

Problem 1

$$(1a) \quad V_{p,min} = V_{IN} - V_{Th} = V_{GSO} + V_{GS1} - V_{Th} \\ = (V_{GSO} - V_{Th}) + (V_{GS1} - V_{Th}) + V_{Th}.$$

That is overdrive voltage of M_0 and M_1 , plus additional one threshold voltage.

(1b) This is a cascode amplifier,

$$R_{out} = (1 + g_{m3} r_{o3}) r_{o2} + r_{o3} \\ \approx g_{m3} r_{o3} r_{o2}$$

(1c) For M_1 ,

$$I_{REF} = k_1 V_{OD,1}^2 \quad \text{where } k = \frac{1}{2} \mu_n C_{ox} \left(\frac{W}{L} \right)$$

For M_2

$$I_{out} = k_2 V_{OD,2}^2 \\ = k_2 V_{OD,1}^2$$

$$\text{Note } \begin{cases} V_{GS1} = V_{GS2} \\ V_{Th1} = V_{Th2} \end{cases} \quad \therefore V_{OD,1} = V_{OD,2}$$

$$\therefore \frac{k_2}{k_1} = \frac{(W/L)_2}{(W/L)_1} = 10., \text{ the ratio between } M_{0,3} \\ \text{is the same as } M_{1,2}, \text{ so}$$

$$\frac{(W/L)_3}{(W/L)_0} = 10.$$

$$(1d) \quad V_{p,min} = V_{DD0} + V_{OD1} + V_{Th} = 0.5V$$

$$V_{DD0} + V_{OD1} = 0.2V$$

$$\therefore V_{OD} = 0.1V$$

For M_2 and M_3

$$I_D = \frac{1}{2} \mu_n C_{ox} \left(\frac{W}{L} \right) V_{OD}^2 \Rightarrow \left(\frac{W}{L} \right)_{2,3} = 400$$

$$So \left(\frac{W}{L} \right)_{0,1} = \frac{400}{10} = 40.$$

$$(1e) \quad V_N = V_{GS1} + V_{GS0}$$

$$I_{D0,1} = \frac{1}{2} \mu_n C_{ox} \left(\frac{W}{L} \right)_{0,1} V_{OD0,1}^2 \Rightarrow$$

$$V_{OD0,1} = \sqrt{\frac{2 I_{D0,1}}{\mu_n C_{ox} \left(\frac{W}{L} \right)_{0,1}}} = \sqrt{\frac{2 \times 0.1 \times 10^{-3}}{5 \times 10^{-4} \times 40}} = 0.1V$$

$$\therefore V_{GS0,1} = V_{OD0,1} + V_{Th} = 0.4V$$

$$\therefore V_N = 0.8V$$

$$\therefore R = \frac{V_{DD} - V_N}{I_{D0,1}} = \frac{1.5 - 0.8}{0.1 \times 10^{-3}} = 7k\Omega$$

Problem 2

(2a) $M_{3,4}$ are current source load of the common source half circuit.

M_7 is used to bias V_b .

(2b)

$$\frac{(W/L)_5}{(W/L)_8} = \frac{I_{SS}}{I_{ref}} = 5 \Rightarrow (W/L)_5 = 5 \times 50 = 250.$$

Now, the current flowing through M_6 is given by

$$(W/L)_6 = I_{D6}/I_{ref} \cdot (W/L)_8$$

$$\text{Let } I_{D6} = 2.5 I_{ref} \Rightarrow (W/L)_6 = 2.5 \times 50 = 125$$

$$\text{As } (W/L)_7 = 50 \text{ and } I_{D3} = I_{D4} = \frac{1}{2} I_{SS} = 2.5 I_{ref} = I_{D6} = I_{D7}$$

$$\text{So } (W/L)_3 = (W/L)_4 = (W/L)_7 = 50$$

$$(2c) \quad I_{ref} = 125 \mu A \Rightarrow I_1 = I_2 = 2.5 \times 125 \mu A = 312.5 \mu A$$

$$I_{SS} = 125 \mu A \times 5 = 625 \mu A$$

For M_5 :

$$I_{SS} = \frac{1}{2} \mu_n C_{ox} (W/L)_5 V_{OD5}^2$$

$$\Rightarrow V_{OD5} = \sqrt{\frac{2 I_{SS}}{\mu_n C_{ox} (W/L)_5}}$$

$$= \sqrt{\frac{2 \times 625 \times 10^{-6}}{5 \times 10^{-4} \times 250}}$$

$$= 0.1 V$$

$$V_{out \min} = V_{OD5} + V_{GS1} - V_{Th}$$

$$= V_{GS1} - 0.2 V \quad (W/L)_{1,2} \text{ is not given}$$

Another limit posted by M_3 in saturation

$$V_{out,max} = V_{DD} - V_{D3}$$

$$V_{D3} = \sqrt{\frac{2I_1}{\mu_p C_{ox} \left(\frac{W}{L}\right)_3}} = \sqrt{\frac{I_{SS}}{2.5 \times 10^{-4} \times 50}} = 0.22V$$

So the output range is $V_{GS1} - 0.2 \sim 1.28V$

$$(2d) \quad V_{in,CM} \geq 0.1V + V_{GS1}$$

$$\begin{aligned} V_{in,CM} &\leq V_{out} - V_{Th} \\ &\leq 0.98V \end{aligned}$$

$$\therefore, 0.1V + V_{GS1} \leq V_{in,CM} \leq 0.98V$$

(2e) The output voltage is not well defined. It will depend on the specific value for the common mode input voltage

Problem 3.

$$(3a) \quad |A_v| = g_{m1,2} (r_{D2} || r_{D4})$$

At DC (or common mode), $V_{in1} = V_{in2}$.

$$\text{So } I_{D1} = I_{D2} = \frac{1}{2} I_D = 0.5mA$$

$$r_{D2} = \frac{1}{\lambda_n I_{D2}} = \frac{1}{0.1 \times 0.5 \times 10^{-3}} = 20k\Omega$$

$$r_{o4} = \frac{1}{\lambda_p I_{D4}} = \frac{1}{0.2 \times 0.5 \times 10^{-3}} = 10 \text{ k}\Omega$$

$$|A_v| = g_{m_{1,2}} (r_{o2} \parallel r_{o4}) = 20.$$

$$g_{m_{1,2}} = \frac{20}{20/3 \text{ k}\Omega} = 3 \times 10^{-3} \text{ V}^{-1}$$

$$= \sqrt{2 \mu_n C_{ox} \left(\frac{W}{L}\right)_{1,2} I_{D1}} \quad g_m = \frac{2I_D}{V_{OD}}$$

$$\begin{aligned} \left(\frac{W}{L}\right)_{1,2} &= \frac{g_{m_{1,2}}^2}{2 \mu_n C_{ox} I_{D1}} \\ &= \frac{9 \times 10^{-6}}{2 \times 5 \times 10^{-4} \times 0.5 \times 10^{-3}} = 18. \end{aligned}$$

(3b)

$$\begin{aligned} \left(\frac{W}{L}\right)_{3,4} &= \frac{2 I_{D3,4}}{\mu_p C_{ox} V_{OD}^3} \\ &= \frac{1 \times 10^{-3}}{2.5 \times 10^{-4} \times 0.2^3} \\ &= 100 \end{aligned}$$

$$\begin{aligned} V_{OD1,2} &= \frac{2I_{D1,2}}{g_{m_{1,2}}} = \frac{2 \times 10^{-3}}{3 \times 10^{-3}} = 0.67 \text{ V} \\ V_{OD3,4} &= 0.2 \text{ V} \\ g_{m_{3,4}} &= \frac{2I_{D3,4}}{V_{OD3,4}} = \frac{1 \times 10^{-3}}{0.2} \\ &= 5 \times 10^{-3} \text{ V}^{-1} \end{aligned}$$

$$(3c) \quad V_{out, \min} \geq V_{OD5} + V_{OD2}$$

$$V_{out, \max} \leq V_{DD} - V_{OD4} = 1.3 \text{ V}$$

(3d) The output DC bias is

$$V_{out} = V_{DD} - V_{SG4}$$

So the maximum swing is

$$V_{DD} - V_{OD4} - (V_{DD} - V_{SG4}) = V_{SG4} - V_{OD4} = |V_{thp}| = 0.3 \text{ V}$$

$$(3e) \quad r_{o5} = \frac{1}{\lambda_n I_D} = \frac{1}{0.1 \times 1 \times 10^{-3}} = 10 \text{ k}\Omega$$

$$r_{o1,2} = \frac{1}{\lambda_n I_{D1,2}} = \frac{1}{0.1 \times 0.5 \times 10^{-3}} = 20 \text{ k}\Omega$$

$$r_{o3,4} = \frac{1}{\lambda_p I_{D3,4}} = \frac{1}{0.2 \times 0.5 \times 10^{-3}} = 10 \text{ k}\Omega$$

$$CMRR = (1 + 2g_{m1,2} r_{o5}) g_{m3,4} (r_{o1,2} \parallel r_{o3,4})$$

$$\approx 2g_{m1,2} r_{o5} g_{m3,4} (r_{o1,2} \parallel r_{o3,4})$$

$$= \cancel{2 \times 3 \times 10^{-3}} \times \cancel{10 \times 10^{-3}} \times \cancel{5 \times 10^{-3}} \times \frac{200}{30} \times \cancel{10^3}$$

$$= 2000$$