

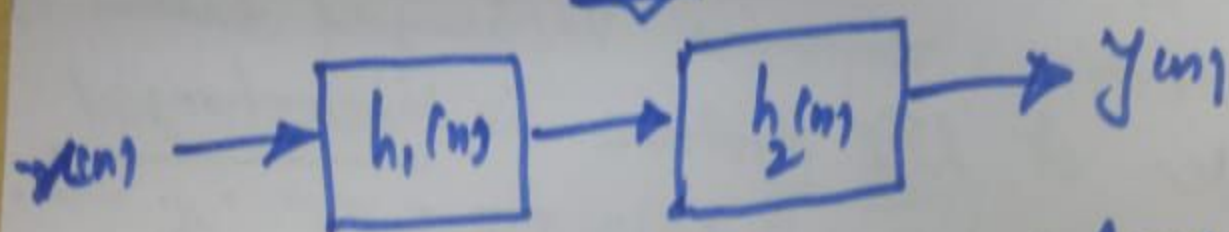
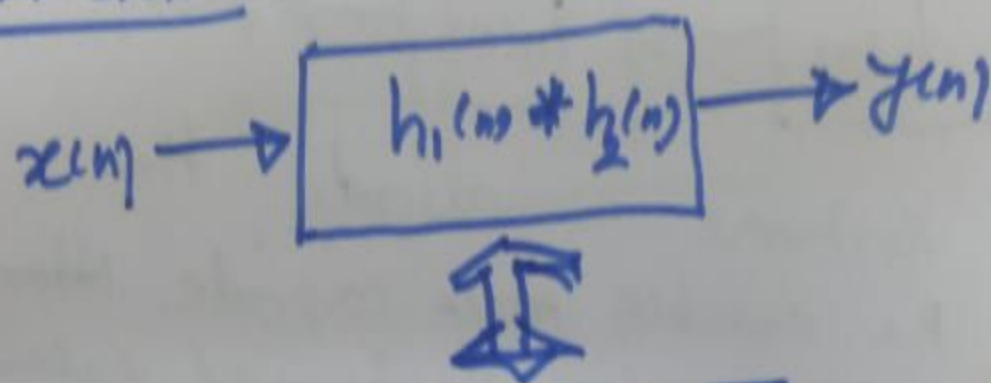
③ Associative Property:

Lecture 18

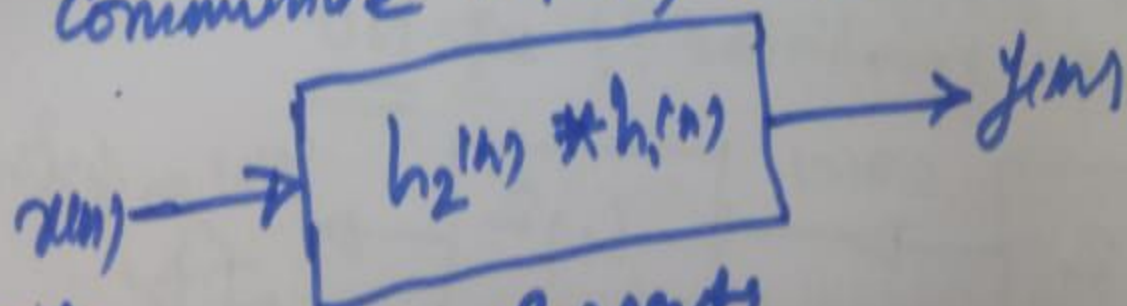
$$x(m) * \underbrace{h_1(m)} * \underbrace{h_2(m)} = [x(m) * h_1(m)] * h_2(m) \\ = x(m) * \underbrace{[h_1(m) * h_2(m)]}$$

