

# STM32L151xD STM32L152xD

Ultralow power ARM-based 32-bit MCU with 384 KB Flash, RTC, LCD, USB, analog functions, 10 serial ports, memory I/F

### **Features**

- Operating conditions
  - Operating power supply range: 1.65 V to 3.6 V (without BOR) or 1.8 V to 3.6 V
- Low power features
  - 7 modes: Sleep, Low-power run (11 μA at 32 kHz) , Low-power sleep (4.4 μA), Stop with RTC, Stop (650 nA), Standby with RTC, Standby (300 nA)
  - Dynamic core voltage scaling down to 233 μA/MHz
  - Ultralow leakage per I/O: 50 nA max
  - Fast wakeup time from Stop: 8 µs
  - Three wakeup pins
- Core: ARM 32-bit Cortex<sup>™</sup>-M3 CPU
  - 32 MHz maximum frequency,
     33.3 DMIPS peak (Dhrystone 2.1)
  - Memory protection unit
- Reset and supply management
  - Low power, ultrasafe BOR (brownout reset)
  - Ultralow power POR/PDR
  - Programmable voltage detector (PVD)
- Clock management
  - 1 to 24 MHz crystal oscillator
  - 32 kHz oscillator for RTC with calibration
  - Internal 16 MHz factory-trimmed RC
  - Internal 37 kHz low consumption RC
  - Internal multispeed low power RC, 65 kHz to 4.2 MHz
  - PLL for CPU clock and USB (48 MHz)
- Memories
  - 384 Kbytes of Flash memory with ECC, split into two banks allowing Read While Write
  - 12 Kbytes of data EEPROM with ECC
  - NVM in 2 banks enabling Read While Write
  - 48 Kbytes of RAM
  - Flexible static memory controller that supports SRAM, PSRAM and NOR Flash





- Low power calendar RTC
  - Alarm, periodic wakeup from Stop/Standby
- Up to 116 fast I/Os (102 of which are 5 V-tolerant)
- DMA: 12-channel DMA controller
- LCD 8 × 40 or 4 × 44 with step-up converter
- 3 operational amplifiers
- 12-bit ADC up to 1 Msps and 40 channels
  - Operational amplifier output, temperature sensor and internal voltage reference
  - Operates down to 1.8 V
- Two 12-bit DACs with output buffers
- Two ultralow power comparators
  - Window mode and wakeup capability
- 11 timers: one 32-bit and six 16-bit generalpurpose timers, two 16-bit basic timers, two watchdog timers (independent and window)
- Up to 12 communication interfaces
  - Up to two I2C interfaces (SMBus/PMBus)
  - Up to five USARTs
  - Up to three SPIs (16 Mbit/s), two with I2S
  - USB 2.0 full-speed interface
  - SDIO interface
- Up to 36 capacitive sensing channels supporting touch, proximity, linear and rotary sensors
- 32-bit CRC calculation unit, 96-bit unique ID

Table 1. Device summary

Reference	Part number
STM32L151xx	STM32L151QD STM32L151RD STM32L151VD STM32L151ZD
STM32L152xx	STM32L152QD STM32L152RD STM32L152VD STM32L152ZD

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### 1 Introduction

This datasheet provides the ordering information and mechanical device characteristics of the STM32L151xD and STM32L152xD ultralow power ARM Cortex<sup>™</sup>-based microcontrollers product line.

The ultralow power STM32L15xxD family includes devices in 4 different package types: from 64 pins to 144 pins. Depending on the device chosen, different sets of peripherals are included, the description below gives an overview of the complete range of peripherals proposed in this family.

These features make the ultralow power STM32L15xxD microcontroller family suitable for a wide range of applications:

- Medical and handheld equipment
- Application control and user interface
- PC peripherals, gaming, GPS and sport equipment
- Alarm systems, wired and wireless sensors, Video intercom
- Utility metering

For information on the Cortex<sup>™</sup>-M3 core please refer to the Cortex<sup>™</sup>-M3 Technical Reference Manual, available from the *www.arm.com* website at the following address: http://infocenter.arm.com/help/index.jsp?topic=/com.arm.doc.ddi0337g.

Figure 1 shows the general block diagram of the device family.

# 2 Description

The ultralow power STM32L15xxD incorporates the connectivity power of the universal serial bus (USB) with the high-performance ARM Cortex <sup>™</sup>-M3 32-bit RISC core operating at a 32 MHz frequency, a memory protection unit (MPU), high-speed embedded memories (Flash memory up to 384 Kbytes and RAM up to 48 Kbytes), a flexible static memory controller (FSMC) interface (for devices with packages of 100 pins and more) and an extensive range of enhanced I/Os and peripherals connected to two APB buses.

All devices offer three operational amplifiers, one 12-bit ADC, two DACs, two ultralow power comparators, one general-purpose 32-bit timer, six general-purpose 16-bit timers and two basic timers, which can be used as time bases.

Moreover, the STM32L15xxD devices contain standard and advanced communication interfaces: up to two I<sup>2</sup>Cs, three SPIs, two I2S, one SDIO, three USARTs, two UARTs and a USB. Up to 36 channels are available for capacitive sensing directly driven through GPIOs and general purpose timers.

They also include a real-time clock and a set of backup registers that remain powered in Standby mode.

Finally, the integrated LCD controller has a built-in LCD voltage generator that allows you to drive up to 8 multiplexed LCDs with contrast independent of the supply voltage.

The ultralow power STM32L15xxD operates from a 1.8 to 3.6 V power supply (down to 1.65 V at power down) with BOR and from a 1.65 to 3.6 V power supply without BOR option. It is available in the -40 to +85 °C temperature range. A comprehensive set of power-saving modes allows the design of low-power applications.







# 2.1 Device overview

Table 2. Ultralow power STM32L15xxD device features and peripheral counts

	pheral	STM32L15xRx	STM32L15xVx	STM32L15xZx STM32L15xQx		
Flash - Kbytes		384	384	384		
Data EEPROM		12	12	12		
RAM - Kbytes		48	48	48		
FSMC		No	multiplexed only	Yes		
	32 bit	1	1	1		
Timers	General-purpose	6	6	6		
	Basic	2	2	2		
	SPI/(I2S)	3/(2)	3/(2)	3/(2)		
	I <sup>2</sup> C	2	2	2		
Communication interfaces	USART	5	5	5		
	USB	1	1	1		
	SDIO	1	1	1		
GPIOs		51	83	115 <sup>(1)</sup>		
Operation amplifi	ers	3	3	3		
12-bit synchroniz Number of chann		1 21	1 25	1 40		
12-bit DAC Number of chann	els	2 2	2 2	2 2		
LCD <sup>(2)</sup> COM x SEG		1 4x32 or 8x28	1 4x44 or 8x40	1 4x44 or 8x40		
Comparators		2	2	2		
Capacitive sensir		30/10	30/10	36/11		
CPU frequency		32 MHz				
Operating voltage	•	1.8 V to 3.6 V (down to 1.65 V at power-down) with BOR option 1.65 V to 3.6 V without BOR option				
Operating temper	atures	Ambient temperature: -40 to +85 °C Junction temperature: -40 to + 105 °C				
Packages		LQFP64				

<sup>1. 109</sup> GPIOs in BGA132 package.

<sup>2.</sup> STM32L152xx devices only.

### 2.2 Ultralow power device continuum

The ultralow power STM32L151xD and STM32L152xD are fully pin-to-pin, software and feature compatible. Besides the full compatibility within the family, the devices are part of STMicroelectronics microcontrollers ultralow power strategy which also includes STM8L101xx and STM8L15xx devices. The STM8L and STM32L families allow a continuum of performance, peripherals, system architecture and features.

They are all based on STMicroelectronics 0.13 µm ultralow leakage process.

Note:

The ultralow power STM32L and general-purpose STM32Fxxxx families are pin-to-pin compatible. The STM8L15xxx devices are pin-to-pin compatible with the STM8L101xx devices. Please refer to the STM32F and STM8L documentation for more information on these devices.

### 2.2.1 Performance

All families incorporate highly energy-efficient cores with both Harvard architecture and pipelined execution: advanced STM8 core for STM8L families and ARM Cortex<sup>™</sup>-M3 core for STM32L family. In addition specific care for the design architecture has been taken to optimize the mA/DMIPS and mA/MHz ratios.

This allows the ultralow power performance to range from 5 up to 33.3 DMIPs.

### 2.2.2 Shared peripherals

STM8L15xxx and STM32L15xxx share identical peripherals which ensure a very easy migration from one family to another:

- Analog peripherals: ADC, DAC, and comparators
- Digital peripherals: RTC and some communication interfaces

### 2.2.3 Common system strategy

To offer flexibility and optimize performance, the STM8L15xxx and STM32L15xxx families use a common architecture:

- Same power supply range from 1.65 V to 3.6 V.
- Architecture optimized to reach ultralow consumption both in low power modes and Run mode
- Fast startup strategy from low power modes
- Flexible system clock
- Ultrasafe reset: same reset strategy including power-on reset, power-down reset, brownout reset and programmable voltage detector.

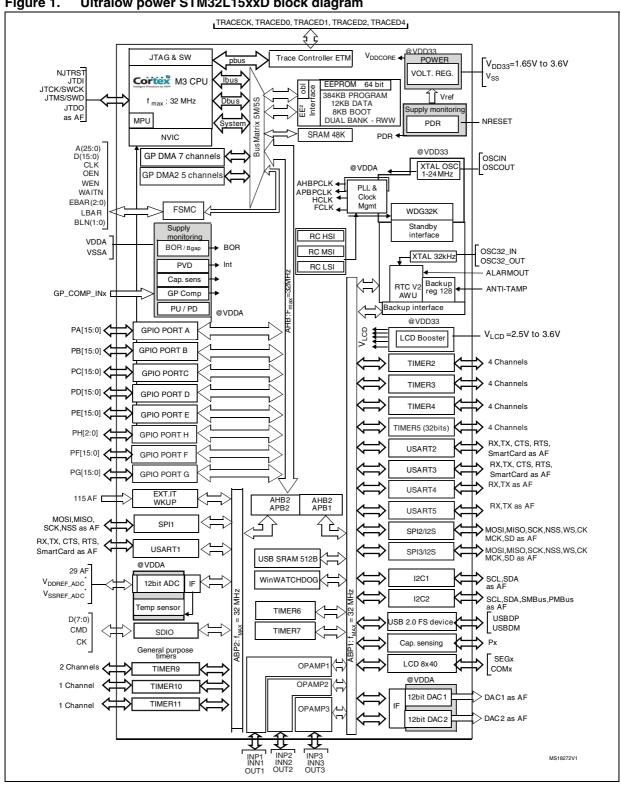
#### 2.2.4 Features

ST ultralow power continuum also lies in feature compatibility:

- More than 10 packages with pin count from 20 to 144 pins and size down to 3 x 3 mm
- Memory density ranging from 4 to 384 Kbytes

#### 3 **Functional overview**

Figure 1. Ultralow power STM32L15xxD block diagram



1. Legend:

AF: alternate function

ADC: analog-to-digital converter BOR: brown out reset DMA: direct memory access

DAC: digital-to-analog converter I<sup>2</sup>C: inter-integrated circuit multimaster interface

# 3.1 Low power modes

The ultralow power STM32L15xxD supports dynamic voltage scaling to optimize its power consumption in run mode. The voltage from the internal low-drop regulator that supplies the logic can be adjusted according to the system's maximum operating frequency and the external voltage supply.

There are three power consumption ranges:

- Range 1 (V<sub>DD</sub> range limited to 2.0-3.6 V), with the CPU running at up to 32 MHz
- Range 2 (full V<sub>DD</sub> range), with a maximum CPU frequency of 16 MHz
- Range 3 (full V<sub>DD</sub> range), with a maximum CPU frequency limited to 4 MHz (generated only with the multispeed internal RC oscillator clock source).

Seven low power modes are provided to achieve the best compromise between low power consumption, short startup time and available wakeup sources:

### Sleep mode

In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs. The Sleep mode power consumption at 16 MHz is of about 1 mA with all peripherals off.

### • Low power run mode

This mode is achieved with the multispeed internal (MSI) RC oscillator set to the minimum clock (128 kHz), execution from SRAM or Flash memory, and internal regulator in low power mode to minimize the regulator's operating current. In the Low power run mode, the clock frequency and the number of enabled peripherals are both limited.

### Low power sleep mode

This mode is achieved by entering the Sleep mode with the internal voltage regulator in Low power mode to minimize the regulator's operating current. In the Low power sleep mode, both the clock frequency and the number of enabled peripherals are limited; a typical example would be to have a timer running at 32 kHz.

When wakeup is triggered by an event or an interrupt, the system reverts to the run mode with the regulator on.

#### Stop mode with RTC

This Stop mode achieves the lowest power consumption while retaining the RAM and register contents and real time clock. All clocks in the  $V_{CORE}$  domain are stopped, the PLL, MSI RC, HSI RC and HSE crystal oscillators are disabled. The LSE or LSI is still running. The voltage regulator is in the low power mode.

The device can be woken up from the Stop mode by any of the EXTI line, in 8 µs. The EXTI line source can be one of the 16 external lines. It can be the PVD output, the Comparator 1 event or Comparator 2 event (if internal reference voltage is on), It can be, the RTC alarm(s), the USB wakeup, the RTC tamper events, the RTC timestamp event, the RTC Wakeup.

### Stop mode without RTC

The Stop mode achieves the lowest power consumption while retaining the RAM and register contents. All clocks are stopped, the PLL, MSI RC, HSI and LSI RC, LSE and HSE crystal oscillators are disabled. The voltage regulator is in the low power mode. The device can be woken up from the Stop mode by any of the EXTI line, in 8  $\mu$ s. The EXTI line source can be one of the 16 external lines. It can be the PVD output, the Comparator 1 event or Comparator 2 event (if internal reference voltage is on). It can also be waken by the USB wakeup.

### Standby mode with RTC

This Standby mode is used to achieve the lowest power consumption and real time clock. The internal voltage regulator is switched off so that the entire  $V_{CORE}$  domain is powered off. The PLL, MSI RC, HSI RC and HSE crystal oscillators are also switched off. The LSE or LSI is still running. After entering Standby mode, the RAM and register contents are lost except for registers in the Standby circuitry (wakeup logic, IWDG, RTC, LSI, LSE Crystal 32K osc, RCC\_CSR).

The device exits the Standby mode in 60 µs when an external reset (NRST pin), an IWDG reset, a rising edge on one of the three WKUP pins, RTC alarm (Alarm A or Alarm B), RTC tamper event, RTC timestamp event or RTC Wakeup event occurs.

#### Standby mode without RTC

The Standby mode is used to achieve the lowest power consumption. The internal voltage regulator is switched off so that the entire  $V_{CORE}$  domain is powered off. The PLL, MSI RC, HSI and LSI RC, HSE and LSE crystal oscillators are also switched off. After entering Standby mode, the RAM and register contents are lost except for registers in the Standby circuitry (wakeup logic, IWDG, RTC, LSI, LSE Crystal 32K osc, RCC\_CSR).

The device exits the Standby mode in  $60 \mu s$  when an external reset (NRST pin) or a rising edge on one of the three WKUP pin occurs.

Note:

The RTC, the IWDG, and the corresponding clock sources are not stopped automatically by entering the Stop or Standby mode.

# 3.2 ARM<sup>®</sup> Cortex<sup>™</sup>-M3 core with MPU

The ARM Cortex<sup>TM</sup>-M3 processor is the industry leading processor for embedded systems. It has been developed to provide a low-cost platform that meets the needs of MCU implementation, with a reduced pin count and low-power consumption, while delivering outstanding computational performance and an advanced system response to interrupts.

The ARM Cortex<sup>™</sup>-M3 32-bit RISC processor features exceptional code-efficiency, delivering the high-performance expected from an ARM core in the memory size usually associated with 8- and 16-bit devices.

The memory protection unit (MPU) improves system reliability by defining the memory attributes (such as read/write access permissions) for different memory regions. It provides up to eight different regions and an optional predefined background region.

Owing to its embedded ARM core, the STM32L15xxD is compatible with all ARM tools and software.

### **Nested vectored interrupt controller (NVIC)**

The ultralow power STM32L15xxD embeds a nested vectored interrupt controller able to handle up to 56 maskable interrupt channels (not including the 16 interrupt lines of Cortex™-M3) and 16 priority levels.

- Closely coupled NVIC gives low-latency interrupt processing
- Interrupt entry vector table address passed directly to the core
- Closely coupled NVIC core interface
- Allows early processing of interrupts
- Processing of *late arriving*, higher-priority interrupts
- Support for tail-chaining
- Processor state automatically saved
- Interrupt entry restored on interrupt exit with no instruction overhead

This hardware block provides flexible interrupt management features with minimal interrupt latency.

# 3.3 Reset and supply management

### 3.3.1 Power supply schemes

- $V_{DD} = 1.65$  to 3.6 V: external power supply for I/Os and the internal regulator. Provided externally through  $V_{DD}$  pins.
- V<sub>SSA</sub>, V<sub>DDA</sub> = 1.65 to 3.6 V: external analog power supplies for ADC, reset blocks, RCs and PLL (minimum voltage to be applied to V<sub>DDA</sub> is 1.8 V when the ADC is used).
   V<sub>DDA</sub> and V<sub>SSA</sub> must be connected to V<sub>DD</sub> and V<sub>SS</sub>, respectively.

### 3.3.2 Power supply supervisor

The device has an integrated ZEROPOWER power-on reset (POR)/power-down reset (PDR) that can be coupled with a brownout reset (BOR) circuitry.

The device exists in two versions:

- The version with BOR activated at power-on operates between 1.8 V and 3.6 V.
- The other version without BOR at power up operates between 1.65V and 3.6V.

As the BOR can be activated and deactivated at run time, this distinction is important only for power-up phase.

When BOR is active at power-on, it ensures proper operation starting from 1.8 V whatever the power ramp-up phase before it reaches 1.8 V. When BOR is not active at power-up, the power ramp-up should guarantee that 1.65 V is reached on  $V_{DD}$  at least 1 ms after it exits the POR area.

After the  $V_{DD}$  threshold is reached (1.65 V or 1.8 V depending on the BOR which is active or not at power-on), the option byte loading process starts, either to confirm or modify default thresholds, or to disable BOR permanently: in this case, the  $V_{DD}$  min value at power down is 1.65 V.

Five BOR thresholds are available through option bytes, starting from 1.8 V to 3 V. To reduce the power consumption in Stop mode, it is possible to automatically switch off the internal reference voltage ( $V_{REFINT}$ ) in Stop mode. The device remains in reset mode when  $V_{DD}$  is below a specified threshold,  $V_{POR/PDR}$  or  $V_{BOR}$ , without the need for any external reset circuit.

Note:

The start-up time at power-on is typically 3.3 ms when BOR is active at power-up, the start-up time at power-on can be decreased down to 1 ms typically for devices with BOR inactive at power-up.

The device features an embedded programmable voltage detector (PVD) that monitors the  $V_{DD}/V_{DDA}$  power supply and compares it to the  $V_{PVD}$  threshold. This PVD offers 7 different levels between 1.85 V and 3.05 V, chosen by software, with a step around 200 mV. An interrupt can be generated when  $V_{DD}/V_{DDA}$  drops below the  $V_{PVD}$  threshold and/or when  $V_{DD}/V_{DDA}$  is higher than the  $V_{PVD}$  threshold. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state. The PVD is enabled by software.

### 3.3.3 Voltage regulator

The regulator has three operation modes: main (MR), low power (LPR) and power down.

- MR is used in Run mode (nominal regulation)
- LPR is used in the Low power run, Low power sleep and Stop modes
- Power down is used in Standby mode. The regulator output is high impedance, the kernel circuitry is powered down, inducing zero consumption but the contents of the registers and RAM are lost except for the standby circuitry (wakeup logic, IWDG, RTC, LSI, LSE crystal 32K osc, RCC CSR).

#### 3.3.4 Boot modes

At startup, boot pins are used to select one of three boot options:

- Boot from Flash memory
- Boot from System memory
- Boot from embedded RAM

The boot from Flash usually boots at the beginning of the Flash (bank 1). An additional boot mechanism is available through user option byte, to allow booting from bank 2 when bank 2 contains valid code. This dual boot capability can be used to easily implement a secure field software update mechanism.

The boot loader is located in System memory. It is used to reprogram the Flash memory by using USART1, USART2 or USB.

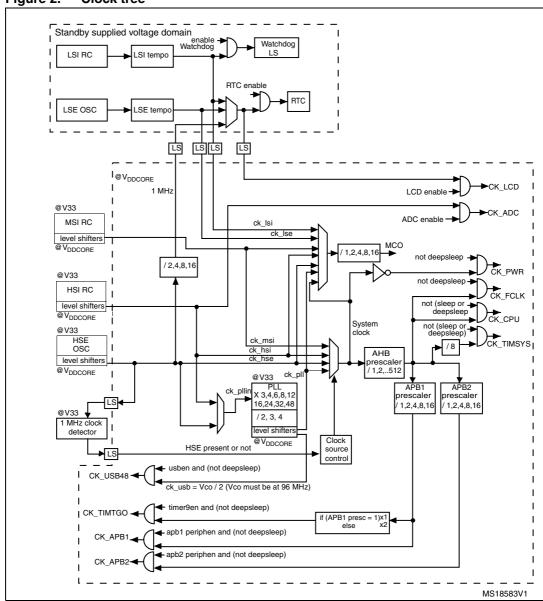
# 3.4 Clock management

The clock controller distributes the clocks coming from different oscillators to the core and the peripherals. It also manages clock gating for low power modes and ensures clock robustness. It features:

- Clock prescaler: to get the best trade-off between speed and current consumption, the clock frequency to the CPU and peripherals can be adjusted by a programmable prescaler
- Safe clock switching: clock sources can be changed safely on the fly in run mode through a configuration register.
- **Clock management**: to reduce power consumption, the clock controller can stop the clock to the core, individual peripherals or memory.
- System clock source: three different clock sources can be used to drive the master clock SYSCLK:
  - 1-24 MHz high-speed external crystal (HSE), that can supply a PLL
  - 16 MHz high-speed internal RC oscillator (HSI), trimmable by software, that can supply a PLL
  - Multispeed internal RC oscillator (MSI), trimmable by software, able to generate 7 frequencies (65 kHz, 131 kHz, 262 kHz, 524 kHz, 1.05 MHz, 2.1 MHz, 4.2 MHz).
     When a 32.768 kHz clock source is available in the system (LSE), the MSI frequency can be trimmed by software down to a ±0.5% accuracy.
- Auxiliary clock source: two ultralow power clock sources that can be used to drive the LCD controller and the real-time clock:
  - 32.768 kHz low-speed external crystal (LSE)
  - 37 kHz low-speed internal RC (LSI), also used to drive the independent watchdog.
     The LSI clock can be measured using the high-speed internal RC oscillator for greater precision.
- RTC and LCD clock sources: the LSI, LSE or HSE sources can be chosen to clock the RTC and the LCD, whatever the system clock.
- USB clock source: the embedded PLL has a dedicated 48 MHz clock output to supply the USB interface.
- Startup clock: after reset, the microcontroller restarts by default with an internal 2 MHz clock (MSI). The prescaler ratio and clock source can be changed by the application program as soon as the code execution starts.
- Clock security system (CSS): this feature can be enabled by software. If a HSE clock failure occurs, the master clock is automatically switched to HSI and a software interrupt is generated if enabled.
- Clock-out capability (MCO: microcontroller clock output): it outputs one of the internal clocks for external use by the application.

Several prescalers allow the configuration of the AHB frequency, each APB (APB1 and APB2) domains. The maximum frequency of the AHB and the APB domains is 32 MHz. See *Figure 2* for details on the clock tree.

Figure 2. Clock tree



For the USB function to be available, both HSE and PLL must be enabled, with the CPU running at either 24 MHz or 32 MHz.

## 3.5 Low power real-time clock and backup registers

The real-time clock (RTC) is an independent BCD timer/counter. Dedicated registers contain the sub-second, second, minute, hour (12/24 hour), week day, date, month, year, in BCD (binary-coded decimal) format. Correction for 28, 29 (leap year), 30, and 31 day of the month are made automatically. The RTC provides two programmable alarms and programmable periodic interrupts with wakeup from Stop and Standby modes.

The programmable wakeup time ranges from 120 µs to 36 hours.

The RTC can be calibrated with an external 512 Hz output, and a digital compensation circuit helps reduce drift due to crystal deviation.

The RTC can also be automatically corrected with a 50/60Hz stable powerline.

The RTC calendar can be updated on the fly down to sub second precision, which enables network system synchronisation.

A time stamp can record an external event occurrence, and generates an interrupt.

There are thirty-two 32-bit backup registers provided to store 128 bytes of user application data. They are cleared in case of tamper detection.

Three pins can be used to detect tamper events. A change on one of these pins can reset backup register and generate an interrupt. To prevent false tamper event, like ESD event, these three tamper inputs can be digitally filtered.

# 3.6 GPIOs (general-purpose inputs/outputs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain), as input (with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions, and can be individually remapped using dedicated AFIO registers. All GPIOs are high-current-capable except for analog pins. The alternate function configuration of I/Os can be locked if needed following a specific sequence in order to avoid spurious writing to the I/O registers. The I/O controller is connected to the AHB with a toggling speed of up to 16 MHz.

### External interrupt/event controller (EXTI)

The external interrupt/event controller consists of 23 edge detector lines used to generate interrupt/event requests. Each line can be individually configured to select the trigger event (rising edge, falling edge, both) and can be masked independently. A pending register maintains the status of the interrupt requests. The EXTI can detect an external line with a pulse width shorter than the Internal APB2 clock period. Up to 115 GPIOs can be connected to the 16 external interrupt lines. The 7 other lines are connected to RTC, PVD, USB or Comparator events.

### 3.7 Memories

The STM32L15xxD devices have the following features:

- 48 Kbyte of embedded RAM accessed (read/write) at CPU clock speed with 0 wait states. With the enhanced bus matrix, operating the RAM does not lead to any performance penalty during accesses to the system bus (AHB and APB buses).
- The non-volatile memory is divided into three arrays:
  - 384 Kbyte of embedded Flash program memory
  - 12 Kbyte of data EEPROM
  - Options bytes

Flash program and data EEPROM are divided in two banks, this enables writing in one bank while running code or reading data in the other bank.

The options bytes are used to write-protect the memory (with 4 KB granularity) and/or readout-protect the whole memory with the following options:

- Level 0: no readout protection
- Level 1: memory readout protection, the Flash memory cannot be read from or written to if either debug features are connected or boot in RAM is selected
- Level 2: chip readout protection, debug features (Cortex-M3 JTAG and serial wire) and boot in RAM selection disabled (JTAG fuse)

The whole non-volatile memory embeds the error correction code (ECC) feature.

# 3.8 FSMC (flexible static memory controller)

The FSMC supports the following modes: SRAM, PSRAM, NOR Flash.

Functionality overview:

- Up to 26 bit address bus
- Up to 16-bit data bus
- Write FIFO
- Burst mode
- Code execution from external memory
- Four chip select signals
- Up to 32 MHz external access (TBC)

# 3.9 DMA (direct memory access)

The flexible 12-channel, general-purpose DMA is able to manage memory-to-memory, peripheral-to-memory and memory-to-peripheral transfers. The DMA controller supports circular buffer management, avoiding the generation of interrupts when the controller reaches the end of the buffer.

Each channel is connected to dedicated hardware DMA requests, with software trigger support for each channel. Configuration is done by software and transfer sizes between source and destination are independent.

The DMA can be used with the main peripherals: SPI,  $I^2C$ , USART, SDIO, general-purpose timers, DAC and ADC.

## 3.10 LCD (liquid crystal display)

The LCD drives up to 8 common terminals and 44 segment terminals to drive up to 320 pixels.

- Internal step-up converter to guarantee functionality and contrast control irrespective of V<sub>DD</sub>. This converter can be deactivated, in which case the V<sub>LCD</sub> pin is used to provide the voltage to the LCD
- Supports static, 1/2, 1/3, 1/4 and 1/8 duty
- Supports static, 1/2, 1/3 and 1/4 bias
- Phase inversion to reduce power consumption and EMI
- Up to 8 pixels can be programmed to blink
- Unneeded segments and common pins can be used as general I/O pins
- LCD RAM can be updated at any time owing to a double-buffer
- The LCD controller can operate in Stop mode

# 3.11 ADC (analog-to-digital converter)

A 12-bit analog-to-digital converters is embedded into STM32L15xxD devices with up to 40 external channels, performing conversions in single-shot or scan mode. In scan mode, automatic conversion is performed on a selected group of analog inputs with up to 29 external channel in a group.

The ADC can be served by the DMA controller.

An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all scanned channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

The events generated by the general-purpose timers (TIMx) can be internally connected to the ADC start triggers, to allow the application to synchronize A/D conversions and timers. An injection mode allows high priority conversions to be done by interrupting a scan mode which runs in as a background task.

The ADC includes a specific low power mode. The converter is able to operate at maximum speed even if the CPU is operating at a very low frequency and has an auto-shutdown function. The ADC's runtime and analog front-end current consumption are thus minimized whatever the MCU operating mode.

### **Temperature sensor**

The temperature sensor has to generate a voltage that varies linearly with temperature. The conversion range is between 1.8 V < V<sub>DDA</sub> < 3.6 V. The temperature sensor is internally connected to the ADC\_IN16 input channel.

### Voltage reference

An internal precise voltage reference can be measured through the ADC. It enables accurate monitoring of the  $V_{DD}$  value (when no external voltage,  $V_{REF+}$ , is available for ADC).

### 3.12 DAC (digital-to-analog converter)

The two 12-bit buffered DAC channels can be used to convert two digital signals into two analog voltage signal outputs. The chosen design structure is composed of integrated resistor strings and an amplifier in non-inverting configuration.

This dual digital Interface supports the following features:

- Two DAC converters: one for each output channel
- Up to 10-bit output
- Left or right data alignment in 12-bit mode
- Synchronized update capability
- Noise-wave generation
- Triangular-wave generation
- Dual DAC channels' independent or simultaneous conversions
- DMA capability for each channel (including the underrun interrupt)
- External triggers for conversion
- Input reference voltage V<sub>REF+</sub>

Eight DAC trigger inputs are used in the STM32L15xxD. The DAC channels are triggered through the timer update outputs that are also connected to different DMA channels.

