Lecture 9: The Register File Design

Based on material prepared by prof. Visvesh S. Sathe

ELECTRICAL & COMPUTER ENGINEERING

Acknowledgements

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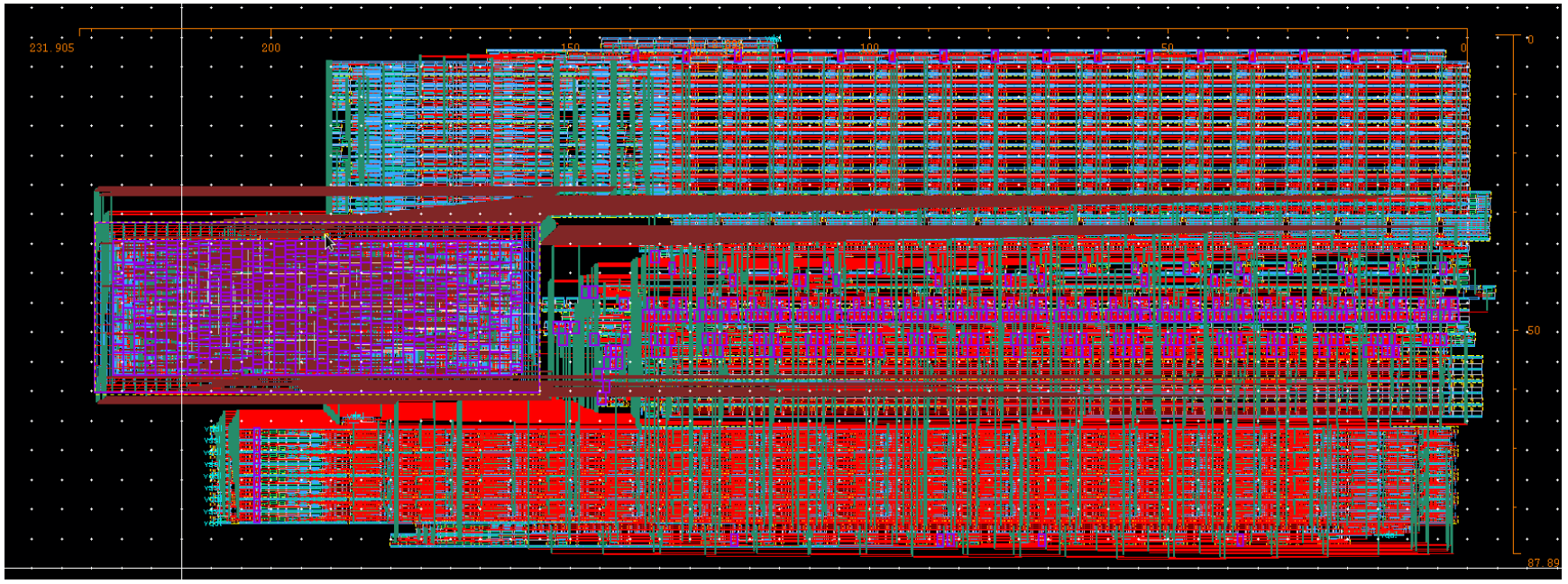
UW (2013-2022)
GaTech (2022-present)

Example Microprocessor Layout



- Datapath layed-out in a regular structure
- "Bitslice" is an important concept/metric
  - Aligning modules and wires along each of the 16 bits of your design will help
  - Data flow from regfile -> shift/alu/pc, from shift/alu ->regfile occur naturally along bit-boundaries.
  - E.g. regfile-out lines up with alu/shift/pc IN
  - Regfile-in lines up with alu/shift/ OUT

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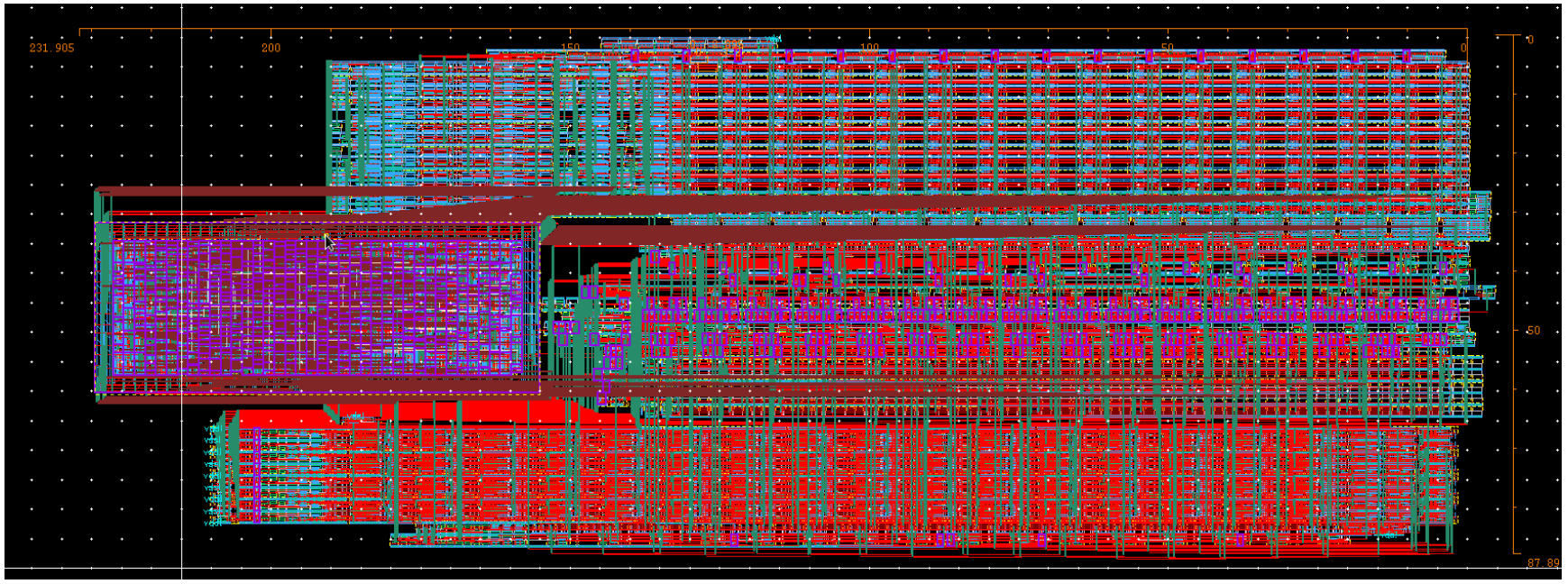

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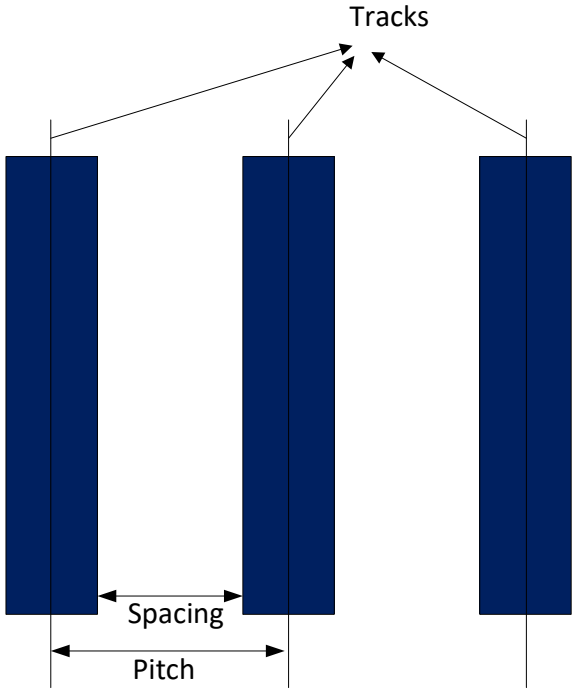
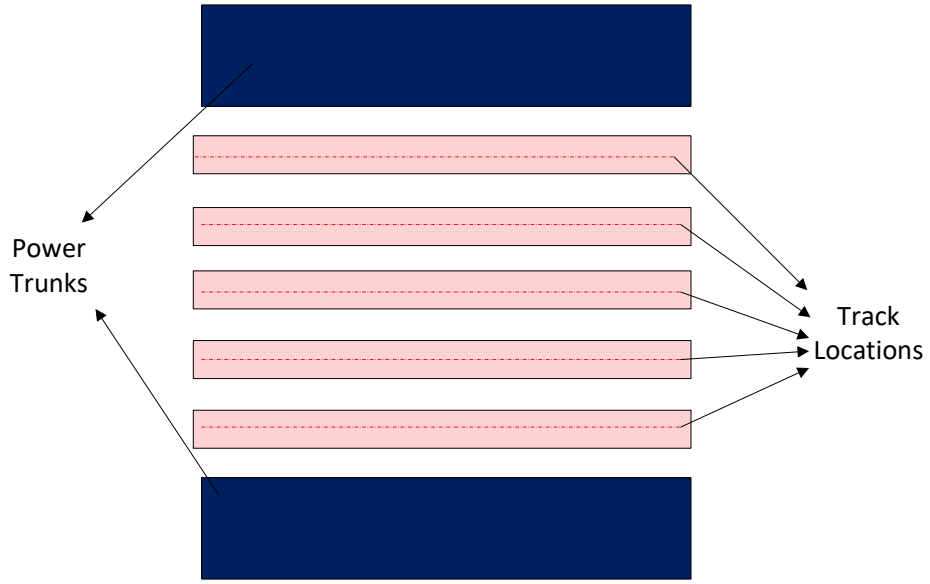
COMMON BITSLICE!!!

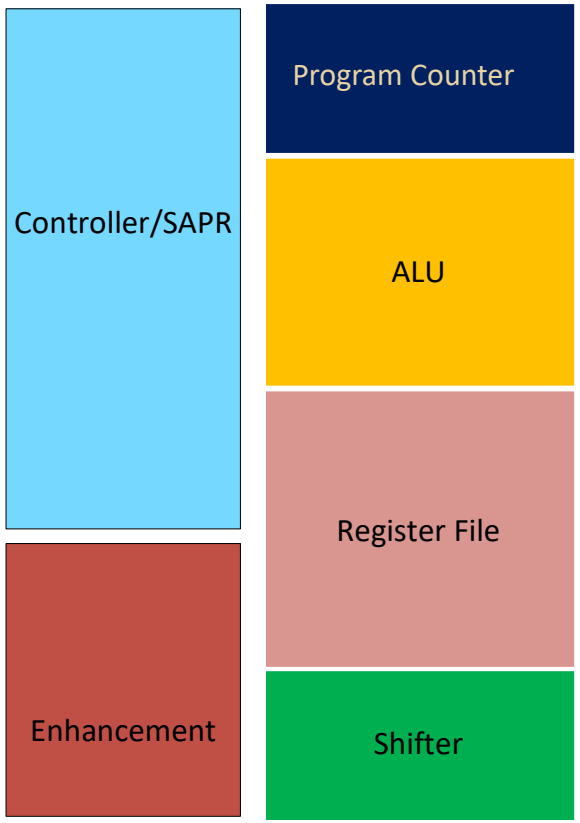
Layout Design Conventions





- Spacing : Space between object edges
- Pitch : Space between object centerlines
- Track : Centerline of region designated for metal route
- Cell-height ： Height of cell from Vdd-centerline to Vss-centerline
- Bit-slice width : **Width** of design corresponding to 1 bit
  - Important to maintain uniform bit-slice width across your structures

Bit-slices and your Datapath



- Bit-slice: Unit section of your datapath corresponding to 1 bit
- Maintain uniform bit-slice across your modules
  - If the structure is narrow, insert space, or shorten height

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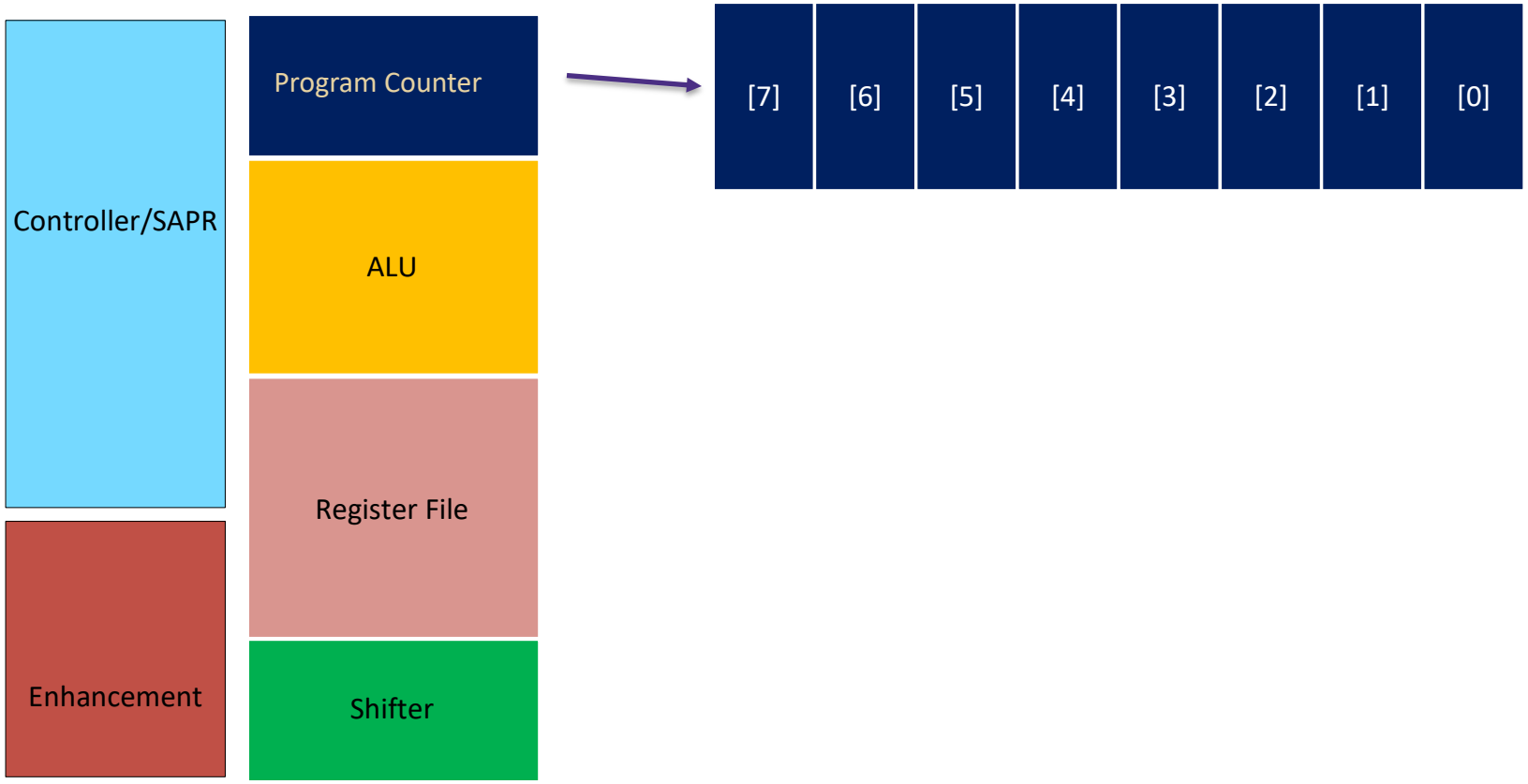

