

Superposition of Waves

[4-1]

- A very simple concept (but extremely powerful) \rightarrow

leads to:
 - polarization
 - interference
 - diffraction
- Remember the wave equation? $\nabla^2 \psi = \frac{1}{v^2} \frac{\partial^2 \psi}{\partial t^2}$
 - if $\psi_1(r, t)$ is a solution, and $\psi_2(r, t)$ is a solution, then $\psi_1 + \psi_2$ is also a solution

- Example: Plane waves $\vec{E}(\vec{r}, t) = \vec{E}_0 \sin(kx - \omega t + \phi)$ $\left[\frac{\omega}{k} = \frac{c}{n} \right]$
 - any linear combination of these is still a solution to wave eqn.
 $\Rightarrow \sum_{i=1}^N E_{0i} \sin(k_i x - \omega_i t + \phi_i)$ $\left[\frac{\omega_i}{k_i} = \frac{c}{n_i} \right]$

- let's consider a few cases to see some of the consequences

Case 1: two waves where $\omega_1 = \omega_2$, but $k_i x + \phi_i \equiv \alpha_i$ may differ

Algebraic Method (Hecht 7.1)

$$E \sin(\omega t + \alpha) = E_1 \sin(\omega t + \alpha_1) + E_2 \sin(\omega t + \alpha_2)$$

apply the sum-angle trig identity:

$$= E_1 (\sin \omega t \cos \alpha_1 + \cos \omega t \sin \alpha_1) + E_2 (\sin \omega t \cos \alpha_2 + \cos \omega t \sin \alpha_2)$$

$$= (E_1 \cos \alpha_1 + E_2 \cos \alpha_2) \sin \omega t + (E_1 \sin \alpha_1 + E_2 \sin \alpha_2) \cos \omega t$$

Complex/Phasor Method (Hecht 7.1.2)

$$E e^{i(\omega t - \alpha)} = E_1 e^{i(\omega t - \alpha_1)} + E_2 e^{i(\omega t - \alpha_2)}$$

⊕ want amplitude $E \rightarrow$ square both sides

$$\begin{aligned} E^2 &= (E_1 e^{i(\omega t - \alpha_1)} + E_2 e^{i(\omega t - \alpha_2)}) \times (E_1 e^{-i(\omega t - \alpha_1)} + E_2 e^{-i(\omega t - \alpha_2)}) \\ &= E_1^2 + E_1 E_2 e^{i(\alpha_1 - \alpha_2)} + E_1 E_2 e^{i(\alpha_2 - \alpha_1)} + E_2^2 \\ &= E_1^2 + E_2^2 + E_1 E_2 (e^{i(\alpha_1 - \alpha_2)} + e^{-i(\alpha_1 - \alpha_2)}) \\ &= E_1^2 + E_2^2 + 2E_1 E_2 \cos(\alpha_1 - \alpha_2) \quad \checkmark \end{aligned}$$

Where the Parenthetical quantities are constant in time \rightarrow these are the parts we want to find amplitude E_0 .

Just call each $(\dots) = E_0 \cos \alpha, E_0 \sin \alpha$, respectively, then square and add them

$$E_0^2 \cos^2 \alpha = E_1^2 \cos^2 \alpha_1 + 2E_1 E_2 \cos \alpha_1 \cos \alpha_2 + E_2^2 \cos^2 \alpha_2$$

$$E_0^2 \sin^2 \alpha = E_1^2 \sin^2 \alpha_1 + 2E_1 E_2 \sin \alpha_1 \sin \alpha_2 + E_2^2 \sin^2 \alpha_2$$

$$\Rightarrow E_0^2 = E_1^2 \cos^2 \alpha + E_1^2 \sin^2 \alpha$$

$$\Rightarrow E_0^2 = E_1^2 + 2E_1 E_2 \cos(\alpha_2 - \alpha_1) + E_2^2$$

Now have amplitude of the field,
What about α ?

$$\Rightarrow \tan \alpha = \frac{E_0 \sin \alpha}{E_0 \cos \alpha} = \frac{E_1 \sin \alpha_1 + E_2 \sin \alpha_2}{E_1 \cos \alpha_1 + E_2 \cos \alpha_2}$$

