

# Relation between Fourier Series & transform.

(DSAA)

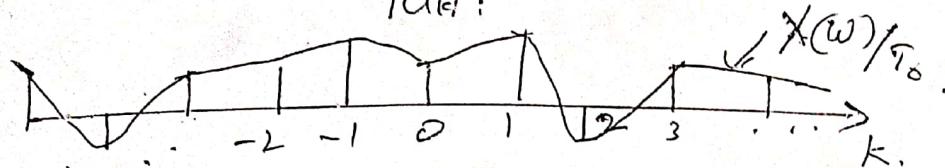
M3

Fourier series:

$$\tilde{x}(t) = \sum_{k=-\infty}^{\infty} a_k e^{j k \omega_0 t}. \rightarrow \text{Synthesis}$$

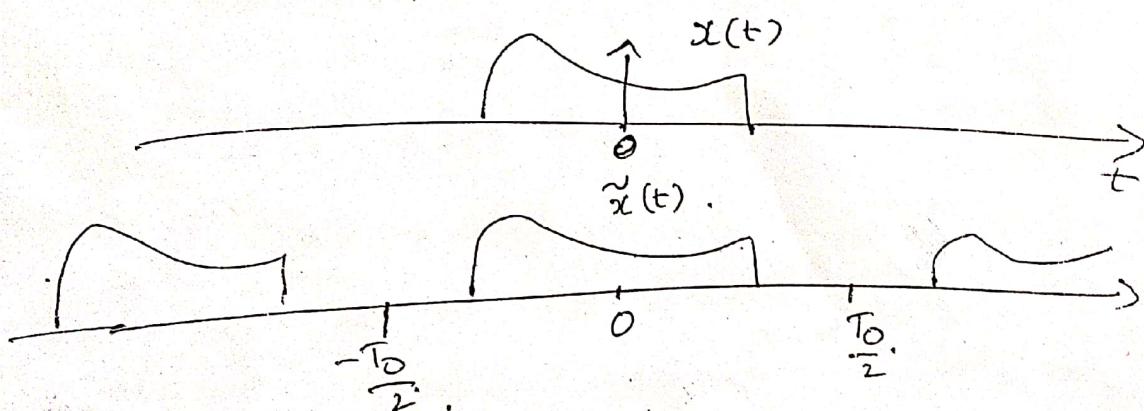
$$a_k = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} \tilde{x}(t) e^{-j k \omega_0 t} dt. \rightarrow \text{analysis.}$$

$|a_k| :$



Consider  $x(t)$ .  $\rightarrow$  aperiodic signal:

Let  $\tilde{x}(t) \rightarrow$  periodic form of  $x(t)$ .



Now.. clearly  $\tilde{x}(t) = x(t)$ , for  $|t| < \frac{T_0}{2}$ . — (2)

$$\text{Now } a_k = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} \tilde{x}(t) e^{-j k \omega_0 t} dt.$$

(2) in (1)

$$a_k = \frac{1}{T_0} \int_{-\infty}^{\infty} x(t) e^{-j k \omega_0 t} dt.$$

Note limits change  
 $x(t) \rightarrow 0$  outside  $t \in [-T_0/2, T_0/2]$

Above  $T_0 = \frac{2\pi}{\omega_0}$

Title: FT  $\rightleftharpoons$  FS - Topic: Signals & Systems Author: A.V.O Section: 4.4.1 Date: 03/01/15

TH

Let us denote the envelope of  $a_k T_0$  as  $X(\omega)$ .

$$\text{then } X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt. \quad \text{--- (3)}$$

↑ envelope.

$$\text{i.e. } x(k\omega_0) = a_k T_0 = \int_{-\infty}^{\infty} x(t) e^{-jk\omega_0 t} dt.$$

i.e;  $\omega$  replaces  $k\omega_0$  for envelope.

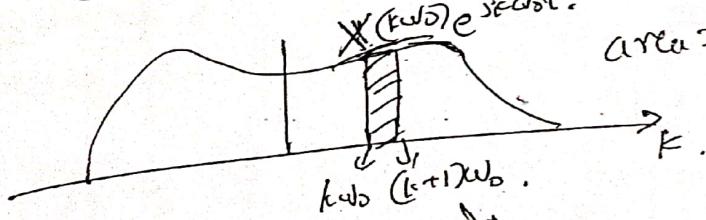
For synthesis ; since  $a_k = \frac{1}{T_0} \downarrow x(k\omega_0) \rightarrow$  a sample form of  $x(\omega)$ .

$$\text{From (1)} \quad \tilde{x}(t) = \sum_{k=-\infty}^{\infty} \frac{1}{T_0} x(k\omega_0) e^{jk\omega_0 t}$$

