

Electromagnetic Waves:

23/03/2022

① Gauss law:

$$\oint \vec{D} \cdot d\vec{s} = Q_{enc} \longrightarrow \nabla \cdot \vec{D} = \rho_v$$
$$= \int_V \rho_v dV$$

$$\oint \vec{B} \cdot d\vec{s} = 0 \longrightarrow \nabla \cdot \vec{B} = 0$$

② Ampere's law:

$$\oint \vec{H} \cdot d\vec{l} = I_{enc} \quad \left(\vec{B} = \mu \vec{H} \right)$$
$$= \int_S \left(\vec{J} + \frac{\partial \vec{D}}{\partial t} \right) \cdot d\vec{s} \quad \left(\vec{B} = \mu_0 \mu_r \vec{H} \right)$$

$$\nabla \times \vec{H} = \vec{J}_c + \vec{J}_D$$

$$\nabla \times \vec{H} = \vec{J}_c + \frac{\partial \vec{D}}{\partial t}$$

③ Faraday's law:

$$\nabla \times \vec{E} = -\frac{\partial \vec{B}}{\partial t} \quad \left(\vec{D} = \epsilon \vec{E} \right)$$

$$\oint \vec{E} \cdot d\vec{l} = -\int \frac{\partial \vec{B}}{\partial t} \cdot d\vec{s} \quad \left(\vec{D} = \epsilon_0 \epsilon_r \vec{E} \right)$$

$$= -\frac{\partial}{\partial t} \int \vec{B} \cdot d\vec{s}$$

$$\oint \vec{E} \cdot d\vec{l} = -\frac{\partial \Psi_m}{\partial t}$$

→ Harmonics

$$\vec{E} = \vec{E}_s \cdot e^{j\omega t}$$

$$\vec{D} = \vec{D}_s \cdot e^{j\omega t}$$

$$\vec{H} = \vec{H}_s \cdot e^{j\omega t}$$

$$\vec{B} = \vec{B}_s \cdot e^{j\omega t}$$

$$\rho_v = (\rho_v)_s \cdot e^{j\omega t}$$

From the above laws,

$$(i) \quad \nabla \cdot \vec{D}_s = \rho_{vs}$$

$$(ii) \quad \nabla \cdot \vec{B}_s = 0$$

$$(iii) \quad \nabla \times \vec{H}_s = \sigma \vec{E}_s + j\omega \vec{D}_s$$

$$(iv) \quad \nabla \times \vec{E}_s = -j\omega \vec{B}_s$$

Removed
the time dependance
in the equations.

⇒ for a source-free medium,

$$\hookrightarrow \boxed{\rho_v = 0, \sigma \vec{E} = 0}$$

then, (i) $\nabla \cdot \vec{D}_s = 0$

(ii) $\nabla \cdot \vec{B}_s = 0$

(iii) $\nabla \times \vec{H}_s = j\omega \vec{D}_s = j\omega \epsilon \vec{E}_s$

(iv) $\nabla \times \vec{E}_s = -j\omega \vec{B}_s = -j\omega \mu \vec{H}_s$

• Homogeneous → μ, ϵ are constant

• Isotropic → μ, ϵ are not functions of direction.

①

$$\nabla \times \vec{E} = -j\omega\mu\vec{H}$$

$$\nabla \times (\nabla \times \vec{E}) = -j\omega\mu (\nabla \times \vec{H})$$

$$\nabla (\nabla \cdot (\vec{E})) - \nabla^2 \vec{E} = -j\omega\mu (\nabla \times \vec{H})$$

$$\nabla (\nabla \cdot \vec{E}) - \nabla^2 \vec{E} = -j\omega\mu (j\omega\epsilon \vec{E})$$