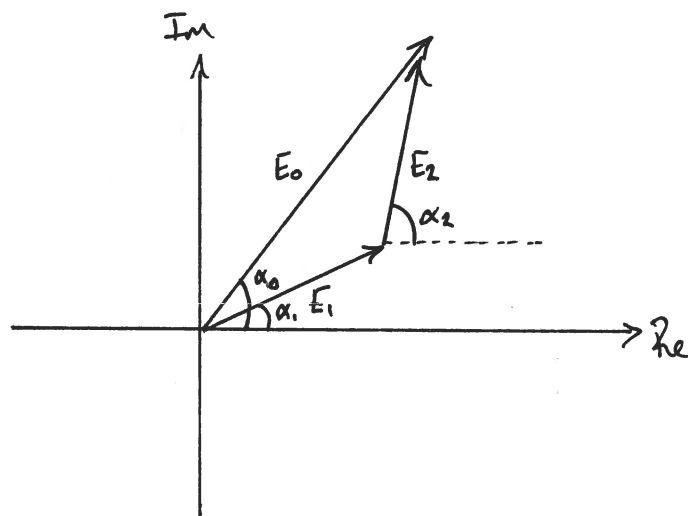
So, finally get

$$E = E_0 \cos \alpha \sin \omega t + E_0 \sin \alpha \cos \omega t$$

$$\boxed{E = E_0 \sin(\omega t + \alpha)}$$
, where we can use these expressions to find E_0 & α

Complex can be much easier! [4-2]

Using Phasors is quite handy, too:

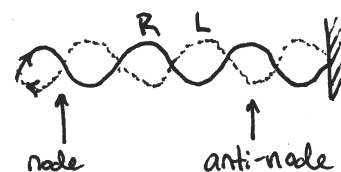


Case 2: two waves where $\omega_1 = -\omega_2$, but $k_1 = k_2$ and phases independent

• this scenario can be used to describe standing waves from reflecting surface

$$E(x,t) = E_L(x,t) + E_R(x,t) \quad \begin{cases} E_L = E_{0L} \sin(kx + \omega t + \phi_L) \\ E_R = E_{0R} \sin(kx - \omega t + \phi_R) \end{cases}$$

if $\delta = \phi_L + \phi_R = \frac{\pi}{2}$, then have situation like:



$$\text{then } E(x,t) = E_{0R} (\sin(kx + \omega t) + \sin(kx - \omega t))$$

$$= 2E_{0R} \sin kx \cos \omega t$$

$$\downarrow \text{ by } \sin \alpha + \sin \beta = 2 \sin \frac{1}{2}(\alpha + \beta) \times \cos \frac{1}{2}(\alpha - \beta)$$

Case 3: two waves where $\omega_1 \neq \omega_2$ & $k_1 \neq k_2$

[4-3]

$$\text{Here } E(x,t) = \underbrace{E_0 \cos(k_1 x - \omega_1 t)}_{E_1} + \underbrace{E_0 \cos(k_2 x - \omega_2 t)}_{E_2}$$

$$= 2E_0 \cos \frac{1}{2}[(k_1 + k_2)x - (\omega_1 + \omega_2)t] \times \cos \frac{1}{2}[(k_1 - k_2)x - (\omega_1 - \omega_2)t]$$

(using trig identity: $\cos \alpha + \cos \beta = 2 \cos \frac{1}{2}(\alpha + \beta) \cos \frac{1}{2}(\alpha - \beta)$)

And now see these average & difference frequencies & wavenumbers

avg. angular freq.:

$$\bar{\omega} = \frac{1}{2}(\omega_1 + \omega_2)$$

avg. wavenumber:

$$\bar{k} = \frac{1}{2}(k_1 + k_2)$$

modulation freq.:

$$\Delta \omega = \frac{1}{2}(\omega_1 - \omega_2)$$

modulation wavenumber:

$$\Delta k = \frac{1}{2}(k_1 - k_2)$$

given these definitions
can always find:

$$\omega_1 = \bar{\omega} + \Delta \omega$$

$$\omega_2 = \bar{\omega} - \Delta \omega$$

$$k_1 = \bar{k} + \Delta k$$

$$k_2 = \bar{k} - \Delta k$$

$$\text{Now } E = 2E_0 \underbrace{\cos(\Delta k x - \Delta \omega t)}_{\text{slow "beat" or "envelope" function}} \underbrace{\cos(\bar{k} x - \bar{\omega} t)}_{\text{fast "carrier" wave}}$$

[See Slide, or Fig. 7.11e Hecht]

- So far, no big surprises.
- However, in general (i.e.: in a dielectric), waves with different frequencies travel at different velocities.
- May even already suspect that the carrier & envelopes have their own characteristic velocities

Velocity of single monochromatic wave: $\left[\psi_1 \propto \cos(k_1 x - \omega_1 t) \right]$ [4-4]
• have already seen this (Phase velocity)

$$\left(\frac{\partial x}{\partial t} \right) = - \frac{(\partial \psi_1 / \partial t)}{(\partial \psi_1 / \partial x)} = \frac{\omega_1}{k_1} = v_p$$

