Progress Report

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1 Design of Digit-parallel Online Adder

2 Design and Optimization of A Digit-parallel Online Multiplier

2.1 Online Multipilication Algorithm

Typically the online multiplication is performed in a recursive manner. The algorithm to generate 1-digit product z_j is given by Algorithm 1, where j is the number of iteration, r denotes the radix and N represents the word-length of input operands.

Algorithm 1 Online Multiplication

Initialization: $X[-\delta] = Y[-\delta] = P[-\delta] = 0$

Recurrence: for $j=-\delta,\ -\delta+1,\ \cdots,\ 1,\ 2,\ \cdots,\ N-1$ do

$$H[j] = r^{-\delta} (x_{j+\delta} \cdot Y[j+1] + y_{j+\delta} \cdot X[j])$$

$$W[j] = P[j] + H[j]$$

$$z_{j} = sel(\widehat{W}[j])$$

$$P[j+1] = r(W[j] - z_{j})$$
(1)

The operands and product are signed number in the range (-1,1), and they can be represented with signed digits from the set $\{-a, \dots, a\}$. For instance in a radix-2 online multiplier, the most common digit set is $\{-1,0,1\}$. The operands and product at iteration j can be represent by (2).

$$X[j] = \sum_{i=0}^{j+\delta-1} x_i r^{-i-1}, \quad Y[j] = \sum_{i=0}^{j+\delta-1} y_i r^{-i-1}, \quad Z[j] = \sum_{i=0}^{j-1} z_i r^{-i-1}$$
 (2)

In addition, $sel(\widehat{W}[j])$ is the function to select the product digit at iteration j. The selection is accomplished by using a discretization algorithm [xxx] based on the knowledge of $\widehat{W}[j]$, which is an estimate of the residue W[j]. In the following of this article, $\widehat{W}[j]$ is obtained by truncating W[j] to its first few MSDs.

2.2 Radix-2 Online Multiplier

Currently we consider the online multiplication algorithm in radix-2 as it is the most commonly used situation in computer arithmetic. In a radix-2 online multiplier, the design choices are summarized in Table 1 [xxx].

Table 1: Design Choice for Radix-2 Online Multiplication.

\overline{r}	Digit Set	$\widehat{W}[j]$	δ
2	$\{-1,0,1\}$	$w_{-2}w_{-1}.w_0$	3

Hence (1) can be modified as given by (3).

$$H[j] = 2^{-3}(x_{j+4} \cdot Y[j+1] + y_{j+4} \cdot X[j])$$

$$W[j] = 2P[j] + H[j]$$

$$z_{j} = sel(\widehat{W}[j])$$

$$P[j+1] = W[j] - z_{j}$$
(3)

The selection function is given by (4) [xxx].

$$sel(\widehat{W[j]}) = \begin{cases} 1 & \text{if } \widehat{W}[j] \geqslant \frac{1}{2} \\ 0 & \text{if } -\frac{1}{2} \leqslant \widehat{W}[j] < \frac{1}{2} \\ -1 & \text{if } \widehat{W}[j] < -\frac{1}{2} \end{cases}$$
(4)

3 Probabilistic Model of Overclocking Error in Online Multiplier

3.1 Annihilation of the Propagation Chain

While the delay in a digit-parallel online multiplier might be derived from many sources such as the computation delay to generate outputs, the overall delay will eventually be determined by the longest propagation delay between stages with increasing operand word-lengths. Let the propagation delay between two adjacent stages in a N-digit olMult be dented by μ , hence the delay of the longest chain which propagates from the MSD to the LSD is given by d_w as shown in (5).

$$d_w = (N + \delta - 1) \cdot \mu \tag{5}$$

However, we note that the chain is annihilated at a certain stage if the propagation inputs of this stage change while the propagation outputs keep stable. This will shrink the value of d_w , and there are two possible cases specifically. As an example for the first case, assume at time t ($t > \mu$) the value of propagation inputs and outputs of a stage S_i to be Pin(t) and Pout(t) respectively. After delay μ , the input value changes to $Pin(t + \mu)$ and stabilizes thereafter, while the output of this stage remains Pout(t). Hence for the next stage S_{i+1} , the input of which varies from $Pout(t - \mu)$ to Pout(t), as illustrated in Figure.xxx. Under this situation, the chain delay to S_{i+1} is reduced by μ because of the annihilation.

For the second case, the current chain will be completely annihilated and a new chain will be generated at a given stage if Pout(t) = Pout(0) for $t = \mu$, 2μ , \cdots . Therefore the worst case delay is given by (6) where d_p denotes the delay of the p^{th} propagation chain.

$$d_w' = \max(d_1, d_2, \cdots) \tag{6}$$

In the following of this section, detailed analysis for both cases will be described and the worst-case delay of the olMult will be discussed.

