

Signals and Systems (CT 203)

Tutorial Sheet-11

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1. Find the Discrete-time Fourier transform (DTFT) of

$$x(n) = (0.2)^n u(n)$$

Solution:-

$$x(n) = (0.2)^n u(n)$$

Taking DTFT, $X(e^{j\omega}) = \sum_{n=-\infty}^{+\infty} x(n) e^{-j\omega n} = \sum_{n=-\infty}^{+\infty} (0.2)^n u(n) e^{-j\omega n}$

$$\therefore X(e^{j\omega}) = \sum_{n=0}^{+\infty} (0.2)^n e^{-j\omega n} \quad \{ \because u(n) = \begin{cases} 1 & n \geq 0 \\ 0 & n < 0 \end{cases} \}$$

$$\therefore X(e^{j\omega}) = \sum_{n=0}^{+\infty} (0.2e^{-j\omega})^n = \frac{1}{1-0.2e^{-j\omega}} \text{ANS} \quad \{ \because \sum_{n=0}^{+\infty} a(r)^n = \frac{1}{1-r}, r < 1 \}$$

2. Prove that the DTFT of discrete-time signal $x(n)$, is periodic, i.e.,

$$X(e^{j(\omega+2\pi)}) = X(e^{j\omega}) \quad \text{where, } X(e^{j\omega}) = \sum_{n=-\infty}^{+\infty} x(n) e^{-j\omega n}$$

Solution:-

$$X(e^{j(\omega+2\pi)}) = \sum_{n=-\infty}^{\infty} x(n) e^{-j(\omega+2\pi)n} = \sum_{n=-\infty}^{\infty} x(n) e^{-j\omega n} e^{-j2\pi n} = \sum_{n=-\infty}^{\infty} x(n) e^{-j\omega n} = X(e^{j\omega}) .$$

$$\{ \because e^{-j2\pi n} = \cos(2\pi n) - j \sin(2\pi n) = 1 - 0 = 1, n \in \mathbb{Z} \}$$

$$\therefore X(e^{j(\omega+2\pi)}) = X(e^{j\omega}) .$$

Hence, proved. Inference: \rightarrow DTFT is periodic with period 2π in the frequency-domain.

3. Prove following *time shifting* and *frequency shifting* property of DTFT

$$x(n - n_o) \xleftrightarrow{F} e^{-j\omega n_o} X(e^{j\omega}) \quad (\text{Time shifting})$$

$$e^{j\omega_o n} x(n) \xleftrightarrow{F} X(e^{j(\omega - \omega_o)}) \quad (\text{Frequency shifting})$$

Solution:-

Time-shifting Property:-

$$F(x(n-n_0)) = \sum_{n=-\infty}^{+\infty} x(n-n_0) e^{-j\omega n}$$

$$\text{Let } m = n - n_0, \quad n = -\infty \Rightarrow m = -\infty, n = \infty \Rightarrow m = \infty$$

$$\therefore F(x(n-n_0)) = \sum_{m=-\infty}^{+\infty} x(m) e^{-j\omega(m+n_0)}$$

$$= e^{-j\omega n_0} \sum_{m=-\infty}^{+\infty} x(m) e^{-j\omega m} \quad \{ \because e^{-j\omega n_0} \text{ is not dependent on summation variable } m \}$$

$$= e^{-j\omega n_0} X(e^{j\omega}) \quad \{ \text{from DTFT definition} \}$$

$$\therefore x(n-n_0) \xleftrightarrow{F} e^{-j\omega n_0} X(e^{j\omega}) . \quad \text{Hence, proved.}$$

Frequency-shifting Property:-

$$F(e^{j\omega_0 n} x(n)) = \sum_{n=-\infty}^{+\infty} e^{j\omega_0 n} x(n) e^{-j\omega n} = \sum_{n=-\infty}^{+\infty} x(n) e^{-j(\omega-\omega_0)n}$$

$$\text{Let } \omega - \omega_0 = \theta \text{ (temporary)}$$

$$F(e^{j\omega_0 n} x(n)) = \sum_{n=-\infty}^{+\infty} x(n) e^{-j\theta n} = X(e^{j\theta}) = X(e^{j(\omega-\omega_0)}) \quad \{ \text{again substitute } \theta = \omega - \omega_0 \}$$

$$\therefore e^{j\omega_0 n} x(n) \xleftrightarrow{F} X(e^{j(\omega-\omega_0)}) . \quad \text{Hence proved.}$$

4. Prove Parseval's relation for DTFT

$$\sum_{n=-\infty}^{+\infty} |x(n)|^2 = \frac{1}{2\pi} \int_{-\pi}^{\pi} |X(e^{j\omega})|^2 d\omega$$

Explain the fundamental concept conveyed by this theorem.