Proof \Rightarrow Homework \checkmark

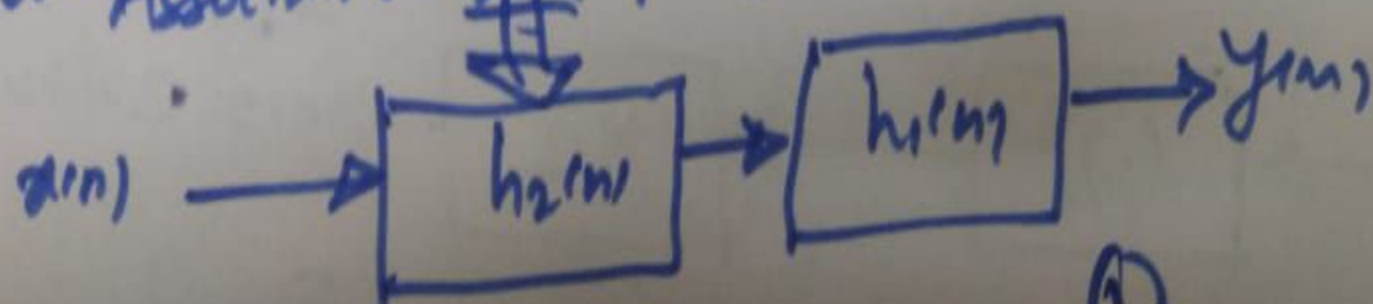
Inference:



Commutative Property of LTZ system.

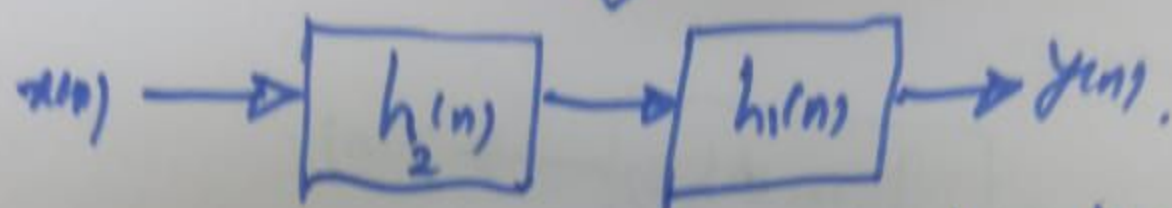
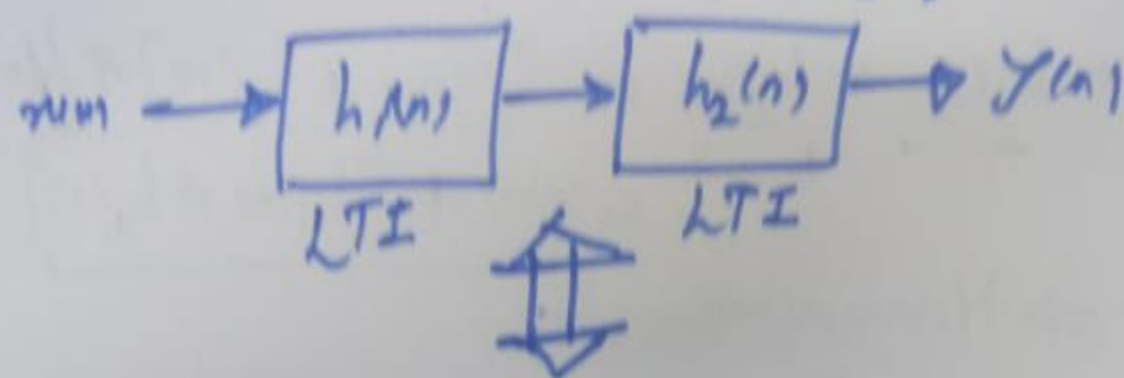