# 3.13 Operational amplifier

The STM32L15xxD embeds three operational amplifiers with external or internal follower routing capability (or even amplifier and filter capability with external components). When one operational amplifier is selected, one external ADC channel is used to enable output measurement.

The operational amplifier features:

- Low input bias current
- Low offset voltage
- Low power mode
- Rail-to-rail input

# 3.14 Ultralow power comparators and reference voltage

The STM32L15xxD embeds two comparators sharing the same current bias and reference voltage. The reference voltage can be internal or external (coming from an I/O).

- One comparator with fixed threshold
- One comparator with rail-to-rail inputs, fast or slow mode. The threshold can be one of the following:
  - DAC output
  - External I/O
  - Internal reference voltage (V<sub>REFINT</sub>) or a submultiple (1/4, 1/2, 3/4)

Both comparators can wake up from Stop mode, and be combined into a window comparator.

The internal reference voltage is available externally via a low power / low current output buffer (driving current capability of 1 µA typical).

# 3.15 System configuration controller and routing interface

The system configuration controller provides the capability to remap some alternate functions on different I/O ports.

The highly flexible routing interface allows the application firmware to control the routing of different I/Os to the TIM2, TIM3 and TIM4 timer input captures. It also controls the routing of internal analog signals to ADC1, COMP1 and COMP2 and the internal reference voltage  $V_{\rm RFFINT}$ .

# 3.16 Touch sensing

The STM32L15xxD devices provide a simple solution for adding capacitive sensing functionality to any application. Capacitive sensing technology is able to detect finger presence near an electrode which is protected from direct touch by a dielectric (glass, plastic, ...). The capacitive variation introduced by the finger (or any conductive object) is measured using a proven implementation based on a surface charge transfer acquisition principle. It consists of charging the electrode capacitance and then transferring a part of the accumulated charges into a sampling capacitor until the voltage across this capacitor has reached a specific threshold. In the STM32L15xxD, this acquisition is managed directly by the GPIOs, timers and analog I/O groups (see Section 3.15: System configuration controller and routing interface).

Reliable touch sensing solution can be quickly and easily implemented using the free STM32 touch sensing firmware library.

# 3.17 Timers and watchdogs

The ultralow power STM32L15xxD devices include seven general-purpose timers, two basic timers and two watchdog timers.

*Table 3* compares the features of the general-purpose and basic timers.

Table 3. Timer feature comparison

Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/compare channels	Complementary outputs
TIM2, TIM3, TIM4	16-bit	Up, down, up/down	Any integer between 1 and 65536	Yes	4	No
TIM5	32-bit	Up, down, up/down	Any integer between 1 and 65536	Yes	4	No
TIM9	16-bit	Up, down, up/down	Any integer between 1 and 65536	No	2	No
TIM10, TIM11	16-bit	Up	Any integer between 1 and 65536	No	1	No
TIM6, TIM7	16-bit	Up	Any integer between 1 and 65536	Yes	0	No

# 3.17.1 General-purpose timers (TIM2, TIM3, TIM4, TIM5, TIM9, TIM10 and TIM11)

There are six synchronizable general-purpose timers embedded in the STM32L15xxD devices (see *Table 3* for differences).

### TIM2, TIM3, TIM4, TIM5

TIM2, TIM3, TIM4 are based on 16-bit auto-reload up/down counter. TIM5 is based on a 32-bit auto-reload up/down counter. They include a 16-bit prescaler. They feature 4 independent channels each for input capture/output compare, PWM or one-pulse mode output. This gives up to 16 input captures/output compares/PWMs on the largest packages.

The TIM2, TIM3, TIM4, TIM5 general-purpose timers can work together or with the TIM10, TIM11 and TIM9 general-purpose timers via the Timer Link feature for synchronization or event chaining. Their counter can be frozen in debug mode. Any of the general-purpose timers can be used to generate PWM outputs.

TIM2, TIM3, TIM4, TIM5 all have independent DMA request generation.

These timers are capable of handling quadrature (incremental) encoder signals and the digital outputs from 1 to 3 hall-effect sensors.

### TIM10, TIM11 and TIM9

TIM10 and TIM11 are based on a 16-bit auto-reload upcounter. TIM9 is based on a 16-bit auto-reload up/down counter. They include a 16-bit prescaler. TIM10 and TIM11 feature one independent channel, whereas TIM9 has two independent channels for input capture/output compare, PWM or one-pulse mode output. They can be synchronized with the TIM2, TIM3, TIM4, TIM5 full-featured general-purpose timers.

They can also be used as simple time bases and be clocked by the LSE clock source (32.768 kHz) to provide time bases independent from the main CPU clock.

### 3.17.2 Basic timers (TIM6 and TIM7)

These timers are mainly used for DAC trigger generation. They can also be used as generic 16-bit time bases.

### 3.17.3 SysTick timer

This timer is dedicated to the OS, but could also be used as a standard downcounter. It is based on a 24-bit downcounter with autoreload capability and a programmable clock source. It features a maskable system interrupt generation when the counter reaches 0.

### 3.17.4 Independent watchdog (IWDG)

The independent watchdog is based on a 12-bit downcounter and 8-bit prescaler. It is clocked from an independent 37 kHz internal RC and, as it operates independently of the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free-running timer for application timeout management. It is hardware- or software-configurable through the option bytes. The counter can be frozen in debug mode.

### 3.17.5 Window watchdog (WWDG)

The window watchdog is based on a 7-bit downcounter that can be set as free-running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from the main clock. It has an early warning interrupt capability and the counter can be frozen in debug mode.

### 3.18 Communication interfaces

### 3.18.1 I2C bus

Up to two I<sup>2</sup>C bus interfaces can operate in multimaster and slave modes. They can support standard and fast modes.

They support dual slave addressing (7-bit only) and both 7- and 10-bit addressing in master mode. A hardware CRC generation/verification is embedded.

They can be served by DMA and they support SM Bus 2.0/PM Bus.

### 3.18.2 Universal synchronous/asynchronous receiver transmitter (USART)

The three USART and two UART interfaces are able to communicate at speeds of up to 4 Mbit/s. They support IrDA SIR ENDEC, are ISO 7816 compliant and have LIN Master/Slave capability. The three USART provide hardware management of the CTS and RTS signals.

All USART/UART interfaces can be served by the DMA controller.

## 3.18.3 Serial peripheral interface (SPI)

Up to three SPIs are able to communicate at up to 16 Mbits/s in slave and master modes in full-duplex and simplex communication modes. The 3-bit prescaler gives 8 master mode frequencies and the frame is configurable to 8 bits or 16 bits. The hardware CRC generation/verification supports basic SD Card/MMC modes.

The SPIs can be served by the DMA controller.

# 3.18.4 Inter-integrated sound (I<sup>2</sup>S)

Two standard I2S interfaces (multiplexed with SPI2 and SPI3) are available. They can operate in master or slave mode, and can be configured to operate with a 16-/32-bit resolution as input or output channels. Audio sampling frequencies from 8 kHz up to 96 kHz are supported. When either or both of the I2S interfaces is/are configured in master mode, the master clock can be output to the external DAC/CODEC at 256 times the sampling frequency.

#### 3.18.5 SDIO

An SD/SDIO/MMC host interface is available, that supports MultiMediaCard System Specification Version 4.2 in three different databus modes: 1-bit (default), 4-bit and 8-bit.

The interface allows data transfer at up to 48 MHz in 8-bit mode, and is compliant with the SD Memory Card Specification Version 2.0.

The SDIO Card Specification Version 2.0 is also supported with two different databus modes: 1-bit (default) and 4-bit.

The current version supports only one SD/SDIO/MMC4.2 card at any one time and a stack of MMC4.1 or previous.

In addition to SD/SDIO/MMC, this interface is fully compliant with the CE-ATA digital protocol Rev1.1.

### 3.18.6 Universal serial bus (USB)

The STM32L15xxD embeds a USB device peripheral compatible with the USB full-speed 12 Mbit/s. The USB interface implements a full-speed (12 Mbit/s) function interface. It has software-configurable endpoint setting and supports suspend/resume. The dedicated 48 MHz clock is generated from the internal main PLL (the clock source must use a HSE crystal oscillator).

# 3.19 CRC (cyclic redundancy check) calculation unit

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code from a 32-bit data word and a fixed generator polynomial.

Among other applications, CRC-based techniques are used to verify data transmission or storage integrity. In the scope of the EN/IEC 60335-1 standard, they offer a means of verifying the Flash memory integrity. The CRC calculation unit helps compute a signature of the software during runtime, to be compared with a reference signature generated at link-time and stored at a given memory location.

# 3.20 Development support

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### Serial wire JTAG debug port (SWJ-DP)

The ARM SWJ-DP interface is embedded, and is a combined JTAG and serial wire debug port that enables either a serial wire debug or a JTAG probe to be connected to the target. The JTAG TMS and TCK pins are shared with SWDIO and SWCLK, respectively, and a specific sequence on the TMS pin is used to switch between JTAG-DP and SW-DP.

The JTAG port can be permanently disabled with a JTAG fuse.

### Embedded Trace Macrocell™

The ARM® Embedded Trace Macrocell provides a greater visibility of the instruction and data flow inside the CPU core by streaming compressed data at a very high rate from the STM32L15xxD through a small number of ETM pins to an external hardware trace port analyzer (TPA) device. The TPA is connected to a host computer using USB, Ethernet, or any other high-speed channel. Real-time instruction and data flow activity can be recorded and then formatted for display on the host computer running debugger software. TPA hardware is commercially available from common development tool vendors. It operates with third party debugger software tools.

# 4 Pin descriptions

Table 4. STM32L15xQD BGA132 7 x7 ballout

	1	2	3	4	5	6	7	8	9	10	11	12
Α	PE3	PE1	PB8	воото	PD7	PD5	PB4	PB3	PA15	PA14	PA13	PA12
В	PE4	PE2	PB9	PB7	PB6	PD6	PD4	PD3	PD1	PC12	PC10	PA11
С	PC13- WKUP2	PE5	PE0	V <sub>DD_3</sub>	PB5	PG14	PG13	PD2	PD0	PC11	PH2	PA10
D	PC14- OSC32 _IN	PE6- WKUP3	V <sub>SS_3</sub>	PF2	PF1	PF0	PG12	PG10	PG9	PA9	PA8	PC9
E	PC15- OSC32 _OUT	VLCD	V <sub>SS_6</sub>	PF3					PG5	PC8	PC7	PC6
F	PH0 OSC_IN	$V_{\rm SS\_5}$	PF4	PF5		V <sub>SS_9</sub>	V <sub>SS_10</sub>		PG3	PG4	$V_{SS\_2}$	V <sub>SS_1</sub>
G	PH1 OSC_ OUT	V <sub>DD_5</sub>	PF6	PF7		V <sub>DD_9</sub>	V <sub>DD_10</sub>		PG1	PG2	V <sub>DD_2</sub>	V <sub>DD_1</sub>
Н	PC0	NRST	$V_{DD\_6}$	PF8					PG0	PD15	PD14	PD13
J	V <sub>SSA</sub>	PC1	PC2	PA4	PA7	PF9	PF12	PF14	PF15	PD12	PD11	PD10
K	OPAMP 3_ VINM	PC3	PA2	PA5	PC4	PF11	PF13	PD9	PD8	PB15	PB14	PB13
L	V <sub>REF+</sub>	PA0- WKUP1	PA3	PA6	PC5	PB2	PE8	PE10	PE12	PB10	PB11	PB12
М	V <sub>DDA</sub>	PA1	OPAM P1_ VINM	OPAMP 2_ VINM	PB0	PB1	PE7	PE9	PE11	PE13	PE14	PE15

Figure 3. STM32L15xZDLQFP144 pinout

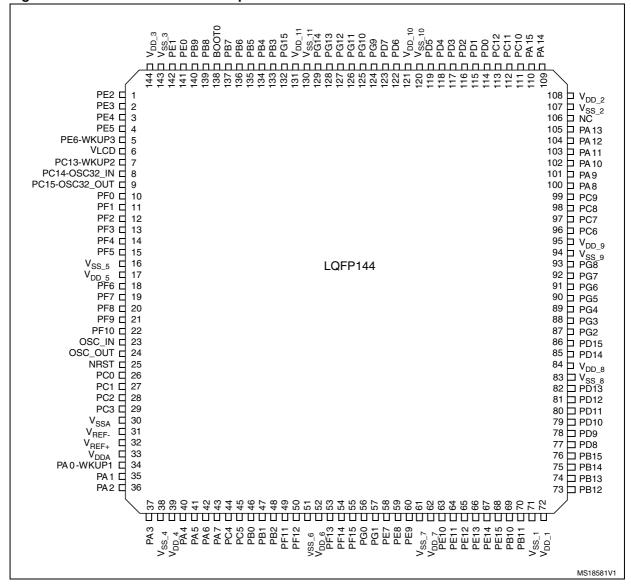


Figure 4. STM32L15xVD LQFP100 pinout

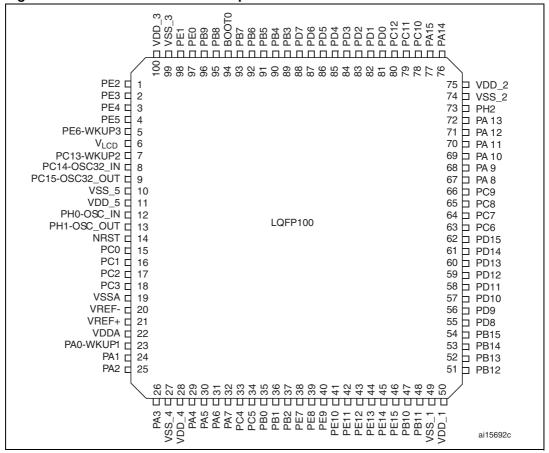


Figure 5. STM32L15xRD LQFP64 pinout

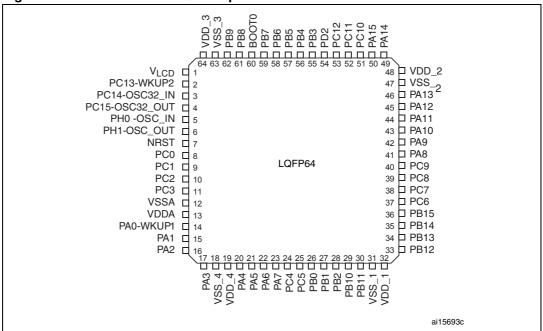


Table 5. STM32L15xxD pin definitions

LQFP144	FBGA132	0						l l
┸	UFBG	LQFP100	LQFP64	Pin name	Type <sup>(1)</sup>	I / O Level <sup>(2)</sup>	Main function <sup>(3)</sup> (after reset)	Alternate functions
1	B2	1	-	PE2	I/O	FT	PE2	TIM3_ETR/LCD_SEG38/ FSMC_A23/TRACECK
2	A1	2	-	PE3	I/O	FT	PE3	TIM3_CH1/LCD_SEG39/ FSMC_A19/TRACED0
3	B1	3	-	PE4	I/O	FT	PE4	TIM3_CH2/FSMC_A20/TRACED1
4	C2	4	ı	PE5	0	FT	PE5	TIM9_CH1/FSMC_A21/TRACED2
5	D2	5	1	PE6- WKUP3	I/O	FT	PE6	WKUP3/TAMPER3/TIM9_CH2/TRACED3
6	E2	6	1	V <sub>LCD</sub> <sup>(4)</sup>	S		$V_{LCD}$	
7	C1	7	2	PC13-WKUP2	I/O	FT	PC13	WKUP2/TAMPER1/RTC_AF1
8	D1	8	3	PC14- OSC32_IN <sup>(5)</sup>	I/O		PC14	OSC32_IN
9	E1	9	4	PC15- OSC32_OUT	I/O		PC15	OSC32_OUT
10	D6	-	-	PF0	I/O	FT	PF0	FSMC_A0
11	D5	-	-	PF1	I/O	FT	PF1	FSMC_A1
12	D4	-	-	PF2	I/O	FT	PF2	FSMC_A2
13	E4	-	-	PF3	I/O	FT	PF3	FSMC_A3
14	F3	-	-	PF4	I/O	FT	PF4	FSMC_A4
15	F4	-	-	PF5	I/O	FT	PF5	FSMC_A5
16	F2	10	1	V <sub>SS_5</sub>	S		$V_{SS_5}$	
17	G2	11	1	$V_{DD_5}$	S		V <sub>DD_5</sub>	
18	G3	-	-	PF6	I/O	FT	PF6	TIM5_CH1/TIM5_ETR/ADC_IN27
19	G4	-	-	PF7	I/O	FT	PF7	TIM5_CH2/ADC_IN28/COMP1_INP
20	H4	-	-	PF8	I/O	FT	PF8	TIM5_CH3/ADC_IN29/COMP1_INP
21	J6	-	1	PF9	I/O	FT	PF9	TIM5_CH4/ADC_IN30/COMP1_INP
22	-	-	-	PF10	I/O	FT	PF10	ADC_IN30/COMP1_INP
23	F1	12	5	PH0-OSC_IN <sup>(6)</sup>	ı		PH0	OSC_IN
24	G1	13	6	PH1- OSC_OUT <sup>(6)</sup>	0		PH1	OSC_VOUT
25	H2	14	7	NRST	I/O		NRST	
26	H1	15	8	PC0	I/O	FT	PC0	LCD_SEG18/ADC_IN10/COMP1_INP
27	J2	16	9	PC1	I/O	FT	PC1	LCD_SEG19/ADC_IN11/COMP1_INP/OPAMP3_VINP
28		17	10	PC2	I/O	FT	PC2	LCD_SEG20/ADC_IN12/COMP1_INP/OPAMP3_VINM
	J3			PC2	1/0	FT	PC2	LCD_SEG20/ADC_IN12/COMP1_INP
	K1			OPAMP3_VINM	I		OPAMP3_VI NM	
29	K2	18	11	PC3	I/O		PC3	LCD_SEG21/ADC_IN13/COMP1_INP/OPAMP3_VOUT
30	J1	19	12	V <sub>SSA</sub>	S		V <sub>SSA</sub>	

Table 5. STM32L15xxD pin definitions (continued)

Tab	Pins		STW32LT3XXD piii				(11111111111111111111111111111111111111	
LQFP144	UFBGA132	LQFP100	LQFP64	Pin name	Type <sup>(1)</sup>	I / O Level <sup>(2)</sup>	Main function <sup>(3)</sup> (after reset)	Alternate functions
31	-	20	-	V <sub>REF-</sub>	S		$V_{REF}$	
32	L1	21	-	$V_{REF+}$	S		$V_{REF+}$	
33	M1	22	13	$V_{DDA}$	S		$V_{DDA}$	
34	L2	23	14	PA0-WKUP1	I/O	FT	PA0	WKUP1/TAMPER2/TIM2_CH1_ETR/TIM5_CH1/ USART2_CTS/ADC_IN0/COMP1_INP
35	M2	24	15	PA1	I/O	FT	PA1	TIM2_CH2/TIM5_CH2/ USART2_RTS/LCD_SEG0/ ADC_IN1/COMP1_INP/OPAMP1_VINP
36		25	16	PA2	I/O	FT	PA2	TIM2_CH3/TIM5_CH3/TIM9_CH1/USART2_TX/ LCD_SEG1/ADC_IN2/ COMP1_INP/OPAMP1_VINM
	КЗ			PA2	I/O	FT	PA2	TIM2_CH3/TIM5_CH3/TIM9_CH1/USART2_TX/ LCD_SEG1/ADC_IN2/ COMP1_INP
	МЗ			OPAMP1_VINM	I		OPAMP1_VI NM	
37	L3	26	17	PA3	I/O		PA3	TIM2_CH4/TIM5_CH4/TIM9_CH2/USART2_RX/ LCD_SEG2/ ADC_IN3/COMP1_INP/OPAMP1_OUT
38	-	27	18	$V_{SS\_4}$	S		$V_{SS\_4}$	
39	-	28	19	$V_{DD\_4}$	S		$V_{DD\_4}$	
40	J4	29	20	PA4	I/O		PA4	SPI1_NSS/SPI3_NSS/ I2S3_WS/USART2_CK/ ADC_IN4/DAC_OUT1/COMP1_INP
41	K4	30	21	PA5	I/O		PA5	TIM2_CH1_ETR/SPI1_SCK/ADC_IN5/DAC_OUT2/ COMP1_INP
42	L4	31	22	PA6	I/O	FT	PA6	TIM3_CH1/TIM10_CH1/SPI1_MISO/LCD_SEG3/ ADC_IN6/COMP1_INP/OPAMP2_VINP
43		32	23	PA7	I/O	FT	PA7	TIM3_CH2/TIM11_CH1/ SPI1_MOSI/LCD_SEG4/ ADC_IN7/COMP1_INP/OPAMP2_VINM
	J5			PA7	I/O	FT	PA7	TIM3_CH2/TIM11_CH1/ SPI1_MOSI/LCD_SEG4/ ADC_IN7/COMP1_INP
	M4			OPAMP2_VINM	I		OPAMP2_VI NM	
44	K5	33	24	PC4	I/O	FT	PC4	LCD_SEG22/ADC_IN14/COMP1_INP
45	L5	34	25	PC5	I/O	FT	PC5	LCD_SEG23/ADC_IN15/COMP1_INP
46	M5	35	26	PB0	I/O		PB0	TIM3_CH3/LCD_SEG5/ADC_IN8/COMP1_INP/ VREF_OUT/ OPAMP2_VOUT
47	M6	36	27	PB1	I/O	FT	PB1	TIM3_CH4/LCD_SEG6/ADC_IN9/COMP1_INP/ VREF_OUT
48	L6	37	28	PB2	I/O	FT	PB2/BOOT1	ADC_IN0b/COMP1_INP
49	K6		1	PF11	I/O	FT	PF11	ADC_IN1b/COMP1_INP
50	J7	_	-	PF12	I/O	FT	PF12	ADC_IN2b/COMP1_INP/FSMC_A6
51	E3	-	-	V <sub>SS_6</sub>	S		$V_{SS\_6}$	
52	НЗ	-	-	$V_{DD\_6}$	S		$V_{DD_6}$	

Table 5. STM32L15xxD pin definitions (continued)

	Pins		<u> </u>		(2)			
LQFP144	UFBGA132	LQFP100	LQFP64	Pin name	Type <sup>(1)</sup>	I / O Level <sup>(2)</sup>	Main function <sup>(3)</sup> (after reset)	Alternate functions
53	K7	-	-	PF13	1/0	FT	PF13	ADC_IN3b/COMP1_INP/FSMC_A7
54	J8	-	-	PF14	0	FT	PF14	ADC_IN6b/COMP1_INP/FSMC_A8
55	J9	-	-	PF15	0	FT	PF15	ADC_IN7b/COMP1_INP/FSMC_A9
56	Н9	-	-	PG0	0	FT	PG0	ADC_IN8b/COMP1_INP/FSMC_A10
57	G9	-	-	PG1	I/O	FT	PG1	FSMC_A6/ADC_IN9b/COMP1_INP/FSMC_A11
58	M7	38	-	PE7	I/O		PE7	FSMC_D4/ ADC_IN22/COMP1_INP
59	L7	39	-	PE8	I/O		PE8	FSMC_D5/ADC_IN23/COMP1_INP
60	M8	40	-	PE9	I/O		PE9	TIM2_CH1_ETR/FSMC_D6/ ADC_IN24/COMP1_INP
61	-	-	-	V <sub>SS_7</sub>	S		V <sub>SS_7</sub>	
62	-	-	-	$V_{DD_{\_7}}$	S		$V_{DD_{2}}$	
63	L8	41	-	PE10	I/O		PE10	TIM2_CH2/ FSMC_D7/ADC_IN25/COMP1_INP
64	М9	42	-	PE11	I/O	FT	PE11	TIM2_CH3/FSMC_D8
65	L9	43	-	PE12	I/O	FT	PE12	TIM2_CH4/SPI1_NSS/FSMC_D9
66	M10	44	-	PE13	I/O	FT	PE13	SPI1_SCK/FSMC_D10
67	M11	45	-	PE14	I/O	FT	PE14	SPI1_MISO/FSMC_D11
68	M12	46	-	PE15	I/O	FT	PE15	SPI1_MOSI/FSMC_D12
69	L10	47	29	PB10	I/O	FT	PB10	TIM2_CH3/I2C2_SCL/USART3_TX/LCD_SEG10
70	L11	48	30	PB11	I/O	FT	PB11	TIM2_CH4/I2C2_SDA/ USART3_RX/LCD_SEG11
71	F12	49	31	V <sub>SS_1</sub>	S		V <sub>SS_1</sub>	
72	G12	50	32	$V_{DD_1}$	S		$V_{DD_1}$	
73	L12	51	33	PB12	I/O	FT	PB12	TIM10_CH1/I2C2_SMBA/SPI2_NSS/I2S2_WS/ USART3_CK/ LCD_SEG12/ADC_IN18/COMP1_INP
74	K12	52	34	PB13	I/O	FT	PB13	TIM9_CH1/SPI2_SCK/ I2S2_CK/ USART3_CTS/ LCD_SEG13/ADC_IN19/COMP1_INP
75	K11	53	35	PB14	I/O	FT	PB14	TIM9_CH2/SPI2_MISO/ USART3_RTS/LCD_SEG14/ ADC_IN20/COMP1_INP
76	K10	54	36	PB15	I/O	FT	PB15	TIM11_CH1/SPI2_MOSI/I2S2_SD/LCD_SEG15/ ADC_IN21/COMP1_INP/RTC_50_60Hz
77	<b>K</b> 9	55	-	PD8	I/O	FT	PD8	USART3_TX/LCD_SEG28/FSMC_D13
78	K8	56	-	PD9	I/O	FT	PD9	USART3_RX/LCD_SEG29/FSMC_D14
79	J12	57	-	PD10	I/O	FT	PD10	USART3_CK/LCD_SEG30/FSMC_D15
80	J11	58	-	PD11	I/O	FT	PD11	USART3_CTS/LCD_SEG31/FSMC_A16
81	J10	59	-	PD12	I/O	FT	PD12	TIM4_CH1 / USART3_RTS/LCD_SEG32/FSMC_A17
82	H12	60	_	PD13	I/O	FT	PD13	TIM4_CH2/LCD_SEG33/FSMC_A18
83	-	-	-	V <sub>SS_8</sub>	S		$V_{SS_8}$	
84	-	-	-	V <sub>DD_8</sub>	S		$V_{DD_8}$	
85	H11	61	-	PD14	I/O	FT	PD14	TIM4_CH3/LCD_SEG34/FSMC_D0

Table 5. STM32L15xxD pin definitions (continued)

Pins			STWSZETSKAD PIII			(50)	,	
LQFP144	UFBGA132	LQFP100	LQFP64	Pin name	Type <sup>(1)</sup>	I / O Level <sup>(2)</sup>	Main function <sup>(3)</sup> (after reset)	Alternate functions
86	H10	62	-	PD15	1/0	FT	PD15	TIM4_CH4/LCD_SEG35/FSMC_D1
87	G10	•	-	PG2	1/0	FT	PG2	FSMC_A12/ADC_IN10b/COMP1_INP
88	F9	-	-	PG3	I/O	FT	PG3	FSMC_A13/ADC_IN11b/COMP1_INP
89	F10	-	-	PG4	I/O	FT	PG4	FSMC_A14/ADC_IN12b/COMP1_INP
90	E9	•	-	PG5	1/0	FT	PG5	FSMC_A15
91	-	ı	-	PG6	0	FT	PG6	
92	-	-	-	PG7	I/O	FT	PG7	
93	-	-	-	PG8	I/O	FT	PG8	
94	F6	•	-	$V_{SS\_9}$	S		$V_{SS_9}$	
95	G6	-	-	$V_{DD_9}$	S		$V_{DD_9}$	
96	E12	63	37	PC6	I/O	FT	PC6	TIM3_CH1/I2S2_MCK/LCD_SEG24/FSMC_D6/ SDIO_D6
97	E11	64	38	PC7	I/O	FT	PC7	TIM3_CH2/I2S3_MCK/LCD_SEG25/FSMC_D7/ SDIO_D7
98	E10	65	39	PC8	1/0	FT	PC8	TIM3_CH3/LCD_SEG26/FSMC_D0/SDIO_D0
99	D12	66	40	PC9	0	FT	PC9	TIM3_CH4/LCD_SEG27/FSMC_D1/SDIO_D1
100	D11	67	41	PA8	I/O	FT	PA8	USART1_CK/MCO/LCD_COM0
101	D10	68	42	PA9	I/O	FT	PA9	USART1_TX / LCD_COM1
102	C12	69	43	PA10	I/O	FT	PA10	USART1_RX / LCD_COM2
103	B12	70	44	PA11	I/O	FT	PA11	USART1_CTS/ USB_DM/SPI1_MISO
104	A12	71	45	PA12	I/O	FT	PA12	USART1_RTS/USB_DP/SPI1_MOSI
105	A11	72	46	PA13	I/O	FT	JTMS- SWDIO	PA13
106	C11	73	-	PH2	I/O	FT	PH2	FSMC_A22
107	F11	74	47	V <sub>SS_2</sub>	S		V <sub>SS_2</sub>	
108	G11	75	48	V <sub>DD_2</sub>	S		$V_{DD_2}$	
109	A10	76	49	PA14	I/O	FT	JTCK- SWCLK	PA14
110	<b>A</b> 9	77	50	PA15	I/O	FT	JTDI	TIM2_CH1_ETR/ SPI1_NSS/SPI3_NSS/ I2S3_WS/LCD_SEG17
111	B11	78	51	PC10	I/O	FT	PC10	SPI3_SCK/I2S3_CK/USART3_TX/ UART4_TX/ LCD_SEG28/LCD_SEG40/LCD_COM4/FSMC_D2/ SDIO_D2
112	C10	79	52	PC11	I/O	FT	PC11	SPI3_MISO/USART3_RX/UART4_RX/ LCD_SEG29/LCD_SEG41/LCD_COM5/FSMC_D3/ SDIO_D3
113	B10	80	53	PC12	I/O	FT	PC12	SPI3_MOSI/I2S3_SD/USART3_CK/ UART5_TX/ LCD_SEG30/ LCD_SEG42/LCD_COM6/SDIO_CK
114	C9	81	-	PD0	I/O	FT	PD0	TIM9_CH1/SPI2_NSS/I2S2_WS/ FSMC_D2

