

# MAE 275 - Homework 6

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## 1 Defining the System

The state-space system can be defined,

$$\begin{aligned}\dot{\vec{x}} &= A\vec{x} + B\vec{u} \\ \vec{y} &= C\vec{x} + D\vec{u}\end{aligned}$$

where the linearized lateral aircraft equations of motion can be expressed in state space form.

$$\vec{x} = \begin{bmatrix} v \\ p \\ r \\ \phi \\ \psi \end{bmatrix}, \quad \vec{u} = \begin{bmatrix} \delta_a \\ \delta_r \end{bmatrix}, \quad \vec{y} = \begin{bmatrix} p \\ \beta \end{bmatrix}$$

Using the lateral equations of motion for F-89 at flight condition 8901, the resultant system is

$$\begin{aligned}\dot{\vec{x}} &= \begin{bmatrix} -8.2900e-2 & 0 & -6.6000e+2 & +3.2200e+1 & 0 \\ -6.8939e-3 & -1.7000e+0 & +1.7200e-1 & 0 & 0 \\ +5.1212e-3 & -6.5400e-2 & -8.9300e-2 & 0 & 0 \\ 0 & +1 & 0 & 0 & 0 \\ 0 & 0 & +1 & 0 & 0 \end{bmatrix} \vec{x} + \begin{bmatrix} 0 & +7.6500e+0 \\ +2.7300e+1 & +5.7600e-1 \\ +3.9300e-1 & -1.3600e+0 \\ 0 & 0 \\ 0 & 0 \end{bmatrix} \vec{u} \\ \vec{y} &= \begin{bmatrix} 0 & +1 & 0 & 0 & 0 \\ +1.5152e-3 & 0 & 0 & 0 & 0 \end{bmatrix} \vec{x} + \begin{bmatrix} 0 & 0 \\ 0 & 0 \end{bmatrix} \vec{u}\end{aligned}$$

Coupling numerators can be used to determine input-output pairing and which loops to close in which order, and appropriate compensators can then be designed.

## 2 Coupling Numerators

The coupling numerators can be derived using the notes in the assignment and the ICE example in the coupling numerators handout. They are

$$\boxed{\frac{p}{\delta_a} = \frac{27.3s(s^2 + 0.1747s + 3.453)}{(s + 1.781)(s - 0.001359)(s^2 + 0.09275s + 3.529)}}$$

$$\frac{p}{\delta_r} = \frac{0.576s(s - 2.885)(s + 2.56)}{(s + 1.781)(s - 0.001359)(s^2 + 0.09275s + 3.529)}$$

$$\frac{\beta}{\delta_a} = \frac{-0.393(s - 6.282)(s + 0.04952)}{(s + 1.781)(s - 0.001359)(s^2 + 0.09275s + 3.529)}$$

$$\frac{\beta}{\delta_r} = \frac{0.011591(s + 117.4)(s + 1.753)(s - 0.003733)}{(s + 1.781)(s - 0.001359)(s^2 + 0.09275s + 3.529)}$$

$$\left. \frac{p}{\delta_a} \right|_{\beta \rightarrow \delta_r} = \frac{27.3s(s + 118.1)}{(s + 117.4)(s + 1.753)(s - 0.003733)}$$

$$\boxed{\left. \frac{\beta}{\delta_r} \right|_{p \rightarrow \delta_a} = \frac{0.011591(s + 118.1)}{(s^2 + 0.1747s + 3.453)}}$$

$$\left. \frac{p}{\delta_r} \right|_{\beta \rightarrow \delta_a} = \frac{0.54936(s + 118.1)}{(s - 2.885)(s + 2.56)}$$

$$\left. \frac{\beta}{\delta_a} \right|_{p \rightarrow \delta_r} = \frac{-0.80517s(s + 118.1)}{(s - 6.282)(s + 0.04952)}$$

From these transfer functions we can:

1. rule out controlling  $p$  with  $\delta_r$  first, as there is a non-minimum phase zero (s-2.885) in the transfer function that would limit the crossover frequency to values significantly below 2.885
2. rule out  $\left. \frac{p}{\delta_r} \right|_{\beta \rightarrow \delta_a}$  and  $\left. \frac{\beta}{\delta_a} \right|_{p \rightarrow \delta_r}$  due to the closed-loop unstable poles that would be produced

This leaves only one viable option: to first close  $p$  with  $\delta_a$ , then close  $\beta$  with  $\delta_r$  with the  $p - \delta_a$  loop closed. As such, the ailerons are paired to the roll-rate, and the rudder is paired to the sideslip.

