

Computerized control - state feedback

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Goal of today's lecture

- ▶ Understand state feedback design

Stability

A system

$$x(k+1) = \Phi x(k), \quad x(0) = x_0$$

is *asymptotically stable* if $\lim_{t \rightarrow \infty} x(kh) = 0$ for all $x_0 \in \mathbb{R}^n$.

A system is asymptotically stable if and only if **all eigenvalues of Φ are inside the unit circle.**

Reachability (controllability)

Reachability is the answer to the question "Can we by choosing a suitable input sequence $u(k)$, $k = 0, 1, 2, \dots, n - 1$ reach any point in the state space?"

Consider

$$x(k + 1) = \Phi x(k) + \Gamma u(k).$$

With initial state $x(0)$ given. The solution at time n where n is the order of the system is

$$\begin{aligned} x(n) &= \Phi^n x(0) + \Phi^{n-1} \Gamma u(0) + \dots + \Gamma u(n-1) \\ &= \Phi^n x(0) + W_c U, \end{aligned} \tag{1}$$

where

$$\begin{aligned} W_c &= [\Gamma \quad \Phi \Gamma \quad \dots \quad \Phi^{n-1} \Gamma] \\ U &= [u(n-1) \quad u(n-2) \quad \dots \quad u(0)]^T \end{aligned}$$

Reachability (controllability), contd

To find the input sequence that takes the state to $x(n) = x_d$ we solve the equation

$$x_d = \Phi^n x(0) + W_c U$$

for U .

$$U = W_c^{-1} (x_d - \Phi^n x(0))$$

This requires the matrix W_x to be **invertible**. This gives Theorem 3.7 in Å&W:

THEOREM 3.7 REACHABILITY The state space system above is reachable if and only if the matrix W_c has rank n .

This is equivalent to

$$\det W_c \neq 0.$$

State feedback

Have state space model

$$\begin{aligned}x(k+1) &= \Phi x(k) + \Gamma u(k) \\ y(k) &= Cx(k)\end{aligned}\tag{2}$$

and measurements (or estimates) of the state vector $x(k)$.

Linear state feedback is the control law

$$\begin{aligned}u(k) &= f((x(k), u_c(k))) = -l_1 x_1(k) - l_2 x_2(k) - \cdots - l_n x_n(k) + m u_c(k) \\ &= -Lx(k) + m u_c(k),\end{aligned}$$

where

$$L = \begin{bmatrix} l_1 & l_2 & \cdots & l_n \end{bmatrix}.$$

Insert the control law into the state space model (3) to get

State feedback

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Linear state feedback is the control law

$$u(k) = -l_1 x_1(k) - l_2 x_2(k) - \cdots - l_n x_n(k) + m u_c(k) = -Lx(k) + m u_c(k),$$

where

$$L = [l_1 \quad l_2 \quad \cdots \quad l_n] .$$

Insert the control law into the state space model (3) to get

$$\begin{aligned}x(k+1) &= (\Phi - \Gamma L) x(k) + m \Gamma u_c(k) \\ y(k) &= Cx(k)\end{aligned}\tag{4}$$

Pole placement by state feedback

Assume the desired performance of the control system is given as a set of desired closed loop poles p_1, p_2, \dots, p_n , corresponding to the desired characteristic polynomial

$$a_c(z) = (z - p_1)(z - p_2) \cdots (z - p_n) = z^n + \alpha_1 z^{n-1} + \cdots + \alpha_n. \quad (5)$$

With state feedback we get the the closed-loop system

$$\begin{aligned} x(k+1) &= (\Phi - \Gamma L)x(k) + m\Gamma u_c(k) \\ y(k) &= Cx(k) \end{aligned} \quad (6)$$

with characteristic equation

$$\det(zI - (\Phi - \Gamma L)) = z^n + \beta_1(l_1, \dots, l_n)z^{n-1} + \cdots + \beta_n(l_1, \dots, l_n). \quad (7)$$

Equate the coefficients in (5) and (7) to get the system of equations

$$\beta_1(l_1, \dots, l_n) = \alpha_1$$

$$\beta_2(l_1, \dots, l_n) = \alpha_2$$

$$\vdots$$

$$\beta_n(l_1, \dots, l_n) = \alpha_n$$

Pole placement by state feedback, contd.

The system of equations

$$\beta_1(l_1, \dots, l_n) = \alpha_1$$

$$\beta_2(l_1, \dots, l_n) = \alpha_2$$

$$\vdots$$

$$\beta_n(l_1, \dots, l_n) = \alpha_n$$

is always linear in the unknown controller parameters, so it can be written

$$AL^T = \alpha,$$

Where $\alpha^T = [\alpha_1 \quad \alpha_2 \quad \cdots \quad \alpha_n]$.

Pole placement and reachability

It can be shown that the controllability matrix W_c is a factor of the matrix A

$$A = \bar{A}W_c.$$

Hence, in general the system of equations

$$\bar{A}W_cL^T = \alpha \tag{8}$$

has a solution only if W_c is invertible, i.e. the system is *reachable*.

Note that equation (8) can still have a solution for unreachable systems if α is in the *column space of A* , i.e. α can be written

$$\alpha = b_1A_{:,1} + b_2A_{:,2} + \cdots + b_mA_{:,m}, \quad m < n$$