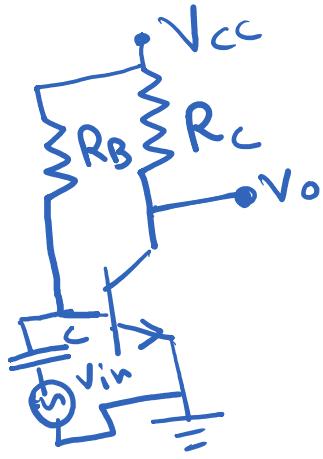
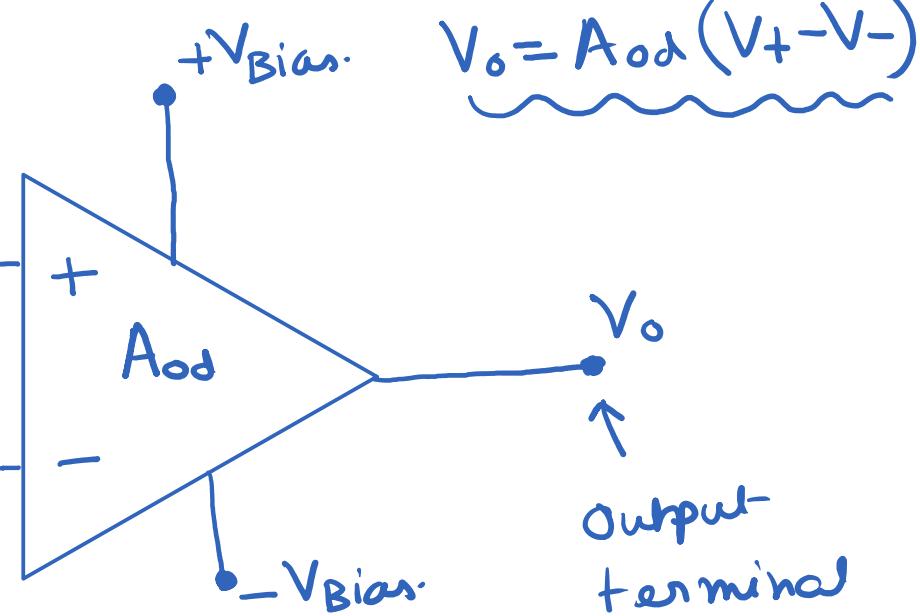


## Op-Amp (Operational Amplifier)

These can be used to carry out mathematical operations such as addition, multiplication, etc. Op-Amp is not a fundamental block or device. Rather, there are multiple transistors inside the Op-Amp that give rise to its properties.



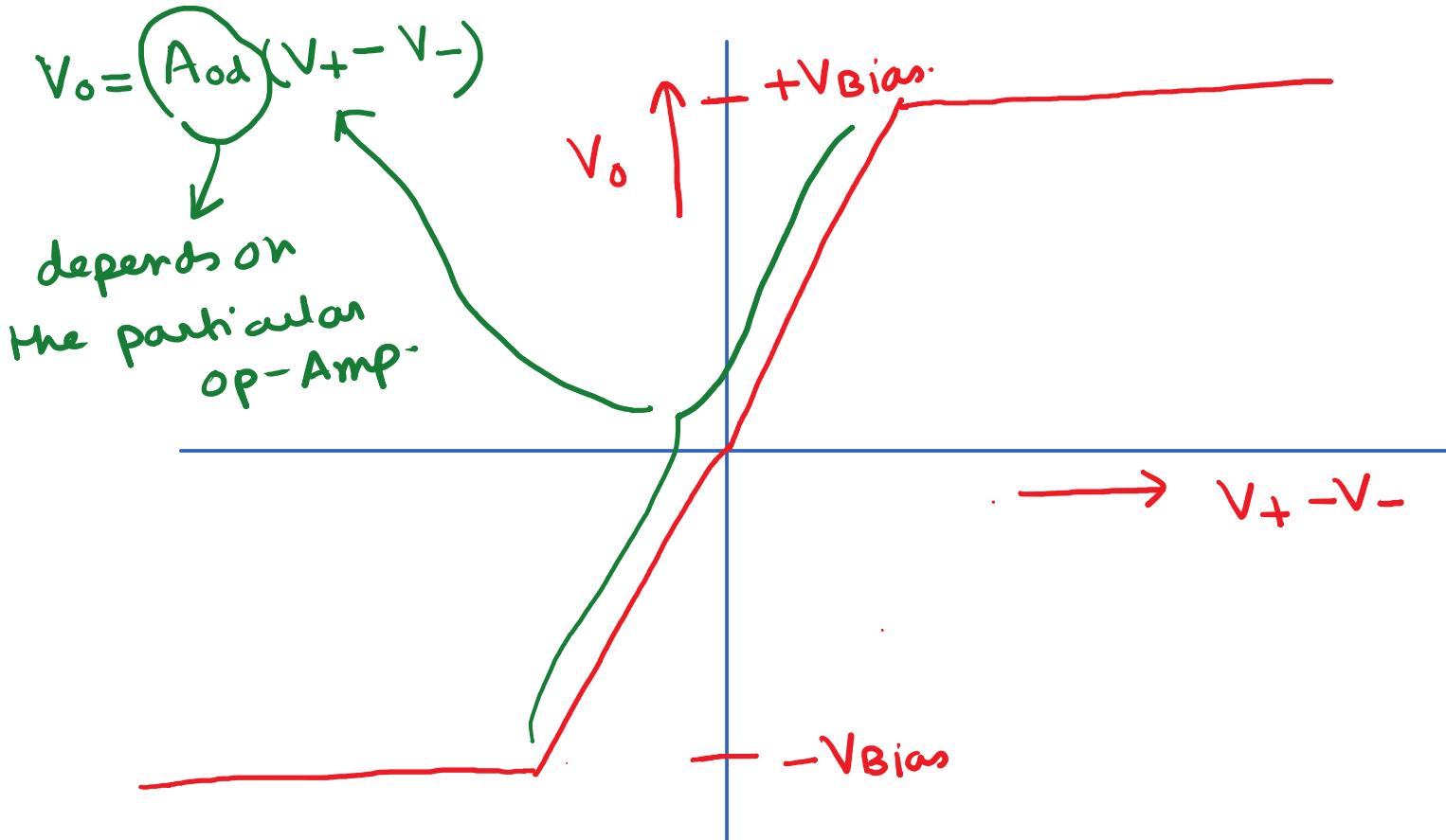
Input terminals {  $V_+$   
  $V_-$



+ → non-inverting terminal } input  
- → inverting terminal }

$A_{od}$  → Open loop gain of the op-Amp.  $A_{od}$  is very large and can be taken as  $\infty$  for mathematical derivations.

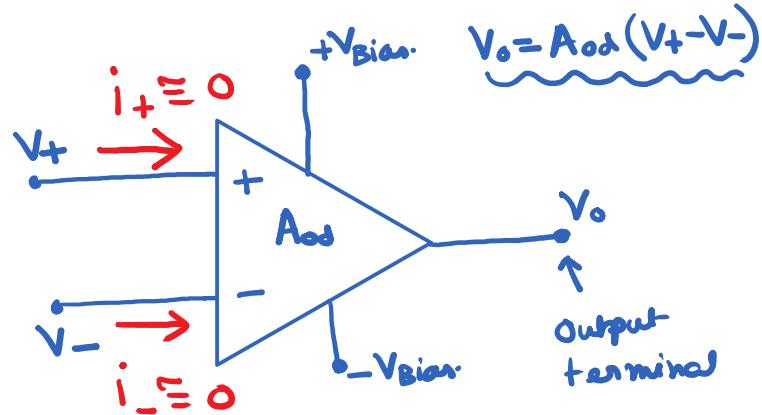
The maximum and minimum voltage across the output of an op-Amp is limited to  $+V_{Bias}$  and  $-V_{Bias}$  respectively. Generally  $+V_{Bias}$  lies in the range of  $+12V$  to  $+15V$  and  $-V_{Bias}$  lies in the range of  $-15V$  to  $-12V$ .



