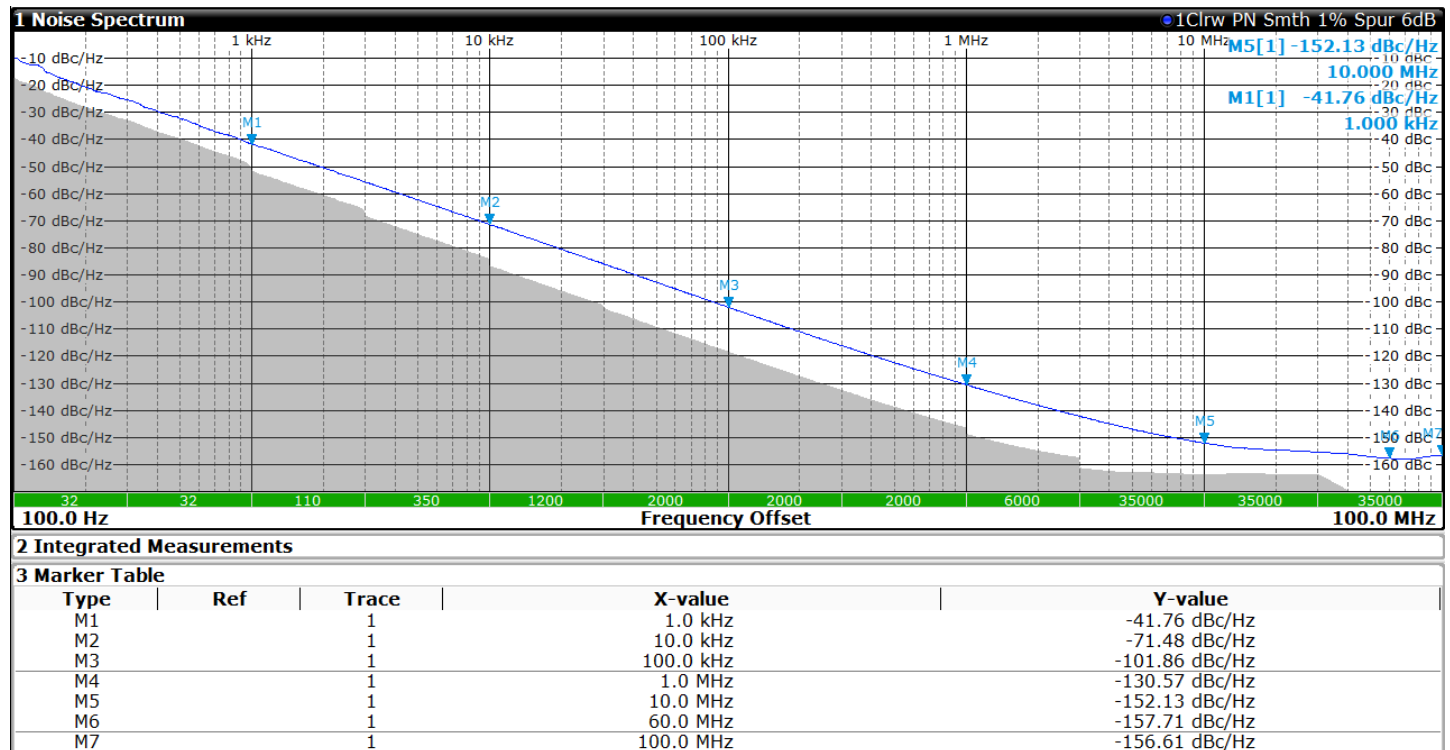


Figure 8. Typical Output Power vs. RF Output Frequency (Over Different Loads)

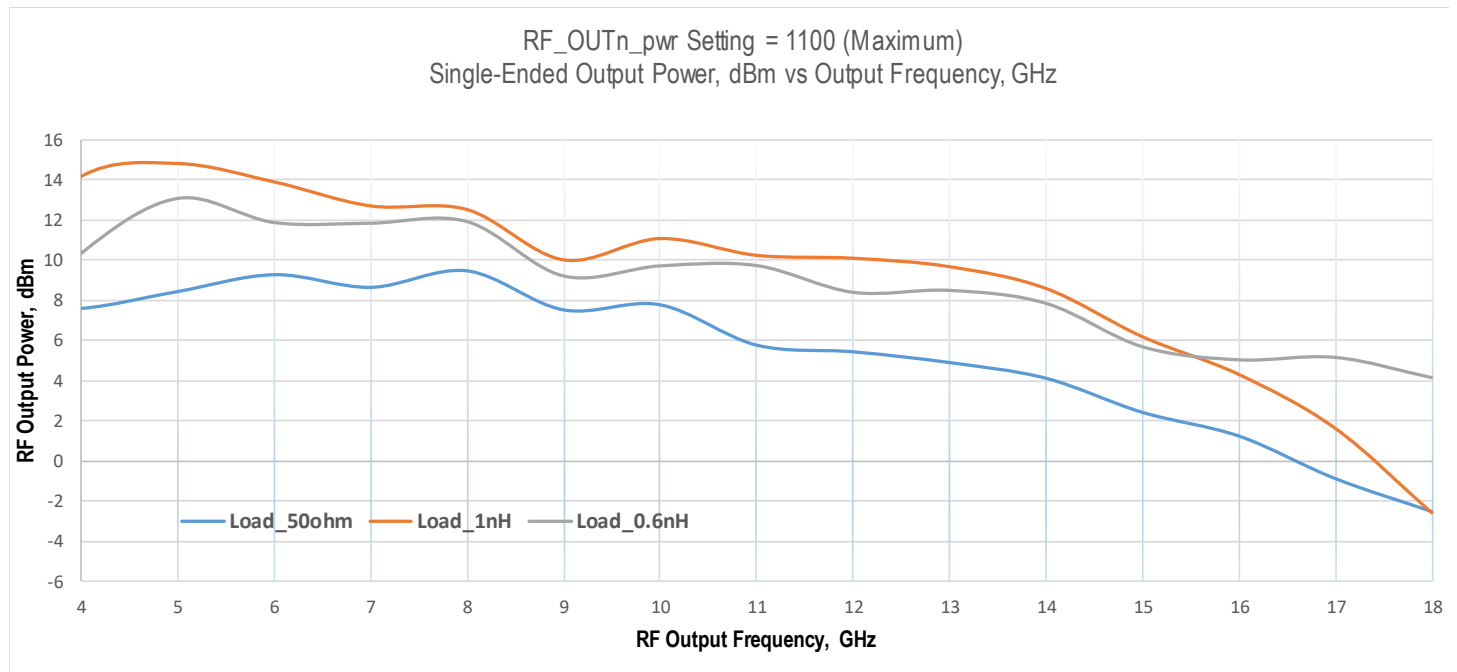
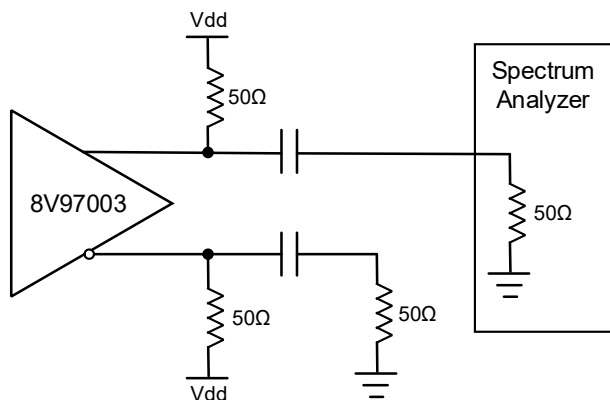


Figure 9. Test Setup for RF Output Power Measurement



Theory of Operation

The 8V97003 is a high-performance frequency synthesizer with an integrated wideband VCO for wide frequency coverage. The VCO provides an octave frequency from 5.5GHz to 11GHz. An optional VCO-doubler is used to generate frequencies larger than the maximum VCO frequency (11GHz) up to 18GHz and an optional output divider can be used to divide the VCO frequency to an output frequency as low as 171.875 MHz. The input reference can support frequencies from 10MHz to 1600MHz. The phase detector (PFD) can support frequencies from 10MHz to 250MHz in fractional mode, and up to 500MHz in Integer mode. A Delta Sigma Modulator (DSM) controls the feedback divider of the PLL in order to create fractional N-divider values. The fractional numerator and the modulus are programmable to 32-bit long, allowing a very fine frequency resolution. The device provides two outputs with individually programmable RF output power (for more information, see [Table 55](#)). The digital logic is a 3- or 4-wire SPI interface that is 1.8V compatible and 3.3V tolerant.

Synthesizer Programming

The Fractional-N divider architecture is implemented via a cascaded programmable dual modulus prescaler, controlled by a DSM. The N divider offers a division ratio in the feedback path of the Phase Lock Loop (PLL), and is given by programming the value of INT, FRAC, and MOD in the following equation:

$$N = \text{INT} + \text{FRAC}/\text{MOD}$$

where:

- *INT* is the divide ratio of the binary 16-bit counter (see [Table 25](#)).
- *FRAC* is the numerator value of the fractional divide ratio. It is programmable from 0 to (MOD – 1) (see [Table 25](#)).
- *MOD* is the 32-bit modulus. It is programmable from 2 to 4,294,967,295 (see [Table 25](#)).

The VCO frequency (f_{VCO}) at RF_OUTA or RF_OUTB is given by the following equation:

$$f_{\text{VCO}} = f_{\text{PFD}} \times (\text{INT} + \text{FRAC}/\text{MOD}) \quad (1)$$

f_{PFD} is the frequency at the input of the Phase and Frequency Detector (PFD).

The 8V97003 supports an Integer mode. It is enabled by programming the FRAC value to 0.

The device's VCO is separated into several frequency bands in order to cover the entire range with sufficient margin for process, voltage, and temperature variations. These are automatically selected by invoking the Autocal feature. The charge pump current is also programmable via the ICP_SETTING register for maximum flexibility.

The [Output Control Registers](#) can be used to enable RF_OUTA, or RF_OUTB, or both outputs.

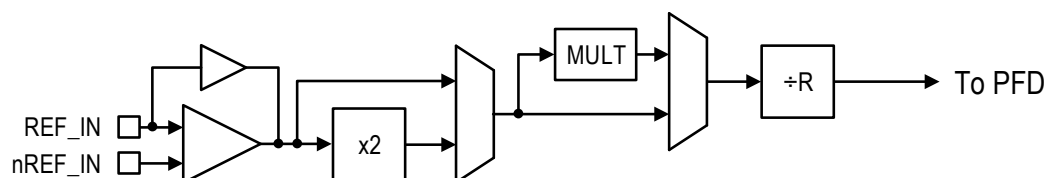
Reference Input Stage

The 8V97003 features one differential reference clock input, REF_IN. This differential input can also be configured as one single-ended input, and it can be driven by an AC-coupled sine wave or square wave.

The input type (Differential or Single-ended) can be programmed via the bit Input_Type in the [Input Control Registers](#). In Power-down mode (set pin CE = 0), this input is set to High-Impedance to prevent loading of the reference source.

The reference input signal path also includes a reference divider (R) and an optional doubler (D), as well as an optional multiplier (MULT) that allows accommodating a higher PFD frequency when the input reference frequency is low (see [Figure 10](#)). Having a high enough PFD frequency is typically better for phase noise performance. However, note that the MULT multiplier also adds its own additive noise. In both cases (doubler and / or multiplier enabled), the maximum PFD frequency is limited to 500MHz in Integer mode and to 250MHz in Fractional mode.

Figure 10. Input Reference Path



$$f_{\text{PFD}} = \text{REFIN} \times \frac{1 + D}{R} \times \text{MULT}$$

- *REFIN* is the input reference (REF_IN) frequency
- *D* is the input reference doubler (0 if not active, or 1 if active)
- *MULT* is the multiplication factor of the input Multiplier "MULT". It is equal to 1 if it is bypassed.

Input Reference Divider (R)

A 10-bit input reference divider (R Divider) is available to scale the input reference frequency to a frequency suitable for the PFD.

Reference Doubler

In order to improve the phase noise performance of the device, the reference doubler can be used. By using the doubler, the PFD frequency is also doubled to allow a more frequent update of the VCO, which may optimize the phase noise performance.

When operating the device in Fractional mode, the speed of the Sigma Delta modulator of the N counter is limited to 250MHz, which is also the maximum PFD frequency that can be used in fractional mode. When the part operates in integer-N mode, the PFD frequency is limited to 500MHz.

Reference Multiplier (MULT)

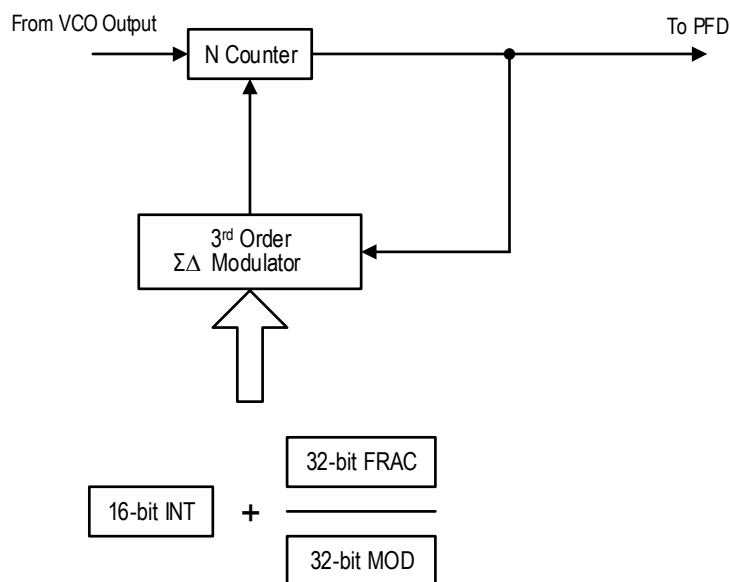
The 8V97003 input path offers an optional frequency multiplier that can multiply the input reference frequency (or the frequency after the optional reference doubler D) to a frequency that must be between 160MHz and 250MHz. That multiplied frequency is used as the PFD frequency. When possible, enabling the doubler is recommended in order to provide a higher input frequency to the MULT multiplier, and thus, optimize its phase noise performance. When it is used, the optional MULT multiplier may degrade the in-band phase noise within the loop bandwidth of the 8V97003.

Note: The input reference multiplier can only output 160MHz to 250MHz before reaching the input reference divider on the input path. Using the input reference multiplier to output a frequency out of that range will not guarantee a lock.

Feedback Divider

The feedback divider N supports fractional division capability in the PLL feedback path. It consists in an integer N divider of 16 bits, and a Fractional divider of 32 bits (FRAC) over 32-bits (MOD).

Figure 11. RF Feedback N Divider



The 16 INT bits (NInt[15:0]) in the [Feedback Divider Control Registers](#) set the integer part of the feedback division ratio. The 32 FRAC bits (Bit NFrac[31:0]) in the registers set the numerator of the fraction that goes into the Sigma Delta modulator. The 32 MOD bits (NMod[31:0]) in the registers set the denominator of the fraction that goes into the Sigma Delta modulator.

From the relation (1), the VCO minimum step frequency is determined by $(1/\text{MOD}) \times f_{\text{PFD}}$.

FRAC values from 0 to (MOD – 1) cover channels over a frequency range equal to the PFD reference frequency. The PFD frequency is calculated as follows:

$$f_{\text{PFD}} = \text{REFIN} \times \frac{1 + D}{R}$$

- *REFIN* is the input reference (REF_IN) frequency
- *D* is the input reference doubler (0 if not active, or 1 if active)
- *R* is the 10-bit programmable input reference pre-divider

This formula assumes that the MULT input multiplier is bypassed.

