

# Digital Signal Processing

Class 6  
02/06/2025

# ENGR 71

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- Class Overview
  - Discrete-Time Signals and Systems
- Assignments
  - Reading:  
Chapter 2: Discrete-Time Signals and Systems
  - Lab 1 – Aliasing lab
    - Will be up on Moodle this afternoon

- Lab 1-Aliasing Lab
  - Find a short piece of music to download
  - Subsample to demonstrate aliasing
  - Repeat the subsampling, but prior to subsampling, apply a low-pass anti-aliasing filter.
  - Compare the results
- More details and sample code will be placed on Moodle

# Class Information

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- Topics in Discrete-Time Signals and Systems
  - Discrete-Time Signals
  - Discrete-Time Systems
  - Analysis of Linear Time-Invariant Systems
  - Description of Systems by Difference Equations
  - Implementation of Discrete-Time Systems
  - Correlation of Discrete-Time Systems

Some aspects of these topics have already been discussed in the review of continuous systems

# Discrete-Time Signals

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- **Signals**

(already covered in review of continuous signals)

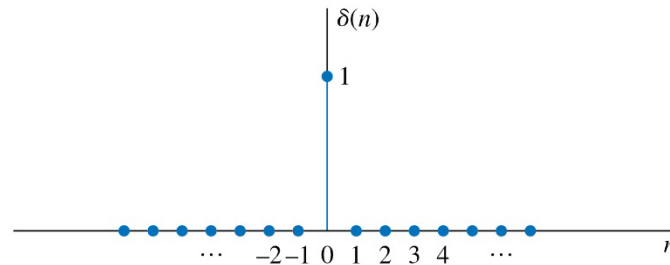
- Discrete-signals are function of an integer index
- Signal described by  $x(n)$  where  $n$  is an integer indicating the sample number.

# Discrete-Time Signals

- Elementary discrete-time signals
  - Unit sample sequence (impulse)

$$\delta(n) = 0, \quad n \neq 0$$

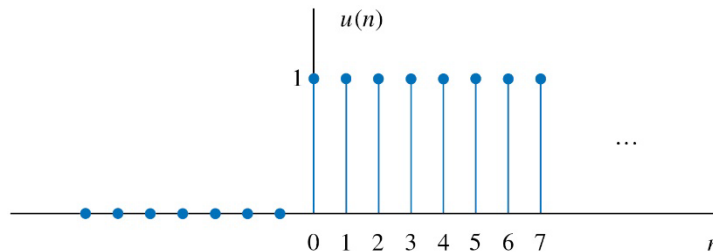
$$\delta(n) = 1, \quad n = 0$$



- Unit step signal

$$u(n) = 1, \quad n \geq 0$$

$$u(n) = 0, \quad n < 0$$



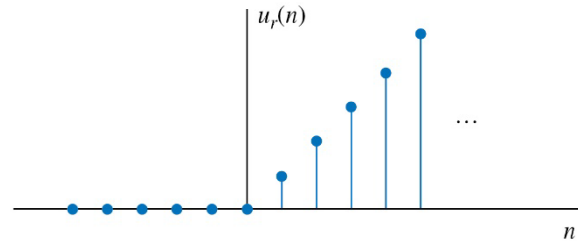
# Discrete-Time Signals

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- Elementary discrete-time signals
  - Unit ramp

$$u_r(n) = n, \quad n \geq 0$$

$$u_r(n) = 0, \quad n < 0$$

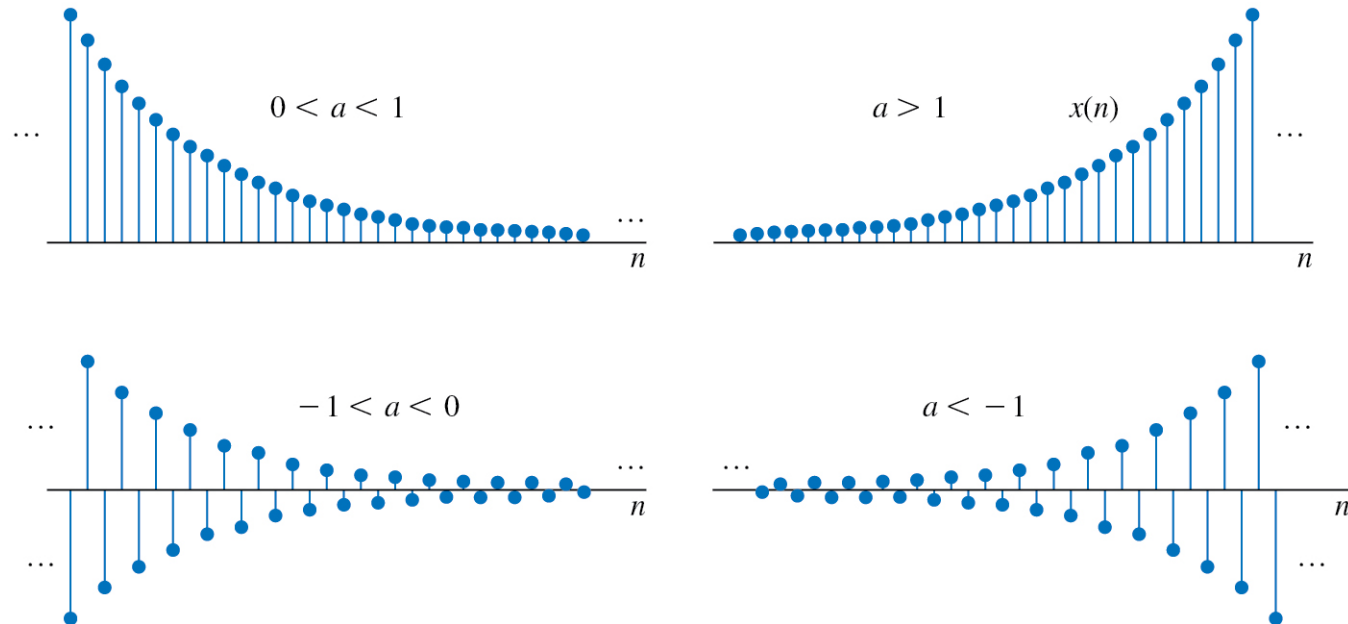


# Discrete-Time Signals

- Elementary discrete-time signals
  - Exponential signal (one we haven't discussed)

$$x(n) = a^n, \text{ for all } n$$

If  $a$  is real,  $x(n)$  is real-valued signal.