Velocity of the "carrier": $\left[\bar{\psi} \propto \cos(\bar{k} x - \bar{\omega} t) \right]$

$$\bar{v} = - \frac{(\partial \bar{\psi} / \partial t)}{(\partial \bar{\psi} / \partial x)} = \frac{\bar{\omega}}{\bar{k}} = \frac{\omega_1 + \omega_2}{k_1 + k_2}$$

Velocity of the "envelope": $\left[\Delta \psi \propto \cos(\Delta k x - \Delta \omega t) \right]$

$$v_g = - \frac{(\partial \Delta \psi / \partial t)}{(\partial \Delta \psi / \partial x)} = \frac{\Delta \omega}{\Delta k} = \frac{\omega_1 - \omega_2}{k_1 - k_2}$$

• for EM waves in vacuum, everything travels at c

$$\frac{\omega_1}{k_1} = \frac{\omega_2}{k_2} = c = \bar{v} = v_g$$

• in general, though, refractive index varies with frequency

$$v_1 = \frac{\omega_1}{k_1} = \frac{c}{n_1} \neq \frac{c}{n_2} = \frac{\omega_2}{k_2} = v_2$$

• ultimately, we'll take this to the infinitesimal limit

and see that $v_g = \left(\frac{d\omega}{dk} \right)_{\bar{\omega}}$ ← group velocity of a wave

Dispersion Relations

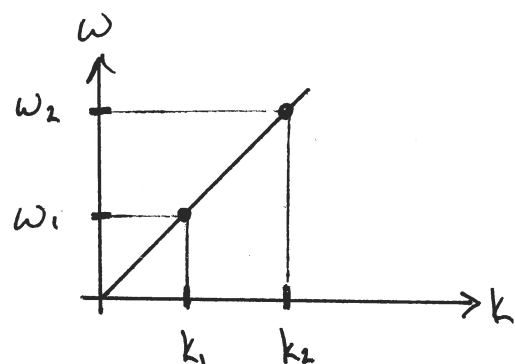
- In physics, we'll often encounter relations between energy and momentum - so-called dispersion relations.
- These are essentially a relation between frequency and wavelength of a particle/field/whatever (In quantum have: $E = \hbar\omega$ & $p = \hbar k$)

Ex: Linear Dispersion (photons in vacuum)

$$\omega(k) = kc$$

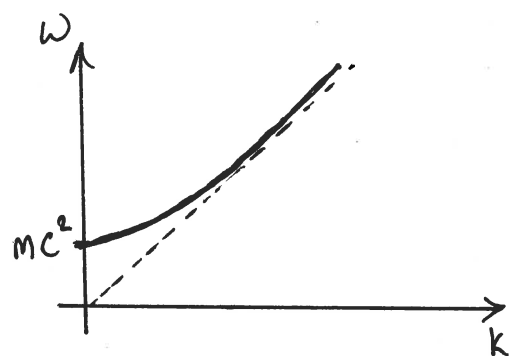
- Could even define indices of refraction for phase & group velocity components

$$n = \frac{c}{v} \Rightarrow n_{\phi} = c/(\omega/k) \text{ \& \ } n_g = c/(\frac{\partial\omega}{\partial k}) \rightarrow \text{both are 1 in vac.}$$



Ex: Massive Free Particle (ex: electron)

$$\omega(k) = \frac{1}{\hbar} \left(m^2 c^4 + \frac{\hbar^2 k^2}{2m} \right)^{\frac{1}{2}}$$

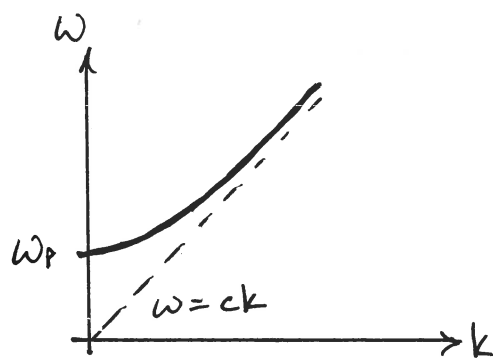


Ex: Dispersion of EM waves in a plasma

- neglecting Ohmic losses (damping)

$$\omega(k) = (\omega_p^2 + c^2 k^2)^{\frac{1}{2}}$$

$$\text{remember: } \omega_p = \sqrt{\frac{Nq^2}{\epsilon_0 m}}$$



Group Velocity

[4-6]

