

# **Chapter 7**

# **Analog to Digital conversion**

Part 1 of 4



### Introduction



- Digital communication is the dominant form of communication.
- A lot of information in modern communications is in digital form.

e.g. binary data

 However, analog signals like audio and video signals are not compatible with digital processing and signal transmission of digital communication systems

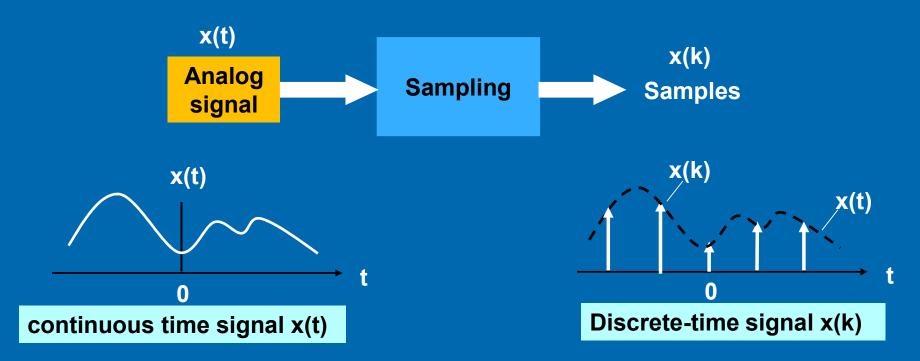
Audio and video signals are two of the most common information signals

 Analog-to-digital conversion (ADC) is required for transmitting analog signals using digital communication systems.





- The first step towards the conversion from analog to digital communications. and the bridge between analog and digital signals.
- Converts an analog signal into a corresponding train of samples that are spaced uniformly in time.





#### Reasons for sampling an analog signal:

- 1. To convert an analog signal into a digital signal that is compatible with digital transmission.
  - To be compatible with digital transmission.
  - To allow an analog signal to be digitally processed.
  - To allow TDM (time division multiplexing) which is the simultaneous transmission of several signals over the same channel.
- 2. Certain signal processing devices (for example high-power microwave tubes; laser) can operate better on a pulse basis.
- 3. Reduction in power needed to transmit a signal as with sampling, signal is transmitted in burst.

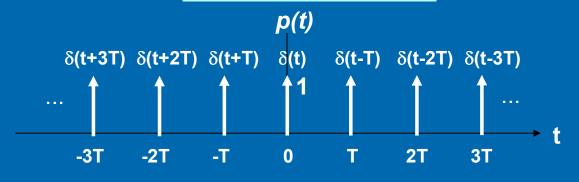




#### Ideal Sampling / impulse train sampling

Ideal or impulse-train sampling of is carried as follows:

#### an unit impulse train



#### sampling frequency

 $f_0$  = fundamental frequency of p(t)

$$f_s = f_0 = 1/T$$

$$x(t) \longrightarrow x_p(t) \longrightarrow x(t) \times p(t)$$
analog signal sampled signal

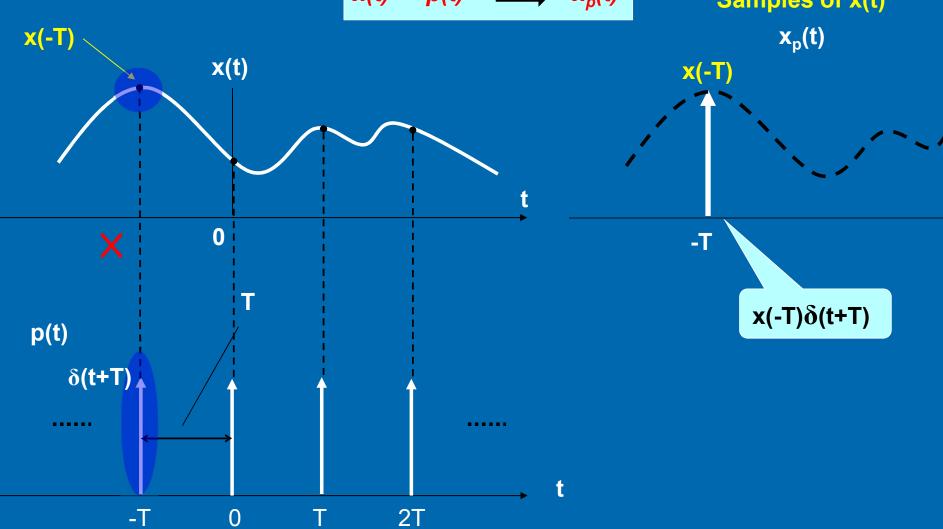
Ideal sampling takes a sample of a signal without any modification. Can be done only with an unit impulse train.



x(t)