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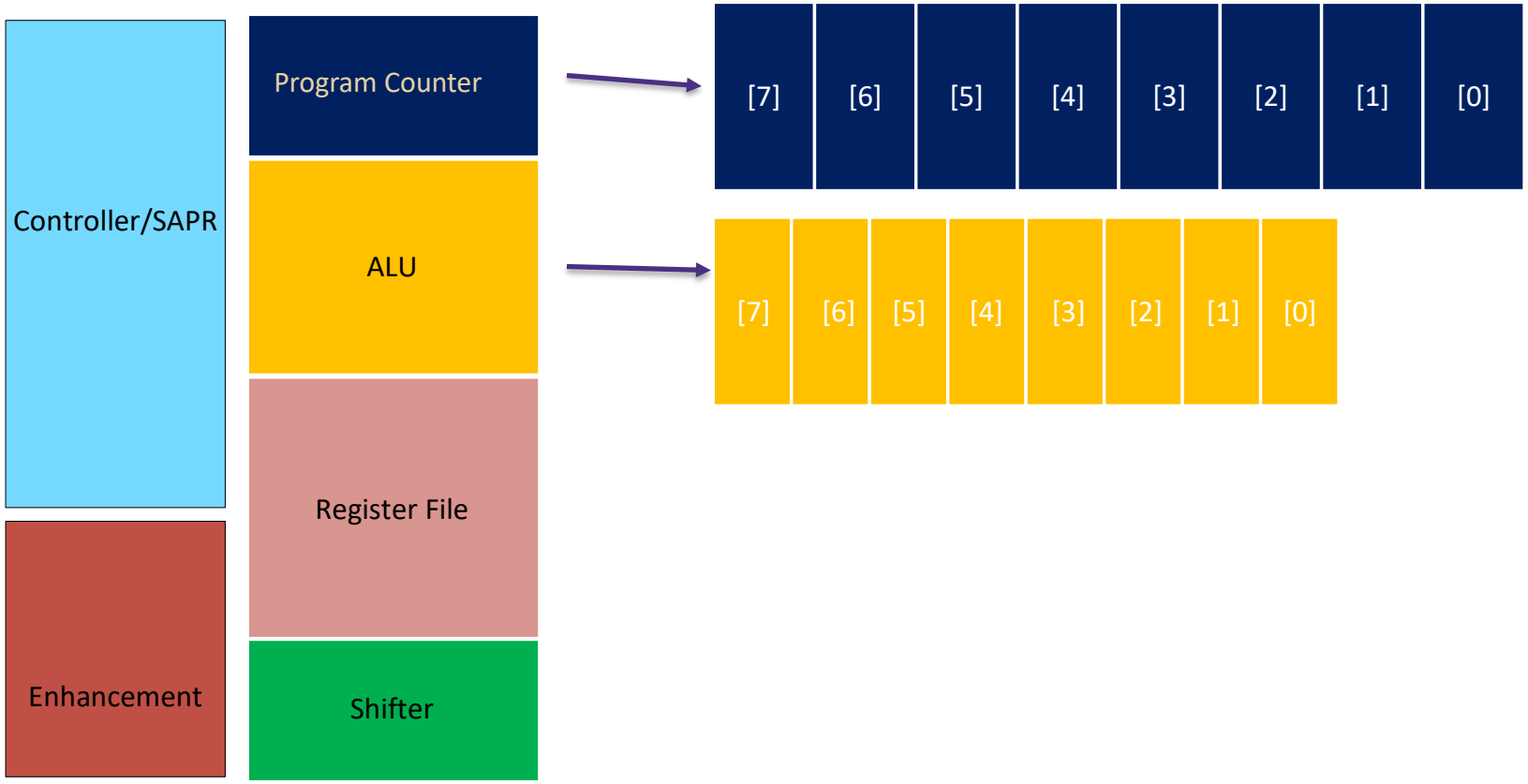

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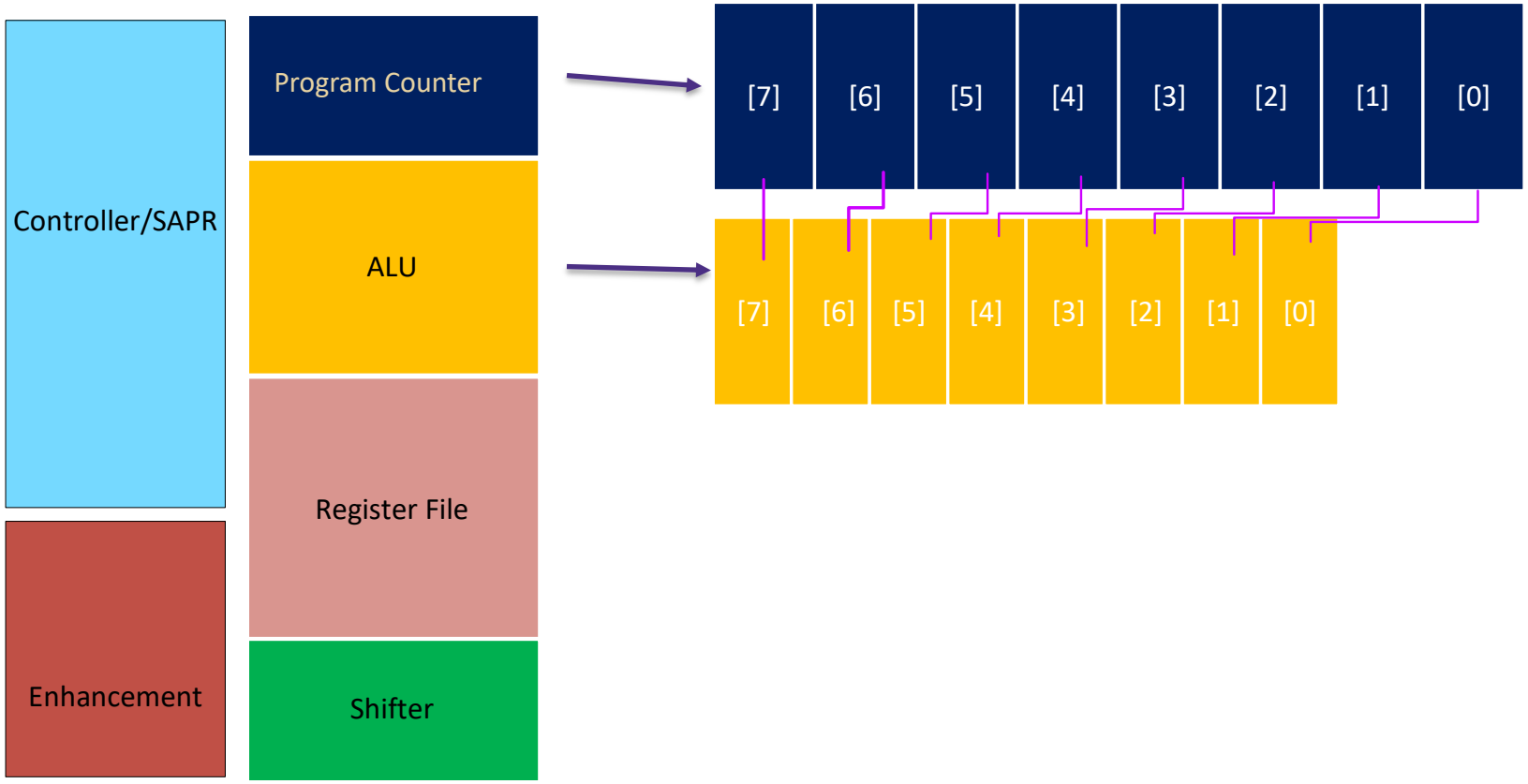

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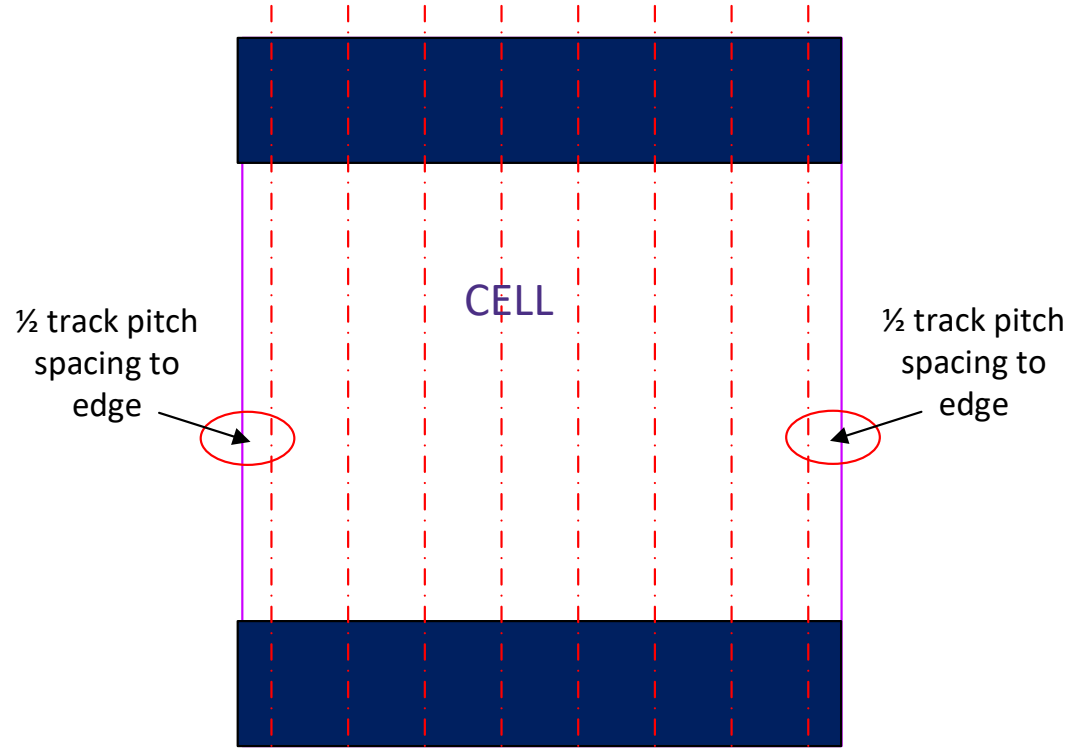
Managing Power Trunks and Tracks



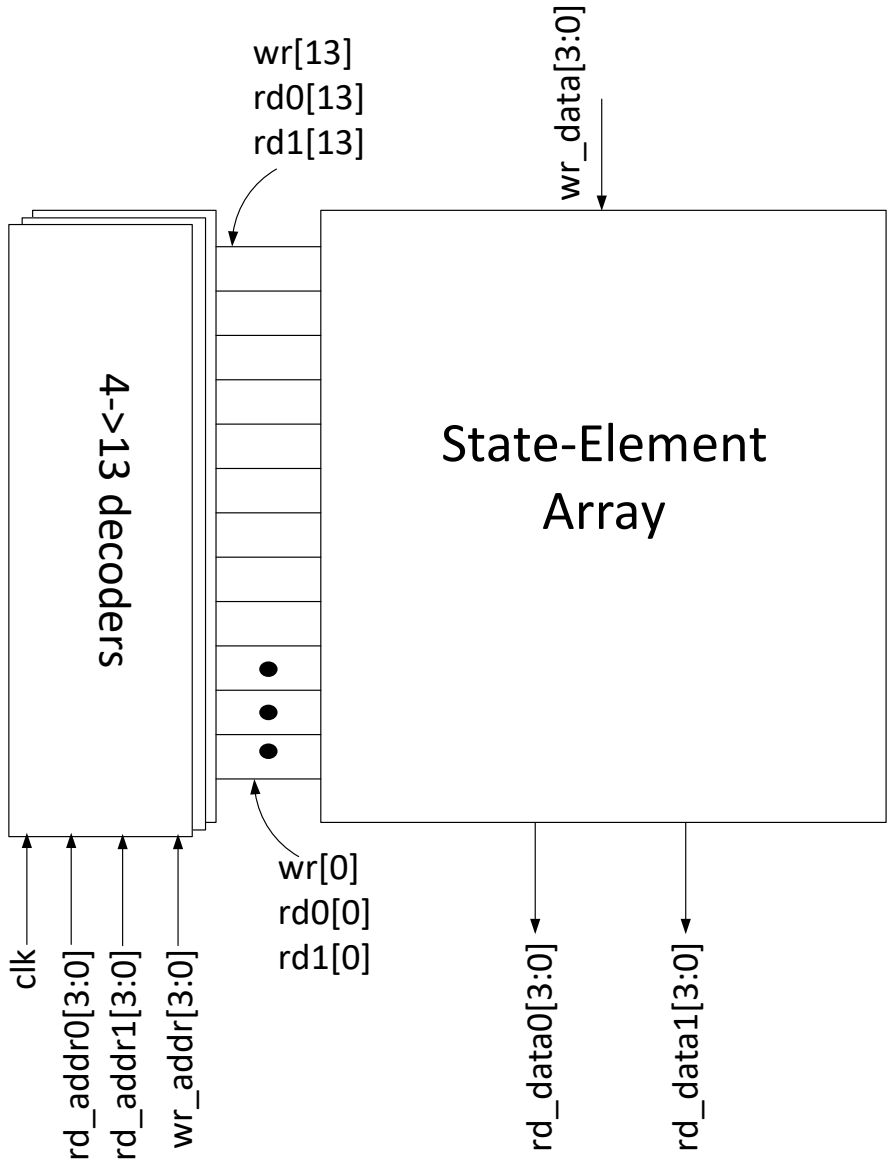
- For example: Min-width = 0.1um, Min-spacing = 0.1um
- Track pitch = 0.2um
- Recommended : Cell boundary 0.5 track pitches away from track location ➔ Cell width is a multiple of track pitch (0.2um in this example)
- **Route on track locations**. Number/Plan tracks