### 3 Compensators

Two compensators were chosen. After both loops were closed, the reported gain and phase margins, and the crossover bandwidths were noted at

$$\frac{p}{\delta_a} = \frac{27.3s(s^2 + 0.1747s + 3.453)}{(s + 1.781)(s - 0.001359)(s^2 + 0.09275s + 3.529)}$$

$$GC_p = 0.18 \times \frac{(s + 2)}{s}$$

$$G_m = Inf \text{ dB}$$

$$P_m = 86.7^\circ \text{ at } 5.06 \text{ rad/s}$$

$$\omega_{BW} = 5.06 \text{ rad/s}$$

$$\left. \frac{\beta}{\delta_r} \right|_{p \rightarrow \delta_a} = \frac{0.011591(s + 117.3)(s + 4.498)(s + 2.198)}{(s + 4.417)(s + 2.186)(s^2 + 0.1828s + 3.513)}$$

$$GC_\beta = 74.4 \times \frac{(s^2 + .2s + 3.5)}{s(s + 20)}$$

$$G_m = Inf \text{ dB}$$

$$P_m = 74.4^\circ \text{ at } 4.96 \text{ rad/s}$$

$$\omega_{BW} = 4.96 \text{ rad/s}$$

Note that the actual value for  $\left. \frac{\beta}{\delta_r} \right|_{p \rightarrow \delta_a}$  is very close to the value predicted by the coupling numerators approximation.

The  $GC_\beta$  compensator approximately cancels the Dutch-Roll mode with a numerator of  $(s^2 + .2s + 3.5)$ . Open and closed-loop bode plots for both compensators are listed on the following pages.

