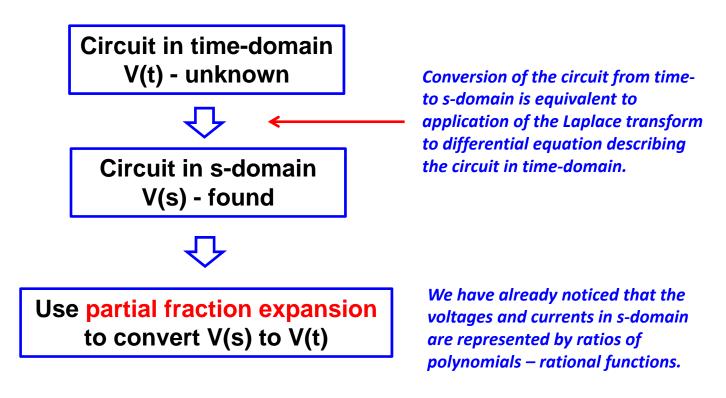
Last time: Laplace transform for circuit analysis.

- 1. Find the Laplace transforms of an important functions and build a table for future reference.
- **ALMOST DONE**
- 2. Develop technique to go from s-domain back to time-domain.
- **EXAMPLES WERE CONSIDERED**

3. Develop circuit analysis techniques in s-domain.

STILL WORKING

Let's discuss again how to get back to time domain



Partial fraction expansion of proper rational functions.

$$V(s) = Coust. \frac{s^m + a_{m-1} \cdot s^{m-1} + \dots + a_1 \cdot s + a_0}{s^n + b_{n-1} \cdot s^{n-1} + \dots + b_1 \cdot s + b_0}$$
; $m < N$

1. Find the roots of the denominator (poles) and represent V(s) as:

$$V(s) = Count. \frac{s^m + a_{m-1} - s^{m-1} + \dots + a_i \cdot s + a_0}{(s-p_i)^{k_i}}$$
, $N = \sum_{i=1}^{\ell} \kappa_i$

2. Find the coefficients in the partial fraction expansion for each pole:

$$\frac{\lambda_{k_1}}{\lambda_{k_2}} = \frac{A_{k_1}}{(s-p_1)^{k_1}} + \frac{A_{k_2-1}}{(s-p_1)^{k_2-1}} + \dots + \frac{A_1}{s-p_1} + \dots$$

3. Find the time-domain function corresponding to each term in partial fraction expansion:

$$L\left[e^{-\alpha t},h(t)\right] = \frac{1}{s+\alpha} \quad \text{and} \quad L\left[\frac{t^{n-1}}{(n-1)!}e^{-\alpha t}\right] = \frac{1}{(s+\alpha)^n}$$

Complex roots always come in conjugate pairs.

$$\begin{aligned}
& P_{z} = P_{1}^{*} \quad ; P_{1} = x + j P_{3} \\
& \tilde{V}(s) = \frac{1}{(s - p_{1})(s - p_{2})} = \frac{A}{s - p_{1}} + \frac{B}{s - p_{2}} \\
& A = \widetilde{V}(s)(s - p_{1}) - B \cdot \frac{s - p_{1}}{s - p_{2}} = \frac{1}{s - p_{2}} - B \cdot \frac{s - p_{1}}{s - p_{2}} = \frac{1}{p_{1} - p_{2}} \\
& OR \quad A = \widetilde{V}(s)(s - p_{1}) \Big|_{s = p_{1}} = \frac{1}{P_{1} - p_{2}} \\
& B = \widetilde{V}(s)(s - p_{2}) \Big|_{s = p_{2}} = \frac{1}{p_{2} - p_{1}} = A^{*}
\end{aligned}$$

$$V(t) = A e^{p_{1}t} + A^{*}e^{p_{2}t} = A e^{\alpha t}e^{ipt} + A^{*}e^{\alpha t}e^{-jnt}$$

$$A = |A|e^{ipt}, \quad A^{*} = |A|e^{-jpt}$$

$$V(t) = |A|e^{\alpha t} \left[e^{i(nt+pt)} \right] = 2|A|e^{\alpha t}\cos(\omega t + p)$$