The Register File (Regfile)

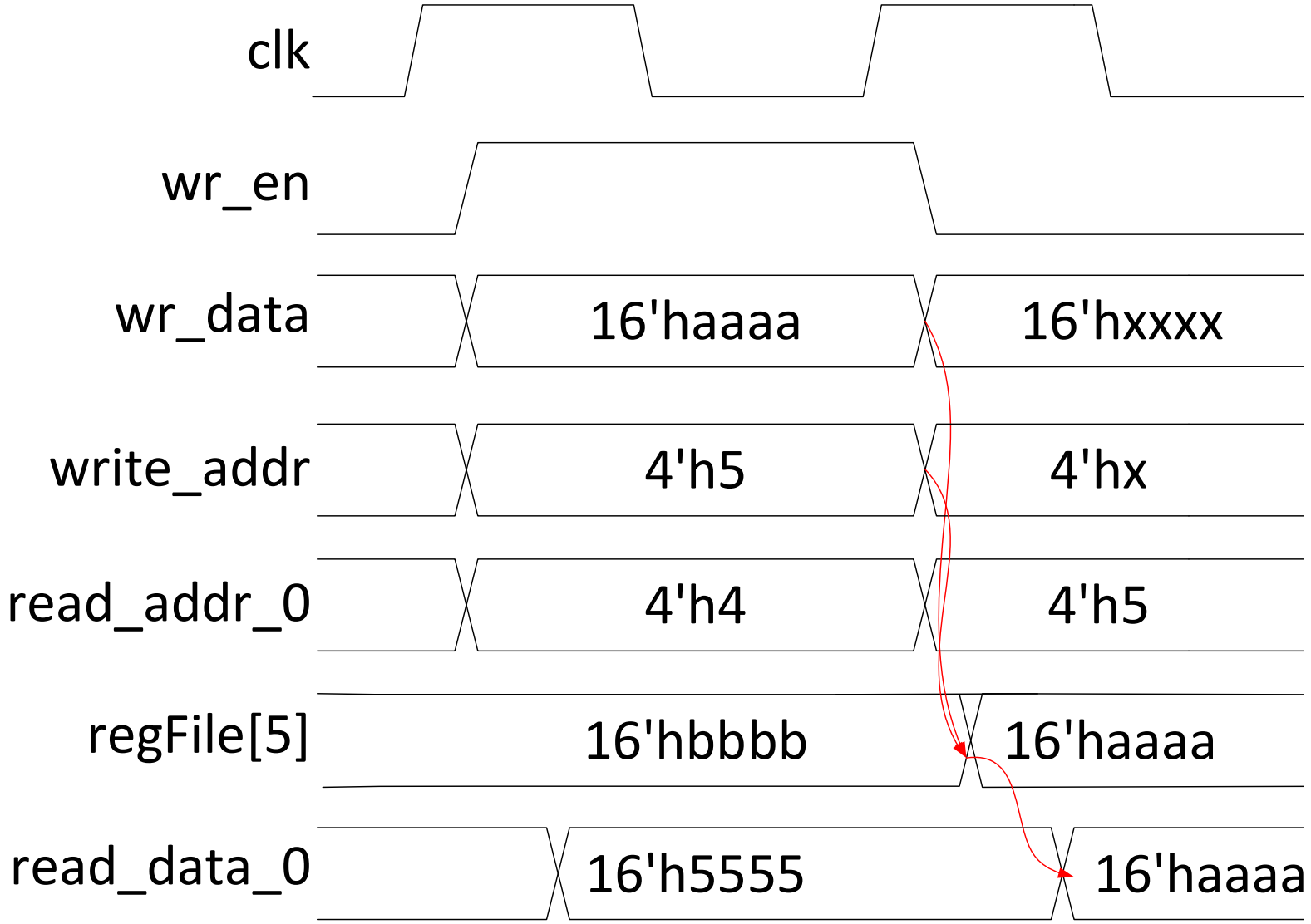
- Store register values (operands for processor instructions)
  - 13-entry (R0-R12), 16-bit
  - 1 write port, 2 read ports
  - Write-before-read regfile operation

- Ports
  - VDD!
  - GND!
  - clk (synchronize the write operation)
  - read_addr_0 (Read address for port 0. 4-bits)
  - read_addr_1 (Read address for port 1. 4-bits)
  - wr_en (Write enable)
  - wr_data (write data, 16 bits)
  - regfile_data0 (Data from read port 0)
  - regfile_data1 (Data from read port 1)