Solution:-

$$\text{LHS} = \sum_{n=-\infty}^{+\infty} |x(n)|^2 = \sum_{n=-\infty}^{+\infty} x(n) x^*(n)$$

$$\begin{aligned}
&= \sum_{n=-\infty}^{+\infty} x(n) \left\{ \frac{1}{2\pi} \int_{2\pi} X(e^{j\omega}) e^{j\omega n} d\omega \right\}^* \quad \{\text{Using, IDTFT definition, i.e.,} \\
&\quad x(n) = \frac{1}{2\pi} \int_{2\pi} X(e^{j\omega}) e^{j\omega n} d\omega, \text{ in the second term}\} \\
&= \sum_{n=-\infty}^{+\infty} x(n) \left\{ \frac{1}{2\pi} \int_{2\pi} X^*(e^{j\omega}) e^{-j\omega n} d\omega \right\} \quad \{\text{Interchange } \sum \text{ and } \int \text{ operators, carefully!!}\} \\
&= \frac{1}{2\pi} \int_{2\pi} X^*(e^{j\omega}) \left\{ \sum_{n=-\infty}^{+\infty} x(n) e^{-j\omega n} \right\} d\omega \quad \{\text{Using DTFT definition for the term} \\
&\quad \text{which is inside the bracket}\} \\
&= \frac{1}{2\pi} \int_{2\pi} X^*(e^{j\omega}) X(e^{j\omega}) d\omega = \frac{1}{2\pi} \int_{2\pi} |X(e^{j\omega})|^2 d\omega = \mathbf{RHS}
\end{aligned}$$

Hence proved.

Inference -> This *Parseval's* relation says, the energy of the signal remains same in time-domain and frequency-domain, i.e., it is energy conservation principle in DTFT framework.

5. Prove the *convolution* property (theorem) for DTFT, if $y(n) = x(n) * h(n)$, then

$$Y(e^{j\omega}) = X(e^{j\omega}) H(e^{j\omega})$$

Solution

$$\begin{aligned}
y(n) &= x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k) h(n-k) \\
Y(e^{j\omega}) &= \sum_{n=-\infty}^{+\infty} y(n) e^{-j\omega n} = \sum_{n=-\infty}^{+\infty} \left(\sum_{k=-\infty}^{+\infty} x(k) h(n-k) \right) e^{-j\omega n} \\
\therefore Y(e^{j\omega}) &= \sum_{n=-\infty}^{+\infty} \left(\sum_{k=-\infty}^{+\infty} x(k) e^{-j\omega k} \right) h(n-k) e^{-j\omega n} e^{j\omega k} \quad \{\text{due to DTFT definition}\} \\
\therefore Y(e^{j\omega}) &= \sum_{n=-\infty}^{+\infty} \left(X(e^{j\omega}) \right) h(n-k) e^{-j\omega(n-k)} = X(e^{j\omega}) \sum_{n=-\infty}^{+\infty} h(n-k) e^{-j\omega(n-k)} \\
&\quad \{\because X(e^{j\omega}) \text{ is independent from the summation variable } n\}
\end{aligned}$$

Taking, $n-k = m$, $n = -\infty \Rightarrow m = -\infty, n = \infty \Rightarrow m = \infty$

$$\therefore Y(e^{j\omega}) = X(e^{j\omega}) \sum_{m=-\infty}^{+\infty} h(m)e^{-j\omega m} = X(e^{j\omega}) H(e^{j\omega})$$

$$\therefore Y(e^{j\omega}) = X(e^{j\omega}) H(e^{j\omega}). \text{ Hence proved.}$$

Inference-> Convolution in (discrete) time-domain corresponds to the multiplication in frequency-domain.