Resistor in s-domain.

time-domain
$$\frac{+ \bigvee_{R}(t)}{\prod_{R}(t)} = \frac{\bigvee_{R}(t)}{R}$$

$$L \prod_{R}(t) = L \left[\frac{\bigvee_{R}(t)}{R} \right] = \frac{L \prod_{R}(t)}{R}$$

$$\frac{Z_{R}(s) = R}{\bigvee_{R}(s)} \qquad \qquad \frac{Z_{R}(s)}{\bigcap_{R}(s)} \qquad \qquad \frac{Z_{R}(s)}{\bigcap_{R}(s)} \qquad \qquad \frac{Z_{R}(s)}{\bigcap_{R}(s)} \qquad \qquad Z_{R}(s)$$

$$= \frac{Z_{R}(s)}{\bigcap_{R}(s)} \qquad \qquad Z_{R}(s) = \frac{Z_{R}(s)}{\bigcap_{R}(s)} \qquad \qquad Z_{R}(s)$$

$$= \frac{Z_{R}(s)}{\bigcap_{R}(s)} \qquad \qquad Z_{R}(s) = \frac{Z_{R}(s)}{\bigcap_{R}(s)} \qquad \qquad Z_{R}(s)$$

$$= \frac{Z_{R}(s)}{\bigcap_{R}(s)} \qquad \qquad Z_{R}(s) = \frac{Z_{R}(s)}{\bigcap_{R}(s)} \qquad \qquad Z_{R}(s)$$

Capacitor in s-domain.

time-domain
$$\frac{+ \bigvee_{c}(t)}{\exists c(t)} = C \frac{d \bigvee_{c}(t)}{dt}$$

$$L [\exists c(t)] = C L [\frac{d}{dt} \bigvee_{c}] = C \cdot (S \cdot L [\bigvee_{c}(t)] - \bigvee_{c}(0 -))$$

$$\exists c(s) = (C \cdot S) \cdot \bigvee_{c}(s) - (C \cdot \bigvee_{c}(0 -))$$

$$z_{c}(s) = \frac{1}{S \cdot C}$$

$$z_{c}(s) = \frac{1}{S \cdot C}$$

$$\exists c(s) = \frac{1}{S \cdot C} - C \cdot \bigvee_{c}(0 -)$$
s-domain
$$\exists c(s) = \frac{1}{S \cdot C} - C \cdot \bigvee_{c}(0 -)$$

$$z_{c}(s) = \frac{1}{S \cdot C} - Impedance in s-domain$$

Inductor in s-domain.

time-domain
$$\frac{+ \bigvee_{L(t)} - \bigvee_{L$$

s-domain
$$Z_{L}(s)=s \cdot L \qquad L \cdot I_{L}(0-) \qquad V_{L}(s) = Z_{L}(s) \cdot \widetilde{I}_{L}(s) - L \cdot I_{L}(0-) \\
\widetilde{I}_{L}(s) \qquad \widetilde{V}_{L}(s) \qquad Z_{L}(s)=s \cdot L \qquad Impedance in s-domain$$

Example.

Example - cont.

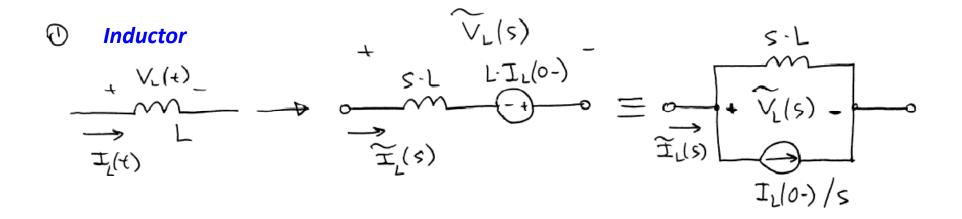
$$\widetilde{I}(s) = 3\left(\frac{1}{s+1} - \frac{1}{s+2}\right) ; \quad L\left[\bar{e}^{x+1}\right] = \frac{1}{s+x}$$

