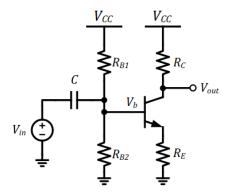
# **Assignment 02**

EE 538 Spring 2020 Analog Circuits for Sensor Systems University of Washington Electrical & Computer Engineering

Due: April 18, 2020 Author: Kevin Egedy

```
In [1]: # Imports
    import os
    import sys
    import math
    import matplotlib.pyplot as plt
    import matplotlib
    import numpy as np
    import pandas as pd
    import ltspice
    import sympy as sp
    from scipy import signal
    %matplotlib inline
    from IPython.core.interactiveshell import InteractiveShell
    InteractiveShell.ast_node_interactivity = "all"
To [2]: def read ltspice trans(file page))
```

### Problem 1: Common-emitter amplifier



For the following, use the figure and the simplified NPN model from Section 2.1.1 in AoE.  $V_{CC}$  = 5V, C = 1 $\mu$ F, and  $\beta$  = 100. Use the npn Ltspice component for your simulations.

- a) Select values for  $R_{B1}$  and  $R_{B2}$  to achieve  $V_B$  = 1V and a high-pass corner frequency ( $f_{3dB}$ ) of 10Hz. You may ignore base current for this step.
- b) Choose  $R_C$  and  $R_E$  for a gain of -10V/V (at 10kHz) and a collector current of 1mA. What is the DC value of  $V_{CE}$ ?
- c) Derive the transfer function for the small-signal gain  $v_{out}/v_{in}$  and plot the frequency response (magnitude and phase) in MATLAB/Python. It may be convenient to separate the transfer function into two parts and compute their product:

$$\frac{v_{out}}{v_{in}} = \left(\frac{v_{out}}{v_b}\right) \left(\frac{v_b}{v_{in}}\right)$$

d) Perform an AC simulation of the circuit in Ltspice. Export the frequency response data for comparison to the transfer function you derived in Part c and plot them together. Explain any observed discrepancies between the two.

## Part A

Given:

 $f_{
m 3dB,HP}=10{
m Hz}$  and  $V_B=1V$  .

Solve

$$egin{align*} f_{
m 3dB} &= rac{1}{2\pi RC} \ &10 {
m Hz} = rac{1}{2\pi R_{
m eq} C}igg|_{C=1 {
m pF}} \ &R_{
m eq} &= R_{B_2}//R_{B_1} = 15915 \Omega \ \end{array}$$

and

$$egin{array}{ll} V_{
m out} &= V_{
m in} rac{R_2}{R_1 + R_2} \ V_b &= V_{
m in} rac{R_{B_2}}{R_{B_1} + R_{B_2}} \ 1 &= 5 rac{R_{B_2}}{R_{B_1} + R_{B_2}} \end{array}$$

then

$$R_{B_2}=19893, R_{B_1}=79572$$

# Part B

Find 
$$\frac{v_{\mathrm{out}}}{v_B}$$

$$egin{array}{lll} I_E & = eta I_B & = rac{V_E}{R_E} \ I_C & = (eta+1)I_B & = rac{V_{CC}-V_{
m out}}{R_C} \ V_E & = V_B-0.6 & pprox V_S-0.6 \ I_E & pprox I_C & {
m for} \ eta>= 100 {
m then} \ rac{V_S-0.6}{R_E} & = rac{V_{CC}}{V_{
m out}} \ V_{
m out} & = V_{CC} - rac{R_C}{R_E} (V_S-0.6) \ V_{
m out} & = -rac{R_C}{R_E} V_S + V_{DC} \ \end{array}$$