Table 5. STM32L15xxD pin definitions (continued)

Pins			3 I WISZE I SAAD PIII				(001111111	,
LQFP144	UFBGA132	LQFP100	LQFP64	Pin name	Type <sup>(1)</sup>	I / O Level <sup>(2)</sup>	Main function <sup>(3)</sup> (after reset)	Alternate functions
115	B9	82	-	PD1	I/O	FT	PD1	SPI2_SCK/I2S2_CK/FSMC_D3
116	C8	83	54	PD2	I/O	FT	PD2	TIM3_ETR/UART5_RX/LCD_SEG31/LCD_SEG43/ LCD_COM7/SDIO_CMD
117	B8	84	1	PD3	I/O	FT	PD3	SPI2_MISO/USART2_CTS/FSMC_CLK
118	В7	85	-	PD4	I/O	FT	PD4	SPI2_MOSI/I2S2_SD/ USART2_RTS/FSMC_NOE
119	A6	86		PD5	I/O	FT	PD5	USART2_TX/FSMC_NWE
120	F7	ı	ı	V <sub>SS_10</sub>	S		V <sub>SS_10</sub>	
121	G7	ı	ı	$V_{DD\_10}$	Ø		$V_{DD\_10}$	
122	B6	87	ı	PD6	9	FT	PD6	USART2_RX/FSMC_NWAIT
123	A5	88	1	PD7	I/O	FT	PD7	TIM9_CH2/USART2_CK/FSMC_EBAR0
124	D9	-	-	PG9	I/O	FT	PG9	FSMC_EBAR1
125	D8	-	-	PG10	I/O	FT	PG10	FSMC_EBAR2
126	-	-	-	PG11	I/O	FT	PG11	
127	D7	-	-	PG12	I/O	FT	PG12	FSMC_EBAR3
128	C7	1	1	PG13	I/O	FT	PG13	FSMC_A24
129	C6	1	1	PG14	I/O	FT	PG14	FSMC_A25
130	•	ı	ı	V <sub>SS_11</sub>	S		V <sub>SS_11</sub>	
131	-	•	•	$V_{DD\_11}$	S		$V_{DD\_11}$	
132	-	-	-	PG15	I/O	FT	PG15	
133	A8	89	55	PB3	I/O	FT	JTDO	TIM2_CH2/SPI1_SCK/SPI3_SCK/ I2S3_CK/ LCD_SEG7/COMP2_INM/TRACESWO
134	A7	90	56	PB4	I/O	FT	NJTRST	TIM3_CH1/ SPI1_MISO/SPI3_MISO/LCD_SEG8/ COMP2_INP
135	C5	91	57	PB5	I/O	FT	PB5	TIM3_CH2 /I2C1_SMBAI/SPI1_MOSI/SPI3_MOSI/ I2S3_SD/LCD_SEG9/COMP2_INP
136	B5	92	58	PB6	I/O	FT	PB6	TIM4_CH1/I2C1_SCL/USART1_TX/COMP2_INP
137	B4	93	59	PB7	I/O	FT	PB7	TIM4_CH2/I2C1_SDA/USART1_RX/PVD_IN/ FSMC_LBAR/ COMP2_INP
138	A4	94	60	BOOT0	I		BOOT0	
139	АЗ	95	61	PB8	I/O	FT	PB8	TIM4_CH3/TIM10_CH1/I2C1_SCL/LCD_SEG16/ FSMC_D4/ SDIO_D4
140	ВЗ	96	62	PB9	I/O	FT	PB9	TIM4_CH4/ TIM11_CH1/I2C1_SDA/LCD_COM3/ FSMC_D5/SDIO_D5
141	СЗ	97	1	PE0	I/O	FT	PE0	TIM4_ETR/TIM10_CH1/LCD_SEG36 /FSMC_NBL0
142	A2	98	-	PE1	I/O	FT	PE1	TIM11_CH1/LCD_SEG37/FSMC_NBL1
143	D3	99	63	V <sub>SS_3</sub>	S		$V_{SS\_3}$	
144	C4	100	64	V <sub>DD_3</sub>	S		$V_{DD_3}$	

- 1. I = input, O = output, S = supply.
- 2. FT = 5 V tolerant.
- 3. Function availability depends on the chosen device.
- 4. Applicable to STM32L152xD devices only. In STM32L151xD devices, this pin should be connected to  $V_{DD}$ .
- 5. The PC14 and PC15 I/Os are only configured as OSC32\_IN/OSC32\_OUT when the LSE oscillator is ON (by setting the LSEON bit in the RCC\_CSR register). The LSE oscillator pins OSC32\_IN/OSC32\_OUT can be used as general-purpose PH0/PH1 I/Os, respectively, when the LSE oscillator is off (after reset, the LSE oscillator is off). The LSE has priority over the GPIO function. For more details, refer to Using the OSC32\_IN/OSC32\_OUT pins as GPIO PC14/PC15 port pins section in the STM32L15xxx reference manual (RM0072).
- The PH0 and PH1 I/Os are only configured as OSC\_IN/OSC\_OUT when the HSE oscillator is ON (by setting the HSEON bit in the RCC\_CR register). The HSE oscillator pins OSC\_IN/OSC\_OUT can be used as general-purpose PH0/PH1 I/Os, respectively, when the HSE oscillator is off ( after reset, the HSE oscillator is off ). The HSE has priority over the GPIO function.



Table 6. Alternate function input/output

						Digit	al alterna	te function r	number					
	AFIO0	AFIO1	AFIO2	AFIO3	AFIO4	AFIO5	AFIO6	AFIO7	AFIO8 .	. AFIO10	AFIO11	AFIO12	. AFIO14	AFIO15
Port name			•			•	Altern	ate function		•				•
	SYSTEM	TIM2	TIM3/4/5	TIM9/ 10/11	I2C1/2	SPI1/2	SPI3	USART1/2/3	UART4/5	USB	LCD	FSMC/ SDIO	CPRI	SYSTEM
воото	воото													EVENT OUT
NRST	NRST													
PA0- WKUP1	WKUP1/ TAMPER2	TIM2_CH1_ETR	TIM5_CH1					USART2_CTS					COMP1_INP/ TIMx_IC1_0/ G1IO1	EVENT OUT
PA1		TIM2_CH2	TIM5_CH2					USART2_RTS			SEG0		COMP1_INP/ TIMx_IC2_0 G1IO2	EVENT OUT
PA2		TIM2_CH3	TIM5_CH3	TIM9_CH1				USART2_TX			SEG1		COMP1_INP/ TIMx_IC3_0/ G1IO3	EVENT OUT
PA3		TIM2_CH4	TIM5_CH4	TIM9_CH2				USART2_RX			SEG2		COMP1_INP/ TIMx_IC4_0/ G1IO4	EVENT OUT
PA4						SPI1_NSS	SPI3_NSS I2S3_WS	USART2_CK					COMP1_INP/ TIMx_IC1_1	EVENT OUT
PA5		TIM2_CH1_ETR*				SPI1_SCK							COMP1_INP/ TIMx_IC2_1	EVENT OUT
PA6			TIM3_CH1	TIM10_ CH1		SPI1_MISO					SEG3		COMP1_INP/ TIMx_IC3_1 G2IO1	EVENT OUT
PA7			TIM3_CH2	TIM11_ CH1		SPI1_MOSI					SEG4		COMP1_INP/ TIMx_IC4_1/ G2IO2	EVENT OUT
PA8	мсо							USART1_CK			СОМ0		TIMx_IC1_2/ G4IO1	EVENT OUT
PA9								USART1_TX			COM1		TIMx_IC2_2/ G4IO2	EVENT OUT
PA10								USART1_RX			COM2		TIMx_IC3_2/ G4IO3	EVENT OUT

 Table 6.
 Alternate function input/output (continued)

						Digit	al alterna	te function n	number					
	AFIO0	AFIO1	AFIO2	AFIO3	AFIO4	AFIO5	AFIO6	AFIO7	AFIO8	AFIO10	AFIO11	AFIO12	AFIO14	AFIO15
Port name			•		•		Altern	ate function		•	•		•	
	SYSTEM	TIM2	TIM3/4/5	TIM9/ 10/11	I2C1/2	SPI1/2	SPI3	USART1/2/3	UART4/5	USB	LCD	FSMC/ SDIO	CPRI	SYSTEM
PA11						SPI1_MISO		USART1_CTS		USBDM			TIMx_IC4_2/ G4IO4	EVENT OUT
PA12						SPI1_MOSI		USART1_RTS		USBDP			TIMx_IC1_3/	EVENT OUT
PA13	JTMS-SWDIO												TIMx_IC2_3/ G5IO1	EVENT OUT
PA14	JTCK-SWCLK												TIMx_IC3_3/ G5IO2	EVEN TOUT
PA15	JTDI	TIM2_CH1_ETR				SPI1_NSS	SPI3_NSS I2S3_WS				SEG17		TIMx_IC4_3/ G5IO3	EVEN TOUT
PB0			TIM3_CH3								SEG5		COMP1_INP/ G3IO1	EVEN TOUT
PB1			TIM3_CH4								SEG6		COMP1_INP/ G3IO2	EVENT OUT
PB2	BOOT1												COMP1_INP/ G3IO3	EVENT OUT
PB3	JTDO	TIM2_CH2				SPI1_SCK	SPI3_SCK I2S3_CK				SEG7			EVENT OUT
PB4	JTRST		TIM3_CH1			SPI1_MISO	SPI3_MISO				SEG8		G6IO1	EVENT OUT
PB5			TIM3_CH2		I2C1_ SMBA	SPI1_MOSI	SPI3_MOSI I2S3_SD				SEG9		G6IO2	EVENT OUT
PB6			TIM4_CH1		I2C1_SCL			USART1_TX					G6IO3	EVENT OUT
PB7			TIM4_CH2		I2C1_SDA			USART1_RX				LBAR	G6IO4	EVENT OUT
PB8			TIM4_CH3	TIM10_ CH1	I2C1_SCL						SEG16	SDIO_D4		EVENT OUT
PB9			TIM4_CH4	TIM11_ CH1	I2C1_SDA						СОМЗ	SDIO_D5		EVENT OUT





 Table 6.
 Alternate function input/output (continued)

						Digita	al alterna	te function r	number					
	AFIO0	AFIO1	AFIO2	AFIO3	AFIO4	AFIO5	AFIO6	AFIO7	AFIO8	AFIO10	AFIO11	AFIO12	AFIO14	AFIO15
Port name		<u> </u>	1	l			Altern	ate function	<u> </u>		ı		1	<b>'</b>
	SYSTEM	TIM2	TIM3/4/5	TIM9/ 10/11	I2C1/2	SPI1/2	SPI3	USART1/2/3	UART4/5	USB	LCD	FSMC/ SDIO	CPRI	SYSTEM
PB10		TIM2_CH3			I2C_SCL			USART3_TX			SEG10			EVENT OUT
PB11		TIM2_CH4			I2C_SDA			USART3_RX			SEG11			EVENT OUT
PB12				TIM10_ CH1	I2C_SMBA	SPI2_NSS I2S2_WS		USART3_CK			SEG12		COMP1_INP/ G7IO1	EVENT OUT
PB13				TIM9_ CH1		SPI2_SCK I2S2_CK		USART3_CTS			SEG13		COMP1_INP/ G7IO2	EVENT OUT
PB14				TIM9_ CH2		SPI2_MISO		USART3_RTS			SEG14		COMP1_INP/ G7IO3	EVENT OUT
PB15	RTC 50/60 Hz			TIM11_ CH1		SPI2_MOSI I2S2_SD					SEG15		COMP1_INP/ G7IO4	EVENT OUT
PC0											SEG18		COMP1_INP/ TIMx_IC1_4/ G8IO1	EVENT OUT
PC1											SEG19		COMP1_INP/ TIMx_IC2_4/ G8IO2	EVENT OUT
PC2											SEG20		COMP1_INP/ TIMx_IC3_4/ G8IO3	EVENT OUT
PC3											SEG21		COMP1_INP/ TIMx_IC4_4/ G8IO4	EVENT OUT
PC4											SEG22		COMP1_INP/ TIMx_IC1_5/ G9IO1	EVENT OUT
PC5											SEG23		COMP1_INP/ TIMx_IC2_5/ G9IO2	EVENT OUT
PC6			TIM3_CH1			I2S2_MCK					SEG24	SDIO_D6	TIMx_IC3_5/ G10IO1	EVENT OUT

Pin descriptions

 Table 6.
 Alternate function input/output (continued)

						Digit	al alterna	te function r	number					
	AFIO0	AFIO1	AFIO2	AFIO3	AFIO4	AFIO5	AFIO6	AFIO7	AFIO8	AFIO10	AFIO11	AFIO12 .	AFIO14	AFIO15
Port name			<b>.</b>	l .		ı	Altern	ate function	<u></u>	II.	l	<u> </u>	1	1
	SYSTEM	TIM2	TIM3/4/5	TIM9/ 10/11	I2C1/2	SPI1/2	SPI3	USART1/2/3	UART4/5	USB	LCD	FSMC/ SDIO	CPRI	SYSTEM
PC7			TIM3_CH2				I2S3_MCK				SEG25	SDIO_D7	TIMx_IC4_5/ G10IO2	EVENT OUT
PC8			TIM3_CH3								SEG26	SDIO_D0	TIMx_IC1_6/ G10IO3	EVENT OUT
PC9			TIM3_CH4								SEG27	SDIO_D1	TIMx_IC2_6/ G10IO4	EVENT OUT
PC10							SPI3_SCK I2S3_CK	USART3_TX	UART4_TX		COM4/ SEG28/ SEG40	SDIO_D2	TIMx_IC3_6/ G5IO4	EVENT OUT
PC11							SPI3_MISO	USART3_RX	UART4_RX		COM5/ SEG29 /SEG41	SDIO_D3	TIMx_IC4_6	EVENT OUT
PC12							SPI3_MOSI I2S3_SD	USART3_CK	UART5_TX		COM6/ SEG30/ SEG42	SDIO_CK	TIMx_IC1_7	EVENT OUT
PC13- WKUP2	WKUP2/ TAMPER1/ TIMESTAMP/ ALARM_OUT/ 512Hz												TIMx_IC2_7	EVENT OUT
PC14 OSC32_I N	OSC32_IN												TIMx_IC3_7	EVENT OUT
PC15 OSC32_ OUT	OSC32_OUT												TIMx_IC4_7	EVENT OUT
PD0				TIM9_CH1		SPI2_NSS I2S2_WS						D2 / DA2	TIMx_IC1_8	EVENT OUT
PD1						SPI2 SCK I2S2_CK						D3 / DA3	TIMx_IC2_8	EVENT OUT
PD2			TIM3_ETR						UART5_RX		COM7/ SEG31/ SEG43	SDIO_ CMD	TIMx_IC3_8	EVENT OUT



 Table 6.
 Alternate function input/output (continued)

						Digita	al alterna	te function r	number					
	AFIO0	AFIO1	AFIO2	AFIO3	AFIO4	AFIO5	AFIO6	AFIO7	AFIO8	AFIO10	AFIO11	AFIO12 .	. AFIO14	AFIO15
Port name	1						Altern	ate function	1	<b>-1</b>	1	1	1	1
	SYSTEM	TIM2	TIM3/4/5	TIM9/ 10/11	I2C1/2	SPI1/2	SPI3	USART1/2/3	UART4/5	USB	LCD	FSMC/ SDIO	CPRI	SYSTEM
PD3						SPI2_MISO		USART2_CTS				CLK	TIMx_IC4_8	EVENT OUT
PD4						SPI2_MOSI I2S2_SD		USART2_RTS				OEN	TIMx_IC1_9	EVENT OUT
PD5								USART2_TX				WEN	TIMx_IC2_9	EVENT OUT
PD6								USART2_RX				WAITN	TIMx_IC3_9	EVENT OUT
PD7				TIM9_CH2				USART2_CK				EBAR0	TIMx_IC4_9	EVENT OUT
PD8								USART3_TX			SEG28	D13/DA13	TIMx_IC1_10	EVENT OUT
PD9								USART3_RX			SEG29	D14/DA14	TIMx_IC2_10	EVENT OUT
PD10								USART3_CK			SEG30	D15/DA15	TIMx_IC3_10	EVENT OUT
PD11								USART3_CTS			SEG31	A16	TIMx_IC4_10	EVENT OUT
PD12			TIM4_CH1					USART3_RTS			SEG32	A17	TIMx_IC1_11	EVENT OUT
PD13			TIM4_CH2								SEG33	A18	TIMx_IC2_11	EVENT OUT
PD14			TIM4_CH3								SEG34	D0/DA0	TIMx_IC3_11	EVENT OUT
PD15			TIM4_CH4								SEG35	D1/DA1	TIMx_IC4_11	EVENT OUT
PE0			TIM4_ETR	TIM10_ CH1							SEG36	BLN0	TIMx_IC1_12	EVENT OUT
PE1				TIM11_ CH1							SEG37	BLN1	TIMx_IC2_12	EVENT OUT

Pin descriptions

Table 6. Alternate function input/output (continued)

				•	<u> </u>	Digita	al alterna	ite function r	number						
	AFIO0	AFIO1	AFIO2	AFIO3	AFIO4	AFIO5	AFIO6	AFIO7	AFIO8	/	AFIO10	AFIO11	AFIO12	AFIO14	AFIO15
Port name		-	l	l		l .	Altern	ate function	l					1	
	SYSTEM	TIM2	TIM3/4/5	TIM9/ 10/11	I2C1/2	SPI1/2	SPI3	USART1/2/3	UART4/5		USB	LCD	FSMC/ SDIO	CPRI	SYSTEM
PE2	TRACECK		TIM3_ETR									SEG 38	A23	TIMx_IC3_12	EVENT OUT
PE3	TRACED0		TIM3_CH1									SEG 39	A19	TIMx_IC4_12	EVENT OUT
PE4	TRACED1		TIM3_CH2										A20	TIMx_IC1_13	EVENT OUT
PE5	TRACED2			TIM9_CH1									A21	TIMx_IC2_13	EVENT OUT
PE6- WKUP3	WKUP3/ TAMPER3 / TRACED3			TIM9_CH2										TIMx_IC3_13	EVENT OUT
PE7													D4/DA4	COMP1_INP/ TIMx_IC4_13	EVENT OUT
PE8													D5/DA5	COMP1_INP/ TIMx_IC1_14	EVENT OUT
PE9		TIM2_CH1_ETR											D6/DA6	COMP1_INP/ TIMx_IC2_14	EVENT OUT
PE10		TIM2_CH2											D7/DA7	COMP1_INP/ TIMx_IC3_14	EVENT OUT
PE11		TIM2_CH3											D8/DA8	TIMx_IC4_14	EVENT OUT
PE12		TIM2_CH4				SPI1_NSS							D9/DA9	TIMx_IC1_15	EVENT OUT
PE13						SPI1_SCK							D10/DA10	TIMx_IC2_15	EVENT OUT
PE14						SPI1_MISO							D11/DA11	TIMx_IC3_15	EVENT OUT
PE15						SPI1_MOSI							D12/DA12	TIMx_IC4_15	EVENT OUT
PF0													A0		EVENT OUT



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 Table 6.
 Alternate function input/output (continued)

						Digit	al alterna	te function r	number					
	AFIO0	AFIO1	AFIO2	AFIO3	AFIO4	AFIO5	AFIO6	AFIO7	AFIO8	AFIO10	AFIO11	AFIO12	. AFIO14	AFIO15
Port name	1					I.	Altern	ate function	l I		l		-	
	SYSTEM	TIM2	TIM3/4/5	TIM9/ 10/11	I2C1/2	SPI1/2	SPI3	USART1/2/3	UART4/5	USB	LCD	FSMC/ SDIO	CPRI	SYSTEM
PF1												A1		EVENT OUT
PF2												A2		EVENT OUT
PF3												А3		EVENT OUT
PF4												A4		EVENT OUT
PF5												A5		EVENT OUT
PF6			TIM5_CH1 _ETR										COMP1_INP G11IO1	EVENT OUT
PF7			TIM5_CH2										COMP1_INP G11IO2	EVENT OUT
PF8			TIM5_CH3										COMP1_INP G11IO3	EVENT OUT
PF9			TIM5_CH4										COMP1_INP G11IO4	EVENT OUT
PF10													COMP1_INP G11IO5	EVENT OUT
PF11													COMP1_INP G3IO4	EVENT OUT
PF12												A6	G3IO5	EVENT OUT
PF13												A7	G9IO3	EVENT OUT
PF14												A8	G9IO4	EVENT OUT
PF15												A9	G2IO3	EVENT OUT

 Table 6.
 Alternate function input/output (continued)

	Digital alternate function number													
	AFIO0	AFIO1	AFIO2	AFIO3	AFIO4	AFIO5	AFIO6	AFIO7	AFIO8	AFIO10	AFIO11	AFIO12	AFIO14	AFIO15
Port name							Altern	ate function		•				•
	SYSTEM	TIM2	TIM3/4/5	TIM9/ 10/11	I2C1/2	SPI1/2	SPI3	USART1/2/3	UART4/5	USB	LCD	FSMC/ SDIO	CPRI	SYSTEM
PG0												A10	G2IO4	EVENT OUT
PG1												A11	G2IO5	EVENT OUT
PG2												A12	G7IO5	EVENT OUT
PG3												A13	G7IO6	EVENT OUT
PG4												A14	G7IO7	EVENT OUT
PG5												A15		EVENT OUT
PG6														EVENT OUT
PG7														EVENT OUT
PG8														EVENT OUT
PG9												EBAR1		EVENT OUT
PG10												EBAR2		EVENT OUT
PG11														EVENT OUT
PG12												EBAR3		EVENT OUT
PG13												A24		EVENT OUT
PG14												A25		EVENT OUT



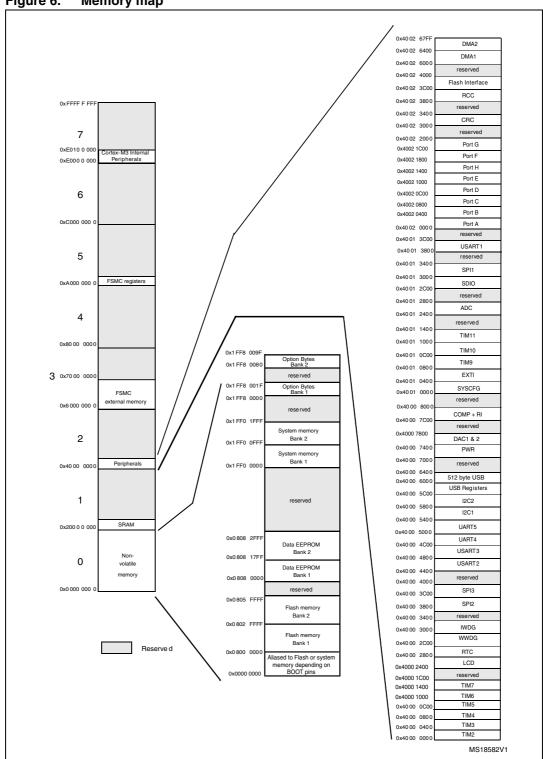


 Table 6.
 Alternate function input/output (continued)

						Digita	al alterna	te function n	number							
	AFIO0	AFIO1	AFIO2	AFIO3	AFIO4	AFIO5	AFIO6	AFIO7	AFIO8	AFIO10	AFIO11	AFIO12	. AFIO14	AFIO15		
Port name		Alternate function														
	SYSTEM	TIM2	TIM3/4/5	TIM9/ 10/11	I2C1/2	SPI1/2	SPI3	USART1/2/3	UART4/5	USB	LCD	FSMC/ SDIO	CPRI	SYSTEM		
PG15														EVENT OUT		
PH0OSC IN	OSC_IN															
PH1OSC _OUT	OSC_OUT															
PH2												A22				

#### 5 **Memory mapping**





# 6 Electrical characteristics

#### 6.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to V<sub>SS</sub>.

#### 6.1.1 Minimum and maximum values

Unless otherwise specified the minimum and maximum values are guaranteed in the worst conditions of ambient temperature, supply voltage and frequencies by tests in production on 100% of the devices with an ambient temperature at  $T_A = 25$  °C and  $T_A = T_A$ max (given by the selected temperature range).

Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes and are not tested in production. Based on characterization, the minimum and maximum values refer to sample tests and represent the mean value plus or minus three times the standard deviation (mean $\pm 3\Sigma$ ).

### 6.1.2 Typical values

Unless otherwise specified, typical data are based on  $T_A = 25$  °C,  $V_{DD} = 3.6$  V (for the 1.65 V  $\leq$ V $_{DD} \leq$ 3.6 V voltage range). They are given only as design guidelines and are not tested.

Typical ADC accuracy values are determined by characterization of a batch of samples from a standard diffusion lot over the full temperature range, where 95% of the devices have an error less than or equal to the value indicated (mean $\pm 2\Sigma$ ).

### 6.1.3 Typical curves

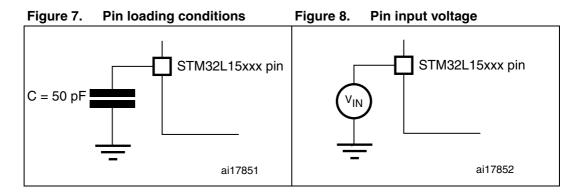
Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

### 6.1.4 Loading capacitor

The loading conditions used for pin parameter measurement are shown in Figure 7.

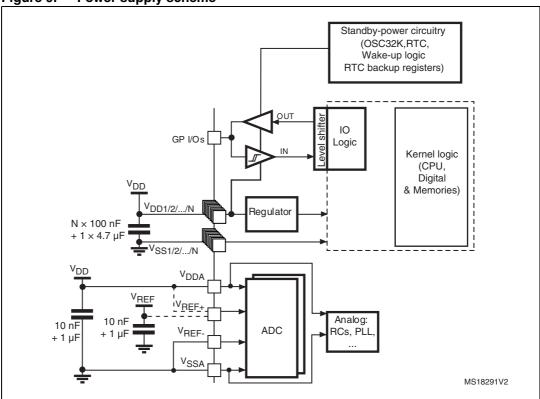
#### 6.1.5 Pin input voltage

The input voltage measurement on a pin of the device is described in Figure 8.



# 6.1.6 Power supply scheme

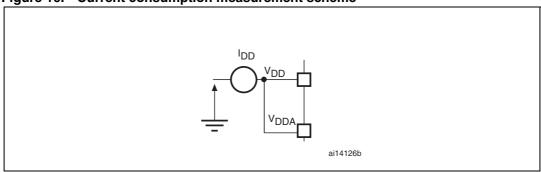
Figure 9. Power supply scheme



Caution: In this figure, the 4.7  $\mu F$  capacitor must be connected to  $V_{DD2}$ .

# 6.1.7 Current consumption measurement

Figure 10. Current consumption measurement scheme



# 6.2 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in *Table 7: Voltage characteristics*, *Table 8: Current characteristics*, and *Table 9: Thermal characteristics* may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Table 7. Voltage characteristics

Symbol	Ratings	Min	Max	Unit
V <sub>DD</sub> -V <sub>SS</sub>	External main supply voltage (including $V_{DDA}$ and $V_{DD}$ ) <sup>(1)</sup>	-0.3	4.0	
V <sub>IN</sub> <sup>(2)</sup>	Input voltage on five-volt tolerant pin	V <sub>SS</sub> -0.3	V <sub>DD</sub> +4.0	V
VIN.	Input voltage on any other pin	V <sub>SS</sub> -0.3	4.0	
l∆V <sub>DDx</sub> l	Variations between different V <sub>DD</sub> power pins		50	mV
IV <sub>SSX</sub> -V <sub>SS</sub> I	Variations between all different ground pins		50	IIIV
V <sub>ESD(HBM)</sub>	Electrostatic discharge voltage (human body model)	see Secti	ion 6.3.11	

All main power (V<sub>DD</sub>, V<sub>DDA</sub>) and ground (V<sub>SS</sub>, V<sub>SSA</sub>) pins must always be connected to the external power supply, in the permitted range.

Table 8. Current characteristics

Symbol	Ratings	Max.	Unit
I <sub>VDD</sub>	Total current into V <sub>DD</sub> /V <sub>DDA</sub> power lines (source) <sup>(1)</sup>	80	
I <sub>vss</sub>	Total current out of V <sub>SS</sub> ground lines (sink) <sup>(1)</sup>	80	
1	Output current sunk by any I/O and control pin	25	
I <sub>IO</sub>	Output current sourced by any I/O and control pin	- 25	mA
(2)	Injected current on five-volt tolerant I/O(3)	+0 /-5	
I <sub>INJ(PIN)</sub> (2)	Injected current on any other pin (4)	± 5	
ΣΙ <sub>ΙΝJ(PIN)</sub>	Total injected current (sum of all I/O and control pins) <sup>(5)</sup>	± 25	

All main power (V<sub>DD</sub>, V<sub>DDA</sub>) and ground (V<sub>SS</sub>, V<sub>SSA</sub>) pins must always be connected to the external power supply, in the permitted range.

- Positive current injection is not possible on these I/Os. A negative injection is induced by V<sub>IN</sub><V<sub>SS</sub>. I<sub>INJ(PIN)</sub> must never be exceeded. Refer to *Table 7* for maximum allowed input voltage values.
- 4. A positive injection is induced by V<sub>IN</sub> > V<sub>DD</sub> while a negative injection is induced by V<sub>IN</sub> < V<sub>SS</sub>. I<sub>INJ(PIN)</sub> must never be exceeded. Refer to *Table 7: Voltage characteristics* for the maximum allowed input voltage values
- When several inputs are submitted to a current injection, the maximum ΣI<sub>INJ(PIN)</sub> is the absolute sum of the
  positive and negative injected currents (instantaneous values).

