

# MAE 275 - Homework 3

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## 1 Defining the System

The longitudinal linearized aircraft equations of motion can be expressed in state space form, with state variables  $\Delta u, \Delta w, \Delta q, \Delta \theta, \Delta h$ , as

$$A = \begin{bmatrix} \frac{X_u}{Z_u} & \frac{X_w}{Z_w} & 0 & -g \cos(\theta_0) & 0 \\ \frac{1 - Z_{\dot{w}}}{1 - Z_{\dot{w}}} & \frac{1 - Z_{\dot{w}}}{1 - Z_{\dot{w}}} & \frac{Z_q + u_0}{1 - Z_{\dot{w}}} & \frac{g \sin \theta_0}{1 - Z_{\dot{w}}} & 0 \\ M_u + \frac{M_{\dot{w}} Z_u}{1 - Z_{\dot{w}}} & M_w + \frac{M_{\dot{w}} Z_w}{1 - Z_{\dot{w}}} & M_q + \frac{M_{\dot{w}}(Z_q + u_0)}{1 - Z_{\dot{w}}} & -\frac{M_{\dot{w}} g \sin \theta_0}{1 - Z_{\dot{w}}} & 0 \\ 0 & 0 & 1 & 0 & 0 \\ 0 & -1 & 0 & u_0 & 0 \end{bmatrix}$$

Plugging in the data for the VZ-4 “Doak” aircraft in Appendix A of **Aircraft Dynamics and Automatic Control** yields

$$A = \begin{bmatrix} -1.3700e - 1 & 0 & 0 & -3.2200e + 1 & 0 \\ 0 & -1.3700e - 1 & 0 & 0 & 0 \\ +1.3600e - 2 & 0 & -4.5200e - 2 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ 0 & -1 & 0 & 0 & 0 \end{bmatrix}$$

Matrices B, C, and D can also be formed

$$B = \begin{bmatrix} \frac{X_{\delta_e}}{Z_{\delta_e}} \\ \frac{1 - Z_{\dot{w}}}{1 - Z_{\dot{w}}} \\ \frac{M_{\dot{w}} Z_{\delta_e}}{1 - Z_{\dot{w}}} + M_{\delta_e} \\ 0 \\ 0 \end{bmatrix} = \begin{bmatrix} 0 \\ 1.08 \\ 1.00 \\ 0 \\ 0 \end{bmatrix}$$

$$C = [0 \quad 0 \quad 1 \quad 1 \quad 0]$$

$$D = [0]$$

## 2 System Properties

We can find the resulting transfer functions,  $\frac{q}{\delta e}(s)$  and  $\frac{\theta}{\delta e}(s)$ , with the following commands

```
1 C = eye(5);  
2 D = [0; 0; 0; 0; 0];  
3  
4 [n, d] = ss2tf(A, B, C, D);  
5 minreal(zpk(tf(n(3, :), d))) % q  
6 minreal(zpk(tf(n(4, :), d))) % theta
```

which results in

```
1      s (s+0.137)  
2      ───────────  
3 (s+0.8223) (s^2 - 0.6401s + 0.5326)
```

and

```
1      (s+0.137)  
2      ───────────  
3 (s+0.8223) (s^2 - 0.6401s + 0.5326)
```

The following MATLAB command is called to identify the characteristic roots and eigenvector elements

```
1 [v, d] = eig(A);
```

resulting in two real eigenvalues and a complex pair of eigenvalues

$$d_1 = -8.2230e - 01$$

$$d_2 = -1.3700e - 01$$

$$d_3 = +3.2005e - 01 \pm i6.5583e - 1$$

and their associated eigenvectors

$$\begin{aligned}
v_1 &= [-9.9962e - 1, \\
&\quad +0.0000e + 0, \\
&\quad +1.7494e - 2, \\
&\quad -2.1275e - 2, \\
&\quad +0.0000e + 0], \\
v_2 &= [+0.0000e + 0, \\
&\quad +1.3573e - 1, \\
&\quad +0.0000e + 0, \\
&\quad +0.0000e + 0, \\
&\quad +9.9075e - 1], \\
v_3 &= [+9.9953e - 1, \\
&\quad +0.0000e + 0, \\
&\quad +8.8107e - 3 \mp i1.5820e - 2, \\
&\quad -1.4187e - 2 \mp i2.0358e - 2, \\
&\quad +0.0000e + 0]
\end{aligned}$$

