

Understanding (Reading) of Fourier Transform: →

1) Graphical Interpretation: →

$$F\{f(t)\} = F(\omega) = \int_{-\infty}^{+\infty} f(t) \cdot e^{-j\omega t} dt$$

$$F(\omega) = \int_{-\infty}^{+\infty} f(t) [\cos(\omega t) - j \sin(\omega t)] dt$$

$$F(\omega) = \underbrace{\int_{-\infty}^{+\infty} f(t) \cdot \cos(\omega t) dt}_{F_R(\omega)} - j \underbrace{\int_{-\infty}^{+\infty} f(t) \sin(\omega t) dt}_{F_I(\omega)}$$

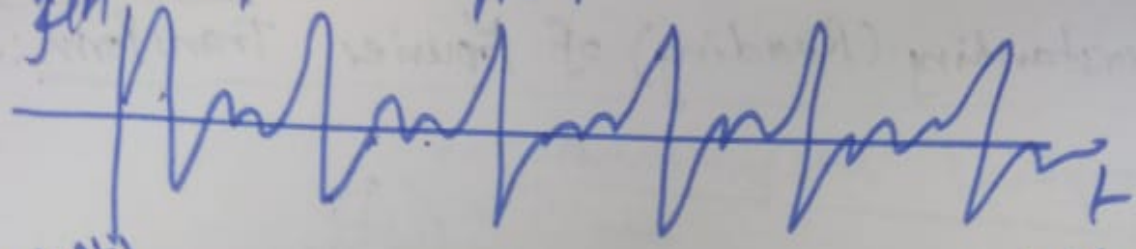
$$F(\omega) = F_R(\omega) + j F_I(\omega)$$

①. Fourier transform of a signal $f(t)$, is in general a complex number, i.e., it has a real part (i.e., $F_R(\omega)$) and an imaginary part (i.e., $F_I(\omega)$).

② $F(\omega)$ has two attributes, namely

- Magnitude spectrum $|F(\omega)| = \sqrt{F_R^2(\omega) + F_I^2(\omega)}$
- Phase spectrum, $\angle F(\omega) = \tan^{-1} \left(\frac{F_I(\omega)}{F_R(\omega)} \right)$

$f(t)$ → speech signal

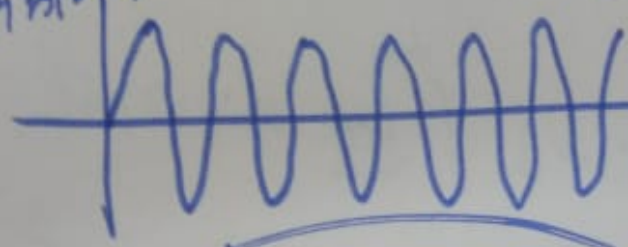


$\cos(\omega t)$



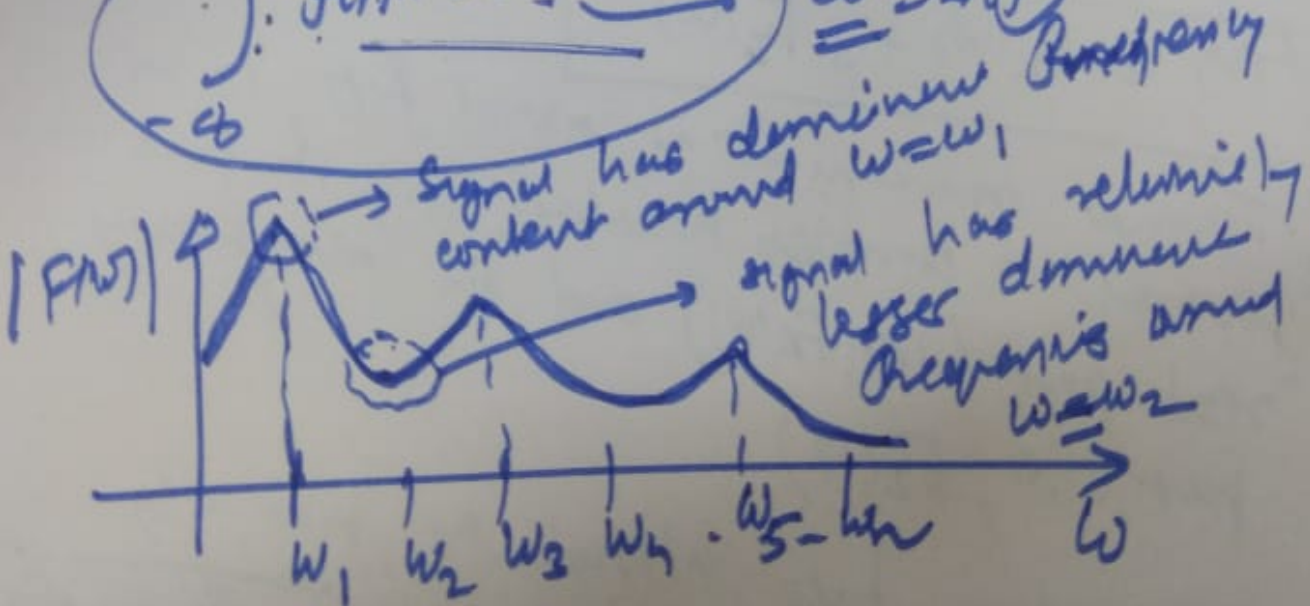
phase difference $\frac{\pi}{2}$

$\sin(\omega t)$



$$\int_{-cb}^{+cb} f(t) \cdot \cos(\omega t) dt$$

$\omega = 2\pi F$



② Inner Product Interpretation: \rightarrow
 Let $f(t)$, $g(t)$ are two signals, then