# Discrete-Time Signals

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- Elementary discrete-time signals
  - Exponential signal (one we haven't discussed)

If  $a$  is complex,  $x(n)$  is complex-valued signal.

$$a = re^{j\theta}$$

$$x(n) = a^n = r^n e^{j\theta n} = r^n (\cos \theta n + j \sin \theta n)$$

$$x_R(n) = r^n \cos \theta n \quad (\text{Real part of signal})$$

$$x_I(n) = r^n \sin \theta n \quad (\text{Imaginary part of signal})$$

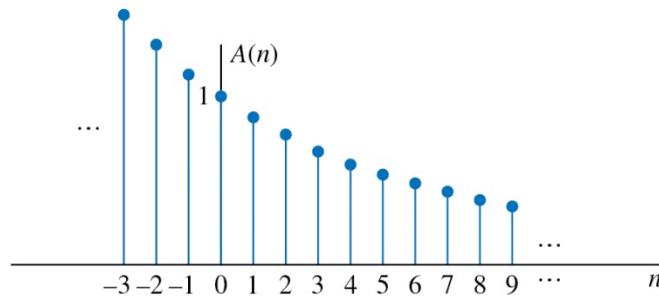
$$|x(n)| = A(n) = r^n \quad (\text{Magnitude of signal})$$

$$\angle x(n) = \phi(n) = \theta n \quad (\text{Phase of signal})$$

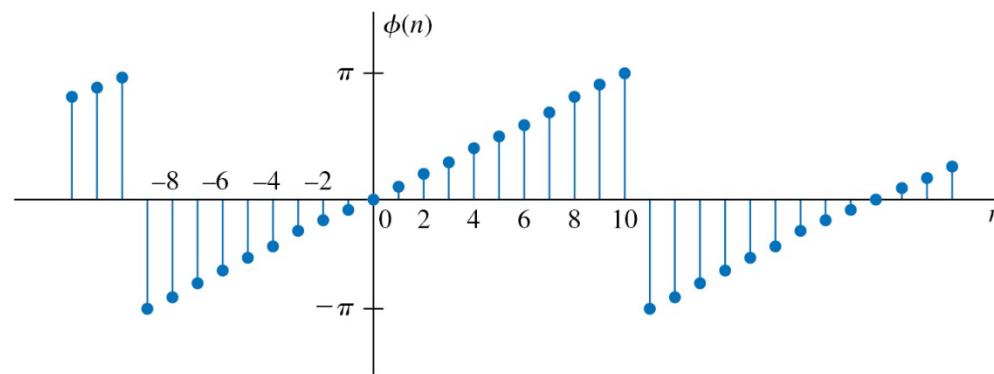
Since it wraps, only consider range from  $-\pi$  to  $\pi$

# Discrete-Time Signals

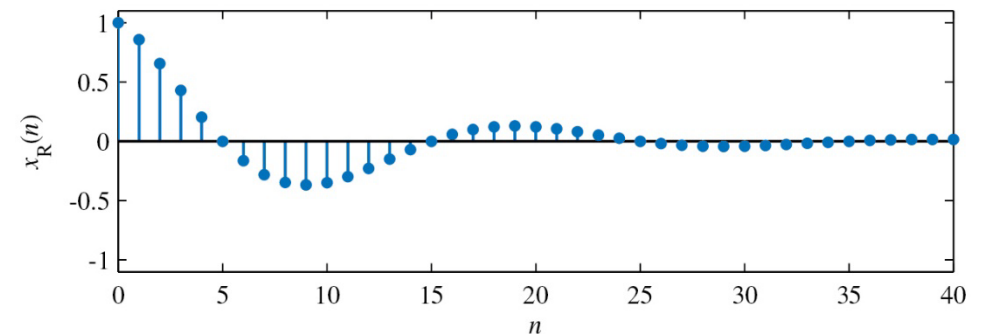
- Elementary discrete-time signals
  - Exponential signal (one we haven't discussed)



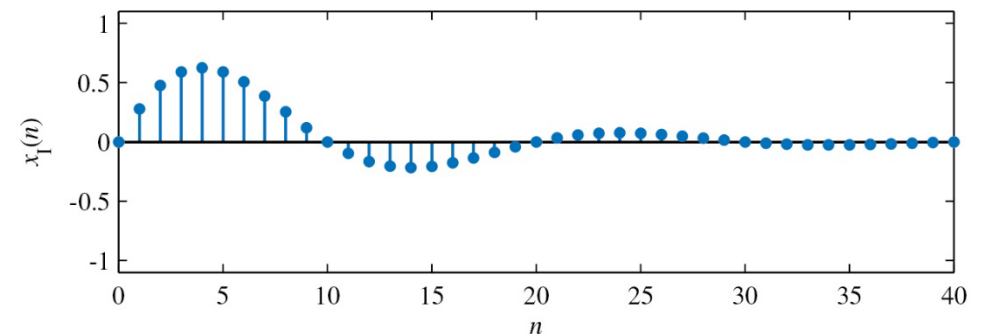
(a) Graph of  $A(n) = r^n$ ,  $r = 0.9$



(b) Graph of  $\phi(n) = \frac{\pi}{10}n$ , modulo  $2\pi$  plotted in the range  $(-\pi, \pi)$



(a)



(b)

# Discrete-Time Signals

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- Classification of signals
  - Energy signals: Finite energy in signal

$$E \equiv \sum_{n=-\infty}^{\infty} |x(n)|^2 < \infty$$

- Power signals: Energy is infinite, but power is finite

$$P \equiv \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=-N}^N |x(n)|^2 < \infty$$

# Discrete-Time Signals

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## – Classification of signals

- Power signal example:

$$x(n) = Ae^{j\omega_0 n}$$

$$P \equiv \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=-N}^N |x(n)|^2 = \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=-N}^N x(n)x^*(n)$$

$$P = \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=-N}^N Ae^{j\omega_0 n} Ae^{-j\omega_0 n} = \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=-N}^N A^2$$

$$= \lim_{N \rightarrow \infty} \frac{1}{2N+1} A^2 (2N+1)$$

$$\boxed{P = A^2}$$

# Discrete-Time Signals

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- Classification of signals
  - Periodic or Aperiodic signals

$x(n + N) = x(n)$  for periodic signal

Infinite energy but finite power

(break sum up into sums over  $N$  samples, where  $N$  is the period)

Power of periodic signal

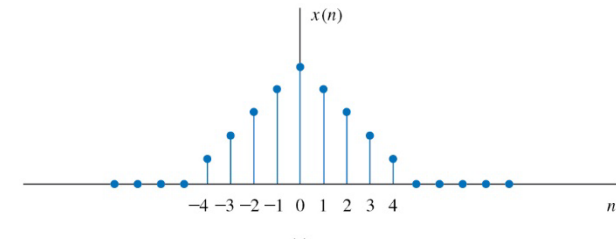
$$P \equiv \frac{1}{N} \sum_{n=0}^{N-1} |x(n)|^2$$

# Discrete-Time Signals

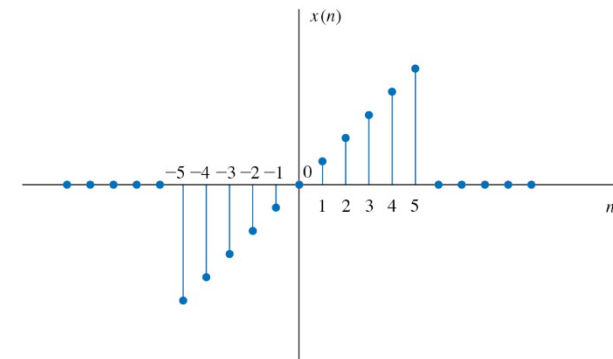
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- Classification of Signals
  - Even and Odd signals (Symmetric / Antisymmetric)

$$x(-n) = x(n) \text{ for even (symmetric)}$$



$$x(-n) = -x(n) \text{ for odd (antisymmetric)}$$



# Discrete-Time Signals

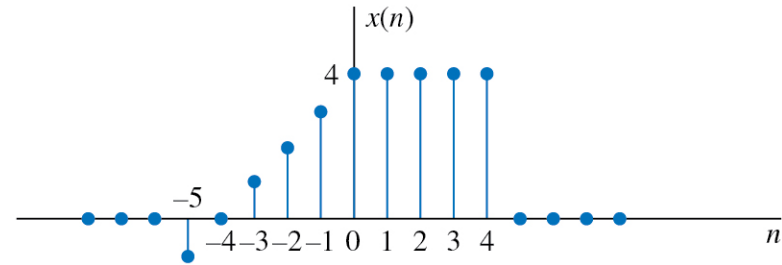
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- Operations on independent variable
  - Shifting (delay or advance)  $y(n) = x(n \pm k)$
  - Folding (reflection)  $y(n) \rightarrow x(-n)$
  - Down sampling  $y(n) \rightarrow x(2n)$

