

EE 458/533 – Power Electronics Controls, Winter 2022
Homework 5

Due Date: Friday March 3rd 2022, 4:30pm PT

Problem 1: Consider a boost converter where the winding and MOSFET on resistances are lumped into R_L . The diagrams in Figs. 1(a) represents the version used in lab and 1(b) represent the improved control structure that implements feed-forward. Below, you will model the improved controller and then compare it to the lab implementation. Assume all sensing path gains are equal to 1 and that a PI controller is used.

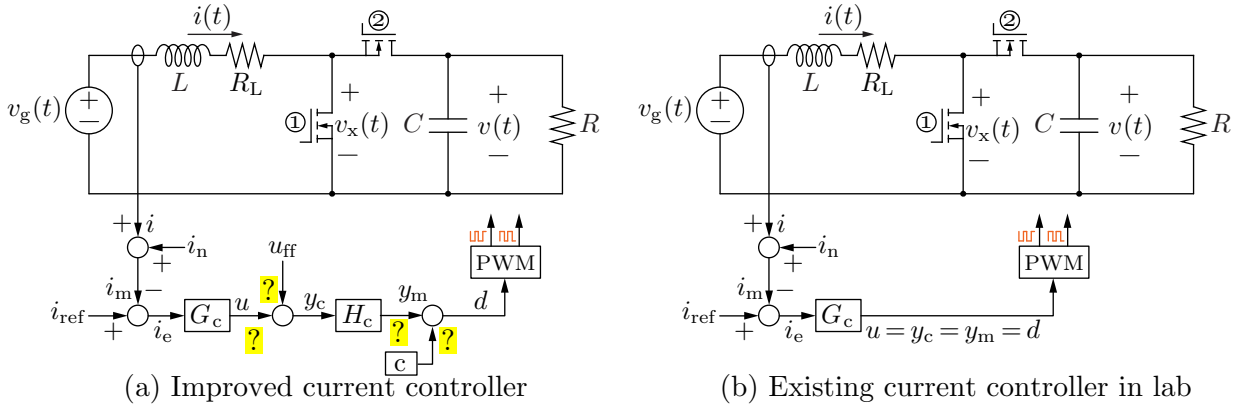


Figure 1: Boost converter with current control.

Perform the following:

- Derive the voltage $\langle v_x(t) \rangle$ (average value of $v_x(t)$ over one switching period) in terms of duty, d and output voltage, $\langle v(t) \rangle$
- Derive the KVL in the input loop to obtain a relationship between $v_g(t) (= \langle v_g(t) \rangle)$, L , R_L , $\langle v_x(t) \rangle$ and $\langle i(t) \rangle$.
- Assume the controller, G_c produces an output, d , verify that Fig. 2 (a) shows the correct implementation of your KVL derived in the preceding question.
- We want to implement feed-forward to simplify our plant transfer function to be only $1/(sL + R_L)$ so that the controller design becomes simplified as in Fig. 2 (c). To achieve this, sketch or provide a numerical relationship between the controller output, u , and the actual duty ratio, d , that goes into the PWM block. (Note under all circumstances, $0 < d < 1$). Compare the relationship you have derived with Fig. 1 (a) to identify H_c , c , u_{ff} and the signs of the sum blocks shown with ?.
- Design your controller, G_c by plant-inversion formulae to achieve a bandwidth of 1/10-th of your switching frequency.

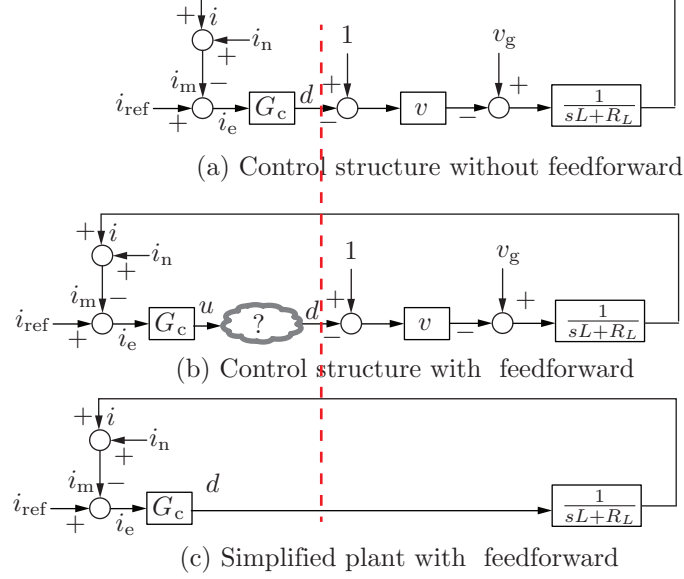


Figure 2: Control simplification with feed-forward. On the right side of the red-dashed line is your plant and on the left is your controller.

Problem 2: Simulate the improved implementation in Problem 1 where $V_g = 24 \text{ V}$, $f_s = 10 \text{ kHz}$, $L = 1.3 \text{ mH}$, $R_L = 60 \text{ m}\Omega$, $C = 250 \mu\text{F}$, and $R = 10 \Omega$. Use the white noise block¹ to simulate the current sensor noise. Initialize your simulation with a reference of $i_{\text{ref}} = 3 \text{ A}$ such that the states are all in steady-state and the error, i_e , is zero. Perform the following:

- Create a switched simulation for improved implementation in Fig. 1 with and without feedforward where the following events below occur. Use your control gains from 2(a). Overlay the $i(t)$ waveforms for with and without feedforward (for without feedforward, use the controller you developed in lab) on the same plot to facilitate comparison. Comment on how they differ.
 - At $t = 5 \text{ ms}$, i_{ref} changes from $3 \text{ A} \rightarrow 5 \text{ A}$.
 - At $t = 10 \text{ ms}$, v_g changes from $24 \text{ V} \rightarrow 28 \text{ V}$.
 - At $t = 15 \text{ ms}$, R changes from $10 \Omega \rightarrow 15 \Omega$.

¹Configure the white noise block with zero mean, 50 mA standard deviation, and sample time T_s .

Problem 3: We will build on the results from Homework 4, Using Faraday's Law, give the expressions for the induced back EMF voltages, e_a , e_b , e_c , within the a, b, and c phase windings, respectively. In addition, draw the equivalent circuits looking into the a, b, and c windings where the back EMFs, e_a , e_b , e_c , are clearly labeled along with the speed-dependent voltages \mathcal{E}_a , \mathcal{E}_b , \mathcal{E}_c and inductive voltage drops. Be careful with all circuit element polarities.