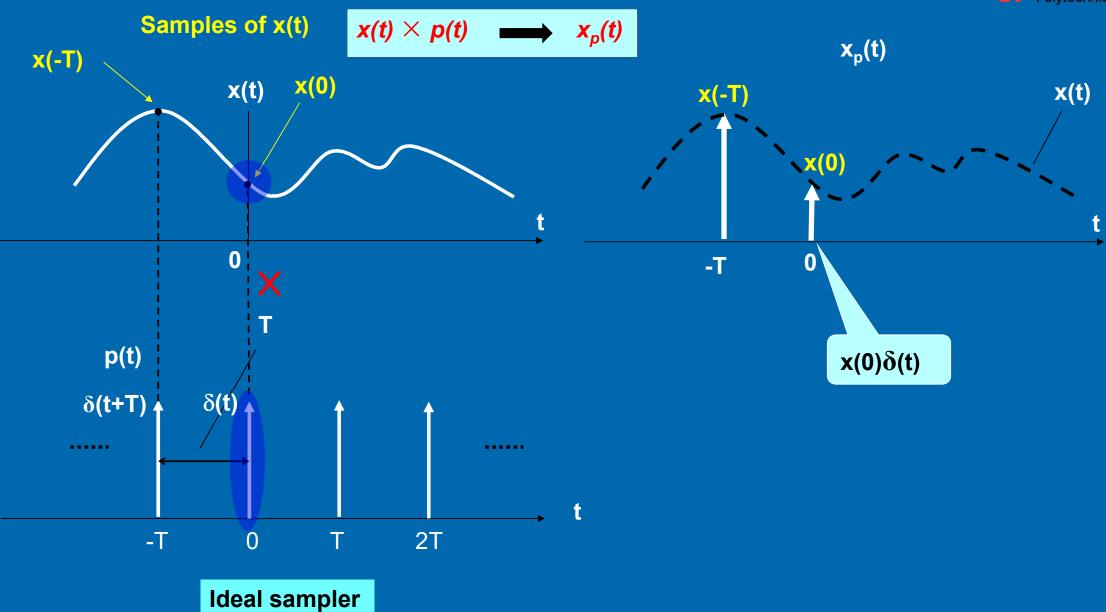


### Samples of x(t)

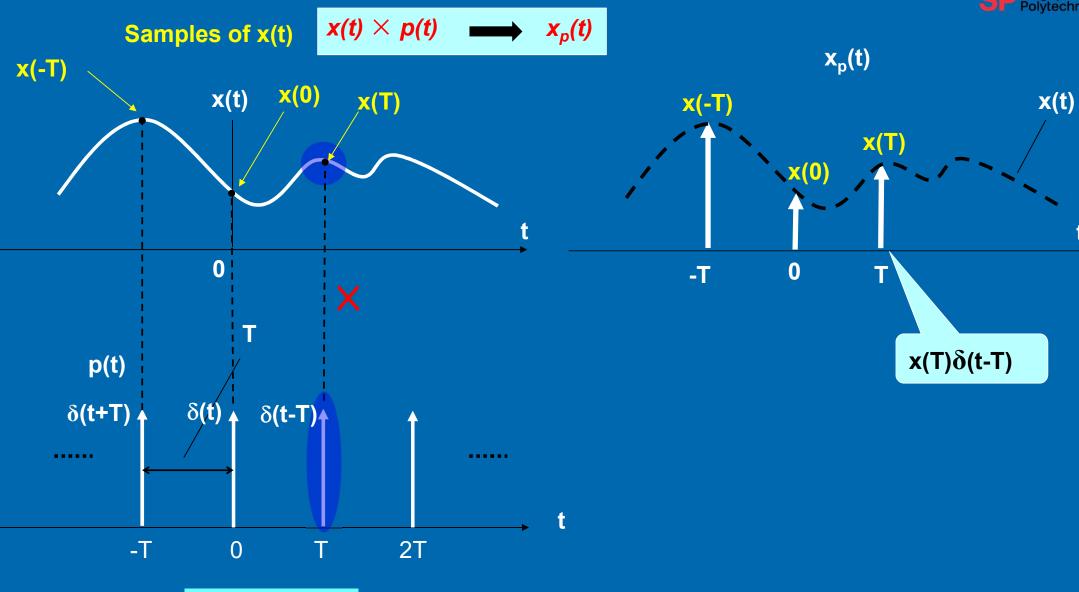


**Ideal sampler** 



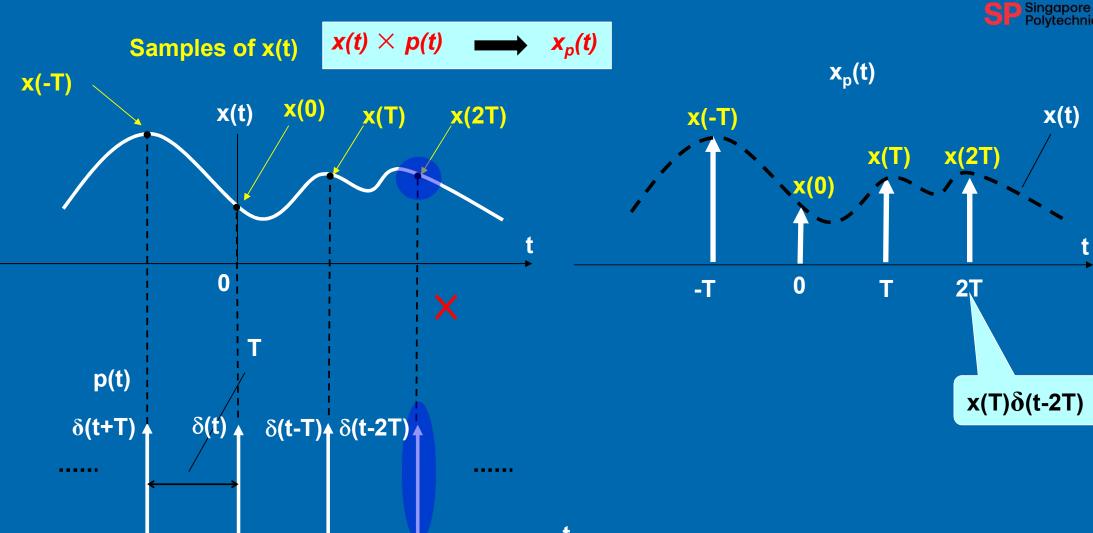






**Ideal sampler** 





**Ideal sampler** 

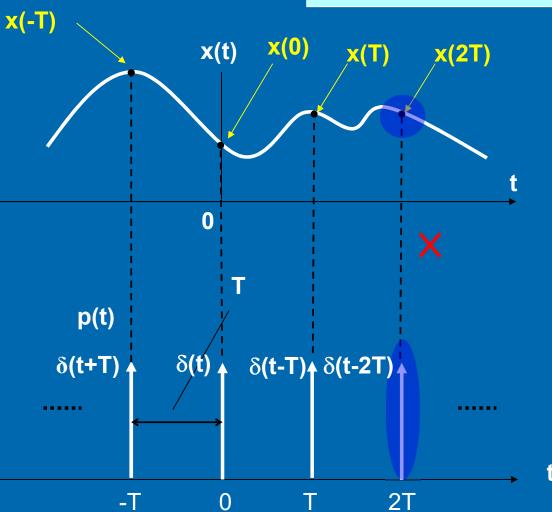
0

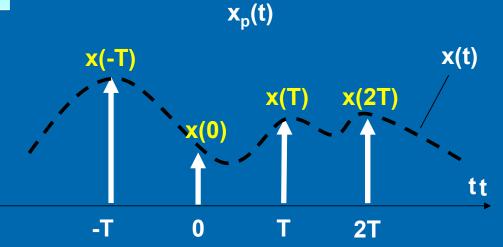
-T

**2T** 









An impulse train whose amplitudes are enveloped by the original signal.