Basic Structure

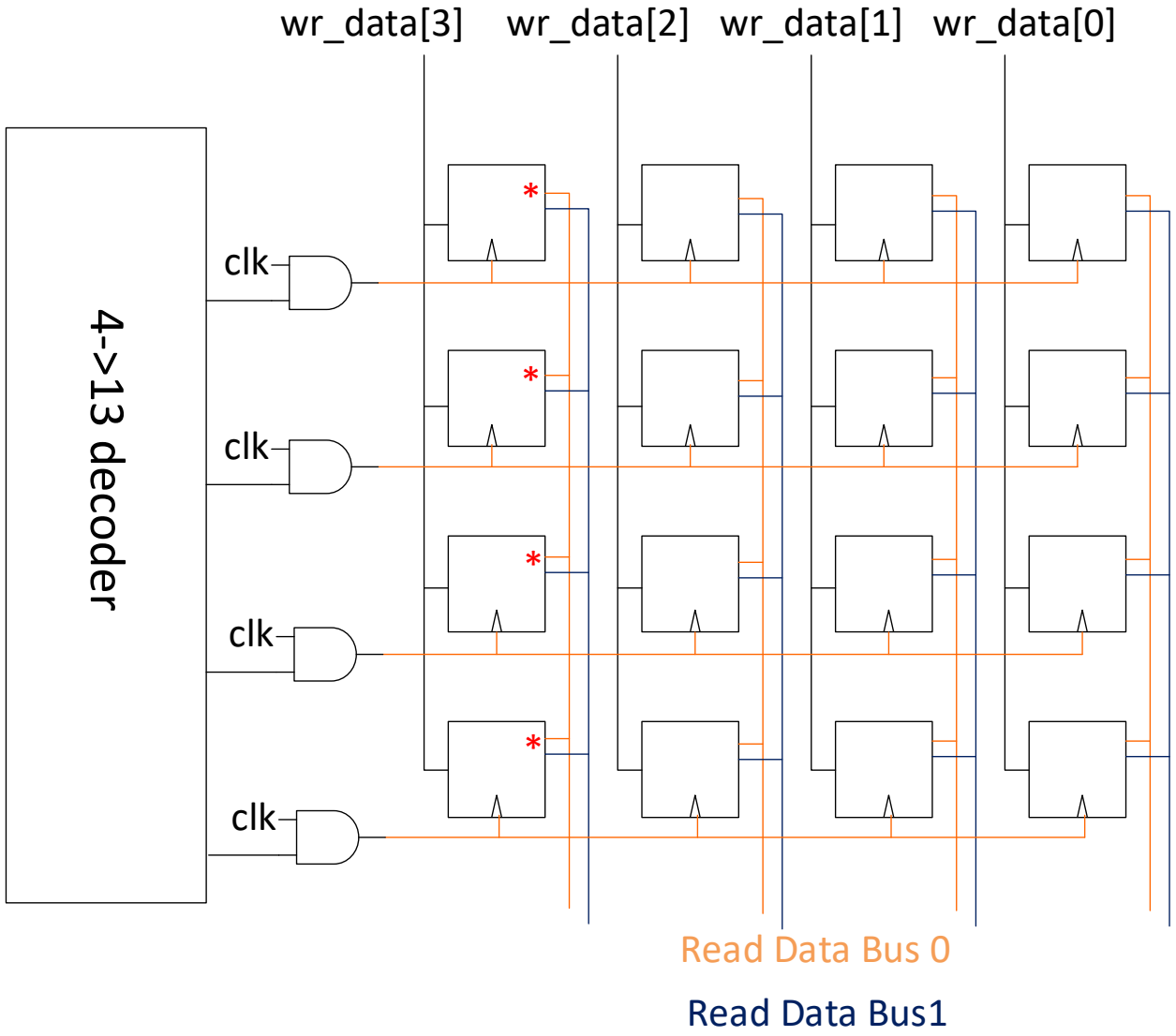


Timing Diagram Walkthrough



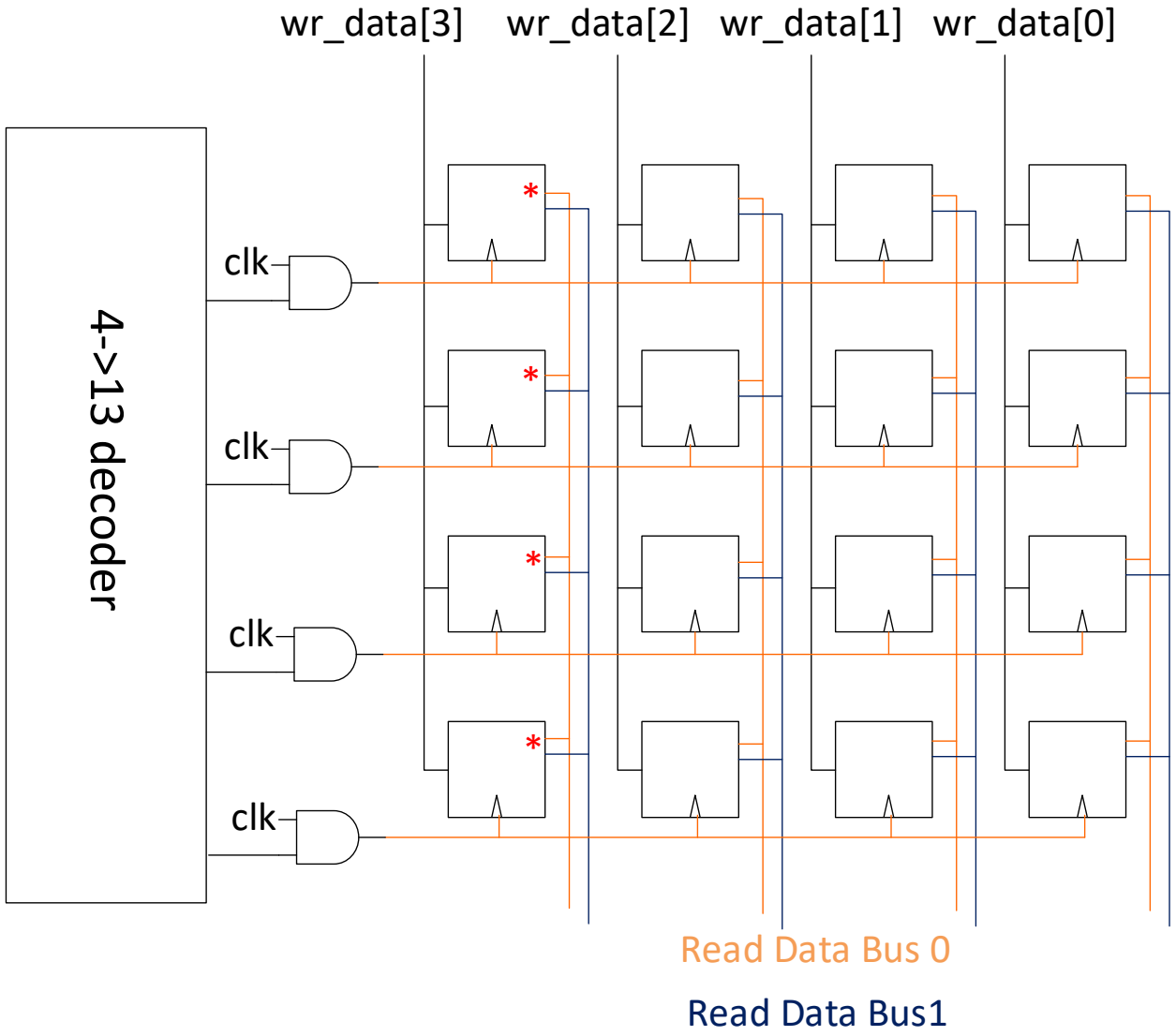
Naïve Implementation

- Array of Flip Flops

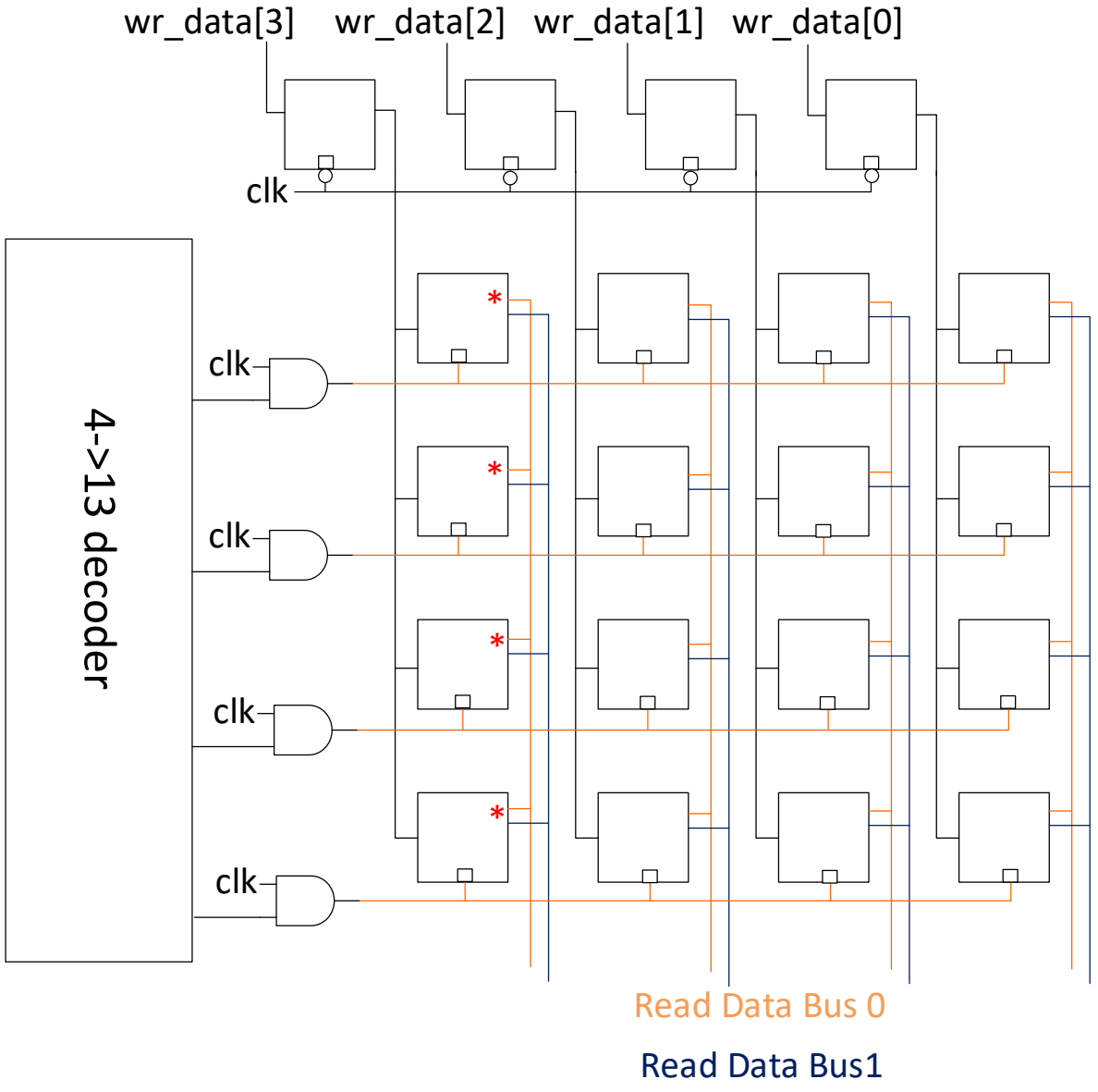


Naïve Implementation

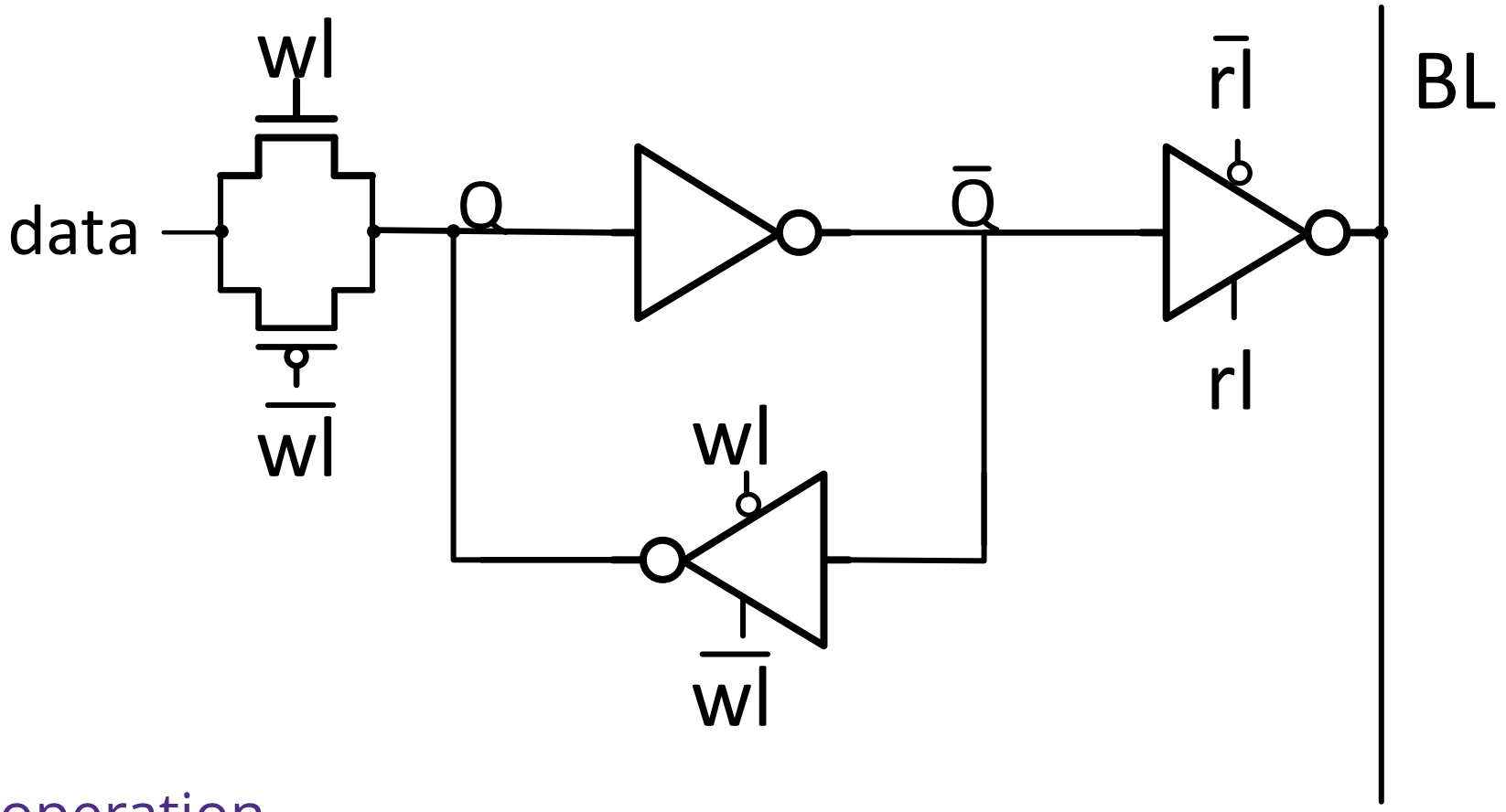
- Array of Flip Flops
- BUT….All master latches are doing the same job!



More Compact Alternative

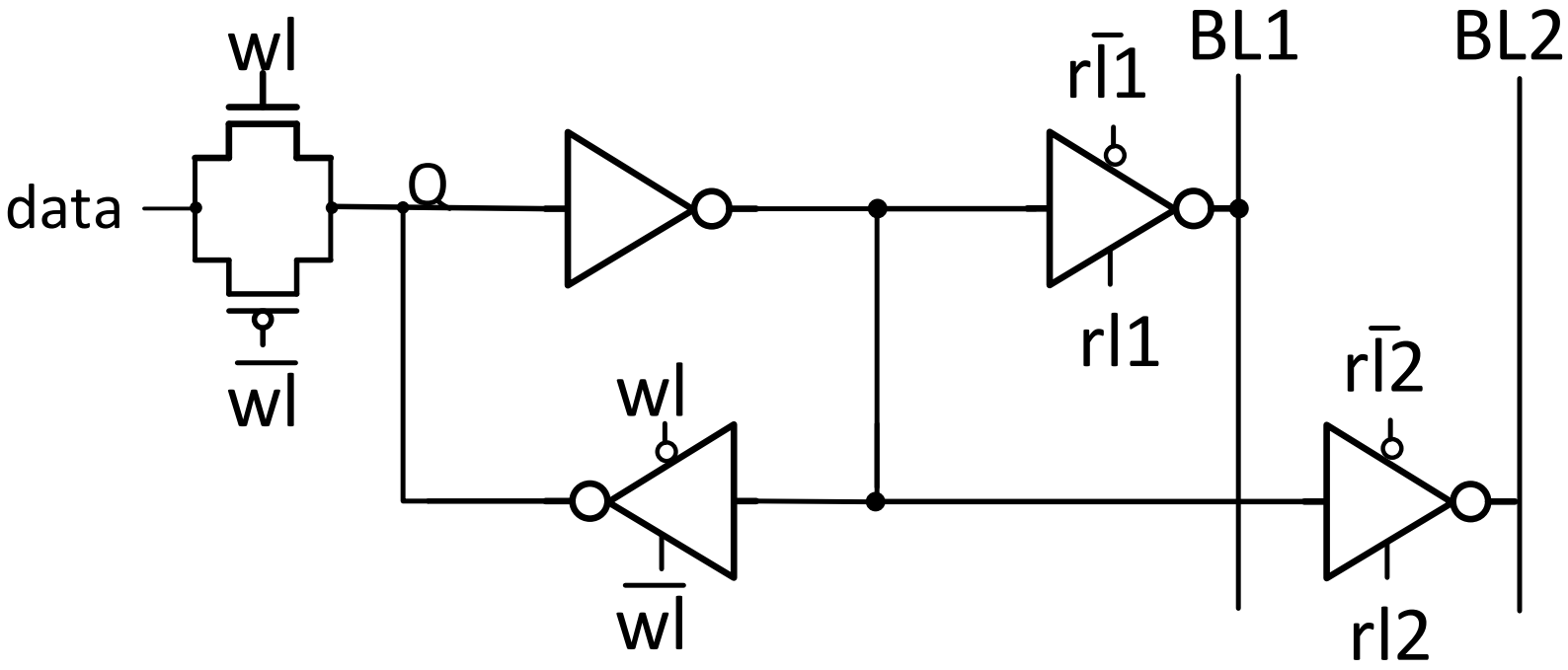


Basic Regfile cell



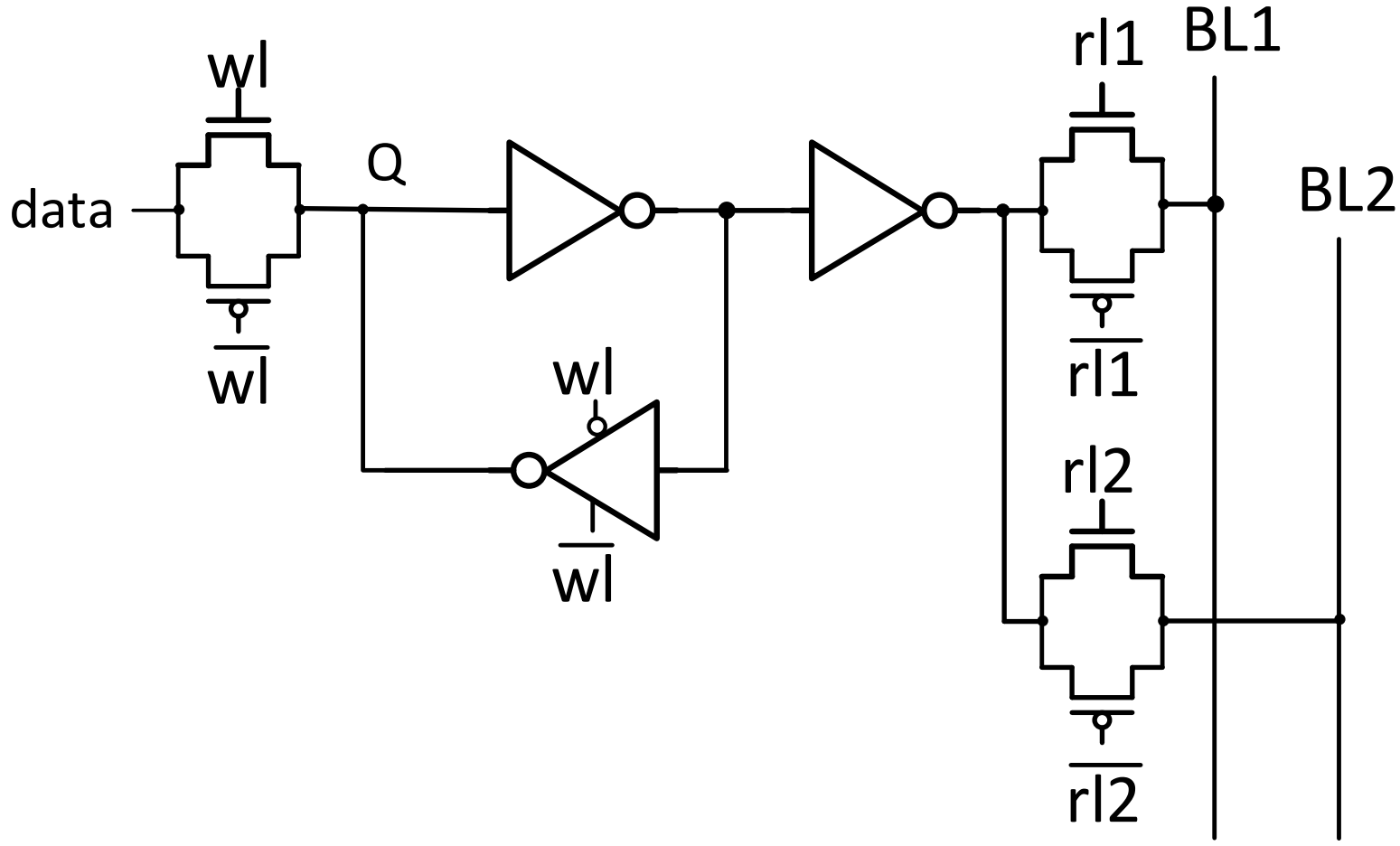
- 12T cell
- Fully static operation

Supporting Multiple Bitlines



- Each bitline is the muxed output connecting all column registers
- What is the capacitance on each bitline?
- How does it impact the size of the drivers?