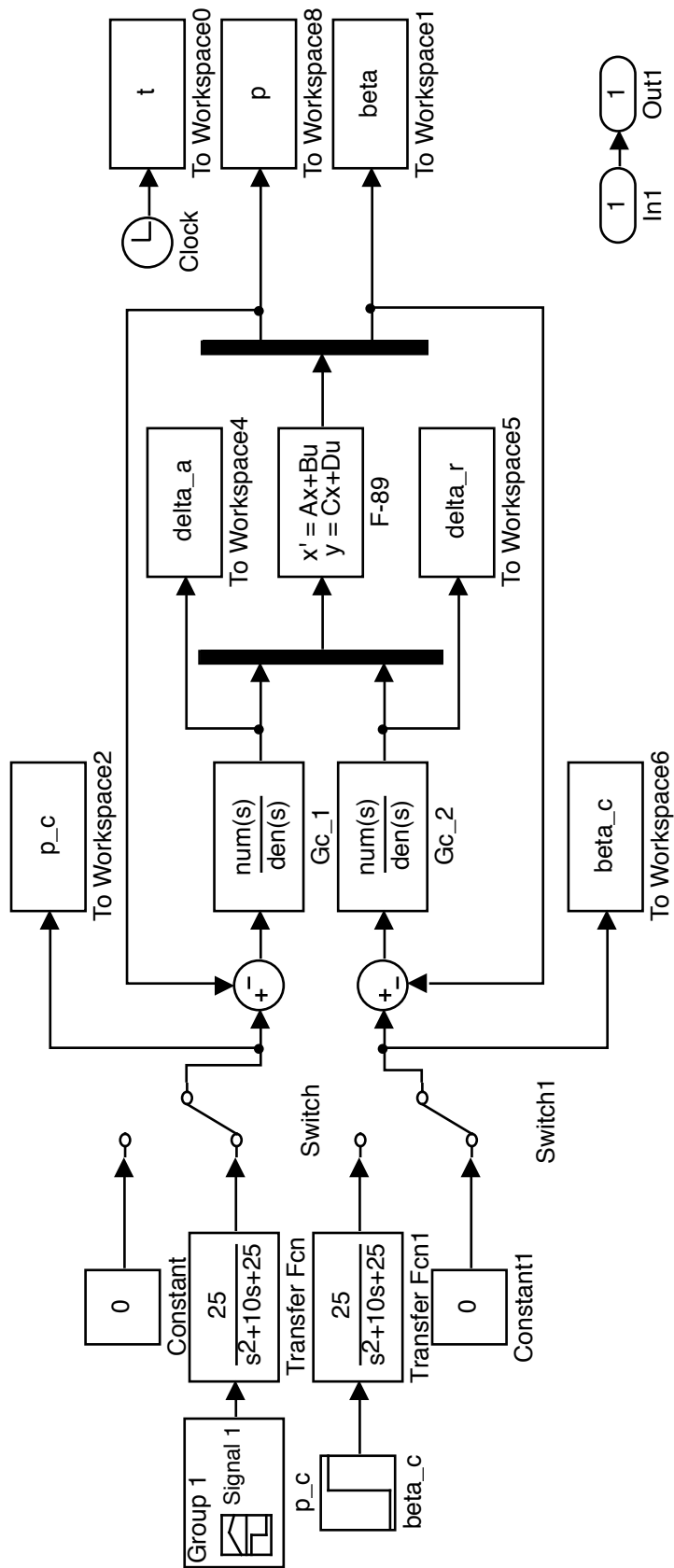


Figure 1: Final simulink diagram with both compensators

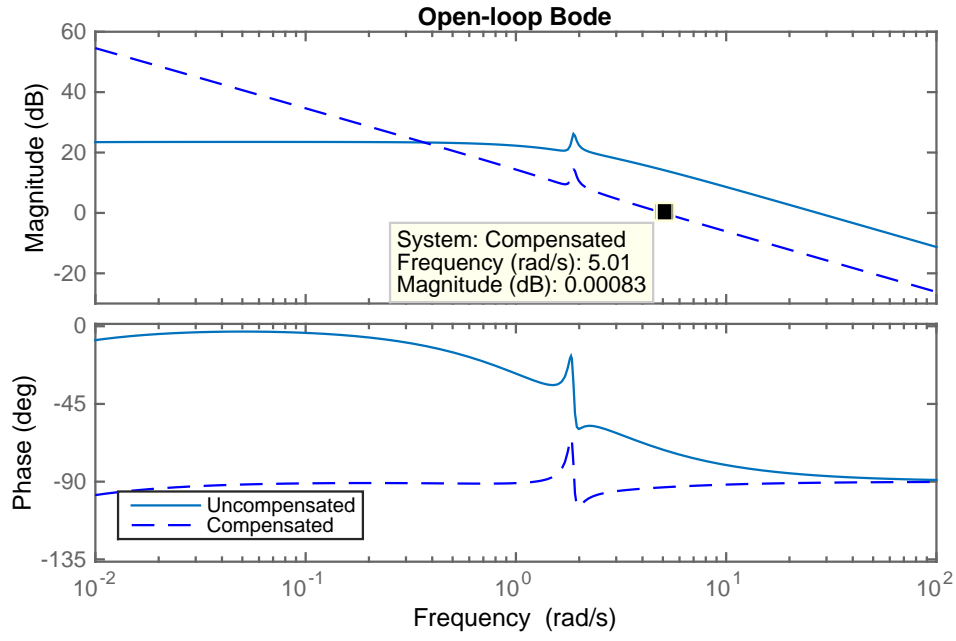


Figure 2: Open-loop Bode Plot for  $p$  loop

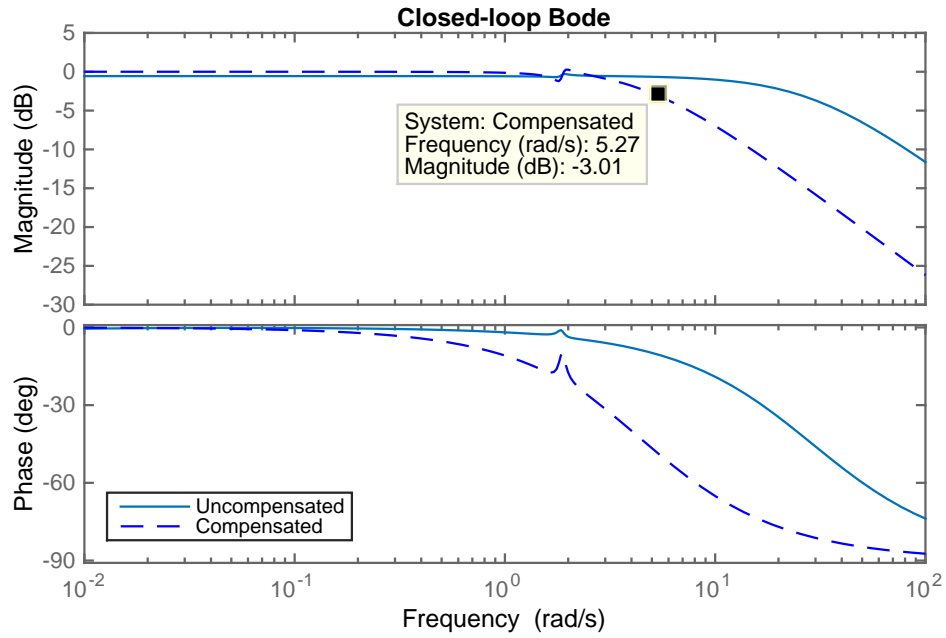


Figure 3: Close-loop Bode Plot for  $p$  loop

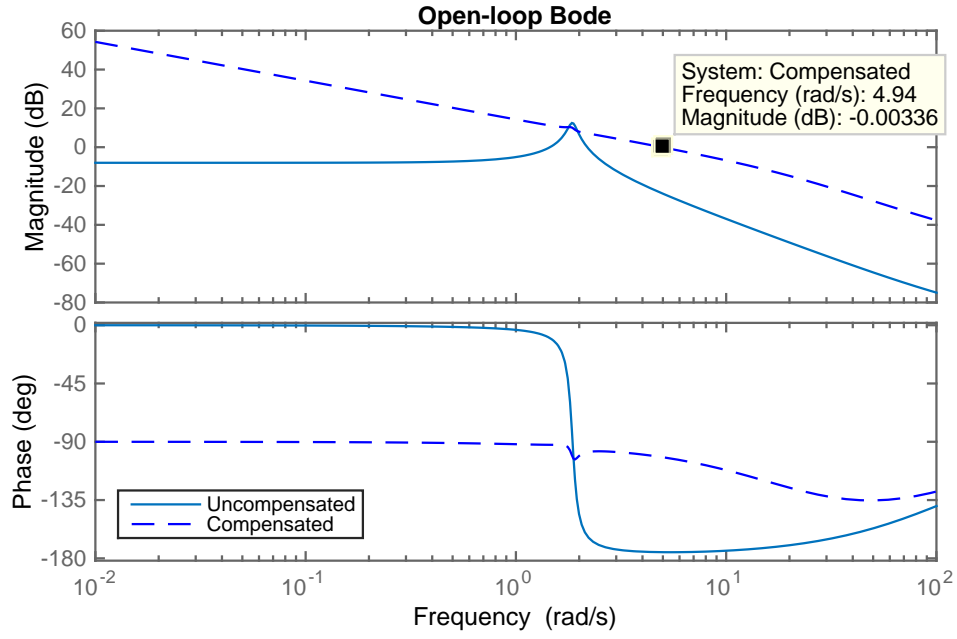


Figure 4: Open-loop Bode Plot for  $\beta$  loop