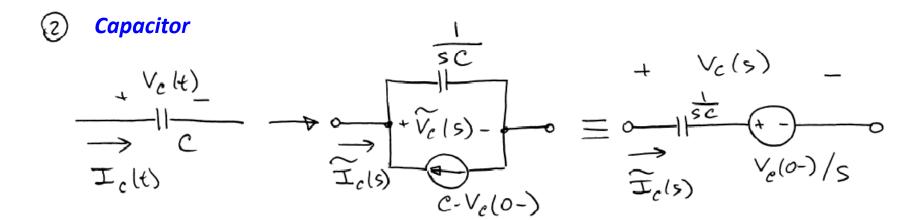
$$I(t) = 3\left(\bar{e}^{-t} - \bar{e}^{-2t}\right), A$$

$$V_c(t) = \frac{1}{c} \int_{0}^{t} I(\tau) d\tau = 2 \cdot 3 \cdot \left(\int_{0}^{t} \bar{e}^{-t} d\tau - \int_{0}^{t} \bar{e}^{-2t} d\tau\right) = 6 \cdot \left(\bar{e}^{-t} \Big|_{t}^{t} - \frac{1}{2} \bar{e}^{-2t}\Big|_{t}^{0}\right) = 6 \cdot \left(1 - \bar{e}^{-t} - \frac{1}{2}\left(1 - \bar{e}^{-2t}\right)\right) = 6 \cdot \left(\frac{1}{2} - \bar{e}^{-t} + \frac{1}{2} \bar{e}^{-2t}\right) = 3 + 3 \cdot \bar{e}^{-2t} - 6 \cdot \bar{e}^{-t}, V$$

Example - cont.

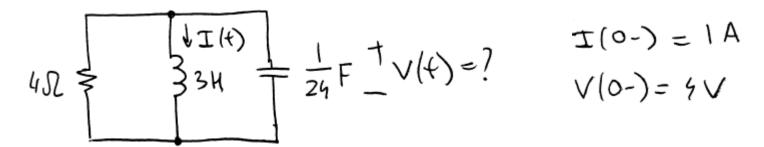
Thevenin and Norton forms for inductor and capacitor.



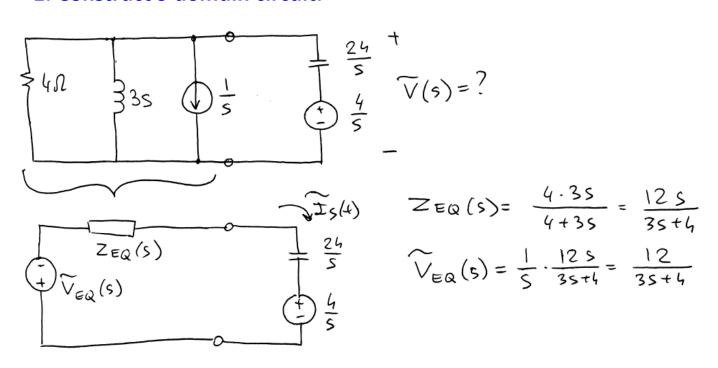


Initial conditions are treated like independent sources in s-domain circuits

Example 1.



1. Construct s-domain circuit.



Example 1 - cont.

2. Solve problem in s-domain.

$$\frac{125}{35+4} \longrightarrow \widehat{T}(5)$$

$$\frac{1}{7} \longrightarrow \frac{12}{5}$$

$$\frac{1}{7} \longrightarrow \frac{1}{7} \longrightarrow \frac{12}{5}$$

$$\frac{1}{7} \longrightarrow \frac{12}{5}$$

$$\frac{1}{$$

Example 1 - cont.