### 3 Designing the Controller

The transfer function of the system,  $VZ - 4$ , was identified as

$$\frac{(s+1) (s+0.137)}{(s+0.8223) (s^2 - 0.6401s + 0.5326)}$$

A controller,  $G_C$  was designed as

$$\frac{10000 (s+0.8223) (s+0.7)^3}{s^2 (s+500) (s+1) (s+0.137)}$$

This controller was determined using loop-shaping principles such that:

- The closed-loop system is stable

- The  $G_C$  has more poles than zeros

- The closed-loop bandwidth (-3dB criterion) is at least 5 rad/sec (23.5 rad/sec)

- ~~The gain margin is at least 6dB (-24 dB)~~

- The phase margin is at least 40 deg (80 deg)

- There is zero steady-state error to a step input  $\theta_C$

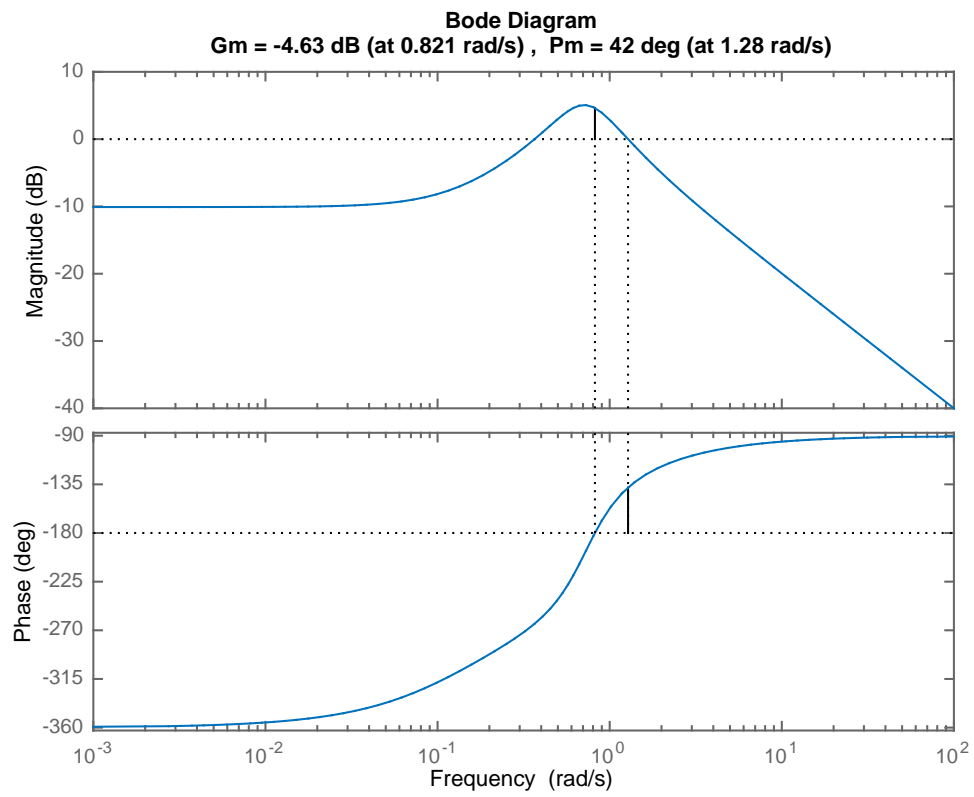


Figure 1: Open Loop Bode Diagram

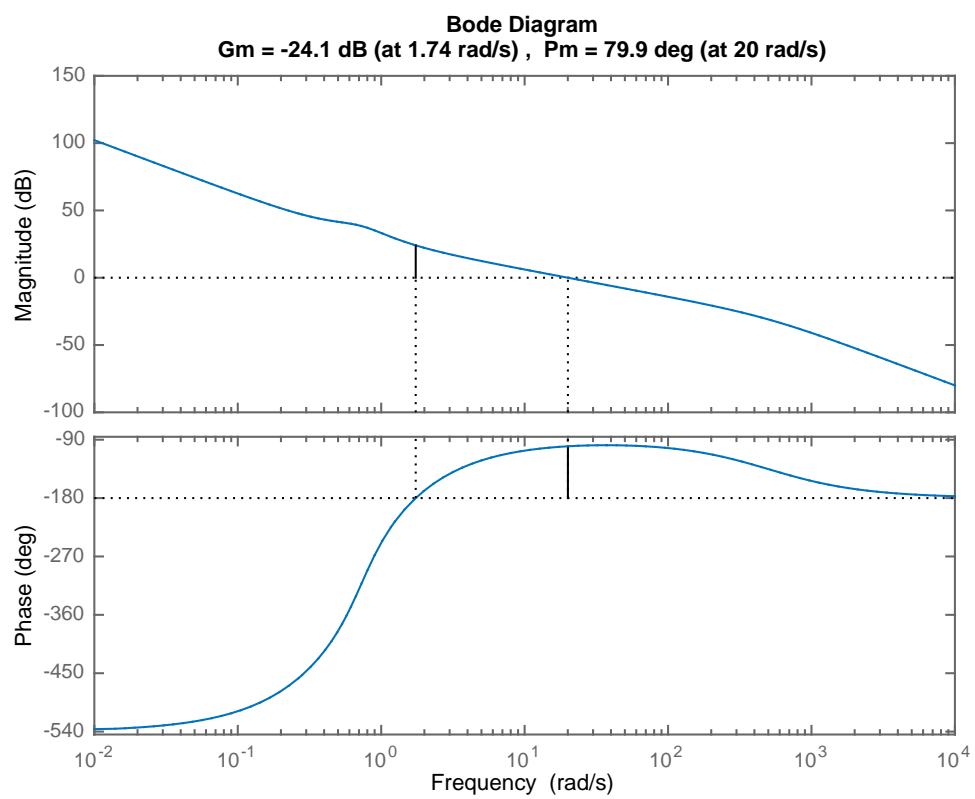


Figure 2: Compensated Open Loop Bode Diagram

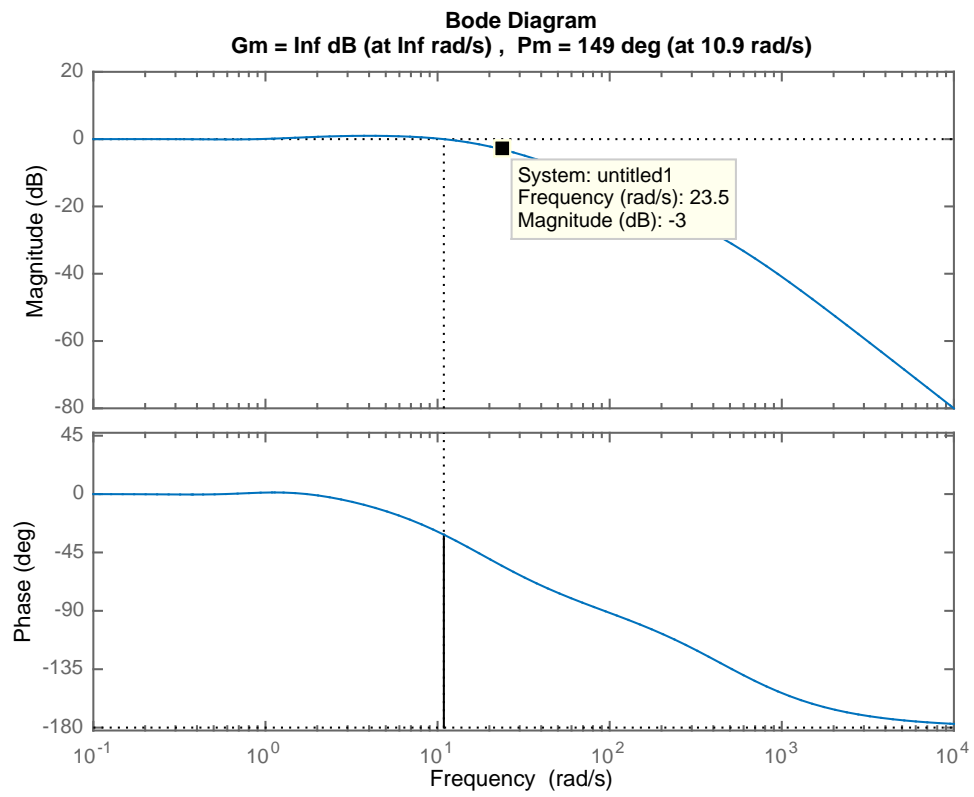