Associative Property



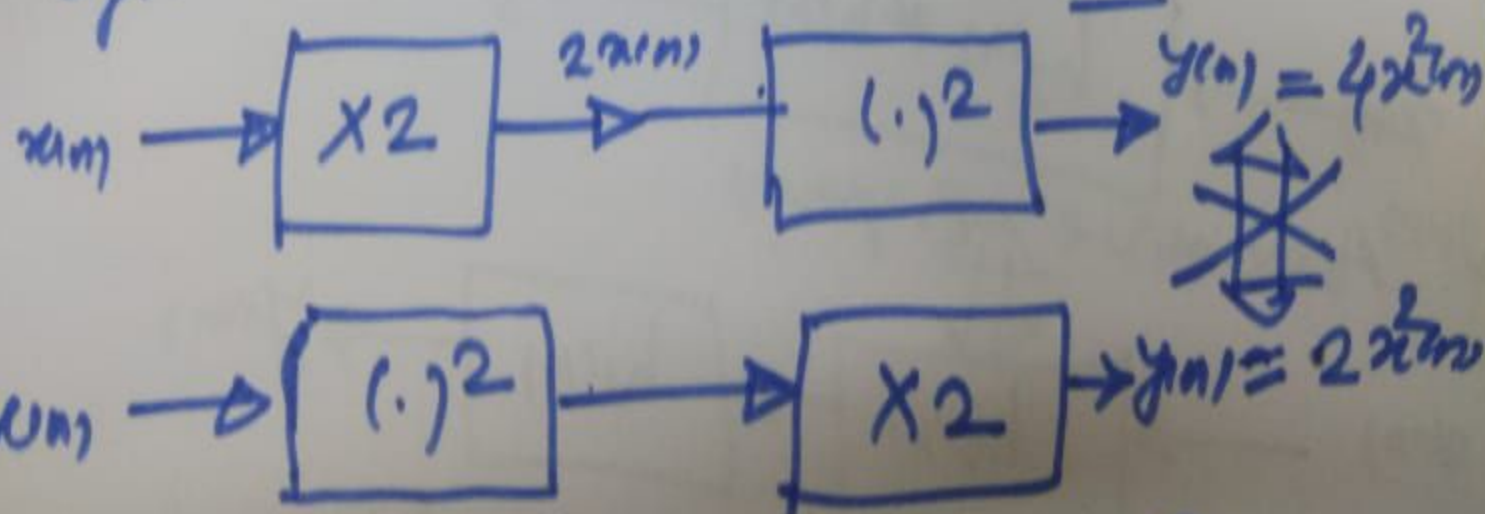
①

Using equivalence $A \Leftrightarrow B \Leftrightarrow C$
 $\therefore A \Leftrightarrow C$.



For LTI systems in cascade, the output of the overall cascade interconnection remains unchanged even if the order of LTI systems is interchanged.

Q: whether above result holds true if systems are nonlinear?? \Rightarrow NO



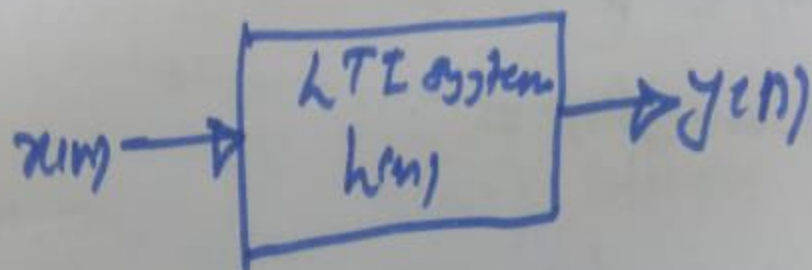
(2)

Properties of LTI systems

i) LTI system with or without memory

Problem: Find the condition on impulse response such that i) LTI system is ^{without} memory
ii) LTI system with memory.

Solⁿ →



$$y(n) = x(n) * h(n)$$

Using commutative property,

$$y(n) = h(n) * x(n)$$

$$y(n) = \sum_{k=-\infty}^{\infty} h(k) \cdot x(n-k)$$

memoryless $y(n) = h(0) \cdot x(n)$

$$y(n) = \dots + h(-1) x(n+1) + h(0) x(n) + h(1) x(n-1) + \dots$$

Input-output relationship for an LTI system
Since the output, $y(n)$, depends on the ~~past~~ past and future values of input,

\therefore the given LTZ system is with memory
 $h(n) \neq 0$ for $n \neq 0$

For memoryless LTZ system, we must choose,

$$\cancel{h(n-1)} = \cancel{h(n-2)} = \dots = 0$$

$$\cancel{y(n)} = \cancel{y(n)} = \dots = 0$$

$$h(n-1) = h(n-2) = \dots = 0$$

$$h(1) = h(2) = \dots = 0$$

$$\therefore h(n) = \begin{cases} K, & n=0 \\ 0 & n \neq 0 \end{cases}$$

$$h(n) = K \cdot \delta(n)$$

condition on
 impulse for
 system
 memoryless

$$y(n) = x(n) * h(n) = x(n) * K \cdot \delta(n), \text{ LTZ}$$

$$= K [x(n) * \delta(n)]$$

$$y(n) = K x(n)$$

(4)

① Find condition on $h[n]$ for
 Soln: Causal LTI system.