3. Partial fraction expansion.

$$\widetilde{V}(s) = -\frac{4}{5} \cdot \frac{12s + 8 - s^2 - 6s - 8}{s^2 + 6s + 8} = \frac{4}{5} \cdot \frac{s^2 - 6s}{s^2 + 6s + 8} = \frac{4(s - 6)}{s^2 + 6s + 8}$$

Find poles:
$$5^2 + 65 + 8 = 0$$

$$s = -3 \pm \sqrt{9-8} = -3 \pm 1 = \left[2 - 2j - 4\right]$$

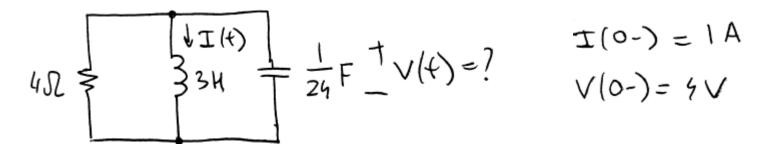
$$\frac{4(s-6)}{(s+2)(s+4)} = \frac{A}{s+2} + \frac{B}{s+4} = \widetilde{V}(s)$$

$$A = V(s)(s+2)|_{s=-2} = \frac{4(-8)}{-2+4} = -16$$

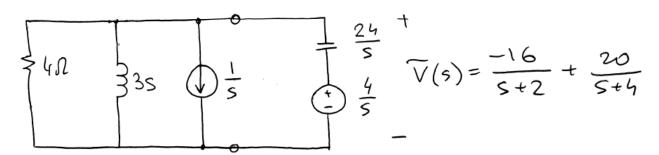
$$B = V(s)(s+4)|_{s=-4} = \frac{4(-10)}{-4+2} = +20$$

$$\widetilde{V}(s) = \frac{-16}{5+2} + \frac{20}{5+4}$$

Example 1 - cont.



Solution in s-domain:

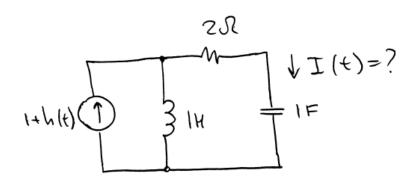


Solution in time-domain:
$$V(t) = -16e^{-2t} + 20e^{-4t}$$

$$t \rightarrow \infty \Rightarrow \lor \longrightarrow O$$
 Circuit contains resistive element and

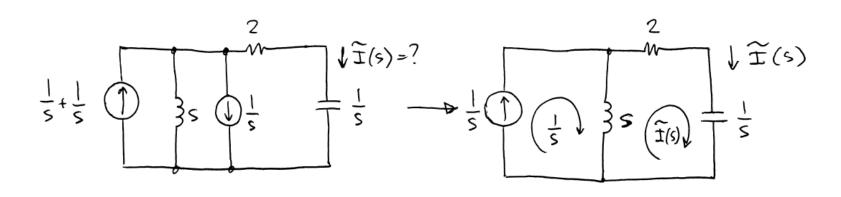
Circuit contains resistive element and no independent sources.

Example 2.



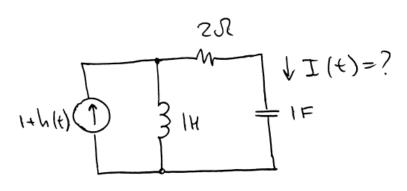
$$\begin{array}{c|c}
2D \\
\hline
1A & JI(0-)=0 \\
\hline
3IH & TIF & V_e(0-)=0 \\
\hline
J_L(0-)=1A
\end{array}$$