- ω is always the same regardless of the material
→ λ (and therefore $k = \frac{2\pi}{\lambda}$) and velocity change, though
- $k(\lambda)$ depends on the refractive index, n , which depends on ω
→ $v = \frac{\omega}{k} = \frac{c}{n} \Rightarrow k = \frac{\omega}{c} n(\omega)$
- above, we defined the group velocity as $v_g = \frac{\partial \omega}{\partial k}$
 - Since we usually have $n(\omega)$, this isn't always the most useful form
- More useful: $\frac{\partial k}{\partial \omega} = \frac{\partial}{\partial \omega} \left(\frac{\omega}{c} n(\omega) \right) = \frac{1}{c} \left(n(\omega) + \omega \frac{\partial n(\omega)}{\partial \omega} \right)$

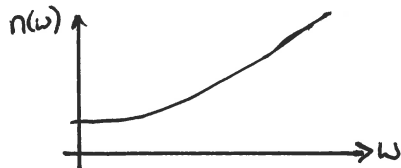
$$\Rightarrow \boxed{v_g = \frac{c}{n + \omega \frac{\partial n}{\partial \omega}} = \frac{c/n}{1 + \frac{\omega}{n} \frac{\partial n}{\partial \omega}} = \frac{v_{\text{phase}}}{1 + \frac{\omega}{n} \frac{\partial n}{\partial \omega}}}$$

- in general, $v_g \neq v_{\text{phase}}$ (and can be very different)

- kind of saw this coming - infinitesimal limit of $v_g = \frac{\Delta \omega}{\Delta k}$, where $\Delta \omega = \frac{\omega_1 - \omega_2}{2}$
and $\Delta k = \frac{k_1 - k_2}{2} = \frac{1}{2} \left(\frac{\omega_1 n(\omega_1)}{c} - \frac{\omega_2 n(\omega_2)}{c} \right)$
- 3 cases to consider:

"Normal" Dispersion

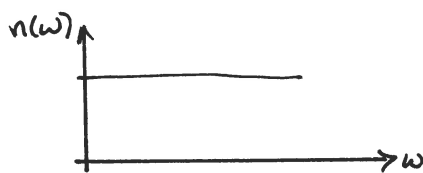
$$\frac{\partial n}{\partial \omega} > 0 \quad \& \quad v_g < v_{\text{phase}}$$



envelope velocity < carrier vel.

Zero Dispersion

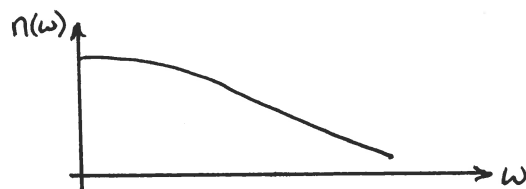
$$\frac{\partial n}{\partial \omega} = 0 \quad \& \quad v_g = v_{\text{phase}}$$



envelope vel. = carrier vel.

"Anomalous" Dispersion

$$\frac{\partial n}{\partial \omega} < 0 \quad \& \quad v_g > v_{\text{phase}}$$



envelope vel. > carrier vel. (!)

Fourier Transform Methods

[4-7]

- the Fourier transform is a way of rewriting a (potentially) complicated function in a (potentially) more manageable or useful form
- the core concept here: Conjugate Variables

$$\left[\begin{array}{cc} f(t) & \longleftrightarrow f(\omega) \\ \text{(time)} & \text{(frequency)} \end{array} \right] \quad \left[\begin{array}{cc} f(x) & \longleftrightarrow f(k) \\ \text{(space)} & \text{("spatial freq.")} \end{array} \right]$$

- for now, we'll stick to time/freq. as the transform variables and limit dimensions to 1D, but we'll expand this to 2D space/spatial freq.
- techniques/expressions for converting between representations varies a bit in their details depending on things like:
 - is the function periodic?
 - is it real or complex valued?
- we'll cover these different cases, but is essentially all the same trick!
- What's the trick?
 - Every well behaved function can be written as a sum of harmonic fn's
 - "well behaved" here is itself a topic of study within Fourier Analysis (we'll stay within very comfortable bounds for our needs, though)

• let's take a look at our "building blocks"