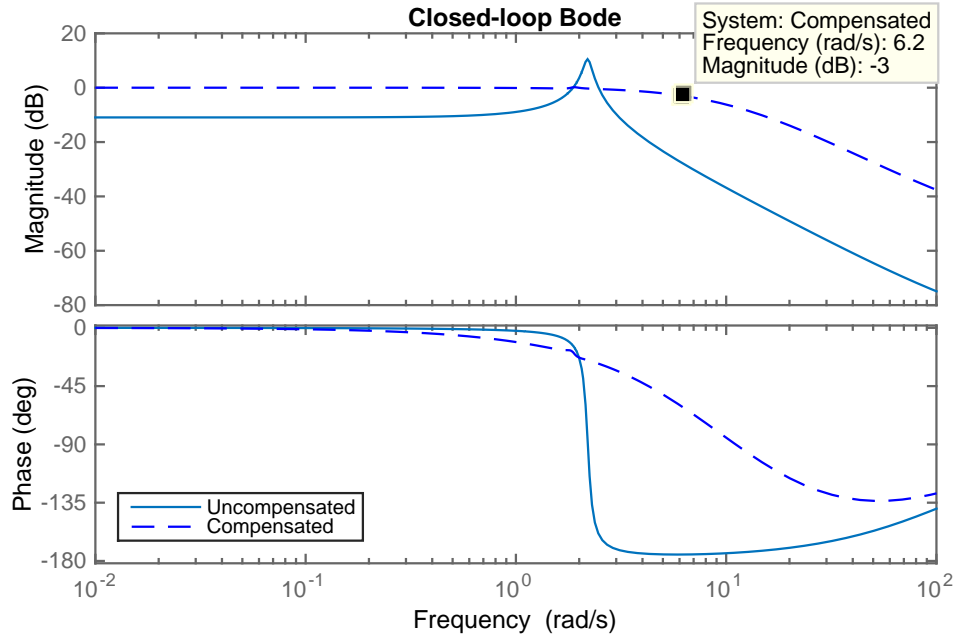


Figure 5: Close-loop Bode Plot for  $\beta$  loop

## 4 Response to Command Inputs

Two initial conditions were investigated (both commands were filtered with a filter of  $\frac{25}{(s^2+10s+25)}$ ):

1. a  $\pm 5$  deg/sec doublet with each of the two pulses lasting 2 sec for the  $p$ -loop with no command for the  $\beta$ -loop
2. a  $+5$  deg/sec command for the  $\beta$ -loop with no command for the  $p$ -loop

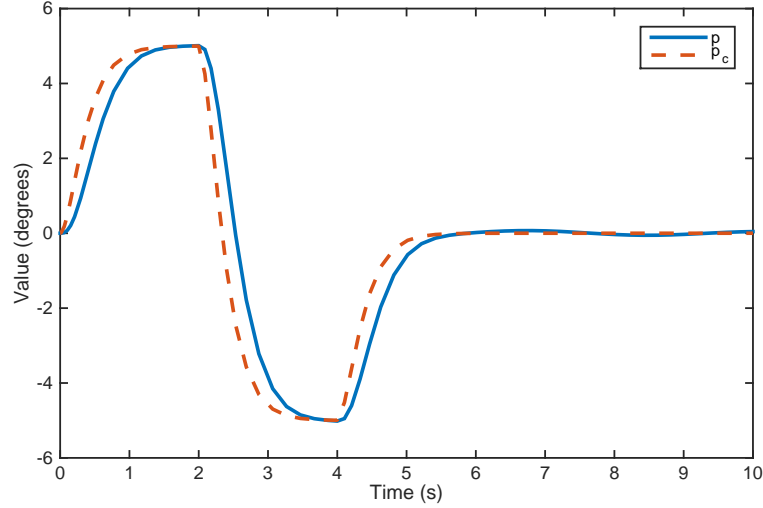


Figure 6:  $p$  Response for Scenario 1

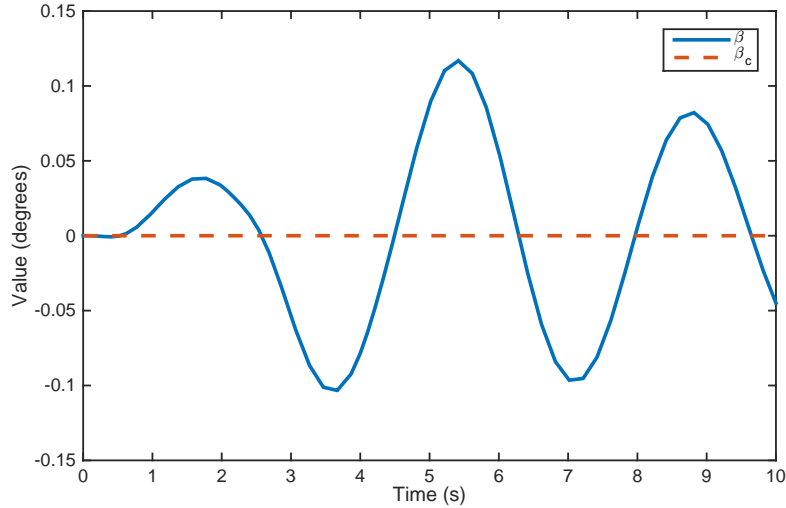


Figure 7:  $\beta$  Response for Scenario 1 (note the scale relative to the  $p$  response and decreasing magnitude of oscillation)