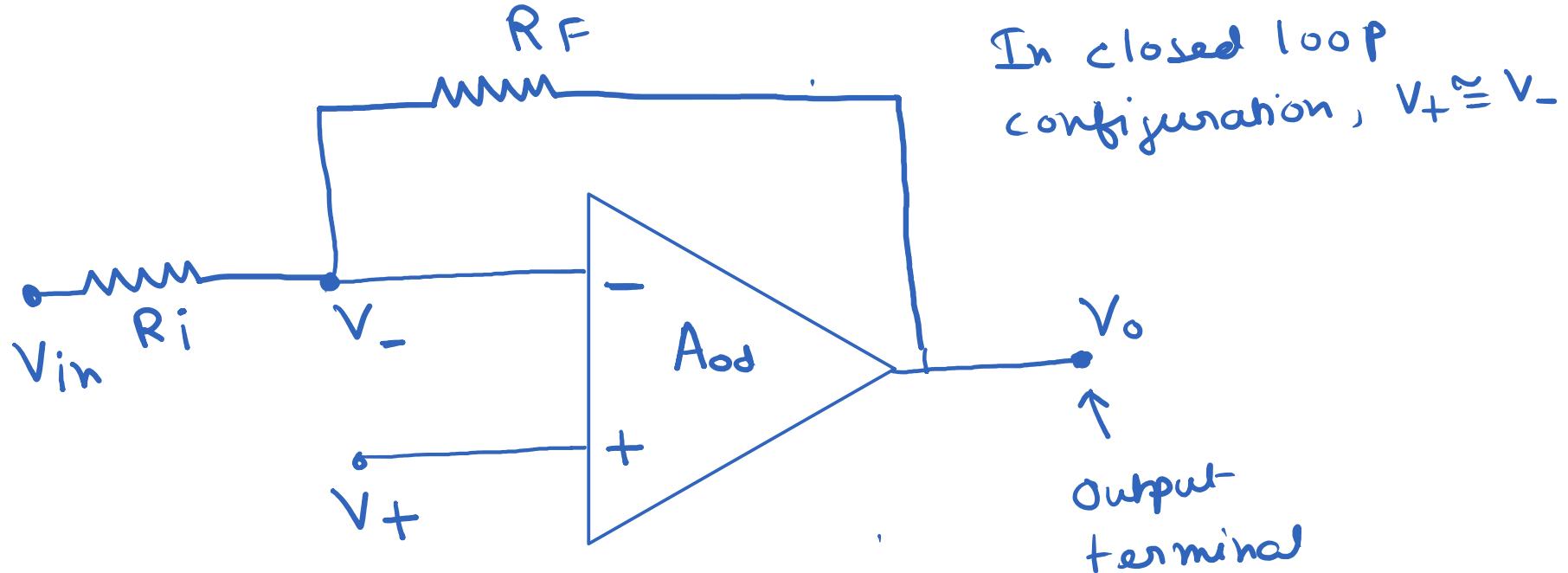
① op-Amp has very large open loop gain.  $A_{od} \rightarrow \infty$ .

② Almost no current can flow into the op-Amp. So,  $i_+ \approx 0$  and  $i_- \approx 0$ . So, an op-Amp has a very large input impedance.

③ op-Amps have very low (almost zero) output impedance. This means that the output voltage does not change with the output current or. the output current can change without a change in the output voltage.



④ In closed loop configuration,  $V_+$  and  $V_-$  should approximately stay at the same voltage, provided that  $-V_{Bias} \leq V_o \leq +V_{Bias}$ . This is known as 'virtual short'



Closed loop configuration of an op-amp is the configuration in which the output and at least one of the input terminals is connected by an external impedance.

$$V_o = A_{od} (V_+ - V_-) \quad \text{--- ①}$$

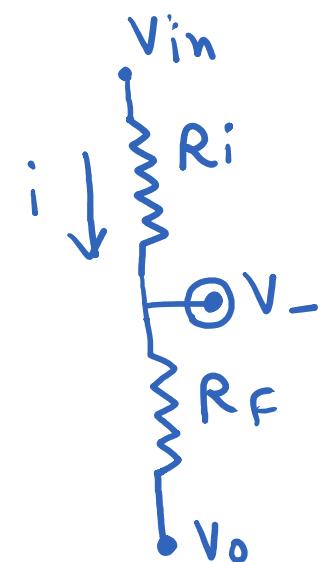
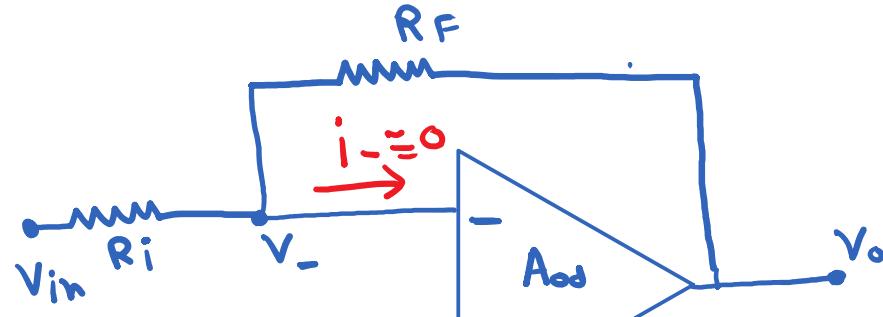
$$V_- = \frac{R_F}{R_i + R_F} V_i + \frac{R_i}{R_i + R_F} V_o \quad \text{--- ②}$$

$$V_o = A_{od} (V_+ - V_-)$$

$$= A_{od} \left( V_+ - \frac{R_F}{R_F + R_i} V_{in} - \frac{R_i}{R_F + R_i} V_o \right)$$

$$\Rightarrow V_o \left\{ 1 + \frac{R_F A_{od}}{R_F + R_i} \right\}$$

$$= A_{od} \left( V_+ - \frac{R_F}{R_F + R_i} V_{in} \right)$$



Dividing both sides by  $A_{od}$ , we get -

$$V_o \left\{ \frac{1}{A_{od}} + \frac{R_i}{R_i + R_F} \right\} = V_+ - \frac{R_F}{R_i + R_F} V_{in}.$$

Since,  $A_{od} \rightarrow \infty$ ,  $\frac{1}{A_{od}} \rightarrow 0$ .

$$\Rightarrow V_o \frac{R_i}{R_i + R_F} = V_+ - \frac{R_F}{R_i + R_F} V_{in}.$$

$$\Rightarrow V_o = \frac{R_i + R_F}{R_i} V_+ - \frac{R_F}{R_i} V_{in} \quad - \textcircled{3}$$

$$V_o = \frac{R_F + R_i}{R_i} V_+ - \frac{R_F}{R_i} V_{in} \quad - \textcircled{3}$$

From  $\textcircled{2}$

$$V_- = \frac{R_F}{R_i + R_F} V_{in} + \frac{R_i}{R_i + R_F} V_o$$

$$= \cancel{\frac{R_F}{R_i + R_F} V_{in}} + V_+ - \cancel{\frac{R_F}{R_i + R_F} V_{in}}$$

$$= V_+$$

$$V_- = V_+ \rightarrow \begin{matrix} \text{This is basically} \\ \text{for virtual} \end{matrix}$$

the condition  
short.

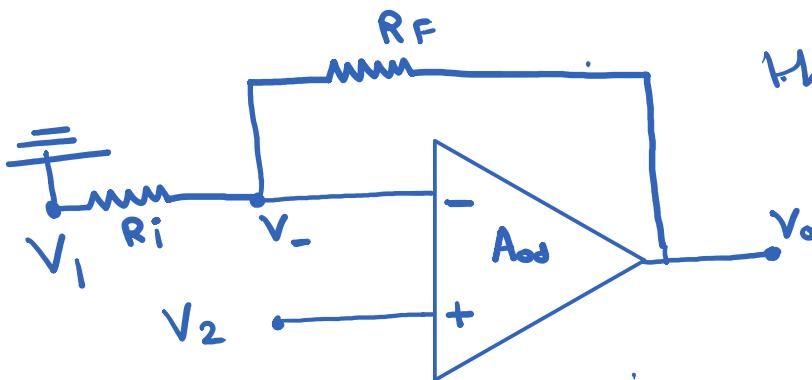
$$V_o = \frac{R_f + R_i}{R_i} V_2 - \frac{R_f}{R_i} V_1$$

Consider  $V_2 = 0$ . In this case,  $V_o = -\frac{R_f}{R_i} V_1$ .

This configuration is known as the inverting configuration of op-Amp Amplifier.

Consider  $V_1 = 0$ . In this case  $V_o = \frac{R_f + R_i}{R_i} V_2$

This configuration is known as the non-inverting configuration of op-Amp Amplifier.