Ideal sampler



#### **Ideal Sampling**

In time domain,

$$x_p(t) = x(t) \times p(t)$$
 where  $p(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT)$ 

$$x_p(t) = \sum_{n=-\infty}^{\infty} x(nT)\delta(t-nT)$$

 $x_p(t)$  is an impulse train with the strength of the impulses equal to the values of x(t) at the sampling instants, at intervals of T.



#### **Ideal Sampling**

In frequency domain (from frequency convolution theorem)

$$X_p(f) = X(f) * P(f)$$

where \* denotes convolution and  $P(f) = \frac{1}{T} \sum_{k=-\infty}^{\infty} \delta(f - kf_s)$ 

$$X(f) * \delta(f - kf_s) = X(f - kf_s)$$
 Chapter 2 eq (2.32)

Hence,

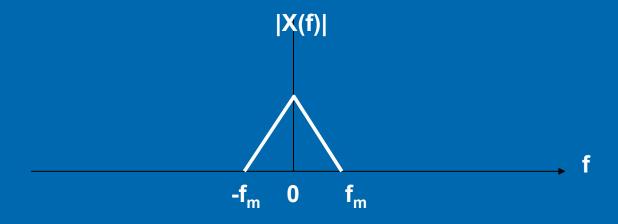
$$X_p(f) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X(f - kf_s)$$

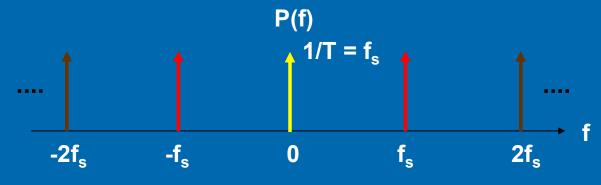
 $X_p(f)$  is a periodic function of f consisting of sum of shifted replicas of X(f), scaled by 1/T.



### **Ideal Sampling**

Let X(f) be the amplitude spectrum of x(t) as shown below with highest frequency component,  $f_m$  and P(f) be the spectrum of p(t).



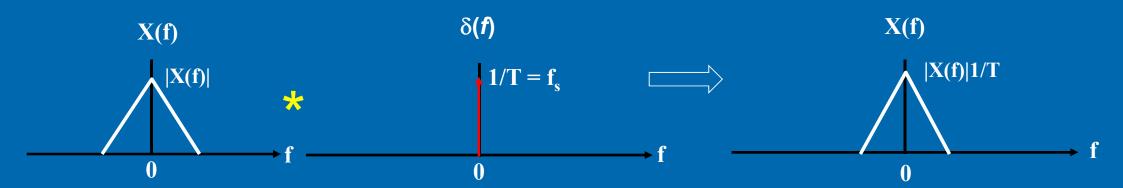


# **Recall (Chapter 2)**



Convolution of any function X(f) with a unit impulse function  $\delta(f)$  gives the function X(f) itself.

$$X(f) * \delta(f) = X(f)$$

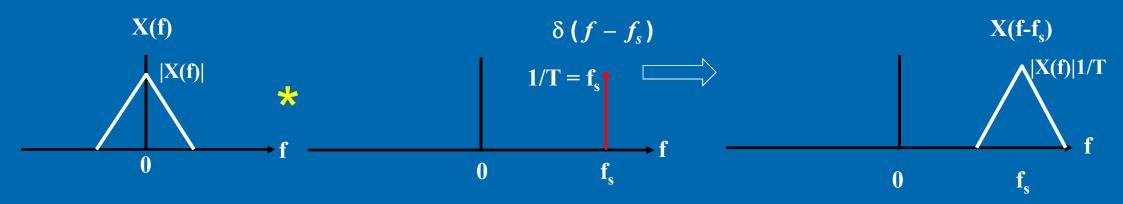


## **Recall (Chapter 2)**



Convolution of any function X(f) with a unit impulse function  $\delta(f-f_s)$  gives the frequency shifted version:  $X(f-f_s)$ 

$$X(f) * \delta(f - f_s) = X(f - f_s)$$

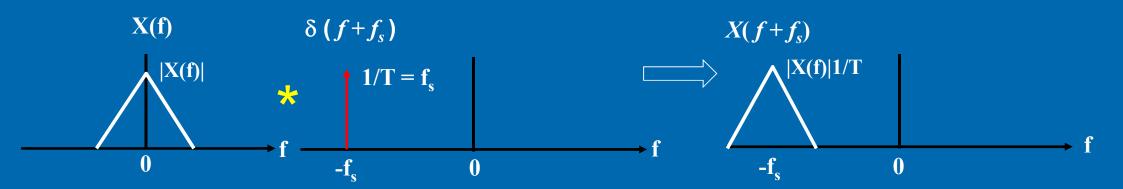


## **Recall (Chapter 2)**

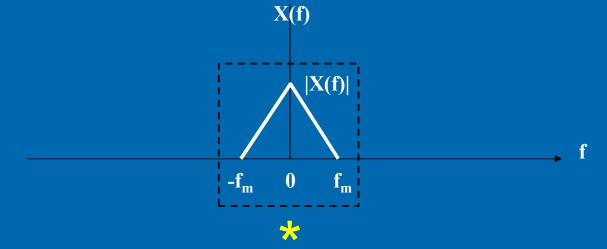


Convolution of any function X(f) with a unit impulse function  $\delta(f+f_s)$  gives the frequency shifted version:  $X(f+f_s)$ 