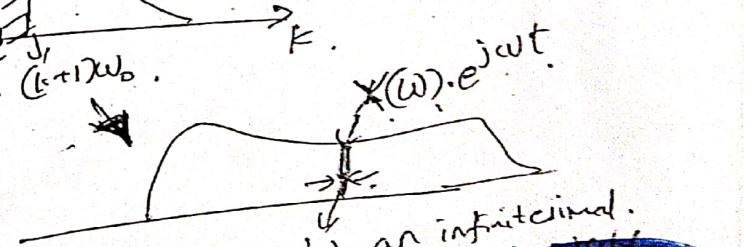
$$\text{Next } T_0 = \frac{2\pi}{\omega_0} \Rightarrow \tilde{x}(t) = \sum_{k=-\infty}^{\infty} \frac{\omega_0}{2\pi} x(k\omega_0) e^{jk\omega_0 t} \quad \text{--- (4)}$$

Now as  $T_0 \rightarrow \infty \Rightarrow \omega_0 \rightarrow 0$ .

$\tilde{x}(t) \rightarrow x(t)$ .  
Next ~~sample~~ term in (4)  $x(k\omega_0) e^{jk\omega_0 t}$  can be seen as



Now as  $\omega_0 \rightarrow 0$



Now, area =  $x(\omega) e^{j\omega t} d\omega$ . and  $\sum_{-\infty}^{\infty} \rightarrow \int_{-\infty}^{\infty}$    
*(infinitesimal)*

$$\therefore x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} x(\omega) e^{j\omega t} d\omega.$$

Total area ..

Ex. 3.5

$$x_2(t) = x_1(1-t) + x_1(t-1). \quad \text{--- } ①$$

Given  $x_1(t) \leftrightarrow a_k$  with freq.  $\omega_1$ .

goal @ find  $\omega_2$  the freq. of  $x_2(t)$ .

and ②  $x_2(t) \leftrightarrow b_k$ . find  $b_k$  in terms of  $a_k$ .

(a) freq. of  $x_2(t)$ .

$$\text{From eq. } x_2(t) = x_1(1-t) + x_1(t-1)$$

$$\text{Since period of } x_1(t) = T_1 = \frac{2\pi}{\omega_1}.$$

period of  $x_1(1-t)$  is also  $T_1$ .

$$\text{Proof: } x_1(1-t+T_1) = x_1(t+T_1) \\ = x_1(t) = x_1(1-t)$$

(let  $t=1-t$ )  
period of  $x_1(t-1)$  is also  $T_1$ . (delay does not alter period.)

$T_1 \rightarrow \text{Lcm of } (T_1, T_1)$ .

$$\therefore \omega_2 = \frac{2\pi}{T_1} = \omega_1.$$

$$(b) \text{ Let } b_k = \frac{1}{T_1} \int_{T_1} x_2(t) e^{-jk\omega_2 t} dt.$$

$$= \frac{1}{T_1} \int_{T_1} x_1(1-t) e^{-jk\omega_2 t} dt + \frac{1}{T_1} \int_{T_1} x_1(t-1) e^{-jk\omega_2 t} dt$$

TQ

$$T_1 \quad \int_{T_1}^1 x_1(r-t) e^{-jk\omega_0 t} dt$$

let  $l-t = t'$  | integration  
 $dt = -dt'$  | limits are reversed.  
 $l \rightarrow 1-T_1$ .

$$= \frac{1}{T_1} \int_{1-T_1}^1 x_1(t') e^{-jk\omega_0(1-t')} dt'.$$

$$= \frac{1}{T_1} e^{-jk\omega_0} \int_{T_1}^1 x_1(t') e^{-jk\omega_0 t'} dt'.$$

$$\therefore T_0 \quad e^{-jk\omega_0} a_{-k}.$$

from time-shifting

$$T(2) = \bar{e}^{jk\omega_0} a_k.$$

$$\therefore b_{lc} = \bar{e}^{-jk\omega_0} a_{-lc} + \bar{e}^{jk\omega_0} a_{lc}.$$

Ex. 3.6.

