

# EE433 INTRODUCTION TO DIGITAL SIGNAL PROCESSING

## Computer Homework Part 2

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### Question 5

Starting from your solution in question 1 of Computer HW Part #1 and using the frequency transformations of low-pass filters described in ed. 3 of the textbook (pp.526-532, section 7.4), design a high-pass filter with cut-off frequency  $\omega_p = 0.7\pi$ :

- (a) Give the system function,  $H(z)$ , of the designed filter.
- (b) Plot the magnitude of frequency response, indicating the corner frequencies corresponding to passband and stopband amplitude design targets given in (1).

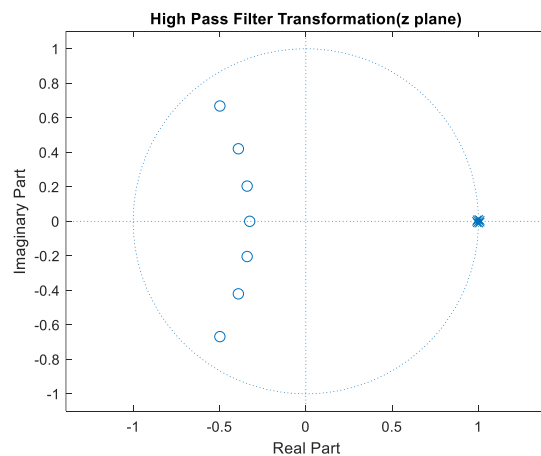
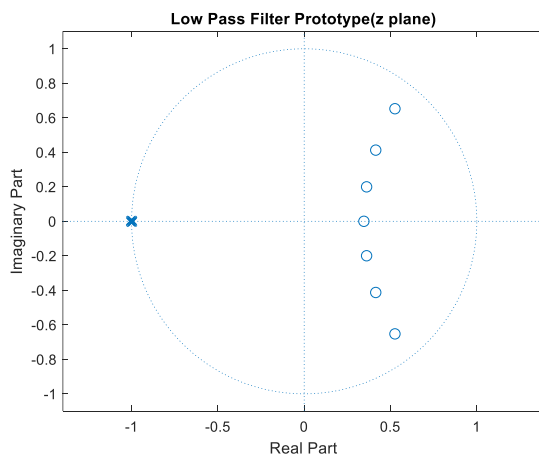
In this part, the code below generates low pass and high pass filter transfer functions. We use this formula writer instead of  $Z^{-1}$  while switch from low-pass filter to high-pass filter and we use this formula for alpha.

$$Z^{-1} = \frac{z^{-1} - \alpha}{1 - \alpha z^{-1}}$$

$$\alpha = \frac{\sin\left(\frac{\theta_p - \omega_p}{2}\right)}{\sin\left(\frac{\theta_p + \omega_p}{2}\right)}$$

Where  $\omega_p$  equal to desired cutoff frequency.

### Simulation



Poles:

$-0.4979 + 0.6685i$

$-0.4979 - 0.6685i$

$-0.3907 + 0.4204i$

$-0.3907 - 0.4204i$

$-0.3399 + 0.2041i$

$-0.3399 - 0.2041i$

$-0.3252 + 0.0000i$

Zeros:

$1.0100 + 0.0049i$

$1.0100 - 0.0049i$

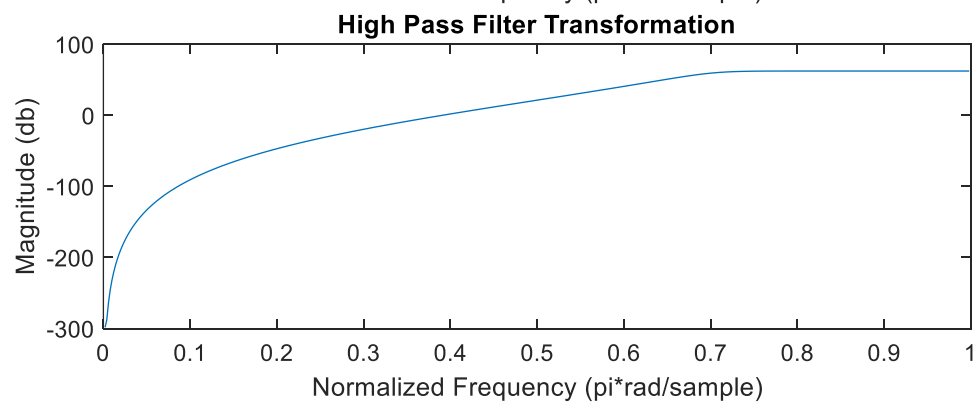
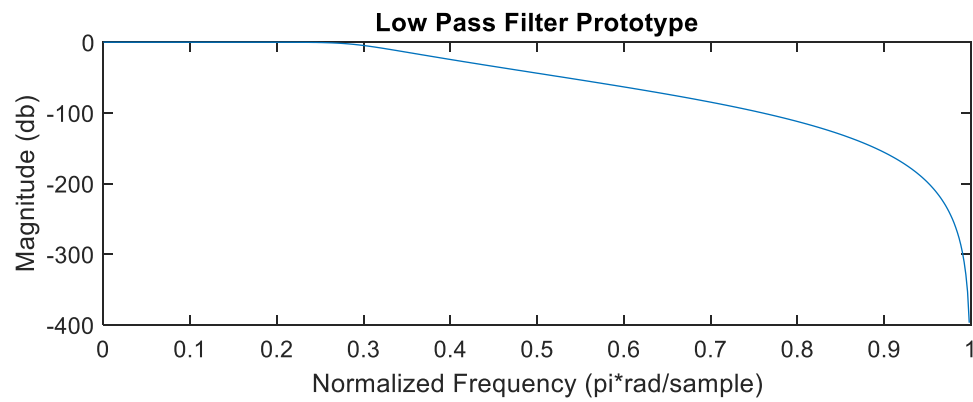
$1.0024 + 0.0109i$

$1.0024 - 0.0109i$

$0.9931 + 0.0086i$

$0.9931 - 0.0086i$

$0.9890 + 0.0000i$



## Source Matlab Code

Design a high-pass filter

```
g_cut=0.905;
w_c=g_cut/pi;
order=7;
[y,x] = butter(order,w_c);
[z,p,k] = butter(order,w_c);
[h_1, w_1] = freqz(y,x);

w_d = 0.7*pi;
alpha = -cos((g_cut + w_d)/2) / cos((g_cut - w_d)/2);

syms Z(z)
Z(z)=-(z^-1 + alpha)/(1 + alpha*z^-1);
zero1 =1-(Z(z)*-1);
pole1 =1-(Z(z)*1.051) + (Z(z)*Z(z)*0.702);
pole2 =1-(Z(z)*0.829) + (Z(z)*Z(z)*0.342);
pole3 =1-(Z(z)*0.723) + (Z(z)*Z(z)*0.171);
pole4 =1-(Z(z)*0.346);

h = zero1^7 / (pole1*pole2*pole3*pole4);

[numerator_eq,denominator_eq] = numden(h);
b_n = sym2poly(numerator_eq);
a_n = sym2poly(denominator_eq);
[hhp, whp] = freqz(b_n, a_n);

figure
zplane(x,y);title('Low Pass Filter Prototype(z plane)')
figure
zplane(a_n,b_n);title('High Pass Filter Transformation(z plane)');

figure
subplot(211); plot(w_1/pi, 20*log10(abs(h_1)));
title('Low Pass Filter Prototype');
xlabel('Normalized Frequency (pi*rad/sample)'); ylabel('Magnitude (db)');
subplot(212); plot(whp/pi, 20*log10(abs(hhp)));
title('High Pass Filter Transformation');
xlabel('Normalized Frequency (pi*rad/sample)'); ylabel('Magnitude (db)');
```

## Question 6

Starting from your solution in question 1 of Computer HW Part #1 and using the frequency transformations of low-pass filters described in ed. 3 of the textbook (pp.526-532, section 7.4), design a band-pass filter with cut-off frequencies  $\omega_{p1} = 0.3\pi$  and  $\omega_{p2} = 0.5\pi$ :

- (a) Give the system function,  $H(z)$ , of the designed filter.
- (b) Plot the magnitude of frequency response, indicating the corner frequencies corresponding to passband and stopband amplitude design targets given in (1).

In this part, the code below generates low pass and high pass filter transfer functions. We use this formula writer instead of  $Z^{-1}$  while switch from low-pass filter to high-pass filter and we use this formula for alpha and k.

$$Z^{-1} = -\frac{z^{-2} - \frac{2\alpha k}{k+1}z^{-1} + \frac{k-1}{k+1}}{\frac{k-1}{k+1}z^{-2} - \frac{2\alpha k}{k+1}z^{-1} + 1} \quad \alpha = \frac{\cos\left(\frac{\omega_{p2} + \omega_{p1}}{2}\right)}{\cos\left(\frac{\omega_{p2} - \omega_{p1}}{2}\right)}$$

$$k = \cot\left(\frac{\omega_{p2} - \omega_{p1}}{2}\right) \tan\left(\frac{\theta_p}{2}\right)$$

Where  $\omega_{p1}$  equal to desired lower cutoff frequency and  $\omega_{p2}$  desired upper cutoff frequency.

| Poles             | Zeros             |
|-------------------|-------------------|
| -0.0442 + 0.9187i | -1.0042 + 0.0020i |
| -0.0442 - 0.9187i | -1.0042 + 0.0020i |
| 0.5802 + 0.7350i  | -1.0010 - 0.0046i |
| 0.5802 - 0.7350i  | -1.0010 - 0.0046i |
| 0.0145 + 0.7835i  | -0.9971 - 0.0037i |
| 0.0145 - 0.7835i  | -0.9971 - 0.0037i |
| 0.4817 + 0.6640i  | -0.9953 + 0.0000i |
| 0.4817 - 0.6640i  | 1.0046 - 0.0022i  |
| 0.1100 + 0.6829i  | 1.0046 - 0.0022i  |
| 0.1100 - 0.6829i  | 1.0011 - 0.0050i  |
| 0.3664 + 0.6239i  | 1.0011 - 0.0050i  |

$0.3664 - 0.6239i$

$0.2353 + 0.6267i$

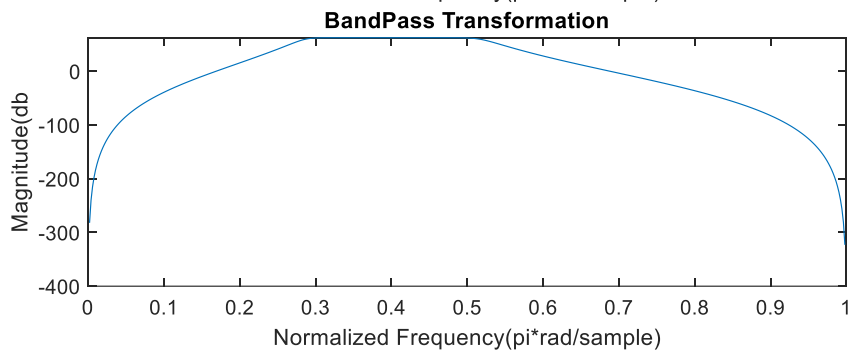
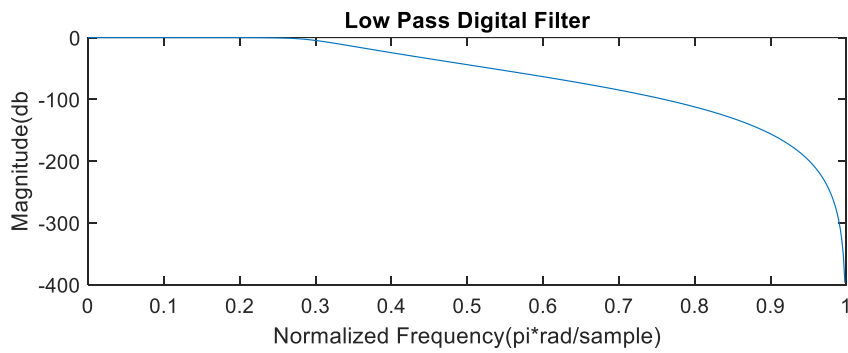
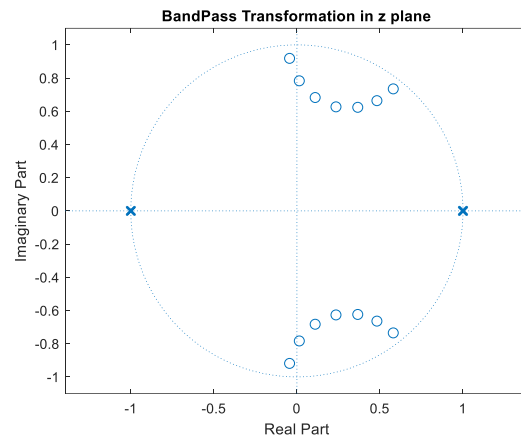
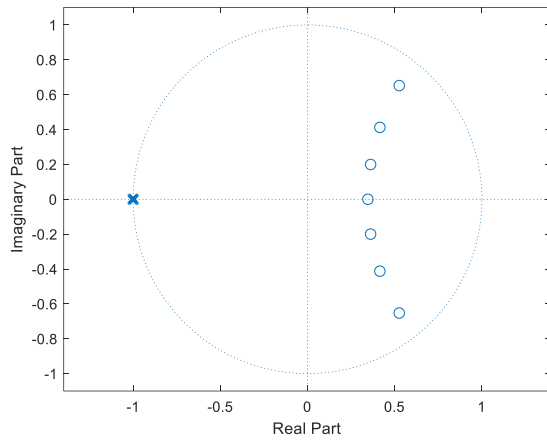
$0.2353 - 0.6267i$

$0.9968 + 0.0040i$

$0.9968 - 0.0040i$

$0.9949 + 0.0000i$

## Simulation



## Source Matlab Code

### Design a band-pass filter

```
wp = 0.25*pi; ws = 0.40*pi;
gammaCut=0.9047; w_c=gammaCut/pi; order=7;
[y,x] = butter(order,w_c);
[z,p,k] = butter(order,w_c);
[h1,w1] = freqz(y,x);

wd_1=0.3*pi; wd_2=0.5*pi;

alpha=cos((wd_2 + wd_1)/2)/cos((wd_2 - wd_1)/2);
d_1=cot((wd_2-wd_1)/2)*tan(wp/2);

syms Z(z)
Z(z) = -(z^-2 - (2*alpha*d_1*z^-1)/(d_1+1) + (d_1-1)/(d_1+1))/((d_1-1)/(d_1+1)*z^-2 - (2*alpha*d_1*z^-1)/(d_1+1) + 1);

% zeros and pole pairs
zero_1=1-(Z(z)*-1);
pole_1=1-(Z(z)*1.051) + (Z(z)*Z(z)*0.702);
pole_2=1-(Z(z)*0.829) + (Z(z)*Z(z)*0.342);
pole_3=1-(Z(z)*0.723) + (Z(z)*Z(z)*0.171);
pole_4=1-(Z(z)*0.346);
res=pole_1*pole_2*pole_3*pole_4;
h=zero_1^7/res;

[numerator_equation, denominator_equation] = numden(h);
y_new = sym2poly(numerator_equation);x_new = sym2poly(denominator_equation);
[hhp, whp]=freqz(y_new, x_new);

figure
zplane(x,y);
figure
zplane(x_new,y_new);title('BandPass Transformation in z plane');
figure
subplot(2,1,1); plot(w1/pi, 20*log10(abs(h1)));
title('Low Pass Digital Filter');
xlabel('Normalized Frequency(pi*rad/sample)'); ylabel('Magnitude(db)');
subplot(2,1,2); plot(whp/pi, 20*log10(abs(hhp)));
title('BandPass Transformation');
xlabel('Normalized Frequency(pi*rad/sample)'); ylabel('Magnitude(db)');
```

## Question 7

Consider the DTFT shown below. That ideal filter will be approximated using a Kaiser window. The parameters of the design are  $A_1/A_2 \geq 15$  dB and  $A_2/A_3 \geq 25$  dB. Additionally, each transition band is not wider than  $0.1\pi$ , that is the separation of successive critical frequencies obey the rule:  $|\omega_i - \omega_{i+1}| \leq 0.1\pi$

(a) Give the formula of  $h[n]$ , the impulse response of the DT-filter.

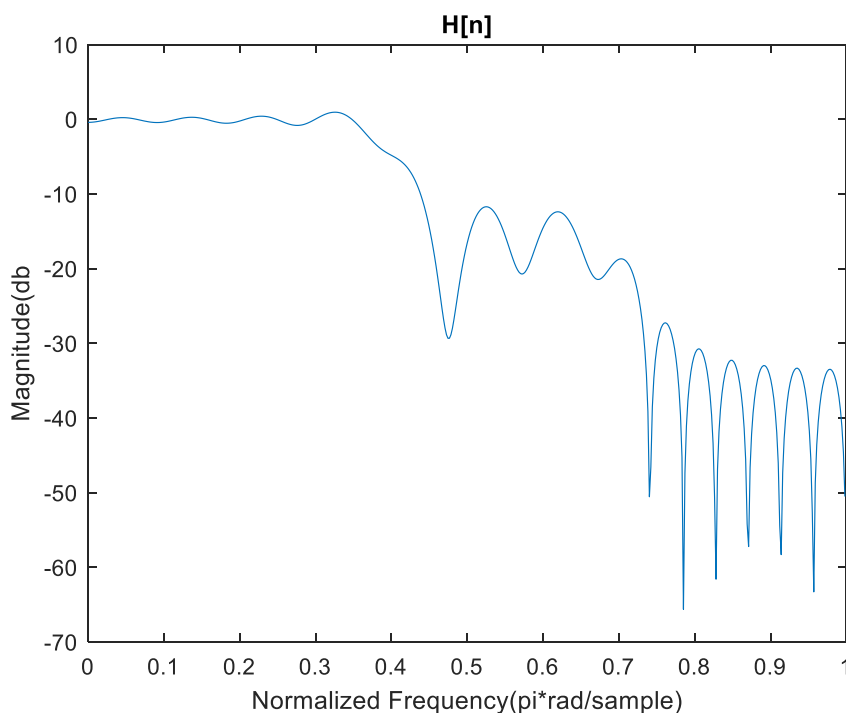
We have to multiply ideal low pass filter equation with Kaiser window.

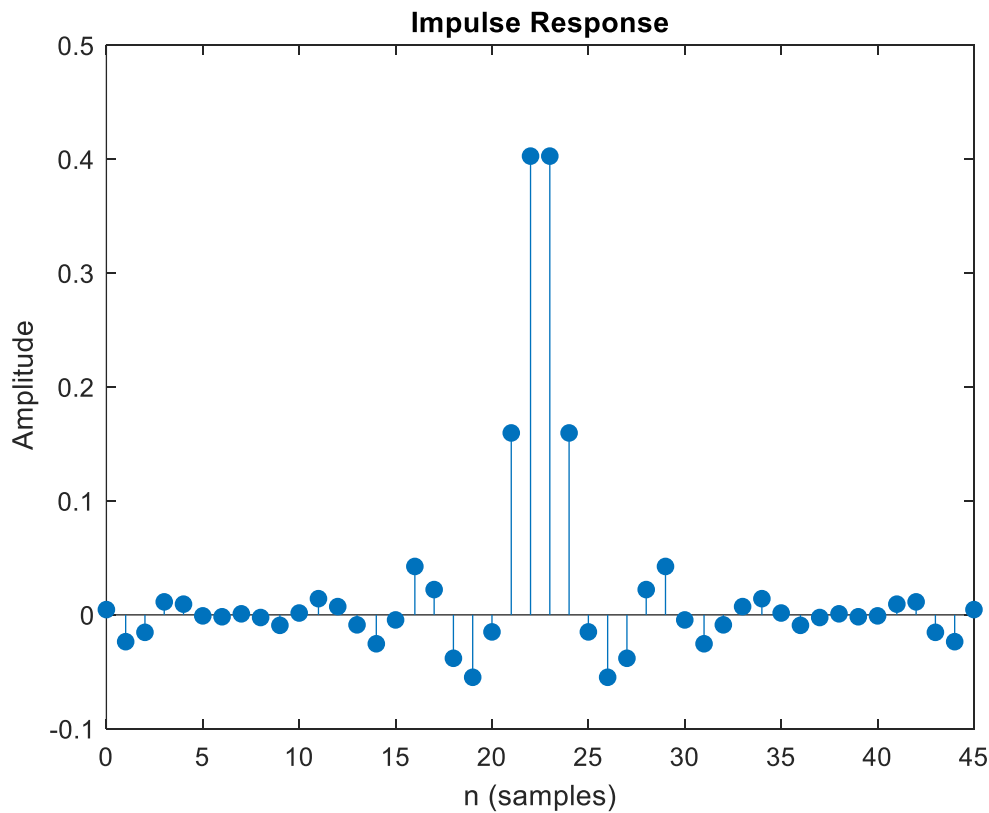
$$h[n] = \sum_{k=2}^{N_{mb}} (A_k - A_{k+1}) \frac{\sin \omega_c (n-a)}{\pi (n-a)} \cdot \frac{I_0 \left[ \beta \left( 1 - \left[ \frac{n-a}{\alpha} \right]^2 \right)^{1/2} \right]}{I_0(\beta)}$$

(b) Plot  $h[n]$ .

(c) Plot the magnitude of the frequency response of the designed DT-filter. Check the gains at the critical frequencies.

## Simulation





## Source Matlab Code

```
parameter_1=1; parameter_2=0.178; parameter_3=(10^-2);
v_1=0.4*pi; v_2=0.7*pi;
theta=3.395; Nmb=2; order=45;
alpha=order/2;
G=[parameter_1 parameter_2];
w=[v_1 v_2];
h1=zeros(order+1,1);
i=1
while i<order+1;

    h1(i)=(G(1)-G(2))*(sin(w(1)*((i-1)-alpha))/(pi*((i-1)-
alpha)))*(besselj(0,theta*(1-((i-1-
alpha)/alpha)^2)^(1/2)))/besselj(0,theta);
    i=i+1;
end

h2=zeros(order,1);
j=1;
for j=1:order+1;

    h2(j)=(G(2)-parameter_3)*(sin(w(2)*((j-1)-alpha))/(pi*((j-1)-
alpha)))*(besselj(0,theta*(1-((j-1-
alpha)/alpha)^2)^(1/2)))/besselj(0,theta);
```



```

    j=j+1;
end

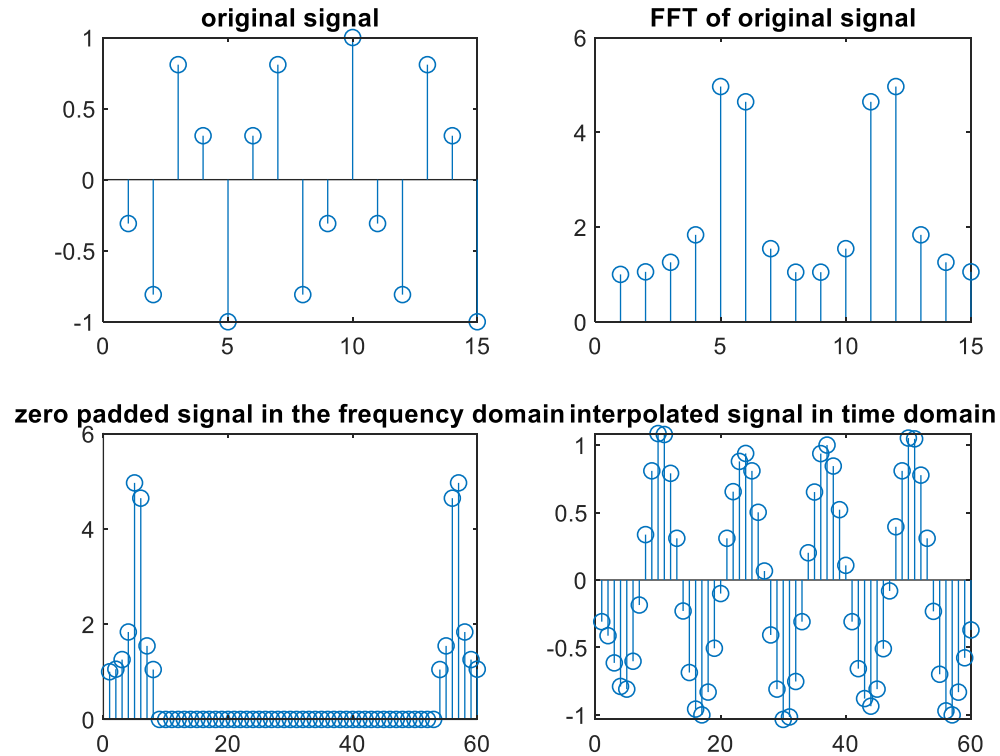
htotal=h1+h2;
[H W]=freqz(htotal);
figure
plot(W/pi,20*log10(abs(H)));
title('H[n]');
xlabel('Normalized Frequency(pi*rad/sample)'); ylabel('Magnitude(db)');

figure
impz(htotal)

```

## Question 8

A signal is defined as  $x[n] = \cos(2 * \pi * 0.3 * n)$  for  $n = 1, 2, \dots, 15$ . Interpolate in time domain by a factor of 4 using zero padding in frequency domain. Plot all of the signals including every intermediate step in this procedure.



## Source Matlab Code

```
n=1:15;
y_t=cos(2*pi*0.3*n);
subplot(2,2,1);
stem(y_t);
title('original signal');
k=4;
pad_len = length(y_t)*(k - 1);
z_len = ceil((length(y_t)+1)/2);
z = (fft(y_t));
subplot(2,2,2);
stem(abs(z));
title('FFT of original signal');
z_p1= [z(1:z_len) zeros(1, pad_len) z(z_len+1:end)];
if ~mod(length(y_t),2)
z_p1(z_len) = z_p1(z_len)/2; z_p1(z_len+pad_len) = z_p1(z_len);
end
subplot(2,2,3);
stem(abs(z_p1));
title('zero padded signal in the frequency domain');
yp = real(ifft(z_p1))*k;
subplot(2,2,4);
stem(yp);
title('interpolated signal in time domain');
```