Transmission Gate Alternative

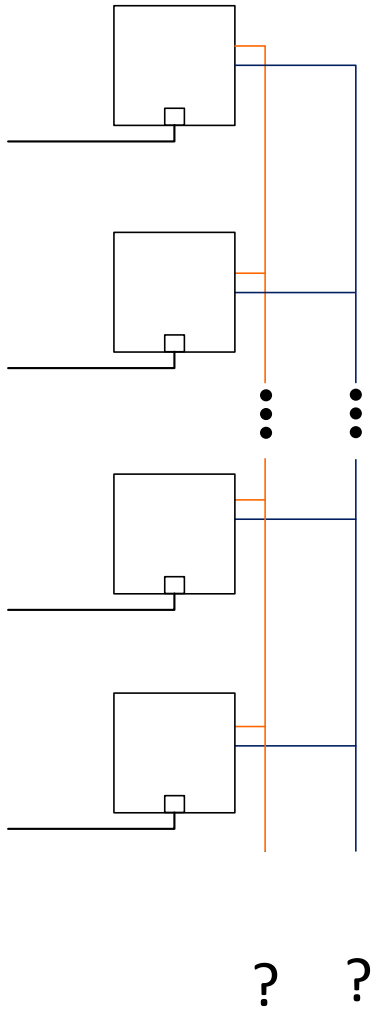


- Advantage: Transmission gate offers lower delay for large loads
- Disadvantage: Critical path loading!
- Is there a third approach to avoid this disadvantage?

Read Delay Analysis

$$T_{read} = T_{clk-q} + T_{bl-drive}$$

- Delay depends on bitcell topology
- Coupling on bitline is important
- Exercise: Order the following delays
  - TX gate output. Bitlines read: (a) Same reg and (b) Different reg
  - Tri-state output. Bitlines read: (a) Same reg and (b) Different reg



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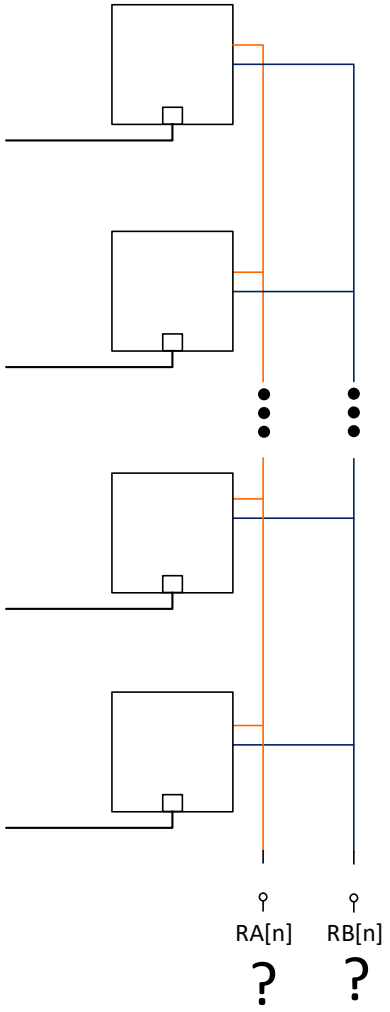

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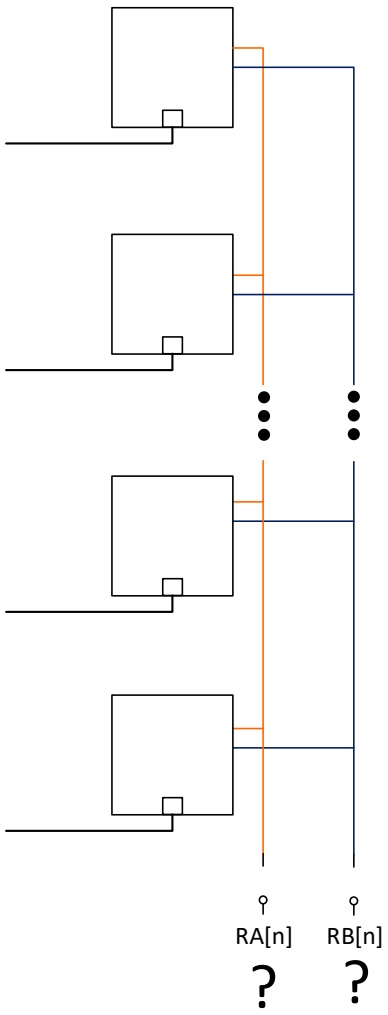

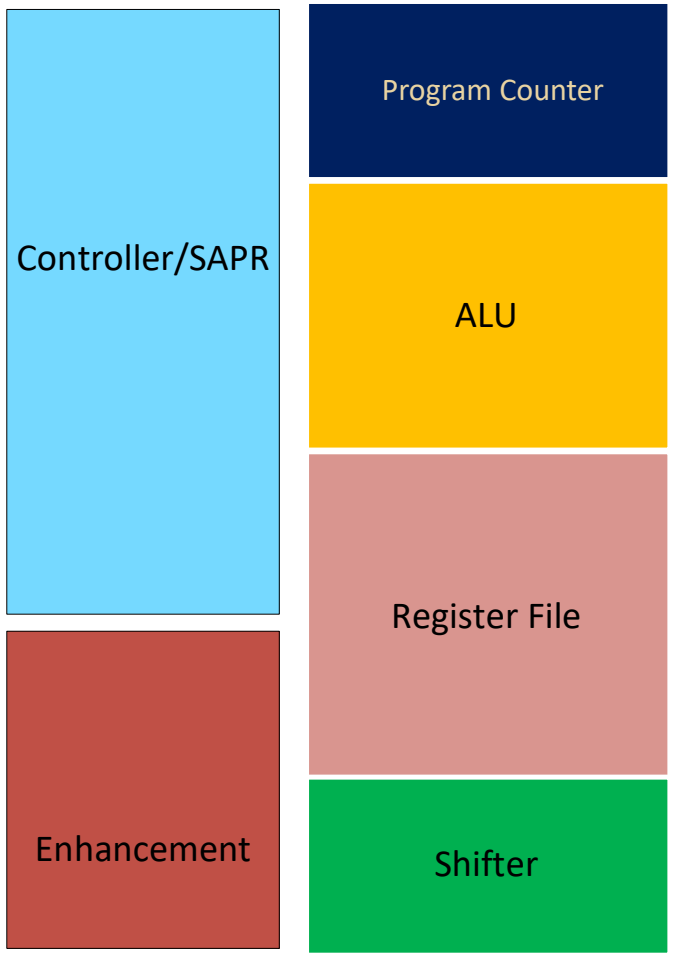

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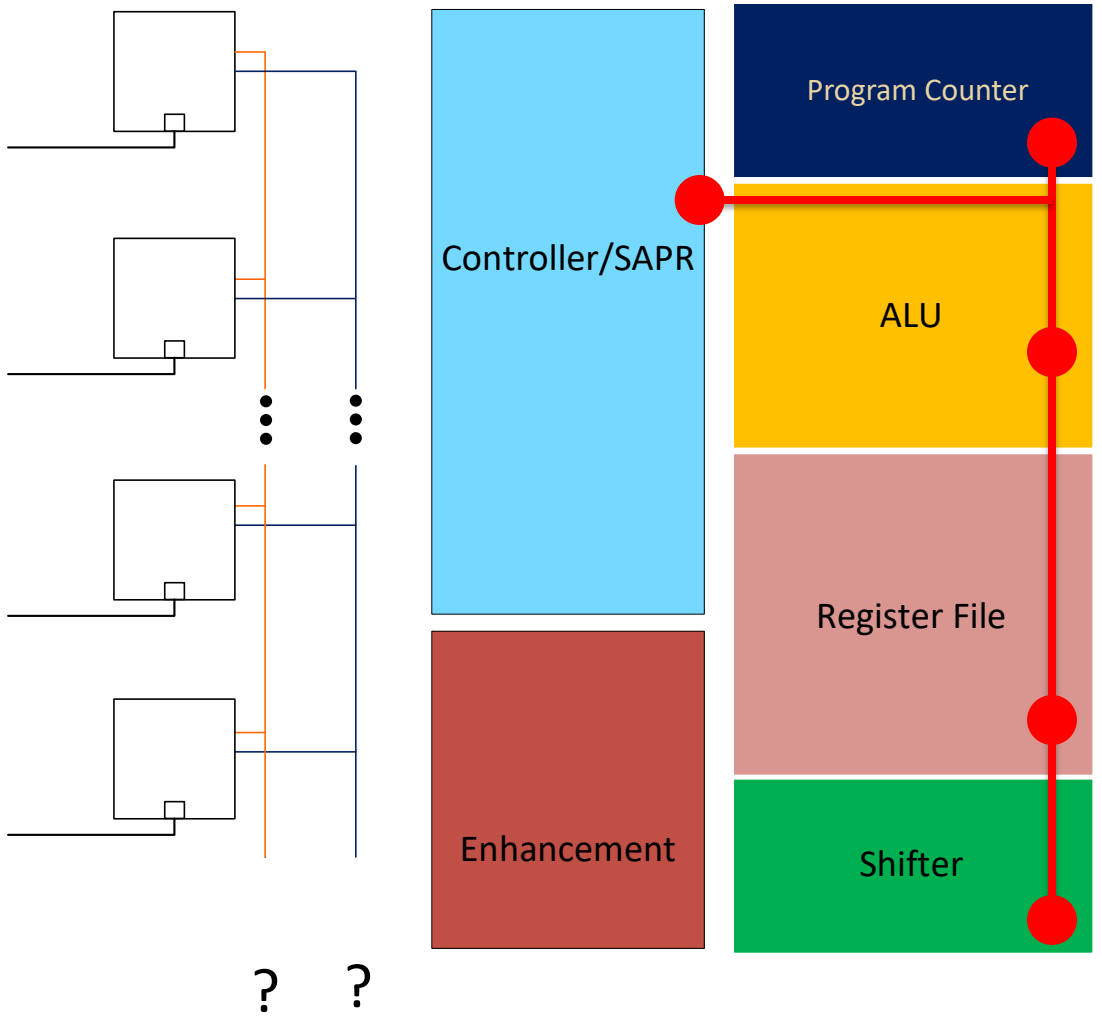

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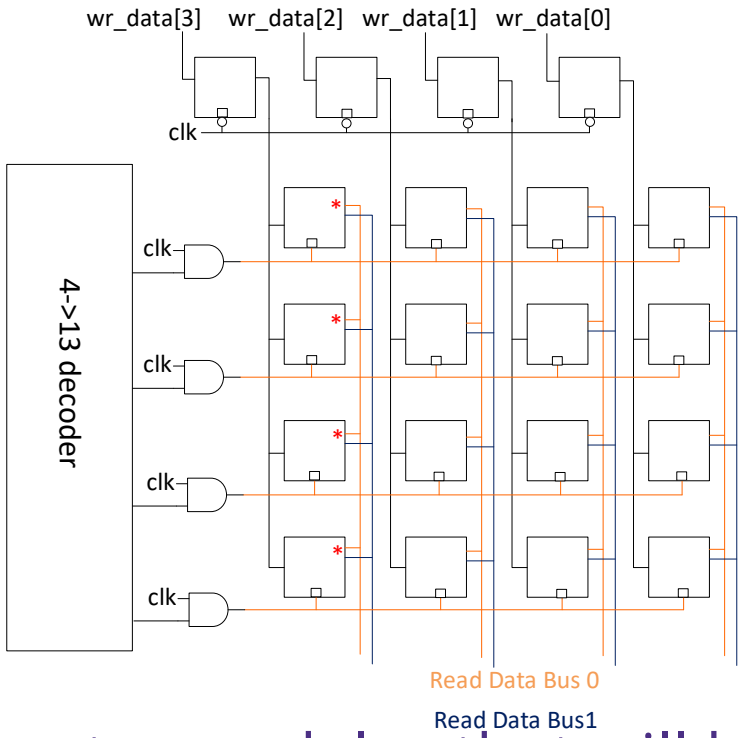
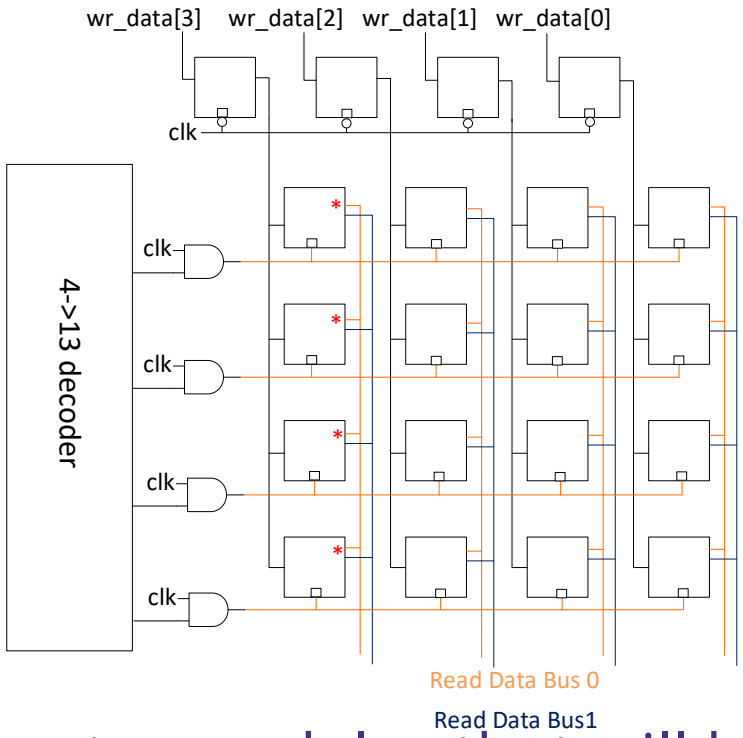


- Exercise: What is the worst-case delay that will be faced by this register file? (Hint: Data written in the previous cycle may be read in the current cycle)

$$T_{clk-out} =$$

- Note: This eqn. applies for when you have a flop-based capture of the write address

Critical Delay Analysis



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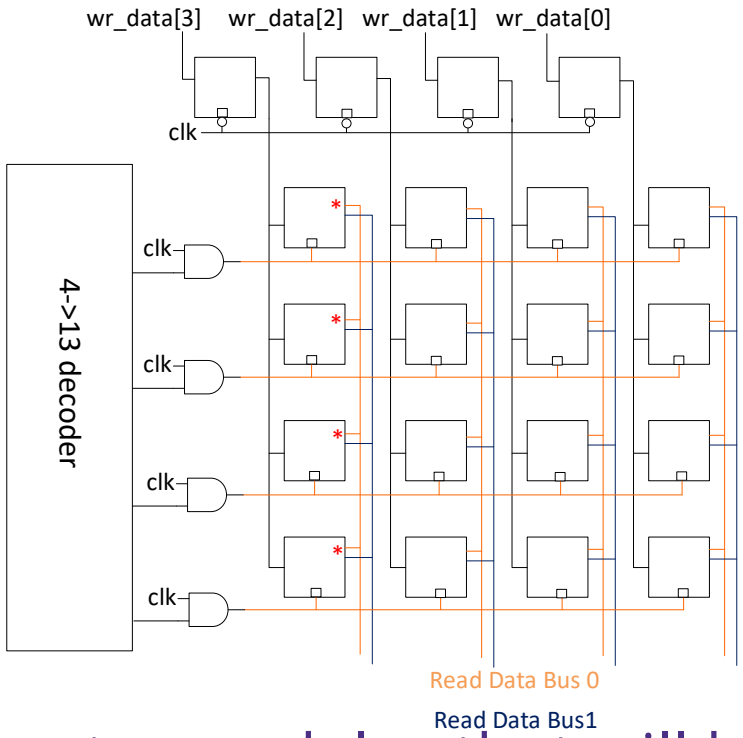

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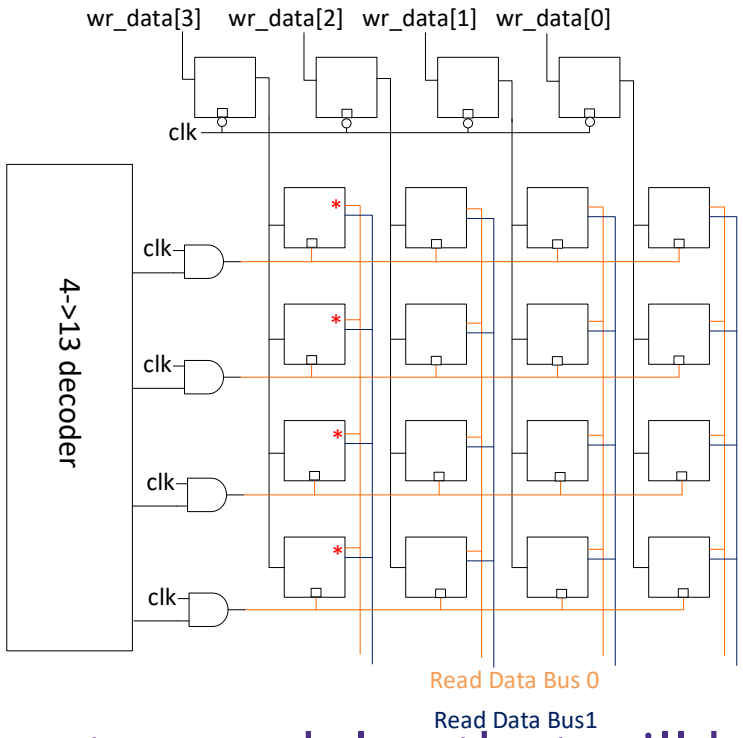

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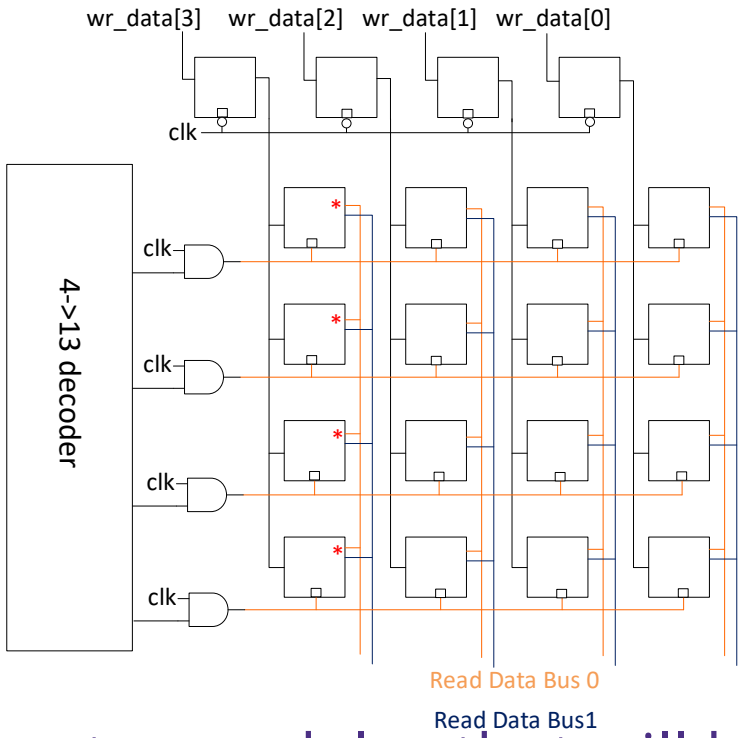

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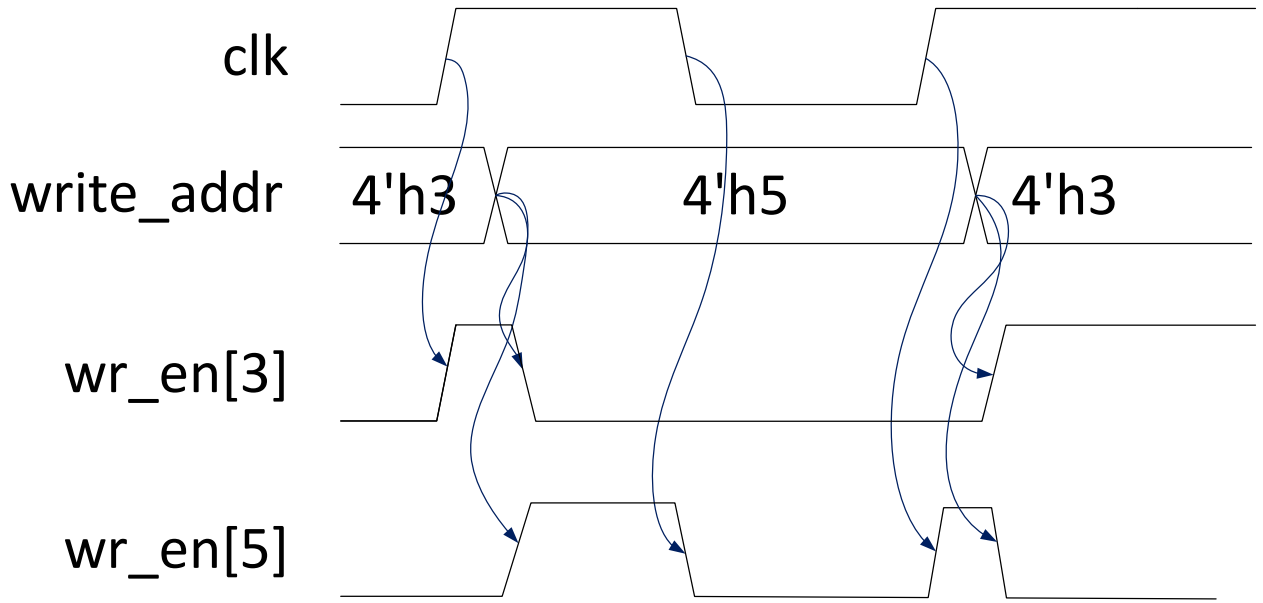
Pitfalls



- 1. What happens when slave clock arrives after master clock
- 2. Writing to multiple slaves due to glitches
  - Common problem if enabling logic is not properly done
- Run timing verification with the right state transitions for worst case delay and slew

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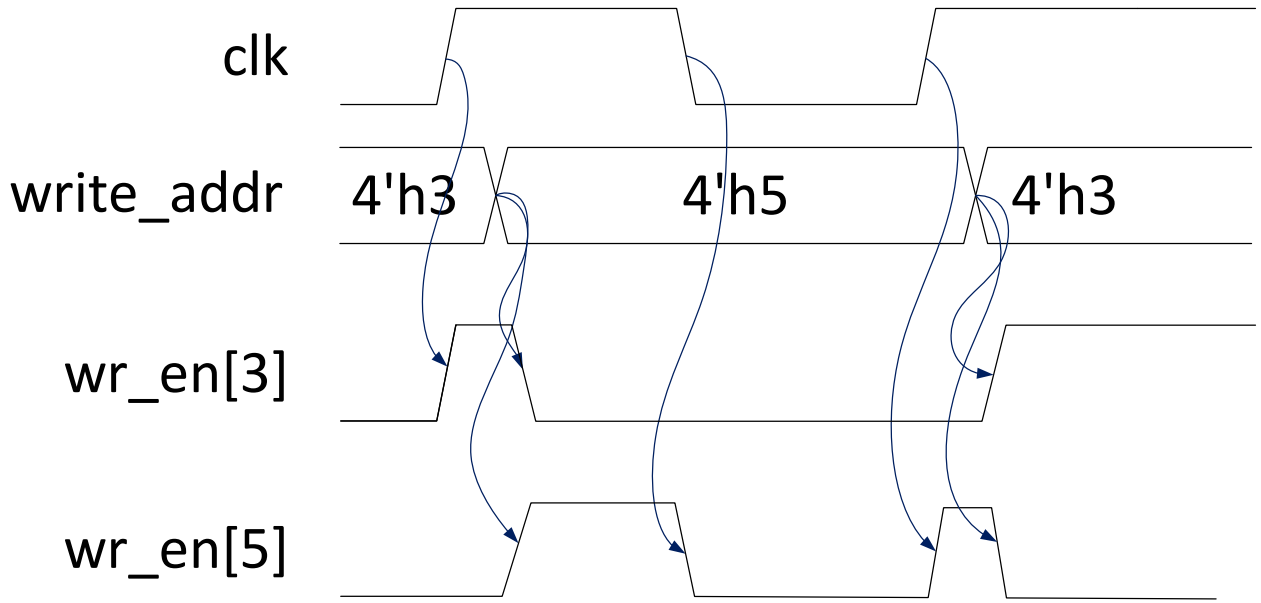

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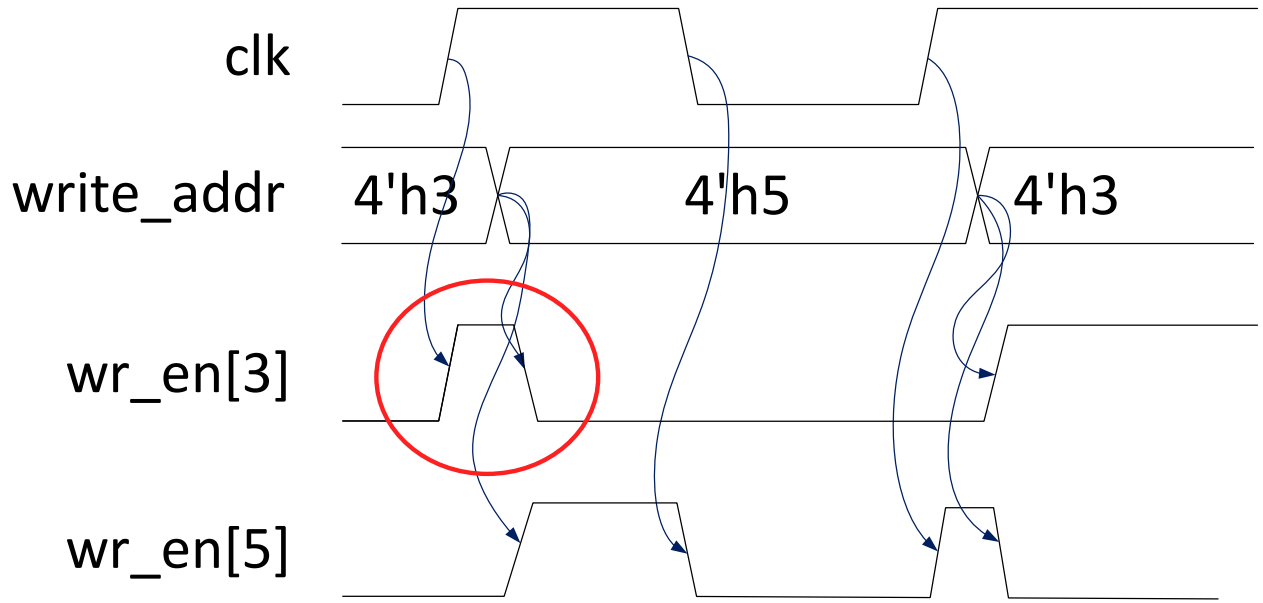