(a). Verify if the signal is real or not.

$$x_1(t) = \sum_{k=0}^{100} \left(\frac{1}{2}\right)^k e^{jk\frac{2\pi}{50}t}.$$

real signal  $a_k^* = a_{-k}$ .

$$a_1 = \frac{1}{2}, \quad a_{-1} = 0 \quad \therefore x_1(t) \text{ is not real.}$$

$$x_2(t) = \sum_{k=-100}^{100} \cos(k\pi) e^{-jk\frac{2\pi}{50}t}.$$

$$a_{|k|} = \cos k\pi. \quad a_{-k} = \cos(-k\pi) \\ = \cos(k\pi) \\ = a_k. \quad \checkmark$$

$\therefore x_2(t) \rightarrow \text{real.}$

$$x_3(t) = \sum_{k=-100}^{100} j \sin \frac{k\pi}{2} e^{-jk\frac{2\pi}{50}t}$$

$$a_{|k|} = j \sin \frac{k\pi}{2}. \quad a_{-k} = -j \sin \frac{|k\pi|}{2}.$$

$$\therefore a_k^* = a_{-k} \quad \therefore$$

$x_3(t) \rightarrow \text{real.}$

Ex 3-7.

Given

$$x(t) \leftrightarrow a_k \quad a_0 = 2.$$

or

$$T = 2.$$

Find

$$g(t) \leftrightarrow b_k$$

$$g(t) = \frac{d}{dt} x(t).$$

$$x(t) = \sum_k a_k e^{jk\omega_0 t} + \sum_k a_k e^{-jk\omega_0 t}.$$

$$\frac{d}{dt} x(t) = \sum_k a_k (jk\omega_0) \cdot e^{-jk\omega_0 t}, \quad b_0 = 0.$$

$$\therefore b_k = jk\omega_0 a_k.$$

Ex 3-8 Find a signal that satisfies.

①  $x(t)$  is real and odd

②  $x(t)$  is periodic  $T = 2$ .

$$x(t) \leftrightarrow a_k$$

③  $a_k = 0 \quad |k| > 1$

④  $\frac{1}{2} \int_0^2 |x(t)|^2 dt = 1.$

Sol:  $x(t) = 2 \sin \pi t$

$$a_1 = -j \quad a_{-1} = j$$

$$\frac{1}{2} \int_0^2 4 \cdot |\sin^2 \pi t|^2 dt = 2 \cdot \frac{1}{2} = 1.$$

Continuous Fourier transforms.  
(time).

$$1. \quad x(t) = e^{j\omega_0 t} \quad \forall t$$

since  $\int_{-\infty}^{\infty} |x(t)| dt \rightarrow \infty$

direct def. Cannot be evaluated.

$$\text{Consider } X(\omega) = 2\pi \delta(\omega - \omega_0).$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega.$$

$$= \frac{1}{2\pi} \int_{-\infty}^{\infty} 2\pi \delta(\omega - \omega_0) e^{j\omega t} d\omega$$

$$\Rightarrow x(t) = e^{j\omega_0 t}.$$

~~$\delta(\omega)$~~   $\quad e^{j\omega_0 t} \neq 2\pi \delta(\omega - \omega_0).$

$$2. \quad x(t) = \sin \omega_0 t.$$

here  $x(t) = \text{Im} \left\{ e^{j\omega_0 t} \right\} = \frac{e^{j\omega_0 t} - e^{-j\omega_0 t}}{2j}$

$$\therefore F\{x(t)\} = f \left\{ \frac{1}{2j} (e^{j\omega_0 t} - e^{-j\omega_0 t}) \right\}$$

$$\therefore X(\omega) = \frac{2\pi}{2j} \left[ \delta(\omega - \omega_0) - \delta(\omega + \omega_0) \right]$$

$$= j\pi \left[ \delta(\omega + \omega_0) - \delta(\omega - \omega_0) \right]$$

$$3. \cos \omega_0 t \stackrel{FT}{\rightarrow} \frac{1}{2} [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)].$$

$$4. x(t) = e^{-at} v(t) \quad a > 0.$$

$$X(\omega) = \int_{-\infty}^{\infty} e^{-at} v(t) e^{-j\omega t} dt.$$

$$= \int_0^{\infty} e^{-(a+j\omega)t} dt$$

$$= \left. \frac{e^{-(a+j\omega)t}}{-(a+j\omega)} \right|_0^{\infty} = \frac{1}{a+j\omega}.$$

$$|X(\omega)| = \frac{1}{\sqrt{a^2 + \omega^2}} \quad \angle X(\omega) = -\tan^{-1}\left(\frac{\omega}{a}\right).$$

~~Amplitude Spec.~~

Magnitude.

phase spec.

$$5. x(t) = \cancel{e^{-t+a t}} e^{-a|t|} \quad a > 0 \quad (\text{real } \times \text{even})$$

$$X(\omega) = \int_{-\infty}^0 e^{j\omega t} e^{at} dt + \int_0^{\infty} e^{-at - j\omega t} dt.$$

$$= \left. \frac{e^{(a+j\omega)t}}{a+j\omega} \right|_{-\infty}^0 + \left. \frac{e^{-(a+j\omega)t}}{-(a+j\omega)} \right|_0^{\infty}$$

$$= \frac{1}{a-j\omega} + \frac{1}{a+j\omega} = \frac{2a}{a^2 + \omega^2}.$$

Exercises

4.1.(a)  $x(t) = e^{-2(t-1)} u(t-1)$ .