Method I
 For causal LTI system, we must have

$$h[-1] = h[-2] = \dots = 0$$

$$\Rightarrow \boxed{h[n] = 0 \text{ for } n < 0} \checkmark$$

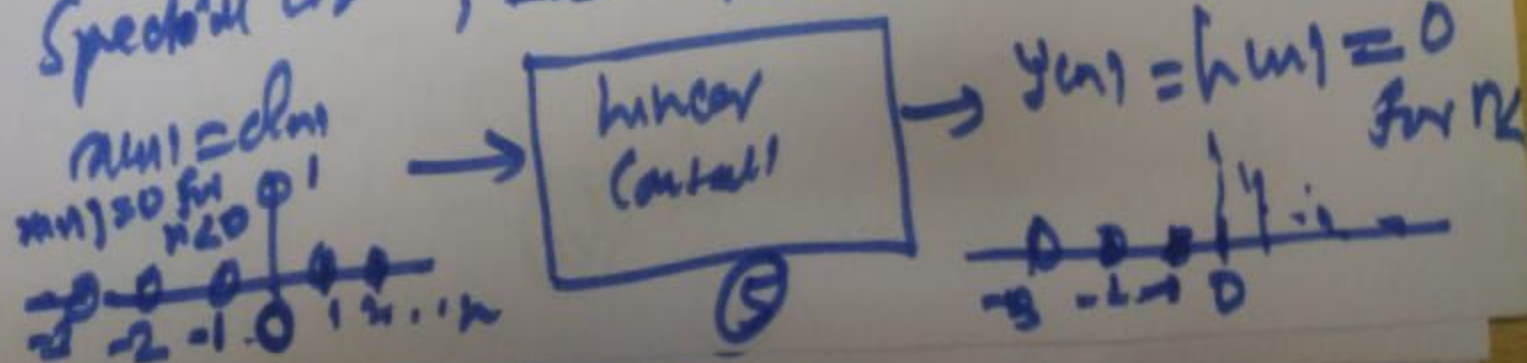
$$y[n] = \sum_{k=-\infty}^{+\infty} h[k] \cdot x[n-k]$$

$$y[n] = \sum_{k=0}^{+\infty} h[k] \cdot x[n-k] \quad \begin{matrix} \text{O/p for} \\ \text{causal} \\ \text{LTI system} \end{matrix}$$