The programmable modulus (MOD) is determined based on the input reference (REF_IN) frequency and the desired channelization (or output frequency resolution). The high resolution provided on the R counter and the Modulus allows the user to choose from several configuration of the PLL to optimize the performance. The high resolution Modulus also allows the use of the same input reference frequency to achieve different channelization requirements. Using a unique PFD frequency for several needed channelization requirements allows the user to design a loop filter for the different needed setups and ensure the stability of the loop.

The channelization is given by $\frac{f_{\text{PFD}}}{\text{MOD}}$

In low noise mode (dither disabled), the Sigma Delta modulator can generate some fractional spurs that are due to the quantization noise. The spurs are located at regular intervals equal to f_{PFD}/L where *L* is the repeat length of the code sequence in the Sigma Delta modulator. That repeat length depends on the MOD value, as described in [Table 13](#).

Table 13. Fractional Spurs Due to the Quantization Noise

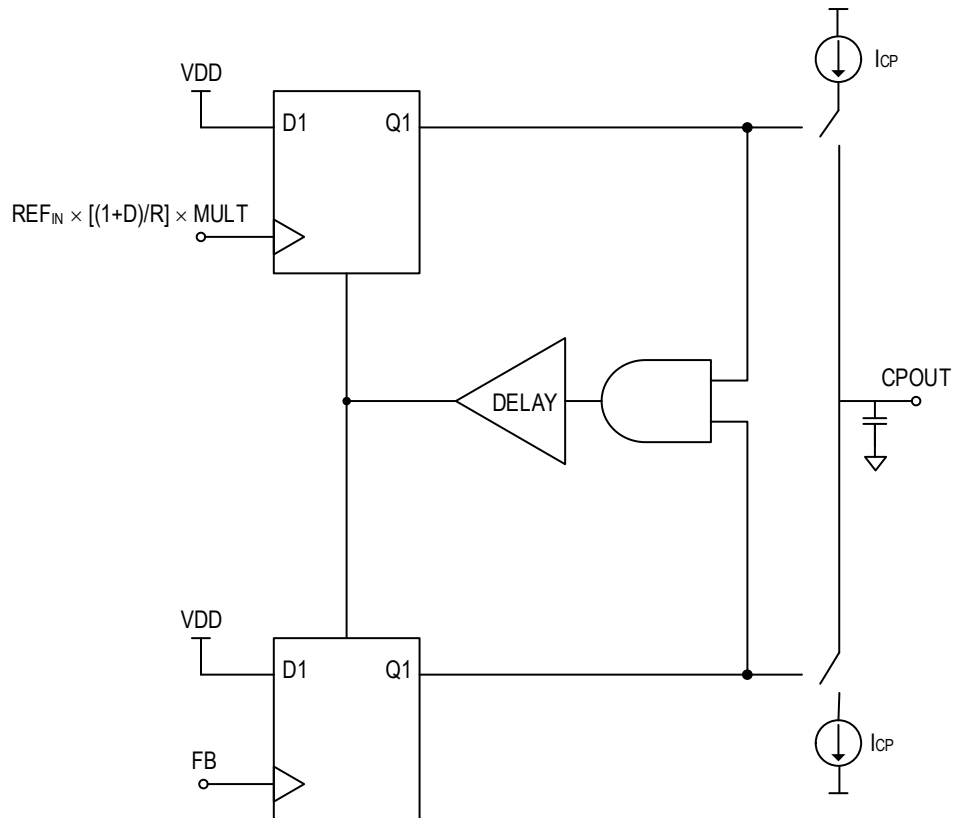
| Condition (Dither Disabled) | L | Spur intervals |
|---------------------------------------|-----------------------|--|
| MOD can be divided by 2, but not by 3 | $2 \times \text{MOD}$ | $f_{\text{PFD}}/(2 \times \text{MOD})$ |
| MOD can be divided by 3, but not by 2 | $3 \times \text{MOD}$ | $f_{\text{PFD}}/(3 \times \text{MOD})$ |
| MOD can be divided by 6 | $6 \times \text{MOD}$ | $f_{\text{PFD}}/(6 \times \text{MOD})$ |
| Other conditions | MOD | $f_{\text{PFD}}/\text{MOD}$ (channel step) |

In order to reduce the spurs, the user can enable the dither function to increase the repeat length of the code sequence in the Sigma Delta Modulator. The increased repeat length is $2^{32} - 1$ cycles so that the resulting quantization error is spread to appear like broadband noise. As a result, the in-band phase noise may be degraded when using the dither function. When the application requires the lowest possible phase noise and when the loop bandwidth is low enough to filter most of the undesirable spurs, or if the spurs will not affect the system performance, it is recommended to use the low noise mode with dither disabled.

Phase and Frequency Detector (PFD) and Charge Pump

The phase detector compares the outputs from the R counter and N counter, and generates an output corresponding to the phase and frequency difference between the two inputs at the PFD. The charge pump current is programmable through the serial port (SPI) to several different levels (see [Table 49](#)).

Figure 12. Simplified PFD Circuit using D-type Flip-Flop



PFD Frequency

The operating frequency for the PFD is up to 500MHz when the device operates in integer mode, and up to 250MHz when the device operates in fractional mode.

External Loop Filter

The 8V97003 requires an external loop filter. The design of that filter is application specific. For more information, see [Applications Information](#).

Charge Pump High-Impedance

In order to put the charge pump into three-state mode, the user must set the bits CP_HiZ (Bit D5) to 1 in Register 47 in the [Charge Pump Control Registers](#). This bit should be set to 0 for normal operation.

Integrated Low Noise VCO

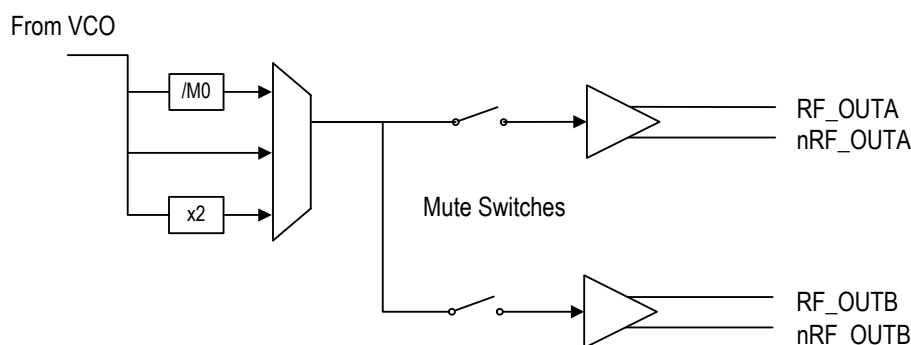
The VCO used in the 8V97003 is divided into several frequency bands. This allows for a lower VCO sensitivity (K_{vco}), which results in the best possible VCO phase noise and spurious performance.

The user does not have to select the different VCO bands. The VCO band select logic of the 8V97003 will automatically select the most suitable band of operation at power up or after programming. In addition, the Force_Relock bit (register 33, bit D7) can be used to automatically select the best frequency band.

Output Clock Distribution and Optional Output Doubler

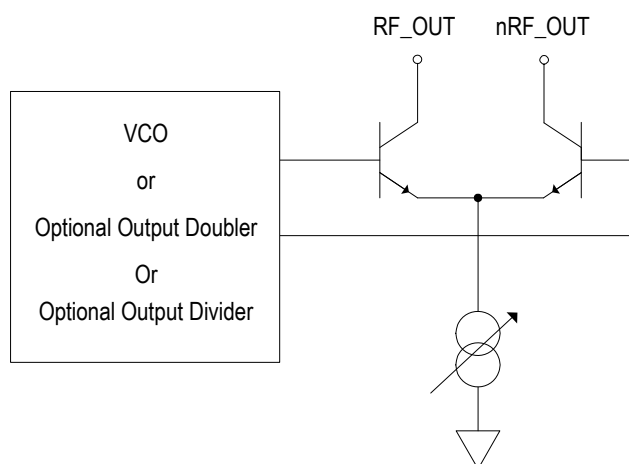
The 8V97003 device provides two differential outputs. Either of the two outputs generates a frequency equal to f_{VCO} when bypassing the optional output doubler and the optional output divider M0, or to $2 \times f_{VCO}$ (up to 18GHz) when using the optional output doubler, or an integer division of the VCO frequency f_{VCO} . The division ratios of the output divider are provided in the [Output Control Registers](#).

Figure 13. Output Clock Distribution



RF_OUT and nRF_OUT are derived from the collector of an NPN differential pair driven by the VCO output (or the output doubler), as displayed in [Figure 14](#).

Figure 14. Output Stage



The 8V97003 offers 4 bits of programmability for the RF output power of each output. The user can configure the RF output power to multiple available settings (see [RF Output Power](#)).

If the auxiliary output (RF_OUTB) is not used, it can be powered down by using the QB_ena bit in the [Output Control Registers](#).

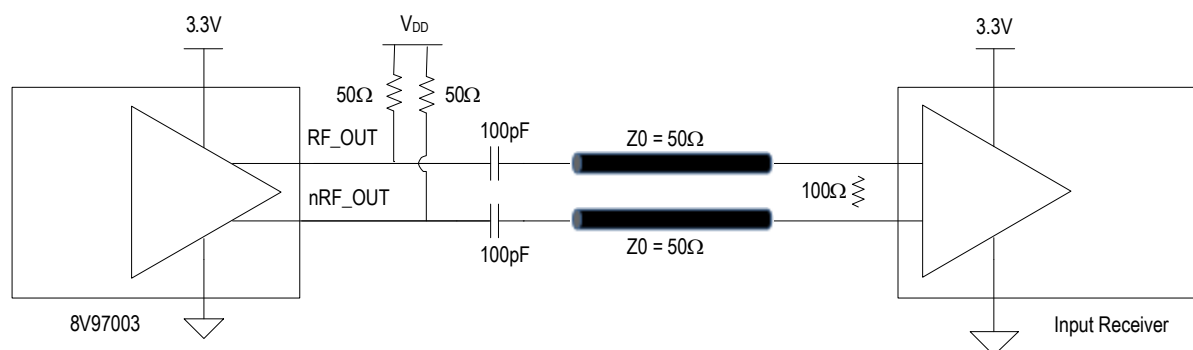
The outputs can be disabled until the part achieves lock. To enable this mode, the user will set the Mute_until_LD bit in the Outputs Control Registers (see [Output Control Registers](#)). The MUTE pin can be used to mute all outputs and be used as a similar function.

Output Matching

The outputs of the 8V97003 are “open collector outputs” and can be matched in different ways.

A simple resistive matching is used to terminate the open collector RF_OUT output with a 50Ω to VDD, and with an AC coupling capacitor in series. Two termination scheme examples are shown in [Figure 15](#) and [Figure 16](#). When considering the frequencies involved from the 8V97003, an inductively loaded configuration is recommended for better performance and optimal power transmission, and to minimize the distortion of the output signal. The resistive matching termination is not ideal to achieve maximum output power transmission, especially for high frequencies.

Figure 15. Resistive Matching Termination

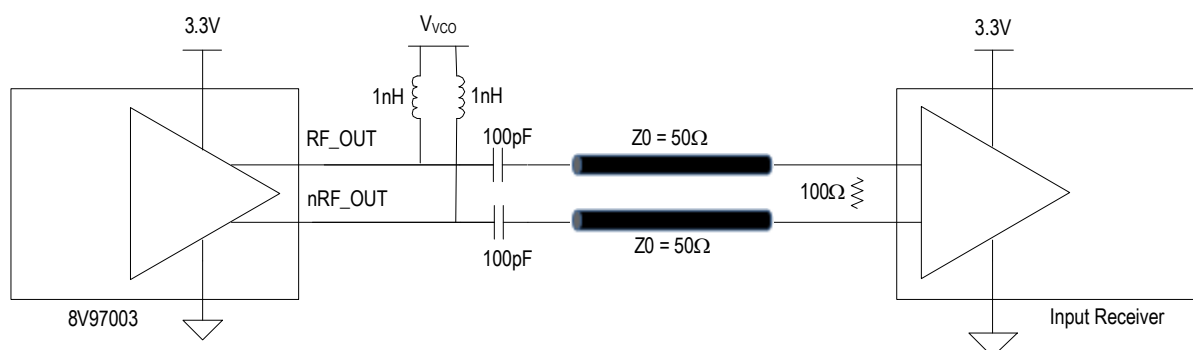


This termination scheme provides one of the selected output powers on the differential pair when connected to a 50Ω load. For additional information about the output power selection, see [RF Output Power](#).

The 50Ω resistor connected to VDD can also be replaced by a choke to provide better performance and optimal power transmission.

The pull-up inductor value is frequency dependent. For impedance matching of 50Ω, the inductance value can be calculated as $L = 50 \div (2\pi f)$, where “f” is the operating frequency. In this example, L = 1nH is for an operating frequency of approximately 6GHz.

Figure 16. Inductively Loaded Termination



For more recommendations on the termination scheme, see [Applications Information](#).

Band Selection Disable

For a given frequency, the output phase can be adjusted when using the BandSelDisable bit (Bit D5 in Register 33; see [Calibration Control Registers](#)). When this bit is enabled (Bit D5 set to 1), the device does not complete a VCO band selection after changing the settings. When the Band_Sel_Disable bit is set to 0, and when the settings are updated, the device proceeds to a VCO band selection. The Band_Sel_Disable bit is useful when the user wants to make small changes in the output frequency (< 1MHz from the nominal frequency) without recalibrating the VCO and minimizing the settling time.

Phase Adjust

The 8V97003 supports adjusting the phase delay between the outputs (RF_OUT/ nRF_OUT) and the input (REF_IN) of the device by shifting the output phase by a fraction of the size of the fractional denominator, when the device is used in fractional mode.

Writing to the [Phase Adjustments Control Registers](#) triggers a phase shift (see [Table 28](#)). The phase adjustment value set by the bits, Phase[31:0], should be less than the fractional-N denominator register, MOD. The actual phase shift can be obtained with the following equation:

$$\text{Phase adjustment (degrees)} = 360 \times \frac{\text{Phase}}{\text{MOD}} \times \frac{T_{\text{VCO}}}{T_{\text{OUT}}}$$

or

$$\text{Phase adjustment (ns)} = T_{\text{VCO}} \times \left(\frac{\text{Phase}}{\text{MOD}} \right)$$

360° represents one cycle of output clock, T_{VCO} is the period of VCO (in ns), T_{OUT} is the period at the output of the 8V97003, and Phase is a programmable value, the same bit length as MOD.

RF Output Power

Each output buffer RF_OUTA and RF_OUTB offers a configurable RF output power. The RF output power can be programmed via the bits RF_OUTA_pwr[3:0] and RF_OUTB_pwr[3:0] in the [Output Control Registers](#).

Output Phase Synchronization

Input-to-Output

The device input-to-output phase relations is deterministic with a fixed phase offset when the PLL feedback divider is integer and the output divider is not used. The phase offset remains the same across power cycles.

Output Phases of Multiple 8V97003 Devices

The output phases of multiple devices can be aligned on the rising signal edge. This is supported for devices with identical configurations and the same input phase and frequency. The devices can use any setting of the input divider, input multiplier, integer and fractional feedback divider, the output frequency doubler, and the 1x frequency output path, but not the output divider.

Phase alignment across devices is established automatically when the PLL feedback divider is integer and the input frequency divider is not used.