$$\begin{aligned}x(\omega) &= \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt \\&= \int_{-\infty}^{\infty} e^{-2(t-1)} \cdot u(t-1) e^{-j\omega t} dt \\&= \int_1^{\infty} e^{-2(t-1)} e^{-j\omega t} dt \\&= \cancel{e^{-2}} \cdot \frac{e^{-(2+j\omega)t}}{-(2+j\omega)} \Big|_1^{\infty} \\&= \frac{e^{-(2+j\omega)\cancel{t}}}{2+j\omega} = \frac{e^{-j\omega}}{2+j\omega}\end{aligned}$$

$$|x(\omega)|_{\text{mag-spec}} = \frac{\cancel{e^{-2}}}{\sqrt{4+\omega^2}} = \frac{1}{\sqrt{4+\omega^2}}$$

$$\angle x(\omega) = \cancel{\tan^{-1}} -\omega - \tan^{-1} \frac{\omega}{2}$$

Properties  
of Fourier transform.

(1) Linearity.

Given  $x_1(t) \leftrightarrow X_1(\omega)$ .

$$x_2(t) \leftrightarrow X_2(\omega).$$

$$\text{then } a x_1(t) + b x_2(t) \leftrightarrow a X_1(\omega) + b X_2(\omega)$$

$$F\left\{ a x_1(t) + b x_2(t) \right\}.$$

$$= \int_{-\infty}^{\infty} (a x_1(t) + b x_2(t)) e^{-j\omega t} dt$$

$$= a \left[ \int x_1(t) e^{-j\omega t} dt \right] + b \left[ \int x_2(t) e^{-j\omega t} dt \right]$$

$$= a X_1(\omega) + b X_2(\omega)$$

(2) Symmetry Property.

$$X(-\omega) = X^*(\omega) . \quad \text{when } x(t) \rightarrow \text{real}$$

$$X(\omega) = \int x(t) e^{j\omega t} dt$$

$$X(-\omega) = \int x(t) e^{-j\omega t} dt$$

$$= \left[ \int x(t) e^{-j\omega t} dt \right]^* = (X(\omega))^* = \underline{\underline{X^*(\omega)}}.$$

(2) Consequence

$$X(-\omega) = X^*(\omega).$$

$$\Rightarrow \operatorname{Re}\{X(-\omega)\} + j \operatorname{Im}\{X(-\omega)\}$$

$$= \operatorname{Re} X^*(\omega) + j \operatorname{Im} X^*(\omega).$$

$$\therefore \operatorname{Re}\{X(-\omega)\} = \operatorname{Re}\{X^*(\omega)\} = \operatorname{Re}\{X(\omega)\}.$$

$$\therefore \operatorname{Im}\{X(-\omega)\} = \operatorname{Im}\{X^*(\omega)\} = -\operatorname{Im}\{X(\omega)\}.$$

if  $x(t)$  is real & even  $X(\omega) \rightarrow$  real & even.

$$X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt \rightarrow \text{even fn.}$$

$$= \int_{-\infty}^{\infty} x(-t) e^{j\omega t} dt$$

$$t = -t$$

$$= \int_{-\infty}^{\infty} x(t) e^{j\omega t} dt. \Leftarrow X(-\omega)$$

Since  $X(\omega) = X(-\omega)$  but  $X(-\omega) = X^*(\omega)$

$$X(\omega) = X^*(\omega) \Rightarrow \operatorname{Im}[X(\omega)] = 0.$$

hence  $X(\omega) \rightarrow$  real & even.

More Properties  
of FT.

(i) Time Shifting.

$$x(t) \leftrightarrow X(\omega) \text{ then}$$

$$x(t-t_0) \leftrightarrow e^{-j\omega t_0} X(\omega).$$

$$F\{x(t-t_0)\}.$$

$$= \int x(t-t_0) e^{-j\omega t} dt$$

let  $t = t-t_0$ .  
limits won't change

$$= \int_{-\infty}^{\infty} x(t) e^{-j\omega(t+t_0)} dt.$$

$$= e^{-j\omega t_0} \cdot X(\omega) \rightarrow \text{phase shift in freq. domain.}$$

(ii) Prove  $\frac{dx(t)}{dt} \leftrightarrow j\omega X(\omega)$ .