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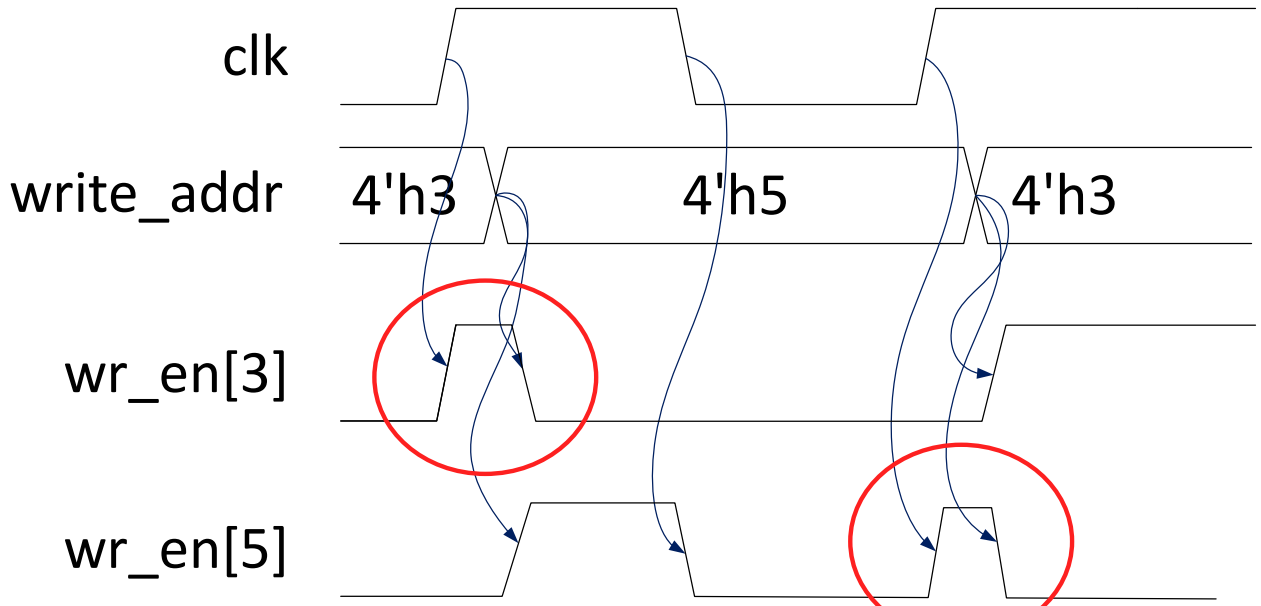

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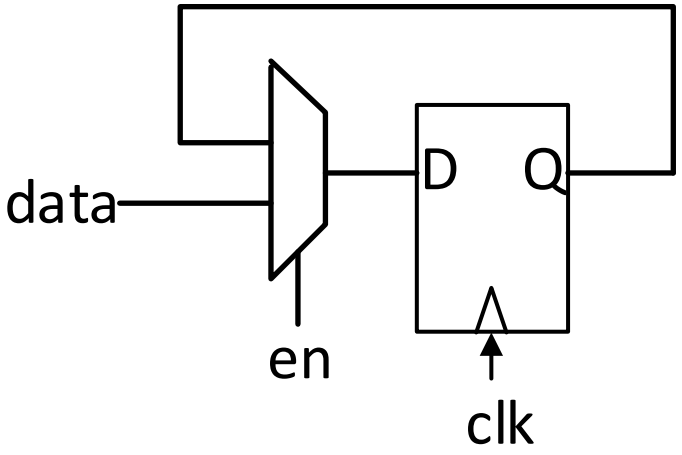
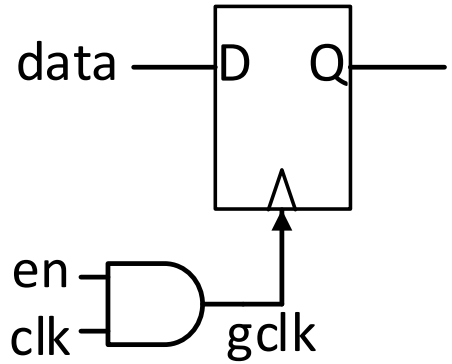
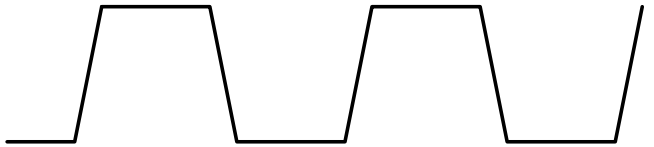
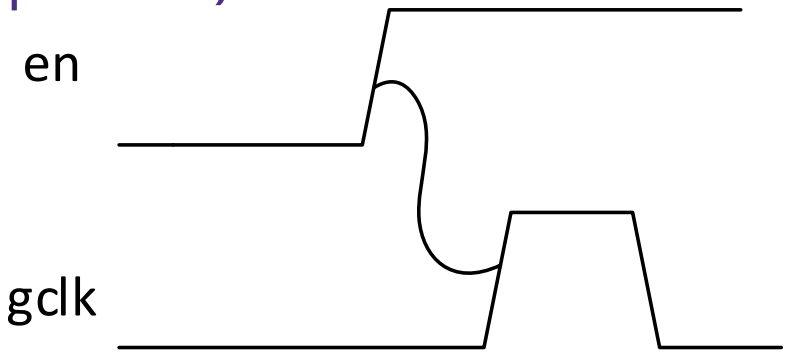
Related Basics : Clock-Gating



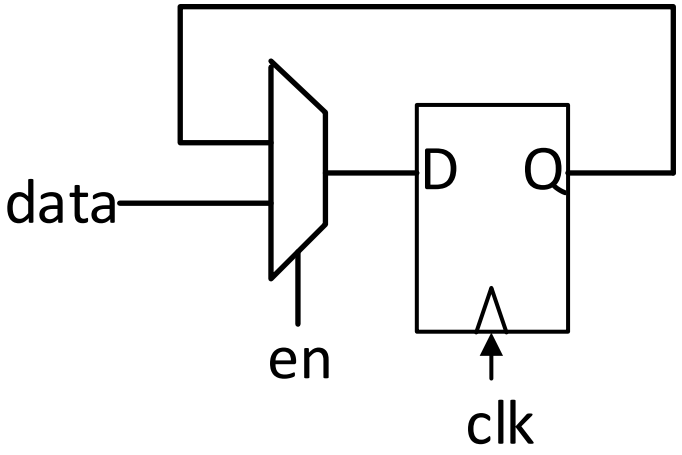


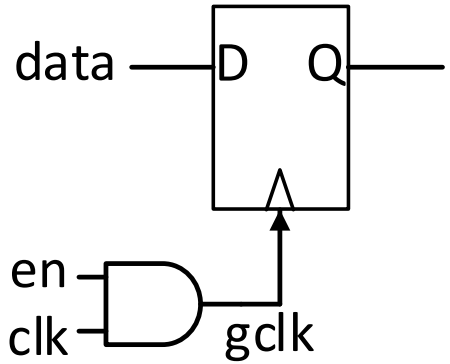


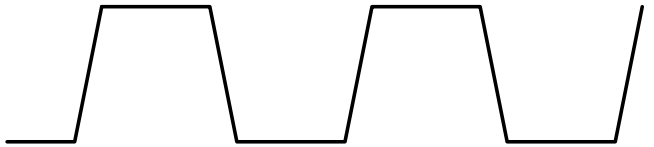
- Conditionally capture state into a timing element (flop/latch)
  - Reduce power



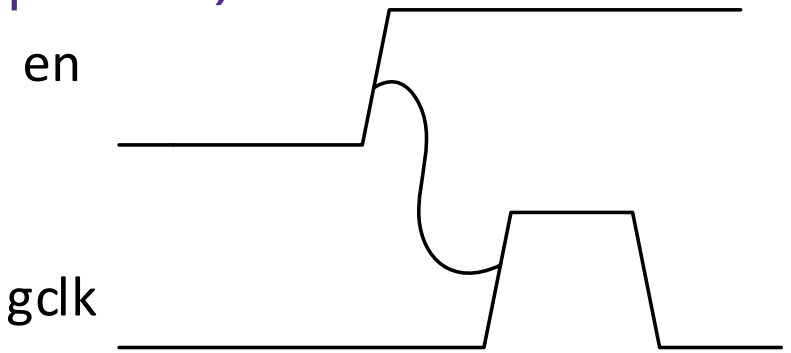
Related Basics : Clock-Gating





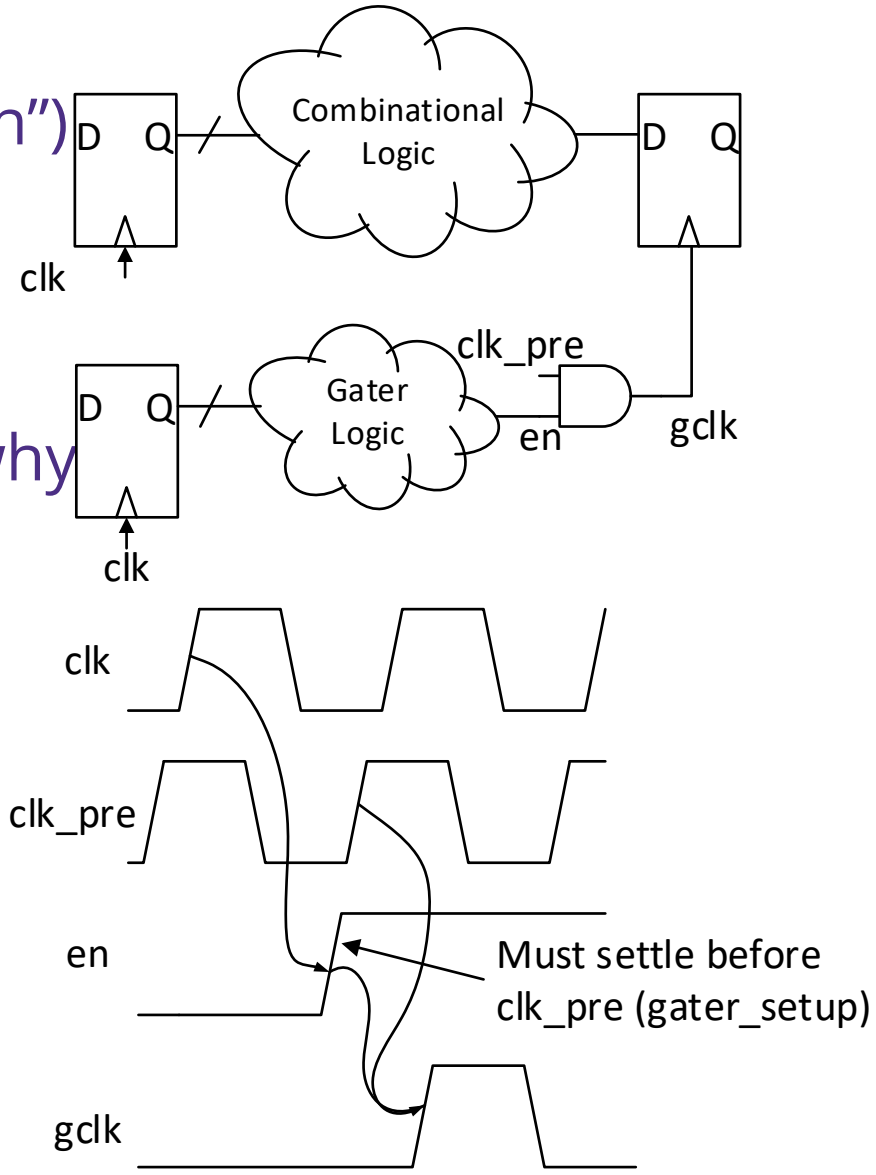


- Conditionally capture state into a timing element (flop/latch)
  - Reduce power
  - Potential area savings (if multiple flops are involved)



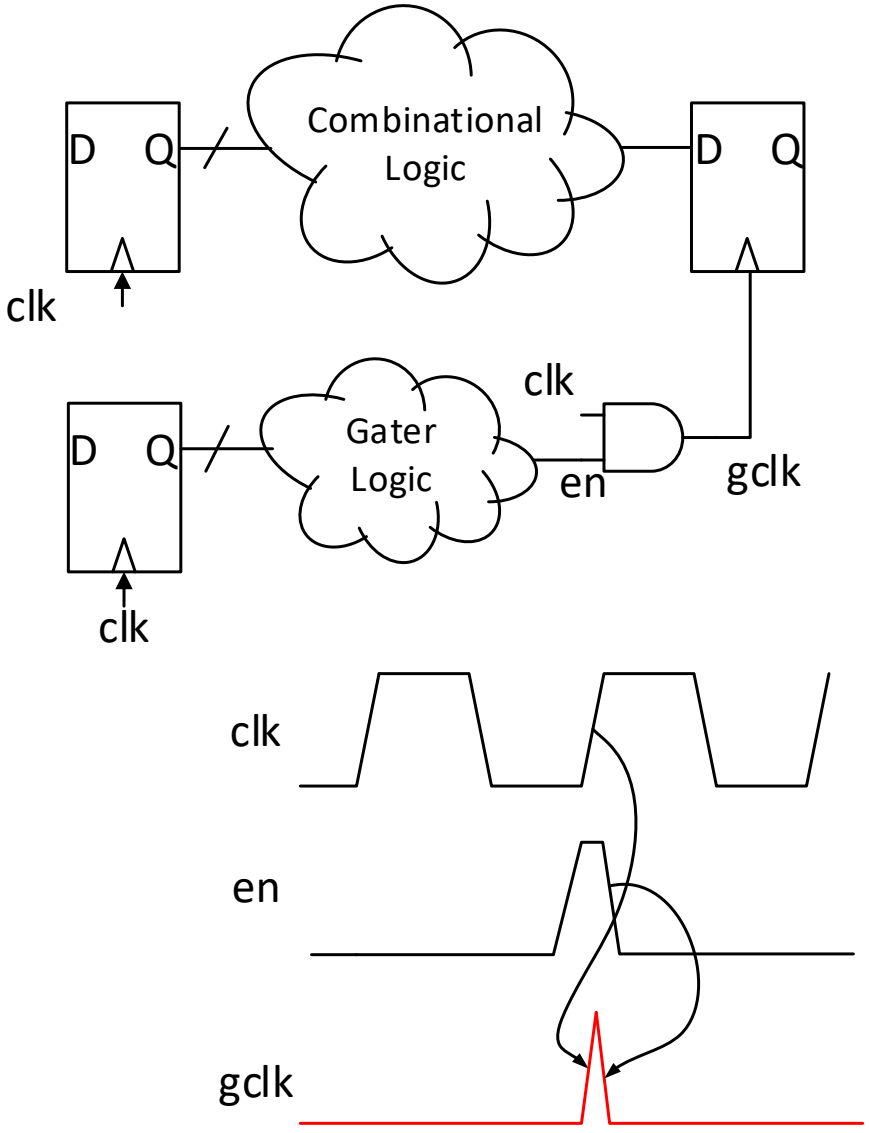
System-context

- State machine generates enable (“en”)
- Clk, or clk_pre gated by en
  - Depends on context
  - You will end up using clk
- En must settle before clk_pre →1 (why
  - Gater-setup