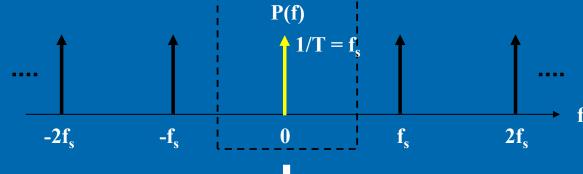
$$X(f) * \delta(f+f_s) = X(f+f_s)$$



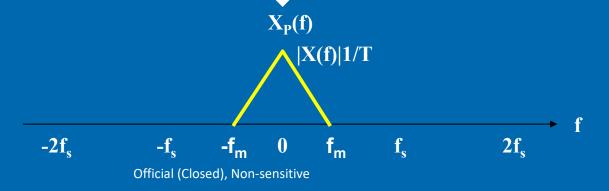




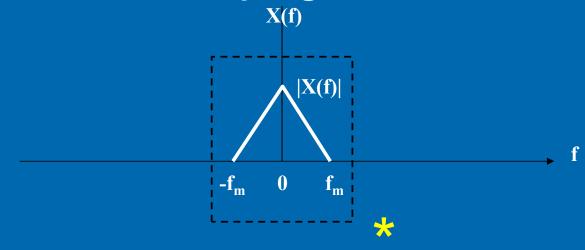




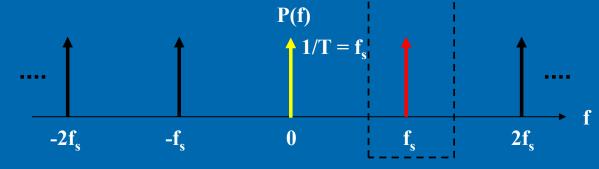
$$X_p(f) = X(f) * P(f)$$



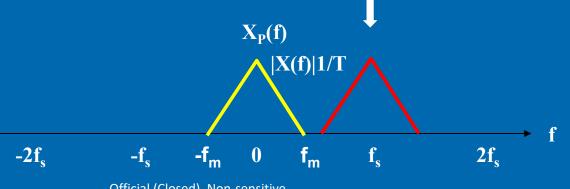




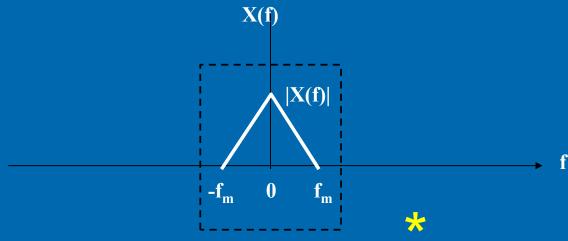
$$P(f) = \frac{1}{T} \sum_{k=-\infty}^{\infty} \delta(f - kf_s)$$



$$X_p(f) = X(f) * P(f)$$

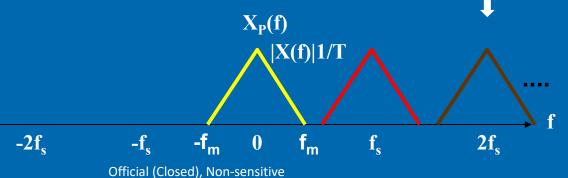




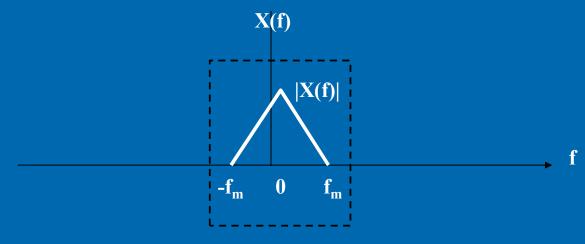


$$P(f) = \frac{1}{T} \sum_{k=-\infty}^{\infty} \delta(f - kf_s) \qquad \qquad \uparrow \qquad \qquad \downarrow \qquad \qquad \downarrow$$

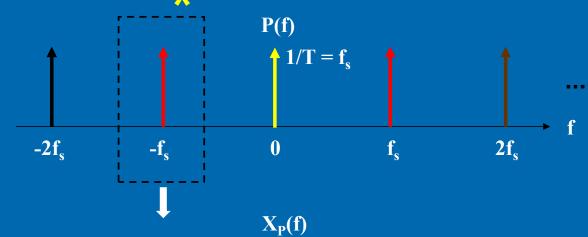
$$X_p(f) = X(f) * P(f) \dots$$



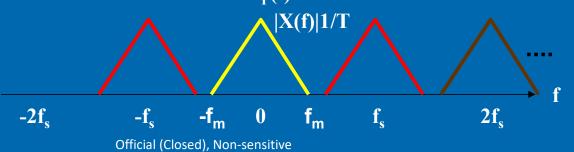




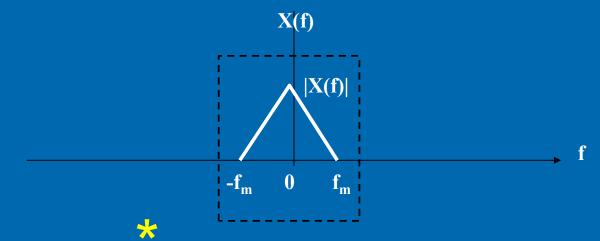
$$P(f) = \frac{1}{T} \sum_{k=-\infty}^{\infty} \delta(f - kf_s)$$