[4-8]

time: $\{ \sin(\omega t), \cos(\omega t), e^{i\omega t} \}$

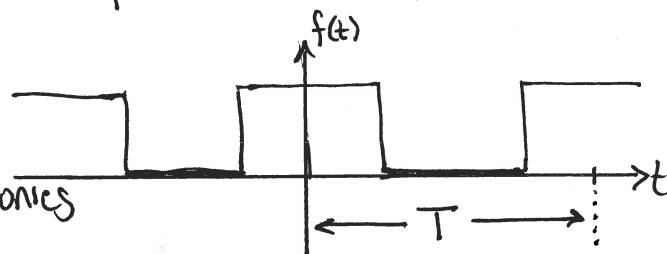
space: $\{ \sin(kx), \cos(kx), e^{ikx} \}$

Periodic Functions

Consider a function $f(t)$ that happens to be periodic with period T

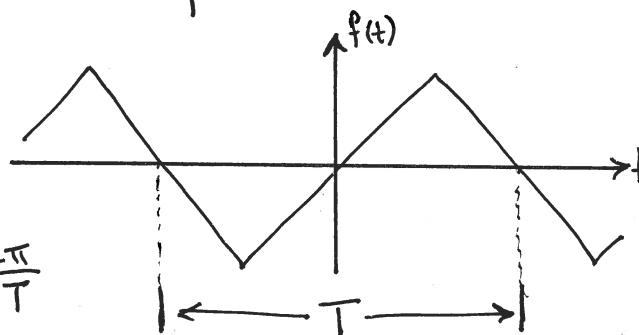
• if it's even, we can express

the function as a sum of cosine harmonics



$$f(t) = \frac{A_0}{2} + \sum_{m=1}^{\infty} A_m \cos(m\omega_0 t), \quad \omega_0 = \frac{2\pi}{T}$$

• if it's odd, we use sine harmonics



$$f(t) = \sum_{m=1}^{\infty} B_m \sin(m\omega_0 t), \quad \omega_0 = \frac{2\pi}{T}$$

[Note: B_0 is always zero for an odd function]

Some general function that isn't particularly even or odd can be expressed as

$$f(t) = \frac{A_0}{2} + \sum_{m=1}^{\infty} A_m \cos(m\omega_0 t) + \sum_{m=1}^{\infty} B_m \sin(m\omega_0 t) \quad \leftarrow \text{(Synthesis fn.)}$$

How accurately does this mimic the 'actual' function?

→ Depends on how many terms we keep in the sum.

(A very important issue for data compression!)

• So, can add up harmonics to produce complicated functions

[4-9]

• Can we do the reverse?

→ given some function, how to find the constituent harmonic components
(how do we find each A_m & B_m ?)

• Obviously need some mathematical "measure" of how much of a certain harmonic appears in a given function (becomes a calculus problem!)

What to do mathematically:

Use the property that the functions $\sin(m\omega t)$ form an "orthogonal basis set" → $\int_0^T \sin(n\omega t) \sin(m\omega t) dt = \delta_{nm} = \begin{cases} \frac{T}{2} & \text{if } n=m \\ 0 & \text{if } n \neq m \end{cases}$

"Kronecker delta"

then can write: $f(t) = \sum_{m=1}^{\infty} B_m \sin(m\omega t)$

$$\Rightarrow f(t) \sin(n\omega t) = \sum_{m=1}^{\infty} B_m \sin(n\omega t) \sin(m\omega t)$$

$$\Rightarrow \int_0^T f(t) \sin(n\omega t) dt = \int_0^T \sum_{m=1}^{\infty} B_m \sin(n\omega t) \sin(m\omega t) dt$$

$$\Rightarrow \int_0^T f(t) \sin(n\omega t) dt = B_n \frac{T}{2} \quad \left[\begin{array}{l} \text{Orthogonality:} \\ = T/2 \text{ if } n=m, \text{ Zero otherwise} \end{array} \right]$$

$$\Rightarrow B_n = \frac{2}{T} \int_0^T f(t) \sin(n\omega t) dt \quad \left. \vphantom{\int_0^T} \right\} \text{(analysis fn's)}$$

$$\text{and same for: } A_n = \frac{2}{T} \int_0^T f(t) \cos(n\omega t) dt \quad \left. \vphantom{\int_0^T} \right\} \text{Fourier Coefficients}$$

What we just did mathematically:

[4-10]

- ① take the function we want to decompose and multiply it by the harmonic whose coefficient we want (basically a measure of the "overlap" b/w fns)
 - ② integrate the product over one period of the function \rightarrow gives a measure of how much of that harmonic function is present in our periodic function
- so have reduced the problem of finding the coefficients to a calculus problem (sometimes a not-so-simple calc. problem)
 - why go through the trouble? many times it's easier to solve a problem in the freq. domain and then convert back to time domain

The Complex Fourier Transform (not in Hecht, but should be)

- seen many times the inconvenience of trig calculations (same is true here)
- start with the same synthesis relation

$$f(t) = A_0 + 2 \sum_{m=1}^{\infty} (A_m \cos(m\omega t) + B_m \sin(m\omega t))$$

- but then let the sum run from $-\infty$ to ∞ instead. What happens?