<sup>2.</sup> V<sub>IN</sub> maximum must always be respected. Refer to *Table 8* for maximum allowed injected current values.

<sup>2.</sup> Negative injection disturbs the analog performance of the device. See note in Section 6.3.17.

Table 9. Thermal characteristics

Symbol	Ratings	Value	Unit
T <sub>STG</sub>	Storage temperature range	-65 to +150	°C
T <sub>J</sub>	Maximum junction temperature	150	°C

# 6.3 Operating conditions

# 6.3.1 General operating conditions

Table 10. General operating conditions

Symbol	Parameter	Conditions	Min	Max	Unit	
f <sub>HCLK</sub>	Internal AHB clock frequency		0	32		
f <sub>PCLK1</sub>	Internal APB1 clock frequency		0	32	MHz	
f <sub>PCLK2</sub>	Internal APB2 clock frequency		0	32		
		BOR detector disabled	1.65	3.6		
V <sub>DD</sub>	Standard operating voltage	BOR detector enabled, at power on	1.8	3.6	V	
		BOR detector disabled, after power on	1.65	.65 3.6		
V <sub>DDA</sub> <sup>(1)</sup>	Analog operating voltage (ADC and DAC not used)	Must be the same voltage	1.65	3.6	V	
V DDA	Analog operating voltage (ADC or DAC used)	as V <sub>DD</sub> <sup>(2)</sup>	1.8	3.6	V	
P <sub>D</sub>	Power dissipation at $T_A = 85  ^{\circ}C^{(3)}$			290	mW	
TA	Tomporatura rango	Maximum power dissipation	-40	85	°C	
IA	Temperature range	Low power dissipation <sup>(4)</sup>	-40	105		
TJ	Junction temperature range	-40 °C ≤T <sub>A</sub> ≤105 °C	-40	105	°C	

<sup>1.</sup> When the ADC is used, refer to Table 59: ADC characteristics.

<sup>2.</sup> It is recommended to power  $V_{DD}$  and  $V_{DDA}$  from the same source. A maximum difference of 300 mV between  $V_{DD}$  and  $V_{DDA}$  can be tolerated during power-up and operation.

<sup>3.</sup> If  $T_A$  is lower, higher  $P_D$  values are allowed as long as  $T_J$  does not exceed  $T_J$  max (see *Table 72: STM32L15xxD Thermal characteristics on page 118*).

<sup>4.</sup> In low power dissipation state,  $T_A$  can be extended to this range as long as  $T_J$  does not exceed  $T_J$  max (see *Table 72: STM32L15xxD Thermal characteristics on page 118*).

Table 11. Functionalities depending on the operating power supply range

	Functi	onalities dep	ending on the op	erating powe	r supply range
Operating power supply range	DAC and ADC operation	USB	V <sub>CORE</sub>	Maximum CPU frequency (f <sub>CPU</sub> max)	I/O operation
V <sub>DD</sub> = 1.65 to 1.8 V	Not functional	Not functional	Range 2 or range 3	16 MHz (1ws) 8MHz (0ws)	Degraded speed performance
V <sub>DD</sub> = 1.8 to 2.0 V	Conversion time up to 500 Ksps	Not functional	Range 2 or range 3	16 MHz (1ws) 8MHz (0ws)	Degraded speed performance
V <sub>DD</sub> = 2.0 to 2.4 V	Conversion time up to 500 Ksps	Functional <sup>(1)</sup>	Range 1, range 2 or range 3	32 MHz (1ws) 16MHz (0ws)	Full speed operation
V <sub>DD</sub> = 2.4 to 3.6 V	Conversion time up to 1 Msps	Functional <sup>(1)</sup>	Range 1, range 2 or range 3	32 MHz (1ws) 16MHz (0ws)	Full speed operation

<sup>1.</sup> To be USB compliant from the IO voltage standpoint, the minimum  $\rm V_{DD}$  is 3.0 V.

# 6.3.2 Embedded reset and power control block characteristics

The parameters given in the following table are derived from the tests performed under the ambient temperature condition summarized in *Table 10*.

Table 12. Embedded reset and power control block characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
	V rice time rate	BOR detector enabled	0		∞		
t <sub>VDD</sub> <sup>(1)</sup>	V <sub>DD</sub> rise time rate BOR detector disabled		0		1000	a/\/	
'VDD' '	V <sub>DD</sub> fall time rate	BOR detector enabled	20		8	— μs/V ∞	
		BOR detector disabled	0		1000		
т . (1)	December of the second	V <sub>DD</sub> rising, BOR enabled		2	3.3	mo	
T <sub>RSTTEMPO</sub> <sup>(1)</sup>	Reset temporization	V <sub>DD</sub> rising, BOR disabled	0.4	0.7	1.6	ms	

Table 12. Embedded reset and power control block characteristics (continued)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
M	Power on/power down reset	Falling edge	1	1.5	1.65	
V <sub>POR/PDR</sub>	threshold	Rising edge	1.3	1.5	1.65	
V	Brown-out reset threshold 0	Falling edge	1.67	1.7	1.74	
$V_{BOR0}$	Brown-out reset tilleshold o	Rising edge	1.69	1.76	1.8	
V	Brown-out reset threshold 1	Falling edge	1.87	1.93	1.97	
V <sub>BOR1</sub>	Brown-out reset tilleshold i	Rising edge	1.96	2.03	2.07	
V.	Brown-out reset threshold 2	Falling edge	2.22	2.30	2.35	
$V_{BOR2}$	Brown-out reset tilleshold 2	Rising edge	2.31	2.41	2.44	
V	Drown out road throshold 0	Falling edge	2.45	2.55	2.60	
$V_{BOR3}$	Brown-out reset threshold 3	Rising edge	2.54	2.66	2.7	
V	Drown out road throabold 4	Falling edge	2.68	2.8	2.85	
$V_{BOR4}$	Brown-out reset threshold 4	Rising edge	2.78	2.9	2.95	
V	Programmable voltage detector	Falling edge	1.8	1.85	1.88	V
$V_{PVD0}$	threshold 0	Rising edge	1.88	1.94	1.99	v
V	DVD three should 1	Falling edge	1.98	2.04	2.09	
V <sub>PVD1</sub>	PVD threshold 1	Rising edge	2.08	2.14	2.18	
V	DVD three should 0	Falling edge	2.20	2.24	2.28	
$V_{PVD2}$	PVD threshold 2	Rising edge	2.28	2.34	2.38	
V	DVD throohold 2	Falling edge	2.39	2.44	2.48	
$V_{PVD3}$	PVD threshold 3	Rising edge	2.47	2.54	2.58	
V	DVD throohold 4	Falling edge	2.57	2.64	2.69	
$V_{PVD4}$	PVD threshold 4	Rising edge	2.68	2.74	2.79	
W	DVD three should 5	Falling edge	2.77	2.83	2.88	
$V_{PVD5}$	PVD threshold 5	Rising edge	2.87	2.94	2.99	
M	DVD throohold 6	Falling edge	2.97	3.05	3.09	
$V_{PVD6}$	PVD threshold 6	Rising edge	3.08	3.15	3.20	
		BOR0 threshold		40		
$V_{hyst}$	Hysteresis voltage	All BOR and PVD thresholds excepting BOR0		100		mV

<sup>1.</sup> Guaranteed by characterisation, not tested in production.

# 6.3.3 Embedded internal reference voltage

The parameters given in *Table 13* are based on characterization results, unless otherwise specified.

Table 13. Embedded internal reference voltage

Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
V <sub>REFINT out</sub> <sup>(1)</sup>	Internal reference voltage	- 40 °C < T <sub>J</sub> < +105 °C	1.202	1.224	1.242	V	
I <sub>REFINT</sub>	Internal reference current consumption			1.4	2.3	μΑ	
T <sub>VREFINT</sub>	Internal reference startup time			2	3	ms	
V <sub>VREF_MEAS</sub>	V <sub>DDA</sub> and V <sub>REF+</sub> voltage during V <sub>REFINT</sub> factory measure		2.99	3	3.01	V	
A <sub>VREF_MEAS</sub>	Accuracy of factory-measured V <sub>REF</sub> value <sup>(2)</sup>	Including uncertainties due to ADC and V <sub>DDA</sub> /V <sub>REF+</sub> values			±5	mV	
T <sub>Coeff</sub> <sup>(3)</sup>	Temperature coefficient	-40 °C < T <sub>J</sub> < +105 °C		20	50	nnm/°C	
	Temperature coemicient	0 °C < T <sub>J</sub> < +50 °C			20	ppm/°C	
A <sub>Coeff</sub> <sup>(3)</sup>	Long-term stability	1000 hours, T= 25 °C			1000	ppm	
VDDCoeff <sup>(3)</sup>	Voltage coefficient	3.0 V < V <sub>DDA</sub> < 3.6 V			2000	ppm/V	
T <sub>S_vrefint</sub> (3)(4)	ADC sampling time when reading the internal reference voltage			5	10	μs	
T <sub>ADC_BUF</sub> (3)	Startup time of reference voltage buffer for ADC				10	μs	
I <sub>BUF_ADC</sub> (3)	Consumption of reference voltage buffer for ADC			13.5	25	μΑ	
I <sub>VREF_OUT</sub> (3)	VREF_OUT output current <sup>(5)</sup>				1	μΑ	
C <sub>VREF_OUT</sub> <sup>(3)</sup>	VREF_OUT output load				50	pF	
I <sub>LPBUF</sub> <sup>(3)</sup>	Consumption of reference voltage buffer for VREF_OUT and COMP			730	1200	nA	
V <sub>REFINT_DIV1</sub> (3)	1/4 reference voltage		24	25	26		
V <sub>REFINT_DIV2</sub> (3)	1/2 reference voltage		49	50	51	% V <sub>REFINT</sub>	
V <sub>REFINT_DIV3</sub> <sup>(3)</sup>	3/4 reference voltage		74	75	76		

<sup>1.</sup> Tested in production;

<sup>2.</sup> The internal  $V_{\mathsf{REF}}$  value is individually measured in production and stored in dedicated EEPROM bytes.

<sup>3.</sup> Guaranteed by design, not tested in production.

<sup>4.</sup> Shortest sampling time can be determined in the application by multiple iterations.

<sup>5.</sup> To guarantee less than 1% VREF\_OUT deviation.

### 6.3.4 Supply current characteristics

The current consumption is a function of several parameters and factors such as the operating voltage, ambient temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code. The current consumption is measured as described in *Figure 10: Current consumption measurement scheme*.

All Run-mode current consumption measurements given in this section are performed with a reduced code that gives a consumption equivalent to Dhrystone 2.1 code.

#### **Maximum current consumption**

The MCU is placed under the following conditions:

- V<sub>DD</sub> = 3.6 V
- All I/O pins are in input mode with a static value at V<sub>DD</sub> or V<sub>SS</sub> (no load)
- All peripherals are disabled except when explicitly mentioned
- The Flash memory access time is adjusted depending on f<sub>HCLK</sub> frequency and voltage range
- Prefetch and 64-bit access are enabled in configurations with 1 wait state

The parameters given in *Table 14*, *Table 10* and *Table 12* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 10*.

Table 14. Current consumption in Run mode, code with data processing running from Flash

Symbol	Parameter	Conditions			Tvn	Max <sup>(1)</sup>			Unit
Syllibol	Parameter	Cond	illions	fHCLK	Тур	55 °C	85 °C	105 °C	Oilit
			Range 3,	1 MHz	360	500	500	500	
			V <sub>CORE</sub> =1.2 V	2 MHz	620	750	750	750	μΑ
		f <sub>HSE</sub> = f <sub>HCLK</sub> up to 16 MHz, included f <sub>HSE</sub> = f <sub>HCLK</sub> /2	VOS[1:0] = 11	4 MHz	1070	1200	1200	1200	
			Range 2,	4 MHz	1.30	1.6	1.6	1.6	
			V <sub>CORE</sub> =1.5 V	8 MHz	2.4	2.9	2.9	2.9	
	above 16 MHz	VOS[1:0] = 10	16 MHz	4.6	5.2	5.2	5.2		
Supply	(PLL ON) <sup>(2)</sup>	Range 1,	8 MHz	2.9	3.5	3.5	3.5		
I <sub>DD</sub>	current in Run mode,		V <sub>CORE</sub> =1.8 V VOS[1:0] = 01	16 MHz	5.7	6.5	6.5	6.5	
(Run from	code			32 MHz	10.4	12	12	12	
Flash)	executed from Flash	HSI clock source	Range 2, V <sub>CORE</sub> =1.5 V VOS[1:0] = 10	16 MHz	4.5	5.2	5.2	5.2	mA
	(16 MHz)	Range 1, V <sub>CORE</sub> =1.8 V VOS[1:0] = 01	32 MHz	10.9	12.3	12.3	12.3		
	MSI clock, 65 kHz MSI clock, 524 kHz	Range 3,	65 kHz	0.05	0.079	0.092	0.13		
		V <sub>CORE</sub> =1.2 V	524 kHz	0.17	0.2	0.21	0.25		
		MSI clock, 4.2 MHz	VOS[1:0] = 11	4.2 MHz	1.0	1.1	1.1	1.2	

<sup>1.</sup> Based on characterization, not tested in production, unless otherwise specified.

<sup>2.</sup> Oscillator bypassed (HSEBYP = 1 in RCC\_CR register).

Table 15. Current consumption in Run mode, code with data processing running from RAM

Symbol	Parameter	Conditions		f <sub>HCLK</sub>	Тур	Max <sup>(1)</sup>			Unit
Symbol	Farameter					55 °C	85 °C	105 °C	Oille
			Range 3,	1 MHz	310	470	470	470	
			V <sub>CORE</sub> =1.2 V	2 MHz	590	780	780	780	μΑ
		$f_{HSE} = f_{HCLK}$ up to 16 MHz, included $f_{HSE} = f_{HCLK}/2$ above 16 MHz	VOS[1:0] = 11	4 MHz	1030	1200	1200	1200 <sup>(3)</sup>	
			Range 2,	4 MHz	1.2	1.5	1.5	1.5	
			V <sub>CORE</sub> =1.5 V	8 MHz	2.3	3	3	3	
			VOS[1:0] = 10	16 MHz	4.3	5	5	5	
Supply current in Run mode,	(PLL ON) <sup>(2)</sup>	Range 1, V <sub>CORE</sub> =1.8 V	8 MHz	2.7	3.5	3.5	3.5		
			16 MHz	5.0	5.55	5.55	5.55		
(Run from	code executed from RAM,	\	VOS[1:0] = 01	32 MHz	9.8	10.9	10.9	10.9	mA
RAM)	Flash switched off	HSI clock source	Range 2, V <sub>CORE</sub> =1.5 V VOS[1:0] = 10	16 MHz	4.3	4.8	4.8	4.8	
	(16 MHz)	Range 1, V <sub>CORE</sub> =1.8 V VOS[1:0] = 01	32 MHz	10.1	11.7	11.7	11.7		
	MSI clock, 65 kHz	Range 3,	65 kHz	40	48.5	63	100		
	<u> </u>	MSI clock, 524 kHz	V <sub>CORE</sub> =1.2 V	524 kHz	148	175	183	215	μΑ
		VOS[1:0] = 11	4.2 MHz	990	1032	1034	1100		

<sup>1.</sup> Based on characterization, not tested in production, unless otherwise specified.

<sup>2.</sup> Oscillator bypassed (HSEBYP = 1 in RCC\_CR register).

<sup>3.</sup> Tested in production.

Table 16. Current consumption in Sleep mode

Cumbal	Davamatav	0	liai o mo		T		Max <sup>(1</sup>	)	l lasia
Symbol	Parameter	Cond	litions	fHCLK	Тур	55 °C	85 °C	105 °C	Unit
			Range 3,	1 MHz	180	220	220	220	
			V <sub>CORE</sub> =1.2 V	2 MHz	225	300	300	300	
		f <sub>HSE</sub> = f <sub>HCLK</sub>	VOS[1:0] = 11	4 MHz	300	380	380	380 <sup>(3)</sup>	
		up to 16 MHz,	Range 2,	4 MHz	360	500	500	500	
		included f <sub>HSE</sub> = f <sub>HCLK</sub> /2	V <sub>CORE</sub> =1.5 V	8 MHz	570	700	700	700	
	Supply current in	above 16 MHz	VOS[1:0] = 10	16 MHz	990	1100	1100	1100	
	Sleep	(PLL ON) <sup>(2)</sup>	Range 1,	8 MHz	675	800	800	800	
	mode, code		V <sub>CORE</sub> =1.8 V	16 MHz	1150	1250	1250	1250	
	executed		VOS[1:0] = 01	32 MHz	2300	2700	2700	2700	μΑ
from RAM, Flash switched OFF	HSI clock source	Range 2, V <sub>CORE</sub> =1.5 V VOS[1:0] = 10	16 MHz	1025	1100	1100	1100		
	(16 MHz)	Range 1, V <sub>CORE</sub> =1.8 V VOS[1:0] = 01	32 MHz	2460	2700	2700	2700		
_		MSI clock, 65 kHz	Range 3,	65 kHz	30	36	46	72	
I <sub>DD</sub> (Sleep)	MSI clock, 524 kHz	V <sub>CORE</sub> =1.2 V	524 kHz	50	58	67	92		
	MSI clock, 4.2 MHz	VOS[1:0] = 11	4.2 MHz	210	245	251	273		
			Range 3,	1 MHz	190	250	250	250	
			V <sub>CORE</sub> =1.2 V VOS[1:0] = 11	2 MHz	235	300	300	300	
				4 MHz	315	380	380	380	
		HSE = 16 MHz <sup>(2)</sup>	Range 2, V <sub>CORE</sub> =1.5 V VOS[1:0] = 10	4 MHz	390	500	500	500	
	Supply	(PLL ON for f <sub>HCLK</sub>		8 MHz	600	700	700	700	
	current in	above 16 MHz)		16 MHz	1000	1120	1120	1120	
	Sleep mode,		Range 1,	8 MHz	690	800	800	800	μΑ
	code		V <sub>CORE</sub> =1.8 V	16 MHz	1160	1300	1300	1300	r
	executed from Flash		VOS[1:0] = 01	32 MHz	2310	2700	2700	2700	
	nom ridon	HSI clock source	Range 2, V <sub>CORE</sub> =1.5 V VOS[1:0] = 10	16 MHz	1040	1160	1160	1160	
		(16 MHz)	Range 1, V <sub>CORE</sub> =1.8 V VOS[1:0] = 01	32 MHz	2500	2800	2800	2800	
	Supply	MSI clock, 65 kHz		65 kHz	42	50	60	90	
	current in Sleep	MSI clock, 524 kHz	Range 3,	524 kHz	63	72	82	110	
Sleep) Sl Sleep) co ex	mode, code executed from Flash	MSI clock, 4.2 MHz	V <sub>CORE</sub> =1.2V VOS[1:0] = 11	4.2 MHz	230	263	265	290	μΑ

- 1. Based on characterization, not tested in production, unless otherwise specified.
- 2. Oscillator bypassed (HSEBYP = 1 in RCC\_CR register)
- 3. Tested in production.

Table 17. Current consumption in Low power run mode

Symbol	Parameter		Conditions		Тур	Max (1)	Unit	
				T <sub>A</sub> = -40 °C to 25 °C	11	14		
		All	MSI clock, 65 kHz f <sub>HCLK</sub> = 32 kHz	T <sub>A</sub> = 85 °C	26	32		
		peripherals	HOLK 92 W.E	T <sub>A</sub> = 105 °C	53	72		
		OFF, code executed		T <sub>A</sub> =-40 °C to 25 °C	18	21		
		from RAM,	MSI clock, 65 kHz f <sub>HCLK</sub> = 65 kHz	T <sub>A</sub> = 85 °C	33	40		
		Flash switched	HOLK OF THE	T <sub>A</sub> = 105 °C	60	78		
		OFF, V <sub>DD</sub>	MSI clock, 131 kHz	$T_A = -40 ^{\circ}\text{C} \text{ to } 25 ^{\circ}\text{C}$	36	41		
		from 1.65 V		T <sub>A</sub> = 55 °C	39	44		
	Supply	to 3.6 V	f <sub>HCLK</sub> = 131 kHz	f <sub>HCLK</sub> = 131 kHz	T <sub>A</sub> = 85 °C	50	58	1
I <sub>DD</sub>	current in			T <sub>A</sub> = 105 °C	78	95		
(LP Run)	Low power run mode			$T_A = -40 ^{\circ}\text{C} \text{ to } 25 ^{\circ}\text{C}$	36	40.5		
	run mode		MSI clock, 65 kHz f <sub>HCLK</sub> = 32 kHz	T <sub>A</sub> = 85 °C	53	60	μΑ	
		All	HOLK	T <sub>A</sub> = 105 °C	81	100		
		peripherals OFF, code		$T_A = -40 ^{\circ}\text{C} \text{ to } 25 ^{\circ}\text{C}$	44	49		
		executed	MSI clock, 65 kHz f <sub>HCLK</sub> = 65 kHz	T <sub>A</sub> = 85 °C	61	67		
		from Flash,	HOLK OF WIL	T <sub>A</sub> = 105 °C	89	107		
		V <sub>DD</sub> from 1.65 V to		$T_A = -40 ^{\circ}\text{C} \text{ to } 25 ^{\circ}\text{C}$	64	71		
		3.6 V	MSI clock, 131 kHz	T <sub>A</sub> = 55 °C	68	73		
			f <sub>HCLK</sub> = 131 kHz	T <sub>A</sub> = 85 °C	80	88		
				T <sub>A</sub> = 105 °C	101	110		
I <sub>DD</sub> max (LP Run)	Max allowed current in Low power run mode	V <sub>DD</sub> from 1.65 V to 3.6 V				200		

<sup>1.</sup> Based on characterization, not tested in production, unless otherwise specified.

Table 18. Current consumption in Low power sleep mode

Symbol	Parameter		Conditions				Unit
			MSI clock, 65 kHz f <sub>HCLK</sub> = 32 kHz Flash OFF	T <sub>A</sub> = -40 °C to 25 °C	4.4		
			MSI clock, 65 kHz	$T_A = -40 ^{\circ}\text{C} \text{ to } 25 ^{\circ}\text{C}$	18	21	
			f <sub>HCLK</sub> = 32 kHz	T <sub>A</sub> = 85 °C	24	27	
		All	Flash ON	T <sub>A</sub> = 105 °C	35	43	
		peripherals OFF, V <sub>DD</sub>	IVISI CIOCK, OS KI IZ	$T_A = -40  ^{\circ}\text{C} \text{ to } 25  ^{\circ}\text{C}$	18.6	21	
		from 1.65 V		T <sub>A</sub> = 85 °C	24.5	28	
		to 3.6 V		T <sub>A</sub> = 105 °C	35	42	
			$T_A = -40 ^{\circ}\text{C} \text{ to } 25 ^{\circ}\text{C}$	22	25		
	I <sub>DD</sub> Supply current in Low power sleep		MSI clock, 131 kHz f <sub>HCLK</sub> = 131 kHz,	T <sub>A</sub> = 55 °C	23.5	26	-
I <sub>DD</sub>			Flash ON	T <sub>A</sub> = 85 °C	28.5	31	
(LP Sleep)				T <sub>A</sub> = 105 °C	39	45	
	mode	TIM9 and	MSI clock, 65 kHz f <sub>HCLK</sub> = 32 kHz	$T_A = -40 ^{\circ}\text{C} \text{ to } 25 ^{\circ}\text{C}$	18	20.5	
				T <sub>A</sub> = 85 °C	24	27	μA
				T <sub>A</sub> = 105 °C	35	43	]
		USART1		$T_A = -40 ^{\circ}\text{C} \text{ to } 25 ^{\circ}\text{C}$	18.6	21	
		enabled, Flash ON,	MSI clock, 65 kHz f <sub>HCLK</sub> = 65 kHz	T <sub>A</sub> = 85 °C	24.5	28	
		V <sub>DD</sub> from	HOLK SS III I	T <sub>A</sub> = 105 °C	35	42	
		1.65 V to		$T_A = -40  ^{\circ}\text{C} \text{ to } 25  ^{\circ}\text{C}$	22	25	
		3.6 V	MSI clock, 131 kHz	T <sub>A</sub> = 55 °C	23.5	26	
			f <sub>HCLK</sub> = 131 kHz	T <sub>A</sub> = 85 °C	28.5	31	
				T <sub>A</sub> = 105 °C	39	45	
I <sub>DD</sub> max (LP Sleep)	Max allowed current in Low power Sleep mode	V <sub>DD</sub> from 1.65 V to 3.6 V				200	

<sup>1.</sup> Based on characterization, not tested in production, unless otherwise specified.

Table 19. Typical and maximum current consumptions in Stop mode

Symbol	Parameter	Conditions				Max <sup>(1)</sup>	Unit
				$T_A = -40^{\circ}C$ to 25°C	1.7	4	
			LCD OFF	T <sub>A</sub> = 55°C	2.7	6	•
			LCD OFF	T <sub>A</sub> = 85°C	7	10	•
				T <sub>A</sub> = 105°C	15	23	•
		RTC clocked by LSI,		$T_A = -40$ °C to 25°C	3.8	6	
		regulator in LP mode, HSI and HSE OFF	LCD ON	T <sub>A</sub> = 55°C	4.7	7	•
		(no independent	(static duty) <sup>(2)</sup>	T <sub>A</sub> = 85°C	9	12	μА
		watchdog)		T <sub>A</sub> = 105°C	19	27	•
				$T_A = -40^{\circ}C$ to 25°C	7.8	10	
			LCD ON (1/8	T <sub>A</sub> = 55°C	8.5	11	•
			duty) <sup>(3)</sup>	T <sub>A</sub> = 85°C	13	16	•
				T <sub>A</sub> = 105°C	26	44	•
	Supply current in Stop mode with RTC enabled			$T_A = -40^{\circ}C$ to 25°C	TBD	TBD	
I <sub>DD</sub>			LCD OFF	T <sub>A</sub> = 55°C	TBD	TBD	μΑ
(Stop with			LOD OFF	T <sub>A</sub> = 85°C	TBD	TBD	
RTC)				T <sub>A</sub> = 105°C	TBD	TBD	
		KHZ), regulator in LP (s		$T_A = -40^{\circ}C$ to $25^{\circ}C$	TBD	TBD	
			LCD ON (static	T <sub>A</sub> = 55°C	TBD	TBD	
		mode, HSI and HSE	duty) <sup>(2)</sup>		TBD	TBD	
		OFF (no independent watchdog)		T <sub>A</sub> = 105°C	TBD	TBD	μΑ
			LCD ON (1/8	$T_A = -40^{\circ}C$ to $25^{\circ}C$	TBD	TBD	
				T <sub>A</sub> = 55°C	TBD	TBD	
			duty) <sup>(3)</sup>	T <sub>A</sub> = 85°C	TBD	TBD	
				T <sub>A</sub> = 105°C	TBD	TBD	
		DTO also de la 150 de		$T_A = -40$ °C to 25°C $V_{DD} = 1.8V$	TBD		
		RTC clocked by LSE (no independent watchdog) <sup>(4)</sup>	LCD OFF	$T_A = -40$ °C to 25°C $V_{DD} = 3.0V$	TBD		
		σ,		$T_A = -40$ °C to 25°C $V_{DD} = 3.6V$	TBD		
		Regulator in LP mode, I HSE OFF, independent and LSI enabled		$T_A = -40^{\circ}\text{C to } 25^{\circ}\text{C}$	1.6	2.2	
I <sub>DD</sub>	Supply current in Stop mode (			$T_A = -40^{\circ}C$ to 25°C	0.65	1	μΑ
(Stop)	RTC disabled)	Regulator in LP mode, I		T <sub>A</sub> = 55°C	1.6	3	, h., ,
		and HSE OFF (no inderwatchdog)	Jenueni	T <sub>A</sub> = 85°C	6	9	•
				T <sub>A</sub> = 105°C	14	22 <sup>(5)</sup>	•

Table 19. Typical and maximum current consumptions in Stop mode (continued)

Symbol	Parameter	Conditions	Тур	Max <sup>(1)</sup>	Unit	
I <sub>DD</sub>	Supply current	MSI = 4.2 MHz		TBD		
(WU	WU during wakeup	MSI = 1.05 MHz	$T_A = -40$ °C to 25°C	TBD		mA
Stop)	from Stop mode	MSI = 65 kHz		TBD		

- 1. Based on characterization, not tested in production, unless otherwise specified
- 2. LCD enabled with external VLCD, static duty, division ratio = 256, all pixels active, no LCD connected
- 3. LCD enabled with external VLCD, 1/4 duty, 1/3 bias, division ratio = 64, all pixels active, no LCD connected.
- 4. Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.
- 5. Tested in production.

Table 20. Typical and maximum current consumptions in Standby mode

Symbol	Parameter	Conditions		Тур	Max <sup>(1)</sup>	Unit
			T <sub>A</sub> = -40 °C to 25 °C	1.3	1.9	
		RTC clocked by LSI (no	T <sub>A</sub> = 55 °C	1.5	2.2	
		independent watchdog)	T <sub>A</sub> = 85 °C	2.15	4	
I <sub>DD</sub>	Supply current in Standby		T <sub>A</sub> = 105 °C	3.8	8.3 <sup>(2)</sup>	
(Standby) with RTC)	mode with RTC enabled		$T_A = -40 ^{\circ}\text{C} \text{ to } 25 ^{\circ}\text{C}$	TBD		
		RTC clocked by LSE (no	T <sub>A</sub> = 55 °C	TBD		
		independent watchdog) <sup>(3)</sup>	T <sub>A</sub> = 85 °C	TBD		
			T <sub>A</sub> = 105 °C	TBD		
		Independent watchdog and LSI enabled	T <sub>A</sub> = -40 °C to 25 °C	1.3	1.7	μΑ
I <sub>DD</sub>	Supply current in Standby		T <sub>A</sub> = -40 °C to 25 °C	0.35	0.6	
		Independent watchdog and	T <sub>A</sub> = 55 °C	0.47	0.9	
		LSI OFF	T <sub>A</sub> = 85 °C	1.2	2.75	
			T <sub>A</sub> = 105 °C	2.9	7 <sup>(2)</sup>	
I <sub>DD</sub> (WU from Standby)	Supply current during wakeup from Standby mode		T <sub>A</sub> = -40 °C to 25 °C	TBD		

- 1. Based on characterization, not tested in production, unless otherwise specified
- 2. Tested in production.
- Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8pF loading capacitors.