Soln: Method II: Using Proposition 1.3

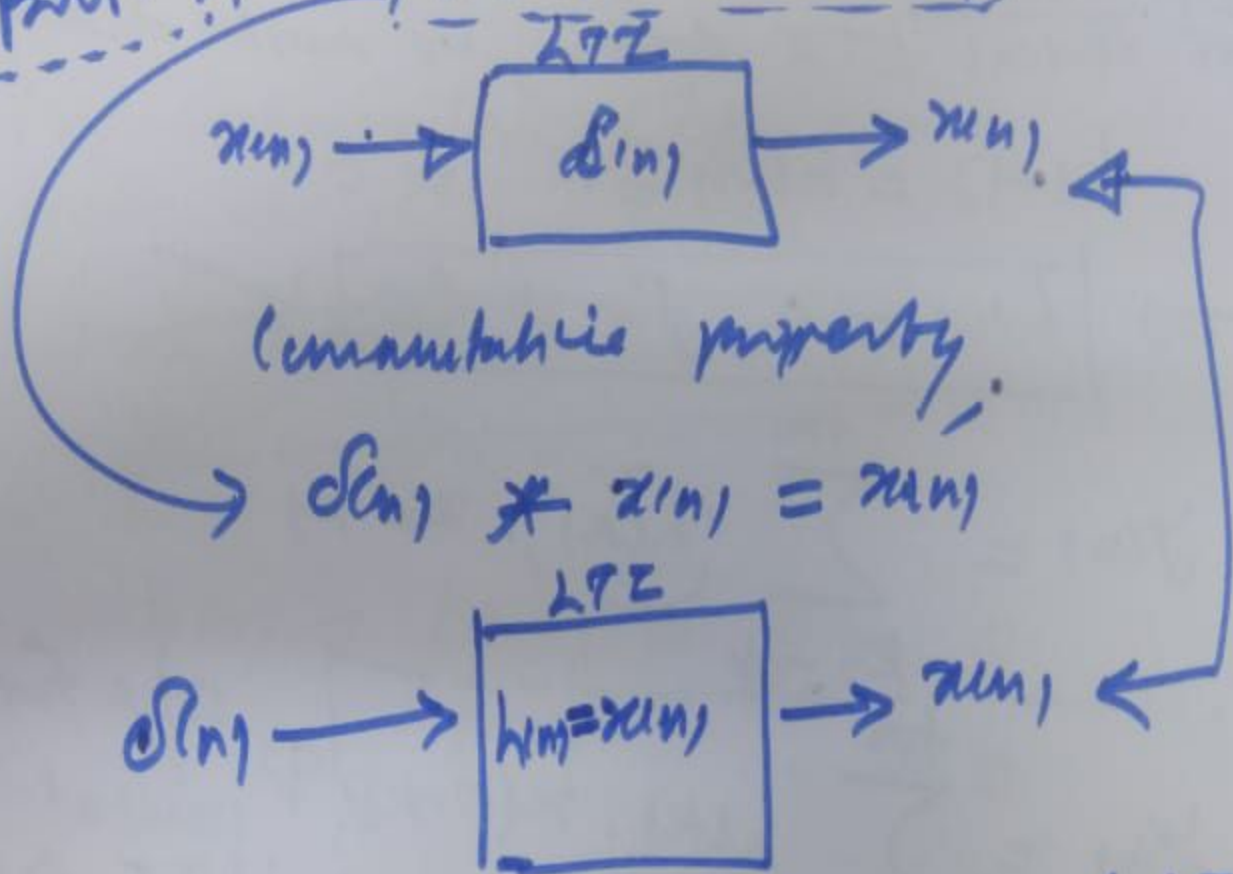
Proposition 1.3 For linear and causal system
 if $x[n] = 0$ for $n \leq n_0$ then $y[n] = 0$ for $n \leq n_0$

Special case, $x[n] = \delta[n] \Rightarrow y[n] = h[n]$



Problem For causal signal, $x[n] = 0$ for $n < 0$

Proof \rightarrow $\boxed{x[n] * \delta[n] = x[n]}$



Using the result that for causal LTI system, $h[n] = 0$ for $n < 0$
 $\Rightarrow x[n] = 0$ for $n < 0$

Problem 6 Find condition in $h(n)$,
for stable LTI system.

Solⁿ: \rightarrow We have $y(n) = \sum_{k=-\infty}^{+\infty} h(k) \cdot x(n-k)$ — ①

For BIBO stability criterion let us
assume that the input $x(n]$ is bounded
i.e. $|x(n)| \leq B_2 < +\infty, \forall n$

$$\therefore |y(n)| = \left| \sum_{k=-\infty}^{+\infty} h(k) \cdot x(n-k) \right|$$

$$|y(n)| = \quad ? ?$$

$$\boxed{\sum_{k=-\infty}^{+\infty} |h(k)| < +\infty}$$

For stable LTI system the impulse
response of LTI system is absolutely
summable. ②

Problem For LTI system,

$$h(n) = a^n \cdot u(n) \quad \checkmark$$

Case I) $|a| < 1 \rightarrow$ ^{stable} LTI, Case II $|a| > 1$

$$\sum_{n=-\infty}^{+\infty} |h(n)| < +\infty$$

Unstable
LTI
system.