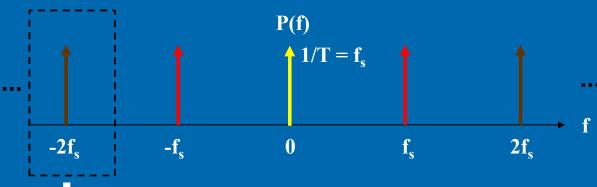
$$X_p(f) = X(f) * P(f)$$



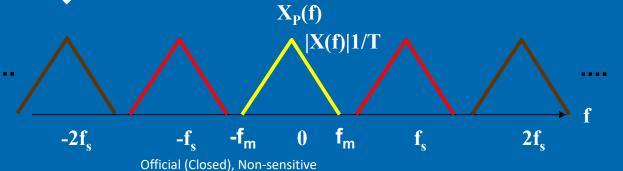








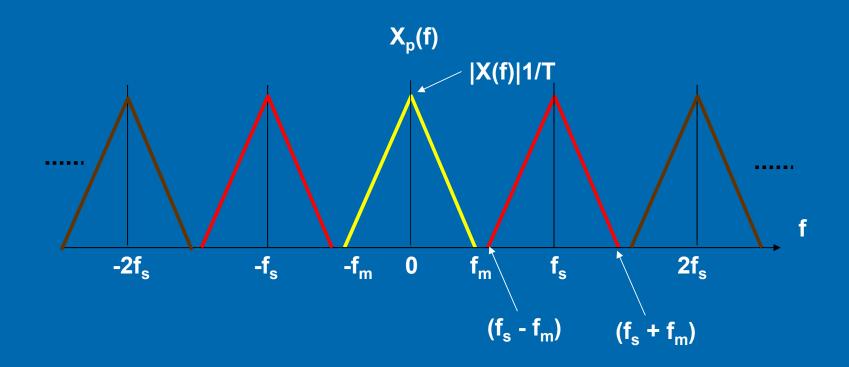
$$X_p(f) = X(f) * P(f)$$





### **Ideal Sampling**

### Amplitude spectrum of the sampled signal





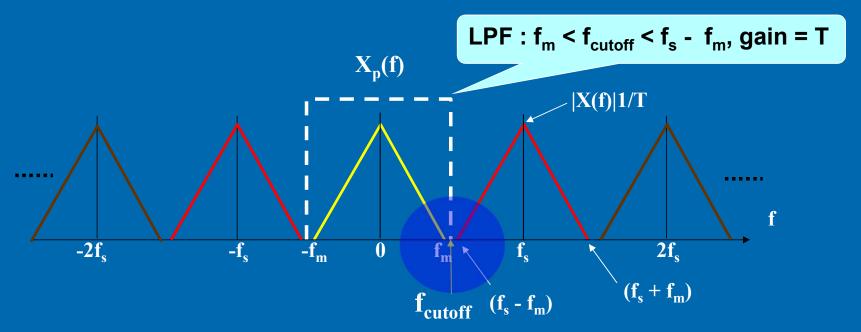
### Ideal Sampling

If 
$$f_s > 2 f_m$$
,  $f_s > f_m + f_m \implies f_s - f_m > f_m$ 

When  $f_s > 2f_m$ , i.e.  $(f_s - f_m) > f_m$  there is no overlap and x(t) can be recovered exactly, by processing  $x_p(t)$  through an ideal LPF of gain T and  $f_m < f_c < (f_s - f_m)$ .

Observation 1

Exact recovery of original signal from sampled signal is possible

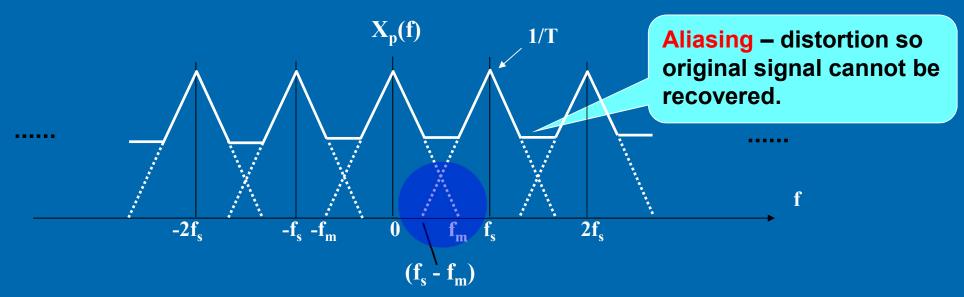




#### **Ideal Sampling**

When  $f_s < 2f_m$ , i.e.  $(f_s - f_m) < f_m$  there is overlap and x(t) cannot be recovered from the sampled signal. Such overlap of replicas of X(t) is known as aliasing.

### **Observation 2: Aliasing**



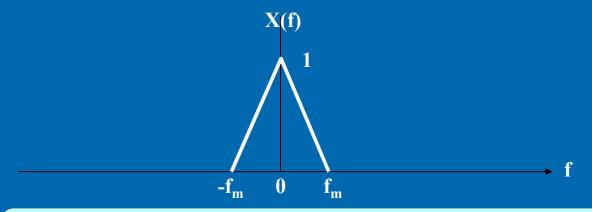




#### **Ideal Sampling**

#### **Uniform Sampling Theorem I**

If x(t) is bandlimited to  $f_m$ , it must be sampled at  $f_s \ge 2f_m$ , in order for it to be recovered completely from its samples.



#### **Bandlimited baseband signal**

i.e. signals having components from dc to about 1 MHz, e.g.

It is common in communication system design, to perform an anti-aliasing filtering before any sampling operation to avoid the occurrence of aliasing.

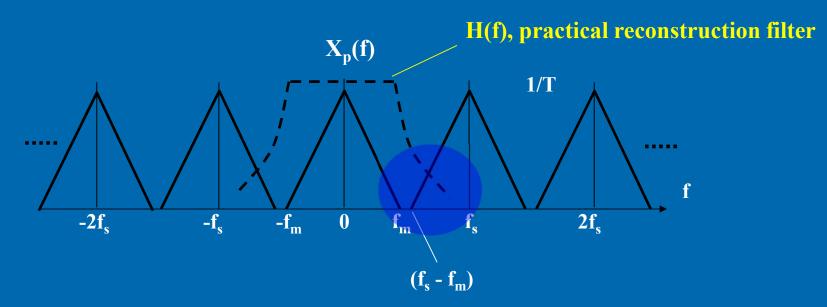


- Nyquist frequency and Nyquist rate
  - The minimum sampling frequency,  $f_s = 2f_m$  is referred to as Nyquist Frequency or Nyquist Rate and
  - Minimum sampling interval,  $T_s = 1/f_s = 1/(2f_m)$  as Nyquist Interval.





#### The effects of using Non-ideal reconstruction filter

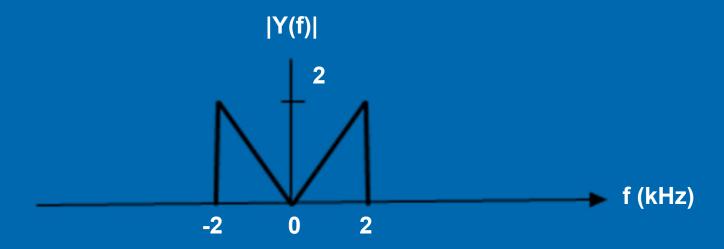


- In practice, difficult to build filters with a very sharp roll-off at f<sub>c</sub>. Some spurious frequency components are allowed by the reconstruction filter to reach the output.
- To minimise these spurious frequencies, a higher sampling frequency should be used.



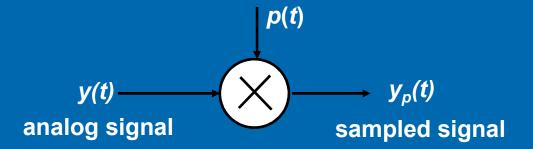
#### **Worked Example 7.1**

A signal y(t) with an amplitude spectrum shown below is sampled at 5 kHz by an ideal unit impulse train, p(t). Sketch the amplitude spectrum of the sampled signal.