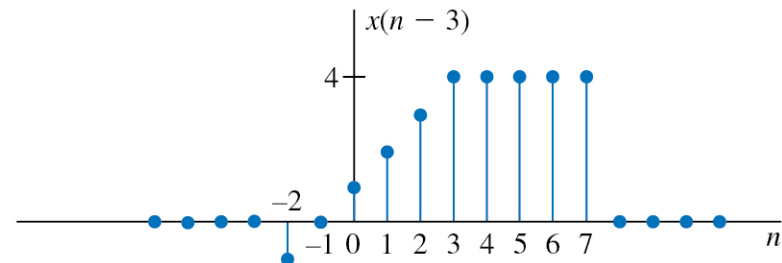
# Discrete-Time Signals

- Shifting

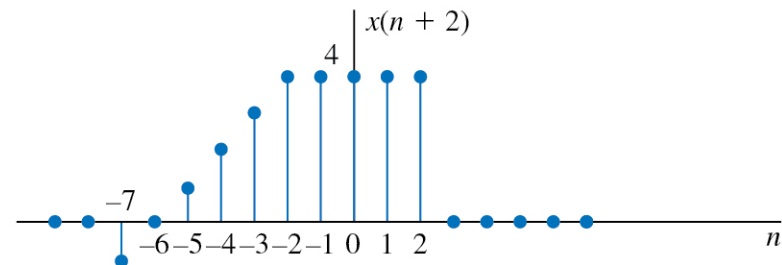
$$y(n] = x(n \pm k)$$



(a)



(b)



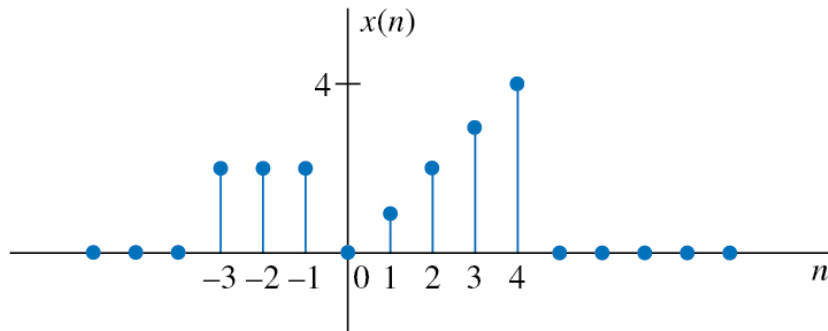
(c)



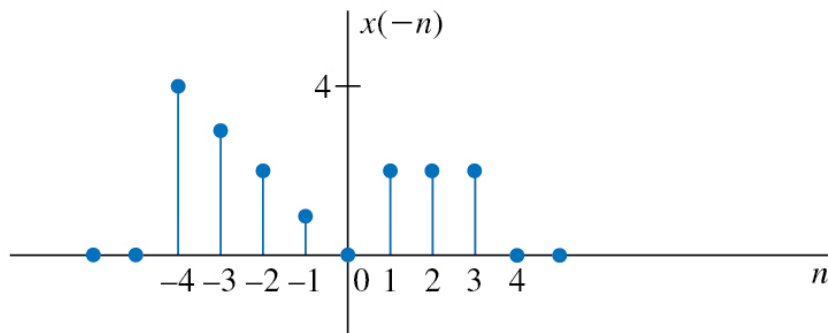
# Discrete-Time Signals

- Folding (reflection)

$$y(n) \rightarrow x(-n)$$

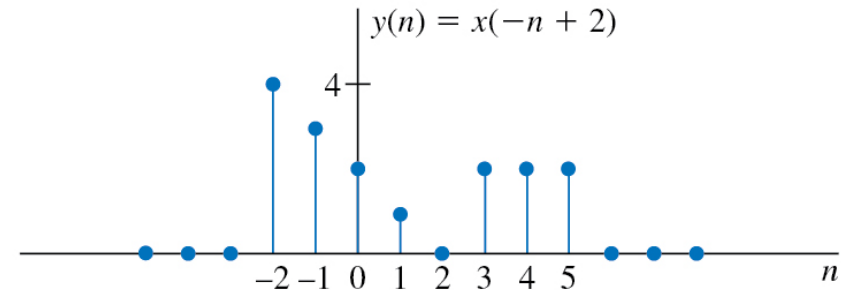


(a)



(b)

Fold and shift

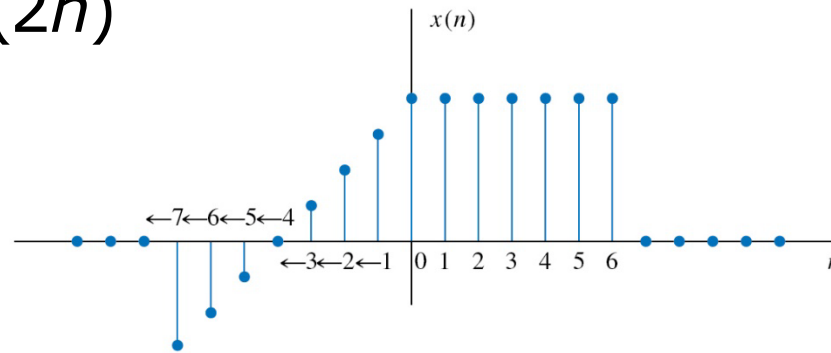


# Discrete-Time Signals

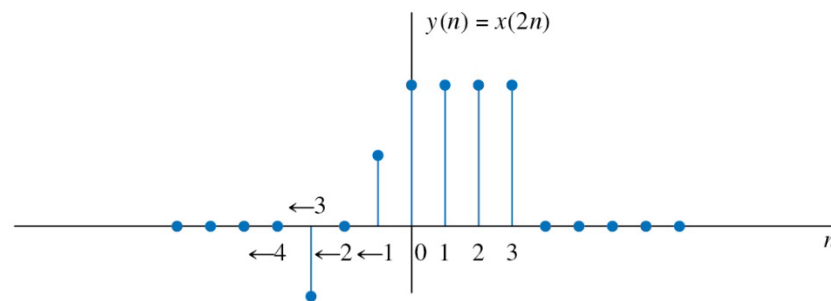
- Down sampling

If sampling, like dividing sampling rate by some integer value

$$y(n) \rightarrow x(2n)$$



(a)



(b)

This is different than continuous time case, Since you can't divide samples or multiply by non-integer values.