Given  $V_B=1\mathrm{V}$  and  $I_C=1\mathrm{mA}$ , solve

$$egin{array}{lll} V_{
m out} & = V_{CC} - rac{R_C}{R_E} (V_S - 0.6) \ & V_{
m out} & = 5 - 10 \cdot (1 - 0.6) \ & V_{
m out} & = 1 {
m V} \ & {
m And} \ & I_C & = rac{V_{CC} - V_{
m out}}{R_C} \ & \end{array}$$

$$I_C = rac{V_{CC} - V_{
m out}}{R_C}$$
 $1 {
m mA} = rac{5-1}{R_C}$ 

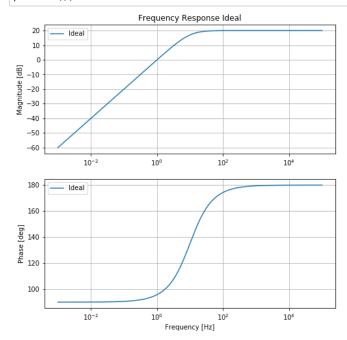
then

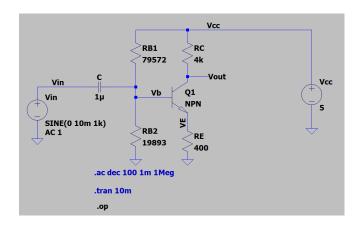
$$R_C = 4k\Omega, R_E = 400\Omega$$

Find  $V_{CE}$ 

$$V_C = V_{
m out} = 1$$
 and  $V_E = 1 - 0.6 = 0.4$  so  $V_{CE} = 0.6$ .

Part C





voltage	0.177253	V(ve)
voltage	5	V(vcc)
voltage	3.24502	V(vout)
voltage	0.930176	V(vb)
voltage	0	V(vin)
device_current	0.000438746	Ic(Q1)
device_current	4.38746e-06	lb(Q1)
device_current	-0.000443134	le(Q1)
device_current	9.30176e-19	I(C)
device_current	4.6759e-05	I(Rb2)
device_current	5.11464e-05	I(Rb1)
device_current	0.000438746	I(Rc)
device_current	0.000443133	I(Re)
device_current	9.30176e-19	I(Vin)
device_current	-0.000489892	I(Vcc)

The DC operating point has key differences from the quick calculations.

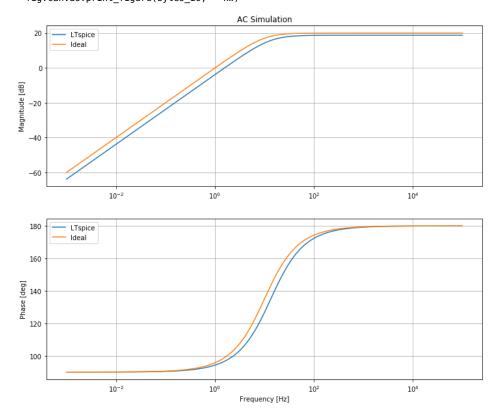
Simulated Results	Calculated Results
$V_{ m out}=3.24$ V	$V_{ m out}=1$ V
$V_{ m BE}=0.75$ V	$V_{ m BE}=0.6$ V
$V_{ m B}=0.93$ V	$V_{ m B}=1$ V
$I_C=438\mu$ A	$I_C=1$ mA

```
In [4]: filepath = 'data/common_emitter.txt'
    df = read_ltspice_tran(filepath)
    df['Vout_Mag'] = df['V(vout)'].apply(lambda x: x.split(',')[0])
    df['Vout_Mag'] = df['Vout_Mag'].apply(lambda x: x[1:-2])
    df['Vout_Mag'] = df['Vout_Mag'].astype('float64')
    df['Vout_Phase'] = df['V(vout)'].apply(lambda x: x.split(',')[1])
    df['Vout_Phase'] = df['Vout_Phase'].apply(lambda x: x[1:-2])
    df['Vout_Phase'] = df['Vout_Phase'].astype('float64')
    df['Freq.'] = df['Freq.'].astype('float64')
    df = df[df['Freq.']<=1e5]</pre>
```

```
In [5]:
    fig, axs = plt.subplots(2,figsize=(12,10))
    freq = df['Freq.']
    mag = df['Vout_Mag']
    ang = df['Vout_Phase']

    axs[0].set_title('AC Simulation')
    axs[0].semilogx(freq, mag, label='LTspice')
    axs[0].semilogx(f, 20*np.log10(abs(H)),label='Ideal')
    axs[0].set_ylabel('Magnitude [dB]')
    axs[0].grid()
    axs[0].legend()
    axs[1].semilogx(freq, ang, label='LTspice')
    axs[1].semilogx(f, -np.angle(H,deg=True),label='Ideal')
    axs[1].set_ylabel('Phase [deg]')
    axs[1].set_xlabel('Frequency [Hz]')
    axs[1].legend()
    plt.show();
```