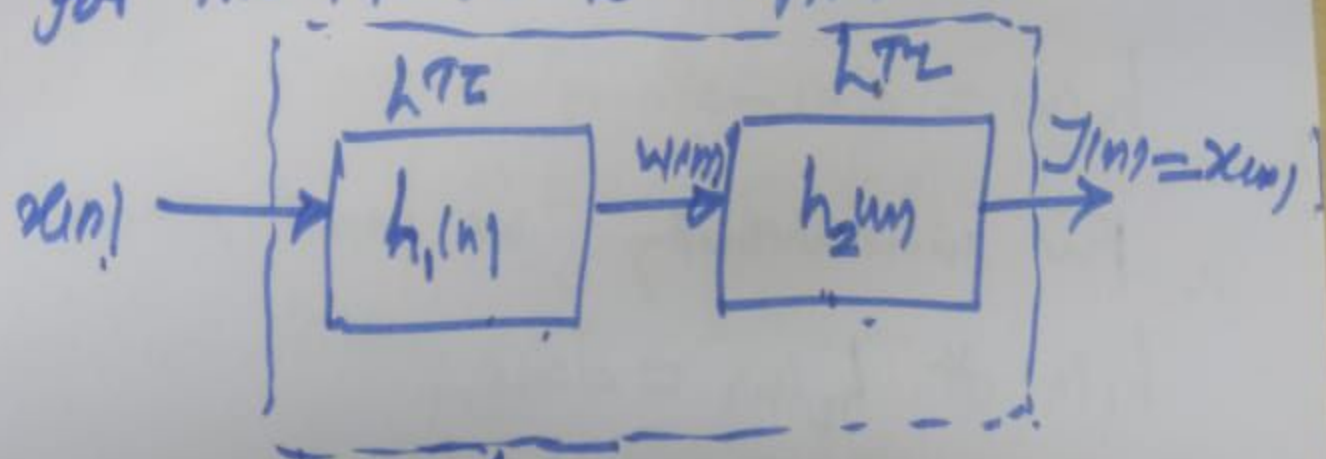
~~h(n)~~ $h(n) = \begin{cases} \frac{1}{3} & -1 \leq n \leq +1 \\ 0 & \text{elsewhere} \end{cases}$

stable LTI system

FIR systems are always stable LTI systems.

Important property.

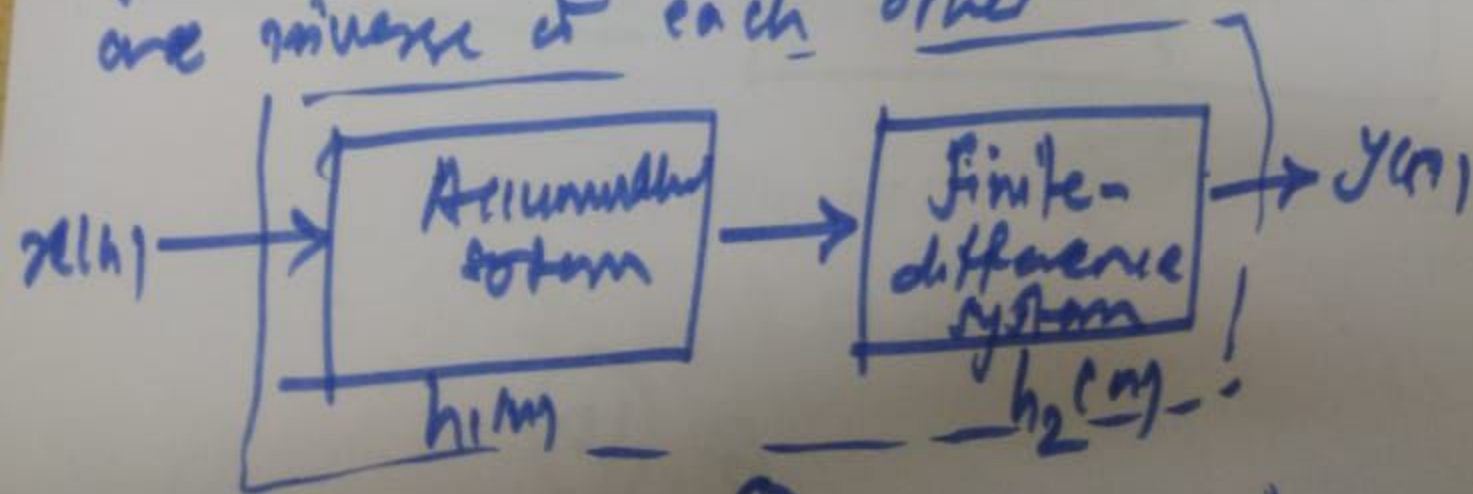
Problem (4) Find the condition on h_1 for invertible LTI system.