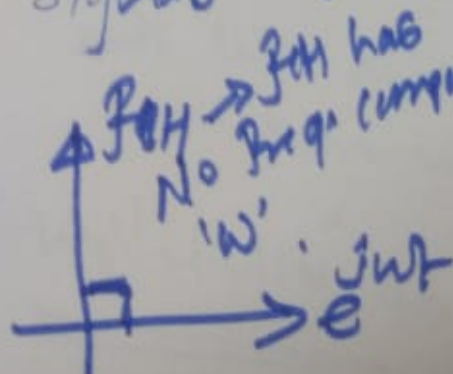
$$\langle f(t), g(t) \rangle = \int_{-\infty}^{\infty} f(t) \cdot g^*(t) dt \quad \text{--- ①}$$

$$\mathcal{F}\{f(t)\} = \int_{-\infty}^{\infty} f(t) \cdot e^{-j\omega t} dt \quad \text{--- ②}$$

$$\boxed{F(\omega) = \langle f(t), e^{j\omega t} \rangle}$$

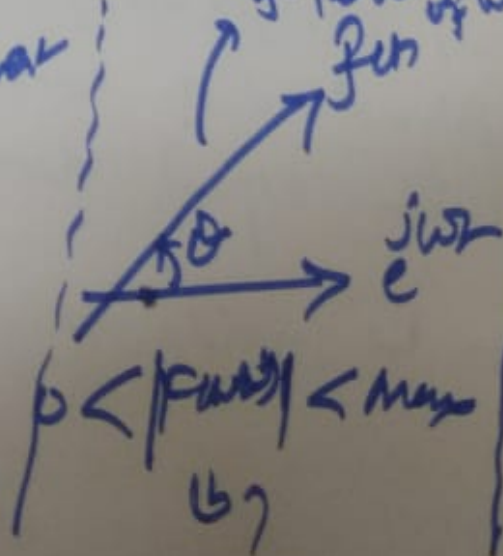
geometric structure.

For signals and systems, we model signals as vectors



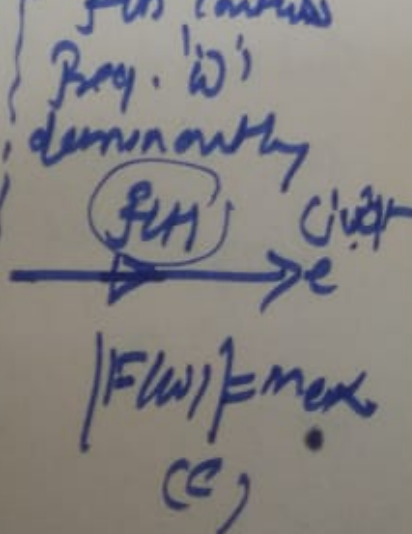
$$|F(\omega)| = 0$$

(a)



$$0 < |F(\omega)| < \max$$

(b)



$$|F(\omega)| = \max$$

(c)

③

Problems on Fourier Transform

① Find Fourier transform of $f(t) = \delta(t)$.

Soln. $f(t) = \delta(t)$

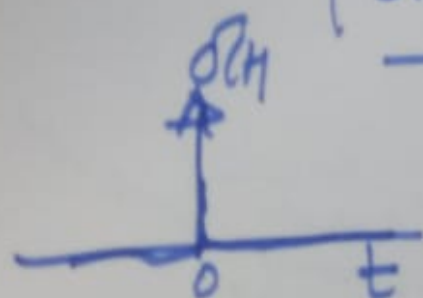
$$\therefore F(\omega) = F\{f(t)\} = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$$

$$\therefore F\{\delta(t)\} = \delta(\omega) = \int_{-\infty}^{\infty} \delta(t) \cdot e^{-j\omega t} dt$$