## Inverting Amplifier:-

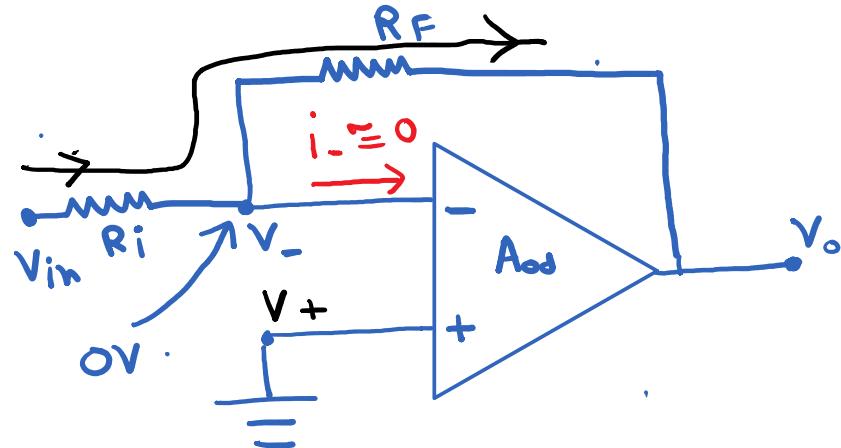
$$V_+ = 0$$

$$V_- \approx V_+ = 0$$

$$I_{R_i} = \frac{V_{in} - V_-}{R_i} = \frac{V_{in}}{R_i}$$

$$I_{R_F} = I_{R_i} = \frac{V_{in}}{R_i}$$

$$V_o = V_- - I_{R_F} \times R_F = 0 - \frac{V_{in}}{R_i} \times R_F$$



$$= - \frac{R_F}{R_i} \times V_{in}$$

Closed loop gain is independent of  $A_{od}$

$$A_{cl} = V_o / V_{in} = - \frac{R_F}{R_i}$$

## Non-inverting Amplifier:-

$$V_+ = V_{in}$$

$$V_- \approx V_+ = V_{in}$$

$$I_{Ri} = \frac{V_-}{R_i} = \frac{V_{in}}{R_i}$$

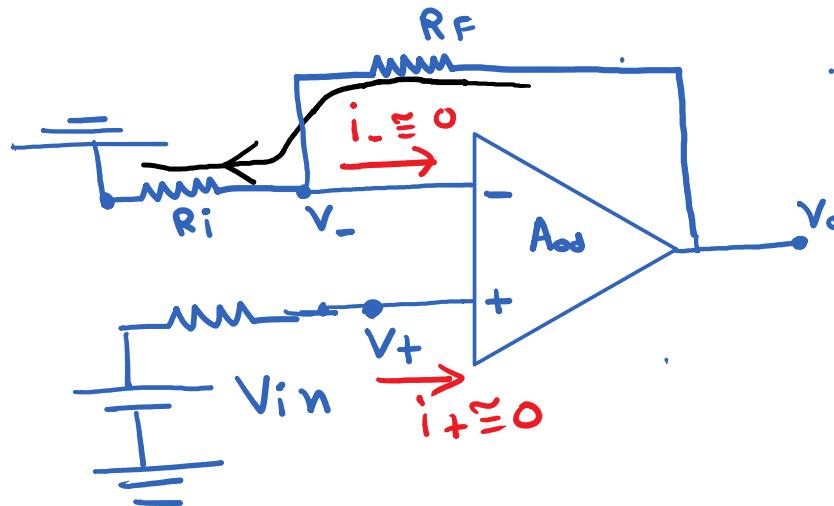
$$I_{RF} = I_{Ri} = \frac{V_{in}}{R_i}$$

$$V_o = V_- + I_{RF} \times R_F = V_{in} + \frac{V_{in}}{R_i} \times R_F$$

$$A_{ci} = \frac{V_o}{V_{in}} = \frac{R_i + R_F}{R_i}$$

$$= V_{in} \left( 1 + \frac{R_F}{R_i} \right)$$

$$= V_{in} \frac{R_i + R_F}{R_i}$$



## Class - 30

- 1) Infinite open loop gain ( $A_{od}$ ).
- 2) Infinite input impedance.
- 3) Zero output impedance.
- 4) In closed loop configuration  $V_+ \equiv V_-$ . This is also known as the property of virtual short.

## Integrator:

$$V_+ = 0$$

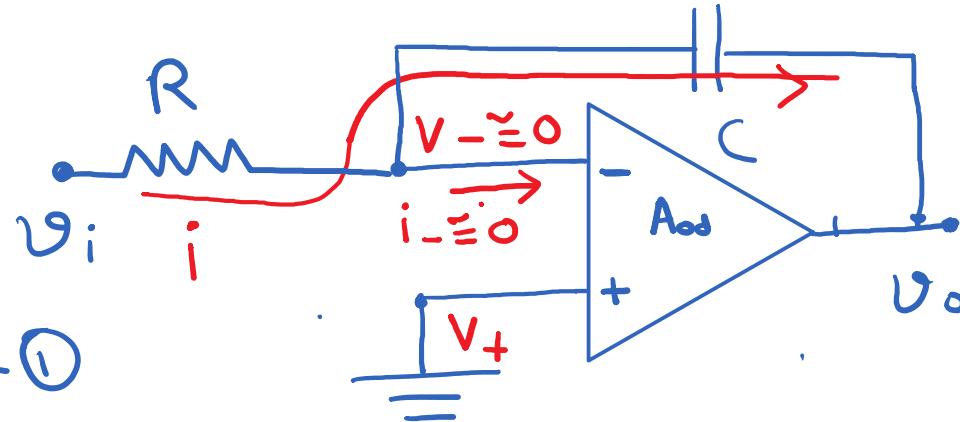
$$V_- \approx V_+ = 0$$

$$i = \frac{V_i - V_-}{R} \approx \frac{V_i}{R} \quad -\textcircled{1}$$

$$i = C \frac{d(V_- - V_o)}{dt} \approx -C \frac{dV_o}{dt} \quad -\textcircled{2}$$

$$-C \frac{dV_o}{dt} = \frac{V_i}{R} \Rightarrow \frac{dV_o}{dt} = -\frac{V_i}{CR}$$

$$V_o = -\frac{1}{CR} \int v_i(t) dt$$



This configuration also acts as a low pass filter

## Differentiator:

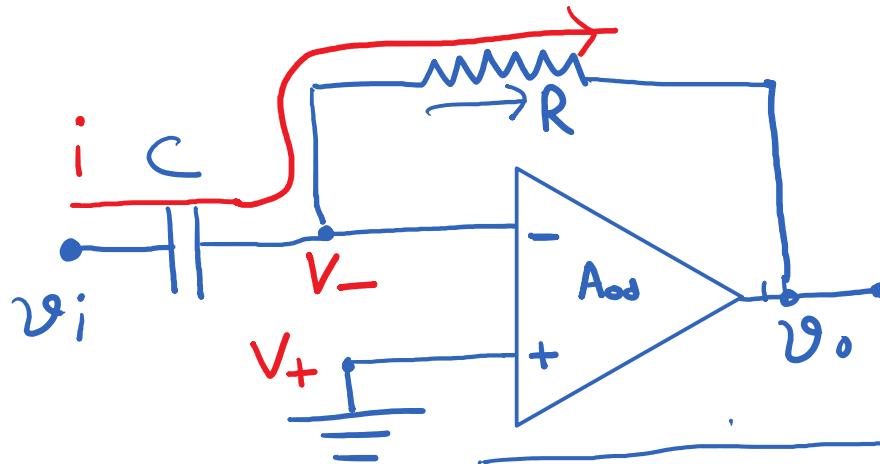
$$V_- \approx V_+ = 0$$

$$i = C \frac{d}{dt} (V_i - V_-)$$

$$\approx C \frac{dV_i}{dt} \quad -\textcircled{1}$$

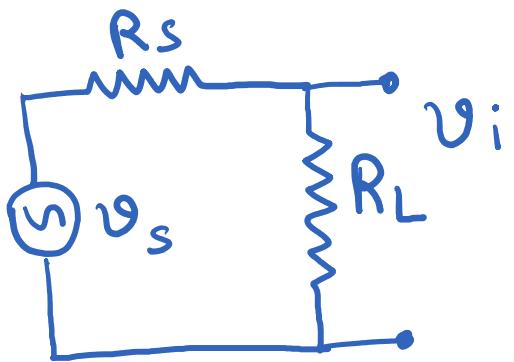
$$i = \frac{V_- - V_o}{R} \approx -\frac{V_o}{R} \quad -\textcircled{2}$$

