Lecture 10: Pipelining and Sequential Design

Acknowledgements

All class materials (lectures, assignments, etc.) based on material prepared by Prof. Visvesh S. Sathe, and reproduced with his permission

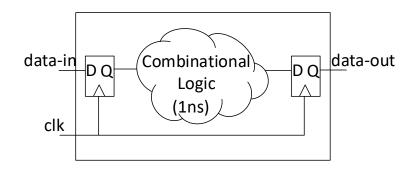


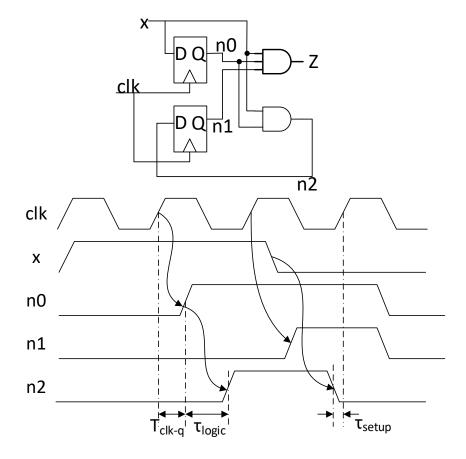
Visvesh S. Sathe
Associate Professor
Georgia Institute of Technology
https://psylab.ece.gatech.edu

UW (2013-2022) GaTech (2022-present)

Two Broad Categories

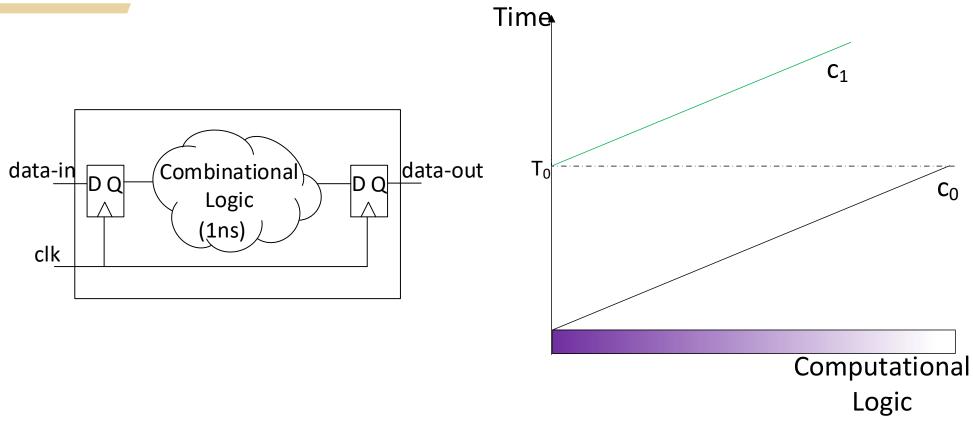
- Combinational logic
 - Output depends only in current inputs
- Sequential design
 - Output depends on a combination of existing and previous inputs
 - Consider a sequential system of 1 output M inputs. Arbitrary dependence on history of past N results leads to equivalent system of inputs.
 - In most practical applications, trace of previous inputs influences system "makeup" (or state), which is a stored quantity.
 - Output depends on current inputs and current state







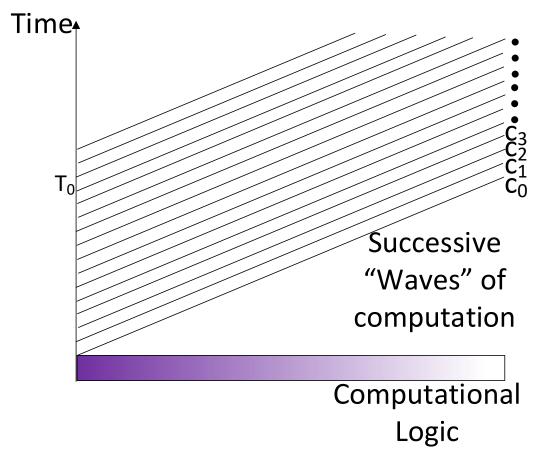
Pipelines In a Little More Detail....



- Synchronous designs operate at a rate determined by a clock signal
 - Clocks control flip-flops to capture and propagate data
- If we wait for computation completion before launching new data:
 - Duration (or latency) of logic evaluation sets processing rate (throughput)
 - Latency L → Throughput T=1/L



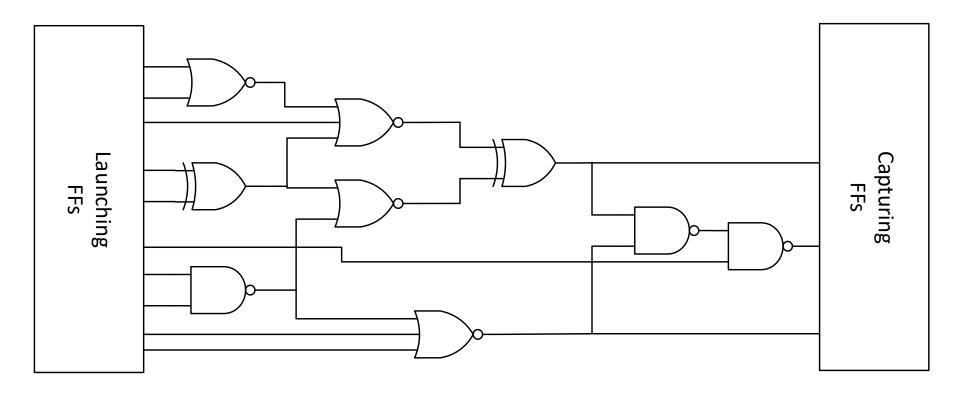
Why Wait?



- Don't wait for a computation to end before starting the next one
 - Process data in "waves" within logic
 - Use a flip-flop to capture and store data (avoid waves from colliding)
 - Wave pipelining



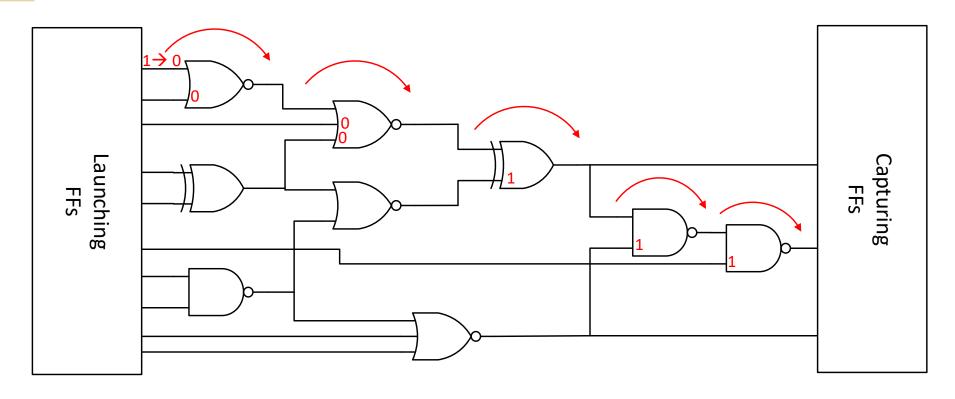
Quick Detour: Max and Min-delays



- Combinational logic evaluation latency occurs over a range of time
 - Latency (Duration) of the computation depends on input data selected
 - Not every input combination excites the critical path
 - Max-delay: Data inputs traverse critical path of the logic
 - Min-delay: Data inputs traverse shortest path (in a delay sense) of logic



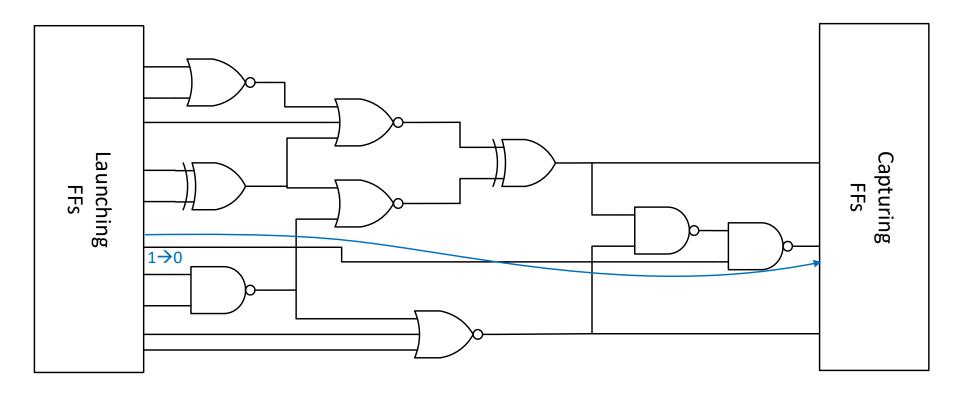
Quick Detour: Max and Min-delays



- Combinational logic evaluation latency occurs over a range of time
 - Latency (Duration) of the computation depends on input data selected
 - Not every input combination excites the critical path
 - Max-delay: Data inputs traverse critical path of the logic
 - Min-delay: Data inputs traverse shortest path (in a delay sense) of logic



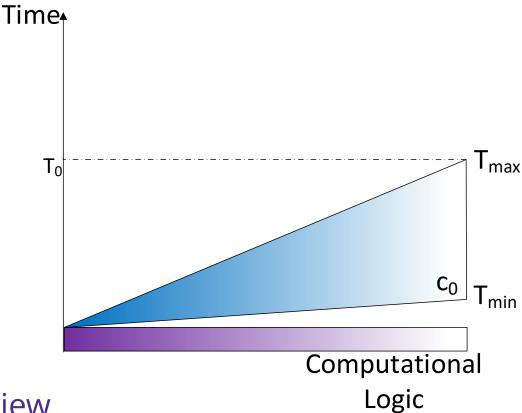
Quick Detour: Max and Min-delays



- Combinational logic evaluation latency occurs over a range of time
 - Latency (Duration) of the computation depends on input data selected
 - Not every input combination excites the critical path
 - Max-delay: Data inputs traverse critical path of the logic
 - Min-delay: Data inputs traverse shortest path (in a delay sense) of logic

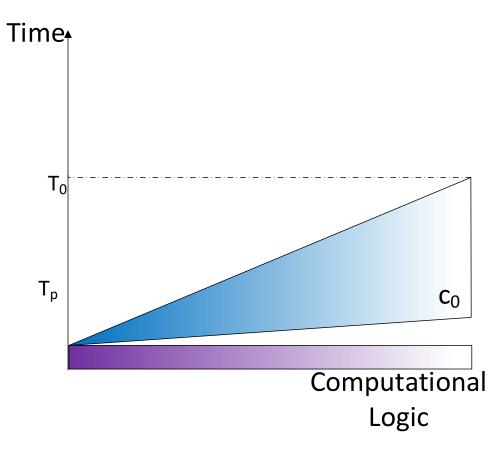


Delay Distribution in Logic



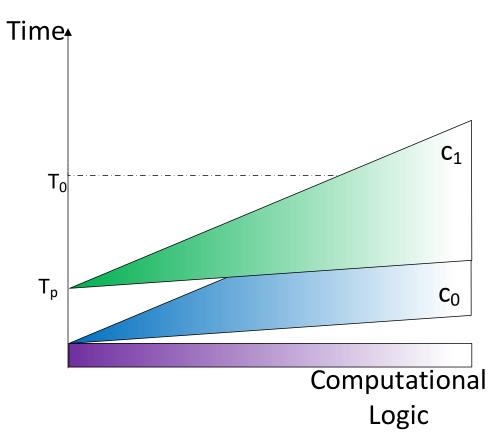
- A more realistic view
 - Logic progression through the combinational block occurs over a range of time
 - Latency depends on the nature of inputs computed
 - Typically, critical paths are excited with a low probability (but must still be designed for!)
 - Note: This figure (showing continuous progression) is still a simplification!





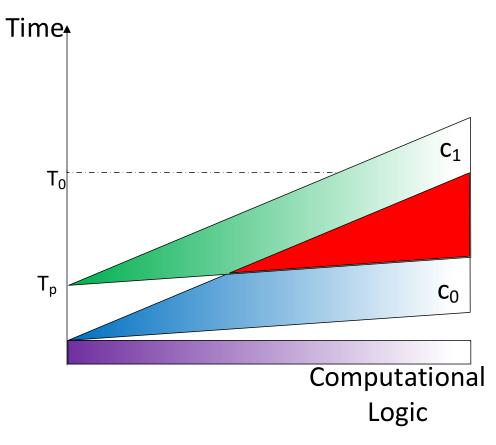
- Prematurely launching new wave before previous wave is "sufficiently far along" risks wave collision (and therefore failure)
- Max/Min delay spread for each output node must be considered





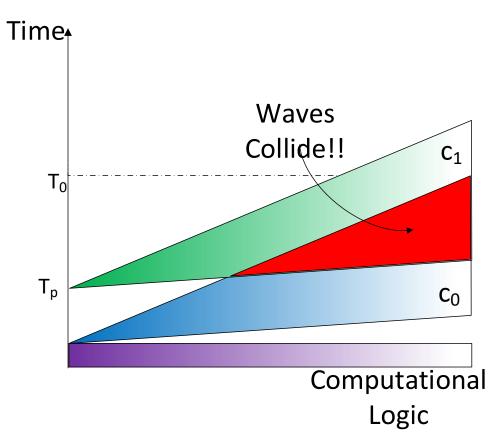
- Prematurely launching new wave before previous wave is "sufficiently far along" risks wave collision (and therefore failure)
- Max/Min delay spread for each output node must be considered





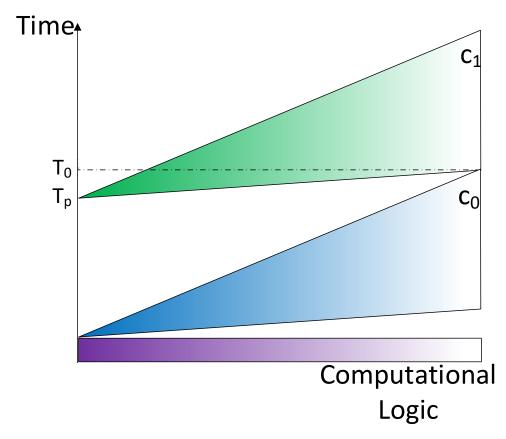
- Prematurely launching new wave before previous wave is "sufficiently far along" risks wave collision (and therefore failure)
- Max/Min delay spread for each output node must be considered





- Prematurely launching new wave before previous wave is "sufficiently far along" risks wave collision (and therefore failure)
- Max/Min delay spread for each output node must be considered

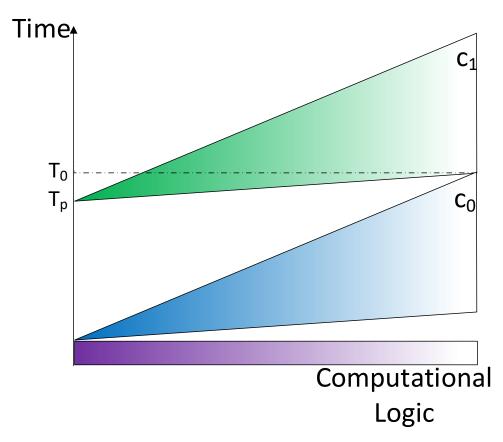




Need to wait for at least

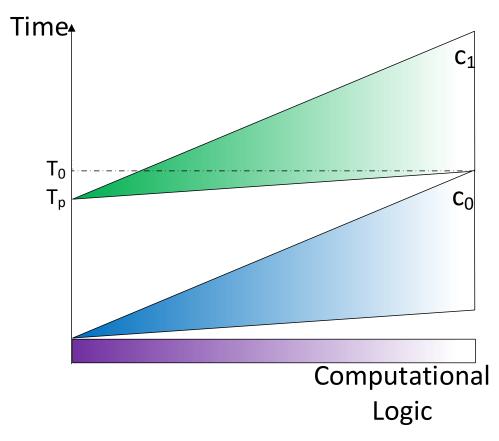
before launching new data





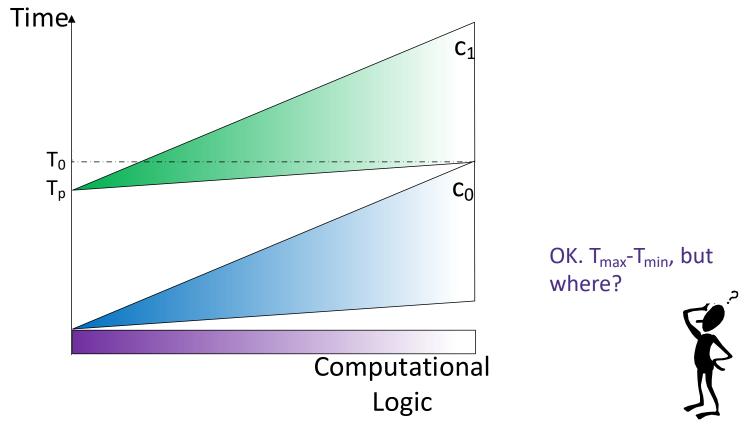
Need to wait for at least T_{max}-T_{min} before launching new data





- Need to wait for at least T_{max}-T_{min} before launching new data
- Max, min latency for any point sets the absolute highest achievable throughput

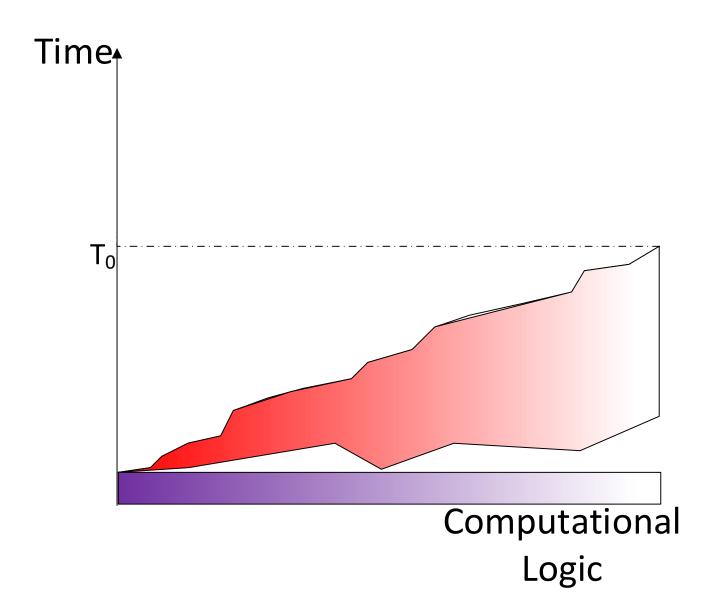




- Need to wait for at least T_{max}-T_{min} before launching new data
- Max, min latency for any point sets the absolute highest achievable throughput



Typical Max and Min Delay Profile





Question

- Given: A combinational logic with n inputs and 1 output. T_{max} range (1.5ns-2.5ns), T_{min} (0,0.5ns). What is the highest achievable throughput using this "wave-pipelining" approach.
 - 3.33 GHz
 - 0.67 GHz
 - -0.4 GHz
 - 0.5 GHz

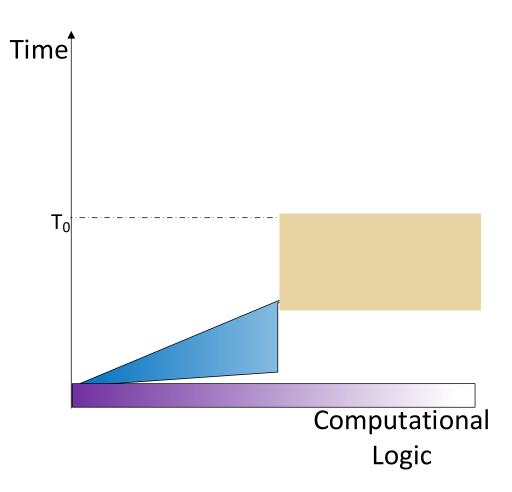


Question

- Given: A combinational logic with n inputs and 1 output. T_{max} range (1.5ns-2.5ns), T_{min} (0,0.5ns). What is the highest achievable throughput using this "wave-pipelining" approach.
 - 3.33 GHz
 - 0.67 GHz
 - 0.4 GHz
 - 0.5 GHz
- Can a general statement be made about computational latency, and the achievable speedup using this approach?

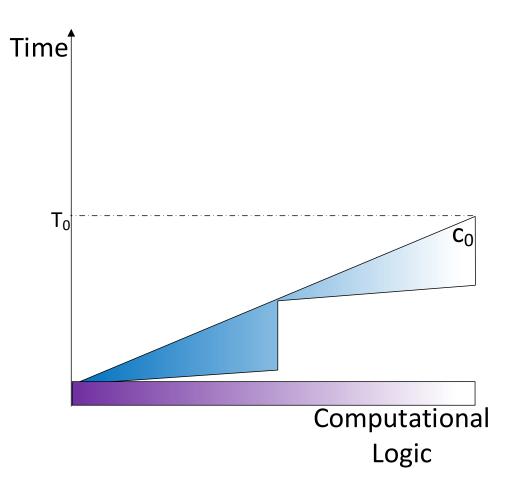






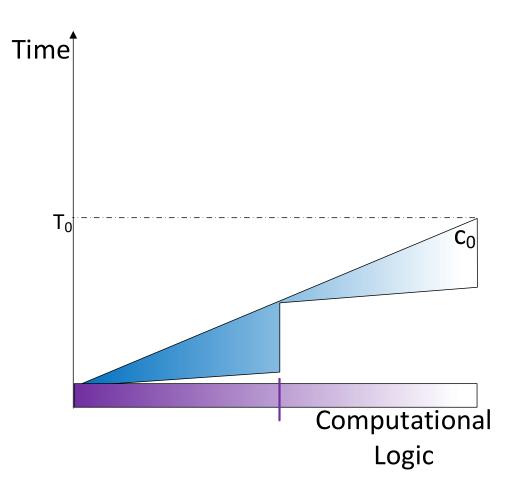
- FF-based pipelines (2-stage example)
 - Insert FFs at points in the logic corresponding to T/2 peak delay





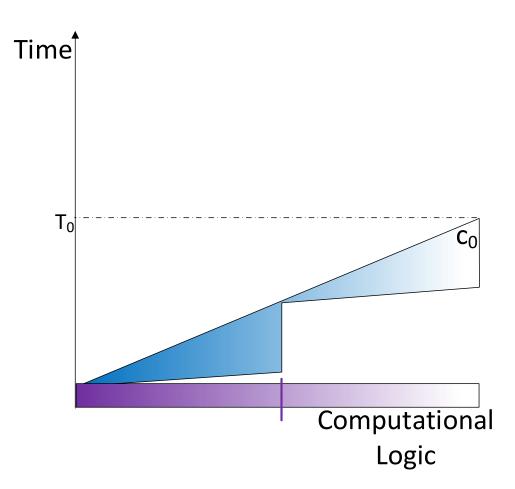
- FF-based pipelines (2-stage example)
 - Insert FFs at points in the logic corresponding to T/2 peak delay





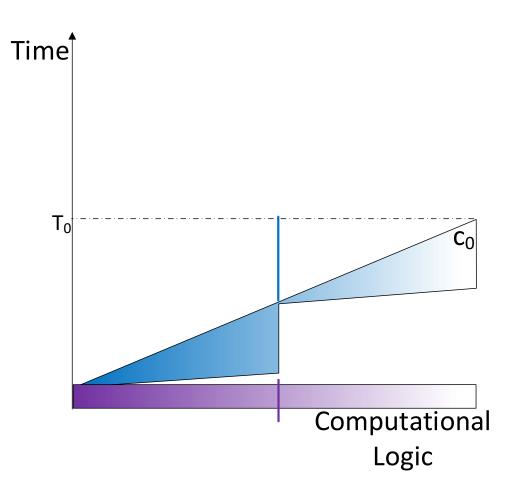
- FF-based pipelines (2-stage example)
 - Insert FFs at points in the logic corresponding to T/2 peak delay





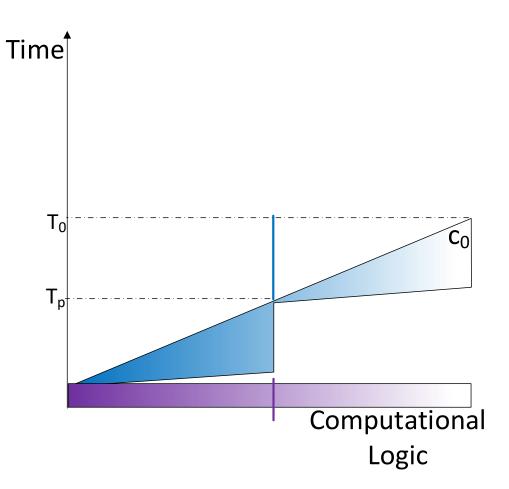
- FF-based pipelines (2-stage example)
 - Insert FFs at points in the logic corresponding to T/2 peak delay
 - FFs capture and store the values, avoiding successive fast-paths from colliding (or even interfering) with them (also referred to as a "race")





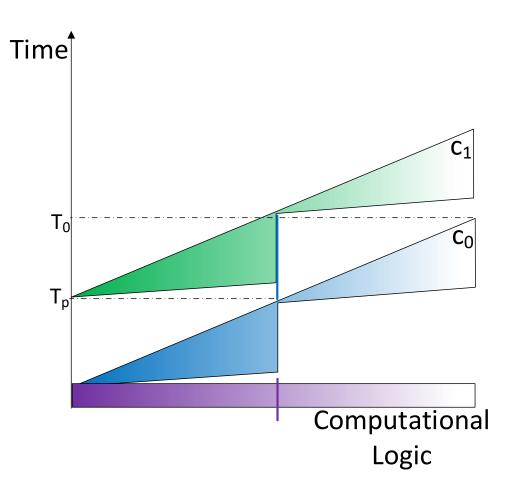
- FF-based pipelines (2-stage example)
 - Insert FFs at points in the logic corresponding to T/2 peak delay
 - FFs capture and store the values, avoiding successive fast-paths from colliding (or even interfering) with them (also referred to as a "race")





- FF-based pipelines (2-stage example)
 - Insert FFs at points in the logic corresponding to T/2 peak delay
 - FFs capture and store the values, avoiding successive fast-paths from colliding (or even interfering) with them (also referred to as a "race")

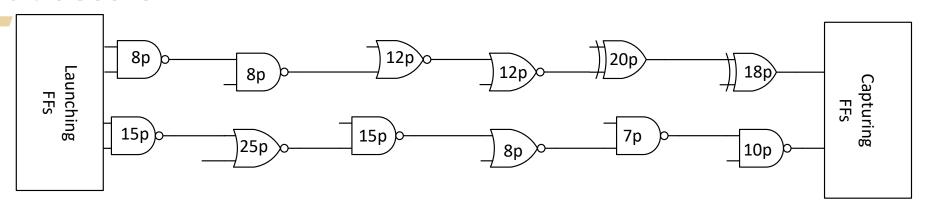




- FF-based pipelines (2-stage example)
 - Insert FFs at points in the logic corresponding to T/2 peak delay

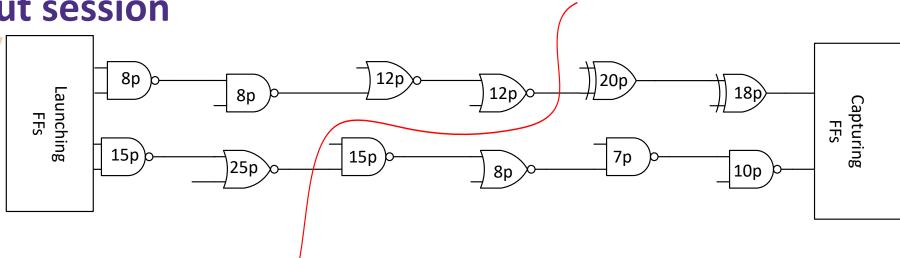
'en





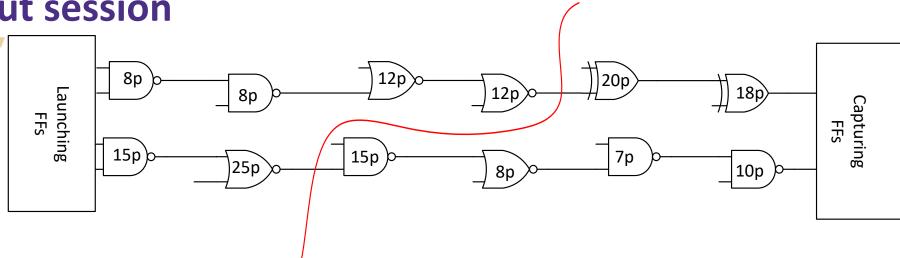
 Consider the following logic paths. Where should flip-flops be inserted to best improve performance for a 2-state design





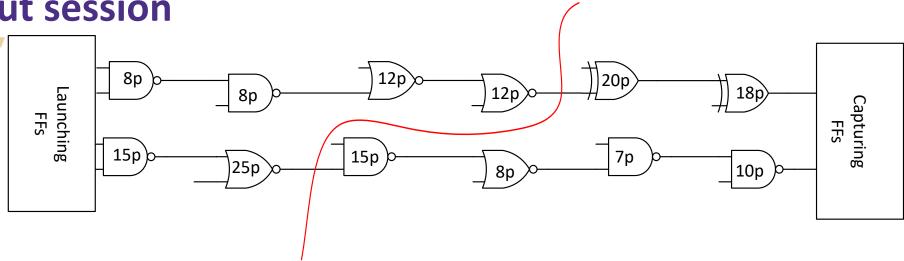
 Consider the following logic paths. Where should flip-flops be inserted to best improve performance for a 2-state design





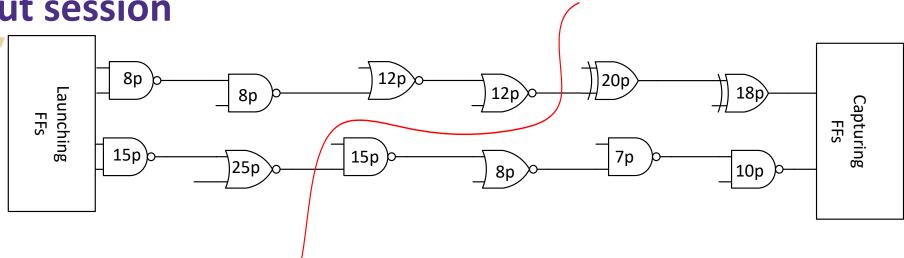
- Consider the following logic paths. Where should flip-flops be inserted to best improve performance for a 2-state design
 - FF insertion at nodes with max. delay of ~T/n
 - T = Total, maximum computational latency
 - n = Number of pipeline stages





- Consider the following logic paths. Where should flip-flops be inserted to best improve performance for a 2-state design
 - FF insertion at nodes with max. delay of ~T/n
 - T = Total, maximum computational latency
 - n = Number of pipeline stages
- What will be the peak performance for a 3-stage design?

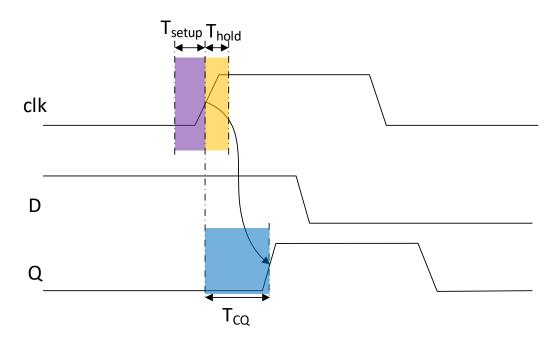




- Consider the following logic paths. Where should flip-flops be inserted to best improve performance for a 2-state design
 - FF insertion at nodes with max. delay of ~T/n
 - T = Total, maximum computational latency
 - n = Number of pipeline stages
- What will be the peak performance for a 3-stage design?
 - Delay-granularity is a consideration, especially for heavily pipelined designs



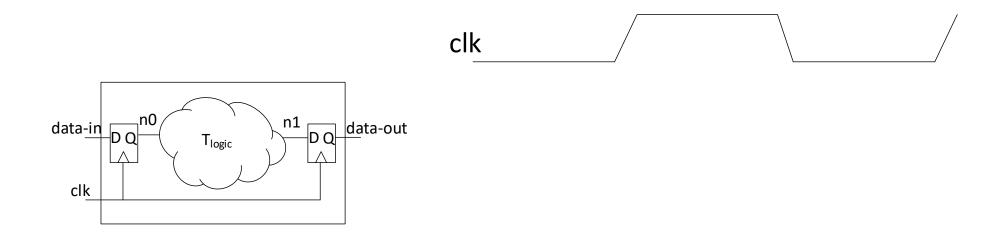
Flip-Flop Timing Properties



- FFs impose timing constraints on input data
 - Data to be latched must arrive no later than T_{setup} before clock edge
 - Data to be latched must be held stable for T_{hold} after clock edge
 - Early-arriving data from previous pipeline stage could degrade timing properties
 - Sampled input data does not get transmitted to the output instantly (T_{CQ})
- T_{setup} + T_{CQ} a good metric for FF timing overhead
- T_{CO} T_{hold} a good metric for race immunity (more on this later)



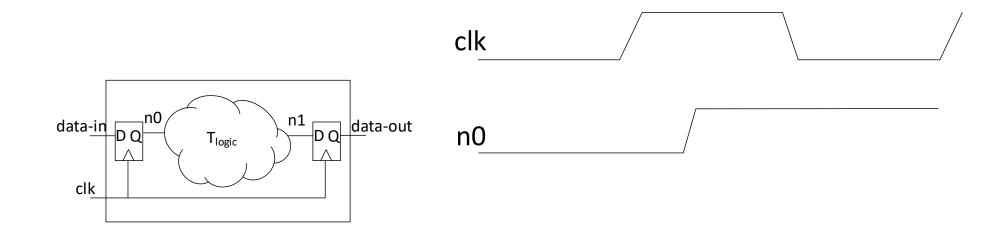
Pipelined Designs: Basic Timing Analysis



- Cycle-time (and throughput) depend on FF parameters and logic
 - $T_{cycle} \ge T_{cq} + T_{logic,max} + T_{setup}$
- A design can consist of multiple such synchronous blocks
 - Blocks synchronized by the same clock said to be in the same clock domain
 - Cycle time of the design is determined by the slowest block.



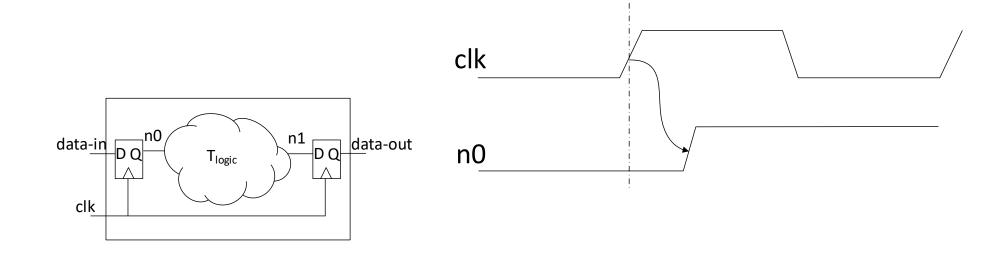
Pipelined Designs: Basic Timing Analysis



- Cycle-time (and throughput) depend on FF parameters and logic
 - $T_{cycle} \ge T_{cq} + T_{logic,max} + T_{setup}$
- A design can consist of multiple such synchronous blocks
 - Blocks synchronized by the same clock said to be in the same clock domain
 - Cycle time of the design is determined by the slowest block.

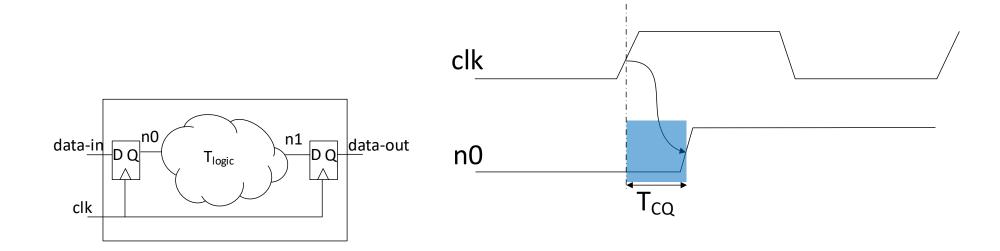


Pipelined Designs: Basic Timing Analysis



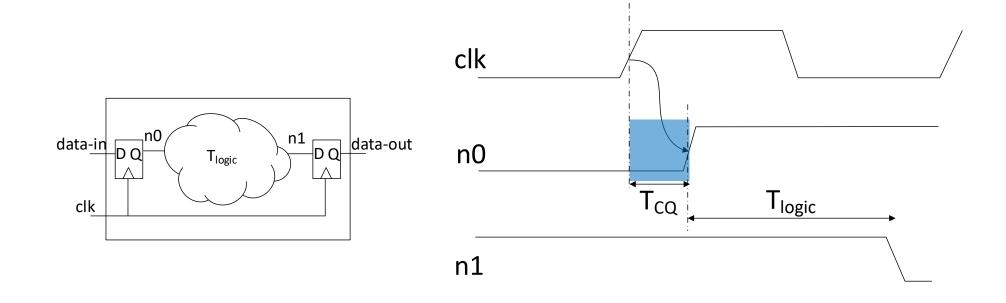
- Cycle-time (and throughput) depend on FF parameters and logic
 - $T_{cycle} \ge T_{cq} + T_{logic,max} + T_{setup}$
- A design can consist of multiple such synchronous blocks
 - Blocks synchronized by the same clock said to be in the same clock domain
 - Cycle time of the design is determined by the slowest block.





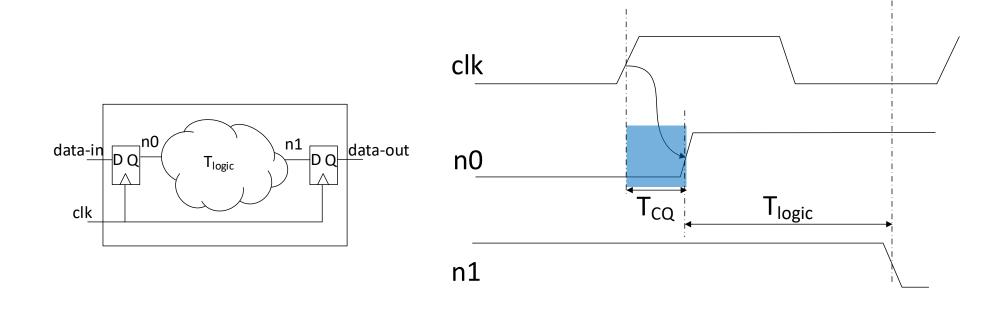
- Cycle-time (and throughput) depend on FF parameters and logic
 - $T_{cycle} \ge T_{cq} + T_{logic,max} + T_{setup}$
- A design can consist of multiple such synchronous blocks
 - Blocks synchronized by the same clock said to be in the same clock domain
 - Cycle time of the design is determined by the slowest block.





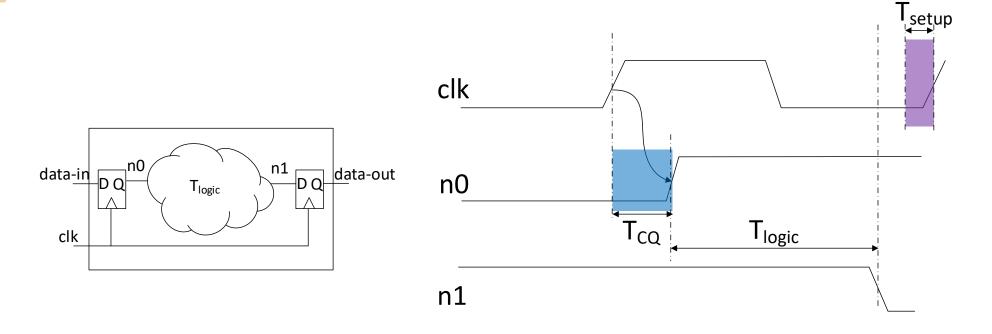
- Cycle-time (and throughput) depend on FF parameters and logic
 - $T_{cycle} \ge T_{cq} + T_{logic,max} + T_{setup}$
- A design can consist of multiple such synchronous blocks
 - Blocks synchronized by the same clock said to be in the same clock domain
 - Cycle time of the design is determined by the slowest block.





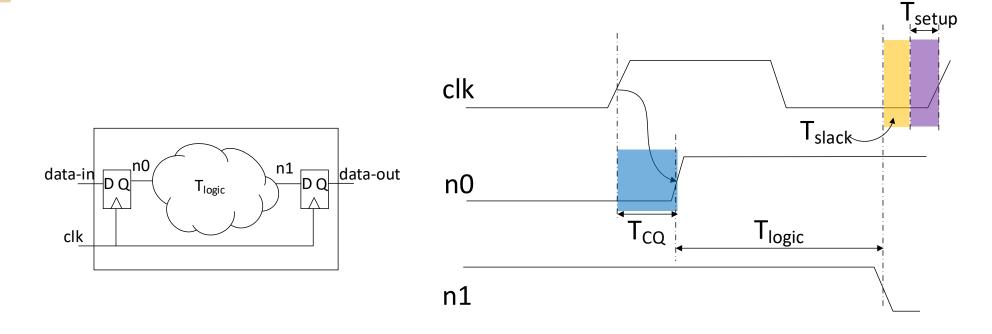
- Cycle-time (and throughput) depend on FF parameters and logic
 - $T_{cycle} \ge T_{cq} + T_{logic,max} + T_{setup}$
- A design can consist of multiple such synchronous blocks
 - Blocks synchronized by the same clock said to be in the same clock domain
 - Cycle time of the design is determined by the slowest block.





- Cycle-time (and throughput) depend on FF parameters and logic
 - $T_{cycle} \ge T_{cq} + T_{logic,max} + T_{setup}$
- A design can consist of multiple such synchronous blocks
 - Blocks synchronized by the same clock said to be in the same clock domain
 - Cycle time of the design is determined by the slowest block.





- Cycle-time (and throughput) depend on FF parameters and logic
 - $T_{cycle} \ge T_{cq} + T_{logic,max} + T_{setup}$
- A design can consist of multiple such synchronous blocks
 - Blocks synchronized by the same clock said to be in the same clock domain
 - Cycle time of the design is determined by the slowest block.



Performance

- For a combinational block, addition of N-pipeline stages impacts several parameters.
 - Total Latency (Ignoring area impact)
 - Absolute Time : $T_{logic} + N*(T_{setup} + T_{cq})$ (†)
 - N*T_{clk} in cycles
 - Throughput improvement (Ignoring area* impact)
 - Perf. Improvement = $\frac{T_{cycle,nom}}{T_{cycle,N-stage}} = \frac{T_{logic} + T_{CQ} + T_{setup}}{\frac{T_{logic}}{N} + T_{CQ} + T_{setup}}$
 - Assumes logic can be equally divided. Remember Longest pipe-stage sets cycle-time
 - Area
 - $A_N = A_{logic} + \sum_{i=1}^{N} A_i$
 - Power*
 - $P_N = P + \sum_{1}^{N} P_i$
- **Exercise**. Consider an unpipelined design with T_{logic} = 2ns. Sketch the plot for cycle time vs pipeline depth, N with $T_{setup} = T_{cq} = 200ps$

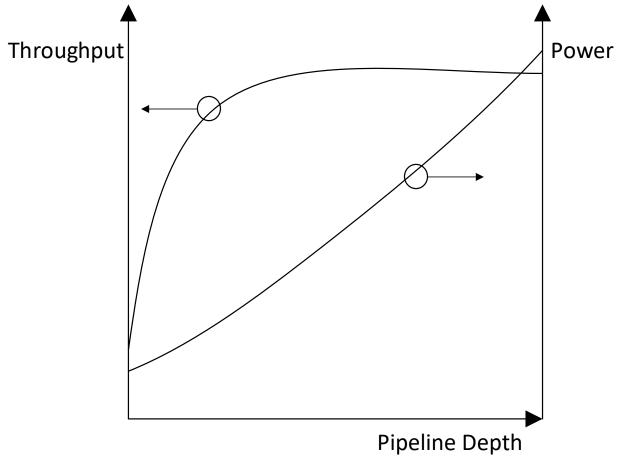


Performance

- For a combinational block, addition of N-pipeline stages impacts several parameters.
 - Total Latency (Ignoring area impact)
 - Absolute Time : $T_{logic} + N*(T_{setup} + T_{cq})$ (†)
 - N*T_{clk} in cycles
 - Throughput improvement (Ignoring area* impact)
 - Perf. Improvement = $\frac{T_{cycle,nom}}{T_{cycle,N-stage}} = \frac{T_{logic} + T_{CQ} + T_{setup}}{\frac{T_{logic}}{N} + T_{CQ} + T_{setup}}$
 - Assumes logic can be equally divided. Remember Longest pipe-stage sets cycle-time
 - Area
 - Power*
 - $P_N = P + \sum_{i}^{N} P_i$
- **Exercise**. Consider an unpipelined design with T_{logic} = 2ns. Sketch the plot for cycle time vs pipeline depth, N with $T_{setup} = T_{cq} = 200ps$

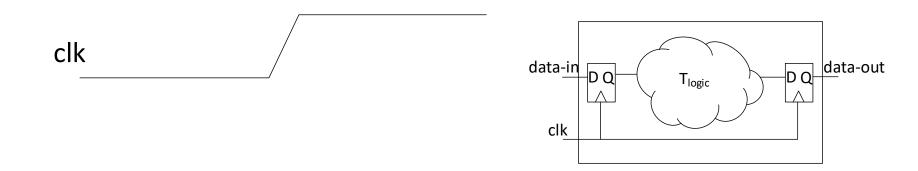


Pipelining Limits



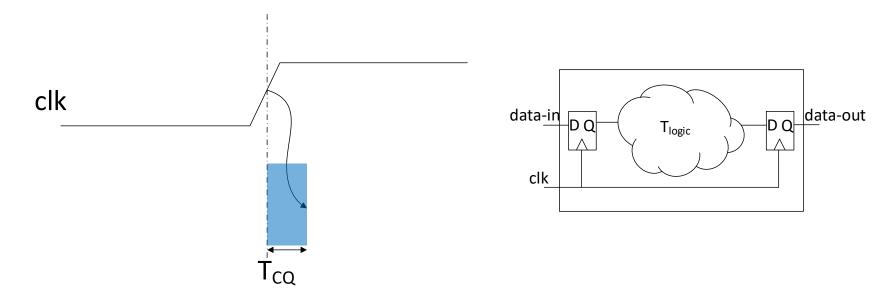
- Ideal flip-flops (Timing and power) → Pipelining is free of cost
- Real flip-flops → Pipelining provides benefits only up to a point
 - Non-zero delays, capacitive loading, power and area count diminish returns





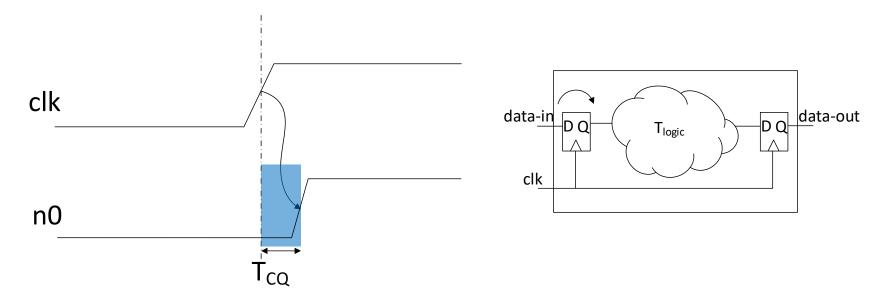
- Analysis examines susceptibility to hold-time violations in FFs
- Sets a minimum delay between launching clock the next FF's A.T
- Consider early arriving data at input of given FF





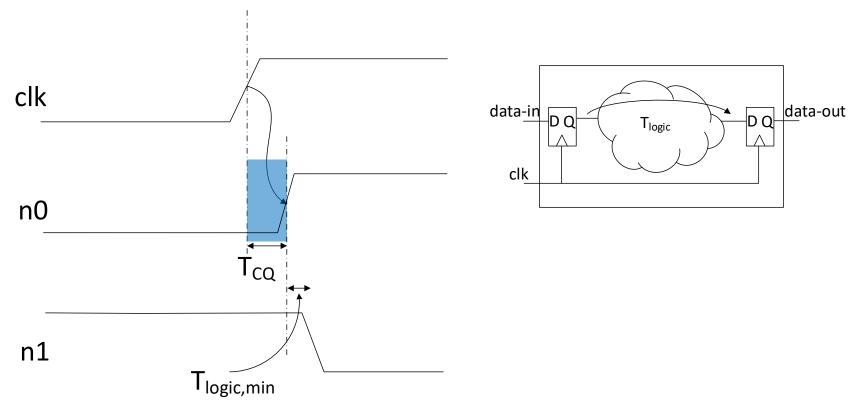
- Analysis examines susceptibility to hold-time violations in FFs
- Sets a minimum delay between launching clock the next FF's A.T
- Consider early arriving data at input of given FF





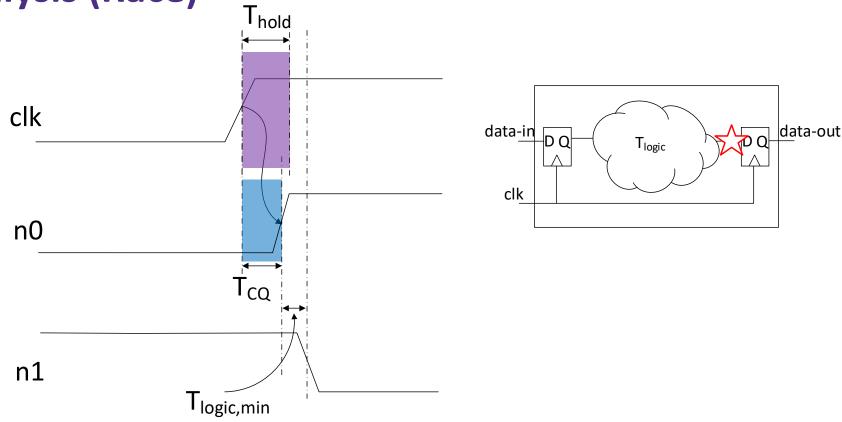
- Analysis examines susceptibility to hold-time violations in FFs
- Sets a minimum delay between launching clock the next FF's A.T
- Consider early arriving data at input of given FF





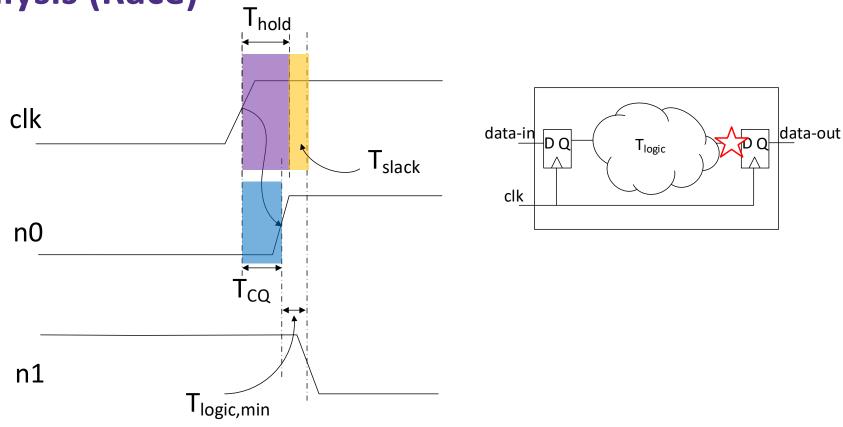
- Analysis examines susceptibility to hold-time violations in FFs
- Sets a minimum delay between launching clock the next FF's A.T
- Consider early arriving data at input of given FF





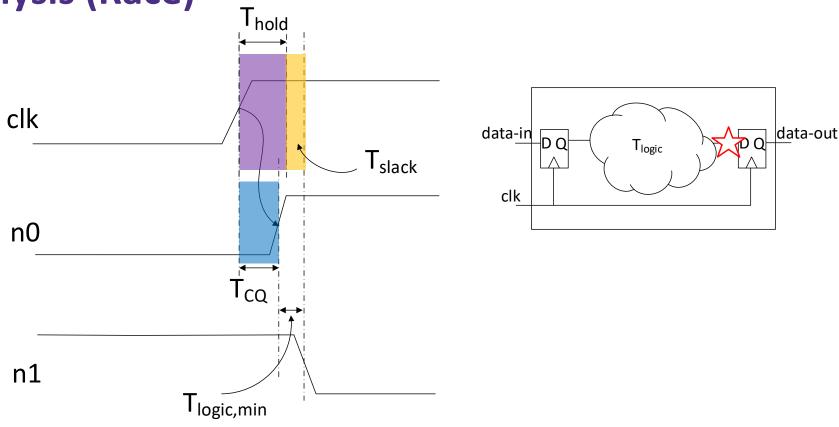
- Analysis examines susceptibility to hold-time violations in FFs
- Sets a minimum delay between launching clock the next FF's A.T
- Consider early arriving data at input of given FF





- Analysis examines susceptibility to hold-time violations in FFs
- Sets a minimum delay between launching clock the next FF's A.T
- Consider early arriving data at input of given FF

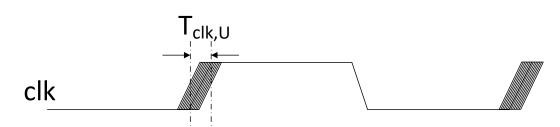




- Analysis examines susceptibility to hold-time violations in FFs
- Sets a minimum delay between launching clock the next FF's A.T
- Consider early arriving data at input of given FF
- $T_{hold} \le T_{CQ} + T_{logic,min}$



Clock Uncertainty



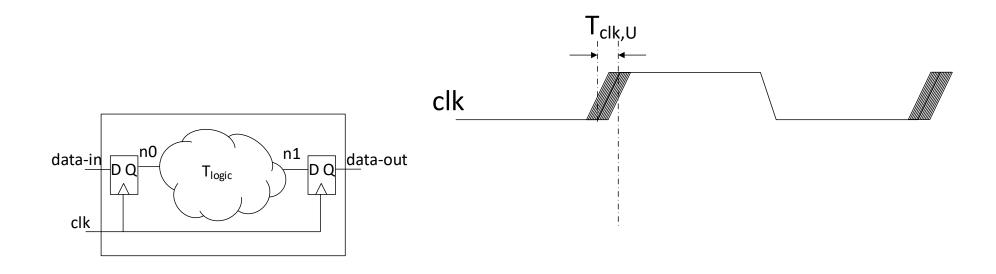
- 2 components: Skew and jitter
- Clock skew
 - Earlier analysis assumed that *clk* arrives at all flops simultaneously
 - Difficult to do in reality
 - In real designs, clk arrives at different times
 - Time invariant component (invariant relative to clock propagation time) is referred to as clock skew: T_{skew}
 - Time-invariant (sort of) component: T_{skew}
- Clock jitter
 - Time-variant component : T_{jitter}
 - Accurate analysis of impact on timing is a little more involved (T_{hold} particularly)
- Simplifying assumption: Worst case clock uncertainty everywhere
 - $T_{clk,U} = T_{skew} + T_{jitter}$





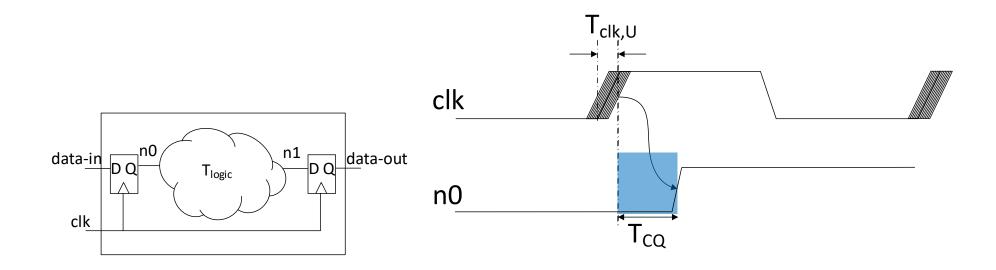
- Setup time analysis: Late launch, early capture
- Cycle-time
 - $T_{cycle} \ge T_{cq} + T_{logic,max} + T_{setup} + T_{clk,U}$





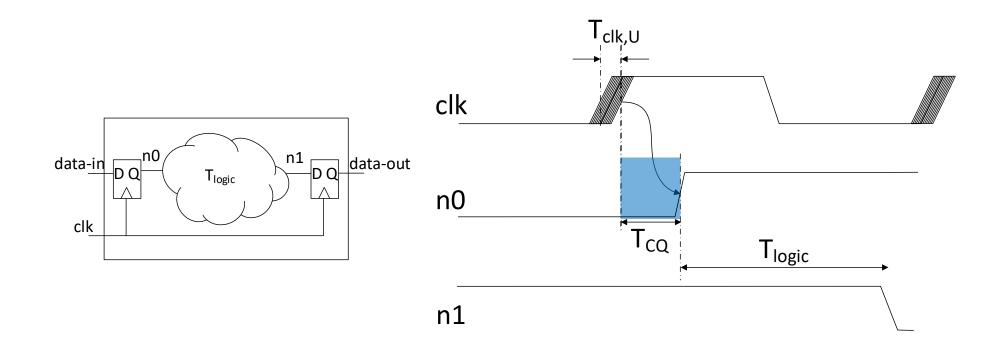
- Setup time analysis: Late launch, early capture
- Cycle-time
 - $T_{cycle} \ge T_{cq} + T_{logic,max} + T_{setup} + T_{clk,U}$





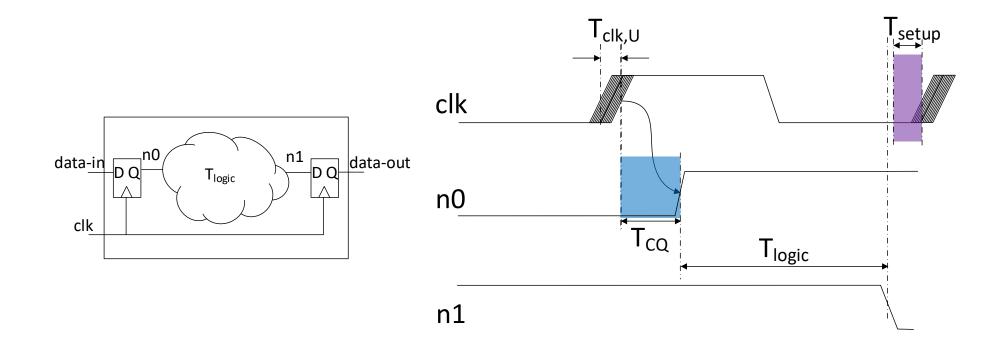
- Setup time analysis: Late launch, early capture
- Cycle-time
 - $T_{cycle} \ge T_{cq} + T_{logic,max} + T_{setup} + T_{clk,U}$





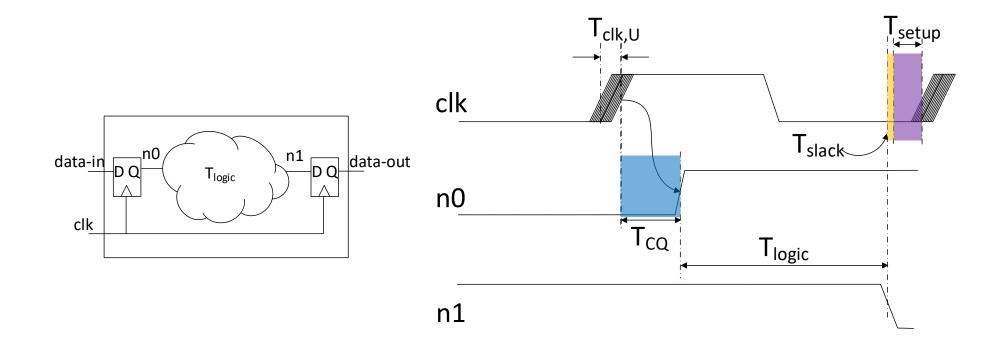
- Setup time analysis: Late launch, early capture
- Cycle-time
 - $T_{cycle} \ge T_{cq} + T_{logic,max} + T_{setup} + T_{clk,U}$





- Setup time analysis: Late launch, early capture
- Cycle-time
 - $T_{cycle} \ge T_{cq} + T_{logic,max} + T_{setup} + T_{clk,U}$





- Setup time analysis: Late launch, early capture
- Cycle-time
 - $T_{cycle} \ge T_{cq} + T_{logic,max} + T_{setup} + T_{clk,U}$



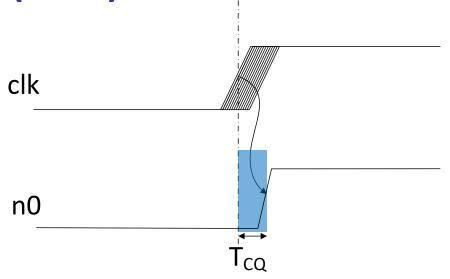
Clock Uncertainty (Hold)

clk

- Analysis examines susceptibility to hold-time violations in FFs
- Sets a minimum delay between launching clock the next FF's A.T
- Consider early arriving data at input of given FF



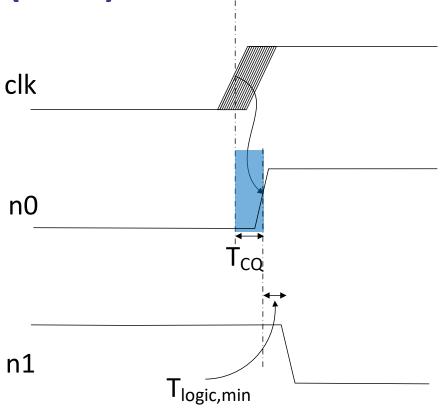
Clock Uncertainty (Hold)



- Analysis examines susceptibility to hold-time violations in FFs
- Sets a minimum delay between launching clock the next FF's A.T
- Consider early arriving data at input of given FF



Clock Uncertainty (Hold)



- Analysis examines susceptibility to hold-time violations in FFs
- Sets a minimum delay between launching clock the next FF's A.T
- Consider early arriving data at input of given FF



Clock Uncertainty (Hold) $\mathsf{T}_{\mathsf{hold}}$ clk n0 n1 I logic, min

- Analysis examines susceptibility to hold-time violations in FFs
- Sets a minimum delay between launching clock the next FF's A.T
- Consider early arriving data at input of given FF



Clock Uncertainty (Hold) $\mathsf{T}_{\mathsf{hold}}$ clk $\mathsf{T}_{\mathsf{slack}}$ (violated) n0 n1 I logic, min

- Analysis examines susceptibility to hold-time violations in FFs
- Sets a minimum delay between launching clock the next FF's A.T
- Consider early arriving data at input of given FF

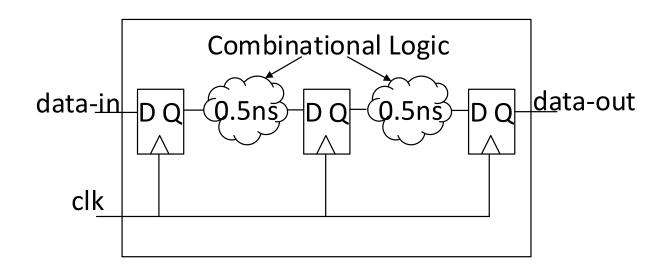


Clock Uncertainty (Hold) $\mathsf{T}_{\mathsf{hold}}$ clk T_{slack} (violated) n0 n1 $T_{logic,min}$

- Analysis examines susceptibility to hold-time violations in FFs
- Sets a minimum delay between launching clock the next FF's A.T
- Consider early arriving data at input of given FF
- $T_{hold} \le T_{CQ} + T_{logic,min} T_{clk,U}$

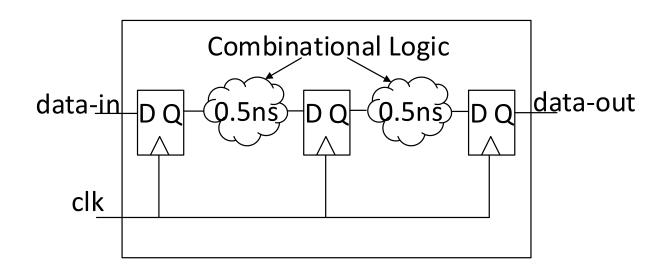


- Consider the following pipeline design
 - Combinational logic is fine-grained (Can be divided into stages easily)
 - $T_{\text{setup}} = 100 \text{ps}$, $T_{\text{cq}} = 100 \text{ps}$, $T_{\text{hold}} = 100 \text{ps}$, $T_{\text{clk,U}} = 50 \text{ps}$
- Determine current cycle time
- Determine T_{logic,min} required for robust operation
- Determine cycle time if one extra pipeline stage were added.



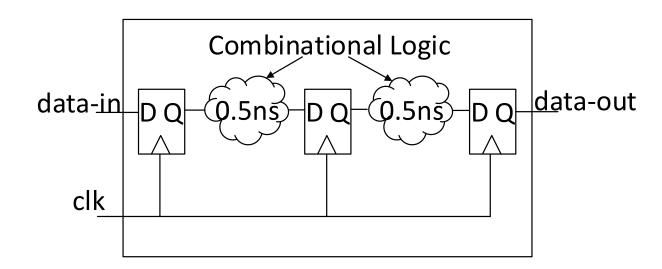


- Consider the following pipeline design
 - Combinational logic is fine-grained (Can be divided into stages easily)
 - $T_{\text{setup}} = 100 \text{ps}$, $T_{\text{cq}} = 100 \text{ps}$, $T_{\text{hold}} = 100 \text{ps}$, $T_{\text{clk,U}} = 50 \text{ps}$
- Determine current cycle time $(T_{cq} + T_{logic} + T_{setup} + T_{clk,U})$
- Determine T_{logic,min} required for robust operation
- Determine cycle time if one extra pipeline stage were added.



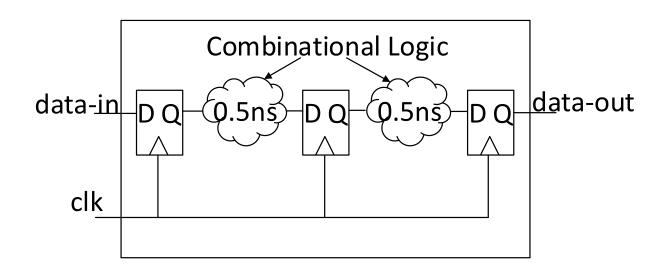


- Consider the following pipeline design
 - Combinational logic is fine-grained (Can be divided into stages easily)
 - $T_{\text{setup}} = 100 \text{ps}$, $T_{\text{cq}} = 100 \text{ps}$, $T_{\text{hold}} = 100 \text{ps}$, $T_{\text{clk,U}} = 50 \text{ps}$
- Determine current cycle time $(T_{cq} + T_{logic} + T_{setup} + T_{clk,U}) = 750ps$
- Determine T_{logic,min} required for robust operation
- Determine cycle time if one extra pipeline stage were added.



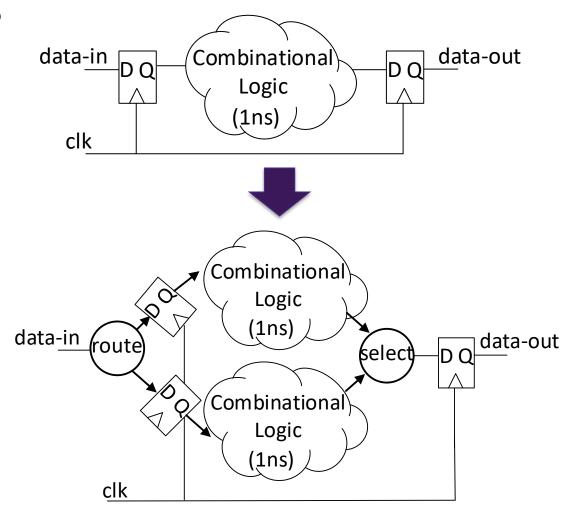


- Consider the following pipeline design
 - Combinational logic is fine-grained (Can be divided into stages easily)
 - $T_{\text{setup}} = 100 \text{ps}$, $T_{\text{cq}} = 100 \text{ps}$, $T_{\text{hold}} = 100 \text{ps}$, $T_{\text{clk,U}} = 50 \text{ps}$
- Determine current cycle time $(T_{cq} + T_{logic} + T_{setup} + T_{clk,U}) = 750ps$
- Determine $T_{logic,min}$ required for robust operation T_{hold} - T_{cq} + $T_{clk,U}$ =50p
- Determine cycle time if one extra pipeline stage were added.





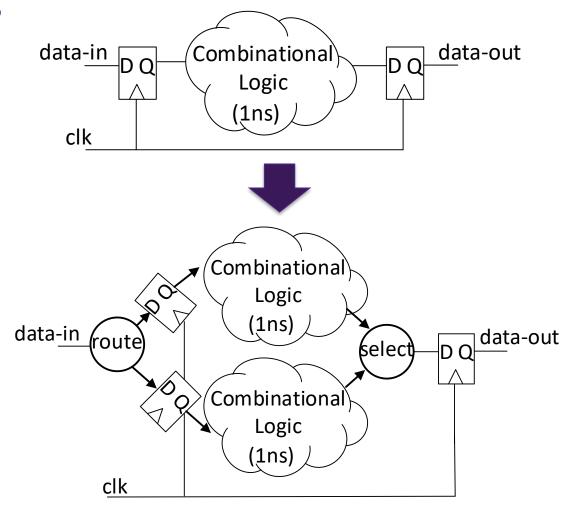
Parallel Processing



- An alternative approach to enhanced processing capability
- Conduct multiple computations in parallel. Combine results



Parallel Processing

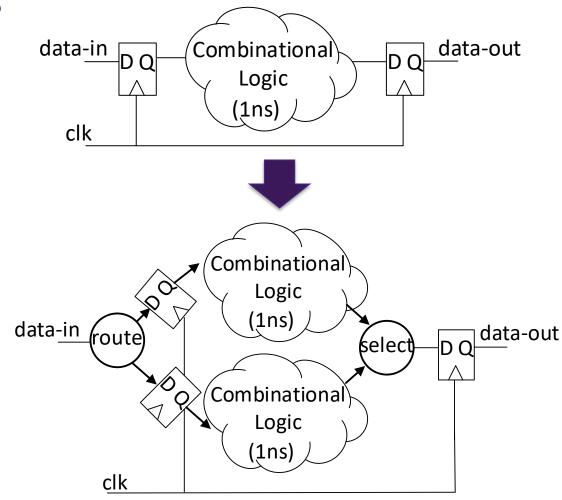


- An alternative approach to enhanced processing capability
- Conduct multiple computations in parallel. Combine results
- There's a bug here ...





Parallel Processing



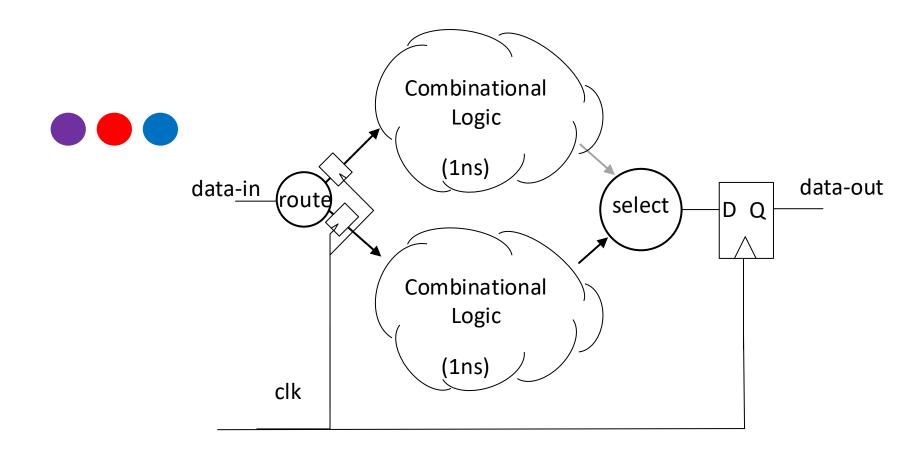
Equal intervals between channel change?



- An alternative approach to enhanced processing capability
- Conduct multiple computations in parallel. Combine results
- There's a bug here ...



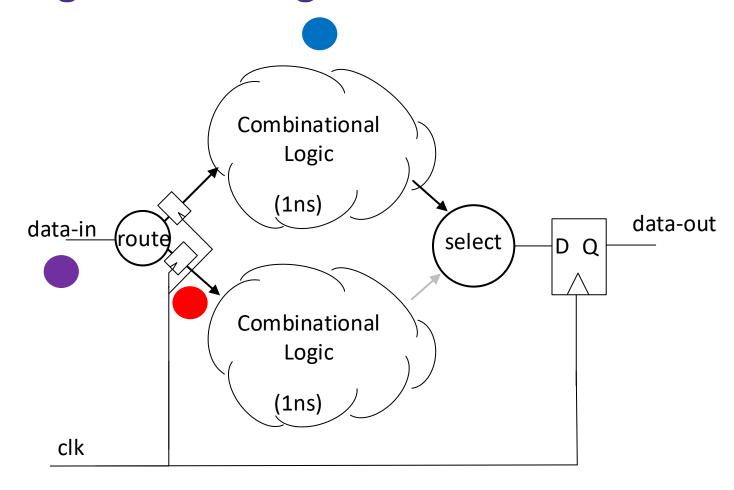
Throughput not determined only be combinational logic latency



Throughput not determined only be combinational logic latency



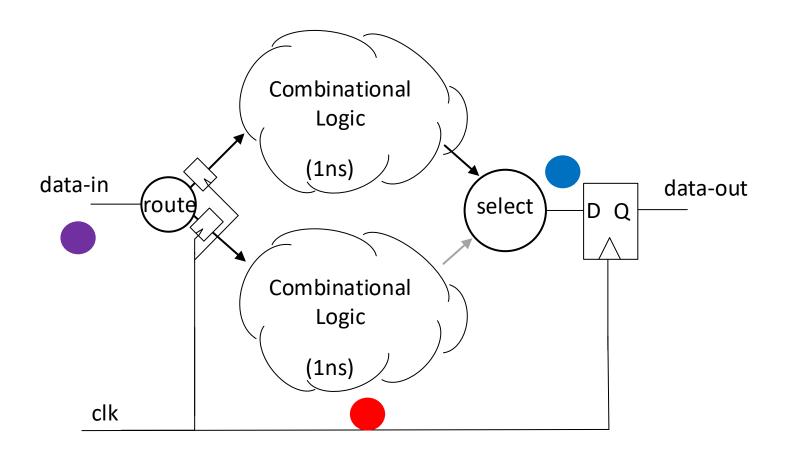
Parallel Processing: Walkthrough



- After ~0.5ns, blue computation still being conducted
- Launch red computation through the other parallel logic

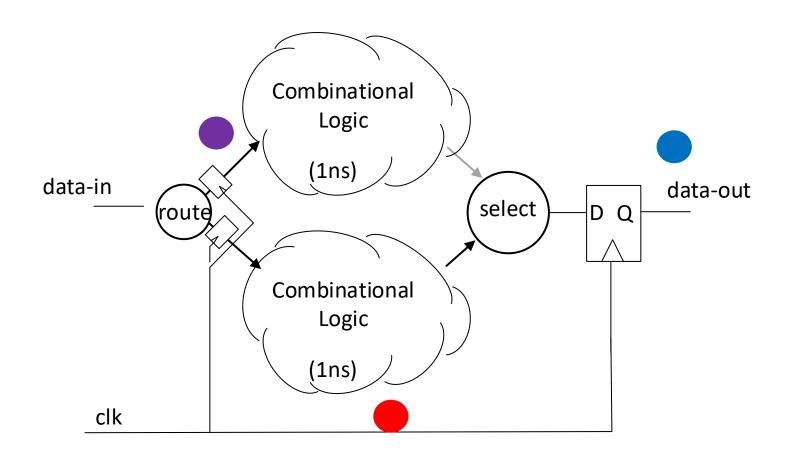


Parallel Processing: Walkthrough





Parallel Processing: Walkthrough



- At T~1ns, blue computation completed
- Ready to launch purple computation
- Red computation still being performed.



Parallel Processing: Gains

- Exercise: Given T_{logic} =1ns, T_{router} = 100ps, T_{select} = 100ps
 - What is the speedup of a 2-way parallel processing computation.

Parallel Processing Limitations

- Inter-computation dependencies
 - Many computations depend on each other (esp. microprocessors)
 - Parallel processing: Send, forget for n cycles, capture
 - Does not provide a mechanism for handling such dependencies
- Other limitations
 - Increased cost (die area)
 - Increased leakage power



Reading assignment

- W&H: 10.1-10.2.3
- Extra Reading: 10.3