# Discrete-Time Signals

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## – Arithmetic operations on signals

- Addition  $y(n) = x_1(n) + x_2(n)$
- Multiplication  $y(n) = x_1(n)x_2(n)$
- Scaling  $y(n) = Ax_2(n)$

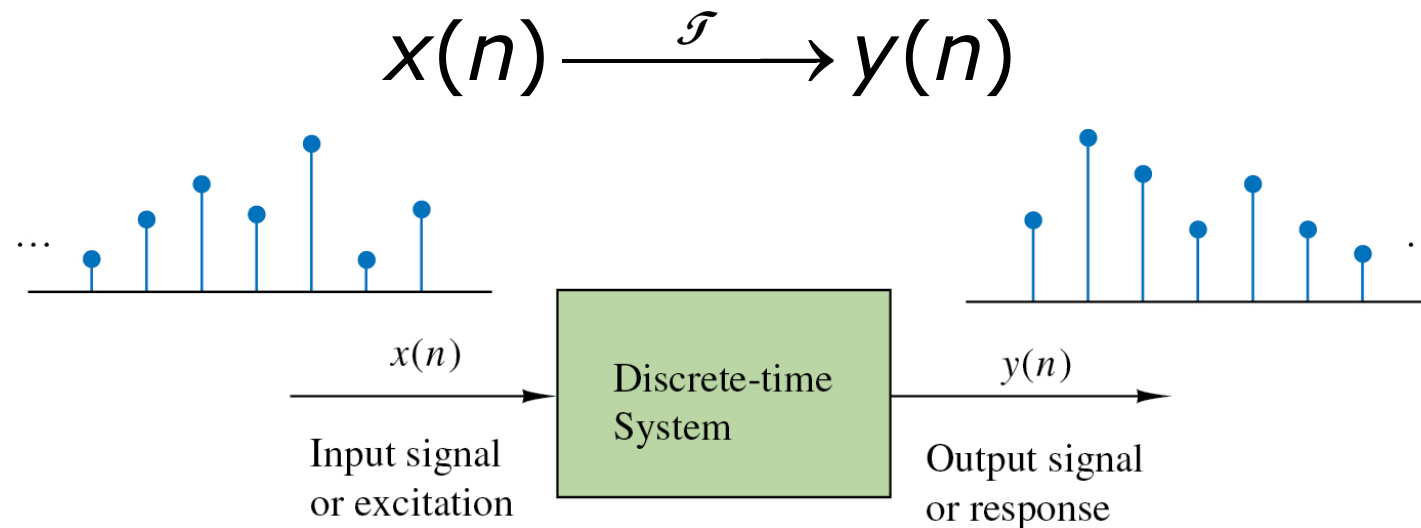
# Discrete-Time Systems

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- **Systems**

(already covered in review of continuous signals)

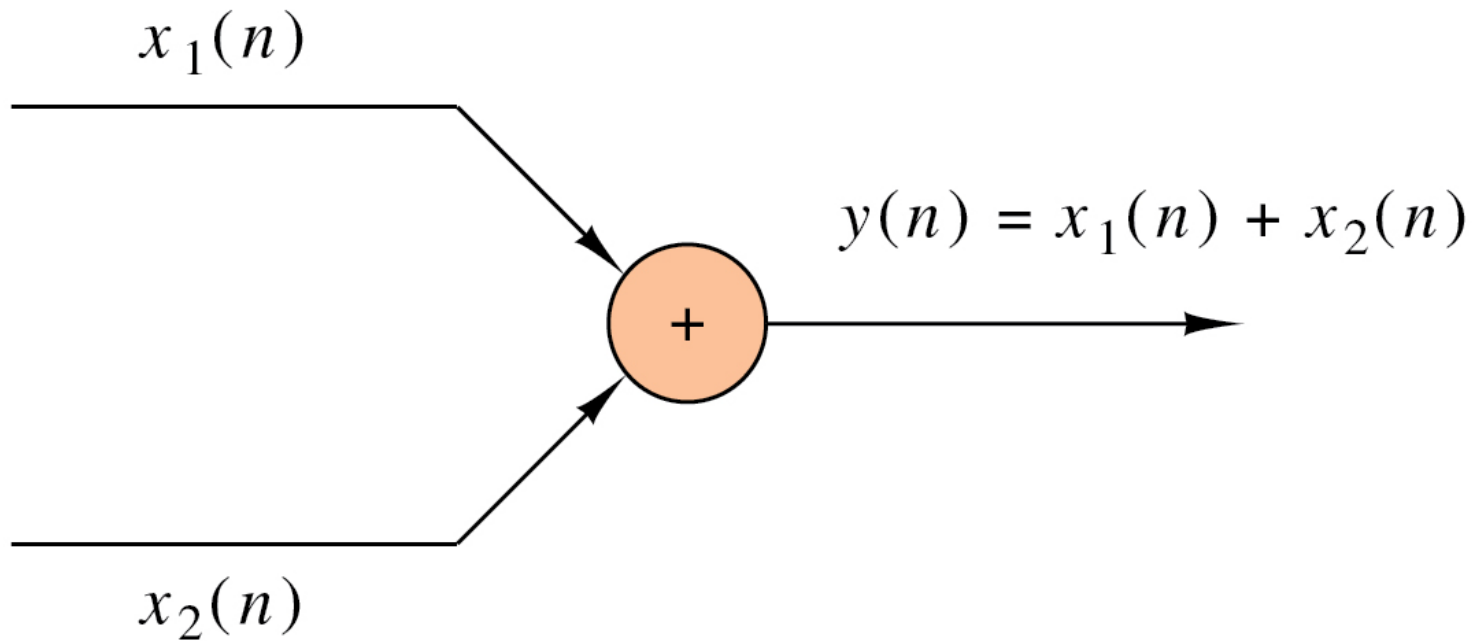
System: Transforms input signal into output signal



# Discrete-Time Systems

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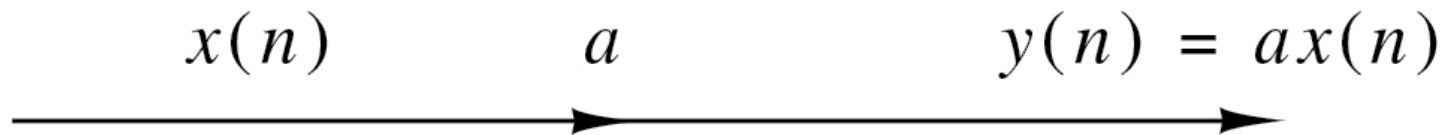
- Block diagrams for systems
  - Adder



# Discrete-Time Systems

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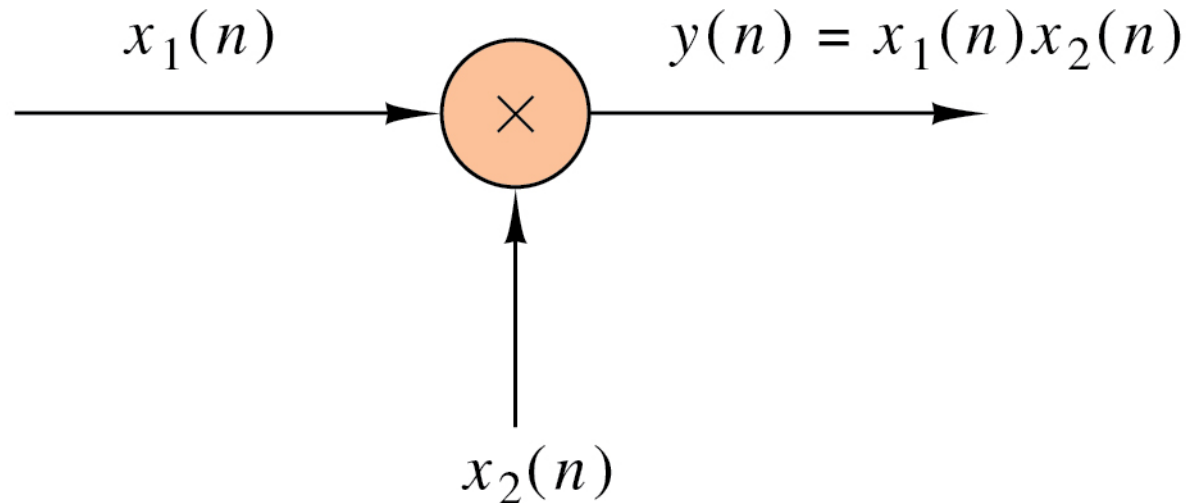
- Block diagrams for systems
  - Constant multiplier