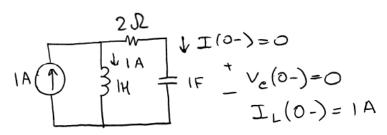
2. Construct s-domain circuit: t>0



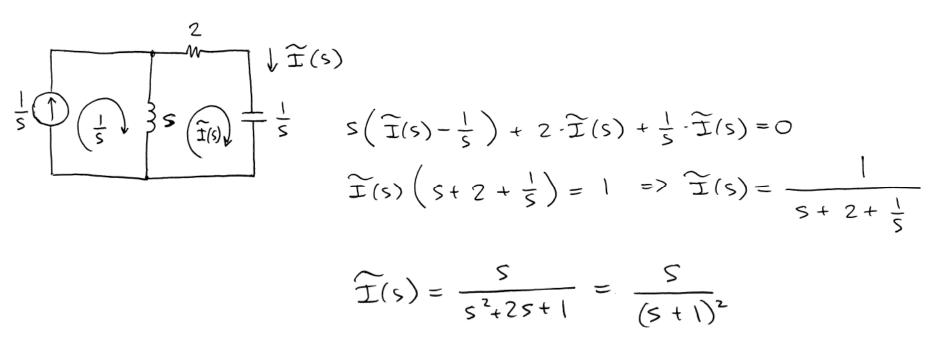
Use mesh analysis – KVL:
$$s\left(\widetilde{\pm}(s) - \frac{1}{s}\right) + 2\widetilde{\pm}(s) + \widetilde{\pm}(s) = 0$$

Example 2 - cont.

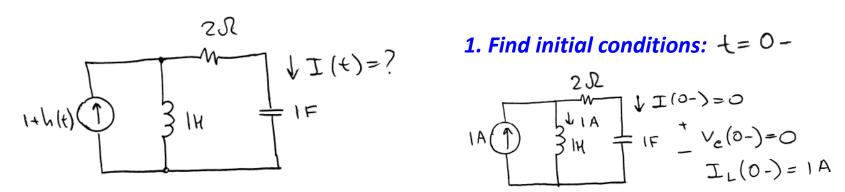


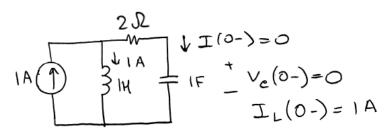


3. Solve in s-domain:



Example 2 - cont.





4. Convert to time-domain:

$$\widehat{T}(s) = \frac{s}{(s+1)^2} = \frac{A}{(s+1)^2} + \frac{B}{s+1}$$

$$\begin{cases}
A = (I(s) \cdot (s+1)^2) |_{s=-1} = -1 \\
B = \frac{d}{ds}(I(s)(s+1)^2) |_{s=-1} = 1
\end{cases}$$

$$\widehat{T}(s) = -\frac{1}{(s+1)^2} + \frac{1}{s+1}$$

$$\begin{cases} B = \frac{d}{ds}(\pm (s)(s+i)) \end{cases}$$

$$\widetilde{\pm}(s) = -\frac{1}{(s+1)^2} + \frac{1}{s+1}$$

Transfer function.

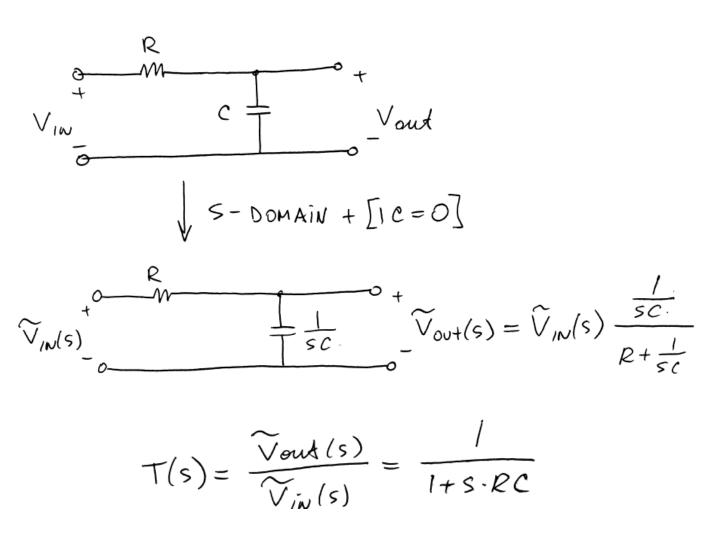
We already considered:
$$\frac{\widetilde{J}(s)}{Z(s)} + \frac{\widetilde{V}(s)}{Z(s)} = \frac{\widetilde{V}(s)}{\widetilde{J}(s)} \Big|_{Z \in PO} + \frac{\operatorname{impedance in}}{\operatorname{s-domain.}}$$