In other configurations, an external pulse to the SYNC input establishes an output phase alignment. The SYNC pulse can be applied to the device at any time after the configuration is loaded with the AutoReSync register bit is set to 1 and the PLL is locked. Internal to the device, this synchronization procedure first resets the input divider and then resets the DSM. For applicable input SYNC and REF_IN timing requirements, see the following table.

Table 14. Timing Requirements

| Symbol | Parameter | Test Conditions | Minimum | Maximum | Unit |
|--------|------------|---|---------|---------|------|
| t_S | Setup Time | Rising edge of SYNC pulse to Rising Edge of REF_IN | 0.5 | | ns |
| t_H | Hold Time | Rising Edge of REF_IN to Falling Edge of SYNC Pulse | 0.5 | | ns |

Power-Down Mode

When power-down is activated, the following events occur:

1. VCO is not powered-down
2. RF_OUT buffers are disabled
3. The input stage is powered down and set to High-Impedance
4. Input registers remain active and capable of loading and latching data
5. The CE pin is set to low level (logic zero) for activating power-down mode. More power-down control bits are available in register 0x0028.

Default Power-Up Conditions

All the RF outputs are muted at power-up. For default values in registers, see [Register Map](#).

VCO Calibration

For proper VCO calibration, the 8V97003 must be programmed with the following recommended settings:

- The band select clock divider (Bits BndSelDiv[12:0] in the [Band Select Clock Divider Control Registers](#)) must be set to divide down the PFD frequency in between 50kHz to 100kHz (PFD Frequency/BandSelDiv[12:0] \leq 100kHz and $>$ 50kHz).
- BandSelAcc[1:0] bits must be set to 10 or 11.

3- or 4-Wire SPI Interface Description

The 8V97003 has a selectable 3/4-wire serial control port that can respond as a slave in an SPI configuration to allow read and write access to any of the internal registers for device programming or read back. The SPI interface consists of SCLK (clock), SDIO (serial data input and output in 3-wire mode, input in 4-wire mode), SDO (output in 4-wire mode), and CSB (chip select). A data transfer contains 16-bit instructions (direction +15 bit address) and any integer multiple of 8 bits data. Internal register data is organized in 8-bit byte.

3/4-Wire Mode

The 3- or 4-wire mode is defined by the SDO Active bit in the device configuration register 0x00 bit3 and bit4. If both bits are set to 0, the device is in a 3-wire mode and the SDIO pin is a bi-directional data input/ output, and the SDO pin is in high-impedance. Otherwise, the device is in a 4-wire mode, the SDIO pin is the data input, and the SDO pin is the data output.

Active Clock Edge

SDIO is always clocked-in on the rising edge of SCLK. SDIO (or SDO if in 4-wire mode) is always clocked-out on the falling edge of SCLK.

Reset

After power-up or reset by the nRESET pin, the SPI engine is reset and all internal registers reset to their default values. The SPI interface is in 3-wire mode with SDO in high-impedance, MSB-first mode, and address is in auto-decrement mode.

The function of SoftReset bit in register 0x00 bit7 and bit0 is similar to the nRESET pin. It resets all the registers to their default values, except registers 0x00 and 0x01.

Least Significant Bit Position

The 8V97003 supports both the least significant bit first and most significant bit first transfers.

- If LSBFirst in register 0x00 bit6 and bit1 is set to 0, data is transferred in this order: transfer direction bit, the address bits A14 to A0, then first data byte D7 to D0, 2nd data byte D7 to D0, and so on until CSB is set to 1.
- If LSBFirst is set to 1, the order is: address bits A0 to A14, then the transfer direction bit, then the first data byte D0 to D7, 2nd data byte D0 to D7, and so on until CSB is set to 1.

By default, LSBFirst is set to 0.

Addressing

The 8V97003 implements registers at the addresses from 0x00 to 0x49. The addressing mode is 15-bit.

During transferring operation, address increments automatically if AddressAcend in register 0x00 bit5 and bit2 is set to 1; otherwise it decrements. In incrementing mode, if address reaches 0x49, it wrap-around to 0x00. In decrementing mode, if address reaches 0x00, it wrap-around to 0x49.

By default, decrementing mode is set.

Read Operation

A SPI operation starts when there is a high to low transition on CSB, and stops when there is a low to high transition on CSB. If the transfer direction bit R/nW is 1, it is a read operation; otherwise it is a write operation.

This device supports multi-byte read or write operations. Bits A14 to A0 refer to the register address. The device reads or writes data to this address and continues as long as CSB is held at low. The device automatically increments or decrements the address depending on AddressAcend bit.

Figure 17 to Figure 22 show the operation protocols.

Mirrored Register Bits

In register 0x00, bits D7–D4 are mirrored with the bits D3–D0. The mirrored bits pair must set to the same value.

Double-Buffered Registers

Configuration registers that are wider than 8 bits are double-buffered for synchronous access. For these registers, it is required to write the multiple-byte setting into the buffered registers first. The new configurations will not take effect until writing 1 to the TransferOn bit in register 0x0F bit0 to transfer them from the buffered registers to the active registers. TransferOn bit is self-clearing. Multiple-byte configuration data can be read-back either from buffered registers or active registers as specified by the BufferReadMode bit in register 0x01 bit5.

Register 0x10 to 0x1D, 0x22 to 0x25, and 0x29 to 0x2C are double buffered.

Operation Protocols

Figure 17. 4-Wire MSB First, Single Byte Write and Read

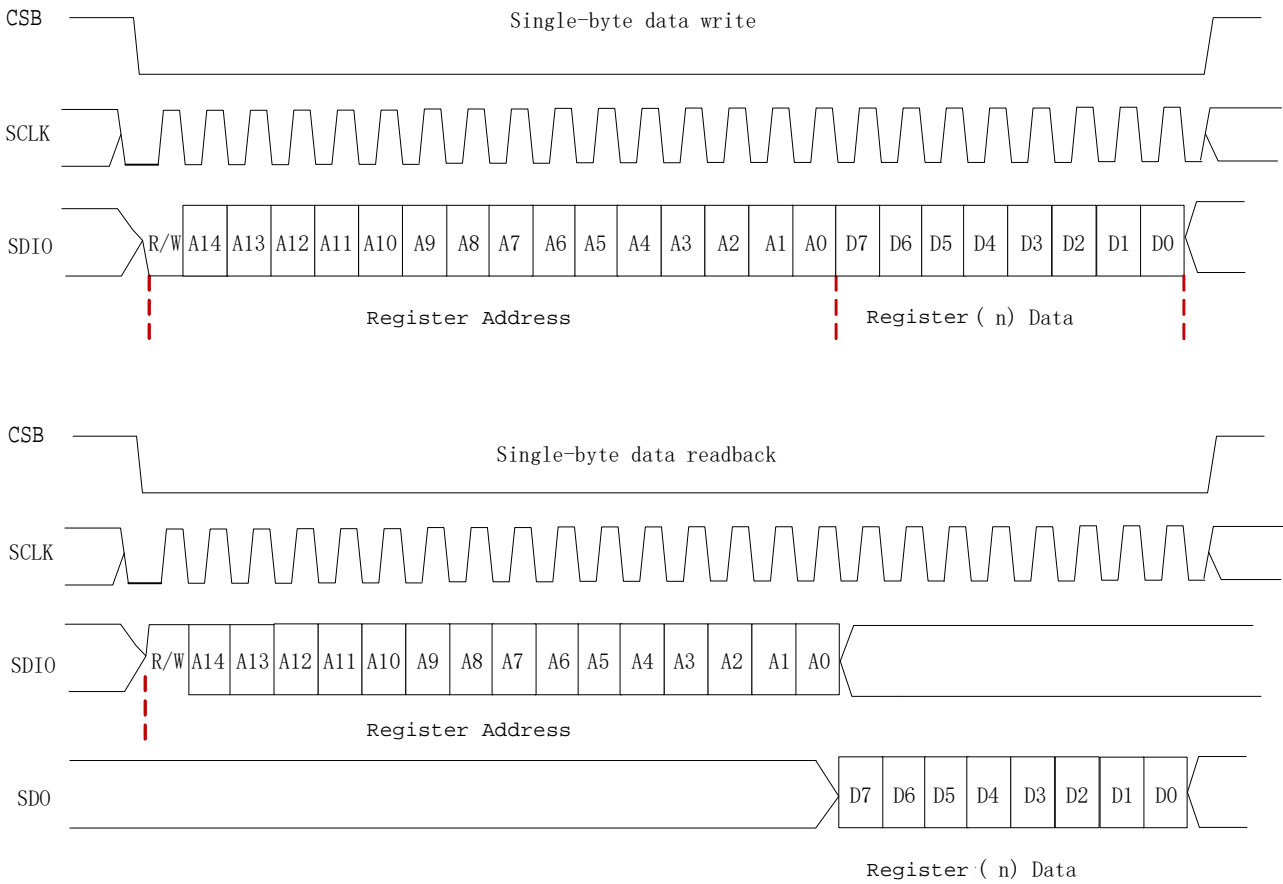


Figure 18. 4-Wire LSB First, Single Byte Write and Read

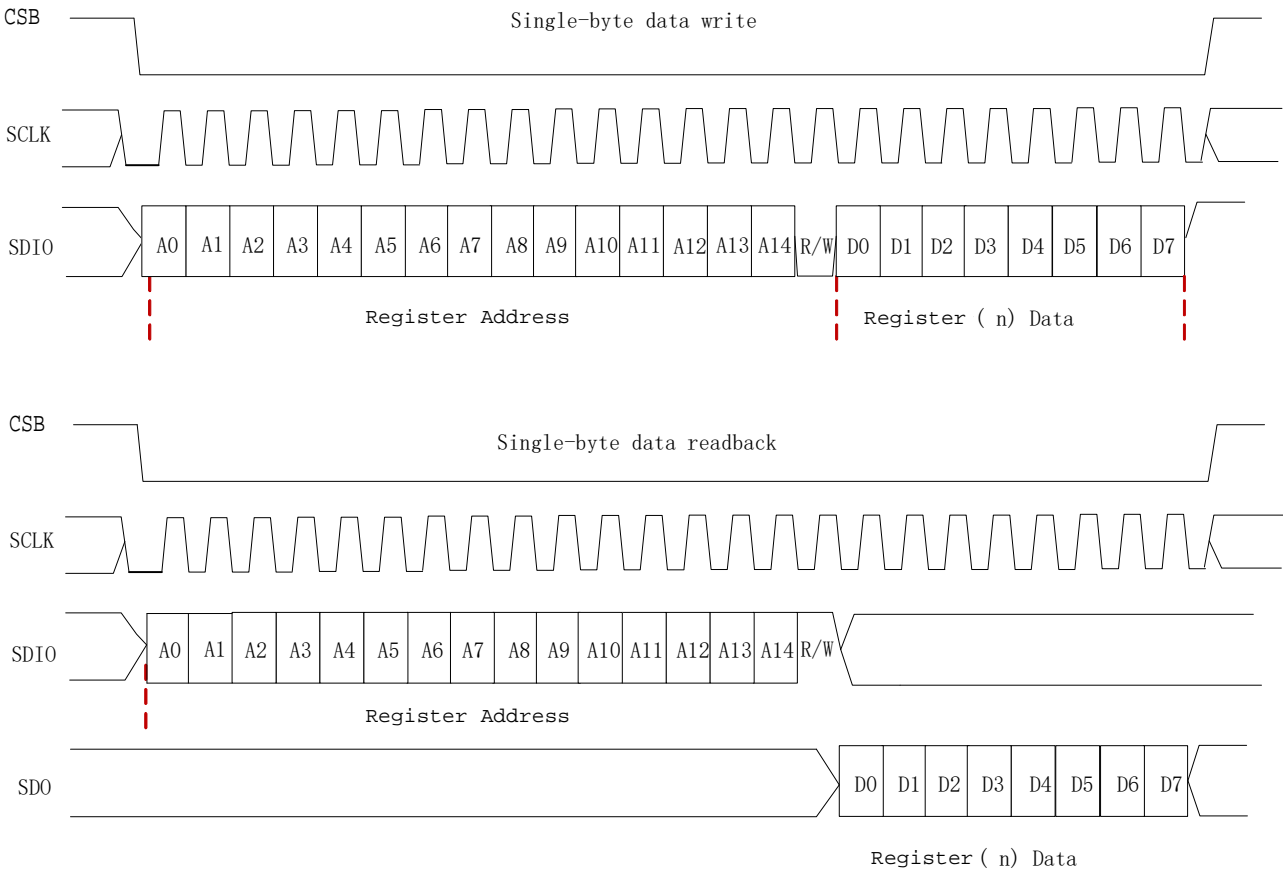


Figure 19. 4-Wire MSB First, Multiple Bytes Write (2 Bytes Shown as Example)

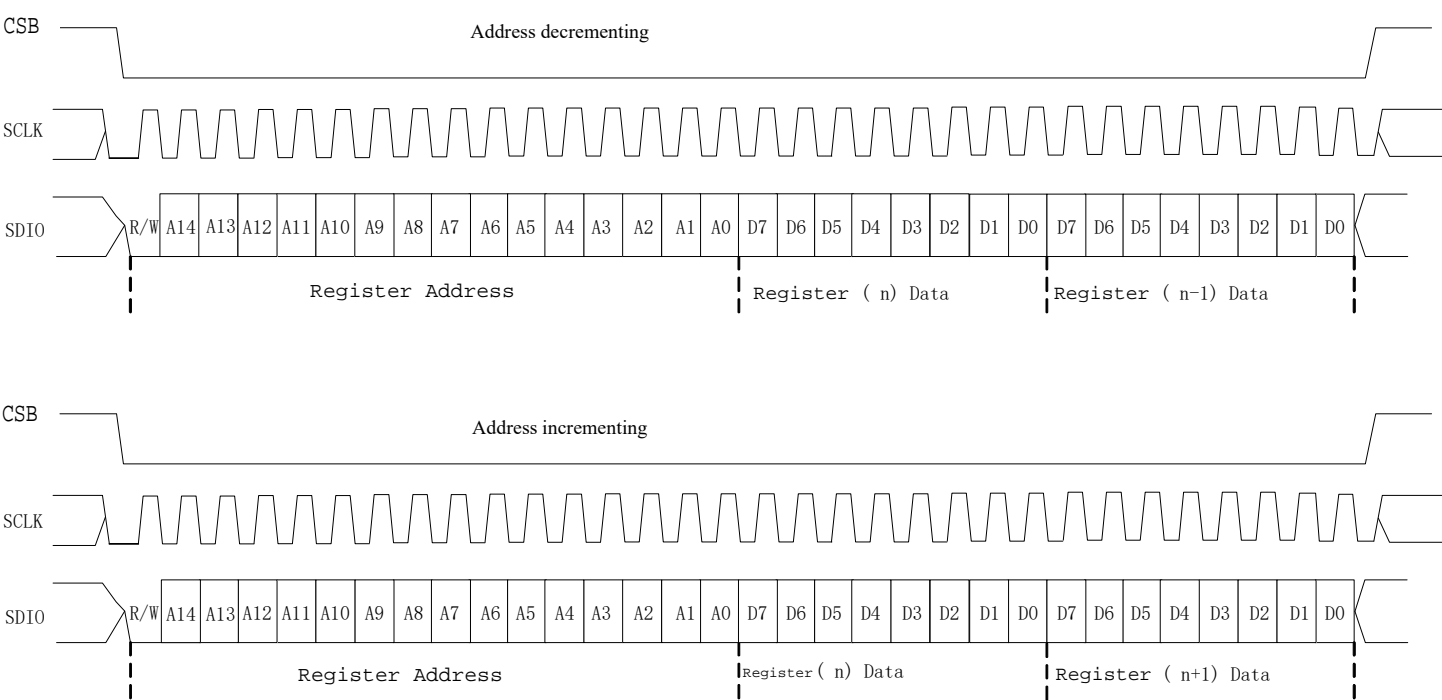


Figure 20. 4-Wire LSB First, Multiple Bytes Read (2 Bytes Shown as Example)