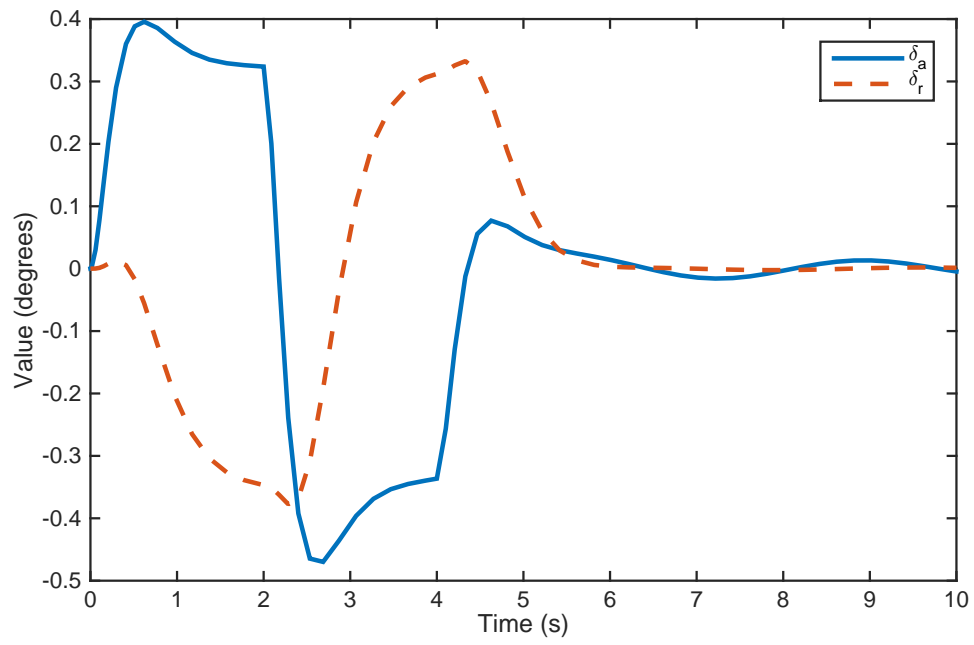


Figure 8: Command Inputs for Scenario 1

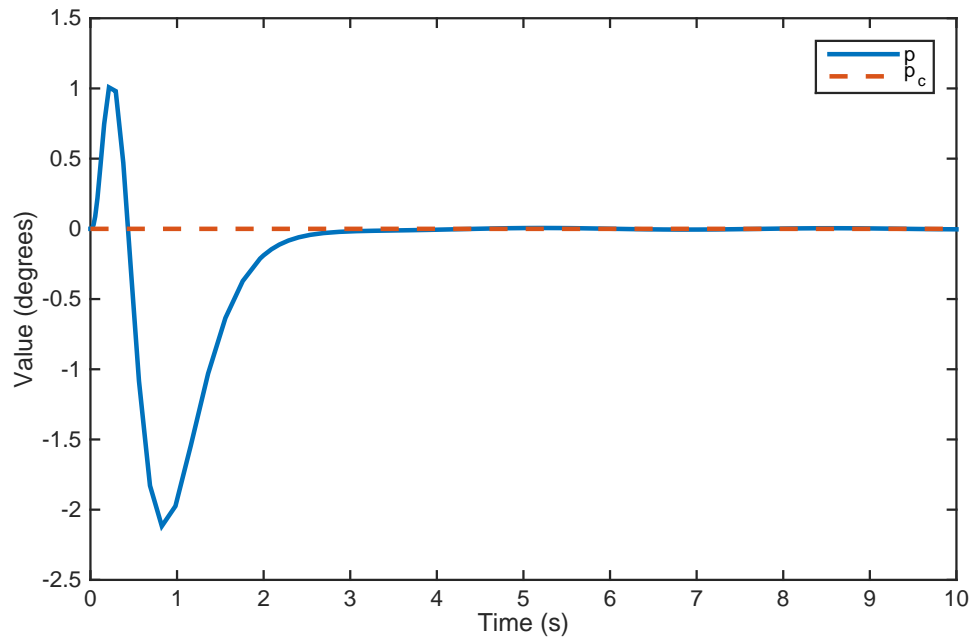


Figure 9:  $p$  Response for Scenario 2