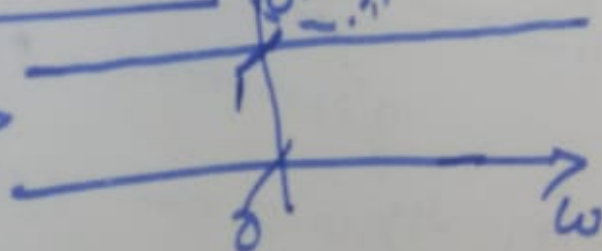
Impulse Function

$$\boxed{\delta(\omega) = 1 \quad \forall \omega}$$

$$\delta(\omega) = F\{\delta(t)\}$$



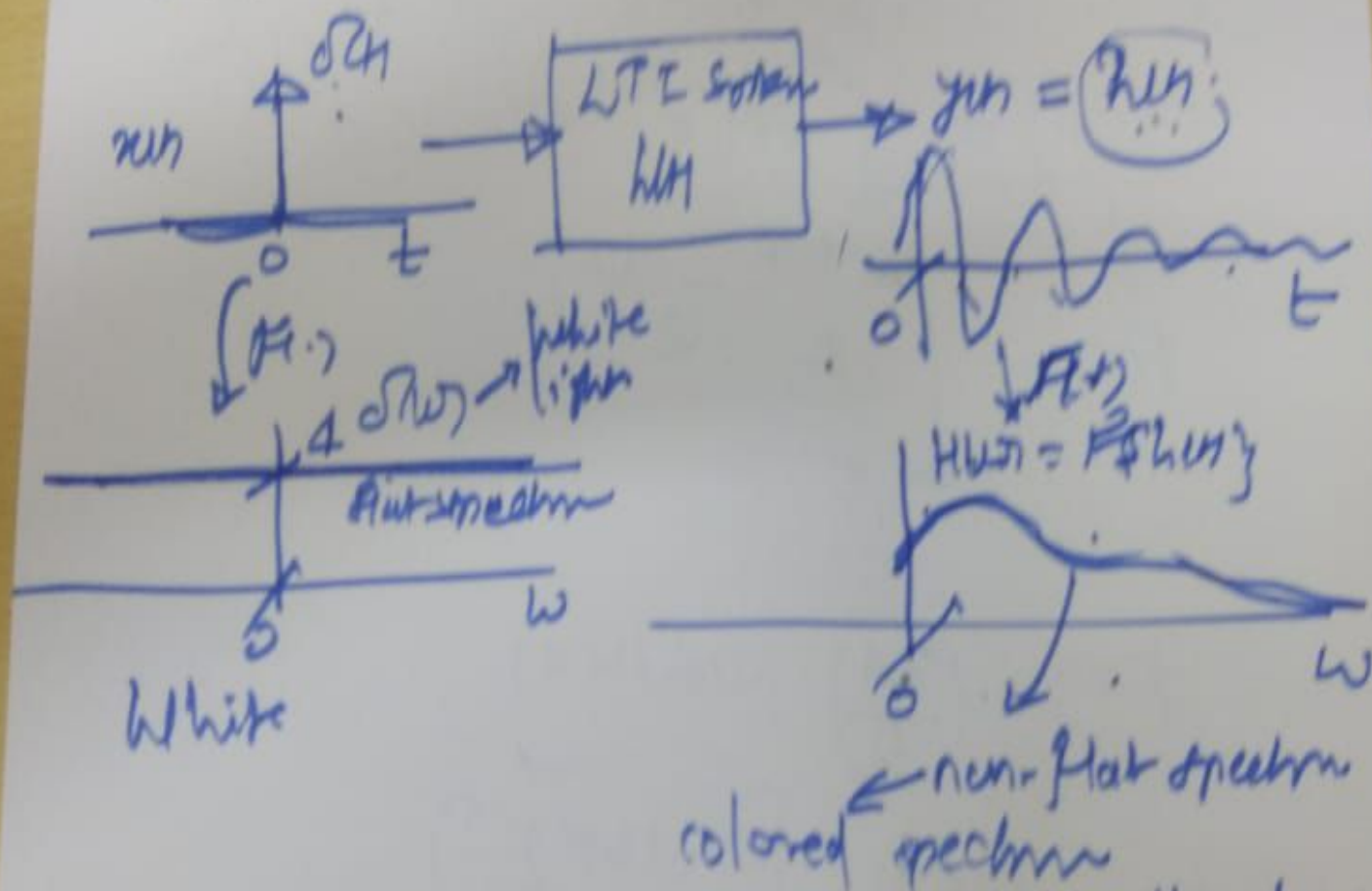
$F(\cdot)$



Interpretation: $\delta(\omega)$ i.e. $F\{\delta(t)\}$ is $1 \quad \forall \omega$ means the impulse signal $\delta(t)$ contains all the frequencies with 'equal dominance'.

\Rightarrow ' $\delta(t)$ ' can be considered as 'whitelight'

② Vonis Farines transfer of δh , practical impulse response of an LTI system characterizes an LTI system completely.

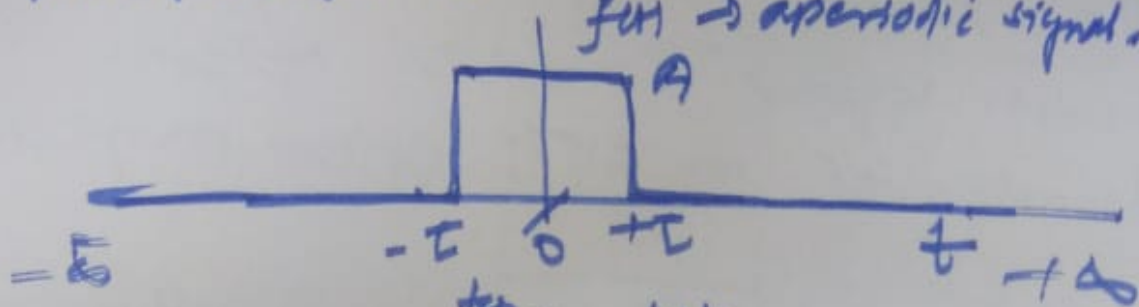


⇒ The LTI system under consideration is responsible to introduce spectral colours in its impulse response, $h(t)$.

⇒ Hence, $h(t)$ completely characterise LTI system.

⑤

③ Find F.W for pulse signal
 for \rightarrow aperiodic signal.



$$F\{f(t)\} = \int_{-\infty}^{+\infty} f(t) e^{-j\omega t} dt$$

$$\downarrow$$

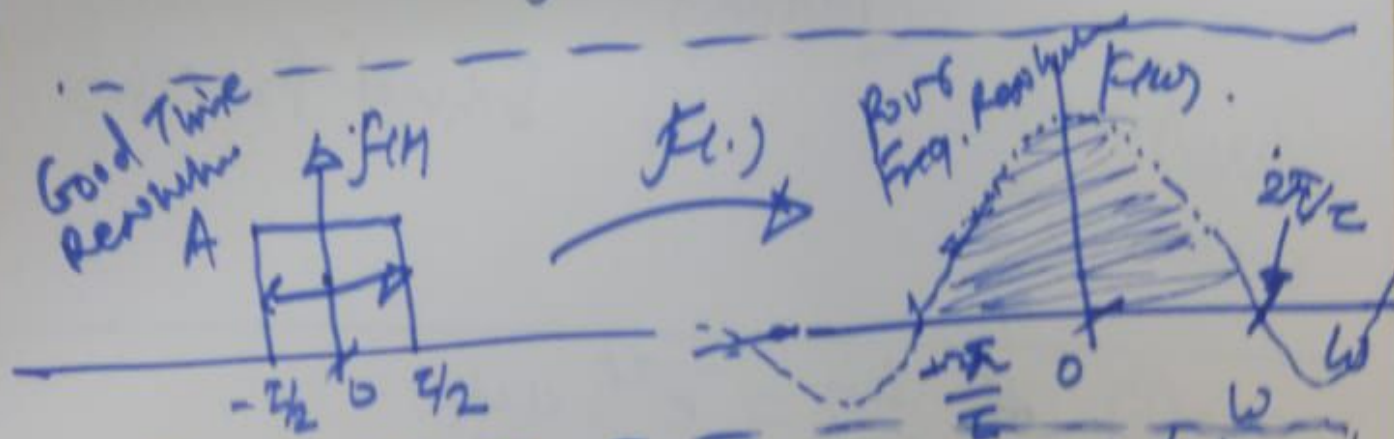
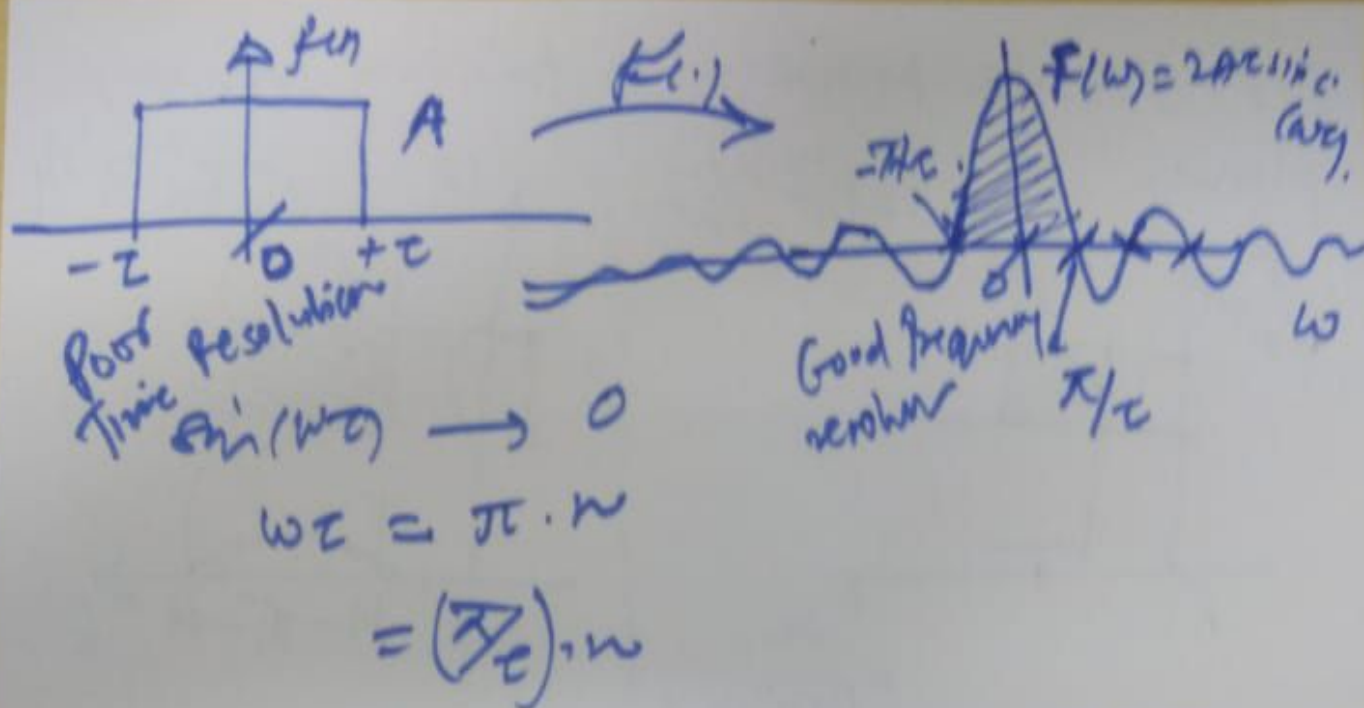
$$F(\omega) = \int_{-\tau}^{+\tau} A \cdot e^{-j\omega t} dt$$