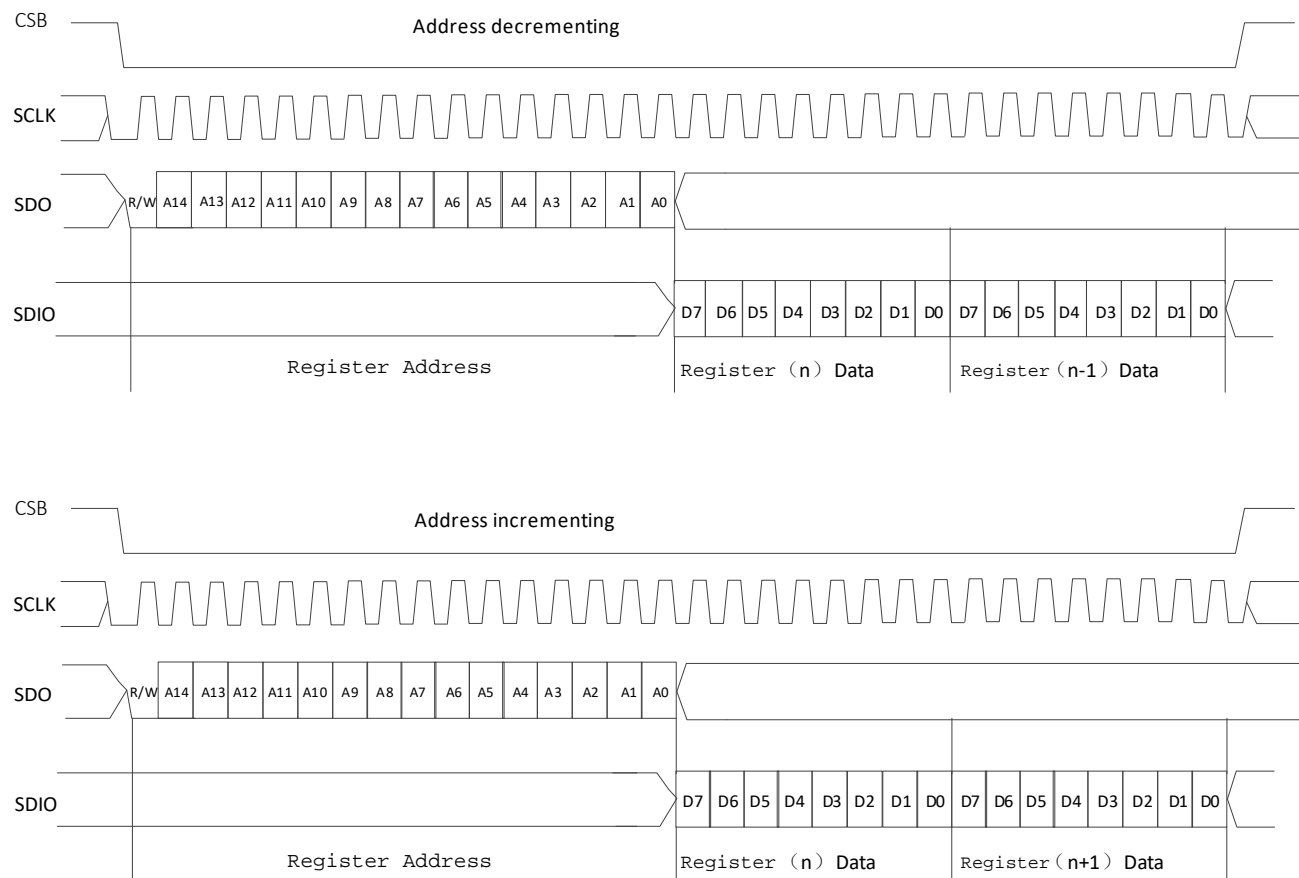


Figure 21. 3-Wire MSB First, Single Byte Read and Write

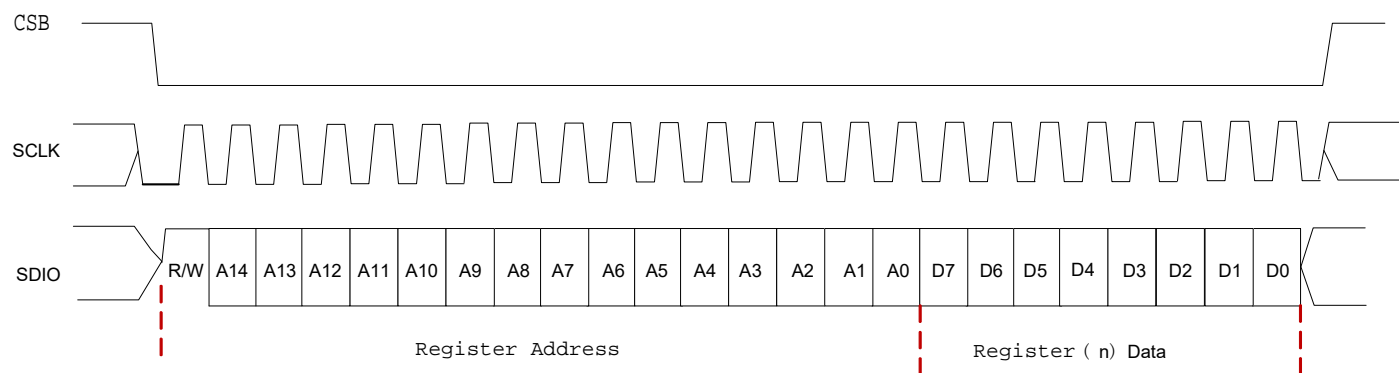
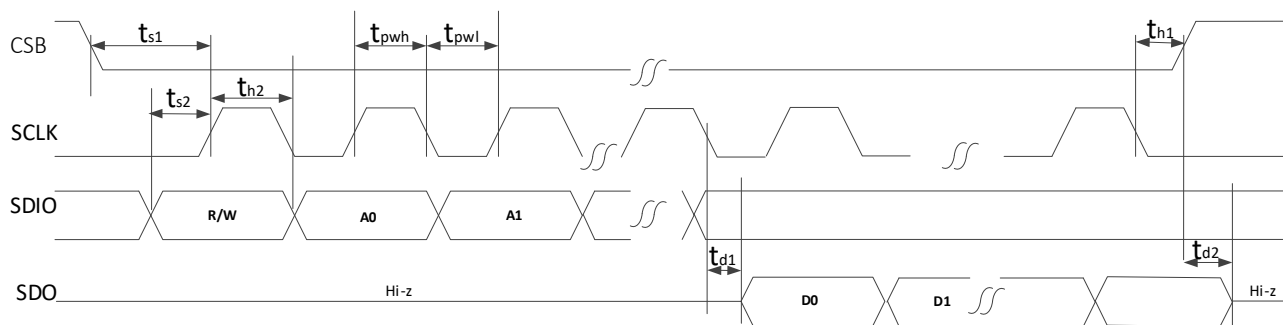


Figure 22. SPI Timing Diagram

Table 15. SPI Read / Write Cycle Timing Parameters

| | Parameter | Minimum | Maximum | Units |
|-------------------|--|---------|---------|-------|
| f_{SCLK} | SCLK Frequency | | 20 | MHz |
| t_{pwh} | SCLK High Pulse Width | 25 | | ns |
| t_{pwl} | SCLK Low Pulse Width | 25 | | ns |
| t_{s1} | Setup Time, CSB falling to SCLK rising | 10 | | ns |
| t_{h1} | Hold Time, SCLK falling to CSB rising | 30 | | ns |
| t_{s2} | Setup Time, SDIO input edge to SCLK rising | 8 | | ns |
| t_{h2} | Hold Time, SCLK rising to SDIO input edge | 8 | | ns |
| t_{d1} | SCLK Falling Edge to valid readback data on SDIO if in 3-wire mode, or SDO if in 4-wire mode | | 10 | ns |
| t_{d2} | CSB Rising Edge to High-Impedance on SDIO if in 3-wire mode, or SDO if in 4-wire mode | | 10 | ns |

Register Map

Table 16. Preface Registers

| Register | Addr | Type | Default Value | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|----------|------|------|---------------|--------------------|--------------------|--------------------|--------------------|--------------------|--------------------|--------------------|--------------------|
| 0 | 0000 | R/W | 0000_0000 | SoftReset | LSBFirst | AddressAs cend | SDOActive | SDOActive | AddressAs cend | LSBFirst | SoftReset |
| 1 | 0001 | R/W | 0000_0000 | Unused | Unused | BufferRead Mode | Unused | Unused | Unused | Unused | Unused |
| 2 | 0002 | R/W | 0000_0000 | Unused | Unused | Unused | Unused | Unused | Unused | Unused | Unused |
| 3 | 0003 | R | 0000_0110 | ChipType<7> | ChipType<6> | ChipType<5> | ChipType<4> | ChipType<3> | ChipType<2> | ChipType<1> | ChipType<0> |
| 4 | 0004 | R | 0000_0000 | ChiplD<7> | ChiplD<6> | ChiplD<5> | ChiplD<4> | ChiplD<3> | ChiplD<2> | ChiplD<1> | ChiplD<0> |
| 5 | 0005 | R | 0000_0000 | ChiplD<15> | ChiplD<14> | ChiplD<13> | ChiplD<12> | ChiplD<11> | ChiplD<10> | ChiplD<9> | ChiplD<8> |
| 6 | 0006 | R | 0000_0000 | ChipVersion <7> | ChipVersion <6> | ChipVersion <5> | ChipVersion <4> | ChipVersion <3> | ChipVersion <2> | ChipVersion <1> | ChipVersion <0> |
| 7 | 0007 | R | 0001_0001 | ChipOption <7> | ChipOption <6> | ChipOption <5> | ChipOption <4> | ChipOption <3> | ChipOption <2> | ChipOption <1> | ChipOption <0> |
| 8 | 0008 | R/W | 0000_0000 | Unused | Unused | Unused | Unused | Unused | Unused | Unused | Unused |
| 9 | 0009 | R/W | 0000_0000 | Unused | Unused | Unused | Unused | Unused | Unused | Unused | Unused |
| 10 | 000A | R/W | 0000_0000 | Unused | Unused | Unused | Unused | Unused | Unused | Unused | Unused |
| 11 | 000B | R/W | 0000_0000 | Unused | Unused | Unused | Unused | Unused | Unused | Unused | Unused |
| 12 | 000C | R | 0010_0110 | VendorID<7> | VendorID<6> | VendorID<5> | VendorID<4> | VendorID<3> | VendorID<2> | VendorID<1> | VendorID<0> |
| 13 | 000D | R | 0000_0100 | VendorID<15> | VendorID<14> | VendorID<13> | VendorID<12> | VendorID<11> | VendorID<10> | VendorID<9> | VendorID<8> |
| 14 | 000E | R/W | 0000_0000 | Unused | Unused | Unused | Unused | Unused | Unused | Unused | Unused |
| 15 | 000F | R/W | 0000_0000 | Unused | Unused | Unused | Unused | Unused | Unused | Unused | TransferOn |

Table 17. Control Registers^[a]

| Register | Addr | Type | Default Value | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|----------|------|------|---------------|-----------|-----------|-----------|-----------|-----------|-----------|-----------|-----------|
| 16 | 0010 | R/W | 0001_1111 | NInt<7> | NInt<6> | NInt<5> | NInt<4> | NInt<3> | NInt<2> | NInt<1> | NInt<0> |
| 17 | 0011 | R/W | 0000_0000 | NInt<15> | NInt<14> | NInt<13> | NInt<12> | NInt<11> | NInt<10> | NInt<9> | NInt<8> |
| 18 | 0012 | R/W | 1000_0110 | NFrac<7> | NFrac<6> | NFrac<5> | NFrac<4> | NFrac<3> | NFrac<2> | NFrac<1> | NFrac<0> |
| 19 | 0013 | R/W | 1111_1110 | NFrac<15> | NFrac<14> | NFrac<13> | NFrac<12> | NFrac<11> | NFrac<10> | NFrac<9> | NFrac<8> |
| 20 | 0014 | R/W | 1111_1111 | NFrac<23> | NFrac<22> | NFrac<21> | NFrac<20> | NFrac<19> | NFrac<18> | NFrac<17> | NFrac<16> |
| 21 | 0015 | R/W | 1011_1100 | NFrac<31> | NFrac<30> | NFrac<29> | NFrac<28> | NFrac<27> | NFrac<26> | NFrac<25> | NFrac<24> |
| 22 | 0016 | R/W | 0000_0000 | NMod<7> | NMod<6> | NMod<5> | NMod<4> | NMod<3> | NMod<2> | NMod<1> | NMod<0> |
| 23 | 0017 | R/W | 1111_1110 | NMod<15> | NMod<14> | NMod<13> | NMod<12> | NMod<11> | NMod<10> | NMod<9> | NMod<8> |
| 24 | 0018 | R/W | 1111_1111 | NMod<23> | NMod<22> | NMod<21> | NMod<20> | NMod<19> | NMod<18> | NMod<17> | NMod<16> |
| 25 | 0019 | R/W | 1111_1111 | NMod<31> | NMod<30> | NMod<29> | NMod<28> | NMod<27> | NMod<26> | NMod<25> | NMod<24> |
| 26 | 001A | R/W | 0000_0001 | Phase<7> | Phase<6> | Phase<5> | Phase<4> | Phase<3> | Phase<2> | Phase<1> | Phase<0> |
| 27 | 001B | R/W | 0000_0000 | Phase<15> | Phase<14> | Phase<13> | Phase<12> | Phase<11> | Phase<10> | Phase<9> | Phase<8> |

Table 17. Control Registers^[a] (Cont.)