# Discrete-Time Systems

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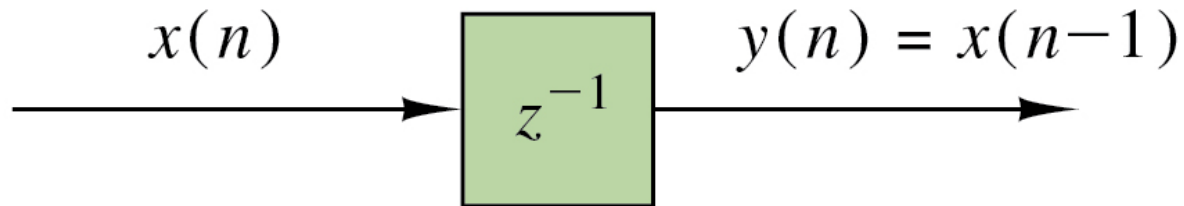
- Block diagrams for systems
  - Signal Multiplier



# Discrete-Time Systems

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- Block diagrams for systems
  - Unit delay element

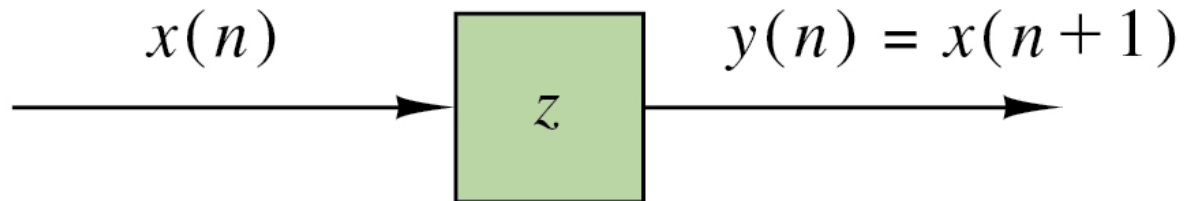




# Discrete-Time Systems

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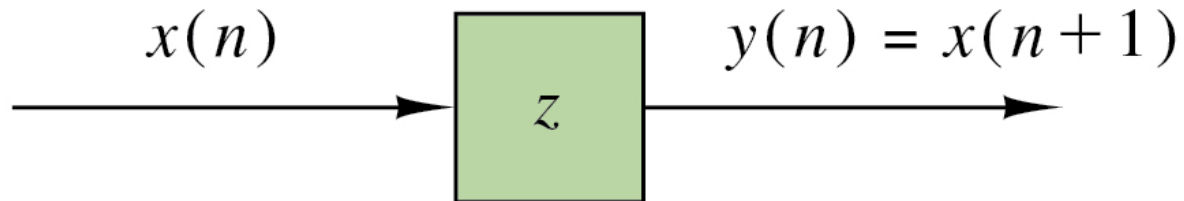
- Block diagrams for systems
  - Unit advance element



# Discrete-Time Systems

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- Block diagrams for systems
  - Unit advance element

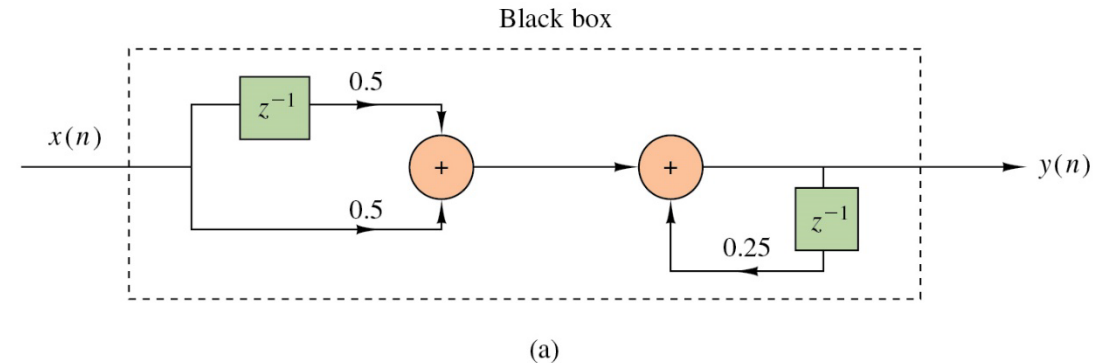


# Discrete-Time Systems

## – Building up a system from blocks

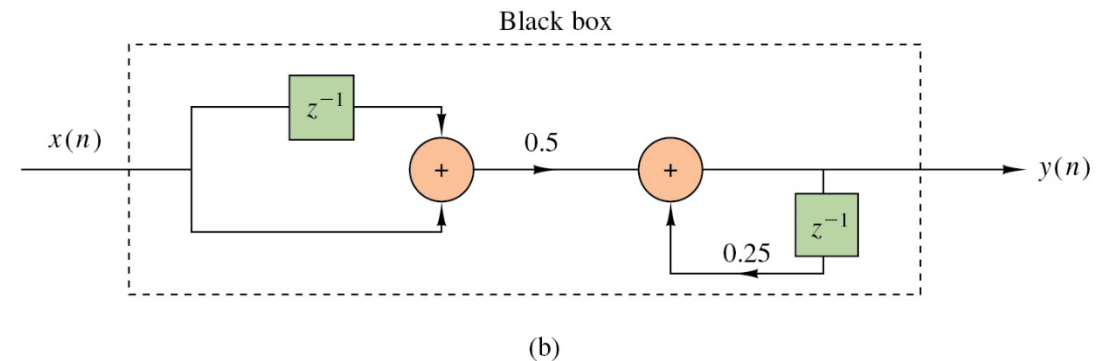
Realization 1 :

$$y(n] = \frac{1}{4} y(n - 1) + \frac{1}{2} x(n) + \frac{1}{2} x(n - 1]$$



Realization 2 :

$$y(n] = \frac{1}{4} y(n - 1) + \frac{1}{2} [x(n) + x(n - 1)]$$



# Discrete-Time Systems

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## – Classification of Systems

(already covered in review of continuous systems)

- Static versus Dynamic Systems

- Static is memoryless: output only depends on current input
- Dynamic has memory: output may depend on previous and current inputs

- Time Invariant (also known as Shift Invariant)

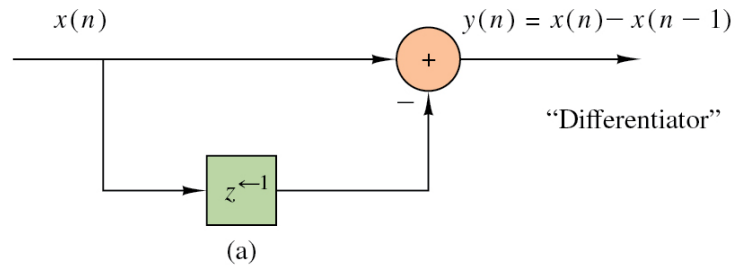
- Input-output characteristics do not change with time (or sample #)
- For “relaxed” system (i.e., no initial conditions)

$$x(n) \xrightarrow{\mathcal{T}} y(n) \quad \Rightarrow \quad x(n-k) \xrightarrow{\mathcal{T}} y(n-k)$$

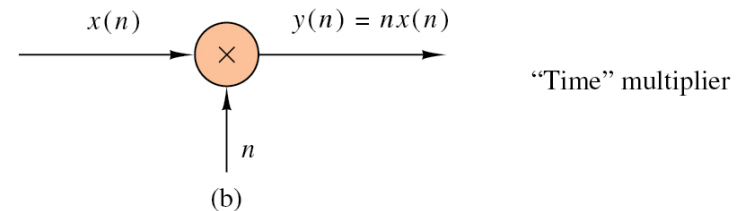
# Discrete-Time Systems

- Time Invariant/Variant examples

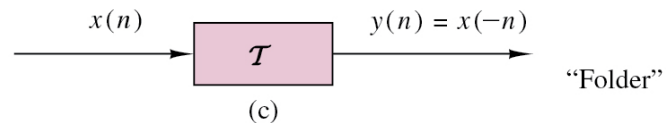
Time invariant



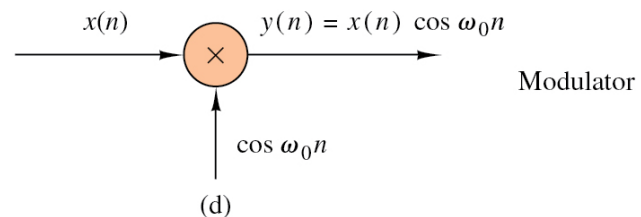
Time variant



Time variant



Time variant

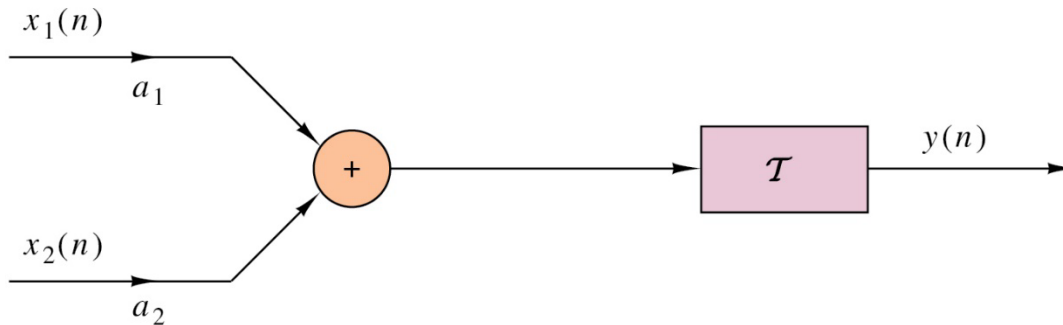


# Discrete-Time Systems

- Linear versus Non-Linear Systems

- Linear:

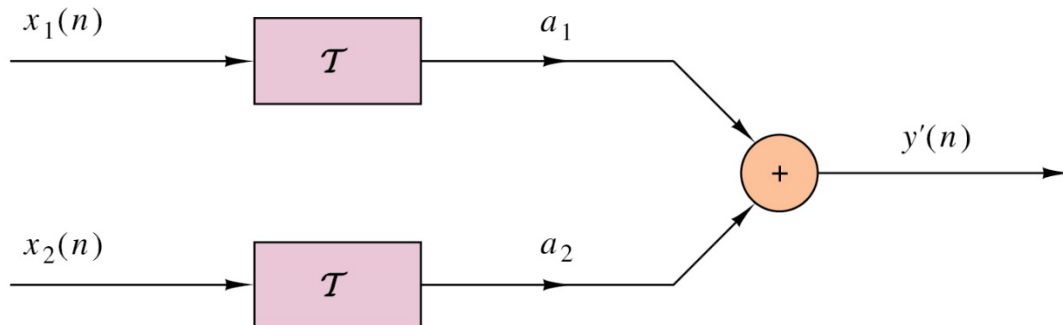
$$S[a_1x_1(t) + a_2x_2(t)] = a_1S[x_1(t)] + a_2S[x_2(t)]$$



Works for multiple sums

$$S\left[x(n) = \sum_{k=1}^{M-1} a_k x_k(t)\right] = y(n) = \sum_{k=1}^{M-1} a_k y_k(t)$$

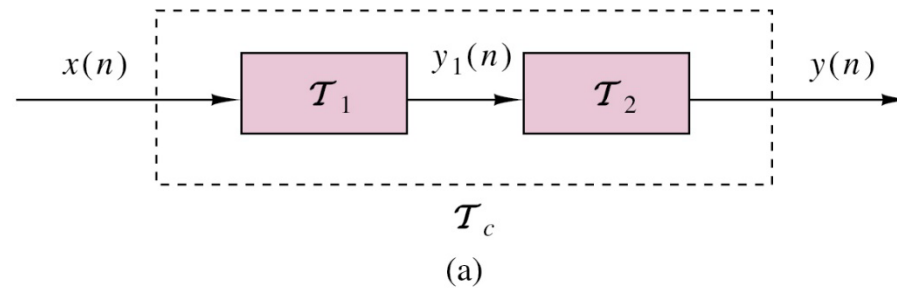
Principle of Superposition



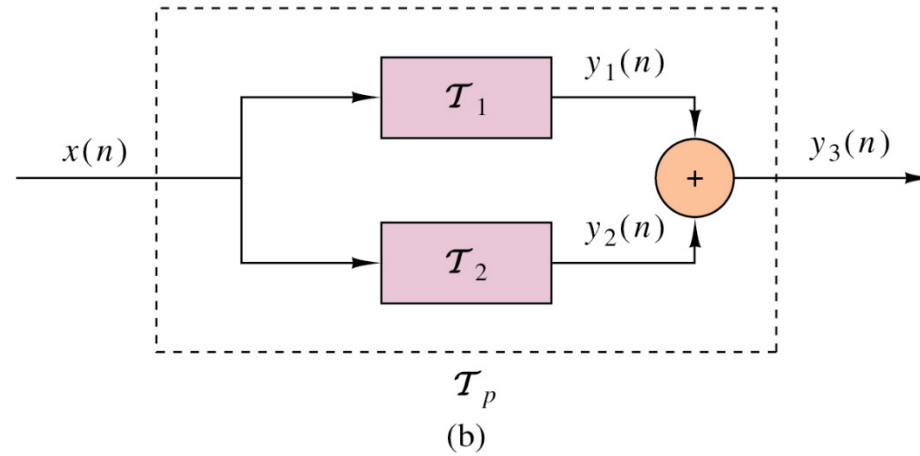
# Discrete-Time Systems

- Interconnections of systems

Series (cascade)



Parallel



# Discrete-Time Systems

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## – Analysis of Linear Systems

- Two methods for analyzing linear systems

- Direct method: (Think “time-domain”)

- Solve equation describing system for  $y(n)$ :

$$y(n) = F[y(n-1), y(n-2), \dots, y(n-N), x(n), x(n-1), \dots, x(n-M)]$$

- Transform method: (Think “frequency-domain”)

- We will discuss z-transform and discrete Fourier transform later



# Discrete-Time Systems

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## – Analysis of Linear Systems

### • Direct Method

- The most general form for Linear Time Invariant (LTI) systems is a difference equation:

$$y(n) = -\sum_{k=1}^N a_k y(n-k) + \sum_{k=1}^M b_k x(n-k)$$

- Can be solved by decomposing input into weighted sum of elementary signals, where the response to the elementary signals is known. The use superposition