$$h(n) = \delta(n) = h_1(n) * h_2(n)$$

$$h_1(n) * h_2(n) = \delta(n)$$

Problem: Prove that an accumulator system and finite-difference system are inverse of each other



(9)

$$h_1(n) = u(n)$$

$$h_2(n) = \delta(n) - \delta(n-1)$$

\therefore For invertibility, we must prove

$$h_1(n) * h_2(n) = \delta(n)$$

$$\therefore \text{LHS} = h_1(n) * h_2(n)$$

$$= u(n) * [\delta(n) - \delta(n-1)]$$

Distributive property

$$= \underbrace{u(n) * \delta(n)} - u(n) * \delta(n-1)$$

$$= u(n) - u(n-1)$$

$$\boxed{h_1(n) * h_2(n) = \delta(n)}$$

Hence proved.

For continuous-time LTI systems

1) For causal LTI system,

$$h(t) = 0 \text{ for } t < 0. \text{ [Homework]}$$

2) For stable LTI system

$$\int_{-\infty}^{+\infty} |h(t)| dt < +\infty$$

\Rightarrow Impulse response is absolutely integrable.

3) For memoryless LTI system

$$h(t) = k \cdot \delta(t)$$

$$y(t) = k \cdot x(t).$$

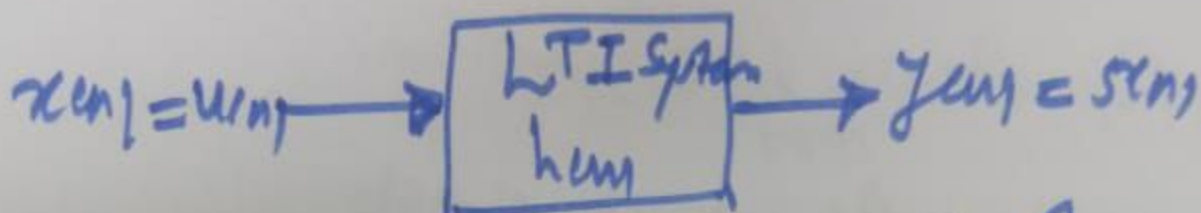
4) For invertible LTI system

$$h_1(t) * h_2(t) = \delta(t)$$

(ii)

Unit Step Response of an LTI system

Application: Used in electronic circuit theory, control systems, power systems.



Output of a system for unit step excitation source is called unit step response, $s(n)$.

$$y(n) = x(n) * h(n)$$
$$\downarrow \quad \quad \downarrow$$
$$s(n) = u(n) * h(n)$$

Commutative property.

$$s(n) = h(n) * u(n)$$

$$s(n) = \sum_{k=-\infty}^{+\infty} h(k) (u(n-k))$$

$$s(n) = \sum_{k=-\infty}^{+\infty} h(k) \rightarrow \text{Step response of an LTI system}$$

Step response of an LTI system is dependent upon the response via accumulation operation.

$$\Rightarrow h(n) = s(n) - s(n-1)$$

For continuous-time system,

$$s(t) = \int_{-\infty}^t h(\tau) d\tau$$

$$\therefore h(t) = \frac{ds(t)}{dt}$$

Since step response of an LTI system depends upon system's impulse response, $h(t)$. Thus, $h(t)$ is the key characterization for an LTI system.