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega.$$

$$\frac{dx(t)}{dt} = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) \frac{d}{dt} e^{j\omega t} d\omega$$

$$= j\omega \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega$$

$$= j\omega X(\omega)$$

FT Properties	Topic	Author	Section	Date
Center.	Signals	A.V.O.	4.8.4.	23/03/15

Fourier Series Coefficients. as samples  
of Fourier Transform. of one period.

Given

$\leftarrow \quad x(t) = \begin{cases} \tilde{x}(t), & -\frac{T_0}{2} < t < \frac{T_0}{2} \\ 0 & -\frac{T_0}{2} < t, \quad t > \frac{T_0}{2}. \end{cases}$

In one period

Fourier Coeff.

$$\therefore a_k = \frac{1}{T_0} \int_{-\frac{T_0}{2}}^{\frac{T_0}{2}} \tilde{x}(t) e^{-j k \omega_0 t} dt.$$

$$= \frac{1}{T_0} \int_{-\frac{T_0}{2}}^{\frac{T_0}{2}} x(t) \cdot \tilde{e}^{-j k \omega_0 t} dt. \quad \text{--- } \textcircled{O}$$

but.  $x(\omega) = \int x(t) e^{j \omega t} dt$ . For given  $x(t)$

$$= \int_{-\frac{T_0}{2}}^{\frac{T_0}{2}} x(t) e^{j \omega t} dt.$$

From  $\textcircled{O}$

$$\therefore a_k = \frac{1}{T_0} X(k \omega_0). \quad \text{--- } \textcircled{D}$$

Also note that  $x(t)$  can be defined as.

$\textcircled{D}$  is true for.

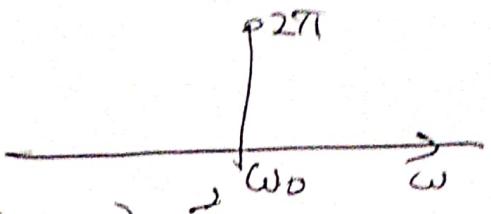
$$x(t) = \begin{cases} \tilde{x}(t) & s \leq t \leq s + T_0 \\ 0 & t < s, \quad t > s + T_0 \end{cases}$$

Fourier Transform  
for periodic signals.

For a single  
scaled sinusoid.

at  $\omega_0$

$$X(\omega) = 2\pi \delta(\omega - \omega_0)$$



①

$$\Rightarrow x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega$$

$$= \frac{1}{2\pi} \int 2\pi \delta(\omega - \omega_0) e^{j\omega t} d\omega$$

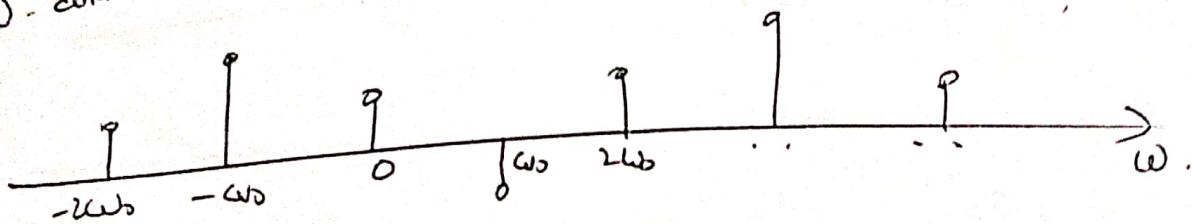
$$\therefore x(t) = e^{j\omega_0 t} \xrightarrow{\text{Fourier series from F.T.}}$$

For a  
train of  
weighted  
impulses  
in  $\omega$ -domain

$$X(\omega) = \sum_{k=-\infty}^{\infty} 2\pi a_k \delta(\omega - k\omega_0).$$

②

$X(\omega)$



then

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega$$

$$= \frac{1}{2\pi} \int_{-\infty}^{\infty} \sum_{k=-\infty}^{\infty} 2\pi a_k \delta(\omega - k\omega_0) e^{j\omega t} d\omega.$$

$$\therefore x(t) = \sum_{k=-\infty}^{\infty} a_k \cdot \int f(\omega - k\omega_0) e^{j\omega t} d\omega$$

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t}$$

↳ synthesis eq.  
for Fourier Series

Thus from ②  $X(\omega) = \sum 2\pi a_k \delta(\omega - k\omega_0)$

✓.  $\xrightarrow{\text{FT of}}$

Train of weighted impulses

$\downarrow$   
 $k^{\text{th}} \text{ area} \rightarrow 2\pi \cdot a_k$ .

a periodic  
signal.

$$\left( T_0 = \frac{2\pi}{\omega_0} \right)$$