0 (since it is source-free)

$$\therefore \boxed{\nabla^2 \vec{E} = -\omega^2 \mu \epsilon \vec{E}}$$

②

$$\nabla \times (\nabla \times \vec{H}) = j\omega\epsilon (\nabla \times \vec{E})$$

$$\nabla (\nabla \cdot \vec{H}) - \nabla^2 \vec{H} = j\omega\epsilon (-j\omega\mu\vec{H})$$

$$\therefore \boxed{\nabla^2 \vec{H} = -\omega^2 \mu \epsilon \vec{H}}$$

Wave: Any physical quantity that varies with time and space is called as wave.

$$\frac{d^2 V}{dx^2} = \frac{1}{u^2} \frac{d^2 V}{dt^2}$$

$$V = V^+ e^{j(\omega t - \beta x)} + V^- e^{j(\omega t + \beta x)}$$

$$(V = V_s e^{j\omega t})$$

$$\frac{d^2 V_s}{dx^2} = -\frac{\omega^2}{u^2} \cdot V_s$$

$$\frac{d^2 V_s}{dx^2} + \frac{\omega^2}{u^2} V_s = 0$$

$$\frac{d^2 V_s}{dx^2} + \beta^2 V_s = 0$$

$$\boxed{V_s = V^+ e^{-j\beta x} + V^- e^{j\beta x}}$$

$$\frac{\omega}{u} = \beta$$

$u \rightarrow$ velocity of wave

$\beta \rightarrow$ phase constant

$$u = \omega / \beta$$

$$\beta = 2\pi / \lambda$$

\vec{H}, \vec{E} are oriented in one direction,

$$\vec{E} = E_x \hat{x} \quad (E_y = E_z = 0)$$

$$\vec{H} = H_y \hat{y} \quad (H_x = H_z = 0)$$

\vec{E} is varying only in the direction of propagation

$$\therefore \left(\frac{\partial}{\partial x} = \frac{\partial}{\partial y} = 0 \right)$$

$$\rightarrow \nabla^2 E_x = -\omega^2 \mu \epsilon E_x$$

$$\therefore \left[\frac{\partial^2}{\partial z^2} E_x = -\omega^2 \mu \epsilon E_x \right]$$

$$(\beta^2 = \omega^2 \mu \epsilon)$$

$$\frac{d^2 E_x}{dz^2} + \beta^2 E_x = 0$$

$$\therefore E_x = E_x^+ e^{-j\beta z} + E_x^- e^{j\beta z}$$

$$E_x \cdot e^{j\omega t} = E_x^+ e^{j(\omega t - \beta z)} + E_x^- e^{j(\omega t + \beta z)}$$

$$\nabla \times \vec{E} = -j\omega\mu\vec{H}$$

$$\begin{vmatrix} \hat{x} & \hat{y} & \hat{z} \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ E_x & E_y & E_z \end{vmatrix} = -\hat{y} \left(-\frac{\partial}{\partial z} E_x \right)$$

$$\Rightarrow \frac{\partial E_x}{\partial z} \hat{y} = -j\omega\mu\vec{H}$$

$$\begin{aligned} \frac{\partial E_x}{\partial z} \hat{y} &= -j\omega\mu (\cancel{H_x \hat{x}} + H_y \hat{y} + \cancel{H_z \hat{z}}) \\ &= -j\omega\mu H_y \hat{y} \end{aligned}$$

$$\boxed{\frac{\partial E_x}{\partial z} = -j\omega\mu H_y}$$

$$\begin{aligned} \frac{\partial}{\partial z} E_x &= -j\beta E_x^+ (e^{j\beta z}) + j\beta E_x^- (e^{j\beta z}) \quad \left. \vphantom{\frac{\partial}{\partial z} E_x} \right\} \text{--- (1)} \\ &= -j\omega\mu H_y \end{aligned}$$

$$H_y = \frac{+j\beta}{+j\omega\mu} E_x^+ e^{-j\beta z} + \frac{-j\beta}{+j\omega\mu} E_x^- e^{j\beta z}$$

$$\begin{aligned} H_y &= H_y^+ e^{-j\beta z} - H_y^- e^{j\beta z} \quad \left. \vphantom{H_y} \right\} \text{--- (2)} \\ &= \frac{\beta}{\omega\mu} E_x^+ e^{-j\beta z} - \frac{\beta}{\omega\mu} E_x^- e^{j\beta z} \end{aligned}$$