| Register | Addr | Type | Default Value | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|----------|------|------|---------------|-----------------|------------------|-------------------|--------------------|-------------------|----------------------|----------------|----------------|
| 28 | 001C | R/W | 0000_0000 | Phase<23> | Phase<22> | Phase<21> | Phase<20> | Phase<19> | Phase<18> | Phase<17> | Phase<16> |
| 29 | 001D | R/W | 0000_0000 | Phase<31> | Phase<30> | Phase<29> | Phase<28> | Phase<27> | Phase<26> | Phase<25> | Phase<24> |
| 30 | 001E | R/W | 0111_0010 | 0 | DSMOrder<2> | DSMOrder<1> | DSMOrder<0> | DitherG<1> | DitherG<0> | ShapeDitherEn | DitherEn |
| 31 | 001F | R/W | 0000_1000 | ManualBandEn | 0 | 0 | 0 | VCOManu<3> | VCOManu<2> | VCOManu<1> | VCOManu<0> |
| 32 | 0020 | R/W | 0100_0000 | Unused | BandManu<6> | BandManu<5> | BandManu<4> | BandManu<3> | BandManu<2> | BandManu<1> | BandManu<0> |
| 33 | 0021 | R/W | 0000_0011 | ForceRelock | PhAdj | BandSelDivisible | ManualResync | 0 | 0 | BandSelAcc<1> | BandSelAcc<0> |
| 34 | 0022 | R/W | 0000_0000 | BndSelDiv<7> | BndSelDiv<6> | BndSelDiv<5> | BndSelDiv<4> | BndSelDiv<3> | BndSelDiv<2> | BndSelDiv<1> | BndSelDiv<0> |
| 35 | 0023 | R/W | 0000_1010 | Unused | Unused | Unused | BndSelDiv<12> | BndSelDiv<11> | BndSelDiv<10> | BndSelDiv<9> | BndSelDiv<8> |
| 36 | 0024 | R/W | 0000_0000 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| 37 | 0025 | R/W | 0000_0000 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| 38 | 0026 | R/W | 0000_0000 | Unused | Unused | Unused | LD_Enable | AutoRecalEn | 0 | 0 | 0 |
| 39 | 0027 | R/W | 0000_0000 | Unused | Unused | LDPinMode<1> | LDPinMode<0> | Unused | LDP<2> | LDP<1> | LDP<0> |
| 40 | 0028 | R/W | 0000_0001 | Reserved | ref_vreg_pwrdown | pdcp_vreg_pwrdown | fb_vreg_pwrdown | outA_vreg_pwrdown | outBbuf_vreg_pwrdown | PDNAnaRegu | VCO_En |
| 41 | 0029 | R/W | 0000_0001 | R<7> | R<6> | R<5> | R<4> | R<3> | R<2> | R<1> | R<0> |
| 42 | 002A | R/W | 0000_1100 | Unused | Unused | Unused | RefDoubleRateDelay | Input_Type | RefDoubleRateEn | R<9> | R<8> |
| 43 | 002B | R/W | 0000_0000 | Mult_En | Mult_reset | Mult<5> | Mult<4> | Mult<3> | Mult<2> | Mult<1> | Mult<0> |
| 44 | 002C | R/W | 0000_0000 | nMultpwrdown<2> | nMultpwrdown<1> | nMultpwrdown<0> | Mult_force_vchi | Mult_force_vclow | 1 | 0 | 0 |
| 45 | 002D | R/W | 0001_1101 | Unused | Unused | lcp_pmos<5> | lcp_pmos<4> | lcp_pmos<3> | lcp_pmos<2> | lcp_pmos<1> | lcp_pmos<0> |
| 46 | 002E | R/W | 0001_1101 | Unused | Unused | lcp_nmos<5> | lcp_nmos<4> | lcp_nmos<3> | lcp_nmos<2> | lcp_nmos<1> | lcp_nmos<0> |
| 47 | 002F | R/W | 0000_0000 | CP_HiZ | lcp_bleeder<6> | lcp_bleeder<5> | lcp_bleeder<4> | lcp_bleeder<3> | lcp_bleeder<2> | lcp_bleeder<1> | lcp_bleeder<0> |
| 48 | 0030 | R/W | 0110_0010 | Unused | Unused | Unused | Unused | pfd_pw<1> | pfd_pw<0> | 1 | Unused |
| 49 | 0031 | R/W | 0000_0000 | 1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 |
| 50 | 0032 | R/W | 0000_0000 | AutoResync | 0 | 0 | 1 | 0 | 0 | 1 | 1 |
| 51 | 0033 | R/W | 0000_0000 | Unused | Unused | Unused | Unused | RF_OUTA_pwr<3> | RF_OUTA_pwr<2> | RF_OUTA_pwr<1> | RF_OUTA_pwr<0> |
| 52 | 0034 | R/W | 0000_1000 | 1 | Unused | Mute_until_LD | RF_OUTA_ena | 1 | 1 | 1 | 0 |
| 53 | 0035 | R/W | 0000_0000 | Unused | Unused | Unused | Unused | RF_OUTB_pwr<3> | RF_OUTB_pwr<2> | RF_OUTB_pwr<1> | RF_OUTB_pwr<0> |

Table 17. Control Registers^[a] (Cont.)

| Register | Addr | Type | Default Value | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|----------|------|------|---------------|---------------|----------------|----------------|-------------|----------|-----------|-----------|-----------|
| 54 | 0036 | R/W | 0000_1000 | 1 | Unused | Unused | RF_OUTB_ena | 1 | 1 | 1 | 0 |
| 55 | 0037 | R/W | 0000_0000 | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved |
| 56 | 0038 | R/W | 0000_0000 | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved |
| 57 | 0039 | R/W | 0000_0000 | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved |
| 58 | 003A | R/W | 0000_0000 | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved |
| 59 | 003B | R/W | 0000_0000 | OutDoublr_Ena | OutDivider_Ena | OutDoublr_Freq | Unused | Unused | OutDiv<2> | OutDiv<1> | OutDiv<0> |
| 60 | 003C | R/W | 0000_0000 | Unused | Unused | Unused | Unused | 0 | 0 | 1 | 0 |
| 61 | 003D | R/W | 0000_0000 | Unused | Unused | Unused | Reserved | Reserved | 0 | 0 | 0 |
| 62 | 003E | R/W | 0000_0000 | Unused | Unused | Unused | Unused | Unused | Unused | Unused | Unused |
| 63 | 003F | R/W | 0000_0000 | Unused | Unused | Unused | Unused | Unused | Unused | Unused | Reserved |
| 64 | 0040 | R/W | 0000_0000 | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved |
| 65 | 0041 | R/W | 0110_0010 | 0 | 1 | 1 | 0 | Unused | Unused | 1 | 0 |
| 66 | 0042 | R/W | 0010_0010 | Unused | Unused | 1 | 0 | Unused | Unused | 1 | 0 |
| 67 | 0043 | R/W | 0010_0010 | Unused | Unused | 1 | 0 | Unused | Unused | 1 | 0 |

[a] It is recommended to write 0 for Unused and Reserved bits when writing.

Table 18. Status Registers

| Register | Addr | Type | Default Value | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|----------|------|------|---------------|---------|-------------|------------|------------|------------|------------|------------|------------|
| 68 | 0044 | R | 0000_0000 | DigLock | BandSelDone | Unused | Unused | VcoSts<3> | VcoSts<2> | VcoSts<1> | VcoSts<0> |
| 69 | 0045 | R | 0000_0000 | Unused | BandSts<6> | BandSts<5> | BandSts<4> | BandSts<3> | BandSts<2> | BandSts<1> | BandSts<0> |
| 70 | 0046 | R | 0000_0000 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| 71 | 0047 | R | 0000_0000 | Unused | Unused | 0 | 0 | 0 | 0 | 0 | 0 |
| 72 | 0048 | R | 0000_0000 | Unused | Unused | Unused | Unused | Unused | Unused | Unused | Unused |
| 73 | 0049 | R/W | 0000_0000 | Unused | Unused | Unused | Unused | Unused | LossLock | Unused | Unused |

Register Block Descriptions

Table 19. Register Block Descriptions

| | Register Blocks (Hex) | Register Block Descriptions |
|-------------------|-----------------------|---|
| Preface Registers | 0000–0001 | Startup Control Registers |
| | 0002 | Reserved |
| | 0003–0007 | Device Type, ID Version, and Option Registers |
| | 0008–000B | Unused |
| | 000C–000D | Vendor ID Control Registers |
| | 000E–000F | Unused |
| Control Registers | 0010–0019 | Feedback Divider Control Registers |
| | 001A–001D | Phase Adjustments Control Registers |
| | 001E | DSM Control Registers |
| | 001F–0020 | Manual VCO and Digital Band Selection |
| | 0021 | Calibration Control Registers |
| | 0022–0023 | Band Select Clock Divider Control Registers |
| | 0024–0025 | Reserved |
| | 0026–0027 | Lock Detect Control Registers |
| | 0028 | Power Down Control Registers |
| | 0029–002C | Input Control Registers |
| | 002D–002F | Charge Pump Current Control Registers |
| | 0030 | PFD Pulse Width Control Registers |
| | 0031–0032 | Re-Sync Control Registers |
| | 0033–003B | Output Control Registers |
| | 003C | Reserved |
| | 003D | Reserved |
| | 003E–0040 | Unused or Reserved |
| | 0041–0043 | Reserved |
| Status Registers | 0044 | Digital Lock and Calibration and VCO Status Registers |
| | 0045 | Digital Band Status Registers |
| | 0046–0048 | Reserved or Unused |
| | 0049 | Loss of Lock Status Registers |

Preface Registers

Table 20. Preface Register Block

| Register Blocks (Hex) | Register Block Descriptions |
|-----------------------|---------------------------------------|
| 0000–0001 | Startup Control Registers |
| 0002 | Reserved |
| 0003–0006 | Device Type, ID and Version Registers |
| 0007 | Reserved |
| 0008–000B | Unused |
| 000C 000D | Vendor ID Control Registers |
| 000E–000F | Unused |

Table 21. Preface Register Bits

| Addr | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|------|----------------|----------------|----------------|----------------|----------------|-----------------|----------------|----------------|
| 0000 | SoftReset | LSBFirst | AddressAscend | SDOActive | <SDOActive> | <AddressAscend> | <LSBFirst> | <SoftReset> |
| 0001 | Unused | Unused | BufferReadMode | Unused | Unused | Unused | Unused | Unused |
| 0002 | Unused | Unused | Unused | Unused | Unused | Unused | Unused | Unused |
| 0003 | ChipType<7> | ChipType<6> | ChipType<5> | ChipType<4> | ChipType<3> | ChipType<2> | ChipType<1> | ChipType<0> |
| 0004 | ChipID<7> | ChipID<6> | ChipID<5> | ChipID<4> | ChipID<3> | ChipID<2> | ChipID<1> | ChipID<0> |
| 0005 | ChipID<15> | ChipID<14> | ChipID<13> | ChipID<12> | ChipID<11> | ChipID<10> | ChipID<9> | ChipID<8> |
| 0006 | ChipVersion<7> | ChipVersion<6> | ChipVersion<5> | ChipVersion<4> | ChipVersion<3> | ChipVersion<2> | ChipVersion<1> | ChipVersion<0> |
| 0007 | ChipOption<7> | ChipOption<6> | ChipOption<5> | ChipOption<4> | ChipOption<3> | ChipOption<2> | ChipOption<1> | ChipOption<0> |
| 0008 | Unused | Unused | Unused | Unused | Unused | Unused | Unused | Unused |
| 0009 | Unused | Unused | Unused | Unused | Unused | Unused | Unused | Unused |
| 000A | Unused | Unused | Unused | Unused | Unused | Unused | Unused | Unused |
| 000B | Unused | Unused | Unused | Unused | Unused | Unused | Unused | Unused |
| 000C | VendorID<7> | VendorID<6> | VendorID<5> | VendorID<4> | VendorID<3> | VendorID<2> | VendorID<1> | VendorID<0> |
| 000D | VendorID<15> | VendorID<14> | VendorID<13> | VendorID<12> | VendorID<11> | VendorID<10> | VendorID<9> | VendorID<8> |
| 000E | Unused | Unused | Unused | Unused | Unused | Unused | Unused | Unused |
| 000F | Unused | Unused | Unused | Unused | Unused | Unused | Unused | TransferOn |

Table 22. Preface Register Description

| Bit Field Name | Field Type | Default (Binary) | Description |
|----------------------------------|-------------------|------------------|--|
| SoftReset <SoftReset> | R/W Auto-Clear | 0 | Soft Reset Function: 0 = Normal operation 1 = Register reset. The device loads the default values into the registers, 0002 - 0049. The content of the register addresses 0000 and 0001 and the SPI engine are not reset. SoftReset bit D7 is mirrored with <SoftReset> in bit position D0. Register reset requires setting both SoftReset and <SoftReset> bits. |
| LSBFirst <LSBFirst> | R/W | 0 | Least Significant Bit Position: Defines the bit transmitted first in SPI transfers between slave and master. 0 = The most significant bit (D7) first 1 = The least significant bit (D0) first The LSBFirst bit D6 is mirrored with the <LSBFirst> in bit position D1. Changing LSBFirst to most significant bit requires setting both LSBFirst and <LSBFirst> bits. |
| AddressAscend <AddressAscend> | R/W | 0 | Address Ascend On: 0 = Address Ascend is off (addresses auto-decrement in streaming SPI mode) 1 = Address Ascend is on (addresses auto-increment in streaming SPI mode) The AddressAscend bit specifies whether addresses are incremented or decremented in streaming SPI transfers. The AddressAscend bit D5 is mirrored with the <AddressAscend> in bit position D2. Changing AddressAscend to "ON" requires setting both the AddressAscend and <AddressAscend> bits. |
| SDOActive <SDOActive> | R/W | 0 | SPI 3/4-Wire Mode: Selects the unidirectional or bidirectional data transfer mode for the SDIO pin. 0 = SPI 3-wire mode: – SDIO is the SPI bidirectional data I/O pin – SDO pin is not used and is in high-impedance 1 = SPI 4-wire mode – SDIO is the SPI data input pin – SDO is the SPI data output pin The SDOActive bit D4 is mirrored with <SDOActive> in bit position D3. Changing SDOActive to SPI 4-wire mode requires setting both the SDOActive and <SDOActive> bits. |
| BufferReadMode | R/W | 0 | Read Back Mode of the Buffer Registers: 0 = Read from active registers 1 = Read from the Buffer Register (case of Doubled Buffer Registers); If the register being read is not doubled buffered, a 1 value will read from the active register. |
| ChipType[7:0] | R only | 0000 0110 | Device (Chip) Type: Reads 00000110 (RF Synthesizer / PLL) after power-up and reset. |

Table 22. Preface Register Description (Cont.)

| Bit Field Name | Field Type | Default (Binary) | Description |
|------------------|-------------------|------------------------|---|
| ChipID[15:0] | R only | 0000000000000001 | Device (Chip) ID |
| ChipVersion[7:0] | R only | 0000 0010 | Device (Chip) Version |
| ChipOption[7:0] | R only | 0001 0001 | Device (Chip) Option |
| VendorID[15:0] | R only | 0000 0100 0010 0110 | Vendor ID: 0x0426 (IDT/Renesas). Reads 0x0426 (IDT/Renesas) after power-up and reset. |
| TransferOn | R/W Auto-Clear | 0 | Transfer On Function transfers the buffer registers values into the active registers: 1 = Transfer ON 0 = Transfer OFF This bit must be set 1 to transfer the contents of the buffers into the active registers. This bit is self-clearing; that is, it does not have to be set back to 0. When this bit is set to 1 (self-clearing), the device updates the active registers to the contents of the buffer registers simultaneously. This is done in order to avoid taking effect asynchronously during programming a multi-bit value (more than 8 bits) into several registers. |

Feedback Divider Control Registers

Table 23. Feedback Divider Control Block

| Register Blocks (Hex) | Register Block Descriptions |
|-----------------------|------------------------------------|
| 0010–0019 | Feedback Divider Control Registers |

Table 24. Feedback Divider Control Register Bits

| Addr | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|------|-----------|-----------|-----------|-----------|-----------|-----------|-----------|-----------|
| 0010 | NInt<7> | NInt<6> | NInt<5> | NInt<4> | NInt<3> | NInt<2> | NInt<1> | NInt<0> |
| 0011 | NInt<15> | NInt<14> | NInt<13> | NInt<12> | NInt<11> | NInt<10> | NInt<9> | NInt<8> |
| 0012 | NFrac<7> | NFrac<6> | NFrac<5> | NFrac<4> | NFrac<3> | NFrac<2> | NFrac<1> | NFrac<0> |
| 0013 | NFrac<15> | NFrac<14> | NFrac<13> | NFrac<12> | NFrac<11> | NFrac<10> | NFrac<9> | NFrac<8> |
| 0014 | NFrac<23> | NFrac<22> | NFrac<21> | NFrac<20> | NFrac<19> | NFrac<18> | NFrac<17> | NFrac<16> |
| 0015 | NFrac<31> | NFrac<30> | NFrac<29> | NFrac<28> | NFrac<27> | NFrac<26> | NFrac<25> | NFrac<24> |
| 0016 | NMod<7> | NMod<6> | NMod<5> | NMod<4> | NMod<3> | NMod<2> | NMod<1> | NMod<0> |
| 0017 | NMod<15> | NMod<14> | NMod<13> | NMod<12> | NMod<11> | NMod<10> | NMod<9> | NMod<8> |
| 0018 | NMod<23> | NMod<22> | NMod<21> | NMod<20> | NMod<19> | NMod<18> | NMod<17> | NMod<16> |
| 0019 | NMod<31> | NMod<30> | NMod<29> | NMod<28> | NMod<27> | NMod<26> | NMod<25> | NMod<24> |

Table 25. Feedback Divider Control Register Description

| Register Description | | | |
|----------------------|------------|---|--|
| Bit Field Name | Field Type | Default (Binary) | Description |
| Nint[15:0] | R/W | 0000 0000 0001 1111 = d'31 | PLL Feedback Divider N Integer Portion INT: Minimum divide ratio is 12 0000 0000 0000 0000 = Not allowed ... 0000 0000 0000 1011 = Not allowed 0000 0000 0000 0111 = 12 ... 0000 0000 0001 1111 = 31 (default) ... 1111 1111 1111 1111 = 65,535 |
| Nfrac[31:0] | R/W | 1011 1100 1111 1111 1111 1110 1000 0110 = d'3,170,893,446 | PLL Feedback Divider N Fractional Portion FRAC ^[a] : 0000 0000 0000 0000 0000 0000 0000 0000 = d'0 ... 1011 1100 1111 1111 1111 1110 1000 0110 = d'3,170,893,446 (default) ... 1111 1111 1111 1111 1111 1111 1111 1111 = d'4,294,967,295 |
| Nmod[31:0] | R/W | 1111 1111 1111 1111 1111 1110 0000 0000 = d'4,294,966,784 | PLL Feedback Divider N Modulus Portion MOD: 0000 0000 0000 0000 0000 0000 0000 0000 = Not allowed 0000 0000 0000 0000 0000 0000 0000 0001 = Not allowed 0000 0000 0000 0000 0000 0000 0000 0010 = d'2 ... 1111 1111 1111 1111 1111 1110 0000 0000 = d'4,294,966,784 (default) ... 1111 1111 1111 1111 1111 1111 1111 1111 = d'4,294,967,295 |

[a] Nfrac is the numerator value of the fractional divide ratio. It is programmable from 0 to (MOD -1).

