

Compensator design - Loop shaping

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Specifications on the frequency properties of the closed-loop system

The design procedure - overview

Specifications on the closed-loop system $G_c(i\omega)$



Specifications on the loop gain $G_o(i\omega)$



Determine $F(i\omega)$ in $G_o(i\omega) = G(i\omega)F(i\omega)$

From specifications on G_c to specifications on G_o

Closed-loop specifications	Loop gain specifications
Bandwidth ω_B	cross-over frequency ω_c
Resonance peak M_p	phase margin φ_m
Static gain $G_c(0) \approx 1$	static gain $G_o(0)$ high

$$e_0 = |G_c(0) - 1| = \left| \frac{G_o(0)}{1 + G_o(0)} - 1 \right| = \left| \frac{1}{1 + G_o(0)} \right| < \epsilon$$
$$\Rightarrow$$

$$G_o(0) > ?$$

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$$G_o(0) > \frac{1}{\epsilon} - 1$$

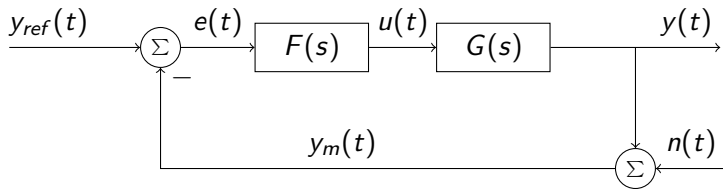
Design procedure in detail

Given $G(i\omega)$ and specifications on $G_o(i\omega)$: ω_c , φ_m , steady-state error e_0 .

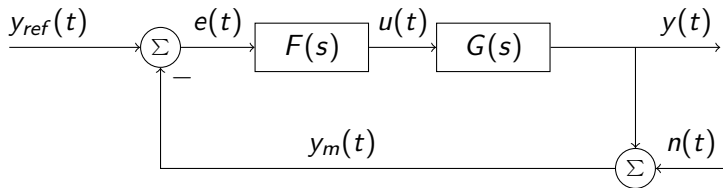
The problem with a PD-controller



The problem with a PD-controller, contd

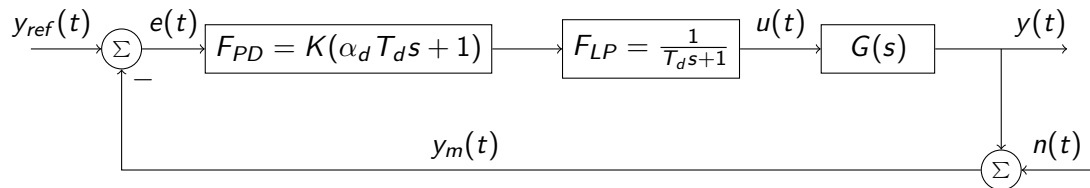


The problem with a PD-controller, contd



High frequency measurement noise entering the system is amplified in the PD-controller $F(s)$

PD-controller + Low-pass filter = lead compensator + gain



$$F(s) = KF_{lead} = K \frac{\alpha T_d s + 1}{T_d s + 1}$$

The lead- and lag filters/compensators

$$F_{lead} = \frac{\alpha_d T_d s + 1}{T_d s + 1}, \alpha_d > 1 \quad F_{lag} = \frac{1}{\alpha_i} \cdot \frac{\alpha_i T_i s + 1}{T_i s + 1}, \alpha_i < 1 \text{ or } F_{lag} = \frac{T_i s + 1}{T_i s}$$

Position control of a radar antenna

Nyquist plot of the plant

Will proportional control work? (The open-loop system is stable)