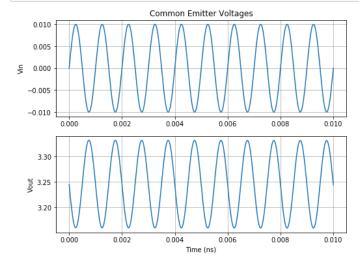
/home/kegedy/anaconda3/lib/python3.7/site-packages/IPython/core/pylabtools.py:128: UserWarning: Creating
legend with loc="best" can be slow with large amounts of data.
fig.canvas.print\_figure(bytes\_io, \*\*kw)



The AC analysis for the simplified calculation and simulation are very similar.

Gain is 8.6501

```
In [7]: fig, axs = plt.subplots(2,figsize=(8,6))
    axs[0].set_title('Common Emitter Voltages')
    axs[0].plot(df['time'], df['V(vin)'],label='Vin')
    axs[0].set_ylabel('Vin')
    axs[0].grid()
    #xs[0].legend()
    axs[1].plot(df['time'], df['V(vout)'],label='Vout')
    axs[1].set_ylabel('Vout')
    axs[1].set_xlabel('Time (ns)')
    axs[1].grid()
    #xs[1].legend()
    plt.show();
```



### Problem 2: Emitter follower

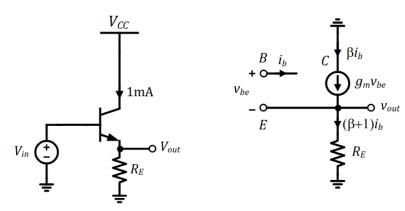


Figure 2a. Emitter follower

Figure 2b. Small-signal equivalent circuit

Use the Ebers-Moll model of the BJT and the figures to answer the following questions.  $V_{CC} = 5V$ ,  $I_S =$  $10^{-16}$ , and  $\beta = 100$ . When determining input/output resistances, connect a test voltage to the smallsignal circuit and determine the resistance as  $r = v_{test}/i_{test}$ . Use the npn Ltspice component for your simulations.

- a) Design the biasing of the emitter follower (i.e. determine the DC value of  $V_{in}$  and the resistance value  $R_E$ ) such that the collector current is 1mA and the DC level of  $V_{out}$  is 1V. You can do this by hand or use a MATLAB/Python script.
- b) Use the small-signal model (Fig. 2b) to determine the small-signal input resistance of the circuit.
- Use the small-signal model (Fig. 2b) to determine the small-signal output resistance of the
- Verify your design in Ltspice and include all relevant SPICE schematics and results in your submission. At minimum, your submission should include:
  - 1. DC simulation results indicating the bias point
  - 2. AC simulation result for input resistance
  - 3. AC simulation result for output resistance

#### Part A

Small signal analysis:

$$V_{
m out} = V_{
m out,DC} + v_{
m out} = V_{
m out,DC} + A_v v_{
m in}$$

Ebers-Moll Model: 
$$I_C = I_S(T)(e^{\frac{V_{BE}}{V_T}}-1) \qquad pprox \ I_Se^{\frac{V_{BE}}{V_T}}igg|_{V_T=\frac{kT}{q}=25.3 \mathrm{mV}} \ V_{BE} = \frac{kT}{q} \mathrm{ln} \, (\frac{I_C}{I_S(T)}+1) \qquad pprox \ \frac{kT}{q} \mathrm{ln} \, (\frac{I_C}{I_S}) \ V_{BE} = 25.3 \cdot 10^{-3} \cdot \mathrm{ln} \, (\frac{1 \cdot 10^{-3}}{10^{-16}}) = 0.757 V$$

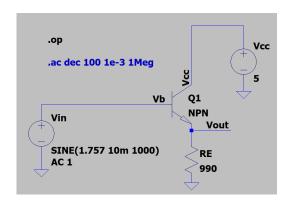
$$\begin{split} &\text{Solve:} \\ &I_{C} = \beta I_{B} \bigg|_{I_{C} = 10^{-3}, \beta = 100} \to I_{B} = 10^{-5} \\ &I_{E} = (\beta + 1)I_{B} \\ &V_{\text{out}} = I_{E} \cdot R_{E} \bigg|_{V_{\text{out}} = 1} \to R_{E} = 990 \\ &V_{\text{in}} = V_{\text{out}} + V_{BE} = 1.757 \text{V} \end{split}$$