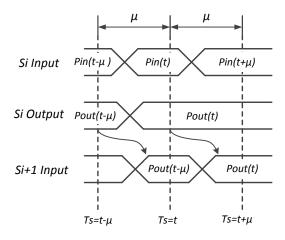


Figure 1: Timing wave.

3.2 Worst-case Analysis in Online Multiplier

3.2.1 Two Observations

From (3) two observations can be made under the assumption that all signals are reset to 0 initially.

Observation 1. The two integer bits of W[j] and 2P[j+1] are either 00 or 11.

This observation can be justified by contradiction. All the combinations of $\widehat{W[j]}$ and the corresponding z_j , the most significant 3 bits of P[j+1] and the 2 integer bits of 2P[j+1] are listed in Table 2. For j=-3, we have (7) according to (3).

$$\begin{cases}
W[-3] = 2^{-3}(y_1 \cdot X[-3]) \\
H[-2] = W[-3]
\end{cases}$$
(7)

Hence $\widehat{W[-3]}$ is either 11.1 or 00.0, with the corresponding 2P[-2] being 00 or 11 which is the propagation input for the next stage. For j > -3, the 3 MSBs of H[j] in (3) is either 00.0 or 11.1 due to the shift of binary point. Also as seen in Table.xxx, if the two integer bits of $\widehat{W[j]}$ are identical, the integer bits of 2P[j+1] will be the same. Therefore the propagation of 2P[j+1] will not generate diverse bits in the integer part of W[j].

Observation 2. P[j+1] will not change if only the integer bits of W[j] change to a new value.

According to Observation 1, there are only four possible cases for the change of W[j] as summarized in Table.xxx. This observation describes the situation where a propagation chain annihilates but not necessarily generating a new chain, because 2P[j+1] may not maintain its initial value at all times.

3.2.2 Annihilation of the Propagation Chain in An olMult

As stated in Section 3.1, the annihilation of propagation chain would leads to a reduction of the propagation delay. It is worthwhile exploring the conditions of occurrence of Observation 2.

Table 2: All the combinations of $\widehat{W[j]}$ and the corresponding z_j , P[j+1] and 2P[j+1].

$\widehat{W[j]}$	$ z_j $	P[j+1] (3 MSBs)	2P[j+1] (2 MSBs)
00.0	0	0.00	00
00.1	1	11.1	11
01.0	1	0.00	00
01.1	1	00.1	01
10.0	$\bar{1}$	11.0	10
10.1	Ī	11.1	11
11.0	1	0.00	00
11.1	0	11.1	11

Table 3: Ob2.

$\widehat{W[j]}$	$ z_j $	P[j+1] (3 MSBs)	2P[j+1] (2 MSBs)
$00.0 \to 11.0$ $00.1 \to 11.1$	$0 \rightarrow \bar{1}$	0.00	00
$00.1 \rightarrow 11.1$	$1 \rightarrow \bar{0}$	11.1	11
$11.0 \rightarrow 00.0$	$\bar{1} \rightarrow 0$	0.00	00
$11.1 \rightarrow 00.1$	$0 \rightarrow 1$	11.1	11

For instance in a 3-digit olMult, there are 6 stages in total, as shown in Fig.xxx. If $x_{-3} \neq 0$ and/or $y_{-3} \neq 0$, no more than 6 MSBs of P[-2] will change with respect to the initial value at t=0. This is because in (7) $y_1 \cdot X[-3]$ contains only 1 fractional bit and 2 integer bits. Hence maximally 5 MSBs of $2P_{[-2]}(0)$ will be updated and thereafter stabilized. It is worth noting that this is irrelevant to the word-length of the olMult. Additionally as assumed previously in the timing model, at t=0 the computation of H[j] ($j \in [-2,2]$) in (3) is finished. This results in $2P_{[j]}(0)$, which will be updated sequentially with a delay of μ , triggered by the propagation of $2P_{[-2]}(0)$. During each propagation, the number of the altered MSBs will be reduced by 1, because of the shifting of the binary point in 2P[j]. For example, $2P_{[-1]}(t)$ stabilizes when $t=\mu$, and the altered MSBs for $2P_{[-1]}(0) \rightarrow 2P_{[-1]}(\mu)$ are 4 bits. This process repeats until S_1 , of which the changed MSBs for $2P_{[1]}(3\mu) \rightarrow 2P_{[1]}(4\mu)$ are 2 bits. According to Observation 2, the chain triggered by $2P_{[-2]}(0)$ is annihilated at S_1 . Therefore the propagation delay from S_{-3} to S_2 is 4μ .

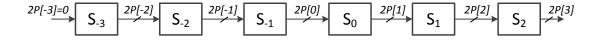


Figure 2: Propagation chain in 3-digit olMult

It is worth noting that in this example, the chain still propagates through S_2 without generating a new chain. Because it is hardly to achieve $2P_{[2]}(t) \equiv 2P_{[2]}(0)$ with $t = \mu$, 2μ , \cdots , due to the propagation of $2P_{[j]}(t)$ where $j \in [-2,1]$. The maximum number of the changed MSBs for $2P_{[2]}(3\mu) \rightarrow 2P_{[2]}(4\mu)$ is 3, because this is actually triggered by $2P_{[-1]}(0)$, which varies 6 MSBs in comparison to its initial value. Likewise there will be maximally 4 MSBs

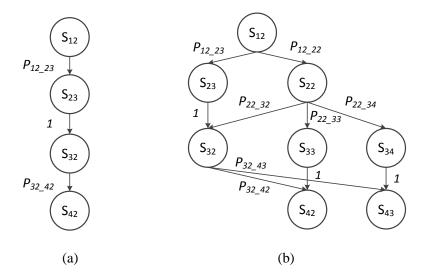


Figure 3: State transition graph in a 5-digit olMult when $2P_{[-2]}(0) \neq 0$. (a) One possible transition path from Stage1 to Stage4 with transition probabilities labelled on the edges. (b) All possible transition paths.

changing for $2P_{[3]}(4\mu) \to 2P_{[3]}(5\mu)$.

The above changing process happens in a 3-digit olMult when $P[-2] \neq 0$, which is only determined by the inputs of the first stage regardless of the other stages. Actually in other situations, the overall propagation delay will not be increased. For example for $t=0, \mu, \cdots$ if $P_{[-2]}(t)=0$ while $P_{[-1]}(t)\neq 0$, the chain starts from $2P_{[-1]}(0)$. Based on a similar analysis, the chain will finally annihilate at S_2 , of which we have 2 MSBs changing for $2P_{[2]}(3\mu) \to 2P_{[2]}(4\mu)$, yielding a delay of 4μ . This value shrinks to 3μ if $P_{[-2]}(t)=P_{[-1]}(t)=0$ while $P_{[0]}(t)\neq 0$ for $t=0, \mu, \cdots$.

3.2.3 State Transition Graph in An olMult

For an olMult with a greater word-length, the aforementioned process is still valid. Annihilation happens at S_i when there are only 2 MSBs of $2P_{[i]}(t)$ update before it is stabilized. Otherwise the propagation continues regardless of the inputs x_{j+4} , y_{j+4} , X[j] and Y[j+1] of S_j . The propagation process from S_1 can be illustrated using a state transition graph. One possible transition path from S_1 in a 5-digit olMult is shown in Fig.xxx, where a state S_{mn} denotes the propagation input of $Stage\ m$ changes n MSBs prior to stabilization. The transition probabilities are labelled on the edges. For instance $P_{12,23}$ denotes the conditional probability $P(S_{23} \mid S_{12})$. Note that $P_{23,32} = 1$ because annihilation only happens when n = 2 in S_{mn} . In this case, the probability of this transition path is given by (8), where $P(S_{12})$ is determined by the value of x_1 and y_1 because S_{12} will happen when $2P_{[-2]}(0) \neq 0$.

$$P_{trans} = P(S_{42}, S_{32}, S_{23}, S_{12})$$

$$= P(S_{42} \mid S_{32}, S_{23}, S_{12})P(S_{32} \mid S_{23}, S_{12})P(S_{23} \mid S_{12})P(S_{12})$$

$$= P_{32_42} \cdot 1 \cdot P_{12_23} \cdot P(S_{12})$$
(8)

The delay of this transition path can also be determined by counting the number of states

 S_{m2} . First, there is no annihilation prior to S_{12} . Second, S_{42} is excluded because it is not necessary to wait for the propagation output of this stage to generate before sampling z_4 . Hence there are totally 2 annihilation states S_{12} and S_{32} in this path. According to (5) the worst case delay without any annihilation is $d_w = 7\mu$. Delay of a specific transition path is given by (9). Hence the delay of this path is 5μ .

$$d = d_w - \mu \cdot AnnihilationNo. \tag{9}$$

Additionally as discussed earlier, there might be other possible states when annihilation happens. The complete transition graph of a 5-digit olMult is shown in Fig.xxx.

- 3.3 Probability of Overclocking Error in Online Multiplier
- 3.4 Magnitude of Overclocking Error in Online Multiplier
- 4 Probabilistic Model of Truncation Error in Online Operators