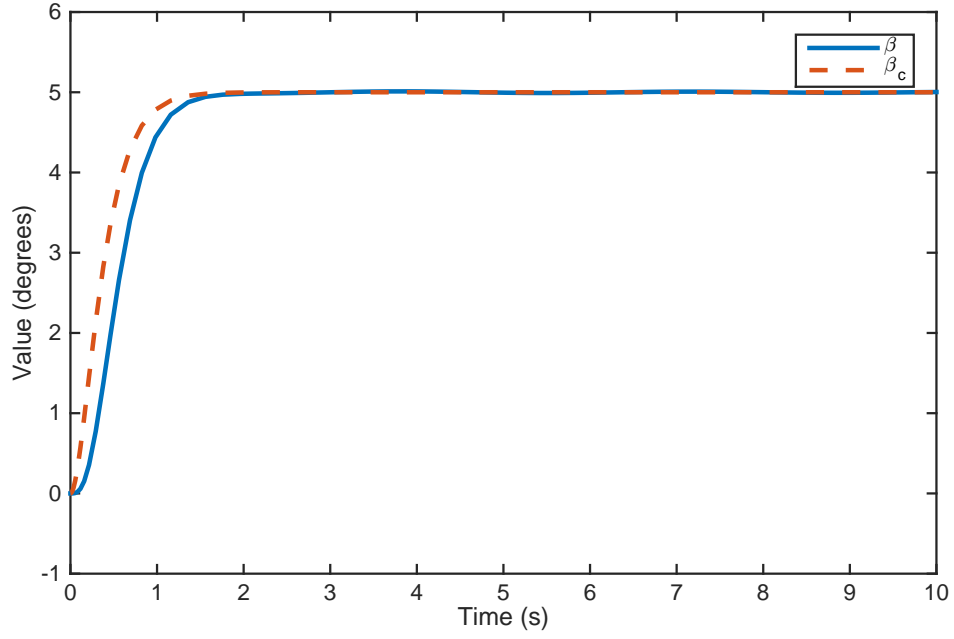


Figure 10:  $\beta$  Response for Scenario 2

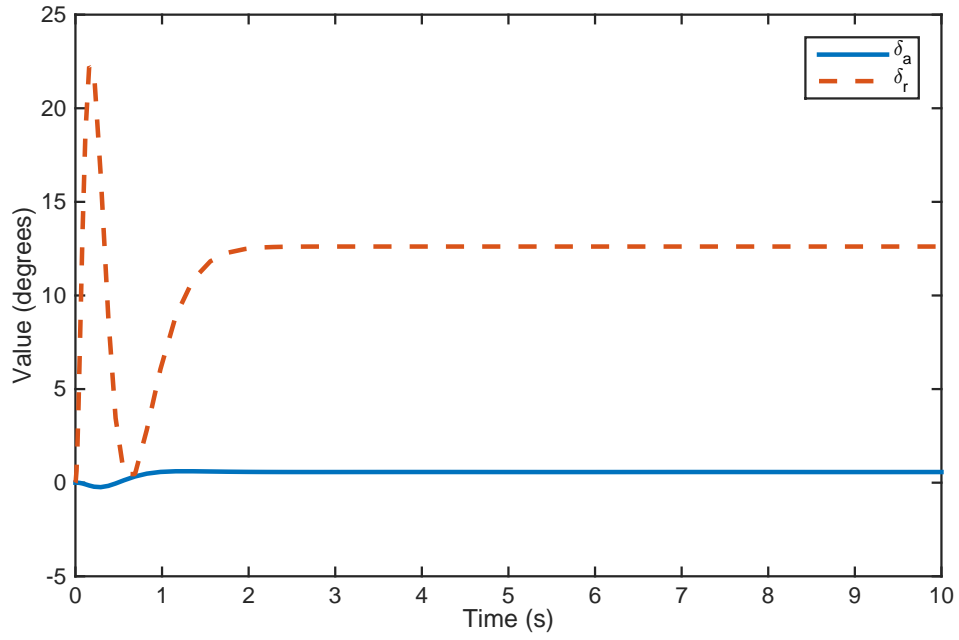


Figure 11: Command Inputs for Scenario 2