From ① & ②

$$\frac{E_x^+}{H_y^+} = \frac{\omega \mu}{\beta} = \eta_0$$

intrinsic impedance of the medium.

$$\frac{E_x^-}{H_y^-} = -\frac{\omega \mu}{\beta} = -\eta_0$$

$$\text{Velocity } (u) = \frac{1}{\sqrt{\mu \epsilon}}$$

$$u = \frac{\omega}{\beta} \rightarrow \beta = \omega \sqrt{\mu \epsilon}$$

$$\eta_0 = \frac{\omega \mu}{\beta} = \frac{\omega \mu}{\omega \sqrt{\mu \epsilon}} = \sqrt{\frac{\mu}{\epsilon}}$$

$$\therefore \eta_0 = \sqrt{\frac{\mu}{\epsilon}}$$

for a free space,

$$\eta_0 = 377 \Omega$$

$$= 120\pi \Omega$$

$$\begin{aligned} \hat{x} \times \hat{y} &= \hat{z} \\ \hat{y} \times \hat{z} &= \hat{x} \\ \hat{z} \times \hat{x} &= \hat{y} \end{aligned}$$

$$\vec{E} \times \vec{H} = \text{direction}$$

The above waves are called
Transverse - electromagnetic wave (TEM wave)

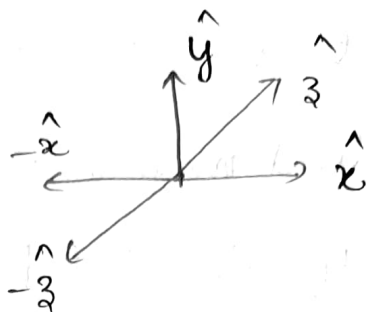
→ \vec{E} is in y-direction and wave is in z-direction

$$\vec{E} = E_y^+ e^{j(\omega t - \beta z)} \hat{y}$$

$$\vec{H} = -H_x^+ e^{j(\omega t - \beta z)} \hat{x}$$

$$\vec{E} = E_{y1}^+ e^{j(\omega t + \beta z)} \hat{y}$$

$$\vec{H} = H_{x1}^+ e^{j(\omega t + \beta z)} \hat{x}$$

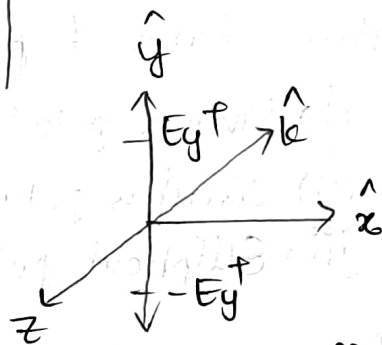


$$\boxed{\frac{E_y^+}{-H_x^+} = \eta_0 = \frac{E_{y1}^+}{H_{x1}^+}}$$

$$\vec{E}_x = E_y^+ \cos(\omega t - \beta z) \hat{y}$$

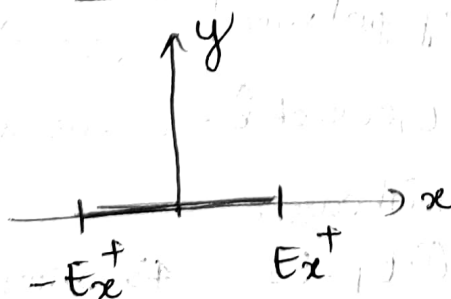
if $z = z_1 = 0$

$\vec{E} = E_y^+ \cos(\omega t) \hat{y} \rightarrow$ locus of electric field in a given plane \perp to direction of propagation is said to be polarisation.