# Part B

$$egin{aligned} i_e &= rac{v_e}{R_E} &= rac{v_{ ext{test}}}{R_E} \ i_{ ext{test}} &= rac{i_e}{eta+1} = rac{v_{ ext{test}}}{R_E(eta+1)} \ r_{ ext{in}} &= rac{v_{ ext{test}}}{i_{ ext{test}}} &= R_E(eta+1) igg|_{R_E=990} = 99990\Omega \end{aligned}$$

#### Part C

$$egin{array}{ll} i_e &= -i_{ ext{test}} \ v_{ ext{test}} &= i_e \cdot R_E \ r_{ ext{out}} &= rac{v_{ ext{test}}}{i_{ ext{test}}} &= R_E = 990 \Omega \end{array}$$



# **DC Operating Point**

voltage	0.983204	V(vout)
voltage	5	V(vcc)
voltage	1.757	V(vb)
device_current	0.000983307	Ic(Q1)
device_current	9.83307e-06	lb(Q1)
device_current	-0.00099314	le(Q1)
device_current	0.000993135	I(Re)
device_current	-9.83302e-06	I(Vin)
device_current	-0.000983302	I(Vcc)

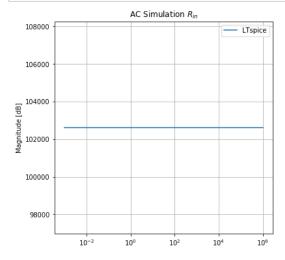
The DC Operating point closes reflects the calculations from Ebers-Moll model. The simulated value of  $I_C=0.98 {
m mA}$  is very close to the design of 1mA. Also simulated  $V_{
m out}=0.98$  is very close to 1V.

#### **AC Simulation for Rin**

```
In [8]: filepath = 'data/emitter_follower_rin.txt'
    df = read_ltspice_tran(filepath)
    df['Rin'] = df['1/Ib(Q1)'].apply(lambda x: x.split(',')[0])
    df['Rin'] = df['Rin'].astype('float64')
    df['Freq.'] = df['Freq.'].astype('float64')
    print(f'Rout is {round(max(df["Rin"]),2)}')
Rout is 102620.39
```

```
In [9]: fig, ax = plt.subplots(1,figsize=(6,6))
    freq = df['Freq.']
    mag = df['Rin']

ax.set_title(r'AC Simulation $R_{in}$')
    ax.semilogx(freq, mag, label='LTspice')
    ax.set_ylabel('Magnitude [dB]')
    ax.grid()
    ax.legend()
    plt.show();
```



The difference between  $R_{
m in}$  calculations and the simulation is  $rac{(102620.39-99990)}{99990}=2.6\%.$ 

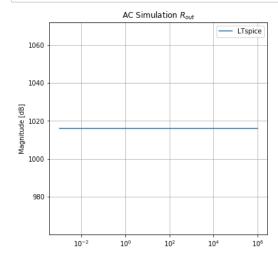
## AC Simulation for Rout

```
In [10]: filepath = 'data/emitter_follower_rout.txt'
    df = read_ltspice_tran(filepath)
    df['Rout'] = df[']/Ie(01)'].apply(lambda x: x.split(',')[0])
    df['Rout'] = df['Rout'].astype('float64')
    df['Freq.'] = df['Freq.'].astype('float64')
    print(f'Rout is {round(-max(df["Rout"]),2)}')
```

Rout is 1016.04

```
In [11]: fig, ax = plt.subplots(1,figsize=(6,6))
    freq = df['Freq.']
    mag = -df['Rout']

ax.set_title(r'AC Simulation $R_{out}$')
    ax.semilogx(freq, mag, label='LTspice')
    ax.set_ylabel('Magnitude [dB]')
    ax.grid()
    ax.legend()
    plt.show();
```



The difference between  $R_{
m out}$  calculations and the simulation is  $rac{(1016.04-990)}{990}=2.6\%$ 

In [ ]: