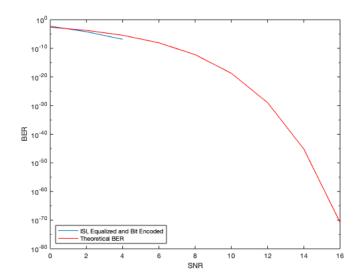
12/21/2019 CommPart2

```
numIter = 5000; % The number of iterations of the simulation. We used 4000 because our code
% is too strong and we need more bits to get observable error
nSym = 1000;
                % The number of symbols per packet
SNR_Vec = 0:2:16;
lenSNR = length(SNR_Vec);
M = 16:
               % The M-ary number, 2 corresponds to binary modulation
%chan = 1:
                    % No channel
chan = [1 .2 .4]; % Somewhat invertible channel impulse response, Moderate ISI
%chan = [0.227 0.460 0.688 0.460 0.227];
                                             % Not so invertible, severe ISI
\ensuremath{\mathtt{\textit{\$}}} Create equalizer object with signal ocnstellation based on \ensuremath{\mathtt{M}}
\mbox{\ensuremath{\$}} We used a Decision Feedback Equalizer instead of linear because it
% produced much better BER
sigConst = qammod(0:M-1,M);
eqlms = dfe(5,3, lms(0.005));
eqlms.SigConst = sigConst;
eqlms.ResetBeforeFiltering = 0;
\ensuremath{\mathtt{\$}} Create a vector to store the BER computed during each iteration
berVec = zeros(numIter, lenSNR);
% Run the simulation numIter amount of times
for i = 1:numIter
    bits = randi([0 1], 1, nSym*log2(M));
                                                 % Generate random bits
    % New bits must be generated at every iteration
     To reduce BER, we use convolutional coding. Convolutional coding was
    % chosen to keep simulation time manageable. BCH would take far too
    % long. We managed to achieve 10^(-7) BER with 2 to 3 convolutional
    % coding. 1 to 2 convolutional coding could not.
    trellis = poly2trellis([5 4],[23 35 0; 0 5 13]); % Constraint Lengths 5 and 4 for two input channels, two sets of Generator Matrices for two input
    convbits = convenc(bits, trellis); % apply convolutional coding
    convbitLen = length(convbits); % length after bit encoding
    trainLen = convbitLen/5; % Use the first 20% of bits to train
    % Here we convert the bits into symbols ranging from 0 to M-1
    msg = zeros(1, convbitLen/log2(M));
    msgindex = 1;
    for bitindex = 1:log2(M):(convbitLen)
        msg(msgindex) = bi2de(convbits(bitindex:bitindex+log2(M)-1), 'left-msb');
        msgindex = msgindex+1;
    end
    %for j = 1:lenSNR % one iteration of the simulation at each SNR Value
    for j = 1:3
        tx = qammod(msg,M); % BPSK modulate the signal
        \ensuremath{\mathtt{\$}} Choose whether or not to filter the signal thorugh a channel with
        if isequal(chan,1)
            txChan = tx;
            txChan = filter(chan,1,tx); % Apply the channel.
        txNoisy = awgn(txChan,SNR_Vec(j)+ 10*log10(log2(M))); % Add AWGN
        eqSig = equalize(eqlms,txNoisy,tx(1:trainLen)); %equalize the signal with trainLen training bits
        rxEq = qamdemod(eqSig,M); % Demodulate
        % Convert symbols back to bits
        rxMSG = zeros(1, convbitLen);
        rxMSGindex = 1;
        for rxEqindex = 1:1:length(rxEq)
            rxMSG(rxMSGindex:rxMSGindex+log2(M)-1) = de2bi(rxEq(rxEqindex), log2(M), 2, 'left-msb');
            rxMSGindex = rxMSGindex + log2(M);
        % decode our convolutional encoding
        deconvbits = vitdec(rxMSG, trellis, 20, 'trunc', 'hard');
        % Compute and store the BER for this iteration
        [~, berVec(i,j)] = biterr(bits(trainLen+1:end), deconvbits(trainLen+1:end)); % We're interested in the BER, which is the 2nd output of BITERR
        \ensuremath{\mathtt{\textit{\$}}} Here we only discard the first trainLen bits since they are used
        % for training
    end % End SNR iteration
end
         % End numIter iteration
% Compute and plot the mean BER
ber = mean(berVec,1);
figure;
semilogy(SNR_Vec, ber)
% Compute the theoretical BER for this scenario
```

12/21/2019 CommPart2

```
berTheory = berawgn(SNR_Vec + 10*log10(log2(M)), 'psk',log2(M), 'nondiff');
hold on
semilogy(SNR_Vec,berTheory,'r');
xlabel("SNR");
ylabel("BER");
legend('ISI, Equalized and Bit Encoded', 'Theoretical BER', 'Location', 'southwest')
```



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