Phase Adjustments Control Registers

Table 26. Phase Adjustments Control Register Block

| Register Blocks (Hex) | Register Block Descriptions |
|-----------------------|-------------------------------------|
| 001A–001D | Phase Adjustments Control Registers |

Table 27. Phase Adjustments Control Register Bits

| Addr | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|------|-----------|-----------|-----------|-----------|-----------|-----------|-----------|-----------|
| 001A | Phase<7> | Phase<6> | Phase<5> | Phase<4> | Phase<3> | Phase<2> | Phase<1> | Phase<0> |
| 001B | Phase<15> | Phase<14> | Phase<13> | Phase<12> | Phase<11> | Phase<10> | Phase<9> | Phase<8> |
| 001C | Phase<23> | Phase<22> | Phase<21> | Phase<20> | Phase<19> | Phase<18> | Phase<17> | Phase<16> |
| 001D | Phase<31> | Phase<30> | Phase<29> | Phase<28> | Phase<27> | Phase<26> | Phase<25> | Phase<24> |

Table 28. Phase Adjustments Control Register Descriptions

| Register Description | | | |
|----------------------|------------|---|--|
| Bit Field Name | Field Type | Default (Binary) | Description |
| Phase[31:0] | R/W | 0000 0000 0000 0000 0000 0000 0000 0001 = d'1 | Phase Adjustments ^[a] 0000 0000 0000 0000 0000 0000 0000 0000 = d'0 0000 0000 0000 0000 0000 0000 0000 0001 = d'1 (default) ... 1111 1111 1111 1111 1111 1111 1111 1111 = d'4,294,967,295 |

[a] Phase adjustment (Phase value) must be less than the Modulus (Nmod value).

DSM Control Registers

Table 29. DSM Control Register Block

| Register Blocks (Hex) | Register Block Descriptions |
|-----------------------|-----------------------------|
| 001E | DSM Control Registers |

Table 30. DSM Control Register Bits

| Addr | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|------|----|-------------|-------------|-------------|------------|------------|---------------|----------|
| 001E | 0 | DSMOrder<2> | DSMOrder<1> | DSMOrder<0> | DitherG<1> | DitherG<0> | ShapeDitherEn | DitherEn |

Table 31. DSM Control Register Descriptions

| Register Description | | | |
|----------------------|------------|------------------|--|
| Bit Field Name | Field Type | Default (Binary) | Description |
| DSMOrder[2:0] | R/W | 111 | DSM Order 000 = OFF. The device operates in integer mode and the fractional part is ignored 001 = 1st order 010 = 2nd order 011 = 3rd order 100 = Reserved 101 = Reserved 110 = Reserved 111 = 3rd order (default) |
| DitherG[1:0] | R/W | 00 | Dither Gain 00 = LSB Dither (Recommended) (default) 01 = LSB x2 Dither 10 = LSB x4 Dither 11 = LSB x8 Dither |
| ShapeDitherEn | R/W | 1 | Shape Dither Enable 0 = Shaped dither disabled 1 = Shaped dither enabled |
| DitherEn | R/W | 0 | Dither Enable 0 = Dither off 1 = Dither on |

Calibration Control Registers

Table 32. Calibration Control Register Block

| Register Blocks (Hex) | Register Block Descriptions |
|-----------------------|-------------------------------|
| 001F–0021 | Calibration Control Registers |

Table 33. Calibration Control Register Bits

| Addr | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|------|-------------|-------------|----------------|--------------|-------------|-------------|---------------|---------------|
| 001F | ManuBandEn | 0 | 0 | 0 | VCOManu<3> | VCOManu<2> | VCOManu<1> | VCOManu<0> |
| 0020 | 0 | BandManu<6> | BandManu<5> | BandManu<4> | BandManu<3> | BandManu<2> | BandManu<1> | BandManu<0> |
| 0021 | ForceRelock | PhAdj | BandSelDisable | ManualReSync | 0 | 0 | BandSelAcc<1> | BandSelAcc<0> |

Table 34. Calibration Control Register Descriptions

| Register Description | | | |
|----------------------|------------|------------------|---|
| Bit Field Name | Field Type | Default (Binary) | Description |
| ManuBandEn | R/W | 0 | Calibration Mode Enable 0 = Automatic VCO and Digital Band Selection 1 = Manual VCO and Digital Band Selection |
| VCOManu[3:0] | R/W | 1000 | Manual VCO Selection (when ManuBandEn=1) 0000 = VCO0 0001 = VCO1 ... 0111 = VCO7 1000 – 1111 = Unused |
| Bandmanu[6:0] | R/W | 10000000 | Manual Digital Band Selection (when ManuBandEn=1) 0000000 = Band 0000001 = Band1 ... 1111111 = Band127 |
| ForceRelock | R/W | 0 | ForceRelock 0 = Normal operation (default) 1 = VCO forced to recalibrate. This bit is self-clearing. Note: When the PLL is used in integer mode, this bit must be set to 1 after programming the feedback divider value. |
| PhAdj | R/W | 0 | Phase Adjust Triggers 0 = Normal operation 1 = Trigger phase adjustment once. This bit is self-clearing. |

Table 34. Calibration Control Register Descriptions

| Register Description | | | |
|----------------------|------------|------------------|--|
| Bit Field Name | Field Type | Default (Binary) | Description |
| BandSelDisable | R/W | 0 | Band Select Disable This bit will prevent a VCO recalibration when registers 0x0010 - 0x0019 are written. 0 = VCO recalibrates when registers 0x0010 - 0x0019 are written 1 = VCO does not recalibrate when registers 0x0010 - 0x0019 are written |
| ManualReSync | R/W | 0 | 0 = Normal operation 1 = Reset the DSM immediately |
| BandSelAcc[1:0] | R/W | 11 | Band select/Calibration resolution 00 = Reserved 01 = Reserved 10 = 4x resolution 11 = 8x resolution (default) |

Band Select Clock Divider Control Registers

Table 35. Band Select Clock Divider Control Register Block

| Register Blocks (Hex) | Register Block Descriptions |
|-----------------------|---|
| 0022–0023 | Band Select Clock Divider Control Registers |

Table 36. Band Select Clock Divider Control Register Bits

| Addr | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|------|--------------|--------------|--------------|---------------|---------------|---------------|--------------|--------------|
| 0022 | BndSelDiv<7> | BndSelDiv<6> | BndSelDiv<5> | BndSelDiv<4> | BndSelDiv<3> | BndSelDiv<2> | BndSelDiv<1> | BndSelDiv<0> |
| 0023 | Unused | Unused | Unused | BndSelDiv<12> | BndSelDiv<11> | BndSelDiv<10> | BndSelDiv<9> | BndSelDiv<8> |

Table 37. Band Select Clock Divider Control Register Descriptions

| Register Description | | | |
|----------------------|------------|-------------------------------|--|
| Bit Field Name | Field Type | Default (Binary) | Description |
| BndSelDiv[12:0] | R/W | 0 1010 0000 0000 = d'2,560 | Band Select Clock Divider ^[a] 0 0000 0000 0000 = Not allowed 0 0000 0000 0001 = d'1 ... 0 1010 0000 0000 = d'2,560 (default) ... 1 1111 1111 1111 = d'8,192 |

[a] This value should be set so that $F_{\text{PFD}} / \text{BndSelDiv}$ is < 100kHz and > 50kHz.

Lock Detect Control Registers

Table 38. Lock Detect Control Register Block

| Register Blocks (Hex) | Register Block Descriptions |
|-----------------------|-------------------------------|
| 0026–0027 | Lock Detect Control Registers |

Table 39. Lock Detect Control Register Bits

| Addr | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|------|--------|--------|--------------|--------------|-------------|--------|--------|--------|
| 0026 | Unused | Unused | Unused | LD_Disable | AutoRecalEn | 0 | 0 | 0 |
| 0027 | Unused | Unused | LDPinMode<1> | LDPinMode<0> | Unused | LDP<2> | LDP<1> | LDP<0> |

Table 40. Lock Detect Control Register Descriptions

| Register Description | | | |
|----------------------|------------|------------------|---|
| Bit Field Name | Field Type | Default (Binary) | Description |
| LD_Disable | R/W | 0 | Lock Detect Disable 0 = Disable lock detect circuitry (default) 1 = Enable lock detect circuitry |
| AutoRecalEn | R/W | 0 | Automatic Recalibration Enable 0 = Disable (default) 1 = Enable (an automatic recalibration occurs if an unlock on LD is detected) |
| LDPinMode[1:0] | R/W | 00 | LD Pin Mode 00 = Digital Lock Detect (default); Normal lock detector function 01 = Calibration done 10 = Low 11 = High |
| LDP[2:0] | R/W | 000 | Lock Detector Precision setting (ns) 000 = 0.375 (default) 001 = 0.75 010 = 1.5 011 = 2.4 100 = 5.2 101 = 5.2 110 = 8.5 111 = 8.5 |

Power Down Control Registers

Table 41. Power Down Control Register Block

| Register Blocks (Hex) | Register Block Descriptions |
|-----------------------|------------------------------|
| 0028 | Power Down Control Registers |

Table 42. Power Down Control Register Bits

| Addr | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|------|----------|------------------|-------------------|-----------------|-------------------|----------------------|-------------|--------|
| 0028 | Reserved | ref_vreg_pwrdown | pdcp_vreg_pwrdown | fb_vreg_pwrdown | outA_vreg_pwrdown | outBbuf_vreg_pwrdown | PDNAAnaRegu | VCO_En |

Table 43. Power Down Control Register Descriptions

| Register Description | | | |
|----------------------|------------|------------------|---|
| Bit Field Name | Field Type | Default (Binary) | Description |
| ref_vreg_pwrdown | R/W | 0 | Reference Input Path Regulator Power Down Control 0 = Regulator enabled (default) 1 = Regulator powered down |
| pdcp_vreg_pwrdown | R/W | 0 | Phase Detector and Charge Pump Regulator Power Down Control 0 = Regulator enabled (default) 1 = Regulator powered down |
| fb_vreg_pwrdown | R/W | 0 | Feedback Divider Regulator Power Down Control 0 = Regulator enabled (default) 1 = Regulator powered down |
| outA_vreg_pwrdown | R/W | 0 | OutputA Regulator Power Down Control 0 = Regulator enabled (default) 1 = Regulator powered down |
| outBbuf_vreg_pwrdown | R/W | 0 | Power Down Control for output path regulator and outputB regulator 0 = Regulator enabled (default) 1 = Regulator powered down |
| PDNAAnaRegu | R/W | 0 | Analog Regulators Power Down Control ^[a] 0 = Normal working mode (default) 1 = Power down all analog regulators |
| VCO_En | R/W | 1 | VCO Enable 0 = Disable all VCOs (reduces VCO supply current I_{VCO} to 46mA (typ.)) 1 = Normal mode |

[a] A value of 1 in PDNAAnaRegu will supersede the values set for ref_vreg_pwrdown, pdcp_vreg_pwrdown, fb_vreg_pwrdown, and out_vreg_pwrdown.

Input Control Registers

Table 44. Input Control Register Block

| Register Blocks (Hex) | Register Block Descriptions |
|-----------------------|-----------------------------|
| 0029–002C | Input Control Registers |

Table 45. Input Control Register Bits

| Addr | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|------|--------------|--------------|-------------|------------------|------------------|---------------|---------|---------|
| 0029 | R<7> | R<6> | R<5> | R<4> | R<3> | R<2> | R<1> | R<0> |
| 002A | Unused | Unused | Unused | RefDoubler_Delay | Input_Type | RefDoubler_En | R<9> | R<8> |
| 002B | Mult_En | nMult_Reset | Mult<5> | Mult<4> | Mult<3> | Mult<2> | Mult<1> | Mult<0> |
| 002C | Mult_mux_ena | Mult_d2s_ena | Mult_cp_ena | Mult_force_vchi | Mult_force_vclow | 1 | 0 | 0 |

Table 46. Input Control Register Descriptions

| Register Description | | | |
|----------------------|------------|--------------------|--|
| Bit Field Name | Field Type | Default (Binary) | Description |
| R[9:0] | R/W | 00 0000 0001 = d'1 | Input reference divide value 00 0000 0000 = d'1 00 0000 0001 = d'1 (default) ... 11 1111 1111 = d'1023 |
| RefDoubler_Delay | R/W | 0 | Selects the standard or extended pulse width delay for the input doubler. 0 = Standard pulse width (default). Use if the input reference frequency is > 50MHz. 1 = Extended pulse width for use with low frequencies (< 50MHz) |
| Input_Type | R/W | 1 | Input type: Selects either differential or single-ended input 0 = Single-ended input 1 = Differential input (default) |
| RefDoubler_En | R/W | 1 | Reference Doubler Enable Enables the Input Reference Doubler 0 = Input reference doubler disabled 1 = Input reference doubler enabled (default) |
| Mult_En | R/W | 0 | MULT Enable 0 = MULT not enabled 1 = MULT enabled |

Table 46. Input Control Register Descriptions (Cont.)

| Register Description | | | |
|----------------------|------------|------------------|--|
| Bit Field Name | Field Type | Default (Binary) | Description |
| Mult_reset | R/W | 0 | Resets the Reference Multiplier Block (MULT) 0 = Multiplier is active (default) 1 = Multiplier is reset ^[a] |
| Mult<5:0> | R/W | 000000 | Frequency multiplication factor for the input clock. When enabled, the multiplier block (Mult) multiplies the input frequency to a higher frequency for the phase detector. 000000 = Unused 000001 = Unused 000010 = Unused 000011 = Multiplication by 3 000100 = Multiplication by 4 ... 111111 = Multiplication by 63 |
| Mult_mux_ena | R/W | 0 | Mux enable for the multiplier 0 = Multiplier mux is disabled (default) 1 = Multiplier mux is enabled |
| Mult_d2s_ena | R/W | 0 | Differential to single-ended block enable for the Multiplier 0 = Multiplier differential to single-ended block disabled (default) 1 = Multiplier differential to single-ended block enabled |
| Mult_cp_ena | R/W | 0 | Charge Pump enable for the multiplier 0 = Multiplier charge pump disabled (default) 1 = Multiplier charge pump enabled |
| Mult_force_vchi | R/W | 0 | Force Multiplier Control Voltage High 0 = Normal operation (default) 1 = Multiplier control voltage is charged to VDD |
| Mult_force_vclow | R/W | 0 | Force Multiplier Control Voltage Low ^[b] 0 = Normal operation (default) 1 = Multiplier control voltage is discharged to GND |

[a] When Input Multiplier (MULT) is being used, it is recommended to program the device with proper MULT settings keeping Mult_reset = 1 and toggle it to low (active), then transfer the data using TransferOn Bit (Register 15, Bit 0) for Doubled-buffered registers and re-lock the PLL (ForceRelock).