$$C \frac{dV_i}{dt} = -\frac{V_o}{R} \Rightarrow V_o = -CR \frac{dV_i}{dt}$$



This configuration  
acts as a  
high pass  
filter

$$v_i = \frac{v_s \times R_L}{R_s + R_L}$$

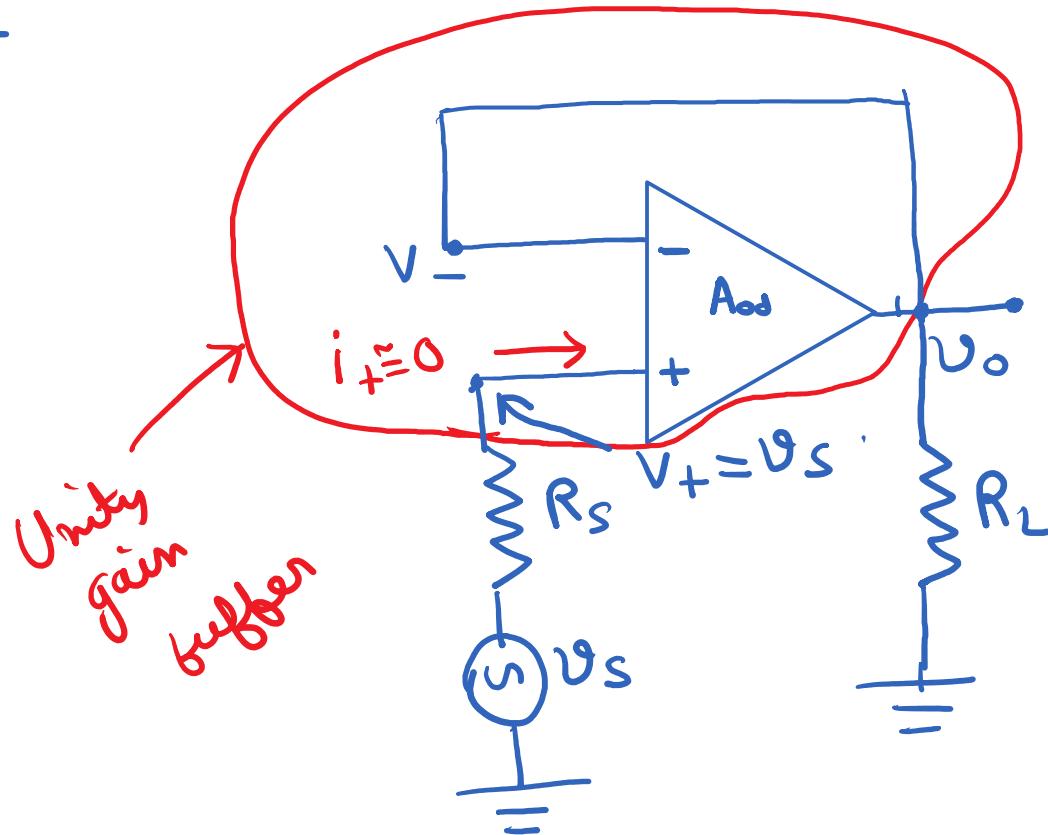


Unity gain buffer:-

$$V_- \approx V_+ = V_s$$

$$V_o = V_- = V_s$$

$$V_{R_L} = V_o = V_s$$



## Adder:

$$V_+ = 0$$

$$V_- \approx V_+ = 0$$

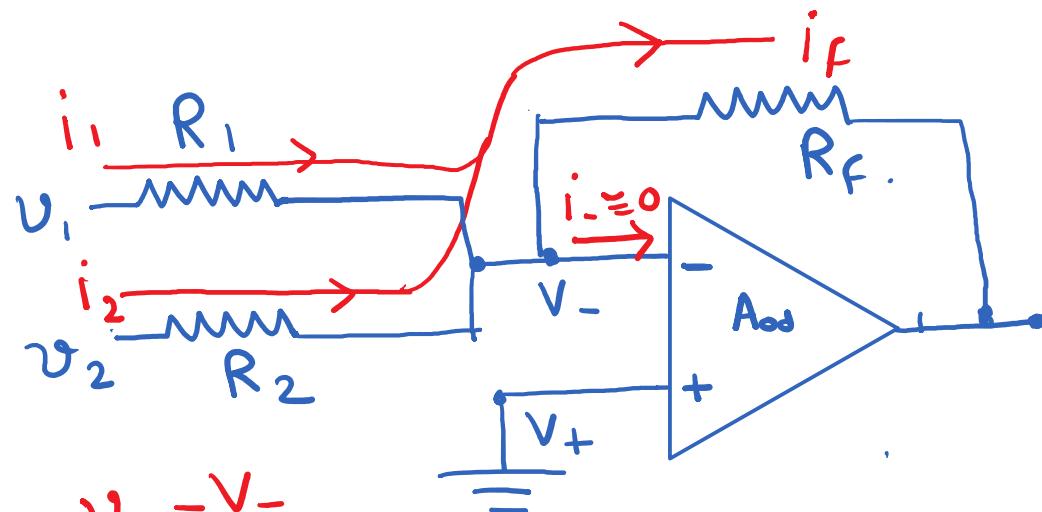
$$i_f = i_1 + i_2$$

$$\frac{V_- - V_o}{R_F} = \frac{V_1 - V_-}{R_1} + \frac{V_2 - V_-}{R_2}$$

$$\Rightarrow -\frac{V_o}{R_F} = \frac{V_1}{R_1} + \frac{V_2}{R_2}$$

$$\Rightarrow V_o = -\frac{R_F}{R_1} V_1 - \frac{R_F}{R_2} V_2$$

$$\text{If } R_1 = R_2 = R, \text{ then } V_o = -\frac{R_F}{R} (V_1 + V_2)$$



## Difference Amplifier :-

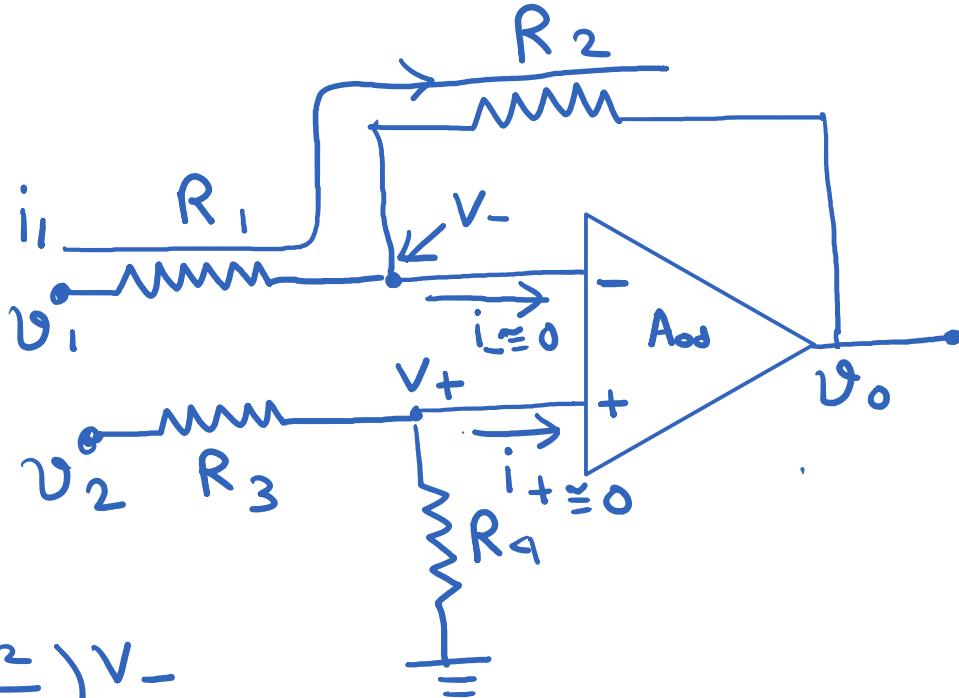
$$V_+ = \frac{R_4}{R_3 + R_4} V_2$$

$$V_- \equiv V_+ = \frac{R_4}{R_3 + R_4} V_2 \quad (1)$$

$$i_1 = \frac{V_1 - V_-}{R_1} = \frac{V_- - V_o}{R_2}$$

$$\Rightarrow V_o = \frac{R_2}{R_1} V_1 - \left(1 + \frac{R_2}{R_1}\right) V_-$$

$$= \frac{R_2}{R_1} V_1 - \left(1 + \frac{R_2}{R_1}\right) \left(\frac{R_4}{R_3 + R_4}\right) V_2$$



$$v_o = \frac{R_2}{R_1} v_1 - \left(1 + \frac{R_2}{R_1}\right) \frac{R_4}{R_3 + R_4} v_2$$

$$= \frac{R_2}{R_1} v_1 - \left(1 + \frac{R_2}{R_1}\right) \frac{\frac{R_4}{R_3}}{1 + \frac{R_4}{R_3}}$$

$$= \frac{R_2}{R_1} v_1 - \frac{R_4}{R_3} \frac{\left(1 + \frac{R_2}{R_1}\right)}{\left(1 + \frac{R_4}{R_3}\right)} v_2$$