Now let's consider somewhat generalized case:

Transfer function:
$$T(s) = \frac{\widetilde{V}_{out}(s)}{\widetilde{V}_{in}(s)}$$
 | ZERO 1C

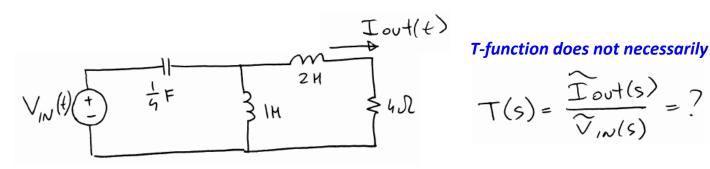
$$\widetilde{V}_{out}(s) = T(s) \cdot \widetilde{V}_{iw}(s)$$

Transfer function of RC integrator.



T-function has single negative real pole: $S = -\frac{1}{RC}$

Example.



T-function does not necessarily means ratio of voltages.

$$T(s) = \frac{\widehat{T}out(s)}{\widehat{V}_{iN}(s)} = ?$$

$$\widehat{I}_{out}(s) = \widehat{V}_{1}(s) / (2s+4) = \widehat{V}_{1N}(s) \frac{s^{2}/2}{s^{3}+2s^{2}+6s+8}$$

$$T(s) = \frac{\widehat{I}_{out}(s)}{\widehat{V}_{1N}(s)} = \frac{s^{2}}{2(s^{3}+2s^{2}+6s+8)}$$

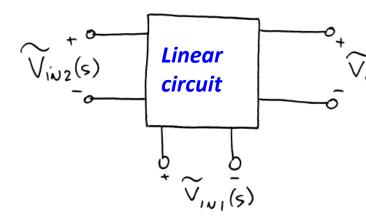
T-function has two "zeros" s = 0 and three "poles".

Circuits with multiple inputs.

For single input circuit:

$$\widetilde{V}_{out}(s) = T(s) \cdot \widetilde{V}_{i\omega}(s)$$
 $\widetilde{V}_{i\omega}(s) = 0 \implies \widetilde{V}_{out}(s) = 0$

But if circuit has independent sources inside then: $(\nabla_{0} + (s)) \neq 0$



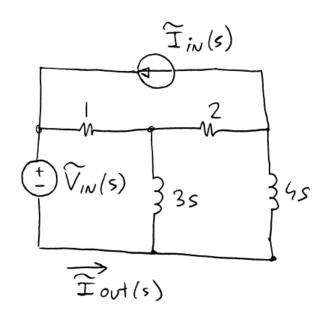
Each source is input.

T-function can be found for each input independently with other inputs "killed".

$$\widetilde{V}_{out}(s) = T_1(s) \cdot \widetilde{V}_{iN_1}(s) + T_2(s) \cdot \widehat{V}_{iN_2}(s)$$

$$V_{iN_1} = 0$$

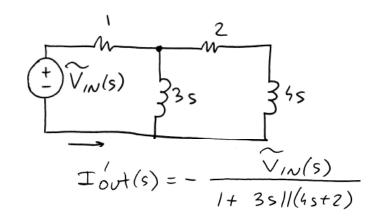
^{*}Always use zero initial conditions to compute T-function.