Figure 3: Compensated Closed Loop Bode Diagram

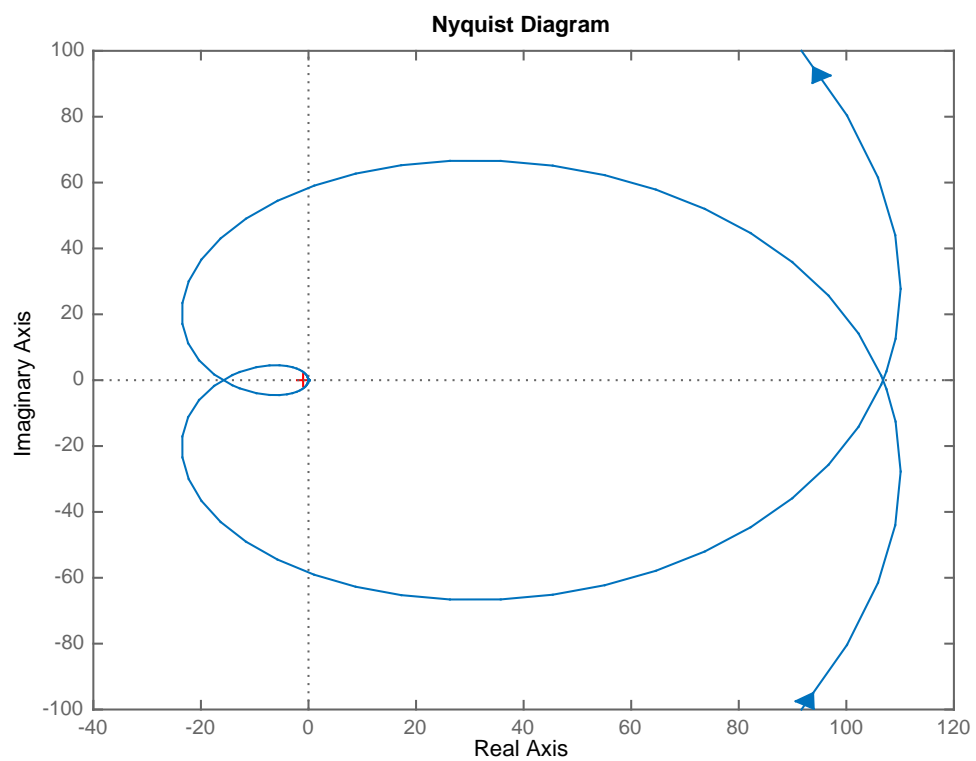


Figure 4: Nyquist Diagram



## 4 Simulink Diagram and Step Response

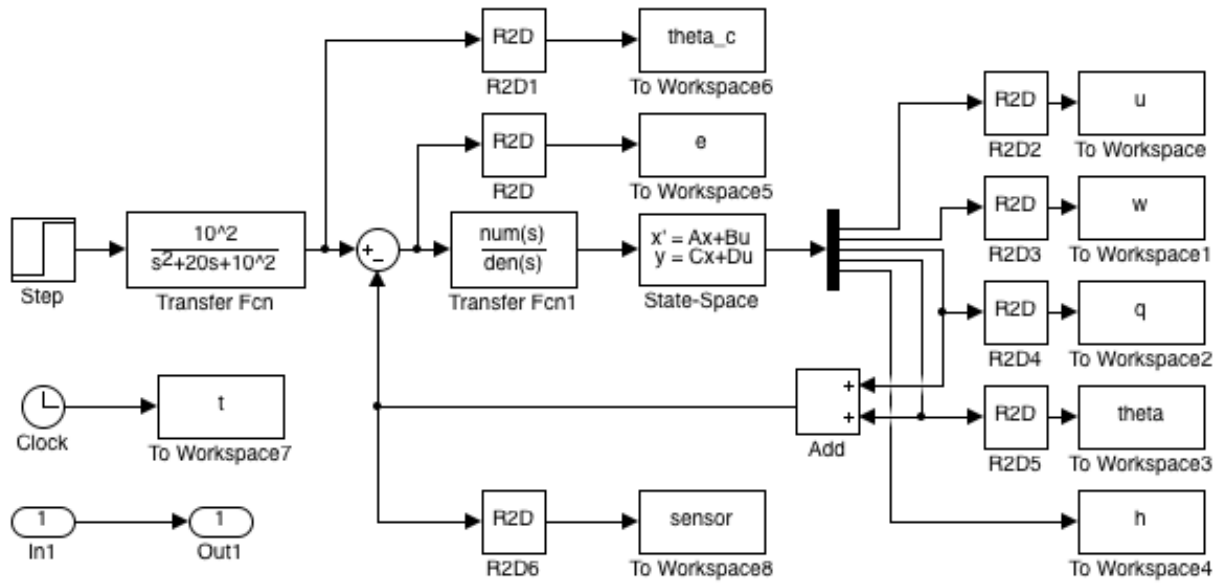


Figure 5: Simulink Diagram

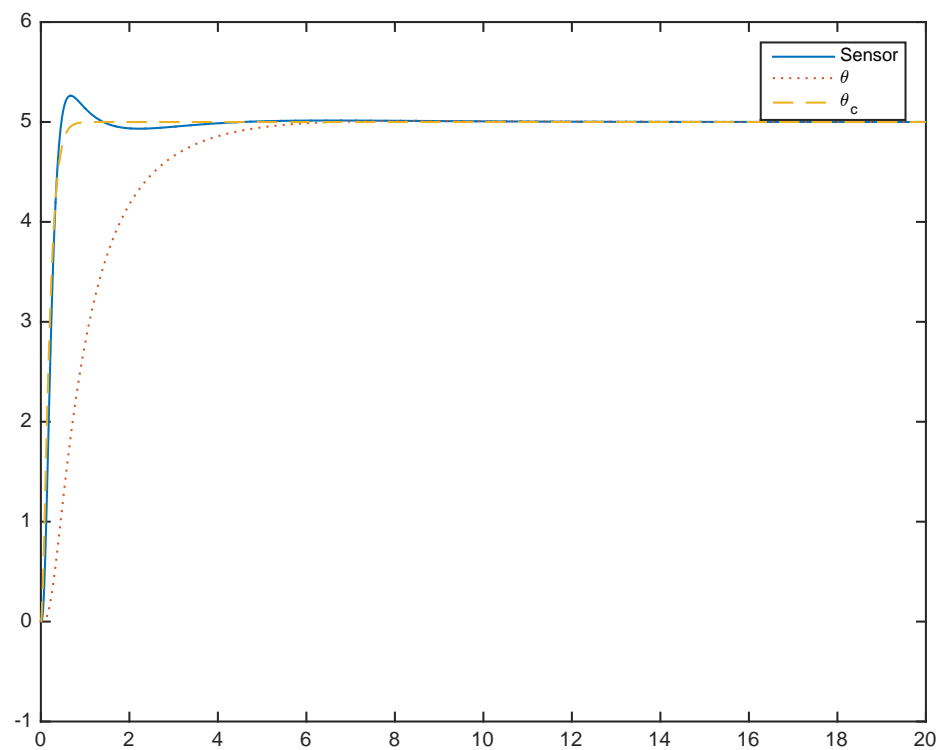


Figure 6: Step Response