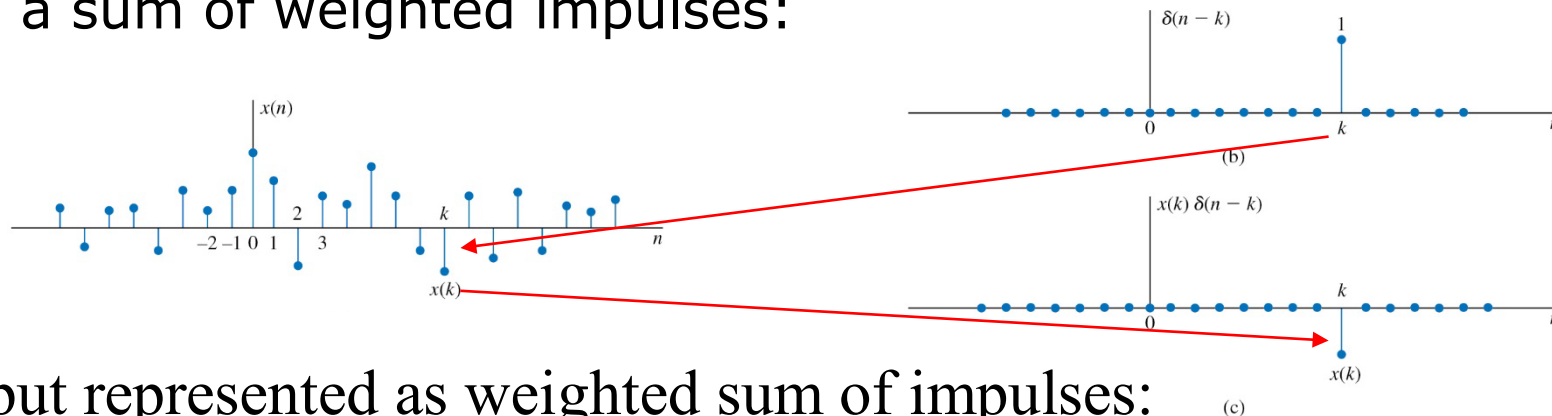
$$x(n) = \sum_k c_k x_k(n) \quad y_k(n) = S[x_k(n)] \quad y = \sum_k c_k y_k(n)$$

# Discrete-Time Systems

- Impulse response:
  - Select elementary functions to be delayed unit impulses

$$x_k(n) = \delta(n - k)$$

- As we've discussed before, any discrete signal can be decomposed into a sum of weighted impulses:



Input represented as weighted sum of impulses:

$$x(n) = \sum_{k=-\infty}^{\infty} x(k)\delta(n - k)$$

# Discrete-Time Systems

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- Impulse response for linear system:

- Linearity:

$$y(n) = S[x(n)] = S\left[\sum_{k=-\infty}^{\infty} x(k)\delta(n-k)\right] = \sum_{k=-\infty}^{\infty} x(k)S[\delta(n-k)]$$

- Output for a system with no initial conditions (“relaxed”)
- In general  $h(n,k) \equiv S[\delta(n-k)]$  could change with sample number if system is not time invariant.
- For time invariant system:

- Linear Time Invariant System:  $h(n-k) \equiv S[\delta(n-k)]$

$$y(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k) \quad \text{where} \quad h(n) = S[\delta(n)] \text{ is impulse response}$$

Discrete Convolution

# Discrete-Time Systems

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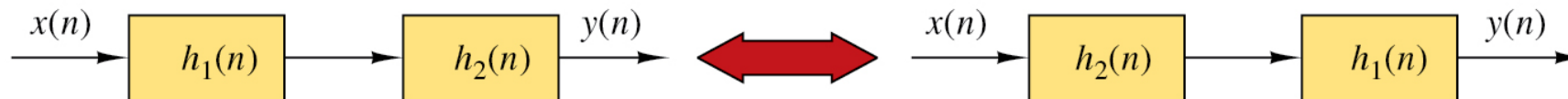
– Properties of Convolution:

- Commutivity:  $x(n) * h(n) = h(n) * x(n)$

$$y(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$

$$y(n) = \sum_{k=-\infty}^{\infty} h(k)x(n-k)$$

$$h_1(n) * h_2(n) = h_2(n) * h_1(n)$$

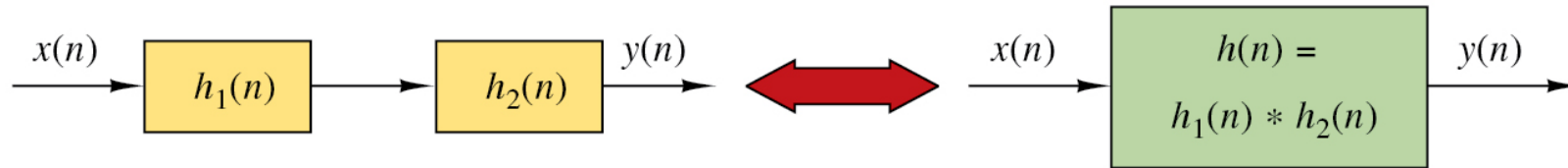


# Discrete-Time Systems

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- Properties of Convolution:
  - Associativity

$$[x(n) * h_1(n)] * h_2(n) = x(n) * [h_1(n) * h_2(n)]$$

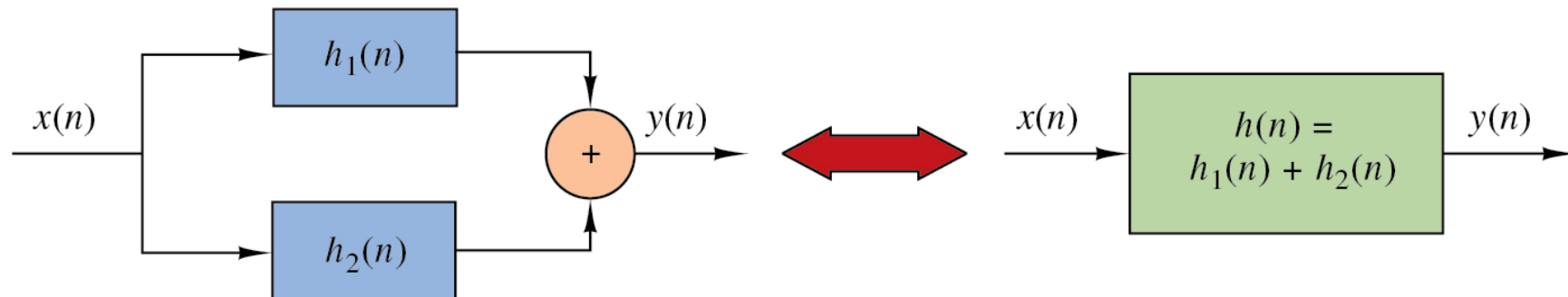


# Discrete-Time Systems

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- Properties of Convolution:
  - Distributivity

$$x(n) * [h_1(n) + h_2(n)] = x(n) * h_1(n) + x(n) * h_2(n)$$



# Discrete-Time Systems

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- Causal LTI Systems:

$$h(n) = 0 \text{ for } n < 0$$

$$y(n) = \sum_{k=0}^{\infty} h(k)x(n-k)$$

# Discrete-Time Systems

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- Stable LTI systems
  - For bounded input to yield bounded output

$$|x(n)| \leq M_x$$

$$|y(n)| \leq M_y$$

$$\sum_{k=-\infty}^{\infty} |h(k)| < \infty$$

Also, impulse response must go to zero as  $n \rightarrow \infty$



# Discrete-Time Systems

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- Stable LTI systems
  - Example of conditions for stability
    - Consider the following causal impulse response:

$$h(n) = a^n u(n)$$

$$\sum_{k=-\infty}^{\infty} |h(k)| < \infty \Rightarrow \sum_{k=-\infty}^{\infty} |a^n u(n)| = \sum_{k=0}^{\infty} |a^n| < \infty$$

Geometric series:  $\sum_{k=0}^{\infty} |a^n| = \frac{1}{1-|a|}$  if  $|a| < 1$

$$\text{If } |a| \geq 1, \quad \sum_{k=0}^{\infty} |a^n| \rightarrow \infty$$

$\therefore$  System with  $h(n) = a^n u(n)$  is stable for  $|a| < 1$ , unstable for  $|a| \geq 1$

# Discrete-Time Systems

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- Finite Impulse Response systems (FIR)

- For causal system,

$$h(n) = 0, \quad n < 0 \text{ and } n \geq M$$

is an FIR system

$$y(n) = \sum_{k=0}^{M-1} h(k)x(n-k)$$

- This is a weighted sum of input values

$$x(n), x(n-1), \dots, x(n-M+2), x(n-M+1)$$

- Finite memory length

# Discrete-Time Systems

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- Infinite Impulse Response systems (IIR)
  - For causal system,  $h(n)$  persists for  $n$  going to infinity

$$y(n) = \sum_{k=0}^{\infty} h(k)x(n-k)$$

- Has infinite memory. Response depends on all previous inputs
  - Results from system with recursion where current output depends on previous outputs.

# Discrete-Time Systems

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- LTI systems characterized by constant-coefficient difference equations

$$y(n) + \sum_{k=1}^N a_k y(n-k) = \sum_{k=0}^M b_k x(n-k)$$

$$y(n) = -\sum_{k=1}^N a_k y(n-k) + \sum_{k=0}^M b_k x(n-k)$$

$N$ 'th order difference equation

$N$ 'th order system

# Discrete-Time Systems

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- Compare this to continuous system

$$\frac{d^n y}{dt^n} + a_{n-1} \frac{d^{n-1} y}{dt^{n-1}} + a_{n-2} \frac{d^{n-2} y}{dt^{n-2}} + \cdots + a_1 \frac{dy}{dt} + a_0 y =$$
$$b_m \frac{d^m x}{dt^m} + b_{m-1} \frac{d^{m-1} x}{dt^{m-1}} + b_{m-2} \frac{d^{m-2} x}{dt^{m-2}} + \cdots + b_1 \frac{dx}{dt} + b_0 x$$

# Discrete-Time Systems

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- If output only depends on previous and current input

$$y(n) = \sum_{k=0}^M b_k x(n-k)$$

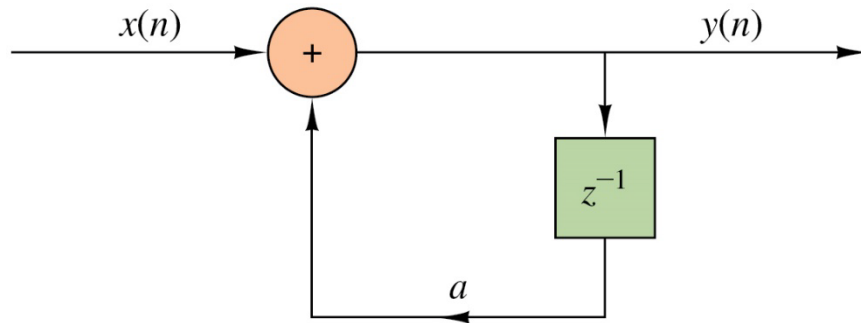
System will be FIR

# Discrete-Time Systems

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- Example of a simple recursive difference equation
  - First order autoregressive model

$$y(n) = ay(n-1) + x(n)$$



# Discrete-Time Systems

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- First order autoregressive model

$$y(n) = ay(n-1) + x(n)$$

$$y(0) = ay(-1) + x(0)$$

$$y(1) = ay(0) + x(1) = a(ay(-1) + x(0)) + x(1) = a^2y(-1) + ax(0) + x(1)$$

$$y(2) = ay(1) + x(2) = a(a^2y(-1) + ax(0) + x(1)) + x(2) = a^3y(-1) + a^2x(0) + ax(1) + x(2)$$

⋮

$$y(n) = a^{n+1}y(-1) + \sum_{k=0}^n a^k x(n-k)$$

Depends on initial condition  $y(-1)$



# Discrete-Time Systems

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- Same drill as before
  - If the system is relaxed (no initial condition)

Zero-state solution

$$y_{ZS}(n) = \sum_{k=0}^n a^k x(n-k)$$

- If the system has no input
- Zero-input solution

$$y_{ZI}(n) = a^{n+1} y(-1)$$

# Discrete-Time Systems

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- Notice the impulse response of this system

$$h(0) = ay(-1) + \delta(0) = 1 \quad (\text{with no initial condition})$$

$$h(1) = ay(0) + \delta(1) = a$$

$$h(2) = ay(1) + \delta(2) = a^2$$

$\vdots$

$$h(n) = a^n u(n)$$

$$y_{ZS}(n) = \sum_{k=0}^n a^k x(n-k) = \sum_{k=0}^n h(k)x(n-k)$$

Zero-state response is convolution of input with impulse response

# Discrete-Time Systems

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- Don't forget about geometric series

$$\sum_{k=0}^n r^k = \frac{1 - r^{n+1}}{1 - r}$$

$$\sum_{k=1}^n r^k = \frac{r(1 - r^n)}{1 - r}$$

$$\sum_{k=0}^{\infty} r^k = \frac{1}{1 - r}$$

for  $|r| < 1$

$$\sum_{k=1}^{\infty} r^k = \frac{r}{1 - r}$$

for  $|r| < 1$

# Discrete-Time Systems

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## – Difference equations

Can be solved in similar manner as differential equations

$$y(n) - ay(n-1) = x(n)$$

Guess solution of homogeneous equation is of the form:  $\lambda^n$

$$y(n) - ay(n-1) = 0$$

$$\lambda^n - a\lambda^{n-1} = 0$$

$$\lambda - a = 0 \quad (\text{Characteristic equation})$$

$$\lambda = a$$

$$y_h(n) = Ca^n$$

Find C from initial condition for impulse:  $x(n) = \delta(n)$

$$y(0) = 1 \Rightarrow 1 = Ca^0 \Rightarrow C = 1$$

Impulse response is:

$$h(n) = a^n$$

Step response is:

$$s(n) = \sum_{k=-\infty}^{\infty} h(k)u(n-k)$$

$$s(n) = \sum_{k=0}^n h(k)u(n-k)$$

$$s(n) = \sum_{k=0}^n h(k) = \sum_{k=0}^n a^k = \frac{1 - a^{n+1}}{1 - a}$$

# Discrete-Time Systems

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## – Difference equations

Can be solved in similar manner as differential equations

$$y(n) + a_1 y(n-1) + a_2 y(n-2) = x(n)$$

Guess solution of homogeneous equation is of the form:  $\lambda^n$

$$\lambda^n + a_1 \lambda^{n-1} + a_2 \lambda^{n-2} = 0$$

$$\lambda^2 + a_1 \lambda + a_2 = 0 \quad (\text{Characteristic equation})$$

$$\lambda_{(1,2)} = \frac{-a_1 \pm \sqrt{a_1^2 - 4a_2}}{2}$$

$$y_h(n) = C_1 \lambda_1^n + C_2 \lambda_2^n$$

Find  $C_1$  and  $C_2$  from initial conditions for impulse:  $x(n) = \delta(n)$

Solve for impulse response, solve for step response.

# Discrete-Time Systems

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– Example:

- Determine the impulse and step response for the following second-order discrete system:

$$y(n) = 0.6y(n-1) - 0.08y(n-2) + x(n)$$