Remember: $\cos(m\omega t) = \cos(-m\omega t)$ so $A_m = A_{-m}$ [even]

$\sin(m\omega t) = -\sin(-m\omega t)$ so $B_m = -B_{-m}$ [odd]

- using Euler, this would now be equivalent to

$$f(t) = A_0 + B_0 + \sum_{m=1}^{\infty} A_m (e^{im\omega t} + e^{-im\omega t}) - i B_m (e^{im\omega t} - e^{-im\omega t})$$

- So, in general, for m over the entire range $-\infty < m < \infty$ [4-11]

$$f(t) = \sum_{m=-\infty}^{\infty} A_m e^{-im\omega_0 t} - iB_m e^{-im\omega_0 t}$$

$$= \sum_{m=-\infty}^{\infty} C_m e^{-im\omega_0 t}, \text{ where } C_m = A_m - iB_m$$

- the corresponding analysis relation is:

$$C_m = \frac{1}{T} \int_0^T f(t) e^{im\omega_0 t} dt$$

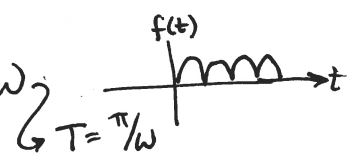
$$\left| \begin{aligned} A_m &= \frac{C_m + C_m^*}{2} = \frac{C_m + C_{-m}}{2} \\ B_m &= \frac{C_m - C_m^*}{2} = \frac{C_m - C_{-m}}{2} \end{aligned} \right.$$

- Bonus: Can handle $\tilde{f}(t)$ when it's complex, too.

Evaluating Fourier Sums

- make use of even/odd symmetries when setting things up
- pay attention to the domain (period T), sometimes can be simplified by "folding" the integration domain
- is always best to find closed-form expressions for Fourier coeff.'s (if able)
- the form of the Fourier sum is not necessarily unique
 - different but equivalent forms by choosing trig or exp expansions, or shifting the sum's starting point (~~ex~~ $m=0$ or $m=1$)

Example: Fourier sum for a rectified sine wave $f(t) = |\sin \omega t|$

- note that the effective frequency is now 2ω , not ω . 
- $$\Rightarrow f(t) = |\sin \omega t| = \sum_{n=-\infty}^{\infty} C_n e^{-in(2\omega)t}$$

- What coefficients are needed to construct this function?

$$C_n = \frac{1}{T} \int_0^T |\sin \omega t| e^{in(2\omega)t} dt \quad [4-12]$$

• but see that $|\sin \omega t| = \sin \omega t = \frac{e^{i\omega t} - e^{-i\omega t}}{2i}$ over interval $[0, \frac{\pi}{\omega}]$

$$\Rightarrow C_n = \frac{1}{T} \int_0^T \frac{1}{2i} (e^{i\omega t} - e^{-i\omega t}) e^{in(2\omega)t} dt$$

$$= \frac{1}{2iT} \int_0^T (e^{i(2n+1)\omega t} - e^{-i(2n-1)\omega t}) dt$$

$$= \frac{1}{2\pi i} \int_0^\pi (e^{i(2n+1)\tau} - e^{i(2n-1)\tau}) d\tau$$

change of variables
 $\tau = \omega t$
 $d\tau = \omega dt$
 $\Rightarrow dt = \frac{1}{\omega} d\tau = \frac{T}{\pi} d\tau$

$$= \frac{1}{2\pi i} \left[\frac{e^{i(2n+1)\pi} - 1}{i(2n+1)} - \frac{e^{i(2n-1)\pi} - 1}{i(2n-1)} \right]$$

• here, we're lucky since $e^{i(2n\pm 1)\pi} = 1$ for all integer n

$$\Rightarrow C_n = \frac{1}{2\pi i} \left[\frac{(-2)}{i(2n+1)} - \frac{(-2)}{i(2n-1)} \right] = \frac{2}{\pi(1-4n^2)}$$

• so, finally have $f(t) = |\sin \omega t| = \sum_{n=-\infty}^{\infty} \frac{2}{\pi(1-4n^2)} e^{-in(2\omega)t}$ ✓

• notice here that $C_n = C_{-n} = C_n^*$. This is because the function is real and even which is a relatively simple case

• not everything of interest is a periodic signal  or

• to consider a finite signal, consider it to be periodic on an infinite time scale! $T \rightarrow \infty$.