Avoiding Glitches

- En logic cone sees delay spread
- Early “en” evaluation causes runt pulse
  - Early de-assertion causes runt pulse



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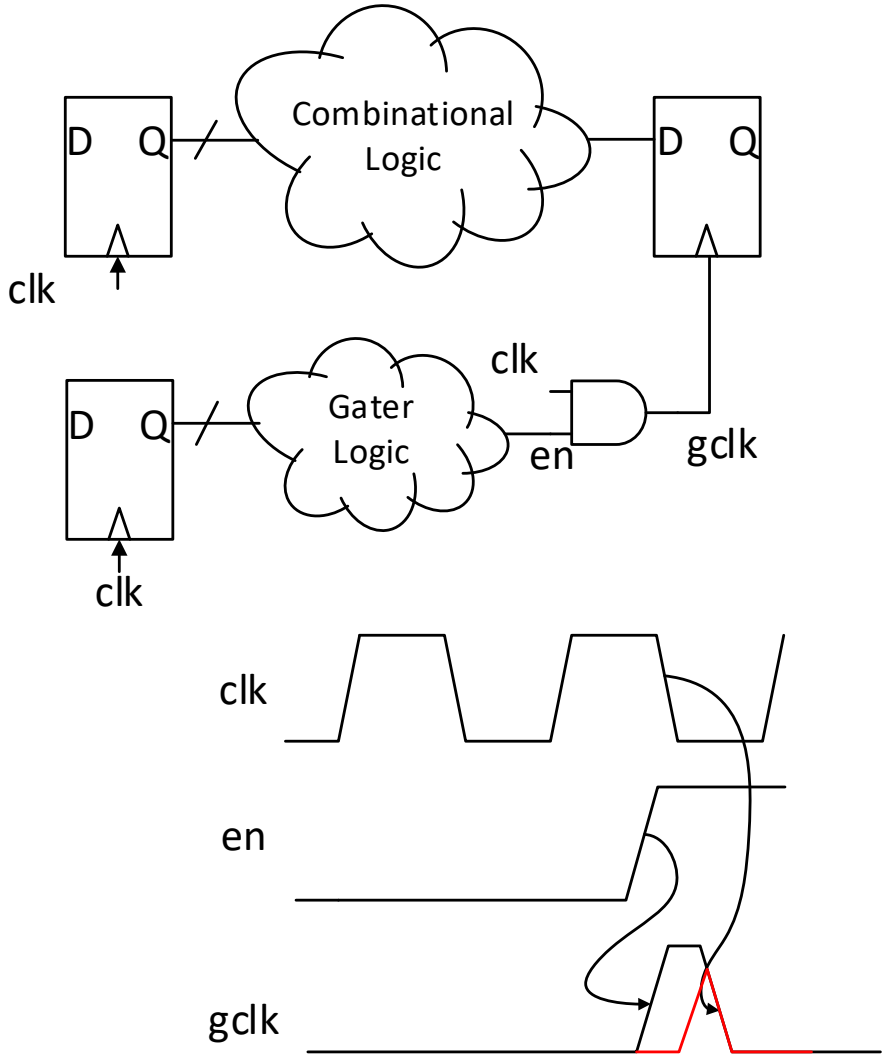
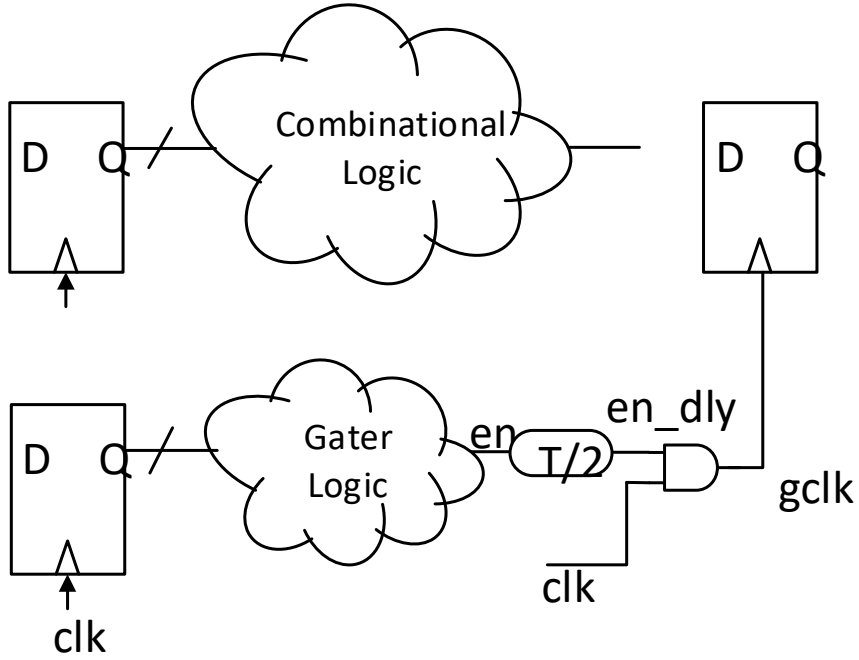
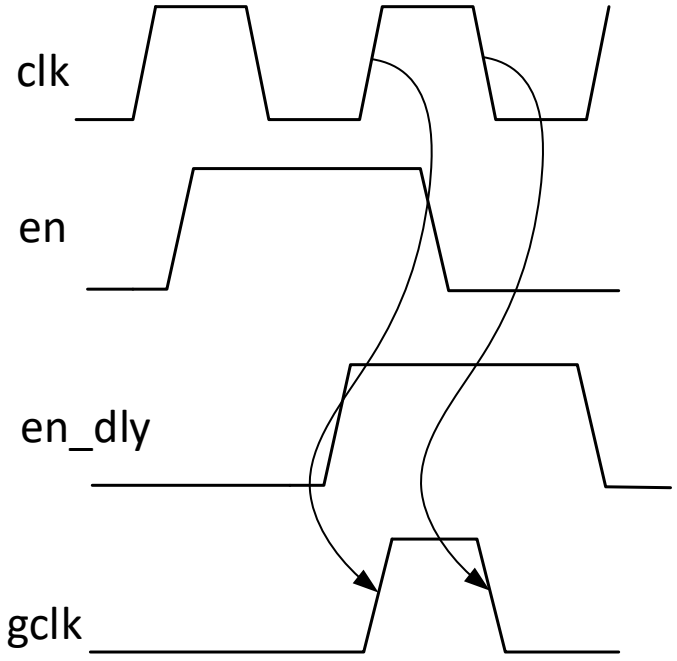


Avoiding Glitches

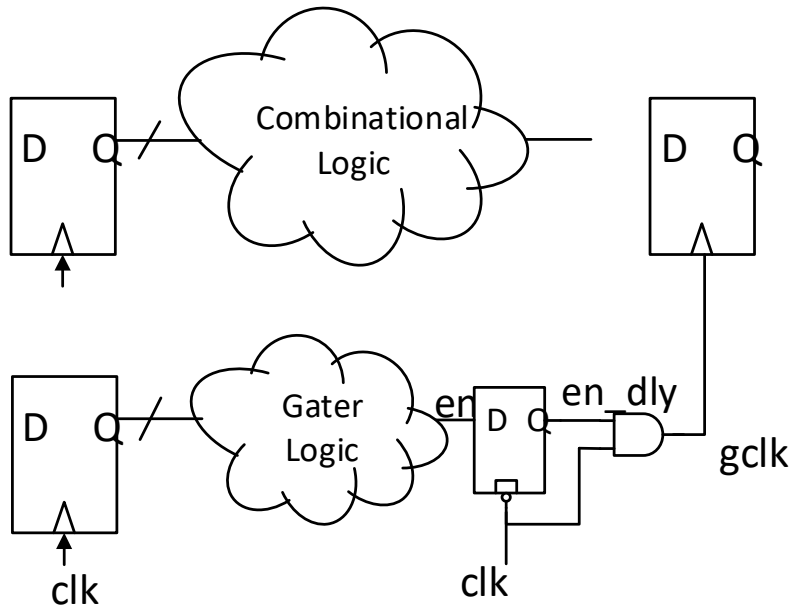
- En logic cone sees delay spread
- Early “en” evaluation causes runt pulse
  - Early de-assertion causes runt pulse
  - Early assertion causes premature clock pulse or premature runt pulse
- Inserting T/2 delay guarantees enable for current cycle will not transition until clk goes low.
- Strictly speaking, correct BUT
  - Adds to critical path!!
  - Dissipative, variable and inefficient

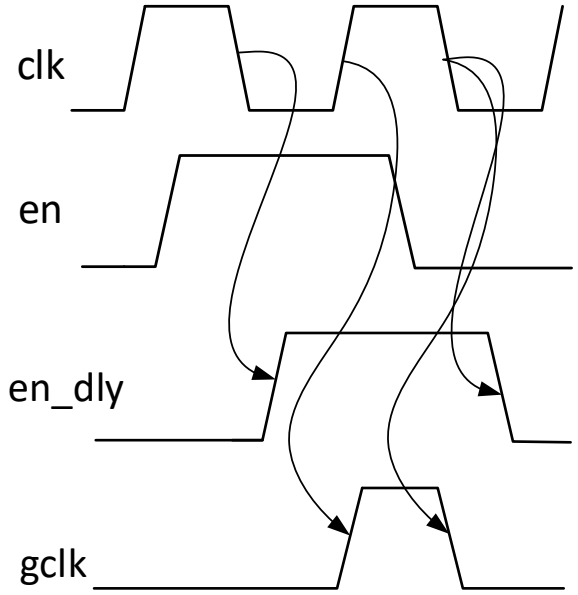




Avoiding Glitches

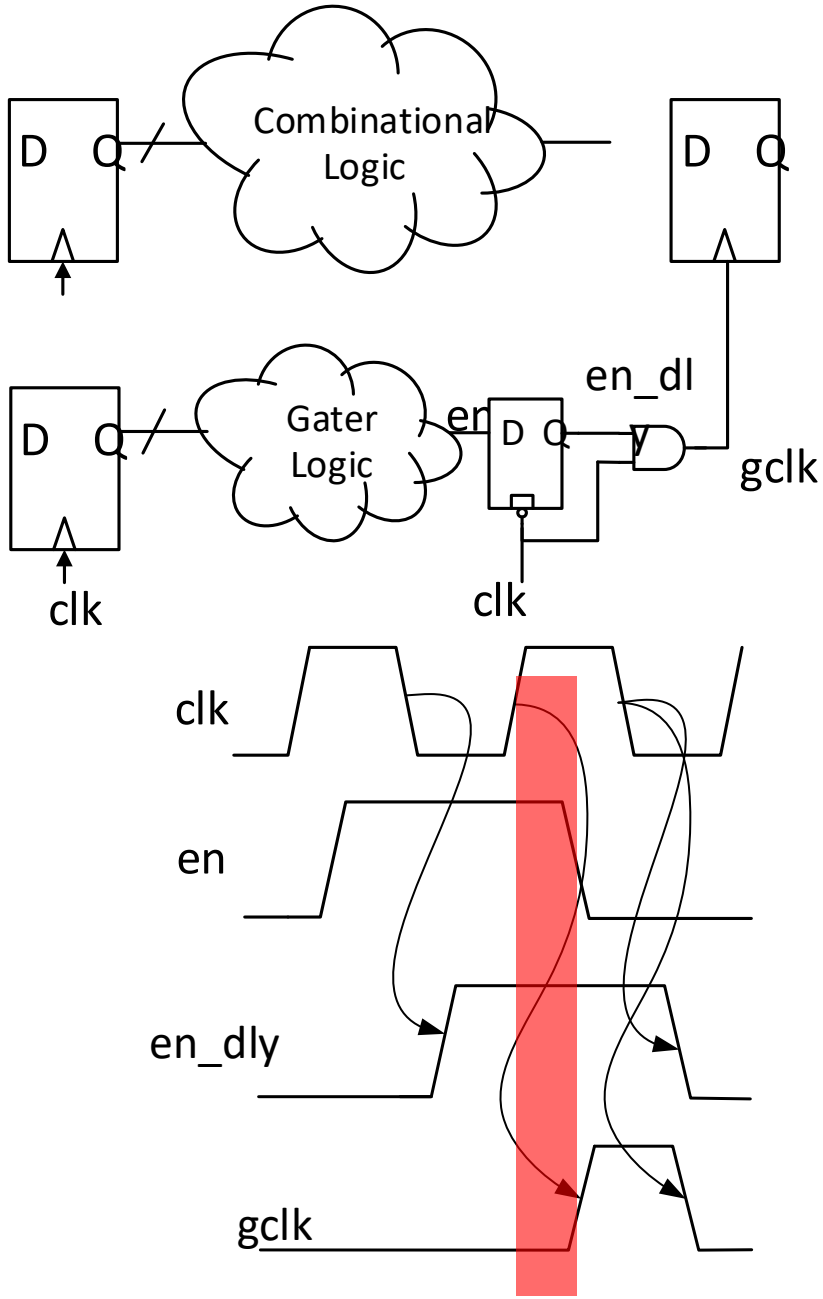
- En logic cone sees delay spread
- Early “en” evaluation causes runt pulse
  - Early de-assertion causes runt pulse
  - Early assertion causes premature clock pulse or premature runt pulse
- Inserting T/2 delay guarantees enable for current cycle will not transition until clk goes low.
- Strictly speaking, correct BUT
  - Adds to critical path!!
  - Dissipative, variable and inefficient
  - (Technically, you could design for min_delay > T/2 instead..)
- Instead,Use a latch!!





One more thing….

- Clock skew often reported for global clocks
  - Good way to control skew BUT most timing elements today are driven by gclk
  - gclk skew ultimately matters!
  - Skew between clk and gclk is important when either launch xor capture is ungated
    - If capture is flip flop



Breakout Session:

- Problem. Construct a clock gating methodology for gating a state machine triggered by the negative edge of the clock.