Example.

$$T_{out}(s) = T_1(s) \cdot \widetilde{V}_{IN}(s) + T_2(s) \cdot \widetilde{T}_{IN}(s)$$

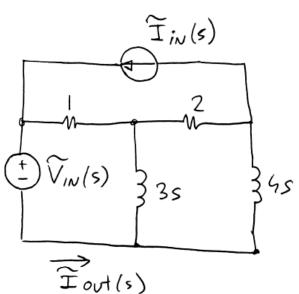
$$T_1(s) = ? \Rightarrow \widetilde{T}_{IN}(s) \Rightarrow 0$$



$$T_{1}(s) = \frac{Tout(s)}{V_{1N}(s)} = -\frac{1}{1+\frac{3s(4s+2)}{7s+2}} = \frac{-(7s+2)}{7s+2+12s^{2}+6s}$$

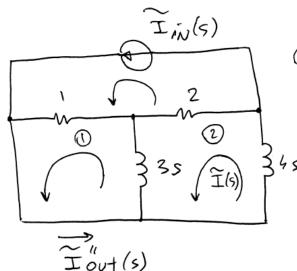
$$T_{1}(s) = \frac{-(7s+2)}{12s^{2}+13s+2}$$





$$\widehat{T}_{out}(s) = T_1(s) \cdot \widehat{V}_{IN}(s) + T_2(s) \cdot \widehat{T}_{IN}(s)$$

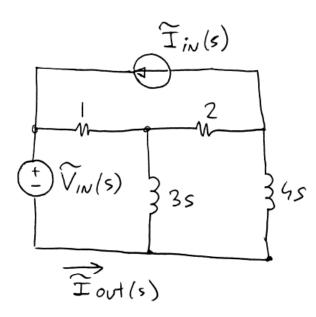
$$\widehat{T}_{out}(s) = \widehat{T}_1(s) \cdot \widehat{V}_{IN}(s) \rightarrow 0$$



$$(1) (\widetilde{J}_{out}^{"}(s) - \widetilde{J}_{"}(s)) - 1 + (\widetilde{J}_{out}^{"}(s) - \widetilde{J}(s)) - 3s = 0$$

$$\begin{cases} 34s \\ \widehat{I}(s) \cdot 4s + \widehat{I}(s) - \widehat{I}i\nu(s) - 2 + \\ + \widehat{I}(s) - \widehat{I}\nu(s) - 3s = 0 \end{cases}$$

$$T_2(s) = \frac{13s+2}{12s^2+13s+2}$$



Example - cont.

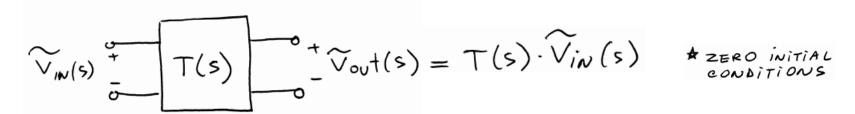
$$\widetilde{T}_{out}(s) = T_1(s) \cdot \widetilde{V}_{IN}(s) + T_2(s) \cdot \widetilde{T}_{IN}(s)$$

$$\widetilde{T}_{out}(s) = -\frac{7s+2}{12s^2+13s+2} \cdot \widetilde{V}_{IN}(s) + \frac{13s+2}{12s^2+13s+2} \cdot \widetilde{T}_{N}(s)$$

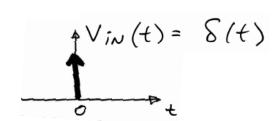
$$\widetilde{\pm}_{N}(s) = \frac{1}{s} Q \widetilde{V}_{N}(s) = \frac{1}{s}$$

$$\widehat{T}_{ov}+(s) = \frac{6}{12s^2+13s+2} = \frac{1}{2} \cdot \frac{1}{s^2+\frac{13}{12}s+\frac{1}{6}}$$