$$E_x = E_x^+ \cos(\omega t - \beta z)$$

for $z=0$ plane,



Generalised electric field expression is,

$$\vec{E} = E_1 \cos(\omega t - \beta z) \hat{x} + E_2 \cos(\omega t - \beta z + \delta) \hat{y}.$$

Generalised magnetic field expression is,

$$\vec{H} = H_1 \cos(\omega t - \beta z) \hat{y} - H_2 \cos(\omega t - \beta z + \delta) \hat{x}.$$

$$\text{So, } \frac{E_1}{H_1} = \eta_0 = \frac{E_2}{H_2}$$

→ Three states of polarisation are:

- (i) Linear polarisation.
- (ii) Circular polarisation.
- (iii) Elliptical polarisation.

→ Polarisation: Locus of \vec{E} in a plane perpendicular to the direction of propagation.

- The states of polarisation is differed by amplitude and phase.

(i) Linear Polarisation:

If locus of \vec{E} is in a straight line then it is said to be linearly polarised. ($\delta = 0$)

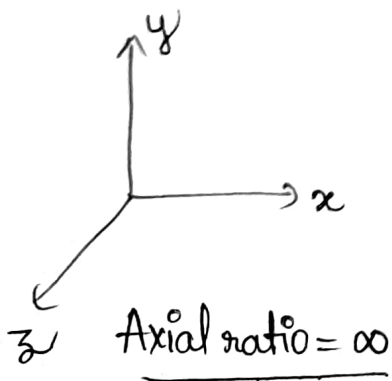
$$\vec{E} = E_1 \cos \omega t \hat{x} + E_2 \cos \omega t \hat{y}$$

Special cases:

① $E_1 = E_2$ 45° oriented.

② If $E_1 = 0$, vertical polarised wave.

③ If $E_2 = 0$, Horizontal polarised wave.



(i) Circular polarisation:

① $E_1 = E_2$ and $\delta = \pm \pi/2$

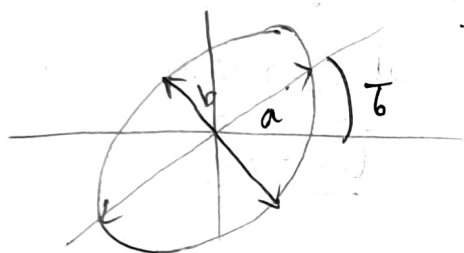


Axial ratio = 1

$0 \leq \tau \leq \pi$

(ii) Elliptical polarisation:

① $E_1 \neq E_2$ and $\delta = \pm \text{angle}$



Axial ratio = a/b

$\tau = \tan^{-1}(E_2/E_1)$

eg: $\vec{E} = 10(\hat{a}_y + j\hat{a}_z) e^{-j(25)x}$

Determine the state of polarisation.

Sol:- $\vec{E} \cdot e^{j\omega t} = 10(\hat{a}_y + j\hat{a}_z) e^{-j(25)x} e^{j\omega t}$

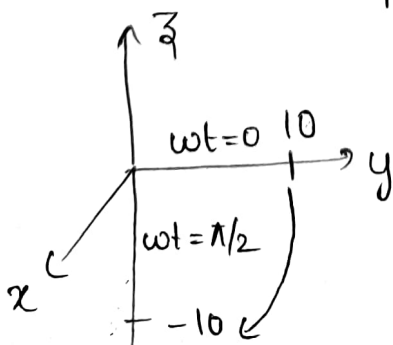
$= 10 \cos(\omega t - 25x) \hat{a}_y + 10 \cos(\omega t - 25x + \pi/2) \hat{a}_z$

if $x=0$, yz -plane

$\vec{E} = 10 \cos(\omega t) \hat{a}_y + 10 \cos(\omega t + \pi/2) \hat{a}_z$

$\omega t = 0 \rightarrow \vec{E} = 10 \hat{a}_y$

$\omega t = \pi/2 \rightarrow \vec{E} = -10 \hat{a}_z$



Right-Hand
Circular polarisation
(R.H.C.P)

eg: $\vec{E} = 5 \cos(2\pi 10^9 t + \beta z) \hat{a}_x + 3 \cos(2\pi 10^9 t + \beta z - \pi/2) \hat{a}_y$

Since $E_1 \neq E_2$
it is Elliptically polarised

if $z=0 \rightarrow XY$ plane

$$\vec{E} = 5 \cos(2\pi 10^9 t) \hat{a}_x + 3 \cos(2\pi 10^9 t - \pi/2) \hat{a}_y$$

at $t_1 = 0$

$$\vec{E} = 5 \hat{a}_x$$

at $t_2 \rightarrow 2\pi 10^9 t = \pi/2$

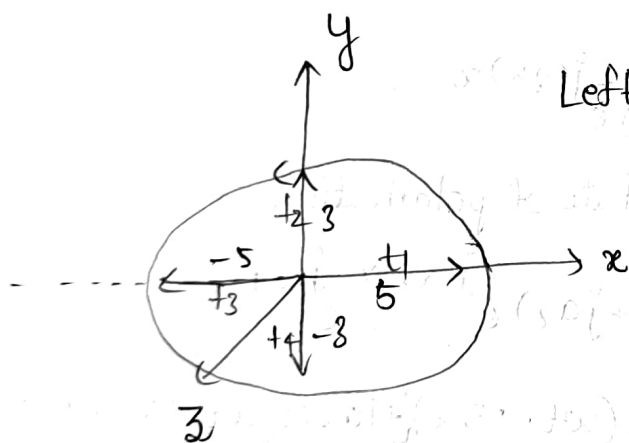
$$\vec{E} = 3 \hat{a}_y$$

at $t_3 \rightarrow 2\pi 10^9 t = \pi$

$$\vec{E} = -5 \hat{a}_x$$

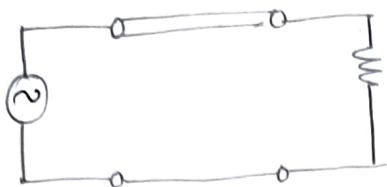
at $t_4 \rightarrow 2\pi 10^9 t = 3\pi/2$

$$\vec{E} = -3 \hat{a}_y$$



Left-Hand Elliptical
polarisation
(L.H.E.P)

Magnetic field - current



→ for the magnetic field to flow in a medium, the medium need not be highly conductive.

$$\nabla \times \vec{H} = \vec{J}_c + \frac{\partial \vec{D}}{\partial t}$$

→ for the current wave to propagate in an electrical circuit, conductivity is essential.

→ EM wave can't propagate in a highly conductive medium (it reflects from the walls of conductive materials).

$$\nabla \times \vec{H} = \vec{J}_c + \vec{J}_D$$

$$= \vec{J}_c + \frac{\partial \vec{D}}{\partial t}$$

$$\nabla \times \vec{H} = \sigma \vec{E} + j\omega \epsilon \vec{E}$$

$$= (\sigma + j\omega \epsilon) \vec{E}$$

$\epsilon \rightarrow$ permittivity of medium, $\epsilon = \epsilon_0 \epsilon_r$.

$$\nabla \times \vec{H} = j\omega \epsilon_0 \left(\epsilon_r - \frac{j\sigma}{\omega \epsilon_0} \right) \vec{E}$$

$\left(\epsilon_r - \frac{j\sigma}{\omega \epsilon_0} \right) \rightarrow$ complex dielectric constant.

→ for a highly, perfect, dielectric, imaginary part = 0

$$\left(\frac{\sigma}{\omega \epsilon_0} = 0 \right) \rightarrow \text{no conductivity} \quad (\sigma = 0)$$

→ Medium can be classified into

(i) Pure dielectric medium

$$\frac{\sigma}{\omega \epsilon_0} = 0$$

(ii) low loss dielectric medium

$$\epsilon_r \gg \frac{\sigma}{\omega \epsilon_0}$$

(iii) lossy dielectric medium

$$\omega \epsilon_0 \epsilon_r \approx \sigma$$

(iv) conductor medium

$$\sigma \gg \omega \epsilon_0 \epsilon_r$$

(v) Perfect conductor

$$\sigma = \infty$$

Complex dielectric constant.

$$\epsilon_{rc} = \left(\epsilon_r - \frac{j\sigma}{\omega \epsilon_0} \right)$$

$$\rightarrow \nabla^2 \vec{E} = -\omega^2 \mu \epsilon_0 \epsilon_{rc} \vec{E}$$

$$= -\omega^2 \mu \epsilon_0 \left(\epsilon_r - \frac{j\sigma}{\omega \epsilon_0} \right) \vec{E}$$

$$\nabla^2 \vec{E} = \gamma^2 \vec{E}$$

$$\text{where, } \gamma^2 = -\omega^2 \mu \epsilon_0 \left[\epsilon_r - \frac{j\sigma}{\omega \epsilon_0} \right]$$

$$\therefore \gamma = j\omega \sqrt{\mu \epsilon_0} \left[\epsilon_r - \frac{j\sigma}{\omega \epsilon_0} \right]^{1/2}$$

Assume, $\gamma = \alpha + j\beta$

$$* \operatorname{Re}(\gamma) = \alpha = \omega \sqrt{\frac{\mu \epsilon_0 \epsilon_r}{2}} \left\{ \sqrt{1 + \frac{\sigma^2}{\omega^2 \epsilon_0^2 \epsilon_r^2}} - 1 \right\}^{1/2}$$

$$* \operatorname{Im}(\gamma) = \beta = \omega \sqrt{\frac{\mu \epsilon_0 \epsilon_r}{2}} \left\{ \sqrt{1 + \frac{\sigma^2}{\omega^2 \epsilon_0^2 \epsilon_r^2}} + 1 \right\}^{1/2}$$

→ In low loss dielectric medium,

$$\therefore \alpha \approx \frac{\sigma}{2} \sqrt{\frac{\mu}{\epsilon_0 \epsilon_r}}$$

$$\beta = \omega \sqrt{\mu \epsilon_0 \epsilon_r} \left\{ 1 + \frac{1}{8} \frac{\sigma^2}{\mu^2 \epsilon_0^2 \epsilon_r^2} \right\}^{1/2}$$

$$\therefore \beta \approx \omega \sqrt{\mu \epsilon}$$

$$\eta = \sqrt{\frac{j\omega\mu}{(\sigma + j\omega\epsilon)}}$$

$$\therefore \underline{\underline{\text{Re}(\eta) = \sqrt{\frac{\mu}{\epsilon}}}}$$

→ Lossy nature of dielectric can be represented using lossy tangent lower the $\tan\delta$, better the dielectric property of the medium.

$$\text{lossy tangent} = \frac{\sigma}{\omega \epsilon_0 \epsilon_r} \quad (\tan\delta)$$

→ for a wave propagation in conductor medium, $(\sigma \gg \omega \epsilon_0 \epsilon_r)$

$$\alpha = \sqrt{\frac{\omega\mu\sigma}{2}} \quad \beta = \sqrt{\frac{\omega\mu\sigma}{2}} \quad (\alpha = \beta)$$

$$\gamma = \alpha + j\beta$$

$$\therefore \gamma = \sqrt{\frac{\omega\mu\sigma}{2}} (1 + j)$$

$$\therefore \eta = \sqrt{\frac{j\omega\mu}{\sigma + j\omega\epsilon}} = \sqrt{\frac{j\omega\mu}{\sigma}} = \sqrt{\frac{\omega\mu}{\sigma}} e^{j\pi/4}$$

$$\therefore \frac{E_0}{H_0} = \eta = \sqrt{\frac{\omega\mu}{\sigma}} e^{j\pi/4}$$

→ Electric field advances magnetic field by $\pi/4$.

→ The wave propagation in a conductor is very similar to lossy transmission line.

→ Electric field gets attenuated by e^{-1}
(in case of wave propagation in conductor medium)

$$0.367 E_0 = E_0 e^{-1} = E_0 e^{-\alpha z_0}$$

$$\downarrow$$
$$\boxed{z_0 = 1/\alpha}$$

• After traveling a distance of ' z_0 ', wave amplitude decays by e^{-1} .

$$\text{Power} = |0.367 E_0|^2$$
$$= 0.135 |E_0|^2$$

• The distance ' z_0 ' is called skindepth

around 70% of power gets attenuated

→ Skindepth of conductor,

$$\text{skin-depth} = \frac{1}{\alpha} = \sqrt{\frac{2}{\omega \mu \sigma}} = \sqrt{\frac{1}{\pi f \mu \sigma}}$$

(wave doesn't propagate in conductor medium)

→ DC resistance offered by wire at low frequency



$$R_{DC} = \frac{l}{\sigma \times (\text{crosst})}$$

a - radius
l - length

$$\therefore R_{DC} = \frac{l}{\sigma \times (\pi a^2)}$$

→ AC resistance offered,

$$\therefore R_{AC} = \frac{l}{\sigma \times 2\pi a \times \delta}$$

δ → skin-depth.

→ At high frequencies, δ ↓ and R_{AC} ↑.

Q) Conductivity of copper $= 5.8 \times 10^7 \text{ S/m}$
(σ)

$$\epsilon_r = \mu_r = 1$$

$$a = 1.2 \text{ mm}, l = 600 \text{ m}$$

Determine R_{DC} , R_{AC} at 100 MHz .

Sol:-

$$R_{DC} = \frac{l}{\sigma \times \pi a^2}$$

$$= \frac{600}{5.8 \times 10^7 \times \pi \times 1.44 \times 10^{-6}}$$

$$\therefore R_{DC} = 2.287 \Omega$$

$$\text{Skin-depth}(\delta) = \sqrt{\frac{1}{\pi f \mu \sigma}}$$

$$\delta = \sqrt{\frac{1}{\pi \times 100 \times 10^6 \times 4\pi \times 10^{-7} \times 5.8 \times 10^7}}$$

$$\delta = \sqrt{\frac{10^{-8}}{4\pi \times 5.8}}$$

$$\therefore \delta = 6.6 \times 10^{-6} \text{ m}$$

$$R_{AC} = \frac{l}{\sigma \times 2\pi a \times \delta}$$

$$= \frac{600}{5.8 \times 10^7 \times 2\pi \times 1.2 \times 10^{-3} \times 6.6 \times 10^{-6}}$$

$$\therefore R_{AC} = 207.61 \Omega$$

(*) Poynting Theorem:

$$\nabla \times \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t} \rightarrow \text{Faraday's law}$$

$$\nabla \times \vec{H} = \sigma \vec{E} + \epsilon \frac{\partial \vec{E}}{\partial t} \rightarrow \text{Ampere's law}$$

$$\vec{E} \cdot (\nabla \times \vec{H}) = \sigma |\vec{E}|^2 + \epsilon \vec{E} \cdot \frac{\partial \vec{E}}{\partial t}$$

(By property,

$$\nabla \cdot (\vec{A} \times \vec{B}) = \vec{B} \cdot (\nabla \times \vec{A}) - \vec{A} \cdot (\nabla \times \vec{B})$$

$$\vec{H} \cdot (\nabla \times \vec{E}) + \nabla \cdot (\vec{H} \times \vec{E}) = \sigma |\vec{E}|^2 + \epsilon \vec{E} \cdot \frac{\partial \vec{E}}{\partial t} \quad \text{--- (1)}$$