If  $\frac{R_2}{R_1} = \frac{R_4}{R_3}$ , then  $v_o = \frac{R_2}{R_1} v_1 - \frac{R_2}{R_1} v_2$

$$= \frac{R_2}{R_1} (v_1 - v_2)$$

## Common-mode rejection ratio (CMRR).

$$\left. \begin{array}{l} v_{cm} = \frac{v_1 + v_2}{2} \\ v_d = (v_1 - v_2) \end{array} \right\} \Rightarrow \begin{array}{l} v_1 = v_{cm} + \frac{v_d}{2} \\ v_2 = v_{cm} - \frac{v_d}{2} \end{array}$$

$$A_{cm} = \frac{v_o}{v_{cm}} \quad \left| \begin{array}{l} v_1 = v_2 \text{ or } v_d = 0 \\ \text{ideal} \end{array} \right. \Rightarrow 0$$

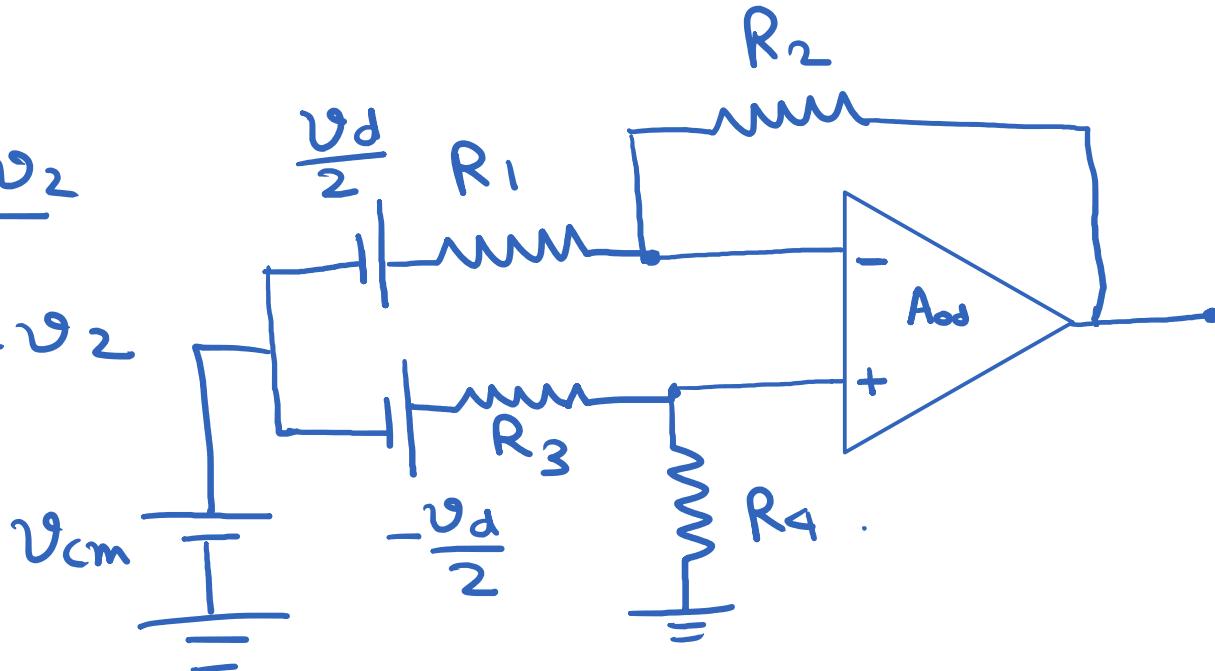
$$A_d = \frac{v_o}{v_d} \quad \left| \begin{array}{l} v_{cm} = 0 \\ \text{ideal} \end{array} \right. \Rightarrow \text{finite value}$$

$$CMRR = \frac{A_d}{A_{cm}} \quad \stackrel{\text{ideal}}{\Rightarrow} \infty$$

Higher value of CMRR means better difference amplifier.

$$V_{cm} = \frac{V_1 + V_2}{2}$$

$$V_d = V_1 - V_2$$



$$\text{CMRR (in dB)} = 20 \log_{10} \left| \frac{A_d}{A_{cm}} \right|.$$

Case I

$$V_1 = 200 \mu V$$

$$V_2 = -200 \mu V$$

$$\text{Output} = V_{o1}$$

$$V_1 - V_2 \\ = 400 \mu V$$

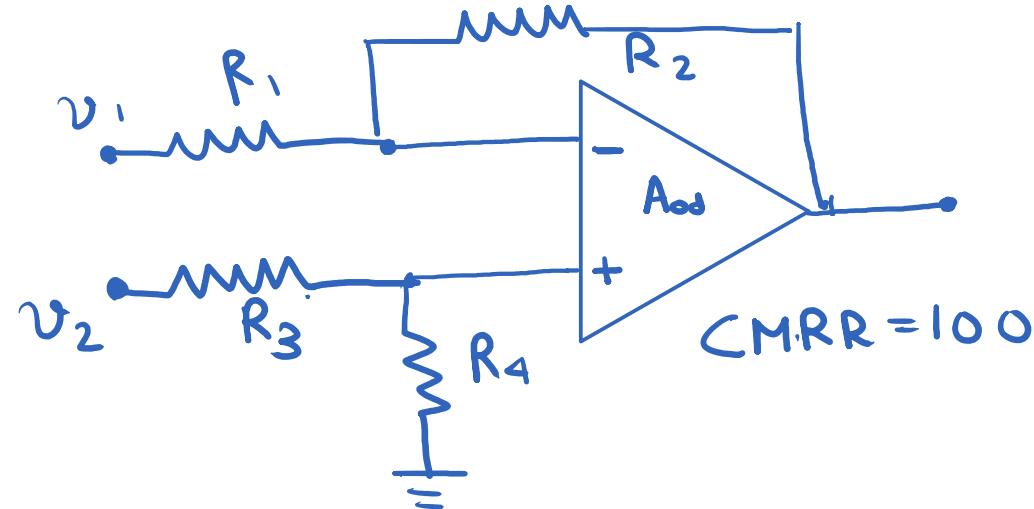
Case II

$$V_1 = 1200 \mu V$$

$$V_2 = 800 \mu V$$

$$\text{Output} = V_{o2}$$

$$V_1 - V_2 \\ = 400 \mu V$$



$$\begin{aligned} V_o &= f_1(V_1) + f_2(V_2) \\ &= f_1(V_{cm} + \frac{V_d}{2}) + f_2(V_{cm} - \frac{V_d}{2}) \\ &= A_{cm}(V_{cm}) + A_d(V_d) \end{aligned}$$