• what changes then?

- Remember had:

[4-13]

$$f(t) = \sum_{m=-\infty}^{\infty} C_m e^{-im\omega t} \quad \& \quad C_m = \frac{1}{T} \int_0^T f(t) e^{im\omega t} dt$$

- for the coefficients, can choose our bounds to be $\int_{-T/2}^{T/2}$ instead
- given a finite signal, nothing exists outside these bounds, so can just as well allow $T \rightarrow \infty$ so we have

$$C_m = \frac{1}{T} \int_{-\infty}^{\infty} f(t) e^{im\omega t} dt \Rightarrow TC_m = \int_{-\infty}^{\infty} f(t) e^{im\omega t} dt$$

- but as $T \rightarrow \infty$, TC_m no longer describes a discrete set of integer multiples of the fundamental freq: $m\omega \rightarrow \omega$ (continuous)

$$\Rightarrow TC_m \Big|_{T \rightarrow \infty} \equiv \underline{\underline{\tilde{f}(\omega) = \int_{-\infty}^{\infty} f(t) e^{i\omega t} dt}} \quad \left[\begin{array}{l} \text{Fourier} \\ \text{Transform} \end{array} \right]$$

- also, can then write $f(t)$ as:

$$f(t) = \sum_{m=-\infty}^{\infty} TC_m e^{-im\omega t} \frac{1}{T}$$

$$= \sum_{m=-\infty}^{\infty} TC_m e^{-im\omega t} \frac{\Delta\omega}{2\pi}$$

$$= \frac{1}{2\pi} \int_{-\infty}^{\infty} \tilde{f}(\omega) e^{-i\omega t} d\omega$$

$\left[\begin{array}{l} \text{Inverse} \\ \text{Fourier} \\ \text{Transform} \end{array} \right]$

$$T = 2\pi/\Delta\omega$$

$$\text{as } T \rightarrow \infty, \Delta\omega \rightarrow 0$$

$$\Rightarrow \Delta\omega \rightarrow d\omega$$

$$\Rightarrow \sum_{m=-\infty}^{\infty} \frac{1}{T} \rightarrow \int_{-\infty}^{\infty} d\omega/2\pi$$

- Here we have the Fourier Transform pair: $f(t) \xleftrightarrow{\mathcal{F}} \tilde{f}(\omega)$

$$\mathcal{F}^{-1}[\mathcal{F}[f(t)]] = f(t)$$

$\left[\begin{array}{l} \text{time} \\ \text{domain} \end{array} \right]$

$\left[\begin{array}{l} \text{freq} \\ \text{domain} \end{array} \right]$

- note that normalization conventions may vary b/w sources. [4-14]

For us, though:
$$\frac{1}{2\pi} \int_{-\infty}^{\infty} \left[\int_{-\infty}^{\infty} f(t) e^{i\omega t} dt \right] e^{-i\omega t} d\omega = f(t)$$

where the $\frac{1}{2\pi}$ needs to be there to balance $d\omega$ & dt

But, could also write in terms of $\nu = \frac{\omega}{2\pi}$, then

$$\int_{-\infty}^{\infty} \left[\int_{-\infty}^{\infty} f(t) e^{i2\pi\nu t} dt \right] e^{-i2\pi\nu t} d\nu$$

- in terms of trig fn's, these would look like

$$f(t) = \frac{1}{\pi} \left[\int_0^{\infty} A(\omega) \cos(\omega t) d\omega + \int_0^{\infty} B(\omega) \sin(\omega t) d\omega \right]$$

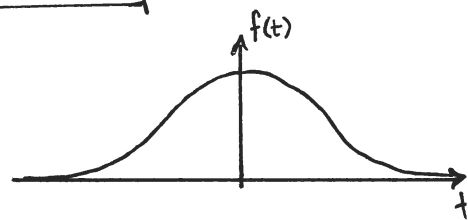
where $A(\omega) = \int_{-\infty}^{\infty} f(t) \cos(\omega t) dt$ & $B(\omega) = \int_{-\infty}^{\infty} f(t) \sin(\omega t) dt$

Wave Packets

- Hecht goes over examples of 2 important cases: the square pulse and cosine wave packet.