[b] If Input Multiplier is not used, it is recommended to program the Mult_force_vclow bit to 1 (High).

Charge Pump Control Registers

Table 47. Charge Pump Control Register Block

| Register Blocks (Hex) | Register Block Descriptions |
|-----------------------|-------------------------------|
| 002D–0030 | Charge Pump Control Registers |

Table 48. Charge Pump Control Register Bits

| Addr | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|------|--------|----------------|----------------|----------------|----------------|----------------|----------------|----------------|
| 002D | Unused | Unused | lcp_pmos<5> | lcp_pmos<4> | lcp_pmos<3> | lcp_pmos<2> | lcp_pmos<1> | lcp_pmos<0> |
| 002E | Unused | Unused | lcp_nmos<5> | lcp_nmos<4> | lcp_nmos<3> | lcp_nmos<2> | lcp_nmos<1> | lcp_nmos<0> |
| 002F | CP_HiZ | lcp_bleeder<6> | lcp_bleeder<5> | lcp_bleeder<4> | lcp_bleeder<3> | lcp_bleeder<2> | lcp_bleeder<1> | lcp_bleeder<0> |
| 0030 | Unused | Unused | Unused | Unused | pfd_pw<1> | pfd_pw<0> | 1 | Unused |

Table 49. Charge Pump Control Register Descriptions

| Register Description | | | |
|------------------------------|------------|------------------|--|
| Bit Field Name | Field Type | Default (Binary) | Description |
| lcp_pmos[5:0] ^[a] | R/W | 01 1101 | Charge Pump Pmos Current Setting 00 0000 = 166uA 00 0001 = 333uA ... 01 1101 = 5mA (default) ... 11 1111 = 10.66mA |
| lcp_nmos[5:0] ^[b] | R/W | 01 1101 | Charge Pump Nmos Current Setting 00 0000 = 166uA 00 0001 = 333uA ... 01 1101 = 5mA (default) ... 11 1111 = 10.66mA |

Table 49. Charge Pump Control Register Descriptions (Cont.)

| Register Description | | | |
|---------------------------------|------------|------------------|--|
| Bit Field Name | Field Type | Default (Binary) | Description |
| CP_HiZ | R/W | 0 | Charge Pump High-Impedance Control 0 = Charge pump active 1 = Charge pump high-impedance |
| Icp_bleeder[6:0] ^[c] | R/W | 000 0000 | Charge Pump Bleeder Current Setting 000 0000 = Off (0uA) (Default) 000 0001 = 20 uA 000 0010 = 40 uA ... 111 1111 = 2540 uA |
| pfd_pw<1:0> | R/W | 00 | PFD Pulse Width Setting 00 = 260ps (default) 01 = 348ps 10 = 487ps 11 = 583ps |

[a] $I_{CP_pmos} = 166.66e - 6 \times (\text{binary value} + 1)$

[b] $I_{CP_pmos} = 166.66e - 6 \times (\text{binary value} + 1)$

[c] $I_{CP_bleeder} = 20e - 6 \times (\text{binary value})$

Re-Sync Control Registers

Table 50. Re-Sync Control Register Block

| Register Blocks (Hex) | Register Block Descriptions |
|-----------------------|-----------------------------|
| 0031-0032 | Re-Sync Control Registers |

Table 51. Re-sync Control Register Bits

| Addr | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|------|------------|----|----|----|----|----|----|----|
| 0031 | 1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 |
| 0032 | AutoReSync | 0 | 0 | 1 | 0 | 0 | 1 | 1 |

Table 52. Re-sync Control Register Descriptions

| Register Description | | | |
|----------------------|------------|------------------|---|
| Bit Field Name | Field Type | Default (Binary) | Description |
| AutoReSync | R/W | 0 | 0 = Normal operation and single-device operation 1 = Required setting for synchronizing the output phase of multiple devices. Waits for a pulse at the SYNC input and then resets the DSM. |

Output Control Registers

Table 53. Output Control Register Block

| Register Blocks (Hex) | Register Block Descriptions |
|-----------------------|-----------------------------|
| 0033–003B | Outputs Control Registers |

Table 54. Output Control Register Bits

| Addr | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|------|----------------|----------------|-----------------|-------------|----------------|----------------|----------------|----------------|
| 0033 | Unused | Unused | Unused | Unused | RF_OUTA_pwr<3> | RF_OUTA_pwr<2> | RF_OUTA_pwr<1> | RF_OUTA_pwr<0> |
| 0034 | 1 | Unused | Mute_until_LD | RF_OUTA_ena | 1 | 1 | 1 | 0 |
| 0035 | Unused | Unused | Unused | Unused | RF_OUTB_pwr<3> | RF_OUTB_pwr<2> | RF_OUTB_pwr<1> | RF_OUTB_pwr<0> |
| 0036 | 1 | Unused | Unused | RF_OUTB_ena | 1 | 1 | 1 | 0 |
| 0037 | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved |
| 0038 | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved |
| 0039 | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved |
| 003A | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved |
| 003B | OutDoubler_Ena | OutDivider_Ena | OutDoubler_Freq | Unused | Unused | OutDiv<2> | OutDiv<1> | OutDiv<0> |

Table 55. Output Control Register Descriptions

| Register Description | | | |
|----------------------|------------|------------------|--|
| Bit Field Name | Field Type | Default (Binary) | Description |
| RF_OUTA_pwr[3:0] | R/W | 0000 | RF_OUTA Power Setting Set the output power for RF_OUTA. Higher setting number provides more output power up to a maximum value, depending on the output loading used. 0000 = OFF (default) 0001 = Minimum output power setting 1100–1111: Maximum output power setting |
| Mute_until_LD | R/W | 0 | Mute until Lock Detect selection 0 = Outputs are enabled independent of Lock Detect (default) 1 = Outputs are enabled only when Lock Detect is high |
| RF_OUTA_ena | R/W | 0 | RF_OUTA Enable 0 = RF_OUTA is disabled (MUTED) 1 = RF_OUTA is enabled |

Table 55. Output Control Register Descriptions (Cont.)

| Register Description | | | |
|----------------------|------------|------------------|--|
| Bit Field Name | Field Type | Default (Binary) | Description |
| RF_OUTB_pwr[3:0] | R/W | 0000 | RF_OUTB Power Setting Set the output power for RF_OUTB. Higher setting number provides more output power up to a maximum value, depending on the output loading used. 0000 = OFF (Default) 0001 = Minimum output power setting 1100–1111: Maximum output power setting |
| RF_OUTB_ena | R/W | 0 | RF_OUTB Enable 0 = RF_OUTB is disabled (MUTED) 1 = RF_OUTB is enabled |
| OutDoubler_Ena | R/W | 0 | RF Output Doubler Enable ^[a] 0 = 1x path (Divide by 1) path Enabled (default) ^[b] 1 = 2x path Enabled (RF Output Doubler Enabled) |
| OutDivider_Ena | R/W | 0 | RF Output Divider Enable 0 = 1x or 2x path Enabled (default) 1 = RF Output Divider Enabled Note: The output divider M0 can only be used (OutDivider_Ena=1) if the output doubler is disabled (OutDoubler_Ena=0). ^[c] |
| OutDoubler_Freq | R/W | 1 | RF Output Doubler Frequency Setting 0 = Use this setting for VCO frequency = 7.0-9.0GHz (default) 1 = Use this setting for VCO frequency = 5.5-7.0GHz |
| OutDiv[2:0] | R/W | 000 | RF Output Divider (M0) Settings 000 = Unused 001 = Div By 2 010 = Div By 4 011 = Div By 8 100 = Div By 16 101 = Div By 32 110 = Unused 111 = Unused |

[a] OutDoubler_Ena can only be set to 1 if the VCO frequency is not greater than 9GHz.

[b] For Divide by 1 (Output Divider and Doubler bypassed), both OutDoubler_Ena Bit and Out_Divider_Ena bit must be set 0 (Low).

[c] Both the OutDoubler_Ena and Out_Divider_Ena bits must not be set 1 (High) at the same time.

Status Registers

Table 56. Status Register Block

| Register Blocks (Hex) | Register Block Descriptions |
|-----------------------|---|
| 0044 | Digital Lock and Calibration and VCO Status Registers |
| 0045 | Digital Band Status Registers |
| 0046–0047 | Reserved |
| 0048 | Unused |
| 0049 | Loss of Lock Status Registers |

Table 57. Status Register Bits

| Addr | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|------|---------|-------------|------------|------------|------------|------------|------------|------------|
| 0044 | DigLock | BandSelDone | Unused | Unused | VcoSts<3> | VcoSts<2> | VcoSts<1> | VcoSts<0> |
| 0045 | Unused | BandSts<6> | BandSts<5> | BandSts<4> | BandSts<3> | BandSts<2> | BandSts<1> | BandSts<0> |
| 0046 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| 0047 | Unused | Unused | 0 | 0 | 0 | 0 | 0 | 0 |
| 0048 | Unused | Unused | Unused | Unused | Unused | Unused | Unused | Unused |
| 0049 | Unused | Unused | Unused | Unused | Unused | LossLock | Unused | Unused |

Table 58. Status Register Descriptions

| Register Description | | | |
|----------------------|------------|------------------|---|
| Bit Field Name | Field Type | Default (Binary) | Description |
| DigLock | R | 0 | Digital Lock 0 = PLL not locked 1 = PLL locked (according to LDP settings in Register 39) |
| BandSelDone | R | 0 | Band Select Done (calibration completed) 0 = Band selection not completed 1 = Band selection completed |
| VcoSts[3:0] | R | 0000 | Status bits reporting the current VCO 0000 = VCO0 0001 = VCO1 ... 0111 = VCO7 1000–1111 = Unused |
| BandSts[6:0] | R | 000 0000 | Status bits reporting the current digital band 000 0000 = Band0 000 0001 = Band1 ... 111 1111 = Band127 |

Table 58. Status Register Descriptions (Cont.)

| Register Description | | | |
|----------------------|------------|------------------|--|
| Bit Field Name | Field Type | Default (Binary) | Description |
| LossLock | R/W1C | 0 | Status bit stating device loss of lock Sticky bit. Write 1 to this bit to clear 0 = Locked since last time register was cleared 1 = Loss of Lock since last time register was cleared |

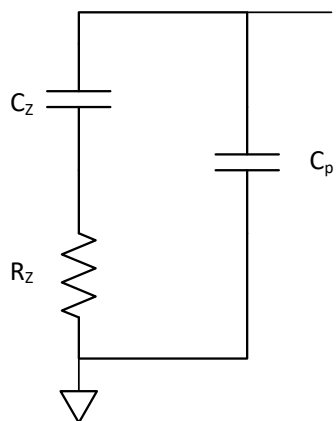
Applications Information

Loop Filter Calculations

2nd Order Loop Filter

This section provides design information for a 2nd order loop filter for the 8V97003. A general 2nd order loop filter is shown in [Figure 23](#). Step-by-step calculations to determine R_Z, C_Z, and C_p values for a desired loop bandwidth are described below. Required parameters are provided. A spreadsheet for calculating the loop filter values is also available.

Figure 23. Typical 2nd Order Loop Filter



1. Determine desired loop bandwidth f_c .
2. Calculate R_Z:

$$R_Z = \frac{2 \times \pi \times f_c \times N}{I_{CP} \times K_{VCO}}$$

Where,

- I_{CP} is charge pump current. I_{CP} is programmable from 166μA to 10.66mA.
- N is effective feedback divider. N must be programmed into the following value.

$$N = \frac{f_{VCO}}{f_{PFD}}$$

- f_{VCO} is VCO frequency. VCO frequency range: 5500 to 11000MHz

- f_{PFD} is phase detector input frequency.

$$f_{\text{PFD}} = \frac{f_{\text{REF}}}{P_V}$$

- f_{REF} is reference clock (REF_IN) input frequency.
- P_V is overall pre-divider or input doubler setting.
- K_{VCO} is VCO gain depends on VCO Frequency (see VCO gain in K_{VCO} in Table 11).

3. Calculate C_Z :

$$C_Z = \frac{\alpha}{2 \times \pi \times f_c \times R_Z}$$

Where,

- $\alpha = f_c / f_z$, user can determine an α number.
- $\alpha > 6$ is recommended.

f_z is frequency at zero.

4. Calculate C_P :

$$C_P = \frac{C_Z}{\alpha \times \beta}$$

Where,

- $\beta = f_p / f_c$, user can determine β number.
- $\beta > 4$ is recommended.

f_p is frequency at pole.

5. Verify Phase Margin (PM).

$$PM = \arctan\left(\frac{b-1}{2 \times \sqrt{b}}\right)$$

Where,

$$b = 1 + \frac{C_Z}{C_P}$$

The phase margin (PM) should be greater than 50° .

A spreadsheet for calculating the loop filter component values is available at www.IDT.com. To use the spreadsheet, simply enter the following parameters:

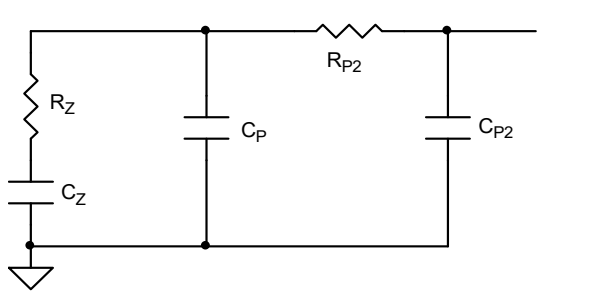
f_c , F_{ref} , P_V , I_{cp} , F_{VCO} , α , and β .

The spreadsheet will provide the component values, R_Z , C_Z , and C_P as the result. The spreadsheet also calculates the maximum phase margin for verification.

3rd Order Loop Filter

This section provides design information for a 3rd order loop filter for the 8V97003. A general 3rd order loop filter is shown in [Figure 24](#).

Figure 24. Typical 3rd Order Loop Filter



The R_Z , C_Z , and C_P can be calculated as 2nd order loop filter. The following equation helps determine the 3rd order loop filter R_{P2} and C_{P2} .

Pick an R_{P2} value. $R_{P2} \sim 1.5 \times R_Z$ is suggested.