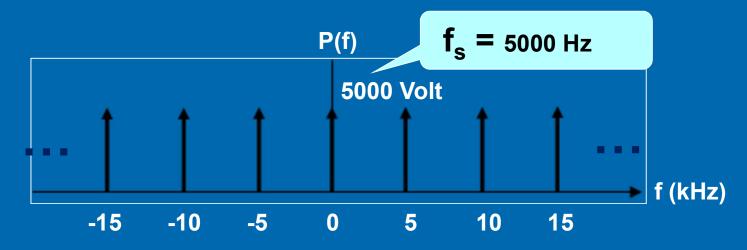


Sampled signal =  $y(t) \times p(t)$ 



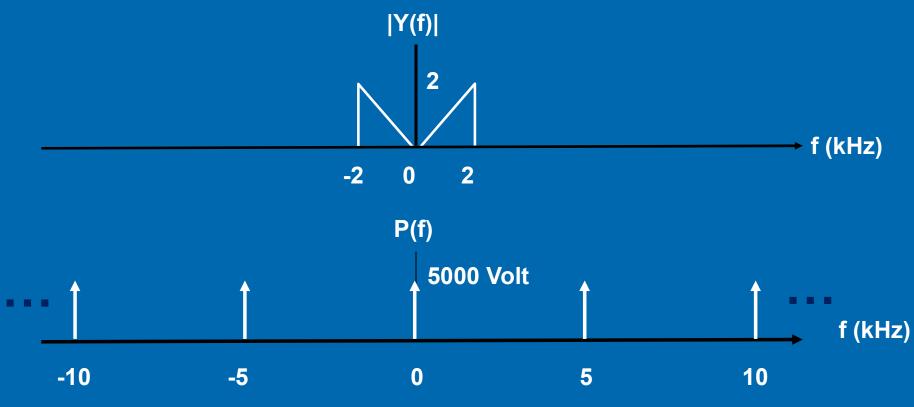
The amplitude spectrum of sampled signal = Y(f)\*P(f)

#### The amplitude spectrum of P(f) is



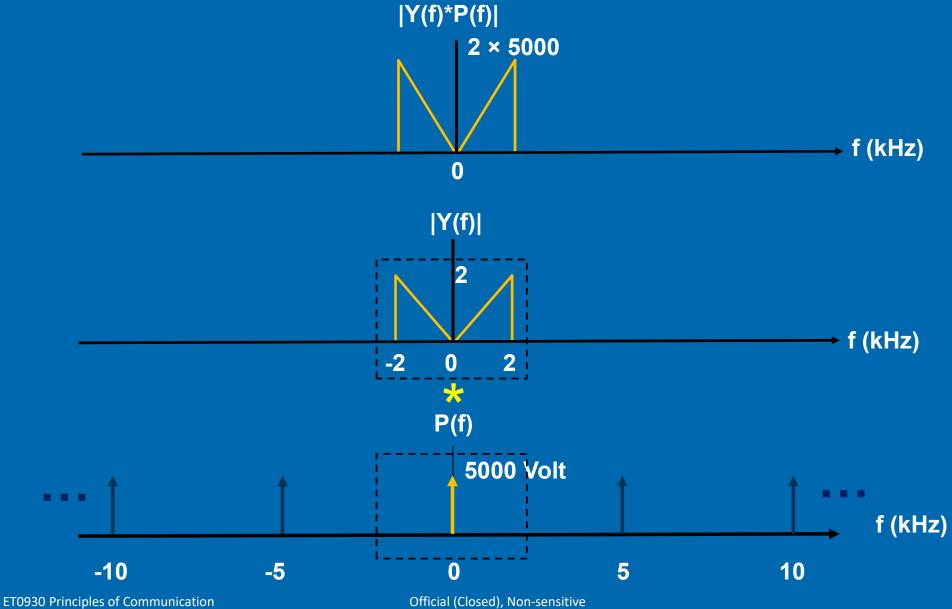


### **Solution** The amplitude spectrum of sampled signal = Y(f)\*P(f)



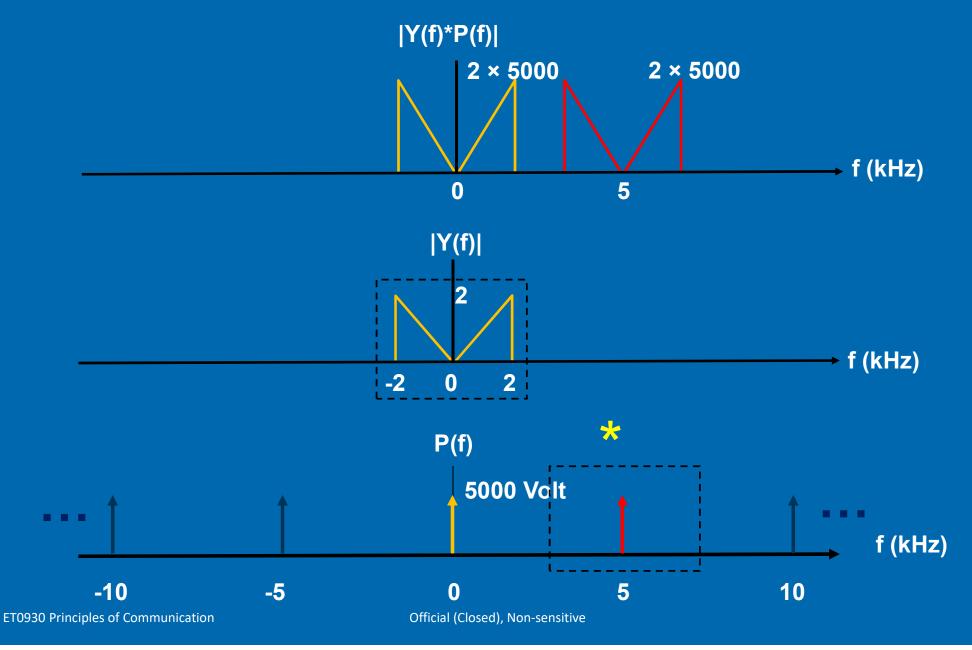
### The amplitude spectrum of sampled signal = Y(f)\*P(f)





