2017/2018 SEMESTER TWO EXAMINATION

Diploma in Electrical and Electronic Engineering 3rd Year Full Time

DIGITAL SIGNAL PROCESSING

<u>Time Allowed</u>: 2 Hours

<u>Instructions to Candidates</u>

- 1. The examination rules as set out on the last page of the answer booklet are to be complied with.
- 2. This paper consists of **TWO** sections:

Section A - 6 Short Questions, 10 marks each.
Section B - 2 Long Questions, 20 marks each.

- 3. ALL questions are COMPULSORY.
- 4. **ALL** questions are to be answered in the answer booklet.
- 5. This paper consists of 6 pages, including 2 pages of mathematical formulae.

2017/2018_S2 Page 1 of 6

SECTION A - SHORT QUESTIONS [10 marks each]

A1. Using z-transform, determine the impulse response of the digital system shown in Figure A1. Comment on the stability of the system. (10 marks)

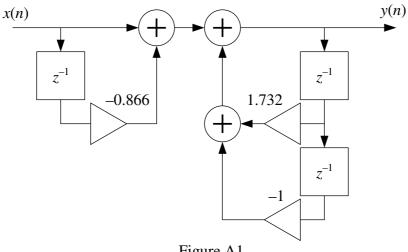


Figure A1

A2 Given $y(n) = \{-1, 5, -4, 2, 1\}$ and impulse response $h(n) = \{1, -1, 1\}$, find the z-transform of y(n) and h(n), hence determine the input x(n) by using the long-division method.

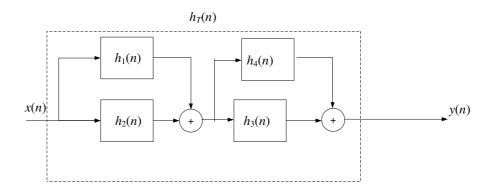
(10 marks)

Find the z-transform of $x(n)=20\sin(0.25\pi n)u(n)$ and **A3** $y(n) = e^{0.2n}\cos(0.25\pi n)u(n)$.

(10 marks)

A4 Evaluate the N = 4-point DFT for X(0) and X(2) if $x(n) = \{0, 2, 0, 2\}$. (10 marks)

2017/2018_S2 Page 2 of 6 A5 The block diagram of a digital system is given as:



- a) Find the overall impulse response of the system, $h_T(n)$ in terms of $h_1(n)$, $h_2(n)$, $h_3(n)$ and $h_4(n)$. Find the z-transform of $h_T(n)$, $H_T(z)$. (6 marks)
- b) If $h_1(n) = h_2(n) = h_3(n) = h_4(n) = \{1,1\}$ respectively, find $h_T(n)$. (4 marks)

A6 The system function of a digital system is given as:

$$X(z) = \frac{z}{3z^2 - 4z + 1}$$

Using partial fraction, find x(n).

(10 marks)

SECTION B - LONG QUESTIONS [20 marks each]

B1.

A certain digital FIR low pass filter was designed with a Barlett window function using the windowing technique. It was noted that the filter coefficients, h(20) = h(30), and h(20) = 0. The sampling frequency used was 10 kHz.

Determine

- (a) the peak approximation error of the filter in dB. (3 marks)
- (b) the number of tap coefficients that the filter had. [Hint: h(n) = h(M n)] (4 marks)
- (c) the width of the transition band in Hz. (4 marks)
- (d) the pass band and stop band frequency ranges in Hz. (9 marks)

B2.

Consider a FIR filter with difference equation given by:

$$y(n) = x(n) - x(n-2) - 2x(n-4)$$

- (a) Compute and sketch the magnitude response of the filter at $\omega = 0, \frac{\pi}{4}, \frac{\pi}{2}, \frac{3\pi}{4}, \pi$. What type of filter is this?
- (b) Find the output of the filter, y(n), when the input $x(n) = \left[\cos\left(\frac{\pi}{2}n\right)\right]u(n)$. Is the steady state output (i.e. when n is large) consistent with the results in part (a) above? Why?