Impulse response



Now assume that



$$L[S(t)] = 1 = \widehat{V}_{in}(s)$$

$$\Rightarrow$$
 $\widetilde{V}_{\text{out}}(s) = T(s) \cdot I = T(s)$

$$L^{-1}[V_{out}(s)] = V_{out}(t) = L^{-1}[T(s)] = imp(t)$$

Impulse response

Impulse response of RC-integrator

$$V_{IN}(t) = \frac{P}{V_{IN}(t)} \cdot \frac{P}{V_{IN}(s)} = \frac{P}{1 + s \cdot PC}$$

$$IMP(t) = \frac{1}{T(s)} = \frac{1}{T(s)} = \frac{1}{1 + s \cdot PC} = \frac{1}{PC} \cdot \frac{1}{PC} \cdot \frac{1}{PC}$$

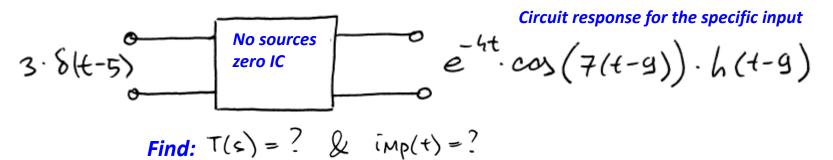
$$IMP(t) = \frac{h(t)}{PC} \cdot exp(-\frac{t}{PC})$$

$$V_{IN}(t) = \frac{h(t)}{PC} \cdot exp(-\frac{t}{PC})$$

Impulse response of RC-differentiator

Again
$$V_{in}(t) = S(t)$$
 $V_{in}(t) = V_{out}(t) = imp(t)$
 $V_{in}(t) = S(t)$
 $V_{in}(t) = S(t)$

Example.



①
$$L[3.8(t-5)] = 3.L[8(t-5)] = 3.e^{-55}$$

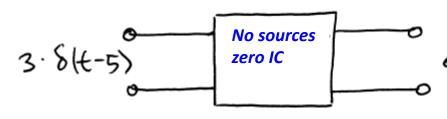
 $L[v(t-to).h(t-to)] = e^{-5.to} \hat{v}(s)$

2
$$L [e^{-4t} \cdot cos(7(t-9)) - h(t-9)] = L [e^{-4(t-9+9)} \cdot cos(7 \cdot (t-9)) - h(t-9)] =$$

 $= e^{-4\cdot 9} \cdot L [e^{-4(t-9)} \cdot cos(7 \cdot (t-9)) - h(t-9)] =$
 $= e^{-36} \cdot e^{-9\cdot 5} \cdot L [e^{-4\cdot t} \cdot cos(7t)] = e^{-36} \cdot e^{-95} \cdot \frac{s+4}{(s+4)^2+7^2}$

Damped cosine – use table

Example - cont.



Circuit response for the specific input

Find:
$$T(s) = ?$$
 & $imp(t) = ?$

$$T(s) = \frac{\widehat{V}_{out}(s)}{\widehat{V}_{in}(s)} = \frac{e^{-36} \cdot \bar{e}^{35} \cdot (s+4)}{(s+4)^2 + 49} / 3 \cdot \bar{e}^{-5s}$$

$$T(s) = \frac{1}{3} \cdot e^{-36} \cdot e^{-4s} \cdot \frac{s+4}{(s+4)^2 + 49}$$

$$T(s) = \frac{1}{3} \cdot e^{-36} \cdot e^{-4s} \cdot \frac{s+4}{(s+4)^2 + 49}$$
Damped cosine time shifted by 4 seconds
$$L\left[e^{-\lambda t}\cos\omega t\right] = \frac{s+\lambda}{(s+\lambda)^2 + \omega^2}$$

$$L\left[v(t+t_0)h(t+t_0)\right] = e^{-s+0} \cdot \widetilde{V}(s)$$

$$imp(t) = L^{-1} [T(s)] = \frac{e^{-36}}{3} \cdot e^{-4(t-4)} \cdot cos(7(t-4)) \cdot h(t-4) =$$

=
$$\frac{1}{3} \cdot e^{-4(t+5)} \cdot cos(7(t-4)) \cdot h(t-4)$$

Observe: IMP (t) = O FOR t < 4