6. Prove the *multiplication* property for DTFT, if $y(n) = x(n)h(n)$, then

$$Y(e^{j\omega}) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\theta}) H(e^{j(\omega-\theta)}) d\theta = \frac{1}{2\pi} [X(e^{j\omega}) * H(e^{j\omega})]$$

Solution

$$y(n) = x(n)h(n)$$

Taking DTFT,

$$Y(e^{j\omega}) = \sum_{n=-\infty}^{+\infty} y(n)e^{-j\omega n} = \sum_{n=-\infty}^{+\infty} (x(n)h(n))e^{-j\omega n}$$

$$\text{Using, IDTFT definition for } x(n), \text{ i.e., } x(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\theta}) e^{j\theta n} d\theta$$

$$\therefore Y(e^{j\omega}) = \sum_{n=-\infty}^{+\infty} \left(\frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\theta}) e^{j\theta n} d\theta \right) h(n) e^{-j\omega n}$$

{ Interchange \sum and \int operators, carefully!! }

$$\therefore Y(e^{j\omega}) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\theta}) \left(\sum_{n=-\infty}^{+\infty} h(n) e^{-j(\omega-\theta)n} \right) d\theta$$

{ Using DTFT definition for the term which is inside the braces, it is DTFT of $h(n)$ with shifted frequency by (θ) }

$$\therefore Y(e^{j\omega}) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\theta}) H(e^{j(\omega-\theta)}) d\theta = \frac{1}{2\pi} [X(e^{j\omega}) * H(e^{j\omega})]. \text{ Hence Proved.}$$

7. A causal LTI system is characterized by the difference equation

$$y(n) - \frac{3}{4}y(n-1) + \frac{1}{8}y(n-2) = 2x(n)$$

Determine the *impulse response*, $h(n)$, of the LTI system.

Solution:-

$$y(n) - \frac{3}{4}y(n-1) + \frac{1}{8}y(n-2) = 2x(n)$$

Taking DTFT both sides and using time-shifting property...

$$\therefore Y(e^{j\omega}) - \frac{3}{4}e^{-j\omega}Y(e^{j\omega}) + \frac{1}{8}e^{-j2\omega}Y(e^{j\omega}) = 2X(e^{j\omega})$$

$$\therefore \left(1 - \frac{3}{4}e^{-j\omega} + \frac{1}{8}e^{-j2\omega}\right)Y(e^{j\omega}) = 2X(e^{j\omega})$$

Now,

$$H(e^{j\omega}) = \frac{Y(e^{j\omega})}{X(e^{j\omega})} = \frac{2}{\left(1 - \frac{3}{4}e^{-j\omega} + \frac{1}{8}e^{-j2\omega}\right)} = \frac{2}{\left(1 - \frac{1}{2}e^{-j\omega}\right)\left(1 - \frac{1}{4}e^{-j\omega}\right)} = \frac{A}{\left(1 - \frac{1}{2}e^{-j\omega}\right)} + \frac{B}{\left(1 - \frac{1}{4}e^{-j\omega}\right)}$$

By residue method for finding A and B

$$A = \frac{\frac{2}{\left(1 - \frac{1}{4}e^{-j\omega}\right)\left(1 - \frac{1}{2}e^{-j\omega}\right)}}{\left(1 - \frac{1}{2}e^{-j\omega}\right)} \bigg|_{e^{-j\omega} = 2} = \frac{2}{\left(1 - \frac{1}{4}(2)\right)} = 4$$

$$B = \frac{\frac{2}{\left(1 - \frac{1}{2}e^{-j\omega}\right)\left(1 - \frac{1}{4}e^{-j\omega}\right)}}{\left(1 - \frac{1}{4}e^{-j\omega}\right)} \bigg|_{e^{-j\omega} = 4} = \frac{2}{\left(1 - \frac{1}{2}(4)\right)} = -2$$

$$H(e^{j\omega}) = \frac{4}{\left(1 - \frac{1}{2}e^{-j\omega}\right)} + \frac{(-2)}{\left(1 - \frac{1}{4}e^{-j\omega}\right)}$$

Taking, IDTFT and using linearity property

$$h(n) = 4F^{-1} \left\{ \frac{1}{\left(1 - \frac{1}{2}e^{-j\omega}\right)} \right\} - 2F^{-1} \left\{ \frac{1}{\left(1 - \frac{1}{4}e^{-j\omega}\right)} \right\}$$

From, problem-1

$$F\left\{(0.2)^n u(n)\right\} = \frac{1}{1 - 0.2e^{-j\omega}}, \text{ in general, } F\left\{(r)^n u(n)\right\} = \frac{1}{1 - re^{-j\omega}}, \quad r < 1 \text{ for convergence}$$

$$\therefore h(n) = 4\left(\frac{1}{2}\right)^n u(n) - 2\left(\frac{1}{4}\right)^n u(n) \quad \underline{\underline{ANS}}$$