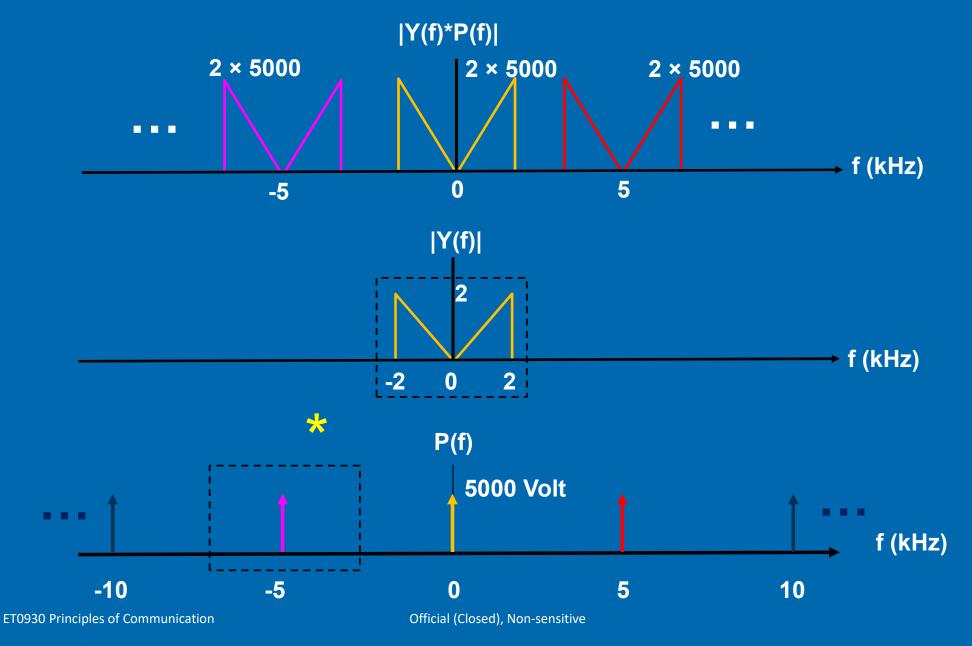
### The amplitude spectrum of sampled signal = Y(f)\*P(f)





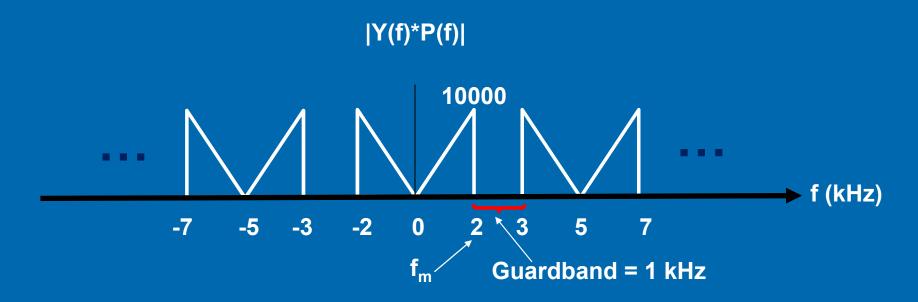
### The amplitude spectrum of sampled signal = Y(f)\*P(f)







Hence, amplitude spectrum of sampled signal is given by



The sample frequency  $f_s = 2 f_m + guardband = 5 kHz$ 



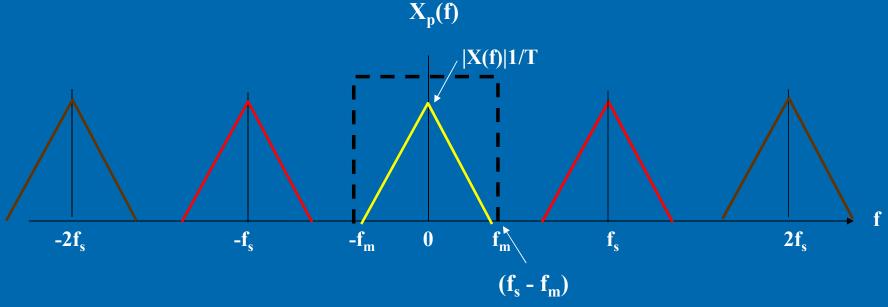
#### The effects of oversampling and undersampling

**Oversampling** 

$$f_s >> f_m + f_m \implies f_s - f_m >> f_m$$

Recovery of original signal is easy but at expense of higher transmission bandwidth.

• For higher sampling freq,  $f_s >> 2 f_m$ , it is wasteful to sample the signal at so <u>high rate</u> as it increases the bandwidth without any gain (most samples are redundant).





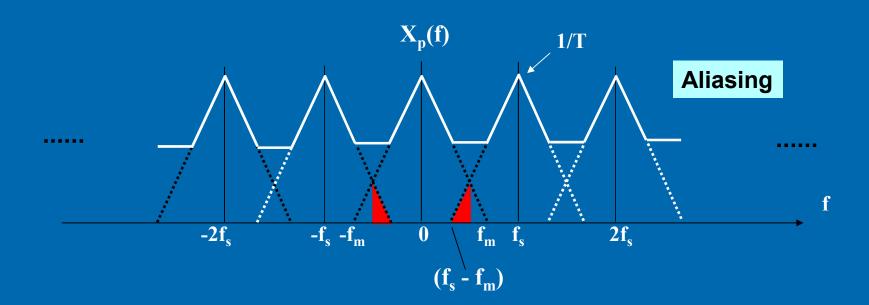
#### The effects of oversampling and undersampling

**Undersampling** 

**Aliasing** 

$$f_{s} < f_{m} + f_{m} \implies f_{s} - f_{m} < f_{m}$$

Recovery of original signal is not possible.



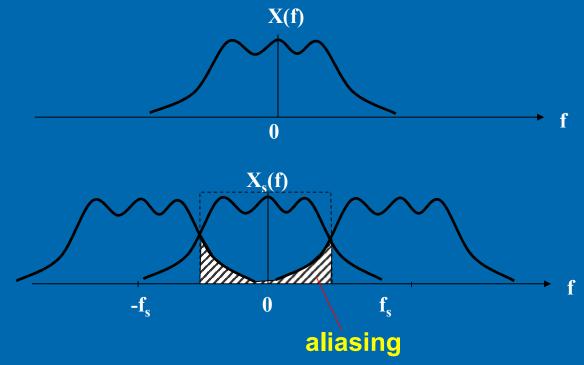


#### The effects of oversampling and undersampling

**Undersampling** 

**Aliasing** 

 Besides undersampling, aliasing can also occur when the signal to be sampled is not bandlimited.