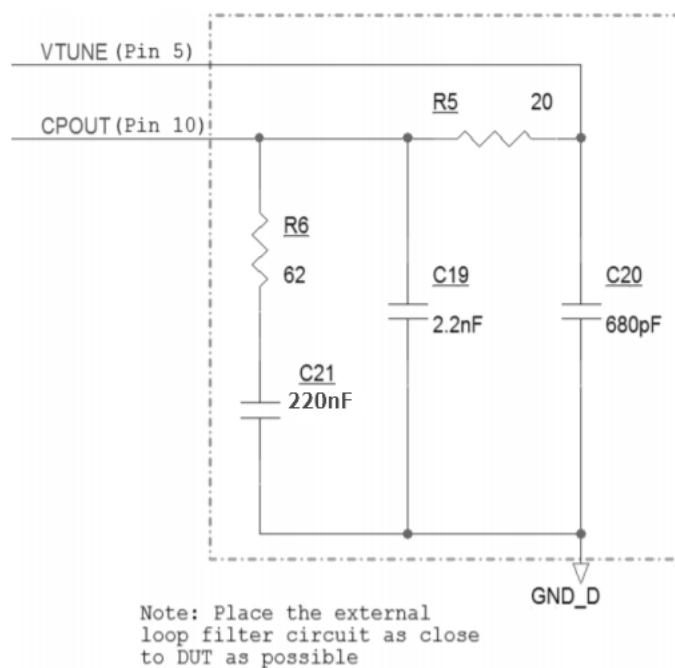
$$C_{P2} = \frac{R_Z * C_P}{R_{P2} * \gamma}$$

Where,

- γ is ratio between the 1st pole frequency and the 2nd pole frequency.
- $\gamma > 4$ is recommended.

[Figure 25](#) shows an example of a loop filter that can be used on the 8V97003.

Figure 25. Loop Filter Example



Recommendations for Unused Input and Output Pins

Inputs

LVC MOS Control Pins

All control pins have internal pullup and pulldown resistors; additional resistance is not required but can be added for additional protection. A 1k Ω resistor can be used.

Outputs

Output Pins

For any unused output, it can be left floating and disabled.

Schematic Example

Figure 26 shows a general application schematic example for the 8V97003.

For power rails, bypass capacitors must be provided to all power supply pins. At least one bypass capacitor per power pin is suggested. Value can range from 0.01 μ F or 0.1 μ F. Mix values of bypass capacitors can help filtering wider range of power supply noise.

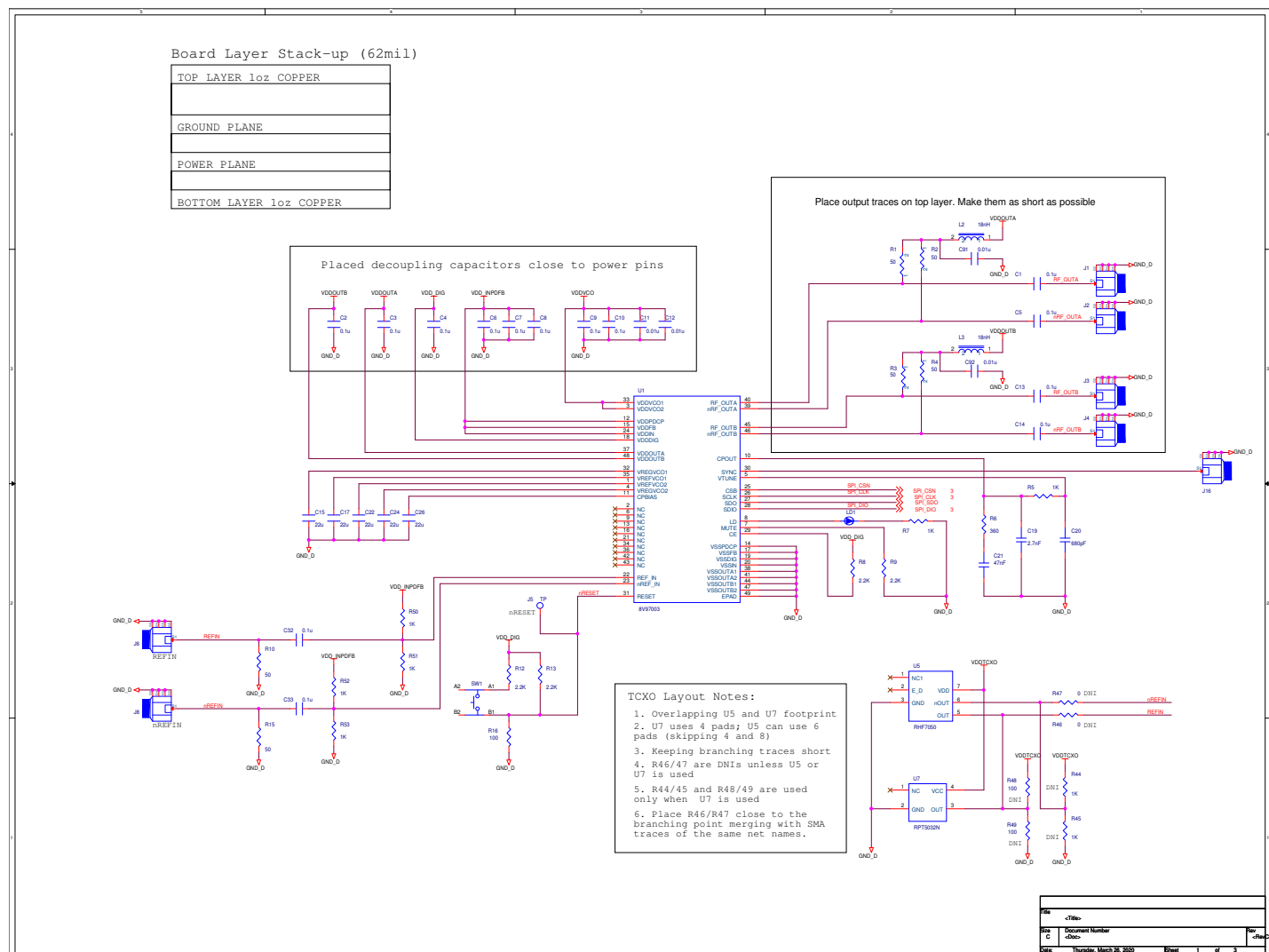
The 8V97003 input is high impedance. The input termination depends on the termination requirement from the driver. There are two input termination examples in the schematic shown in Figure 26; both are designed for transmission line with characteristic impedance $Z_0 = 50\Omega$.

The first example, labeled "Input Reference" at the lower right corner of the schematic, shows an input termination scheme for accepting a reference clock from bench signal generators. 50 Ω resistors, R82 and R83 to GND, present matched loads to signal generator's source impedance; the reference clock signal is then AC-coupled with capacitors C156 and C157 to ensure proper DC-biased to level of $V_{DDx}/2$ by voltage divider networks of R90, R91, R92, and R93. The second example shows an input termination scheme for accepting a reference clock from a TCXO. The values of R96, R96, R85, and R88 in this example are designed for a TCXO with a single-ended CMOS output on pin 5. They can be changed for different TCXO output signal types.

The 8V97003 output pull-up loading can be resistors or inductors. For inductor pull-up loading, the inductor value is frequency dependent. One inductor value cannot cover all the output frequency range. For example, an inductance of $L = 1.3\text{nH}$ that is suitable for approximately 6GHz operating frequency. The output can also drive single ended LO input.

Figure 26 also shows an example of the 8V97003 output driving single-ended LO input of the mixer through an LC balun. The LC balun component values are frequency dependent. These values can be adjusted to optimize the performance. A single-ended LO receiver input also can tap to one side of the differential driver using resistor loading or inductor loading. For single-ended LO input, both sides of the differential driver still need to be loaded with a pull up. The output power level can also be adjusted further through programming.

The loop filter values can be calculated to meet the loop bandwidth requirement (for detailed calculations, see [Loop Filter Calculations](#)).

Figure 26. Schematic Example


Power Considerations

This section provides information on power dissipation and junction temperature for the 8V97003. Equations and example calculations are also provided.

1. Power Dissipation.

The power dissipation for the 8V97003 is the total power minus the power dissipated into the loads. The following is the power dissipation for $V_{DDX} = 3.3V + 5\% = 3.465V$ at ambient temperature of $95^{\circ}C$.

Maximum current at $95^{\circ}C$, $I_{DDX_MAX} = 650mA$ (see Table 8)

- Total Power Dissipation: $Power_MAX = V_{DDX_MAX} * I_{DDX_MAX} = 3.465V * 650mA = 2252mW$
- Power dissipation in external loads for both outputs: $Power (output)MAX = 2 * I_{LOAD}^2 * 50\Omega = 22.5mW$
(Load Current with Out_Pwr = 0101, $I_{LOAD} = 15mA$)

Power Dissipation, PD = $Power_MAX - Power (output)MAX = 2252mW - 22.5mW = 2229.5mW$

2. Junction Temperature.

Junction temperature, T_j , signifies the hottest point on the device and exceeding the specified limit could cause device reliability issues. The maximum recommended junction temperature is 125°C.

For devices like this and in systems where most heat escapes from the bottom exposed pad of the package, θ_{JB} is the primary thermal resistance of interest.

The equation to calculate T_j using θ_{JB} is: $T_j = \theta_{JB} * P_D + T_B$:

T_j = Junction Temperature

θ_{JB} = Junction-to-Board Thermal Resistance

P_D = Device Power Dissipation (example calculation is in section 1 above)

T_B = Board Temperature

In order to calculate junction temperature, the appropriate junction-to-board thermal resistance θ_{JB} must be used. Assuming a 2-ground plane board, the appropriate value of θ_{JB} is 0.76°C/W (see Table 6).

Therefore, T_j for a PCB maintained at 105°C with the outputs switching is:

105°C + 2.2295W * 0.76°C/W = 106.5°C which is below the limit of 125°C.

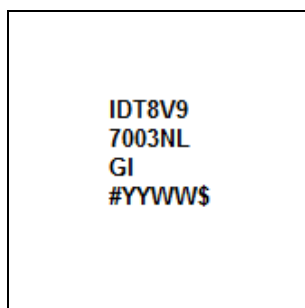
This calculation is only an example. T_j will obviously vary depending on the number of loaded outputs, supply voltage, air flow, heat transfer method, the type of board (multi-layer) and the actual maintained board temperature.

Package Outline Drawings

The package outline drawings are appended at the end of this document and are accessible from the link below. The package information is the most current data available.

www.idt.com/document/psc/48-vfqfpn-package-outline-drawing-70-x70-x-085-mm-body-05mm-pitch-epad-530-x-530-mm-nlg48p3

Marking Diagram



- Lines 1–3 indicate the part number.
- Line 4 indicates the following:
 - “#” denotes stepping.
 - “YY” is the last two digits of the year; “WW” is the work week number when the part was assembled.
 - “\$” denotes the mark code.

Ordering Information

Table 59. Ordering Information

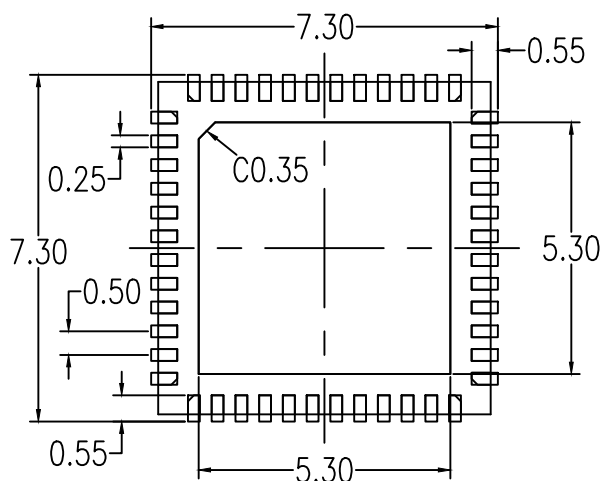
| Orderable Part Number | Package | Shipping Packaging | Temperature |
|-----------------------|--|--------------------|----------------|
| 8V97003NLGI | 7 × 7 mm 48-VFQFPN, Lead-free | Tray | -40°C to +95°C |
| 8V97003NLGI8 | 7 × 7 mm 48-VFQFPN, Lead-free Pin 1 Orientation: Quadrant 1 (EIA-481-C) | Tape and Reel | -40°C to +95°C |
| 8V97003NLGI/W | 7 × 7 mm 48-VFQFPN, Lead-free Pin 1 Orientation: Quadrant 2 (EIA-481-D) | Tape and Reel | -40°C to +95°C |

Table 60. Pin 1 Orientation in Tape and Reel Packaging

| Part Number Suffix | Pin 1 Orientation | Illustration |
|--------------------|------------------------|---|
| NLGI8 | Quadrant 1 (EIA-481-C) | <p>Correct PIN 1 ORIENTATION</p> <p>CARRIER TAPE TOPSIDE (Round Sprocket Holes)</p> <p>USER DIRECTION OF FEED</p> |
| NLGI/W | Quadrant 2 (EIA-481-D) | <p>Correct PIN 1 ORIENTATION</p> <p>CARRIER TAPE TOPSIDE (Round Sprocket Holes)</p> <p>USER DIRECTION OF FEED</p> |

Revision History

| Revision Date | Description of Change |
|--------------------|--|
| April 8, 2021 | <ul style="list-style-type: none"> Added Figure 9 |
| November 17, 2020 | <ul style="list-style-type: none"> Changed the pull-up/pull-down status of pin 31 (nRESET) to PU in Pin Descriptions Updated Figure 25 Added a note to Reference Multiplier (MULT) |
| September 15, 2020 | <ul style="list-style-type: none"> Updated the phase adjustment formulas and the last sentence in Phase Adjust Updated the description of Output Phase Synchronization Updated the first line of each field description in Table 25 Updated the description of ManualReSync in Table 34 Updated the description of AutoReSync in Table 52 |
| April 7, 2020 | <ul style="list-style-type: none"> Updated the description of CE in Table 1 Updated the Power-Down Current parameter in Table 8 Updated Reference Input Stage Changed D7 in register 0x0028 to reserved. Also updated the description of VCO_En in the same register (see Power Down Control Registers) Updated the schematic example in Figure 26 Updated loop filter values and typo corrections |
| January 20, 2020 | <ul style="list-style-type: none"> Corrected a typo for the maximum power setting in the typical current in Table 9 Added a description for register 7 (Chip option) (see Table 16 and Table 22) |
| January 8, 2020 | <ul style="list-style-type: none"> Corrected the product description on page 1 to indicate output frequency support of “171.875MHz to 18GHz” Updated the $f_{RF_OUT} = 8\text{GHz}$ test condition for $t_{jit}(\emptyset)$ in Table 12 Updated Figure 4 Rebranded the document as Renesas |
| December 20, 2019 | Initial release. |



RECOMMENDED LAND PATTERN DIMENSION

NOTES:

1. ALL DIMENSIONS ARE IN MM. ANGLES IN DEGREES.
2. TOP DOWN VIEW. AS VIEWED ON PCB.
3. LAND PATTERN IN BLUE. NSMD PATTERN ASSUMED.
4. LAND PATTERN RECOMMENDATION PER IPC-7351B GENERIC REQUIREMENT FOR SURFACE MOUNT DESIGN AND LAND PATTERN.

| Package Revision History | | |
|--------------------------|---------|----------------------------------|
| Date Created | Rev No. | Description |
| July 9, 2018 | Rev 01 | New Format, Change QFN to VFQFPN |
| July 24, 2018 | Rev 02 | Change P4 to P3 |

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