Case I

$$V_1 = 200 \mu V$$

$$V_2 = -200 \mu V$$

$$V_{cm}^1 = \frac{V_1 + V_2}{2}$$

$$= 0 \mu V$$

$$V_d^1 = V_1 - V_2$$

$$= 400 \mu V$$

$$V_{o1} = A_{cm} V_{cm}^1 + A_d V_d^1$$

$$= A_{cm} \times 0 + A_d \times 400 \mu V.$$

Case II

$$V_1 = 1200 \mu V$$

$$V_2 = 800 \mu V$$

$$V_{cm}^2 = \frac{V_1 + V_2}{2}$$

$$= 1000 \mu V$$

$$V_d^2 = V_1 - V_2$$

$$= 400 \mu V$$

$$V_{o2} = A_{cm} \times (1000 \mu V) + A_d \times 400 \mu V.$$

$V_{o1} - V_{o2}$

$$= A_d \times 400 \mu V$$

$$- A_{cm} \times (1000 \mu V)$$

$$- A_d \times 400 \mu V$$

$$= - A_{cm} \times (1000 \mu V)$$

Percentage difference between the outputs :-

$$\frac{V_{o1} - V_{o2}}{V_{o1}} \times 100\%$$

$$= \frac{-A_{cm} \times (1500 \mu V)}{A_d \times 400 \mu V} \times 100\%$$

$$= \frac{1000}{400} \times \frac{1}{\frac{A_d}{A_{cm}}} \times 100\% = 2.5 \times \frac{1}{CMRR} \times 100\%$$

$$= 2.5 \times \frac{1}{100} \times 100\% = 2.5\%$$

## Class-31

Digital to Analog Converter:-

$$\text{MSB} \rightarrow b_3 b_2 b_1 b_0 \xrightarrow{\text{LSB}} 2^3 \times b_3 + 2^2 \times b_2 + 2^1 \times b_1 + 2^0 \times b_0$$

$$\begin{aligned} 1101 &\rightarrow 1 \times 2^3 + 1 \times 2^2 + 0 \times 2^1 + 1 \times 2^0 \\ &= 8 + 4 + 0 + 1 = 13 \end{aligned}$$

$$V_+ = 0$$

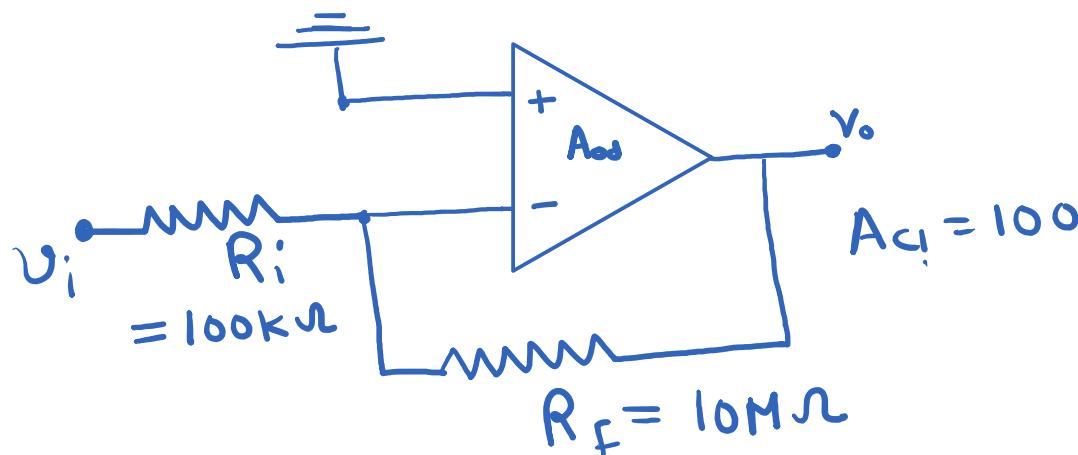
$$V_- \approx V_+ = 0$$

$$\begin{aligned} & \frac{b_3 - V_-}{R} + \frac{b_2 - V_-}{2R} \\ & + \frac{b_1 - V_-}{4R} + \frac{b_0 - V_-}{8R} \\ & = \frac{V_- - V_o}{R_F} \end{aligned}$$

$$\begin{aligned} \Rightarrow V_o &= -\frac{R_F}{R} \left[ b_3 + \frac{b_2}{2} + \frac{b_1}{4} + \frac{b_0}{8} \right] \quad (\because V_- \approx 0) \\ &= -\frac{R_F}{8R} \underbrace{\left[ 2^3 b_3 + 2^2 b_2 + 2 b_1 + 2^0 b_0 \right]}_{\text{Analog value}} \end{aligned}$$

## Amplifier with T-network :-

Gain in inverting config :-  $- \frac{R_f}{R_i}$



$$A_{ci} = 100$$

Amplifier with T-network:-

$$V_+ = 0$$

$$V_- \equiv V_+ = 0$$

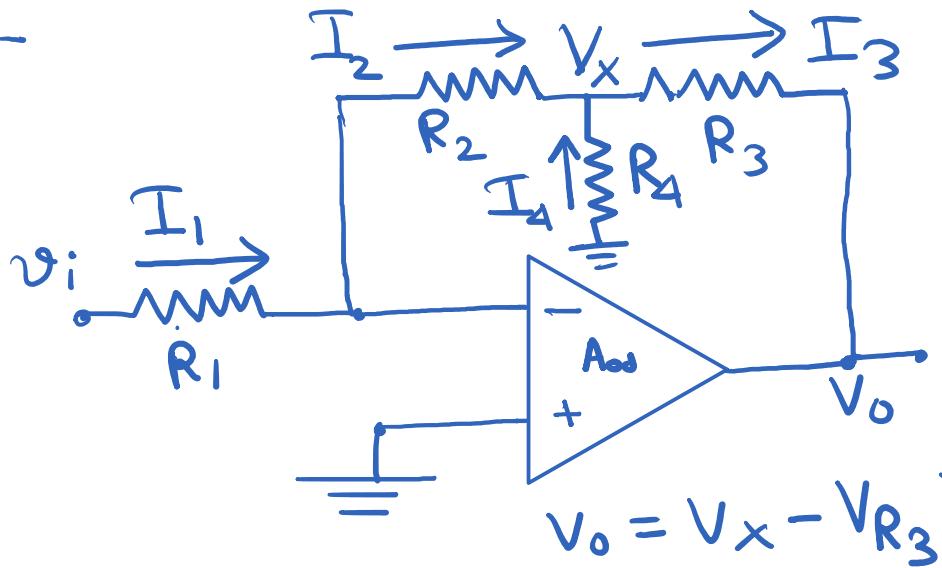
$$I_1 = \frac{V_i - V_-}{R_1} = \frac{V_i}{R_1}$$

$$I_2 = \frac{V_- - V_x}{R_2} = -\frac{V_x}{R_2}$$

$$I_1 = I_2$$

$$\Rightarrow \frac{V_i}{R_1} = -\frac{V_x}{R_2} \Rightarrow V_x = -V_i \times \frac{R_2}{R_1}$$

$$I_4 = \frac{0 - V_x}{R_4} = -\left(-V_i \times \frac{R_2}{R_1}\right) \frac{1}{R_4} = V_i \times \frac{R_2}{R_1 R_4}$$



$$I_3 = I_2 + I_4$$

$$= \frac{v_i}{R_1} + v_i \times \frac{R_2}{R_1 R_4}.$$

$$V_o = V_x - V_{R_3}$$

$$= v_i \times \frac{R_2}{R_1} - I_3 R_3$$

$$= -v_i \times \frac{R_2}{R_1} - \left( \frac{v_i}{R_1} + \frac{v_i R_2}{R_1 R_4} \right) R_3$$

$$= -v_i \left( \frac{R_2}{R_1} + \frac{R_3}{R_1} + \frac{R_2 R_3}{R_1 R_4} \right)$$

$$= -v_i \times \frac{R_2}{R_1} \underbrace{\left( 1 + \frac{R_3}{R_2} + \frac{R_3}{R_4} \right)}$$

Gain increases by this factor.

### Class - 32

$$V_+ = V_1 \times \frac{R_4}{R_3 + R_4} + V_2 \times \frac{R_3}{R_3 + R_4}$$

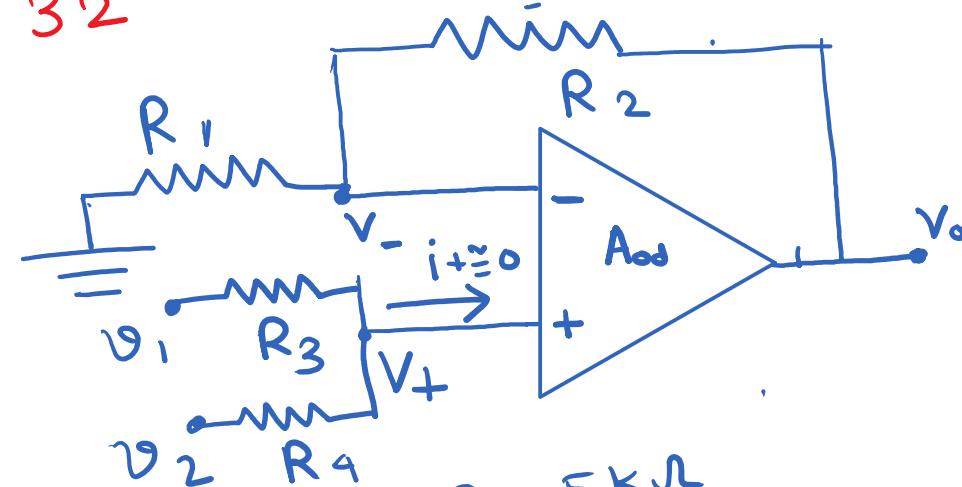
$$V_- \equiv V_+ = \frac{V_1 R_4}{R_3 + R_4} + \frac{V_2 R_3}{R_3 + R_4}$$