### Wakeup time from low-power mode

The wakeup times given in the following table are measured with the MSI RC oscillator. The clock source used to wake up the device depends on the current operating mode:

- Sleep mode: the clock source is the clock that was set before entering Sleep mode
- Stop mode: the clock source is the MSI oscillator in the range configured before entering Stop mode
- Standby mode: the clock source is the MSI oscillator running at 2.1 MHz

All timings are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 10*.

Table 21. Typical and maximum timings in Low power modes

Symbol	Parameter	Conditions	Тур	Max <sup>(1)</sup>	Unit
t <sub>WUSLEEP</sub>	Wakeup from Sleep mode	f <sub>HCLK</sub> = 32 MHz	0.4		
t	Wakeup from Low power sleep mode	f <sub>HCLK</sub> = 262 kHz Flash enabled	46		
twusleep_lp	$f_{HCLK} = 262 \text{ kHz}$	f <sub>HCLK</sub> = 262 kHz Flash switched OFF	46		
	Wakeup from Stop mode, regulator in Run mode	f <sub>HCLK</sub> = f <sub>MSI</sub> = 4.2 MHz	TBD		
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 4.2 MHz Voltage range 1 and 2	7.7	8.9	
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 4.2 MHz Voltage range 3	8.2	13.1	μs
t <sub>WUSTOP</sub>		f <sub>HCLK</sub> = f <sub>MSI</sub> = 2.1 MHz	10.2	13.4	
	regulator in low power mode	$f_{HCLK} = f_{MSI} = 1.05 \text{ MHz}$	16	20	
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 524 kHz	31	37	
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 262 kHz	57	66	
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 131 kHz	112	123	
		f <sub>HCLK</sub> = MSI = 65 kHz	221	236	
tw//otppy	Wakeup from Standby mode FWU bit = 1	THCLK = MSI = 2.1 MHZ	58	104	
t <sub>wustdby</sub>	Wakeup from Standby mode FWU bit = 0	f <sub>HCLK</sub> = MSI = 2.1 MHz	2.6	3.25	ms

<sup>1.</sup> Based on characterization, not tested in production, unless otherwise specified

# On-chip peripheral current consumption

The current consumption of the on-chip peripherals is given in the following table. The MCU is placed under the following conditions:

- $\bullet \quad$  all I/O pins are in input mode with a static value at  $V_{DD}$  or  $V_{SS}$  (no load)
- all peripherals are disabled unless otherwise mentioned
- the given value is calculated by measuring the current consumption
  - with all peripherals clocked off
  - with only one peripheral clocked on

Table 22. Peripheral current consumption<sup>(1)</sup>

		Typical o	consumption,	V <sub>DD</sub> = 3.0 V, T	A = 25 °C	
Peripheral		Range 1, V <sub>CORE</sub> = 1.8 V VOS[1:0] = 01	Range 2, V <sub>CORE</sub> = 1.5 V VOS[1:0] = 10	Range 3, V <sub>CORE</sub> = 1.2 V VOS[1:0] = 11	Low power sleep and run	Unit
	TIM2	TBD	TBD	TBD	TBD	
	TIM3	TBD	TBD	TBD	TBD	
	TIM4	TBD	TBD	TBD	TBD	
	TIM6	TBD	TBD	TBD	TBD	
	TIM7	TBD	TBD	TBD	TBD	
	LCD	TBD	TBD	TBD	TBD	
	WWDG	TBD	TBD	TBD	TBD	
APB1	SPI2	TBD	TBD	TBD	TBD	μΑ/MHz
AFDI	USART2	TBD	TBD	TBD	TBD	(f <sub>HCLK</sub> )
	USART3	TBD	TBD	TBD	TBD	
	I2C1	TBD	TBD	TBD	TBD	
	I2C2	TBD	TBD	TBD	TBD	
	USB	TBD	TBD	TBD	TBD	
	PWR	TBD	TBD	TBD	TBD	
	DAC	TBD	TBD	TBD	TBD	
	COMP	TBD	TBD	TBD	TBD	

Table 22. Peripheral current consumption<sup>(1)</sup> (continued)

		Typical o	onsumption,	V <sub>DD</sub> = 3.0 V, T	<sub>A</sub> = 25 °C	
Peripheral		Range 1, V <sub>CORE</sub> = 1.8 V VOS[1:0] = 01	Range 2, V <sub>CORE</sub> = 1.5 V VOS[1:0] = 10	Range 3, V <sub>CORE</sub> = 1.2 V VOS[1:0] = 11	Low power sleep and run	Unit
	SYSCFG & RI	TBD	TBD	TBD	TBD	
	TIM9	TBD	TBD	TBD	TBD	
APB2	TIM10	TBD	TBD	TBD	TBD	
	TIM11	TBD	TBD	TBD	TBD	
	ADC <sup>(2)</sup>	TBD	TBD	TBD	TBD	
	SPI1	TBD	TBD	TBD	TBD	
	USART1	TBD	TBD	TBD	TBD	
	GPIOA	TBD	TBD	TBD	TBD	μΑ/MHz
	GPIOB	TBD	TBD	TBD	TBD	(f <sub>HCLK</sub> )
	GPIOC	TBD	TBD	TBD	TBD	
	GPIOD	TBD	TBD	TBD	TBD	
AHB	GPIOE	TBD	TBD	TBD	TBD	
	GPIOF	TBD	TBD	TBD	TBD	
	CRC	TBD	TBD	TBD	TBD	
	FLASH	TBD	TBD	TBD	TBD	
	DMA1	TBD	TBD	TBD	TBD	
All enabled	•	TBD	TBD	TBD	TBD	
I <sub>DD (RTC)</sub>			TE	3D		
I <sub>DD (LCD)</sub>			TE	3D		
I <sub>DD (ADC)</sub> <sup>(3)</sup>			TE	3D		
I <sub>DD (DAC)</sub> <sup>(4)</sup>			TE	3D		
I <sub>DD (COMP1)</sub>			TE	3D		μΑ
1	Slow mode		TE	3D		
I <sub>DD</sub> (COMP2)	Fast mode		TE	3D		
I <sub>DD (PVD / BOF</sub>	(5)		TE	3D		
I <sub>DD (IWDG)</sub>			TE	3D		

Data based on differential I<sub>DD</sub> measurement between all peripherals OFF an one peripheral with clock enabled, in the following conditions: f<sub>HCLK</sub> = 32 MHz (range 1), f<sub>HCLK</sub> = 16 MHz (range 2), f<sub>HCLK</sub> = 4 MHz (range 3), f<sub>HCLK</sub> = 64kHz (Low power run/sleep), f<sub>APB1</sub> = f<sub>HCLK</sub>, f<sub>APB2</sub> = f<sub>HCLK</sub>, default prescaler value for each peripheral. The CPU is in Sleep mode in both cases. No I/O pins toggling. Not tested in production.

<sup>2.</sup> HSI oscillator is OFF for this measure.

Data based on a differential IDD measurement between ADC in reset configuration and continuous ADC conversion (HSI consumption not included).

- 4. Data based on a differential IDD measurement between DAC in reset configuration and continuous DAC conversion of VDD/2. DAC is in buffered mode, output is left floating.
- 5. Including supply current of internal reference voltage.

# 6.3.5 External clock source characteristics

### High-speed external user clock generated from an external source

Table 23. High-speed external user clock characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>HSE_ext</sub>	User external clock source frequency		1	8	32	MHz
V <sub>HSEH</sub>	OSC_IN input pin high level voltage		0.7V <sub>DD</sub>		$V_{DD}$	V
V <sub>HSEL</sub>	OSC_IN input pin low level voltage		$V_{SS}$		0.3V <sub>DD</sub>	V
$t_{w(HSE)} \ t_{w(HSE)}$	OSC_IN high or low time		12			ns
t <sub>r(HSE)</sub>	OSC_IN rise or fall time				20	115
C <sub>in(HSE)</sub>	OSC_IN input capacitance			2.6		pF
DuCy <sub>(HSE)</sub>	Duty cycle		45		55	%
IL	OSC_IN Input leakage current	$V_{SS} \leq V_{IN} \leq V_{DD}$			±1	μΑ

<sup>1.</sup> Guaranteed by design, not tested in production.

### Low-speed external user clock generated from an external source

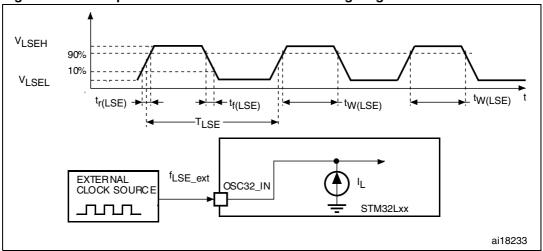
The characteristics given in the following table result from tests performed using a low-speed external clock source, and under ambient temperature and supply voltage conditions summarized in *Table 10*.

Table 24. Low-speed external user clock characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>LSE_ext</sub>	User external clock source frequency		1	32.768	1000	kHz
V <sub>LSEH</sub>	OSC32_IN input pin high level voltage		0.7V <sub>DD</sub>		V <sub>DD</sub>	V
V <sub>LSEL</sub>	OSC32_IN input pin low level voltage		V <sub>SS</sub>		0.3V <sub>DD</sub>	V
t <sub>w(LSE)</sub>	OSC32_IN high or low time		TBD			ns
$\begin{matrix} t_{r(LSE)} \\ t_{f(LSE)} \end{matrix}$	OSC32_IN rise or fall time				TBD	113
C <sub>IN(LSE)</sub>	OSC32_IN input capacitance			0.6		pF
DuCy <sub>(LSE)</sub>	Duty cycle		TBD		TBD	%
ΙL	OSC32_IN Input leakage current	$V_{SS} \leq V_{IN} \leq V_{DD}$			±1	μΑ

<sup>1.</sup> Guaranteed by design, not tested in production

Figure 11. Low-speed external clock source AC timing diagram



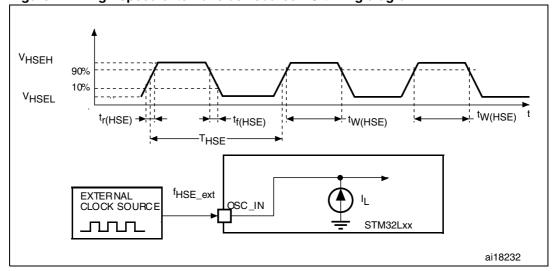


Figure 12. High-speed external clock source AC timing diagram

#### High-speed external clock generated from a crystal/ceramic resonator

The high-speed external (HSE) clock can be supplied with a 1 to 24 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in *Table 25*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
f <sub>OSC_IN</sub>	Oscillator frequency		1		24	MHz	
R <sub>F</sub>	Feedback resistor			200		kΩ	
С	Recommended load capacitance versus equivalent serial resistance of the crystal (R <sub>S</sub> ) <sup>(3)</sup>	R <sub>S</sub> = 30 Ω		20		pF	
I <sub>HSE</sub>	HSE driving current	$V_{DD}$ = 3.3 V, $V_{IN}$ = $V_{SS}$ with 30 pF load			3	mA	
1	HSE oscillator power	C = 20 pF $f_{OSC} = 16 MHz$			2.5 (startup) 0.7 (stabilized)	mΛ	
IDD(HSE)	consumption	C = 10 pF $f_{OSC} = 16 MHz$			2.5 (startup) 0.46 (stabilized)	mA	
9 <sub>m</sub>	Oscillator transconductance	Startup	3.5			mA /V	
t <sub>SU(HSE)</sub>	Startup time	V <sub>DD</sub> is stabilized		1		ms	

Table 25. HSE 1-24 MHz oscillator characteristics<sup>(1)(2)</sup>

- 1. Resonator characteristics given by the crystal/ceramic resonator manufacturer.
- 2. Based on characterization results, not tested in production.
- 3. The relatively low value of the RF resistor offers a good protection against issues resulting from use in a humid environment, due to the induced leakage and the bias condition change. However, it is recommended to take this point into account if the MCU is used in tough humidity conditions.
- t<sub>SU(HSE)</sub> is the startup time measured from the moment it is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer.

For  $C_{L1}$  and  $C_{L2}$ , it is recommended to use high-quality external ceramic capacitors in the 5 pF to 25 pF range (typ.), designed for high-frequency applications, and selected to match the requirements of the crystal or resonator (see *Figure 13*).  $C_{L1}$  and  $C_{L2}$  are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of  $C_{L1}$  and  $C_{L2}$ . PCB and MCU pin capacitance must be included (10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing  $C_{L1}$  and  $C_{L2}$ . Refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

Resonator

CL1

OSC\_IN

Resonator

CL2

OSC\_OUT

STM32

Figure 13. HSE oscillator circuit diagram

1. R<sub>EXT</sub> value depends on the crystal characteristics.

## Low-speed external clock generated from a crystal/ceramic resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in *Table 26*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Table 26.	LSE oscillator characteristics (f <sub>LSE</sub> = 32.76	8 kHz) <sup>(1)</sup>
Table 20.	LSE OSCINATOR CHARACTERISTICS (II SE = 32.70	O KHZJ`

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>LSE</sub>	Low speed external oscillator frequency			32.768		kHz
R <sub>F</sub>	Feedback resistor			1.2		МΩ
C <sup>(2)</sup>	Recommended load capacitance versus equivalent serial resistance of the crystal (R <sub>S</sub> ) <sup>(3)</sup>	R <sub>S</sub> = 30 kΩ		8		pF
I <sub>LSE</sub>	LSE driving current	$V_{DD} = 3.3 \text{ V}, V_{IN} = V_{SS}$			1.1	μA
		V <sub>DD</sub> = 1.8 V		450		
I <sub>DD (LSE)</sub>	LSE oscillator current consumption	V <sub>DD</sub> = 3.0 V		600		nA
	,	V <sub>DD</sub> = 3.6V		750		
9 <sub>m</sub>	Oscillator transconductance		3			μ <b>A</b> /V
t <sub>SU(LSE)</sub> <sup>(4)</sup>	Startup time	V <sub>DD</sub> is stabilized		1		s

- 1. Based on characterization, not tested in production.
- Refer to the note and caution paragraphs below the table, and to the application note AN2867 "Oscillator design guide for ST microcontrollers".
- The oscillator selection can be optimized in terms of supply current using an high quality resonator with small R<sub>S</sub> value for example MSIV-TIN32.768kHz. Refer to crystal manufacturer for more details;
- t<sub>SU(LSE)</sub> is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768 kHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer.

Note:

For  $C_{L1}$  and  $C_{L2}$ , it is recommended to use high-quality ceramic capacitors in the 5 pF to 15 pF range selected to match the requirements of the crystal or resonator (see Figure 14).  $C_{L1}$  and  $C_{L2}$ , are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of  $C_{L1}$  and  $C_{L2}$ .

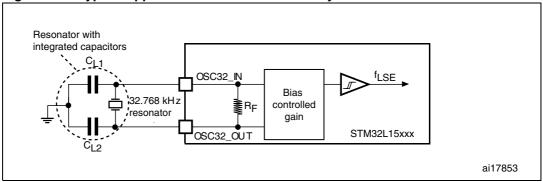
Load capacitance  $C_L$  has the following formula:  $C_L = C_{L1} \times C_{L2} / (C_{L1} + C_{L2}) + C_{stray}$  where  $C_{stray}$  is the pin capacitance and board or trace PCB-related capacitance. Typically, it is between 2 pF and 7 pF.

Caution:

To avoid exceeding the maximum value of  $C_{L1}$  and  $C_{L2}$  (15 pF) it is strongly recommended to use a resonator with a load capacitance  $C_L \le 7$  pF. Never use a resonator with a load capacitance of 12.5 pF.

**Example:** if you choose a resonator with a load capacitance of  $C_L = 6$  pF and  $C_{stray} = 2$  pF, then  $C_{L1} = C_{L2} = 8$  pF.

Figure 14. Typical application with a 32.768 kHz crystal



#### 6.3.6 Internal clock source characteristics

The parameters given in *Table 27* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 10*.

### High-speed internal (HSI) RC oscillator

Table 27. HSI oscillator characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>HSI</sub>	Frequency	V <sub>DD</sub> = 3.0 V		16		MHz
TRIM <sup>(1)(2)</sup>	HSI user-trimmed	Trimming code is not a multiple of 16		±0.4	0.7	%
TRIM	resolution	Trimming code is a multiple of 16			±1.5	%
		$V_{DDA} = 3.0 \text{ V}, T_A = 25 ^{\circ}\text{C}$	-1 <sup>(3)</sup>		1 <sup>(3)</sup>	%
	Accuracy of the factory-calibrated HSI oscillator	V <sub>DDA</sub> = 3.0 V, T <sub>A</sub> = 0 to 55 °C	-1.5		1.5	%
		V <sub>DDA</sub> = 3.0 V, T <sub>A</sub> = -10 to 70 °C	-2		2	%
ACC <sub>HSI</sub> <sup>(2)</sup>		$V_{DDA} = 3.0 \text{ V}, T_A = -10 \text{ to } 85 ^{\circ}\text{C}$	-2.5		2	%
		V <sub>DDA</sub> = 3.0 V, T <sub>A</sub> = -10 to 105 °C	-4		2	%
		V <sub>DDA</sub> = 1.65 V to 3.6 V T <sub>A</sub> = -40 to 105 °C	-4		3	%
t <sub>SU(HSI)</sub> <sup>(2)</sup>	HSI oscillator startup time			3.7	6	μs
I <sub>DD(HSI)</sub> <sup>(2)</sup>	HSI oscillator power consumption			100	140	μΑ

<sup>1.</sup> The trimming step differs depending on the trimming code. It is usually negative on the codes which are multiples of 16 (0x00, 0x10, 0x20, 0x30...0xE0).

### Low-speed internal (LSI) RC oscillator

Table 28. LSI oscillator characteristics

Symbol	Parameter	Min	Тур	Max	Unit
f <sub>LSI</sub> <sup>(1)</sup>	LSI frequency	26	38	56	kHz
D <sub>LSI</sub> <sup>(2)</sup>	LSI oscillator frequency drift 0°C ≤T <sub>A</sub> ≤ 85°C	-10		4	%
t <sub>su(LSI)</sub> <sup>(3)</sup>	LSI oscillator startup time			200	μs
I <sub>DD(LSI)</sub> <sup>(3)</sup>	LSI oscillator power consumption		400	510	nA

<sup>1.</sup> Tested in production.

<sup>2.</sup> Based on characterization, not tested in production.

<sup>3.</sup> Tested in production.

<sup>2.</sup> This is a deviation for an individual part, once the initial frequency has been measured.

<sup>3.</sup> Guaranteed by design, not tested in production.

# Multi-speed internal (MSI) RC oscillator

Table 29. MSI oscillator characteristics

Symbol	Parameter	Condition	Тур	Max	Unit
f <sub>MSI</sub>	Frequency after factory calibration, done at $V_{DD}$ = 3.3 V and $T_A$ = 25 °C	MSI range 0	65.5		- kHz
		MSI range 1	131		
		MSI range 2	262		
		MSI range 3	524		
		MSI range 4	1.05		MHz
		MSI range 5	2.1		
		MSI range 6	4.2		
ACC <sub>MSI</sub>	Frequency error after factory calibration		±0.5		%
D <sub>TEMP(MSI)</sub> <sup>(1)</sup>	MSI oscillator frequency drift 0 °C ≤T <sub>A</sub> ⊴85 °C		±3		%
D <sub>VOLT(MSI)</sub> <sup>(1)</sup>	MSI oscillator frequency drift 1.65 V ≤V <sub>DD</sub> ≤3.6 V, T <sub>A</sub> = 25 °C			2.5	%/V
I <sub>DD(MSI)</sub> <sup>(2)</sup>	MSI oscillator power consumption	MSI range 0	0.75		μА
		MSI range 1	1		
		MSI range 2	1.5		
		MSI range 3	2.5		
		MSI range 4	4.5		
		MSI range 5	8		
		MSI range 6	15		

Table 29. MSI oscillator characteristics (continued)

Symbol	Parameter	Condition	Тур	Max	Unit
<sup>t</sup> sU(MSI)	MSI oscillator startup time	MSI range 0	30		
		MSI range 1	20		
		MSI range 2	15		
		MSI range 3	10		
		MSI range 4	6		
		MSI range 5	5		
		MSI range 6, Voltage range 1 and 2	3.5		
		MSI range 6, Voltage range 3	5		
	MSI oscillator stabilization time	MSI range 0		40	μs
		MSI range 1		20	
		MSI range 2		10	
t <sub>STAB(MSI)</sub> <sup>(2)</sup>		MSI range 3		4	
		MSI range 4		2.5	
STAB(MSI)		MSI range 5		2	
		MSI range 6, Voltage range 1 and 2		2	
		MSI range 3, Voltage range 3		3	
f <sub>OVER(MSI)</sub>	MSI oscillator frequency overshoot	Any range to range 5		4	MHz
		Any range to range 6		6	IVII IZ

<sup>1.</sup> This is a deviation for an individual part, once the initial frequency has been measured.

<sup>2.</sup> Based on characterization, not tested in production.

### 6.3.7 PLL characteristics

The parameters given in *Table 30* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 10*.

Table 30. PLL characteristics

Cumbal	Dovomotov		Unit		
Symbol	Parameter	Min	Тур	Max <sup>(1)</sup>	Unit
f	PLL input clock <sup>(2)</sup>	2		24	MHz
f <sub>PLL_IN</sub>	PLL input clock duty cycle	45		55	%
f <sub>PLL_OUT</sub>	PLL output clock	2		32	MHz
t <sub>LOCK</sub>	Worst case PLL lock time PLL input = 2 MHz PLL VCO = 96 MHz		100	130	μs
Jitter	Cycle-to-cycle jitter			±600	ps
I <sub>DDA</sub> (PLL)	Current consumption on V <sub>DDA</sub>		220	450	
I <sub>DD</sub> (PLL)	Current consumption on V <sub>DD</sub>		120	150	μΑ

<sup>1.</sup> Based on characterization, not tested in production.

# 6.3.8 Memory characteristics

The characteristics are given at  $T_A = -40$  to 105 °C unless otherwise specified.

### **RAM** memory

Table 31. RAM and hardware registers

	Symbol	Parameter	Conditions	Min	Тур	Max	Unit
ſ	VRM	Data retention mode <sup>(1)</sup>	STOP mode (or RESET)	1.65			V

Minimum supply voltage without losing data stored in RAM (in Stop mode or under Reset) or in hardware registers (only in Stop mode).

<sup>2.</sup> Take care of using the appropriate multiplier factors so as to have PLL input clock values compatible with the range defined by  $f_{\text{PLL\_OUT}}$ .

# Flash memory

Table 32. Flash memory characteristics

Symbol	Parameter	Conditions	Min	Тур	Max <sup>(1)</sup>	Unit
V <sub>DD</sub>	Operating voltage Read / Write / Erase		1.65		3.6	٧
t <sub>prog</sub>	Programming time for	Erasing		3.28	3.94	mo
	word or half-page	Programming		3.28	3.94	ms
I <sub>DD</sub>	Average current during the whole programming / erase operation	T 25 °C V 26 V		600	TBD	μΑ
	Maximum current (peak) during the whole pro- gramming / erase opera- tion	T <sub>A</sub> = 25 °C, V <sub>DD</sub> = 3.6 V		1.5	TBD	mA

<sup>1.</sup> Guaranteed by design, not tested in production.

Table 33. Flash memory endurance and data retention

Symbol	Parameter	Conditions	Value			Unit
Symbol	Farameter	Conditions	Min <sup>(1)</sup>	Тур	Max	Oill
N <sub>CYC</sub> <sup>(2)</sup>	Cycling (erase / write ) Program memory	ogram memory $T_A = -40^{\circ}\text{C to}$ cling (erase / write ) $105^{\circ}\text{C}$	10			keyeles
CYC	Cycling (erase / write ) EEPROM data memory		300			kcycles
t <sub>RET</sub> <sup>(2)</sup>	Data retention (program memory) after 10 kcycles at T <sub>A</sub> = 85 °C	T	30			
	Data retention (EEPROM data memory) after 300 kcycles at T <sub>A</sub> = 85 °C	T <sub>RET</sub> = +85 °C	30			Vooro
	Data retention (program memory) after 10 kcycles at T <sub>A</sub> = 105 °C	T <sub>BET</sub> = +105 °C	10			years
	Data retention (EEPROM data memory) after 300 kcycles at T <sub>A</sub> = 105 °C	RET - +105 C	10			

<sup>1.</sup> Based on characterization not tested in production.

<sup>2.</sup> Characterization is done according to JEDEC JESD22-A117.

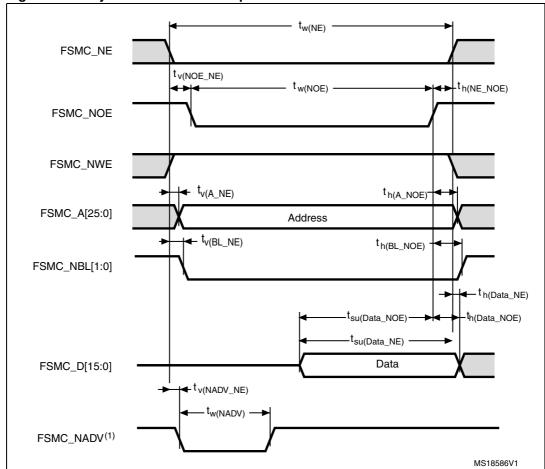
### 6.3.9 FSMC characteristics

# Asynchronous waveforms and timings

Figure 15 through Figure 18 represent asynchronous waveforms and Table 34 through Table 37 provide the corresponding timings. The results shown in these tables are obtained with the following FSMC configuration:

- AddressSetupTime = 0
- AddressHoldTime = 1
- DataSetupTime = 1

Figure 15. Asynchronous non-multiplexed SRAM/PSRAM/NOR read waveforms



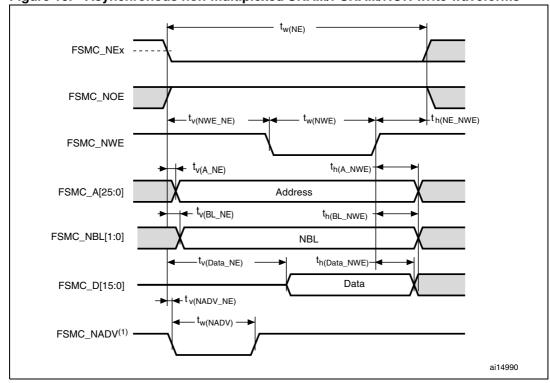
1. Mode 2/B, C and D only. In Mode 1, FSMC\_NADV is not used.

Table 34. Asynchronous non-multiplexed SRAM/PSRAM/NOR read timings<sup>(1)(2)</sup>

Symbol	Parameter	Min	Max	Unit
t <sub>w(NE)</sub>	FSMC_NE low time	TBD	TBD	ns
t <sub>v(NOE_NE)</sub>	FSMC_NEx low to FSMC_NOE low	TBD	TBD	ns
t <sub>w(NOE)</sub>	FSMC_NOE low time	TBD	TBD	ns
t <sub>h(NE_NOE)</sub>	FSMC_NOE high to FSMC_NE high hold time	TBD		ns
t <sub>v(A_NE)</sub>	FSMC_NEx low to FSMC_A valid		0	ns
t <sub>h(A_NOE)</sub>	Address hold time after FSMC_NOE high	TBD		ns
t <sub>v(BL_NE)</sub>	FSMC_NEx low to FSMC_BL valid		0	ns
t <sub>h(BL_NOE)</sub>	FSMC_BL hold time after FSMC_NOE high	0		ns
t <sub>su(Data_NE)</sub>	Data to FSMC_NEx high setup time	TBD		ns
t <sub>su(Data_NOE)</sub>	Data to FSMC_NOEx high setup time	TBD		ns
t <sub>h(Data_NOE)</sub>	Data hold time after FSMC_NOE high	0		ns
t <sub>h(Data_NE)</sub>	Data hold time after FSMC_NEx high	0		ns
t <sub>v(NADV_NE)</sub>	FSMC_NEx low to FSMC_NADV low		5	ns
t <sub>w(NADV)</sub>	FSMC_NADV low time		TBD	ns

<sup>1.</sup>  $C_L = 15 pF$ .

Figure 16. Asynchronous non-multiplexed SRAM/PSRAM/NOR write waveforms



1. Mode 2/B, C and D only. In Mode 1, FSMC\_NADV is not used.

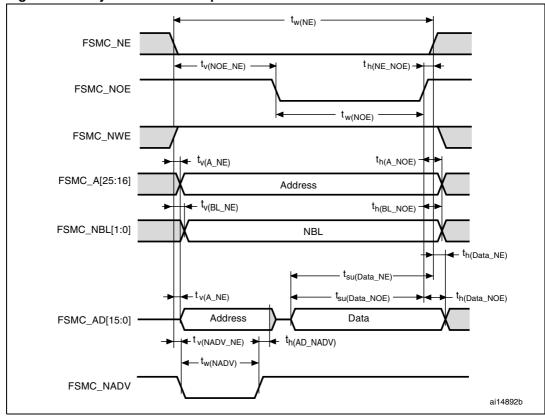
<sup>2.</sup> Preliminary values.