$$F(\omega) = \left(\frac{2A}{\omega} \right) \sin(\omega \tau)$$

$$= (2A\tau) \cdot \left[\frac{\sin(\omega \tau)}{(\omega \tau)} \right]$$

$$\boxed{F(\omega) = (2A\tau) \operatorname{sinc}(\omega \tau)}$$

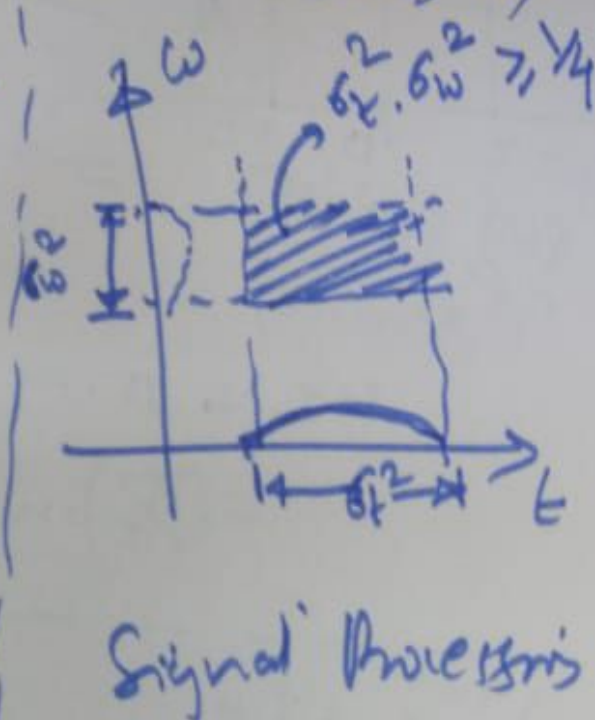
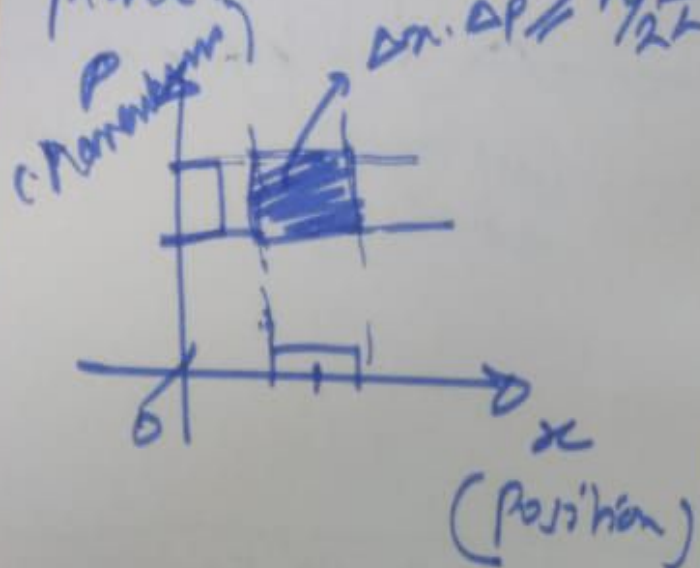
⑥



Inference: We cannot achieve good time resolution and good frequency resolution at the same time.

Reminds Heisenberg's Uncertainty Principle
 in Signal Processing

Let $f(x)$ be periodic in time and frequency plane.



Assume $f(t) = e^{-at}$, with

Soln: $\rightarrow F\{f(t)\} = F(\omega) = \int_{-\infty}^{\infty} f(t) \cdot e^{-j\omega t} dt$

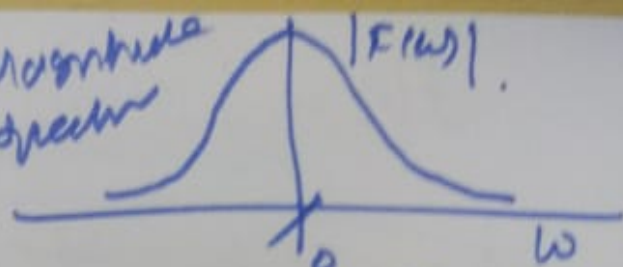
$\therefore F\{f(t)\} = F\{e^{-at} \cdot u(t)\} = \int_{-\infty}^{\infty} [e^{-at} \cdot u(t)] \cdot e^{-j\omega t} dt$

$F(\omega) = \frac{1}{a + j\omega} \rightarrow \underline{\underline{RC \text{ circuit}}}$

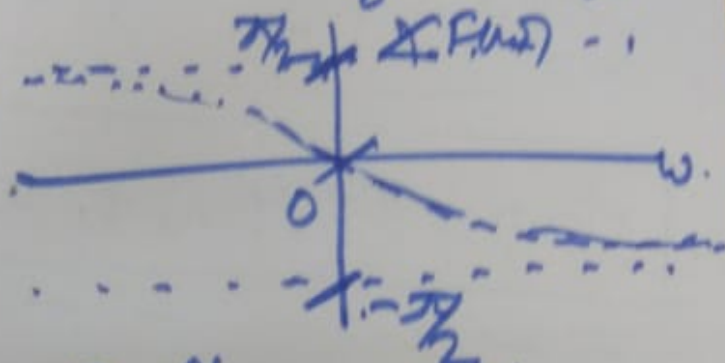
(8)

$$|F(\omega)| = \frac{1}{\sqrt{a^2 + \omega^2}}$$

① Magnitude spectrum

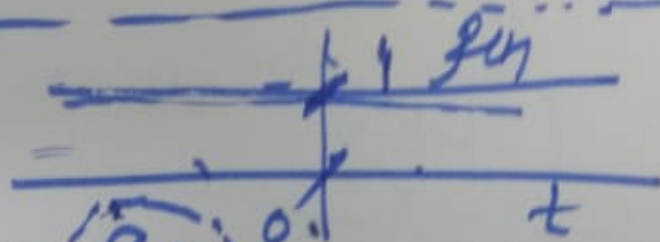


$$\angle F(\omega) = -\tan^{-1}(\frac{\omega}{a})$$



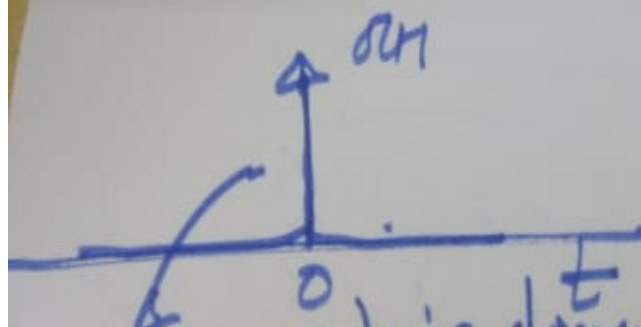
② Phase spectrum.

⑤ $f(t) = 1, \forall t$



Let $F(\omega) = \mathcal{F}\{f(t)\} = 2\pi \delta(\omega)$

Impulse in frequency domain



Impulse in time domain

We have $\mathcal{F}\{F(\omega)\} = f(t) = \left(\frac{1}{2\pi}\right) \int_{-\infty}^{\infty} F(\omega) e^{j\omega t} d\omega$

⑥ $f(t) = \left(\frac{1}{2\pi}\right) \int_{-\infty}^{\infty} (2\pi \delta(\omega)) e^{j\omega t} d\omega$

$$f(t) = \int_{-\infty}^{+\infty} (\delta(\omega)) \cdot e^{j\omega t} d\omega$$

$$F(t) = 1 \quad \forall t \rightarrow \text{impulse}$$

$$\therefore F(\omega) = \mathcal{F}\{F(t)\} = 2\pi \delta(\omega)$$

$$\mathcal{F}\{1\} = 2\pi \delta(\omega)$$

$$\int_{-\infty}^{+\infty} f(t) e^{-j\omega t} dt = 2\pi \delta(\omega)$$

$$\therefore \int_{-\infty}^{+\infty} 1 \cdot e^{-j\omega t} dt = 2\pi \delta(\omega)$$

$$\int_{-\infty}^{+\infty} e^{-j\omega t} dt = 2\pi \delta(\omega)$$

LIATE rule

Impulse in
Frequency domain

⑥

$$\omega \rightarrow \omega - \omega_0$$

$$\therefore \int_{-\infty}^{+\infty} e^{j(\omega - \omega_0)t} dt = 2\pi \delta(\omega - \omega_0)$$

$$\int_{-\infty}^{+\infty} e^{j(\omega - n\omega_0)t} dt = 2\pi \delta(\omega - n\omega_0)$$

$$\mathcal{F}\{e^{jn\omega_0 t}\} = \underline{2\pi \delta(\omega - n\omega_0)}$$

$$\mathcal{F}\{F_n e^{jn\omega_0 t}\} = 2\pi F_n \delta(\omega - n\omega_0)$$

$$n=1 \rightarrow \mathcal{F}\{F_1 e^{j1\omega_0 t}\} = 2\pi F_1 \delta(\omega - 1\omega_0)$$

$$n=2 \rightarrow \mathcal{F}\{F_2 e^{j2\omega_0 t}\} = 2\pi F_2 \delta(\omega - 2\omega_0)$$

$$\mathcal{F}\{F_1 e^{j1\omega_0 t} + F_2 e^{j2\omega_0 t}\} = 2\pi F_1 \delta(\omega - \omega_0) + 2\pi F_2 \delta(\omega - 2\omega_0)$$

$$\mathcal{F}\left\{\sum_{n=-\infty}^{+\infty} F_n e^{jn\omega_0 t}\right\} = \sum_{n=-\infty}^{+\infty} 2\pi F_n \delta(\omega - n\omega_0)$$

(14)

2. $f(t) = \sum_{n=-\infty}^{\infty} F_n e^{jn\omega_0 t}$ where F_n 's are called as Fourier series coefficients.

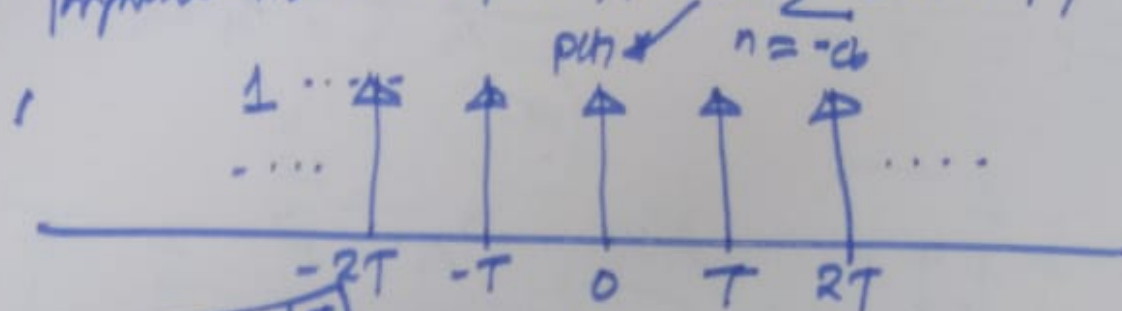
then we can even find Fourier transform of periodic signal, $f(t)$

$$F(\omega) = \sum_{n=-\infty}^{\infty} 2\pi \cdot F_n \delta(\omega - n\omega_0)$$

where, $\omega_0 = \frac{2\pi}{T}$, where $T = \text{fundamental period of periodic signal } f(t)$

with a condition that impulses are allowed to be represented in frequency domain $\{\delta(\omega - n\omega_0)\}_{n \in \mathbb{Z}}$ are permitted.

Problem 6 Find the Fourier transform of impulse-train signal, $p(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT)$



Soln: Method 2 Fig. Impulse-train signal

$$\mathcal{F}\{p(t)\} = P(\omega) = \sum_{n=-\infty}^{\infty} 2\pi P_n \cdot \delta(\omega - n\omega_0)$$

where P_n = Fourier transform coefficient of $p(t)$.

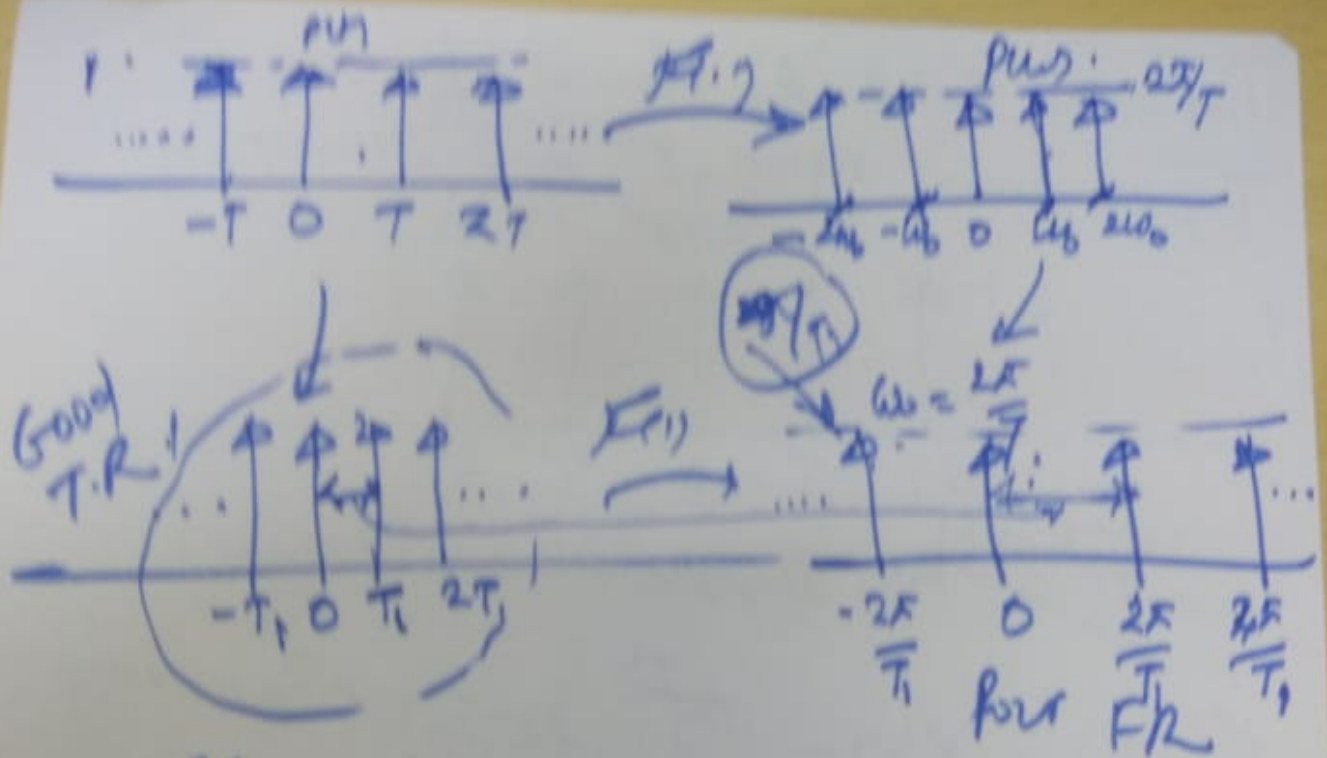
$$P_n = \frac{1}{T}, \quad \forall n$$

$$P(\omega) = \sum_{n=-\infty}^{\infty} 2\pi \left(\frac{1}{T}\right) \cdot \delta(\omega - n\omega_0)$$

$$P(\omega) = \left(\frac{2\pi}{T}\right) \sum_{n=-\infty}^{\infty} \delta(\omega - n\omega_0)$$

Impulses in freq. domain

(13)



⇒ Interms : →

Manifestation of Heisenberg's character
in signal processing domain

Method II) Via Fourier transform expression

$$P(\omega) = \mathcal{F}\{p(t)\} = \int_{-\infty}^{+\infty} p(t) e^{-j\omega t} dt$$

$$\therefore P(\omega) = \int_{-\infty}^{+\infty} \left[\sum_{n=-\infty}^{+\infty} \delta(t - nT) \right] e^{-j\omega t} dt$$

Assuming

$$\int \sum = \sum \int$$

(14)

$$P(\omega) = \sum_{n=-\infty}^{+\infty} \left\{ \int_{-\infty}^{+\infty} \delta(t-nT) \cdot e^{-j\omega t} dt \right\}$$

$$\int_{-\infty}^{+\infty} \delta(t) \cdot e^{-j\omega t} dt = F\{\delta(t)\} = 1$$

$$F\{\delta(t-nT)\} = e^{-jn\omega T}$$

$$P(\omega) = \sum_{n=-\infty}^{+\infty} e^{-jn\omega T}$$

$$\therefore P(\omega) = \frac{2\pi}{T} \sum_{n=-\infty}^{+\infty} \delta(\omega - n\omega_0)$$

$$\therefore \sum_{n=-\infty}^{+\infty} e^{-jn\omega T} = \frac{2\pi}{T} \sum_{n=-\infty}^{+\infty} \delta(\omega - n\omega_0)$$

Poisson's formula

Properties of Fourier Transform: \rightarrow
Modularity. Fourier transform pair

$$f(t) = e^{-at} \cdot u(t) \rightarrow F(\omega) = \frac{1}{a + j\omega}$$

$$f(t) \xleftrightarrow{F} F(\omega) \quad \left\{ \begin{array}{l} e^{-at} \cdot u(t) \xleftrightarrow{F} \frac{1}{a + j\omega} \end{array} \right.$$

1) Linearity. $\left[\begin{array}{l} f(t) = e^{at} u(t) + e^{bt} u(t) \\ F(\omega) = \frac{1}{a + j\omega} + \frac{1}{b + j\omega} \end{array} \right]$

Let $f(t) \xleftrightarrow{F} F(\omega)$ and $g(t) \xleftrightarrow{F} G(\omega)$

Then $z(t) = a_1 f(t) + a_2 g(t)$

$$F[z(t)] = \int_{-\infty}^{\infty} z(t) \cdot e^{-j\omega t} dt = \int_{-\infty}^{\infty} [a_1 f(t) + a_2 g(t)] \cdot e^{-j\omega t} dt$$

$$Z(\omega) = a_1 \left[\int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt \right] + a_2 \left[\int_{-\infty}^{\infty} g(t) e^{-j\omega t} dt \right]$$

$$Z(\omega) = a_1 F(\omega) + a_2 G(\omega)$$

$$\sum_{\langle k \rangle} a_k f_k(t) \xleftrightarrow{F(\omega)} \sum_{\langle k \rangle} a_k F_k(\omega)$$

(16)