$$V_o = \left(1 + \frac{R_2}{R_1}\right) V_+$$

$$= \left(1 + \frac{R_2}{R_1}\right) \left( \frac{V_1 R_4}{R_3 + R_4} + \frac{V_2 R_3}{R_3 + R_4} \right)$$

$$A_{V_1} = \left. \frac{V_o}{V_1} \right|_{V_2=0} = \left(1 + \frac{R_2}{R_1}\right) \frac{R_4}{R_3 + R_4}$$

$$= \left(1 + \frac{70}{75}\right) \times \frac{50}{75} = 10$$



$$R_1 = 5\text{k}\Omega$$

$$R_2 = 70\text{k}\Omega$$

$$R_3 = 25\text{k}\Omega$$

$$R_4 = 50\text{k}\Omega$$

$$A_{V_1} = \left. \frac{V_o}{V_1} \right|_{V_2=0}$$

$$A_{V_2} = \left. \frac{V_o}{V_2} \right|_{V_1=0}$$

$$A v_2 = \frac{V_o}{v_2} |_{v_1=0}$$

$$A_{v_2} = \frac{V_o}{V_2} \Big|_{V_1=0}$$

$$\begin{aligned}& \cancel{R} \left( 1 + \frac{R_2}{R_3 + R_4} \right) \frac{R_3}{R_4} \\&= \left( 1 + \frac{70}{5} \right) \times \frac{25}{75} := 5\end{aligned}$$

$$A_{v_1} = 10$$

$$A_{v_2} = 5$$

$$V_+ = \frac{V_2 R_4}{R_3 + R_4} + \frac{V_3 R_3}{R_3 + R_4}$$

$$V_- \equiv V_+ = \frac{V_2 R_4}{R_3 + R_4} + \frac{V_3 R_3}{R_3 + R_4}$$

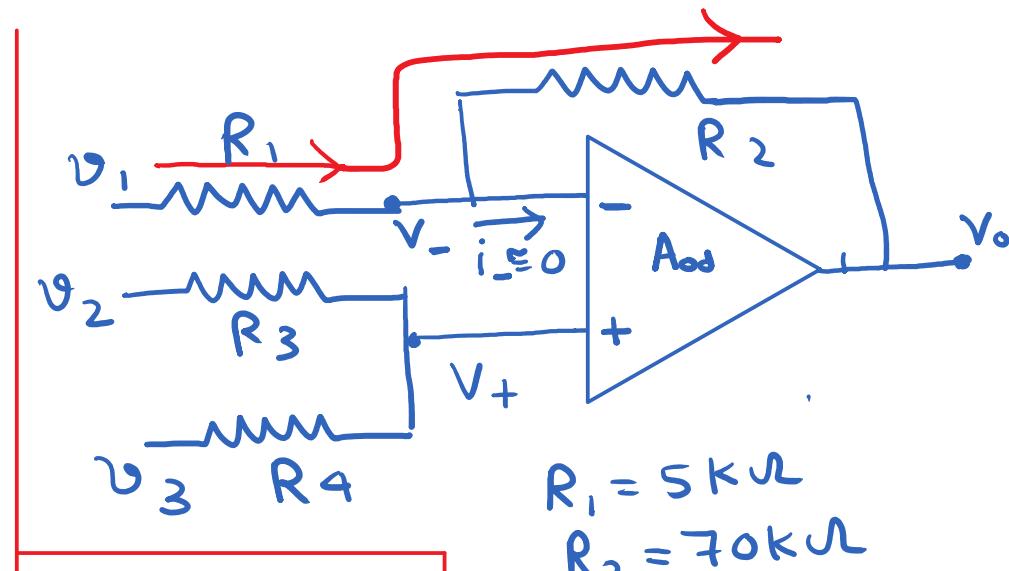
$$\frac{V_1 - V_-}{R_1} = \frac{V_- - V_o}{R_2}$$

$$V_o = V_- + \frac{R_2}{R_1} (V_- - V_1)$$

$$= V_- \left(1 + \frac{R_2}{R_1}\right) - \frac{R_2}{R_1} V_1$$

$$= \left(\frac{V_2 R_4}{R_3 + R_4} + \frac{V_3 R_3}{R_3 + R_4}\right) \left(1 + \frac{R_2}{R_1}\right) - \frac{R_2}{R_1} V_1$$

$$= \left(V_2 \frac{50}{75} + V_3 \times \frac{25}{75}\right) \left(1 + \frac{70}{5}\right) - \frac{70}{5} V_1$$



$$R_1 = 5 \text{ k}\Omega$$

$$R_2 = 70 \text{ k}\Omega$$

$$R_3 = 25 \text{ k}\Omega$$

$$R_4 = 50 \text{ k}\Omega$$

$$A_{V1} = \frac{V_o}{V_1} \Big|_{V_2 = V_3 = 0}$$

$$A_{V2} = \frac{V_o}{V_2} \Big|_{V_1, V_3 = 0}$$

$$A_{V3} = \frac{V_o}{V_3} \Big|_{V_1, V_2 = 0}$$

$$= \left( v_2 \frac{50}{75} + v_3 \times \frac{-1}{75} \right) \left( 1 + \frac{1}{5} \right) - \frac{1}{5} v_1$$

$$A v_3 = \frac{v_0}{v_3} \quad | \quad v_1, v_2 = 0$$

$$= \left( \frac{2}{3}v_2 + \frac{1}{3}v_3 \right) \times 15 - 14v_1$$

$$= -14v_1 + 10v_2 + 5v_3$$

$$Av_1 = \frac{v_o}{v_1} \Big|_{v_2=v_3=0} = -14$$

$$Av_2 = \frac{v_o}{v_2} \Big|_{v_1=v_3=0} = 10$$

$$Av_3 = \frac{v_o}{v_3} \Big|_{v_1=v_2=0} = 5$$

Ans

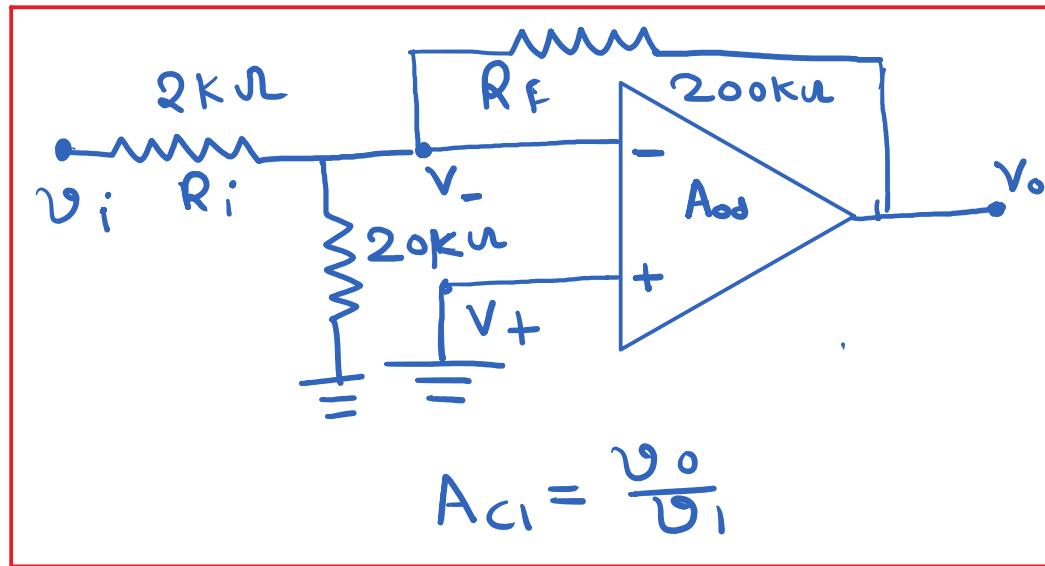
$$V_+ = 0$$

$$V_- \equiv V_+ = 0$$

$$A_{C1} = -\frac{R_F}{R_i}$$

$$= -\frac{200k\Omega}{2k\Omega}$$

$$= -100$$



$$A_{C1} = \frac{v_o}{v_i}$$

$$V_- \approx V_+ = 0$$

$$I_1 = \frac{V_i - V_-}{R_i} = \frac{V_i}{R_i} = \frac{V_i}{2}$$

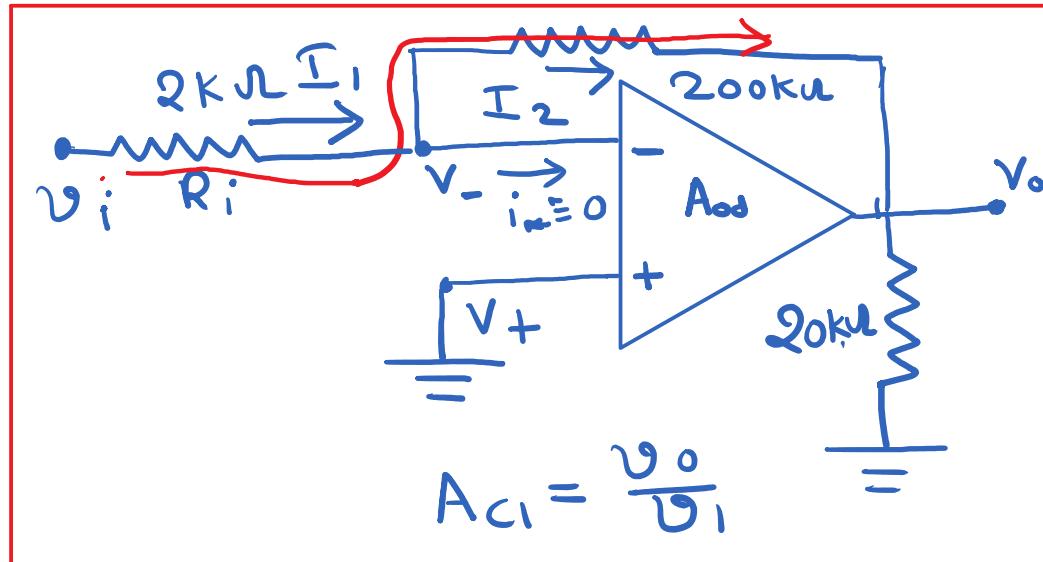
$$I_2 = \frac{V_- - V_o}{200} = -\frac{V_o}{200}$$

$$I_1 = I_2$$

$$\Rightarrow \frac{V_i}{2} = -\frac{V_o}{200}$$

$$\Rightarrow \frac{V_o}{V_i} = -\frac{200}{2} = -100$$

$$A_{ci} = -100$$



$$V_- \cong V_+ = 0$$

$$I_1 = \frac{v_i - v_-}{R} = \frac{v_i}{R}$$

$$I_2 = I_1 = \frac{v_i}{R}$$

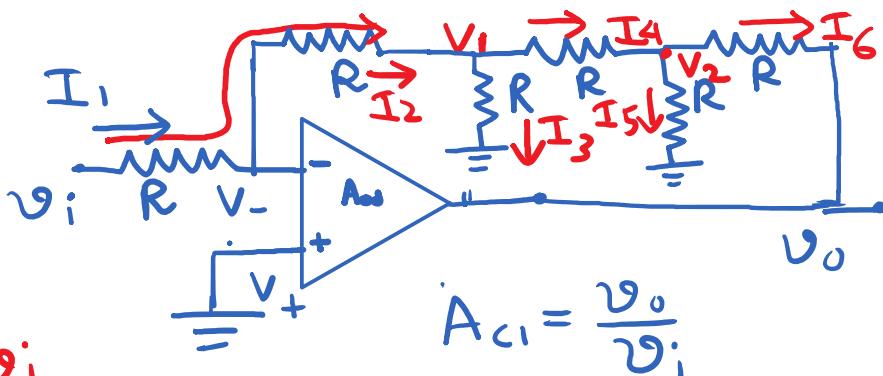
$$V_1 = V_- - I_2 R = - \frac{v_i}{R} \times R = -v_i$$

$$\Rightarrow V_1 = -v_i$$

$$I_3 = \frac{V_1 - 0}{R} = -\frac{v_i}{R}$$

$$I_4 = I_2 - I_3 = \frac{2v_i}{R}$$

$$\begin{aligned} V_2 &= V_1 - I_4 \times R \\ &= -v_i - \frac{2v_i}{R} \times R \\ &= -3v_i \end{aligned}$$



$$I_5 = \frac{V_2 - 0}{R} = -\frac{3v_i}{R}$$

$$I_6 = I_4 - I_5$$

$$= \frac{2v_i}{R} - \left( -\frac{3v_i}{R} \right)$$

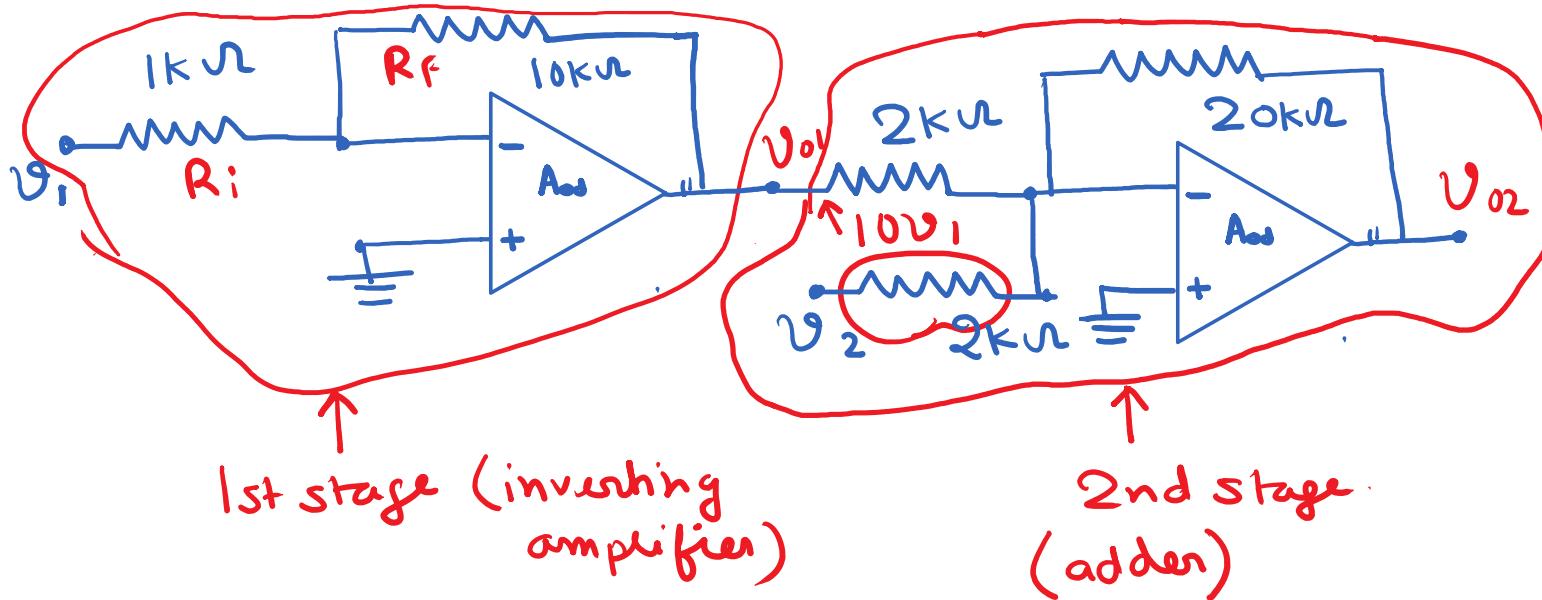
$$= \frac{5v_i}{R}$$

$$\begin{aligned} V_o &= V_2 - I_6 R = -3v_i \\ &\quad - \left( \frac{5v_i}{R} \right) R \end{aligned}$$

$$= -3v_i \quad \left| \begin{array}{l} v_o = v_2 - 16^{\circ} - \left( \frac{5v_i}{R} \right) R \end{array} \right.$$

$$= -8v_i$$

$$A_{ci} = \frac{v_o}{v_i} = -8$$



$$\text{For the 1st stage: } A_{v1,1} = -\frac{R_F}{R_i} = -\frac{10}{1}$$

$$\Rightarrow v_{01} = -10 v_1$$

$$\begin{aligned}
 v_{02} &= -\frac{20}{2} \times v_{01} - \frac{20}{2} \times v_2 = -10v_{01} - 10v_2 \\
 &= -10 \times (-10v_1) - 10v_2 \\
 &= 100v_1 - 10v_2
 \end{aligned}$$