-End of Paper-

2017/2018_S2 Page 4 of 6

Appendix

The *z*-transform is defined as $X(z) = \sum_{n=-\infty}^{\infty} x[n]z^{-n}$

	<i>n</i> =-∞
Sequence	Transform
$\delta[n]$	1
u[n]	1
	$1-z^{-1}$
$\delta[n-m]$	<i>z</i> - ^m
$a^nu[n]$	1
	$1-az^{-1}$
$na^nu[n]$	az^{-1}
	$\overline{(1-az^{-1})^2}$
$[\cos \omega_0 n] u[n]$	$1-[\cos\omega_0]z^{-1}$
	$1 - [2\cos\omega_0]z^{-1} + z^{-2}$
$[\sin \omega_0 n] u[n]$	$[\sin \omega_0]z^{-1}$
	$\frac{1 - [2\cos\omega_0]z^{-1} + z^{-2}}{1 - [2\cos\omega_0]z^{-1} + z^{-2}}$
$[r^n \cos \omega_0 n] u[n]$	$1 - [r\cos\omega_0]z^{-1}$
	$\frac{1 - [2r\cos\omega_0]z^{-1} + r^2z^{-2}}{1 - [2r\cos\omega_0]z^{-1} + r^2z^{-2}}$
$[r^n \sin \omega_0 n] u[n]$	$[r\sin\omega_0]z^{-1}$
Ů	$1 - [2r\cos\omega_0]z^{-1} + r^2z^{-2}$

Some <i>z</i> -transform properties:				
Sequence	Transform			
x[n]	X(z)			
$x_1[n]$	$X_1(z)$			
$x_2[n]$	$X_2(z)$			
$ax_1[n] + bx_2[n]$	$aX_1(z) + bX_2(z)$			
x[n-m]	$z^{-m}X(z)$			

Some trigonometric identities:

$$e^{\pm j\theta} = \cos\theta \pm j \sin\theta$$

$$\cos^2\theta + \sin^2\theta = 1$$

$$\sin(A \pm B) = \sin A \cos B \pm \cos A \sin B$$

$$\cos(A \pm B) = \cos A \cos B \mu \sin A \sin B$$

$$\tan(A \pm B) = \frac{\tan A \pm \tan B}{1 \mu \tan A \tan B}$$

Complex number theory:

$$z = a + jb = r \angle \theta = re^{j\theta}$$
$$r = \sqrt{a^2 + b^2}$$
$$\theta = \tan^{-1} \frac{b}{a}$$

Cariac.

$$\sin(x) = x - \frac{x^3}{3!} + \frac{x^5}{5!} - \frac{x^7}{7!} + \dots$$

Discrete Fourier Transform:

$$X(k) = \sum_{n=0}^{N-1} x(n)e^{-j\frac{2\pi}{N}kn}$$
$$x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k)e^{j\frac{2\pi}{N}kn}$$

Quadratic equation solution:

If
$$ax^2 + bx + c = 0$$

then $x = \frac{-b \pm \sqrt{b^2 - 4ac}}{2a}$

The characteristics of the different windowing functions:

Window Type	Peak approximation	Transition
	Error	Band
	$20 \log_{10} \delta dB$	Δω
Rectangular:	-21	4π
$w(n) = \int 1 0 \le n \le M$		$\overline{M+1}$
$w(n) = \begin{cases} 1 & 0 \le n \le M \\ 0 & \text{otherwise} \end{cases}$		
Bartlett:	-25	8π
Bartlett: $w[n] = \begin{cases} \frac{2n}{M} & 0 \le n \le \frac{M}{2} \\ 2 - \frac{2n}{M} & \frac{M}{2} \le n \le M \\ 0 & \text{otherwise} \end{cases}$		\overline{M}
$w[n] = \begin{cases} 2 - \frac{2n}{M} & \frac{M}{2} \le n \le M \end{cases}$		
0 otherwise		
Hanning:	-44	8π
$w(n) = \begin{cases} 0.5 - 0.5 \cos\left(\frac{2\pi n}{M}\right) & 0 \le n \le M \\ 0 & \text{otherwise} \end{cases}$		M
0 otherwise		
Hamming:	-53	8π
$w(n) = \begin{cases} 0.54 - 0.46 \cos\left(\frac{2\pi n}{M}\right) & 0 \le n \le M \\ 0 & \text{otherwise} \end{cases}$		\overline{M}
0 otherwise		
Blackman:	-74	12π
$w(n) = \begin{cases} 0.42 - 0.5\cos\left(\frac{2\pi n}{M}\right) + 0.08\cos\left(\frac{4\pi n}{M}\right) & 0 \le n \le M \end{cases}$		\overline{M}
0 otherwise		

The impulse response of an ideal low pass filter is:
$$h_d(n) = \frac{\sin\left(\omega_c\left[n - \frac{M}{2}\right]\right)}{\pi\left(n - \frac{M}{2}\right)}$$

2017/2018_S2 Page 6 of 6