Table 35. Asynchronous non-multiplexed SRAM/PSRAM/NOR write timings<sup>(1)(2)</sup>

Symbol	Parameter	Min	Max	Unit
t <sub>w(NE)</sub>	FSMC_NE low time	TBD	TBD	ns
t <sub>v(NWE_NE)</sub>	FSMC_NEx low to FSMC_NWE low	TBD	TBD	ns
t <sub>w(NWE)</sub>	FSMC_NWE low time	TBD	TBD	ns
t <sub>h(NE_NWE)</sub>	FSMC_NWE high to FSMC_NE high hold time	T <sub>HCLK</sub>		ns
t <sub>v(A_NE)</sub>	FSMC_NEx low to FSMC_A valid		TBD	ns
t <sub>h(A_NWE)</sub>	Address hold time after FSMC_NWE high	T <sub>HCLK</sub>		ns
t <sub>v(BL_NE)</sub>	FSMC_NEx low to FSMC_BL valid		TBD	ns
t <sub>h(BL_NWE)</sub>	FSMC_BL hold time after FSMC_NWE high	TBD		ns
t <sub>v(Data_NE)</sub>	FSMC_NEx low to Data valid		TBD	ns
t <sub>h(Data_NWE)</sub>	Data hold time after FSMC_NWE high	T <sub>HCLK</sub>		ns
t <sub>v(NADV_NE)</sub>	FSMC_NEx low to FSMC_NADV low		TBD	ns
t <sub>w(NADV)</sub>	FSMC_NADV low time		TBD	ns

<sup>1.</sup>  $C_L = 15 pF$ .

Figure 17. Asynchronous multiplexed PSRAM/NOR read waveforms



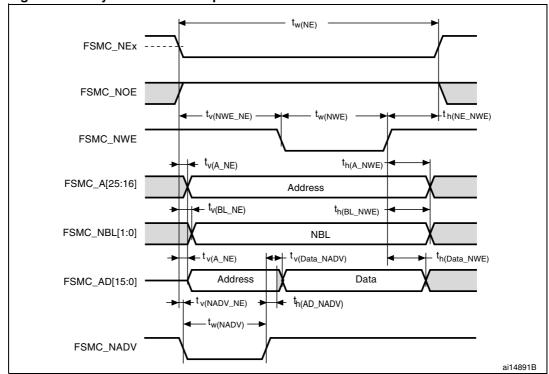
<sup>2.</sup> Preliminary values.

Table 36. Asynchronous multiplexed PSRAM/NOR read timings<sup>(1)(2)</sup>

Symbol	Parameter	Min	Max	Unit
t <sub>w(NE)</sub>	FSMC_NE low time	TBD	TBD	ns
t <sub>v(NOE_NE)</sub>	FSMC_NEx low to FSMC_NOE low	TBD	TBD	ns
t <sub>w(NOE)</sub>	FSMC_NOE low time	TBD	TBD	ns
t <sub>h(NE_NOE)</sub>	FSMC_NOE high to FSMC_NE high hold time	TBD		ns
t <sub>v(A_NE)</sub>	FSMC_NEx low to FSMC_A valid		0	ns
t <sub>v(NADV_NE)</sub>	FSMC_NEx low to FSMC_NADV low	TBD	TBD	ns
t <sub>w(NADV)</sub>	FSMC_NADV low time	TBD	TBD	ns
t <sub>h(AD_NADV)</sub>	FSMC_AD (address) valid hold time after FSMC_NADV high	T <sub>HCLK</sub>		ns
t <sub>h(A_NOE)</sub>	Address hold time after FSMC_NOE high	T <sub>HCLK</sub>		ns
t <sub>h(BL_NOE)</sub>	FSMC_BL hold time after FSMC_NOE high	0		ns
t <sub>v(BL_NE)</sub>	FSMC_NEx low to FSMC_BL valid		0	ns
t <sub>su(Data_NE)</sub>	Data to FSMC_NEx high setup time	TBD		ns
t <sub>su(Data_NOE)</sub>	Data to FSMC_NOE high setup time	TBD		ns
t <sub>h(Data_NE)</sub>	Data hold time after FSMC_NEx high	0		ns
t <sub>h(Data_NOE)</sub>	Data hold time after FSMC_NOE high	0		ns

<sup>1.</sup>  $C_L = 15 pF$ .

Figure 18. Asynchronous multiplexed PSRAM/NOR write waveforms



<sup>2.</sup> Preliminary values.

Table 37. Asynchronous multiplexed PSRAM/NOR write timings<sup>(1)(2)</sup>

Symbol	Parameter	Min	Max	Unit
t <sub>w(NE)</sub>	FSMC_NE low time	TBD	TBD	ns
t <sub>v(NWE_NE)</sub>	FSMC_NEx low to FSMC_NWE low	2T <sub>HCLK</sub>	TBD	ns
t <sub>w(NWE)</sub>	FSMC_NWE low time	TBD	TBD	ns
t <sub>h(NE_NWE)</sub>	FSMC_NWE high to FSMC_NE high hold time	TBD		ns
t <sub>v(A_NE)</sub>	FSMC_NEx low to FSMC_A valid		TBD	ns
t <sub>v(NADV_NE)</sub>	FSMC_NEx low to FSMC_NADV low	TBD	TBD	ns
t <sub>w(NADV)</sub>	FSMC_NADV low time	TBD	TBD	ns
t <sub>h(AD_NADV)</sub>	FSMC_AD (address) valid hold time after FSMC_NADV high	TBD		ns
t <sub>h(A_NWE)</sub>	Address hold time after FSMC_NWE high	TBD		ns
t <sub>v(BL_NE)</sub>	FSMC_NEx low to FSMC_BL valid		TBD	ns
t <sub>h(BL_NWE)</sub>	FSMC_BL hold time after FSMC_NWE high	TBD		ns
t <sub>v(Data_NADV)</sub>	FSMC_NADV high to Data valid		TBD	ns
t <sub>h(Data_NWE)</sub>	Data hold time after FSMC_NWE high	TBD		ns

<sup>1.</sup>  $C_L = 15 pF$ .

<sup>2.</sup> Preliminary values.

# Synchronous waveforms and timings

Figure 19 through Figure 22 represent synchronous waveforms and Table 39 through Table 41 provide the corresponding timings. The results shown in these tables are obtained with the following FSMC configuration:

- BurstAccessMode = FSMC\_BurstAccessMode\_Enable;
- MemoryType = FSMC\_MemoryType\_CRAM;
- WriteBurst = FSMC WriteBurst Enable;
- CLKDivision = 1; (0 is not supported, see the STM32F10xxx reference manual)
- DataLatency = 1 for NOR Flash; DataLatency = 0 for PSRAM

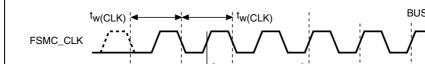


Figure 19. Synchronous multiplexed NOR/PSRAM read timings

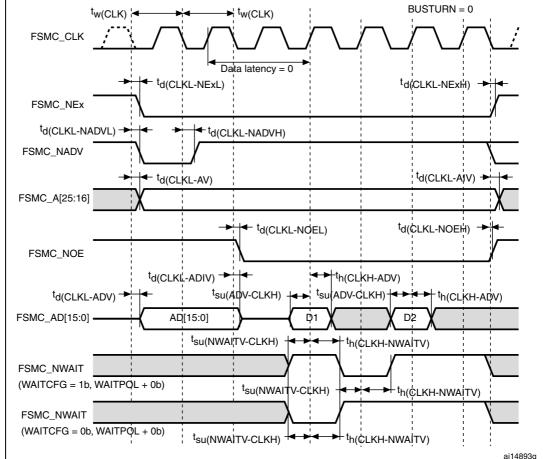


Table 38. Synchronous multiplexed NOR/PSRAM read timings<sup>(1)(2)</sup>

Symbol	Parameter	Min	Max	Unit
t <sub>w(CLK)</sub>	FSMC_CLK period	2*T <sub>HCLK</sub>		ns
t <sub>d(CLKL-NExL)</sub>	FSMC_CLK low to FSMC_NEx low (x = 02)		TBD	ns
t <sub>d(CLKL-NExH)</sub>	FSMC_CLK low to FSMC_NEx high (x = 02)	TBD		ns
t <sub>d(CLKL-NADVL)</sub>	FSMC_CLK low to FSMC_NADV low		TBD	ns
t <sub>d(CLKL-NADVH)</sub>	FSMC_CLK low to FSMC_NADV high	TBD		ns
t <sub>d(CLKL-AV)</sub>	FSMC_CLK low to FSMC_Ax valid (x = 1625)		0	ns
t <sub>d(CLKL-AIV)</sub>	FSMC_CLK low to FSMC_Ax invalid (x = 1625)	TBD		ns
t <sub>d(CLKL-NOEL)</sub>	FSMC_CLK low to FSMC_NOE low		TBD	ns
t <sub>d(CLKL-NOEH)</sub>	FSMC_CLK low to FSMC_NOE high	TBD		ns
t <sub>d(CLKL-ADV)</sub>	FSMC_CLK low to FSMC_AD[15:0] valid		TBD	ns
t <sub>d(CLKL-ADIV)</sub>	FSMC_CLK low to FSMC_AD[15:0] invalid	0		ns
t <sub>su(ADV-CLKH)</sub>	FSMC_A/D[15:0] valid data before FSMC_CLK high	TBD		ns
t <sub>h(CLKH-ADV)</sub>	FSMC_A/D[15:0] valid data after FSMC_CLK high	0		ns
t <sub>su(NWAITV-CLKH)</sub>	FSMC_NWAIT valid before FSMC_CLK high	TBD		ns
t <sub>h(CLKH-NWAITV)</sub>	FSMC_NWAIT valid after FSMC_CLK high	TBD		ns

<sup>1.</sup>  $C_L = 15 pF$ .

<sup>2.</sup> Preliminary values.

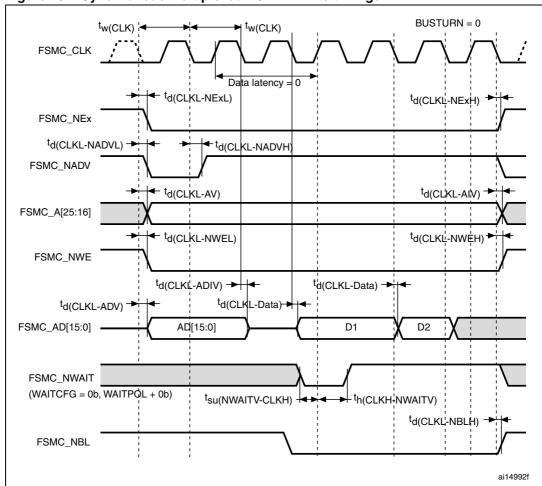


Figure 20. Synchronous multiplexed PSRAM write timings

Table 39. Synchronous multiplexed PSRAM write timings<sup>(1)(2)</sup>

Symbol	Parameter	Min	Max	Unit
t <sub>w(CLK)</sub>	FSMC_CLK period	2*T <sub>HCLK</sub>		ns
t <sub>d(CLKL-NExL)</sub>	FSMC_CLK low to FSMC_Nex low (x = 02)		TBD	ns
t <sub>d(CLKL-NExH)</sub>	FSMC_CLK low to FSMC_NEx high (x = 02)	TBD		ns
t <sub>d(CLKL-NADVL)</sub>	FSMC_CLK low to FSMC_NADV low		TBD	ns
t <sub>d(CLKL-NADVH)</sub>	FSMC_CLK low to FSMC_NADV high	TBD		ns
t <sub>d(CLKL-AV)</sub>	FSMC_CLK low to FSMC_Ax valid (x = 1625)		0	ns
t <sub>d(CLKL-AIV)</sub>	FSMC_CLK low to FSMC_Ax invalid (x = 1625)	TBD		ns
t <sub>d(CLKL-NWEL)</sub>	FSMC_CLK low to FSMC_NWE low		TBD	ns
t <sub>d(CLKL-NWEH)</sub>	FSMC_CLK low to FSMC_NWE high	TBD		ns
t <sub>d(CLKL-ADV)</sub>	FSMC_CLK low to FSMC_AD[15:0] valid		TBD	ns
t <sub>d(CLKL-ADIV)</sub>	FSMC_CLK low to FSMC_AD[15:0] invalid	TBD		ns
t <sub>d(CLKL-Data)</sub>	FSMC_A/D[15:0] valid after FSMC_CLK low		TBD	ns
t <sub>su(NWAITV-CLKH)</sub>	FSMC_NWAIT valid before FSMC_CLK high	TBD		ns
t <sub>h(CLKH-NWAITV)</sub>	FSMC_NWAIT valid after FSMC_CLK high	TBD		ns
t <sub>d(CLKL-NBLH)</sub>	FSMC_CLK low to FSMC_NBL high	TBD		ns

<sup>1.</sup>  $C_L = 15 pF$ .

<sup>2.</sup> Preliminary values

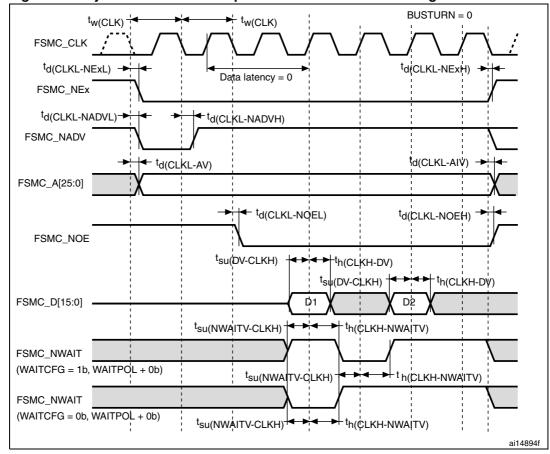


Figure 21. Synchronous non-multiplexed NOR/PSRAM read timings

Table 40. Synchronous non-multiplexed NOR/PSRAM read timings<sup>(1)(2)</sup>

Symbol	Parameter	Min	Max	Unit
t <sub>w(CLK)</sub>	FSMC_CLK period	2*T <sub>HCLK</sub>		ns
t <sub>d(CLKL-NExL)</sub>	FSMC_CLK low to FSMC_NEx low (x = 02)		TBD	ns
t <sub>d(CLKL-NExH)</sub>	FSMC_CLK low to FSMC_NEx high (x = 02)	TBD		ns
t <sub>d(CLKL-NADVL)</sub>	FSMC_CLK low to FSMC_NADV low		TBD	ns
t <sub>d(CLKL-NADVH)</sub>	FSMC_CLK low to FSMC_NADV high	TBD		ns
t <sub>d(CLKL-AV)</sub>	FSMC_CLK low to FSMC_Ax valid (x = 025)		0	ns
t <sub>d(CLKL-AIV)</sub>	FSMC_CLK low to FSMC_Ax invalid (x = 025)	TBD		ns
t <sub>d(CLKL-NOEL)</sub>	FSMC_CLK low to FSMC_NOE low		TBD	ns
t <sub>d(CLKL-NOEH)</sub>	FSMC_CLK low to FSMC_NOE high	TBD		ns
t <sub>su(DV-CLKH)</sub>	FSMC_D[15:0] valid data before FSMC_CLK high	TBD		ns
t <sub>h(CLKH-DV)</sub>	FSMC_D[15:0] valid data after FSMC_CLK high	TBD		ns
t <sub>su(NWAITV-CLKH)</sub>	FSMC_NWAIT valid before FSMC_SMCLK high	TBD		ns
t <sub>h(CLKH-NWAITV)</sub>	FSMC_NWAIT valid after FSMC_CLK high	TBD		ns

<sup>1.</sup>  $C_L = 15 pF$ .

<sup>2.</sup> Preliminary values.

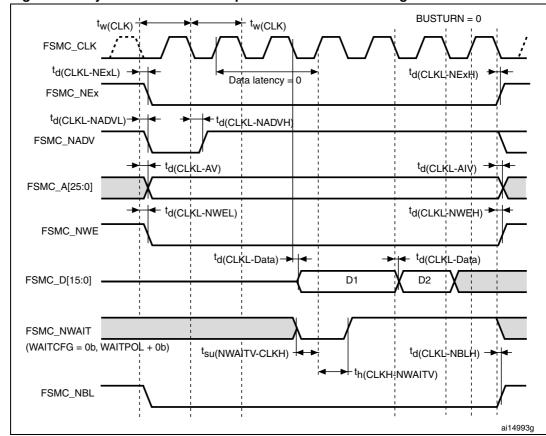


Figure 22. Synchronous non-multiplexed PSRAM write timings

Table 41. Synchronous non-multiplexed PSRAM write timings<sup>(1)(2)</sup>

Symbol	Parameter	Min	Max	Unit
t <sub>w(CLK)</sub>	FSMC_CLK period	2*T <sub>HCLK</sub>		ns
t <sub>d(CLKL-NExL)</sub>	FSMC_CLK low to FSMC_NEx low (x = 02)		TBD	ns
t <sub>d(CLKL-NExH)</sub>	FSMC_CLK low to FSMC_NEx high (x = 02)	TBD		ns
t <sub>d(CLKL-NADVL)</sub>	FSMC_CLK low to FSMC_NADV low		TBD	ns
t <sub>d(CLKL-NADVH)</sub>	FSMC_CLK low to FSMC_NADV high	TBD		ns
t <sub>d(CLKL-AV)</sub>	FSMC_CLK low to FSMC_Ax valid (x = 1625)		0	ns
t <sub>d(CLKL-AIV)</sub>	FSMC_CLK low to FSMC_Ax invalid (x = 1625)	TBD		ns
t <sub>d(CLKL-NWEL)</sub>	FSMC_CLK low to FSMC_NWE low		TBD	ns
t <sub>d(CLKL-NWEH)</sub>	FSMC_CLK low to FSMC_NWE high	TBD		ns
t <sub>d(CLKL-Data)</sub>	FSMC_D[15:0] valid data after FSMC_CLK low		TBD	ns
t <sub>su(NWAITV-CLKH)</sub>	FSMC_NWAIT valid before FSMC_CLK high	TBD		ns
t <sub>h(CLKH-NWAITV)</sub>	FSMC_NWAIT valid after FSMC_CLK high	TBD		ns
t <sub>d(CLKL-NBLH)</sub>	FSMC_CLK low to FSMC_NBL high	TBD		ns

<sup>1.</sup>  $C_L = 15 pF$ .

<sup>2.</sup> Preliminary values.

#### 6.3.10 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

### Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports). the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- Electrostatic discharge (ESD) (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- FTB: A Burst of Fast Transient voltage (positive and negative) is applied to V<sub>DD</sub> and V<sub>SS</sub> through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.

A device reset allows normal operations to be resumed.

The test results are given in *Table 42*. They are based on the EMS levels and classes defined in application note AN1709.

Table 42. EMS characteristics

Symbol	Parameter	Conditions	Level/ Class
V <sub>FESD</sub>	Voltage limits to be applied on any I/O pin to induce a functional disturbance	$V_{DD}$ = 3.3 V, LQFP100, $T_{A}$ = +25 °C, $f_{HCLK}$ = 32 MHz conforms to IEC 61000-4-2	2B
V <sub>EFTB</sub>	Fast transient voltage burst limits to be applied through 100 pF on V <sub>DD</sub> and V <sub>SS</sub> pins to induce a functional disturbance	$V_{DD}$ = 3.3 V, LQFP100, $T_A$ = +25 °C, $f_{HCLK}$ = 32 MHz conforms to IEC 61000-4-4	4A

### Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.

### Software recommendations

The software flowchart must include the management of runaway conditions such as:

- Corrupted program counter
- Unexpected reset
- Critical data corruption (control registers...)

#### **Prequalification trials**

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the oscillator pins for 1 second.

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

# **Electromagnetic Interference (EMI)**

The electromagnetic field emitted by the device are monitored while a simple application is executed (toggling 2 LEDs through the I/O ports). This emission test is compliant with IEC 61967-2 standard which specifies the test board and the pin loading.

Table 43. EMI characteristics

			Max vs.				
Symbol	Parameter	Conditions	Monitored frequency band	4 MHz voltage range 3	16 MHz voltage range 2	voltage	Unit
		$V_{DD} = 3.3 \text{ V},$	0.1 to 30 MHz	3	-6	-5	
6	Peak level	$T_A = 25  ^{\circ}C$	30 to 130 MHz	18	4	-7	dΒμV
SEMI	reak level	compliant with IEC	130 MHz to 1GHz	15	5	-7	
			SAE EMI Level	2.5	2	1	-

# 6.3.11 Absolute maximum ratings (electrical sensitivity)

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

#### Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts  $\times$  (n+1) supply pins). This test conforms to the JESD22-A114/C101 standard.

Table 44. ESD absolute maximum ratings

Symbol	Ratings	Conditions	Class	Maximum value <sup>(1)</sup>	Unit
V <sub>ESD(HBM)</sub>		T <sub>A</sub> = +25 °C, conforming to JESD22-A114	2	2000	V
V <sub>ESD(CDM)</sub>	Electrostatic discharge voltage (charge device model)	T <sub>A</sub> = +25 °C, conforming to JESD22-C101	Ш	500	V

<sup>1.</sup> Based on characterization results, not tested in production.

# Static latch-up

Two complementary static tests are required on six parts to assess the latch-up performance:

- A supply overvoltage is applied to each power supply pin
- A current injection is applied to each input, output and configurable I/O pin

These tests are compliant with EIA/JESD 78A IC latch-up standard.

Table 45. Electrical sensitivities

Symbol	Parameter	Conditions	Class
LU	Static latch-up class	T <sub>A</sub> = +105 °C conforming to JESD78A	II level A

# 6.3.12 I/O current injection characteristics

As a general rule, current injection to the I/O pins, due to external voltage below  $V_{SS}$  or above  $V_{DD}$  (for standard pins) should be avoided during normal product operation. However, in order to give an indication of the robustness of the microcontroller in cases when abnormal injection accidentally happens, susceptibility tests are performed on a sample basis during device characterization.

### Functional susceptibility to I/O current injection

While a simple application is executed on the device, the device is stressed by injecting current into the I/O pins programmed in floating input mode. While current is injected into the I/O pin, one at a time, the device is checked for functional failures.

The failure is indicated by an out of range parameter: ADC error, out of spec current injection on adjacent pins or other functional failure (for example reset, oscillator frequency deviation, LCD levels, etc.).

The test results are given in the following table.

Table 46. I/O current injection susceptibility

		Functional s		
Symbol Description		Negative injection	Positive injection	Unit
	Injected current on true open-drain pins	-5	+0	
I <sub>INJ</sub>	Injected current on all 5 V tolerant (FT) pins	-5	+0	mA
	Injected current on any other pin	-5	+5	

# 6.3.13 I/O port characteristics

# General input/output characteristics

Unless otherwise specified, the parameters given in *Table 47* are derived from tests performed under the conditions summarized in *Table 10*. All I/Os are CMOS and TTL compliant.

Table 47. I/O static characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>IL</sub>	Input low level voltage		V <sub>SS</sub> - 0.3		0.8	
	Standard I/O input high level voltage	TTL ports 2.7 V ≤V <sub>DD</sub> ≤3.6 V	2 <sup>(1)</sup>		V <sub>DD</sub> +0.3	
V <sub>IH</sub>	FT <sup>(2)</sup> I/O input high level voltage		2` '		5.5V	
V <sub>IL</sub>	Input low level voltage	CMOS ports 1.65 V ≤V <sub>DD</sub> ≤ 3.6 V	-0.3		0.3V <sub>DD</sub> <sup>(3)</sup>	
	Standard I/O Input high level voltage	CMOS ports 1.65 V ≤V <sub>DD</sub> ≤ 3.6 V			V <sub>DD</sub> +0.3	V
V <sub>IH</sub>	FT <sup>(5)</sup> I/O input high level voltage	CMOS ports 1.65 V ≤V <sub>DD</sub> ≤ 2.0 V	0.7 V <sub>DD</sub> <sup>(3)(4)</sup>		5.25	
	T 1 - 70 input high level voltage	CMOS ports 2.0 V≤V <sub>DD</sub> ≤ 3.6 V		5.5		
V <sub>hys</sub>	Standard I/O Schmitt trigger voltage hysteresis <sup>(6)</sup>		10% V <sub>DD</sub> <sup>(7)</sup>			
		V <sub>SS</sub> ≤V <sub>IN</sub> ≤V <sub>DD</sub> I/Os with LCD			±50	
		V <sub>SS</sub> ≤V <sub>IN</sub> ≤V <sub>DD</sub> I/Os with analog switches			±50	
I <sub>lkg</sub>	Input leakage current (8)(3)	V <sub>SS</sub> ≤V <sub>IN</sub> ≤V <sub>DD</sub> I/Os with analog switches and LCD			±50	nA
		V <sub>SS</sub> ≤V <sub>IN</sub> ≤V <sub>DD</sub> I/Os with USB			TBD	
		V <sub>SS</sub> ≤V <sub>IN</sub> ≤V <sub>DD</sub> Standard I/Os			±50	
R <sub>PU</sub>	Weak pull-up equivalent resistor <sup>(9)(3)</sup>	$V_{IN} = V_{SS}$	30	45	60	kΩ
R <sub>PD</sub>	Weak pull-down equivalent resistor <sup>(9)(3)</sup>	$V_{IN} = V_{DD}$	30	45	60	kΩ
C <sub>IO</sub>	I/O pin capacitance			5		pF

<sup>1.</sup> Guaranteed by design.

<sup>2.</sup> FT = 5V tolerant. To sustain a voltage higher than VDD +0.5 the internal pull-up/pull-down resistors must be disabled.

<sup>3.</sup> Tested in production

<sup>4. 0.7</sup>V<sub>DD</sub> for 5V-tolerant receiver

<sup>5.</sup> FT = Five-volt tolerant.

<sup>6.</sup> Hysteresis voltage between Schmitt trigger switching levels. Based on characterization, not tested in production.

<sup>7.</sup> With a minimum of 200 mV. Based on characterization, not tested in production.

- 8. The max. value may be exceeded if negative current is injected on adjacent pins.
- Pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS/NMOS. This MOS/NMOS contribution to the series resistance is minimum (~10% order).

### **Output driving current**

The GPIOs (general purpose input/outputs) can sink or source up to  $\pm 8$  mA, and sink or source up to  $\pm 20$  mA with the non-standard  $V_{OL}/V_{OH}$  specifications given in *Table 48*.

in the user application, the number of I/O pins which can drive current must be limited to respect the absolute maximum rating specified in *Section 6.2*:

- The sum of the currents sourced by all the I/Os on V<sub>DD</sub>, plus the maximum Run consumption of the MCU sourced on V<sub>DD</sub>, cannot exceed the absolute maximum rating I<sub>VDD</sub> (see *Table 8*).
- The sum of the currents sunk by all the I/Os on V<sub>SS</sub> plus the maximum Run consumption of the MCU sunk on V<sub>SS</sub> cannot exceed the absolute maximum rating I<sub>VSS</sub> (see *Table 8*).

### **Output voltage levels**

Unless otherwise specified, the parameters given in *Table 48* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 10*. All I/Os are CMOS and TTL compliant.

Table 48. Output voltage characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
V <sub>OL</sub> <sup>(1)(2)</sup>	Output low level voltage for an I/O pin when 8 pins are sunk at same time	I <sub>IO</sub> = +8 mA		0.4	
V <sub>OH</sub> <sup>(3)(2)</sup>	Output high level voltage for an I/O pin when 8 pins are sourced at same time	2.7 V < V <sub>DD</sub> < 3.6 V	2.4		
V <sub>OL</sub> (1)(4)	Output low level voltage for an I/O pin when 8 pins are sunk at same time	I <sub>IO</sub> =+ 4 mA 1.65 V < V <sub>DD</sub> <		0.45	v
V <sub>OH</sub> (3)(4)	Output high level voltage for an I/O pin when 8 pins are sourced at same time	2.7 V	V <sub>DD</sub> -0.45		V
V <sub>OL</sub> <sup>(1)(4)</sup>	Output low level voltage for an I/O pin when 4 pins are sunk at same time	I <sub>IO</sub> = +20 mA		1.3	
V <sub>OH</sub> <sup>(3)(4)</sup>	Output high level voltage for an I/O pin when 4 pins are sourced at same time	$I_{IO} = +20 \text{ mA}$ $2.7 \text{ V} < V_{DD} < 3.6 \text{ V}$	V <sub>DD</sub> -1.3		

The I<sub>IO</sub> current sunk by the device must always respect the absolute maximum rating specified in *Table 8* and the sum of I<sub>IO</sub> (I/O ports and control pins) must not exceed I<sub>VSS</sub>.

4. Based on characterization data, not tested in production.

<sup>2.</sup> Tested in production.

<sup>3.</sup> The  $I_{IO}$  current sourced by the device must always respect the absolute maximum rating specified in *Table 8* and the sum of  $I_{IO}$  (I/O ports and control pins) must not exceed  $I_{VDD}$ .

# Input/output AC characteristics

The definition and values of input/output AC characteristics are given in *Figure 23* and *Table 49*, respectively.

Unless otherwise specified, the parameters given in *Table 49* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 10*.

Table 49. I/O AC characteristics<sup>(1)</sup>

OSPEEDRx [1:0] bit value <sup>(1)</sup>	Symbol	Parameter	Conditions	Min	Max <sup>(2)</sup>	Unit
	4	Maximum frequency <sup>(3)</sup>	$C_L = 50 \text{ pF}, V_{DD} = 2.7 \text{ V to } 3.6 \text{ V}$		400	kHz
00	f <sub>max(IO)out</sub>	Maximum frequency	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 1.65 V to 2.7 V		TBD	KIIZ
00	t <sub>f(IO)out</sub>	Output rise and fall time	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 2.7 V to 3.6 V		625	no
	t <sub>r(IO)out</sub>	Output rise and fail time	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 1.65 V to 2.7 V		TBD	ns
	4	Maximum frequency <sup>(3)</sup>	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 2.7 V to 3.6 V		2	MHz
01	f <sub>max(IO)out</sub>	maximum frequency (**)	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 1.65 V to 2.7 V		1	IVITZ
01	t <sub>f(IO)out</sub>	$\begin{array}{c} t_{f(IO)out} \\ t_{r(IO)out} \end{array} \ \ \text{Output rise and fall time}$	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 2.7 V to 3.6 V		125	no
			C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 1.65 V to 2.7 V		TBD	ns
	Е	Maximum frequency <sup>(3)</sup>	$C_L = 50 \text{ pF}, V_{DD} = 2.7 \text{ V to } 3.6 \text{ V}$		10	MHz
10	F <sub>max(IO)out</sub>	Maximum frequency	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 1.65 V to 2.7 V		2	IVITZ
10	t <sub>f(IO)out</sub>	Output rise and fall time	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 2.7 V to 3.6 V		25	20
	t <sub>r(IO)out</sub>	Output rise and fail time	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 1.65 V to 2.7 V		TBD	ns
	F	Maximum frequency <sup>(3)</sup>	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 2.7 V to 3.6 V		50	MHz
11	F <sub>max(IO)out</sub>	maximum frequency.	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 1.65 V to 2.7 V		8	IVITZ
Į Į	t <sub>f(IO)out</sub>	Output rice and fall time	C <sub>L</sub> = 30 pF, V <sub>DD</sub> = 2.7 V to 3.6 V		5	
	t <sub>r(IO)out</sub>	Output rise and fall time	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 1.65 V to 2.7 V		TBD	
-	t <sub>EXTIpw</sub>	Pulse width of external signals detected by the EXTI controller		8		ns

The I/O speed is configured using the OSPEEDRx[1:0] bits. Refer to the STM32L15xxx reference manual for a description of GPIO Port configuration register.

<sup>2.</sup> Guaranteed by design. Not tested in production.

<sup>3.</sup> The maximum frequency is defined in Figure 23.

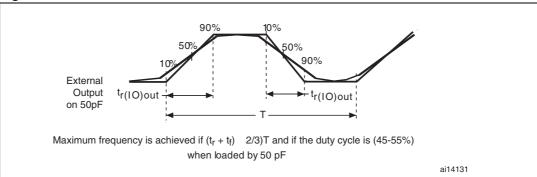


Figure 23. I/O AC characteristics definition

# 6.3.14 NRST pin characteristics

The NRST pin input driver uses CMOS technology.

Unless otherwise specified, the parameters given in *Table 50* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 10*.

Table 50. NRST pin characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>IL(NRST)</sub> <sup>(1)</sup>	NRST input low level voltage		$V_{SS}$		0.8	
V <sub>IH(NRST)</sub> <sup>(1)</sup>	NRST input high level voltage		1.4		$V_{DD}$	
V <sub>OL(NRST)</sub> <sup>(1)</sup>	NRST output low level voltage	$I_{OL} = 2 \text{ mA}$ 2.7 V < $V_{DD}$ < 3.6 V			0.4	٧
		I <sub>OL</sub> = 1.5 mA 1.65 V < V <sub>DD</sub> < 2.7 V			0.4	
V <sub>hys(NRST)</sub> <sup>(1)</sup>	NRST Schmitt trigger voltage hysteresis		10%V <sub>DD</sub> <sup>(2)</sup>			mV
R <sub>PU</sub>	Weak pull-up equivalent resistor <sup>(3)</sup>	$V_{IN} = V_{SS}$	30	45	60	kΩ
V <sub>F(NRST)</sub> <sup>(1)</sup>	NRST input filtered pulse				50	ns
V <sub>NF(NRST)</sub> <sup>(1)</sup>	NRST input not filtered pulse		350			ns

<sup>1.</sup> Guaranteed by design, not tested in production.

<sup>2. 200</sup> mV minimum value

The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance is around 10%.

External reset circuit<sup>(1)</sup>
NRST<sup>(2)</sup>
RPU
Filter Internal reset
STM32L15xxx

Figure 24. Recommended NRST pin protection

- 1. The reset network protects the device against parasitic resets.
- 2. The user must ensure that the level on the NRST pin can go below the  $V_{IL(NRST)}$  max level specified in *Table 50*. Otherwise the reset will not be taken into account by the device.

# 6.3.15 TIM timer characteristics

The parameters given in the following table are guaranteed by design.

Refer to *Section 6.3.12: I/O current injection characteristics* for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

Table 51. TIMx<sup>(1)</sup> characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
t(TIM)	Timer resolution time		1		t <sub>TIMxCLK</sub>
<sup>t</sup> res(TIM)	Timer resolution time	f <sub>TIMxCLK</sub> = 32 MHz	31.25		ns
f <sub>EXT</sub>	Timer external clock		0	f <sub>TIMxCLK</sub> /2	MHz
frequency on CH1 to CH4		f <sub>TIMxCLK</sub> = 32 MHz	0	16	MHz
Res <sub>TIM</sub>	Timer resolution			16	bit
	16-bit counter clock period		1	65536	t <sub>TIMxCLK</sub>
<sup>t</sup> COUNTER	when internal clock is selected (timer's prescaler disabled)	f <sub>TIMxCLK</sub> = 32 MHz	0.0312	2048	μs
t Maximum passible sount				65536 × 65536	t <sub>TIMxCLK</sub>
t <sub>MAX_COUNT</sub>	Maximum possible count	f <sub>TIMxCLK</sub> = 32 MHz		134.2	s

<sup>1.</sup> TIMx is used as a general term to refer to the TIM2, TIM3 and TIM4 timers.

#### 6.3.16 Communications interfaces

### I<sup>2</sup>C interface characteristics

Unless otherwise specified, the parameters given in *Table 52* are derived from tests performed under ambient temperature,  $f_{PCLK1}$  frequency and  $V_{DD}$  supply voltage conditions summarized in *Table 10*.

The STM32L151xD and STM32L152xD product line  $I^2C$  interface meets the requirements of the standard  $I^2C$  communication protocol with the following restrictions: SDA and SCL are not "true" open-drain I/O pins. When configured as open-drain, the PMOS connected between the I/O pin and  $V_{DD}$  is disabled, but is still present.

The I<sup>2</sup>C characteristics are described in *Table 52*. Refer also to *Section 6.3.12: I/O current injection characteristics* for more details on the input/output alternate function characteristics (SDA and SCL).

Table 52. I<sup>2</sup>C characteristics

Cumbal	Parameter	Standard	mode I <sup>2</sup> C <sup>(1)</sup>	Fast mode	e I <sup>2</sup> C <sup>(1)(2)</sup>	Unit
Symbol	Parameter	Min	Max	Min	Max	Unit
t <sub>w(SCLL)</sub>	SCL clock low time	4.7		1.3		
t <sub>w(SCLH)</sub>	SCL clock high time	4.0		0.6		μs
t <sub>su(SDA)</sub>	SDA setup time	250		100		
t <sub>h(SDA)</sub>	SDA data hold time	0(3)		0 <sup>(4)</sup>	900 <sup>(3)</sup>	
t <sub>r(SDA)</sub> t <sub>r(SCL)</sub>	SDA and SCL rise time		1000	20 + 0.1C <sub>b</sub>	300	ns
t <sub>f(SDA)</sub>	SDA and SCL fall time		300		300	
t <sub>h(STA)</sub>	Start condition hold time	4.0		0.6		
t <sub>su(STA)</sub>	Repeated Start condition setup time	4.7		0.6		μs
t <sub>su(STO)</sub>	Stop condition setup time	4.0		0.6		μs
t <sub>w(STO:STA)</sub>	Stop to Start condition time (bus free)	4.7		1.3		μs
C <sub>b</sub>	Capacitive load for each bus line		400		400	pF

<sup>1.</sup> Guaranteed by design, not tested in production.

f<sub>PCLK1</sub> must be higher than 2 MHz to achieve standard mode I<sup>2</sup>C frequencies. It must be higher than 4 MHz
to achieve fast mode I<sup>2</sup>C frequencies. It must be a multiple of 10 MHz to reach the 400 kHz maximum I<sup>2</sup>C
fast mode clock.

<sup>3.</sup> The maximum hold time of the Start condition has only to be met if the interface does not stretch the low period of SCL signal.

The device must internally provide a hold time of at least 300 ns for the SDA signal in order to bridge the undefined region of the falling edge of SCL.

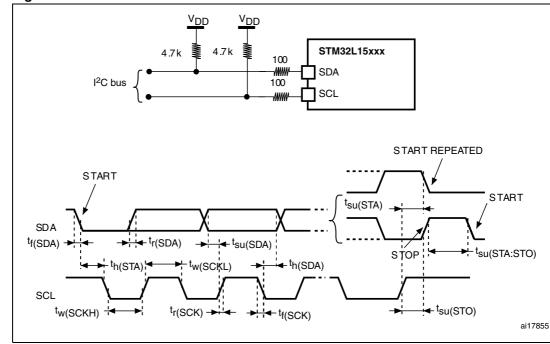


Figure 25. I<sup>2</sup>C bus AC waveforms and measurement circuit

1. Measurement points are done at CMOS levels:  $0.3V_{DD}$  and  $0.7V_{DD}$ .

Table 53. SCL frequency  $(f_{PCLK1} = 32 \text{ MHz}, V_{DD} = 3.3 \text{ V})^{(1)(2)}$ 

f (/rU=\	I2C_CCR value
f <sub>SCL</sub> (kHz)	$R_{p}$ = 4.7 k $\Omega$
400	0x801B
300	0x8024
200	0x8035
100	0x00A0
50	0x0140
20	0x0320

<sup>1.</sup>  $R_P$  = External pull-up resistance,  $f_{SCL} = I^2C$  speed.

For speeds around 200 kHz, the tolerance on the achieved speed is of ±5%. For other speed ranges, the
tolerance on the achieved speed is ±2%. These variations depend on the accuracy of the external
components used to design the application.

#### **SPI** characteristics

Unless otherwise specified, the parameters given in the following table are derived from tests performed under ambient temperature,  $f_{PCLKx}$  frequency and  $V_{DD}$  supply voltage conditions summarized in *Table 10*.

Refer to *Section 6.3.12: I/O current injection characteristics* for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO).

Table 54. SPI characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Max <sup>(2)</sup>	Unit
f <sub>SCK</sub>	SPI clock frequency	Master mode		16	MHz
1/t <sub>c(SCK)</sub>	SFI Clock frequency	Slave mode		16	IVITIZ
t <sub>r(SCK)</sub>	SPI clock rise and fall time	Capacitive load: C = 30 pF		TBD	ns
DuCy(SCK)	SPI slave input clock duty cycle	Slave mode	30	70	%
t <sub>su(NSS)</sub>	NSS setup time	Slave mode	4t <sub>PCLK</sub>		
t <sub>h(NSS)</sub>	NSS hold time	Slave mode	2t <sub>PCLK</sub>		
t <sub>w(SCKH)</sub>	SCK high and low time	Master mode, f <sub>PCLK</sub> = 16 MHz, presc = 4	TBD	TBD	
t <sub>su(MI)</sub>	Data input setup time	Master mode	5		
t <sub>su(SI)</sub>	Data input setup time	Slave mode	5		
t <sub>h(MI)</sub>	Data input hold time	Master mode	5		
t <sub>h(SI)</sub>	Data input noid time	Slave mode	4		ns
t <sub>a(SO)</sub> (3)	Data output access time	Slave mode, f <sub>PCLK</sub> = 20 MHz	0	3t <sub>PCLK</sub>	
t <sub>dis(SO)</sub> <sup>(4)</sup>	Data output disable time	Slave mode	TBD	TBD	
t <sub>v(SO)</sub> (2)	Data output valid time	Slave mode (after enable edge)		TBD	
t <sub>v(MO)</sub> <sup>(2)</sup>	Data output valid time	Master mode (after enable edge)		TBD	
t <sub>h(SO)</sub> <sup>(2)</sup>	Data output hold time	Slave mode (after enable edge)	TBD		
t <sub>h(MO)</sub> <sup>(2)</sup>	Data output hold time	Master mode (after enable edge)	TBD		

<sup>1.</sup> Remapped SPI1 characteristics to be determined.

<sup>2.</sup> Based on characterization, not tested in production.

Min time is for the minimum time to drive the output and max time is for the maximum time to validate the data.

<sup>4.</sup> Min time is for the minimum time to invalidate the output and max time is for the maximum time to put the data in Hi-Z.

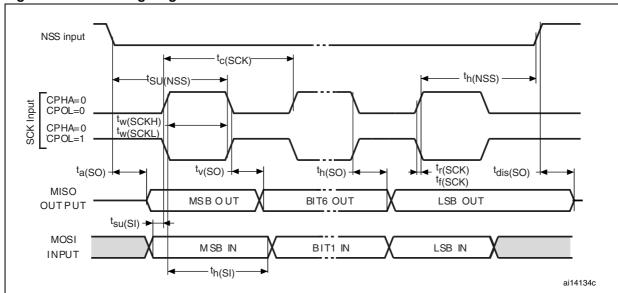
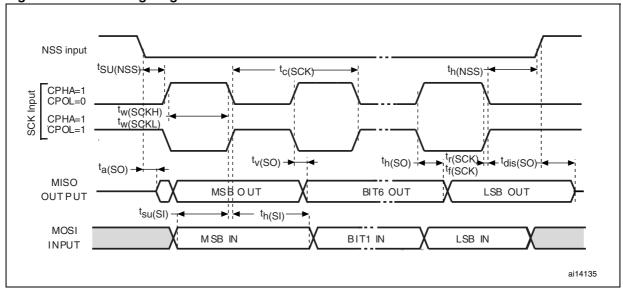


Figure 26. SPI timing diagram - slave mode and CPHA = 0





1. Measurement points are done at CMOS levels:  $0.3V_{DD}$  and  $0.7V_{DD}$ .

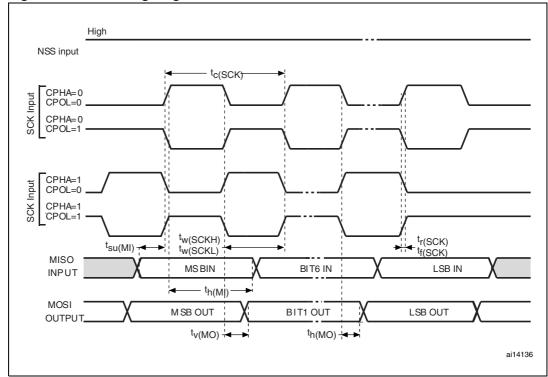


Figure 28. SPI timing diagram - master mode<sup>(1)</sup>

1. Measurement points are done at CMOS levels:  $0.3V_{DD}$  and  $0.7V_{DD}$ .

### **USB** characteristics

The USB interface is USB-IF certified (full speed).

Table 55. USB startup time

Symbol	Parameter	Max	Unit
t <sub>STARTUP</sub> <sup>(1)</sup>	USB transceiver startup time	1	μs

1. Guaranteed by design, not tested in production.

Table 56. USB DC electrical characteristics

Symbol	Parameter	Conditions	Min. <sup>(1)</sup>	Max. <sup>(1)</sup>	Unit
Input levels					
V <sub>DD</sub>	USB operating voltage		3.0	3.6	V
V <sub>DI</sub> <sup>(2)</sup>	Differential input sensitivity	I(USBDP, USBDM)	0.2		
V <sub>CM</sub> <sup>(2)</sup>	Differential common mode range	Includes V <sub>DI</sub> range	0.8	2.5	V
V <sub>SE</sub> <sup>(2)</sup>	Single ended receiver threshold		1.3	2.0	
Output lev	vels				
V <sub>OL</sub> <sup>(3)</sup>	Static output level low	R <sub>L</sub> of 1.5 kΩ to 3.6 V <sup>(4)</sup>		0.3	V
V <sub>OH</sub> <sup>(3)</sup>	Static output level high	$R_L$ of 15 kΩ to $V_{SS}^{(4)}$	2.8	3.6	] <b>v</b>

- 1. All the voltages are measured from the local ground potential.
- 2. Guaranteed by characterization, not tested in production.
- 3. Tested in production.
- 4.  $R_L$  is the load connected on the USB drivers.

Figure 29. USB timings: definition of data signal rise and fall time

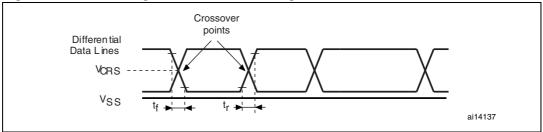


Table 57. USB: full speed electrical characteristics

	Driver characteristics <sup>(1)</sup>							
Symbol	Parameter	Conditions	Min	Max	Unit			
t <sub>r</sub>	Rise time <sup>(2)</sup>	C <sub>L</sub> = 50 pF	4	20	ns			
t <sub>f</sub>	Fall Time <sup>(2)</sup>	C <sub>L</sub> = 50 pF	4	20	ns			
t <sub>rfm</sub>	Rise/ fall time matching	t <sub>r</sub> /t <sub>f</sub>	90	110	%			
V <sub>CRS</sub>	Output signal crossover voltage		1.3	2.0	V			

- 1. Guaranteed by design, not tested in production.
- Measured from 10% to 90% of the data signal. For more detailed informations, please refer to USB Specification - Chapter 7 (version 2.0).

# 6.3.17 12-bit ADC characteristics

Unless otherwise specified, the parameters given in *Table 59* are guaranteed by design.

Table 58. ADC clock frequency

Symbol	Parameter		Conditions		Min	Max	Unit
				$V_{REF+} = V_{DDA}$		16	
f <sub>ADC</sub>			2.4 V ≤V <sub>DDA</sub> ≤3.6 V	$V_{REF+} < V_{DDA}$ $V_{REF+} > 2.4 V$		8	
	ADC clock frequency	Voltage range 1 & 2		$V_{REF+} < V_{DDA}$ $V_{REF+} \le 2.4 \text{ V}$	0.480	4	MHz
			1914 041	$V_{REF+} = V_{DDA}$		8	
			1.8 V ≤V <sub>DDA</sub> ≤2.4 V	$V_{REF+} < V_{DDA}$		4	
			Voltage range 3			4	

Table 59. ADC characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
V <sub>DDA</sub>	Power supply		1.8		3.6		
V <sub>REF+</sub>	Positive reference voltage	2.4 V ≤V <sub>DDA</sub> ≤3.6 V V <sub>REF+</sub> must be below or equal to V <sub>DDA</sub>	1.8 <sup>(1)</sup>		V <sub>DDA</sub>	V	
V <sub>REF-</sub>	Negative reference voltage			$V_{SSA}$			
I <sub>VDDA</sub>	Current on the V <sub>DDA</sub> input pin			1000	1450	μA	
(2)	Current on the V <sub>REF</sub> input	Peak		400	700		
I <sub>VREF</sub> <sup>(2)</sup>	pin	Average		400	450		
V <sub>AIN</sub>	Conversion voltage range <sup>(3)</sup>		0 <sup>(4)</sup>		V <sub>REF+</sub>	V	
	40 lett a annullin a mate	Direct channels	0.03		1	Msps	
	12-bit sampling rate	Multiplexed channels	0.03		0.76	ivisps	
	10-bit sampling rate	Direct channels	0.03		1.07	Msps	
f <sub>S</sub>	10-bit sampling rate	Multiplexed channels	0.03		0.8	ivisps	
'S	9 hit compling rate	Direct channels	0.03		1.23	Mono	
	8-bit sampling rate	Multiplexed channels	0.03		0.89	Msps	
	6 hit campling rate	Direct channels	0.03		1.54	Mana	
	6-bit sampling rate	Multiplexed channels	0.03		1	Msps	

Table 59. ADC characteristics (continued)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
		Direct channels 2.4 V ≤V <sub>DDA</sub> ≤3.6 V	0.25 <sup>(5)</sup>			
		Multiplexed channels 2.4 V ≤V <sub>DDA</sub> ≤3.6 V	0.56 <sup>(5)</sup>			
t <sub>S</sub>	Sampling time	Direct channels 1.8 V ≤V <sub>DDA</sub> ≤2.4 V	0.56 <sup>(5)</sup>			μs
		Multiplexed channels 1.8 V ≤V <sub>DDA</sub> ≤2.4 V	1 <sup>(5)</sup>			
			4		384	1/f <sub>ADC</sub>
		f <sub>ADC</sub> = 16 MHz	1		24.75	μs
<sup>t</sup> CONV	Total conversion time (including sampling time)		4 to 384 phase) approxi	+12 (su	1/f <sub>ADC</sub>	
C <sub>ADC</sub>	Internal sample and hold	Direct channels		16		nΕ
ADC	capacitor	Multiplexed channels		10		pF
f <sub>TRIG</sub>	External trigger frequency	12-bit conversions			Tconv+1	1/f <sub>ADC</sub>
TRIG	Regular sequencer	6/8/10-bit conversions			Tconv	1/f <sub>ADC</sub>
f <sub>TRIG</sub>	External trigger frequency	12-bit conversions			Tconv+2	1/f <sub>ADC</sub>
TRIG	Injected sequencer	6/8/10-bit conversions			Tconv+1	1/f <sub>ADC</sub>
R <sub>AIN</sub> (6)	External input impedance				50	kΩ
PAIN'	External input impedance				0.5	K22
t <sub>lat</sub>	Injection trigger conversion	f <sub>ADC</sub> = 16 MHz	219		281	ns
'lat	latency		3.5		4.5	1/f <sub>ADC</sub>
t <sub>latr</sub>	Regular trigger conversion	f <sub>ADC</sub> = 16 MHz	156		219	ns
чatr	latency		2.5		3.5	1/f <sub>ADC</sub>
t <sub>STAB</sub>	Power-up time				3.5	μs

<sup>1.</sup> The Vref+ input can be grounded iif neither the ADC nor the DAC are used (this allows to shut down an external voltage reference).

- 2. The current consumption through VREF is composed of two parameters:
  - one constant (max 300 μA)
  - one variable (max 400  $\mu\text{A})\text{, only during sampling time} + 2 first conversion pulses.$

So, peak consumption is 300+400 = 700  $\mu A$  and average consumption is 300 + [(4 sampling + 2) /16] x 400 = 450  $\mu A$  at 1Msps

- 3. V<sub>REF+</sub> can be internally connected to V<sub>DDA</sub> and V<sub>REF-</sub> can be internally connected to V<sub>SSA</sub>, depending on the package. Refer to *Section 4: Pin descriptions* for further details.
- 4.  $V_{SSA}$  or  $V_{REF-}$  must be tied to ground.
- 5. Minimum sampling and conversion time is reached for maximum Rext = 0.5  $k\Omega$
- 6. For 1 Msps, maximum Rext is 0.5  $\text{k}\Omega$

ADC accuracy<sup>(1)(2)</sup> Table 60.

Symbol	Parameter	Test conditions	Min <sup>(3)</sup>	Тур	Max <sup>(3)</sup>	Unit
ET	Total unadjusted error		-	2	4	
EO	Offset error	2.4 V ≤V <sub>DDA</sub> ≤ 3.6 V	-	1	2	
EG	Gain error	$2.4 \text{ V} \le \text{V}_{REF+} \le 3.6 \text{ V}$ $f_{ADC} = 8 \text{ MHz}, R_{AIN} = 50 \Omega$	-	1.5	3.5	LSB
ED	Differential linearity error	$T_A = -40 \text{ to } 105 ^{\circ}\text{C}$	-	1	2	
EL	Integral linearity error		-	1.7	3	
ENOB	Effective number of bits	247/27 2267	9.2	10	-	bits
SINAD	Signal-to-noise and distorsion ratio	$-2.4 \text{ V} \leq \text{V}_{DDA} \leq 3.6 \text{ V}$ $\text{V}_{DDA} = \text{V}_{REF+}$ $\text{f}_{ADC} = 16 \text{ MHz}, \text{R}_{AIN} = 50 \Omega$	57.5	62	-	
SNR	Signal-to-noise ratio	T <sub>A</sub> = -40 to 105 ° C	57.5	62	-	dB
THD	Total harmonic distorsion	1 kHz ≤F <sub>input</sub> ≤ 100 kHz	-74	-75	-	
ET	Total unadjusted error		-	4	6.5	
EO	Offset error	2.4 V ≤V <sub>DDA</sub> ≤ 3.6 V	-	2	4	
EG	Gain error	1.8 V $\leq$ V <sub>REF+</sub> $\leq$ 2.4 V $\int$ f <sub>ADC</sub> = 4 MHz, R <sub>AIN</sub> = 50 $\Omega$	-	4	6	LSB
ED	Differential linearity error	$T_A = -40 \text{ to } 105 ^{\circ}\text{C}$	-	1	2	
EL	Integral linearity error		-	1.5	3	
ET	Total unadjusted error			2	3	
EO	Offset error	1.8 V ≤V <sub>DDA</sub> ≤ 2.4 V		1	1.5	
EG	Gain error	1.8 V $\leq$ V <sub>REF+</sub> $\leq$ 2.4 V $f_{ADC}$ = 4 MHz, $R_{AIN}$ = 50 $\Omega$		1.5	2	LSB
ED	Differential linearity error	$T_A = -40 \text{ to } 105 ^{\circ}\text{C}$		1	2	
EL	Integral linearity error			1	1.5	

<sup>1.</sup> ADC DC accuracy values are measured after internal calibration.

ADC accuracy vs. negative injection current: injecting negative current on any of the standard (non-robust) analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to standard analog pins which may potentially inject negative current.

Any positive injection current within the limits specified for I<sub>INJ(PIN)</sub> and  $\Sigma$ I<sub>INJ(PIN)</sub> in *Section 6.3.12* does not affect the ADC accuracy.

<sup>3.</sup> Based on characterization, not tested in production.

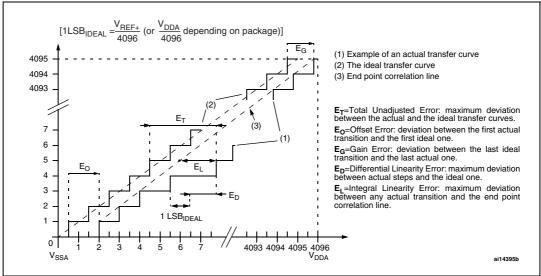
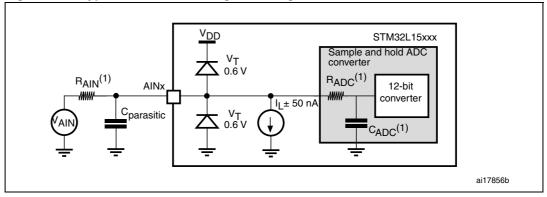


Figure 30. ADC accuracy characteristics

Figure 31. Typical connection diagram using the ADC



- 1. Refer to Table 59 for the values of  $R_{AIN}$ ,  $R_{ADC}$  and  $C_{ADC}$ .
- 2. C<sub>parasitic</sub> represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (roughly 7 pF). A high C<sub>parasitic</sub> value will downgrade conversion accuracy. To remedy this, f<sub>ADC</sub> should be reduced.

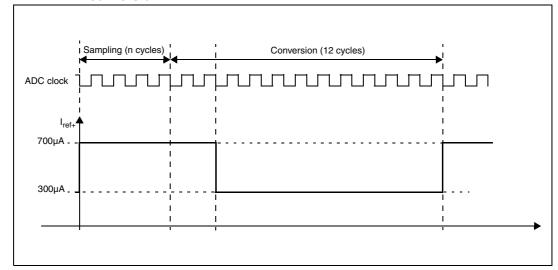


Figure 32. Maximum dynamic current consumption on V<sub>REF+</sub> supply pin during ADC conversion

Table 61.  $R_{AIN}$  max for  $f_{ADC} = 16 \text{ MHz}^{(1)}$ 

	Ally	ADC	R <sub>AIN</sub> m	nax (kΩ)		
Ts (cycles)	Ts (µs)	Multipleyed channels		Direct channels		
		2.4 V < V <sub>DDA</sub> < 3.6 V	1.8 V < V <sub>DDA</sub> < 2.4 V	2.4 V < V <sub>DDA</sub> < 3.3 V	1.8 V < V <sub>DDA</sub> < 2.4 V	
4	0.25	Not allowed	Not allowed	0.7	Not allowed	
9	0.5625	0.8	Not allowed	2.0	1.0	
16	1	2.0	0.8	4.0	3.0	
24	1.5	3.0	1.8	6.0	4.5	
48	3	6.8	4.0	15.0	10.0	
96	6	15.0	10.0	30.0	20.0	
192	12	32.0	25.0	50.0	40.0	
384	24	50.0	50.0	50.0	50.0	

<sup>1.</sup> Guaranteed by design, not tested in production.

# General PCB design guidelines

Power supply decoupling should be performed as shown in *Figure 33* or *Figure 34*, depending on whether  $V_{REF+}$  is connected to  $V_{DDA}$  or not. The 10 nF capacitors should be ceramic (good quality). They should be placed as close as possible to the chip.

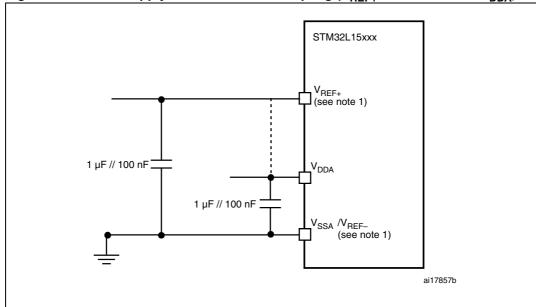


Figure 33. Power supply and reference decoupling ( $V_{REF+}$  not connected to  $V_{DDA}$ )

1.  $V_{\text{REF+}}$  and  $V_{\text{REF-}}$  inputs are available only on 100-pin packages.

Figure 34. Power supply and reference decoupling (V<sub>REF+</sub> connected to V<sub>DDA</sub>)

STM32L15xxx

V<sub>REF+</sub>V<sub>DDA</sub>
(See note 1)

1 μF // 100 nF

V<sub>REF-</sub>V<sub>SSA</sub>
(See note 1)

1.  $V_{\text{REF+}}$  and  $V_{\text{REF-}}$  inputs are available only on 100-pin packages.

# 6.3.18 DAC electrical specifications

Data guaranteed by design, not tested in production, unless otherwise specified.

Table 62. DAC characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
$V_{DDA}$	Analog supply voltage		1.8		3.6	
V <sub>REF+</sub>	Reference supply voltage	V <sub>REF+</sub> must always be below V <sub>DDA</sub>	1.8		3.6	V
V <sub>REF-</sub>	Lower reference voltage			V <sub>SS</sub>	A	
. (1)	Current consumption on	No load, middle code (0x800)		130	220	
I <sub>DDVREF+</sub> (1)	V <sub>REF+</sub> supply V <sub>REF+</sub> = 3.3 V	No load, worst code (0x000)		220	350	^
. (1)	Current consumption on	No load, middle code (0x800)		210	320	μA
I <sub>DDA</sub> <sup>(1)</sup>	V <sub>DDA</sub> supply V <sub>DDA</sub> = 3.3 V	No load, worst code (0xF1C)		320	520	
R <sub>L</sub> <sup>(2)</sup>	Resistive load	DAC output buffor ON	5			kΩ
C <sub>L</sub> <sup>(2)</sup>	Capacitive load	DAC output buffer ON			50	pF
R <sub>O</sub>	Output impedance	DAC output buffer OFF	6	8	10	kΩ
	Voltage on DAC_OUT output	DAC output buffer ON	0.2		V <sub>DDA</sub> – 0.2	٧
V <sub>DAC_OUT</sub>		DAC output buffer OFF	0.5		V <sub>REF+</sub> – 1LSB	mV
DNL <sup>(1)</sup>	Differential non linearity <sup>(3)</sup>	$C_L \le 50$ pF, $R_L \ge 5$ k $\Omega$ DAC output buffer ON		1.5	3	
		No R <sub>LOAD</sub> , $C_L \le 50 \text{ pF}$ DAC output buffer OFF		1.5	3	
INL <sup>(1)</sup>	Integral non linearity <sup>(4)</sup>	$C_L \le 50$ pF, $R_L \ge 5$ k $\Omega$ DAC output buffer ON		2	4	
IINL '	integral non linearity.	No $R_{LOAD}$ , $C_{L} \le 50 pF$ DAC output buffer OFF		2	4	LSB
Offset <sup>(1)</sup>	Offset error at code	$C_L \le 50$ pF, $R_L \ge 5$ k $\Omega$ DAC output buffer ON		±10	±25	
Onser /	0x800 <sup>(5)</sup>	No R <sub>LOAD</sub> , $C_L \le 50 \text{ pF}$ DAC output buffer OFF		±5	±8	
Offset1 <sup>(1)</sup>	Offset error at code 0x001 <sup>(6)</sup>	No R <sub>LOAD</sub> , C <sub>L</sub> ≤ 50 pF DAC output buffer OFF		±1.5	±5	

Table 62. DAC characteristics (continued)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
dOffset/dT <sup>(1)</sup>	Offset error temperature	$V_{DDA} = 3.3V$ $V_{REF+} = 3.0V$ $T_A = 0$ to 50 ° C DAC output buffer OFF	-20	-10	0	μV/°C
dOliseva i V	coefficient (code 0x800)	$V_{DDA} = 3.3V$ $V_{REF+} = 3.0V$ $T_A = 0$ to 50 ° C DAC output buffer ON	0	20	50	μν/ С
Gain <sup>(1)</sup>	Gain arror <sup>(7)</sup>	$C_L \le 50 \text{ pF, } R_L \ge 5 \text{ k}\Omega$ DAC output buffer ON		+0.1 / -0.2%	+0.2 / -0.5%	%
Gain'	Gain error <sup>(7)</sup>	No R <sub>LOAD</sub> , C <sub>L</sub> ≤ 50 pF DAC output buffer OFF		+0 / -0.2%	+0 / -0.4%	70
dGain/dT <sup>(1)</sup>	$ \begin{array}{c} \text{Gain error temperature} \\ \text{Coefficient} \end{array} \begin{array}{c} \text{V}_{DDA} = 3.3 \text{V} \\ \text{V}_{REF_{+}} = 3.0 \text{V} \\ \text{T}_{A} = 0 \text{ to } 50 ^{\circ}\text{C} \\ \text{DAC output buffer OFF} \end{array} \begin{array}{c} -10 \\ \text{V}_{DDA} = 3.3 \text{V} \\ \text{V}_{REF_{+}} = 3.0 \text{V} \\ \text{T}_{A} = 0 \text{ to } 50 ^{\circ}\text{C} \\ \text{DAC output buffer ON} \end{array} \begin{array}{c} -40 \\ \text{-8} \\ 0 \end{array} $	0	μV/°C			
dGain/dT(1)		$V_{REF+} = 3.0V$ $T_A = 0 \text{ to } 50 ^{\circ}\text{C}$	-40	-8	0	μν/ Ο
TUE <sup>(1)</sup>	Tatal was divisted a sure	$C_L \le 50 \text{ pF, } R_L \ge 5 \text{ k}\Omega$ DAC output buffer ON		12	30	LSB
I UEV	Total unadjusted error	No R <sub>LOAD</sub> , C <sub>L</sub> ≤ 50 pF DAC output buffer OFF		8	12	LOD
<sup>t</sup> SETTLING	Settling time (full scale: for a 12-bit code transition between the lowest and the highest input codes till DAC_OUT reaches final value ±1LSB	$C_L \le 50 \text{ pF, } R_L \ge 5 \text{ k}\Omega$		7	12	μs
Update rate	Max frequency for a correct DAC_OUT change (95% of final value) with 1 LSB variation in the input code	$C_L \le 50 \text{ pF, } R_L \ge 5 \text{ k}\Omega$			1	Msps
<sup>t</sup> WAKEUP	Wakeup time from off state (setting the ENx bit in the DAC Control register) <sup>(8)</sup>	$C_L \le 50 \text{ pF, } R_L \ge 5 \text{ k}\Omega$		9	15	μs
PSRR+	V <sub>DDA</sub> supply rejection ratio (static DC measurement)	$C_L \le 50 \text{ pF, } R_L \ge 5 \text{ k}\Omega$		-60	-35	dB

<sup>1.</sup> Data based on characterization results.

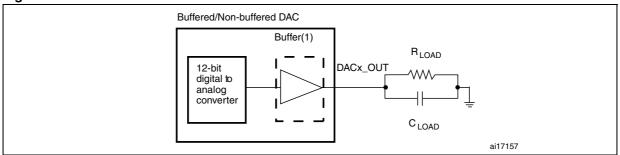
<sup>2.</sup> Connected between DAC\_OUT and  $V_{\mbox{SSA}}$ .

<sup>3.</sup> Difference between two consecutive codes - 1 LSB.

<sup>4.</sup> Difference between measured value at Code i and the value at Code i on a line drawn between Code 0 and last Code 4095.

- 5. Difference between the value measured at Code (0x800) and the ideal value =  $V_{REF+}/2$ .
- 6. Difference between the value measured at Code (0x001) and the ideal value.
- 7. Difference between ideal slope of the transfer function and measured slope computed from code 0x000 and 0xFFF when buffer is OFF, and from code giving 0.2 V and  $(V_{DDA} 0.2)$  V when buffer is ON.
- 8. In buffered mode, the output can overshoot above the final value for low input code (starting from min value).

Figure 35. 12-bit buffered /non-buffered DAC



The DAC integrates an output buffer that can be used to reduce the output impedance and to drive external loads directly
without the use of an external operational amplifier. The buffer can be bypassed by configuring the BOFFx bit in the
DAC\_CR register.

# 6.3.19 Operational amplifier characteristics

Table 63. Operational amplifier characteristics

Symbol	Para	meter	Condition <sup>(1)</sup>	Min <sup>(2)</sup>	Тур	Max <sup>(2)</sup>	Unit
CMIR	Common mode inpu	it range		0		$V_{DD}$	
M	Input offeet voltage	Maximum calibration range				±15	mV
VI <sub>OFFSET</sub>	Input offset voltage	After offset calibration				±1.5	IIIV
A\/  .	Input offset voltage	Normal mode				±40	μV/°C
ΔVI <sub>OFFSET</sub>	drift	Low power mode				±80	
I <sub>IB</sub>		Dedicated input				1	
	Input current bias	General purpose input	75 °C			10	nA
	Drive current	Normal mode				500	
I <sub>LOAD</sub>	Drive current	Low power mode				100	μA
ı	Concumption	Normal mode	No load,		100	220	
I <sub>DD</sub>	Consumption	Low power mode	quiescent mode		30	60	μA
CMDD	Common mode	Normal mode			-85		dB
CMRR	rejection ration	Low power mode			-90		uБ
PSRR	Power supply	Normal mode	- DC		-85		٩D
ronn	rejection ratio	Low power mode	DC		-90		dB

 Table 63.
 Operational amplifier characteristics (continued)

Symbol	Parameter		Condition <sup>(1)</sup>	Min <sup>(2)</sup>	Тур	Max <sup>(2)</sup>	Unit
		Normal mode	V - 24 V	400	1000	3000	
GBW	D de cialda	Low power mode	V <sub>DD</sub> >2.4 V	150	300	800	kHZ
GBW	Bandwidth	Normal mode	V 0.4.V	200	500	2200	
		Low power mode	V <sub>DD</sub> <2.4 V	70	150	800	
		Normal mode	V <sub>DD</sub> >2.4 V (between 0.1 V and V <sub>DD</sub> -0.1 V)		700		
SR	Slew rate	Low power mode	V <sub>DD</sub> >2.4 V		100		V/ms
		Normal mode	V -0.4.V		300		
		Low power mode	─ V <sub>DD</sub> <2.4 V		50		
AO	Onen leen gein	Normal mode		55	100		dB
AU	Open loop gain	Low power mode	w power mode		110		иь
В	B : :: 1 :	Normal mode	V 0.4.V	4			1.0
R <sub>LOAD</sub>	Resistive load	Low power mode	V <sub>DD</sub> <2.4 V	20			kΩ
C <sub>LOAD</sub>	Capacitive load	,				50	pF
VOH <sub>SAT</sub>	High saturation	Normal mode		V <sub>DD</sub> - 100			
0,	voltage	Low power mode	I <sub>LOAD</sub> = max or	V <sub>DD</sub> -50			mV
VOI	Low saturation	Normal mode	$R_{LOAD} = min$			100	
VOL <sub>SAT</sub>	voltage	low power mode				50	
φm	Phase margin	-			60		0
GM	Gain margin				-12		dB
t <sub>OFFTRIM</sub>	Offset trim time: during calibration, minimum time needed between two steps to have 1 mV accuracy				1		ms
twarens	Wakeun time	Normal mode	$\begin{aligned} &C_{LOAD} \leq 50 \text{ pf,} \\ &R_{LOAD} \geq 4 \text{ k}\Omega \end{aligned}$		10		μs
<sup>t</sup> WAKEUP	Wakeup time	Low power mode	$C_{LOAD} \le 50 \text{ pf},$ $R_{LOAD} \ge 20 \text{ k}\Omega$		30		μο

<sup>1.</sup> Operating conditions are limited to junction temperature (0 °C to 105 °C) when V<sub>DD</sub> is below 2 V. Otherwise, the operating temperature range is 105 °C to -40 °C.

<sup>2.</sup> Data based on characterization results, not tested in production.

### 6.3.20 Temperature sensor characteristics

Table 64. TS characteristics

Symbol	Parameter	Min	Тур	Max	Unit
T <sub>L</sub> <sup>(1)</sup>	V <sub>SENSE</sub> linearity with temperature		±1	<u>+2</u>	°C
Avg_Slope <sup>(1)</sup>	Average slope	TBD	1.66	TBD	mV/°C
V <sub>110</sub>	Voltage at 110°C ±5°C <sup>(2)</sup>	612	626.8	641.5	mV
I <sub>DDA</sub> (TEMP) <sup>(3)</sup>	Current consumption		3.4	6	μΑ
t <sub>START</sub> (3)	Startup time			10	
T <sub>S_temp</sub> <sup>(4)(3)</sup>	ADC sampling time when reading the temperature		5	10	μs

- 1. Guaranteed by characterization, not tested in production.
- 2. Measured at  $V_{DD}$  = 3 V ±10 mV. V110 ADC conversion result is stored in the TS\_Factory\_CONV\_V110 byte.
- 3. Guaranteed by design, not tested in production.
- 4. Shortest sampling time can be determined in the application by multiple iterations.

## 6.3.21 Comparator

Table 65. Comparator 1 characteristics

Symbol	Parameter	Conditions	Min <sup>(1)</sup>	Тур	Max <sup>(1)</sup>	Unit
$V_{DDA}$	Analog supply voltage		1.65		3.6	V
R <sub>400K</sub>	R <sub>400K</sub> value			400		kΩ
R <sub>10K</sub>	R <sub>10K</sub> value			10		K22
V <sub>IN</sub>	Comparator 1 input voltage range		0.6		$V_{DDA}$	V
t <sub>START</sub>	Comparator startup time			7	10	110
td	Propagation delay <sup>(2)</sup>			3	10	μs
Voffset	Comparator offset			Œ	±10	mV
d <sub>Voffset</sub> /dt	Comparator offset variation in worst voltage stress conditions	$V_{DDA} = 3.6 \text{ V}$ $V_{IN+} = 0 \text{ V}$ $V_{IN-} = V_{REFINT}$ $T_A = 25 \text{ °C}$	0	1.5	10	mV/1000 h
I <sub>COMP1</sub>	Current consumption <sup>(3)</sup>			160	260	nA

<sup>1.</sup> Based on characterization, not tested in production.

The delay is characterized for 100 mV input step with 10 mV overdrive on the inverting input, the non-inverting input set to the reference.

<sup>3.</sup> Comparator consumption only. Internal reference voltage not included.

Table 66. Comparator 2 characteristics

Symbol	Parameter	Conditions	Min	Тур	Max <sup>(1)</sup>	Unit	
$V_{DDA}$	Analog supply voltage		1.65		3.6	٧	
V <sub>IN</sub>	Comparator 2 input voltage range		0		$V_{DDA}$	V	
+.	Comparator startup time	Fast mode		15	20		
t <sub>START</sub>	Comparator startup time	Slow mode		20	25		
	Propagation delay <sup>(2)</sup> in slow mode	1.65 V ≤V <sub>DDA</sub> ≤2.7 V		1.8	3.5		
t <sub>d slow</sub>	Propagation delay. 7 in slow mode	2.7 V ≤V <sub>DDA</sub> ≤3.6 V		2.5	6	μs	
	Propagation delay <sup>(2)</sup> in fast mode	1.65 V ≤V <sub>DDA</sub> ≤2.7 V		0.8	2		
t <sub>d fast</sub>	Propagation delay. 7 in last mode	2.7 V ≤V <sub>DDA</sub> ≤3.6 V		1.2	4		
V <sub>offset</sub>	Comparator offset error			±4	±20	mV	
dThreshold/ dt	Threshold voltage temperature coefficient	$\begin{split} V_{DDA} &= 3.3 V \\ T_{A} &= 0 \text{ to } 50 \text{ °C} \\ V &= V_{REF+}, 3/4 \\ V_{REF+}, 1/2 \ V_{REF+}, 1/4 \ V_{REF+}. \end{split}$		15	30	ppm /°C	
l	Current consumption <sup>(3)</sup>	Fast mode		3.5	5	μΑ	
I <sub>COMP2</sub>	Current consumptions	Slow mode		0.5	2		

<sup>1.</sup> Based on characterization, not tested in production.

<sup>2.</sup> The delay is characterized for 100 mV input step with 10 mV overdrive on the inverting input, the non-inverting input set to the reference.

<sup>3.</sup> Comparator consumption only. Internal reference voltage (necessary for comparator operation) is not included.

## 6.3.22 LCD controller (STM32L152xD only)

The STM32L152xD embeds a built-in step-up converter to provide a constant LCD reference voltage independently from the  $V_{DD}$  voltage. An external capacitor  $C_{\text{ext}}$  must be connected to the  $V_{\text{LCD}}$  pin to decouple this converter.

Table 67. LCD controller characteristics

Symbol	Parameter	Min	Тур	Max	Unit
$V_{LCD}$	LCD external voltage			3.6	
$V_{LCD0}$	LCD internal reference voltage 0		2.6		
V <sub>LCD1</sub>	LCD internal reference voltage 1		2.73		
V <sub>LCD2</sub>	LCD internal reference voltage 2		2.86		
V <sub>LCD3</sub>	LCD internal reference voltage 3		2.98		V
V <sub>LCD4</sub>	LCD internal reference voltage 4		3.12		
V <sub>LCD5</sub>	LCD internal reference voltage 5		3.26		
V <sub>LCD6</sub>	LCD internal reference voltage 6		3.4		
V <sub>LCD7</sub>	LCD internal reference voltage 7		3.55		
C <sub>ext</sub>	V <sub>LCD</sub> external capacitance	0.1		2	μF
I <sub>LCD</sub> <sup>(1)</sup>	Supply current at V <sub>DD</sub> = 2.2 V		3.3		
ILCD` ′	Supply current at V <sub>DD</sub> = 3.0 V		3.1		μA
R <sub>Htot</sub> <sup>(2)</sup>	Low drive resistive network overall value	5.28	6.6	7.92	МΩ
R <sub>L</sub> <sup>(2)</sup>	High drive resistive network total value	192	240	288	kΩ
V <sub>44</sub>	Segment/Common highest level voltage			$V_{LCD}$	V
V <sub>34</sub>	Segment/Common 3/4 level voltage		3/4 V <sub>LCD</sub>		
V <sub>23</sub>	Segment/Common 2/3 level voltage		2/3 V <sub>LCD</sub>		
V <sub>12</sub>	Segment/Common 1/2 level voltage		1/2 V <sub>LCD</sub>		V
V <sub>13</sub>	Segment/Common 1/3 level voltage		1/3 V <sub>LCD</sub>		] v
V <sub>14</sub>	Segment/Common 1/4 level voltage		1/4 V <sub>LCD</sub>		
V <sub>0</sub>	Segment/Common lowest level voltage	0			
ΔVxx <sup>(3)</sup>	Segment/Common level voltage error T <sub>A</sub> = -40 to 85 ° C			±50	mV

LCD enabled with 3 V internal step-up active, 1/8 duty, 1/4 bias, division ratio= 64, all pixels active, no LCD connected

<sup>2.</sup> Guaranteed by design, not tested in production.

<sup>3.</sup> Based on characterization, not tested in production.

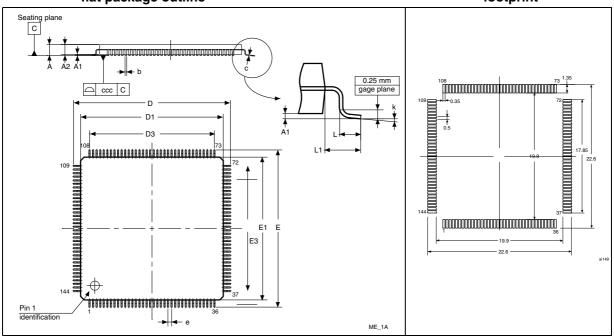
## 7 Package characteristics

## 7.1 Package mechanical data

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK® packages, depending on their level of environmental compliance. ECOPACK® specifications, grade definitions and product status are available at: <a href="https://www.st.com">www.st.com</a>. ECOPACK® is an ST trademark.

Figure 36. LQFP144, 20 x 20 mm, 144-pin low-profile quad flat package outline

Figure 37. Recommended footprint



1. Drawing is not to scale. Dimensions are in millimeters.

Table 68. LQFP144, 20 x 20 mm, 144-pin low-profile quad flat package mechanical data

Cumbal	millimeters			inches <sup>(1)</sup>			
Symbol	Min	Тур	Max	Min	Тур	Max	
Α			1.60			0.063	
A1	0.05		0.15	0.002		0.0059	
A2	1.35	1.40	1.45	0.0531	0.0551	0.0571	
b	0.17	0.22	0.27	0.0067	0.0087	0.0106	
С	0.09		0.20	0.0035		0.0079	
D	21.80	22.00	22.20	0.8583	0.8661	0.874	
D1	19.80	20.00	20.20	0.7795	0.7874	0.7953	
D3		17.50			0.689		
Е	21.80	22.00	22.20	0.8583	0.8661	0.874	
E1	19.80	20.00	20.20	0.7795	0.7874	0.7953	
E3		17.50			0.689		
е		0.50			0.0197		
L	0.45	0.60	0.75	0.0177	0.0236	0.0295	
L1		1.00			0.0394		
k	0°	3.5°	7°	0°	3.5°	7°	
ccc		0.08	•		0.0031	•	

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

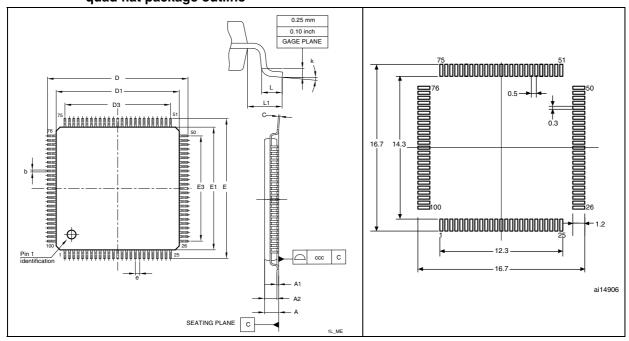


Figure 38. LQFP100, 14 x 14 mm, 100-pin low-profile Figure 39. Recommended footprint quad flat package outline

1. Drawing is not to scale. Dimensions are in millimeters.

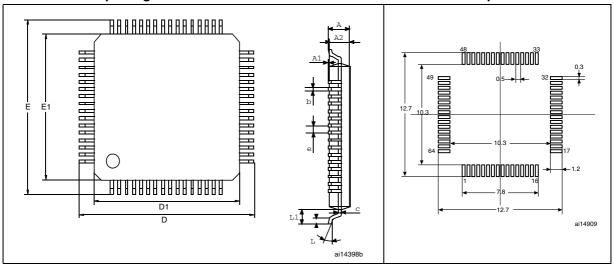
Table 69. LQPF100, 14 x 14 mm, 100-pin low-profile quad flat package mechanical data

Complete		millimeters		inches <sup>(1)</sup>			
Symbol	Min	Тур	Max	Min	Тур	Max	
Α			1.6			0.063	
A1	0.05		0.15	0.002		0.0059	
A2	1.35	1.4	1.45	0.0531	0.0551	0.0571	
b	0.17	0.22	0.27	0.0067	0.0087	0.0106	
С	0.09		0.2	0.0035		0.0079	
D	15.8	16	16.2	0.622	0.6299	0.6378	
D1	13.8	14	14.2	0.5433	0.5512	0.5591	
D3		12			0.4724		
E	15.8	16	16.2	0.622	0.6299	0.6378	
E1	13.8	14	14.2	0.5433	0.5512	0.5591	
E3		12			0.4724		
е		0.5			0.0197		
L	0.45	0.6	0.75	0.0177	0.0236	0.0295	
L1		1			0.0394		
k	0.0°	3.5°	7.0°	0.0°	3.5°	7.0°	
ccc		0.08			0.0031		

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 40. LQFP64, 10 x 10 mm, 64-pin low-profile quad Figure flat package outline

Figure 41. Recommended footprint



1. Drawing is not to scale. Dimensions are in millimeters.

Table 70. LQFP64, 10 x 10 mm, 64-pin low-profile quad flat package mechanical data

O	millimeters			inches <sup>(1)</sup>			
Symbol -	Min	Тур	Max	Min	Тур	Max	
Α			1.60			0.0630	
A1	0.05		0.15	0.0020		0.0059	
A2	1.35	1.40	1.45	0.0531	0.0551	0.0571	
b	0.17	0.22	0.27	0.0067	0.0087	0.0106	
С	0.09		0.20	0.0035		0.0079	
D		12.00			0.4724		
D1		10.00			0.3937		
E		12.00			0.4724		
E1		10.00			0.3937		
е		0.50			0.0197		
θ	0°	3.5°	7°	0°	3.5°	7°	
L	0.45	0.60	0.75	0.0177	0.0236	0.0295	
L1		1.00			0.0394		
N			Numbe	r of pins		•	
IN			(	64			

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

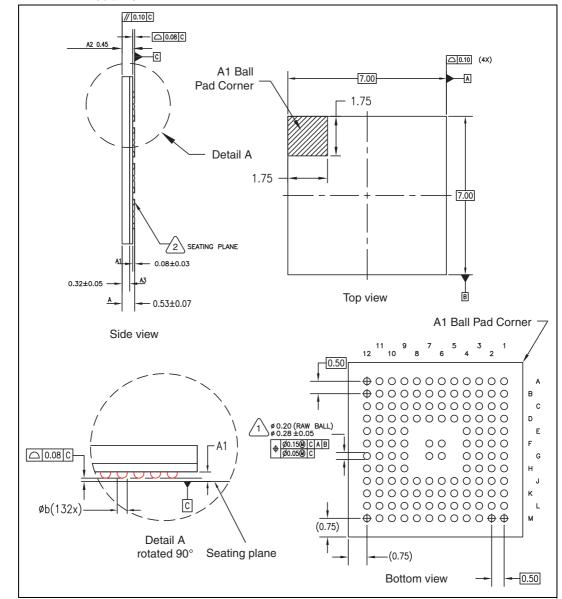


Figure 42. UFBGA132, 7 x 7 mm, 132-ball ultra thin, fine-pitch ball grid array package outline

- 1. Primary datum C and seating plane are defined by the spherical crowns of the solder balls.
- 2. Dimension is measured at the maximum solder ball diameter, parallel to primary datum C.

Table 71. UFBGA132 package mechanical data

Symbol		millimeters				
Symbol	Min	Тур	Max	Min	Тур	Max
Α	0.46	0.53	0.60	0.0181	0.0209	0.0236
A1	0.05	0.08	0.11	0.0020	0.0032	0.0043
A2	0.40	0.45	0.50	0.0157	0.0177	0.0197
b	0.17	0.28	0.33	0.0067	0.0110	0.0130

1. Values in inches are converted from mm and rounded to 4 decimal digits.

### 7.2 Thermal characteristics

The maximum chip-junction temperature,  $T_J$  max, in degrees Celsius, may be calculated using the following equation:

$$T_{J} \max = T_{A} \max + (P_{D} \max \times \Theta_{JA})$$

#### Where:

- ullet T<sub>A</sub> max is the maximum ambient temperature in  ${}^{\circ}$  C,
- Θ<sub>JA</sub> is the package junction-to-ambient thermal resistance, in °C/W,
- $P_D$  max is the sum of  $P_{INT}$  max and  $P_{I/O}$  max ( $P_D$  max =  $P_{INT}$  max +  $P_{I/O}$ max),
- P<sub>INT</sub> max is the product of I<sub>DD</sub> and V<sub>DD</sub>, expressed in Watts. This is the maximum chip internal power.

P<sub>I/O</sub> max represents the maximum power dissipation on output pins where:

$$P_{I/O} \max = \Sigma (V_{OL} \times I_{OL}) + \Sigma((V_{DD} - V_{OH}) \times I_{OH}),$$

taking into account the actual  $V_{OL}$  /  $I_{OL}$  and  $V_{OH}$  /  $I_{OH}$  of the I/Os at low and high level in the application.

Table 72. STM32L15xxD Thermal characteristics

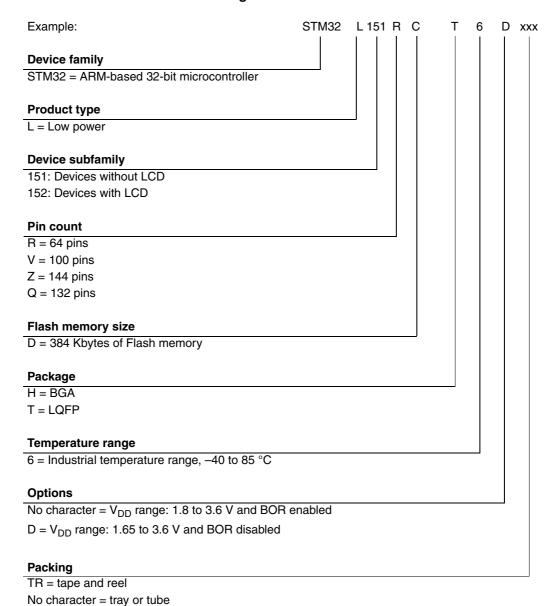
Symbol	Parameter	Value	Unit
	Thermal resistance junction-ambient LQFP144 - 20 x 20 mm / 0.5 mm pitch	40	
	Thermal resistance junction-ambient BGA132 - 7 x 7 mm	60	°C/W
$\Theta_{JA}$	Thermal resistance junction-ambient LQFP100 - 14 x 14 mm / 0.5 mm pitch	43	C/VV
	Thermal resistance junction-ambient LQFP64 - 10 x 10 mm / 0.5 mm pitch	46	

### 7.2.1 Reference document

JESD51-2 Integrated Circuits Thermal Test Method Environment Conditions - Natural Convection (Still Air). Available from www.jedec.org.

## 8 Ordering information scheme

Table 73. STM32L15xxD ordering information scheme



For a list of available options (speed, package, etc.) or for further information on any aspect of this device, please contact your nearest ST sales office.

# 9 Revision history

Table 74. Document revision history

Date	Revision	Changes
03-Oct-2011	1	Initial release.
03-Feb-2012	2	Status of the document changed (datasheet instead of preliminary data).  Updated low power features on page 1. Removed references to devices with 256 KB of Flash memory. GPIOF replaced with GIOPH. Added SDIO in Table 2: Ultralow power STM32L15xxD device features and peripheral counts on page 10 and in Table 6: Alternate function input/output on page 37 (FSMC/SDIO instead of FSMC). Table 2: Ultralow power STM32L15xxD device features and peripheral counts: replaced STM32L15xxD device features and peripheral counts: replaced STM32L15xxD block diagram: updated legend.  Modified Section 3.4: Clock management on page 17. Table 4: STM32L15xQD BGA132 7 x7 ballout: replaced STM32L15xQC/D with STM32L15xQD. Figure 3, Figure 4, Figure 5: updated titles. Table 5: STM32L15xD pin definitions: updated title, updated pins PF0, PF1, PH2, PF12, PF13, PF14, PF15, PG0, PG1, PG12, PG15, PD0, and PD1. Table 6: Alternate function input/output: Modified alternate function for PA13 and PA14; removed EVENT OUT for PH2. Figure 6: Memory map: removed the text "APB memory space". Modified Figure 9: Power supply scheme on page 48. Modified Table 11: Functionalities depending on the operating power supply range on page 51. Table 15: Current consumption in Run mode, code with data processing running from RAM: added footnote 3. Table 16: Current consumption in Sleep mode: updated condition for flase; added footnote 3. Table 20: Typical and maximum current consumptions in Standby mode: modified max values. Table 56: USB DC electrical characteristics: removed two footnotes. Modified Table 32: Flash memory characteristics on page 74. Table 72: STM32L15xxD Thermal characteristics updated "TBDs" with values. Modified tables in Section 6.3.4: Supply current characteristics on page 54.

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