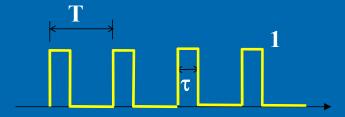
Hence, anti-aliasing filter is required to bandlimit the signal before sampling.





#### **Practical sampling**

- Practical sampling :
  - a) Sampling waveform consists of pulses of finite amplitude and duration.



b) Practical reconstruction filters do not possess ideal characteristics.

Signal recovery is possible using higher sampling frequency (guardband ↑)

- There are two forms of practical sampling:
- 1. Natural sampling
- 2. Flat-top sampling

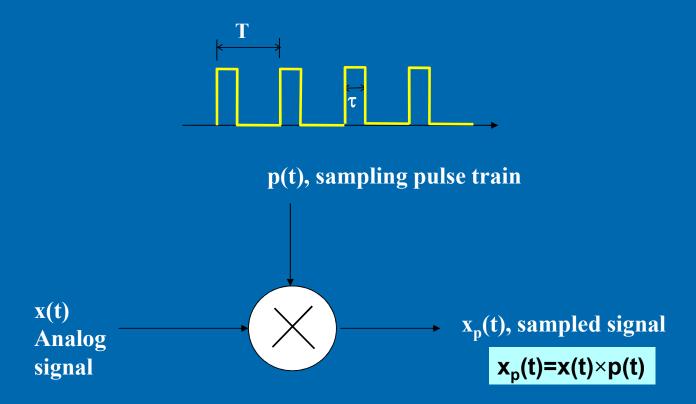




### **Practical sampling**

### **Natural Sampling**

In natural sampling the signal to be sampled is multiplied by a sampling pulse train.

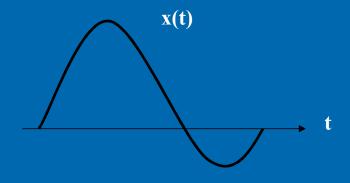




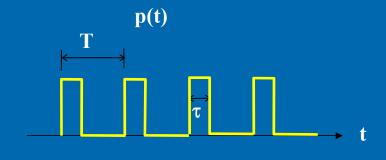
### **Practical sampling**

### **Natural Sampling**

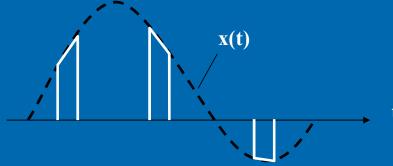
a) message signal



b) sampling train



c) Naturally sampled signal  $x_p(t)=x(t)\times p(t)$ 





#### **Practical sampling**

### **Natural Sampling**

Observation on natural sampling based on the above :

- 1. The sampled signal is a pulse train whose amplitude follows the signal over duration  $\tau$ .
- 2.  $f_s \ge 2f_m$  and hence, no aliasing occurs.
- 3. x(t) can be reconstructed from  $x_p(t)$  by processing it through an ideal LPF:

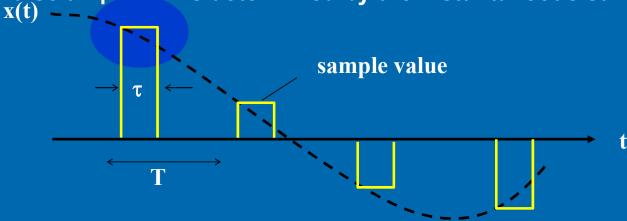
$$H(f) = \begin{cases} T/\tau ; & |f| \le |f_{\rm c}| \\ \theta ; & |f| > |f_{\rm c}| \end{cases} \quad \text{and} \quad |f_{\rm m}| \le |f_{\rm c}| \le |f_{\rm s} - f_{\rm m}|$$



#### **Practical sampling**

#### **Flat-top Sampling**

- Also known as 'instantaneous sampling'.
- Amplitude of each pulse in the sampled pulse train is constant over duration  $\tau$ . Pulse amplitude is determined by the instantaneous sample of x(t).

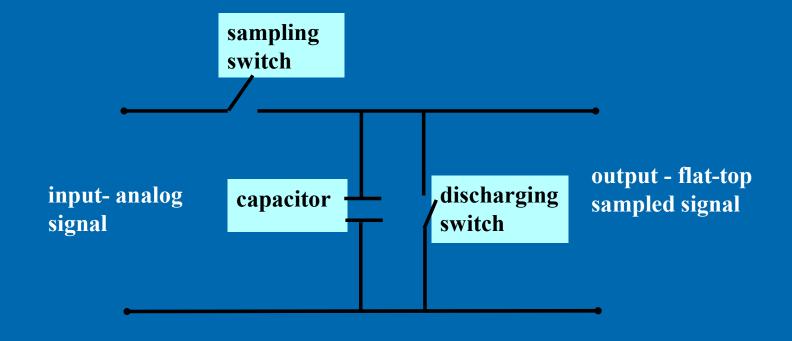


Generated by a Sample-and-Hold circuit.



**Practical sampling** 

**Flat-top Sampling** 



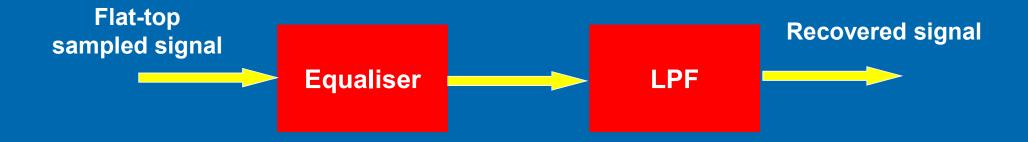
Sample-and-Hold circuit



#### **Practical sampling**

#### **Flat-top Sampling**

- Unlike Natural sampling, Flat-top sampling does not produce a distortion-free signal. This distortion is called <u>Aperture Distortion</u>.
- This distortion may be corrected by adding a second filter the equalising filter, in cascade to the reconstruction filter.







# **END**

**CHAPTER 7** 

(Part 1 of 4)