- Another very common pulse form is the Gaussian pulse

gaussian function: $f(t) = A e^{-\alpha t^2}$



- Apply Fourier Transform:

$$\mathcal{F}[f(t)] = \tilde{f}(\omega) = \int_{-\infty}^{\infty} A e^{-\alpha t^2 + i\omega t} dt$$

- to evaluate this integral, need to use a trick:

Complete the square! where in general

[4-15]

$$at^2 + bt + c = a\left(t + \frac{b}{2a}\right)^2 + c - \frac{b^2}{4a}$$

So, with $a = -\alpha$, $b = i\omega$, and $c = 0$, the exponent is

$$\int_{-\infty}^{\infty} A e^{-\alpha\left(t - \frac{i\omega}{2\alpha}\right)^2 + 0 - \frac{\omega^2}{4\alpha}} dt = A e^{-\frac{\omega^2}{4\alpha}} \int_{-\infty}^{\infty} e^{-\alpha\left(t - \frac{i\omega}{2\alpha}\right)^2} dt$$

and (another trick) see that integral is the same if we substitute

$t \rightarrow t + b$ (okay since limits are $\pm\infty$)

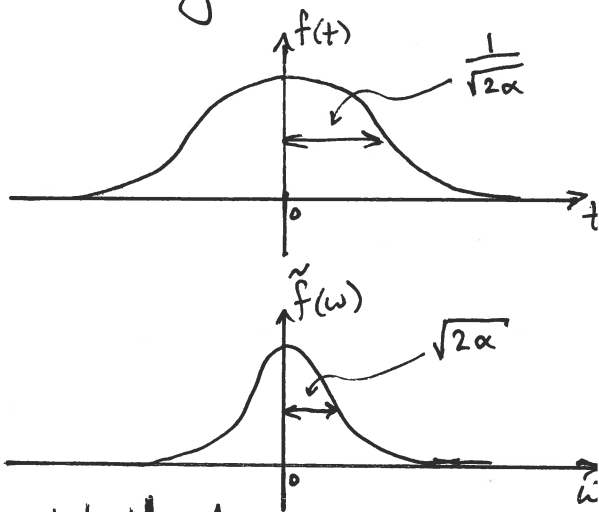
$$\Rightarrow \tilde{f}(\omega) = A e^{-\frac{\omega^2}{4\alpha}} \int_{-\infty}^{\infty} e^{-\alpha t^2} dt = A e^{-\frac{\omega^2}{4\alpha}} \sqrt{\pi/\alpha}$$

So the Fourier Transform of a gaussian is a gaussian!

$$A e^{-\alpha t^2} \xleftrightarrow{\mathcal{F}} A \sqrt{\frac{\pi}{\alpha}} e^{-\frac{\omega^2}{4\alpha}}$$

temporal width of gaussian: $\sigma_t = \frac{1}{\sqrt{2\alpha}}$

spectral width of gaussian: $\sigma_\omega = \sqrt{2\alpha}$



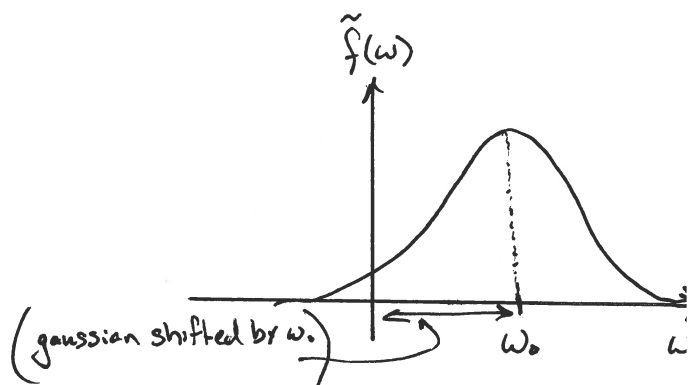
Spectral & temporal widths are inversely related!

Note: Negative freq's!

What about off-center gaussians?

$$f(t) = A e^{-\frac{t^2}{2\tau^2} - i\omega_0 t}$$

$$\tilde{f}(\omega) = \frac{A}{\sqrt{2\pi}} \sqrt{\frac{\pi}{\tau^2/2}} e^{-\frac{(\omega - \omega_0)^2}{4(\tau^2/2)}}$$



Coherence of Optical Fields

[4-16]

- in general, a non-zero spectral width (σ_ω) of a physical optical source means the field isn't perfectly coherent (we'll formalize the idea of coherence a bit more later)

Coherence time: $\delta t_c = \frac{1}{\delta \nu}$

Coherence length: $\delta l_c = c \delta t_c$

$$\nu = \frac{c}{\lambda}, \text{ but } \delta \nu \neq \frac{c}{\delta \lambda}$$

$$\Rightarrow \delta \nu = \frac{c \delta \lambda}{\lambda^2}$$