$$\begin{aligned} \Rightarrow \left(\begin{aligned} \frac{\partial |\vec{E}|^2}{\partial t} &= \frac{\partial \vec{E} \cdot \vec{E}}{\partial t} \\ &= \vec{E} \cdot \frac{\partial \vec{E}}{\partial t} + \vec{E} \cdot \frac{\partial \vec{E}}{\partial t} \\ &= 2 \vec{E} \cdot \frac{\partial \vec{E}}{\partial t} \\ \vec{E} \cdot \frac{\partial \vec{E}}{\partial t} &= \frac{1}{2} \frac{\partial |\vec{E}|^2}{\partial t} \end{aligned} \right) \end{aligned}$$

$$\begin{aligned} \Rightarrow \left(\begin{aligned} \vec{H} \cdot (\nabla \times \vec{E}) &= -\vec{H} \cdot \mu \frac{\partial \vec{H}}{\partial t} \\ &= -\frac{1}{2} \mu \frac{\partial |\vec{H}|^2}{\partial t} \end{aligned} \right) \end{aligned}$$

So, the equation (1) is written as,

$$-\frac{\mu}{2} \frac{\partial}{\partial t} |\vec{H}|^2 - \nabla \cdot (\vec{E} \times \vec{H}) = \sigma |\vec{E}|^2 + \frac{\epsilon}{2} \frac{\partial}{\partial t} |\vec{E}|^2$$

$$\nabla \cdot (\vec{E} \times \vec{H}) = -\sigma |\vec{E}|^2 - \frac{\epsilon}{2} \frac{\partial}{\partial t} |\vec{E}|^2 - \frac{\mu}{2} \frac{\partial}{\partial t} |\vec{H}|^2$$

Applying volume integral,

$$\int_V \nabla \cdot (\vec{E} \times \vec{H}) dv = \int_V -\sigma |\vec{E}|^2 dv - \int_V \frac{\mu}{2} \frac{\partial}{\partial t} |\vec{H}|^2 dv - \int_V \frac{\epsilon}{2} \frac{\partial}{\partial t} |\vec{E}|^2 dv$$

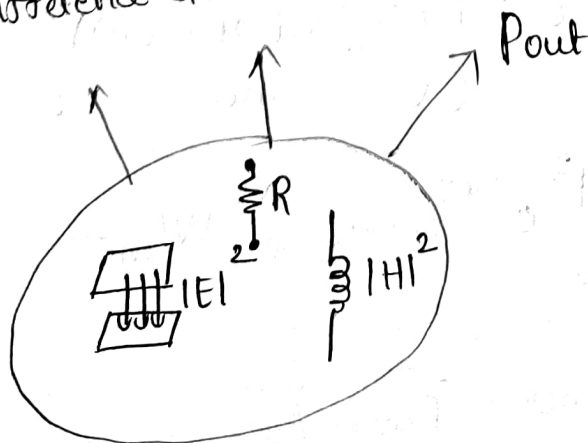
Divergence theorem.

ohmic power loss

Rate of reduction of magnetic energy

$$\oint_S (\vec{E} \times \vec{H}) \cdot d\vec{s} = \int_V -\sigma |\vec{E}|^2 dv - \frac{\mu}{2} \int_V \frac{\partial}{\partial t} |\vec{H}|^2 dv - \frac{\epsilon}{2} \int_V \frac{\partial}{\partial t} |\vec{E}|^2 dv$$

→ For an outward flow of power, there must be reduction in magnetic energy in such a way that difference b/w them is equal to ohmic losses.



Poynting vector.

$$\vec{P} = \vec{E} \times \vec{H}$$

$$\vec{E} = V/m$$

$$\vec{H} = A/m$$

$$E \times H = \text{Watt}/m^2$$

$$\vec{E} = E_0 e^{-\alpha z} \cos(\omega t - \beta z) \hat{x}$$

$$\vec{H} = H_0 e^{-\alpha z} \cos(\omega t - \beta z - \theta_n) \hat{y}$$

$$\left(\frac{E_0}{H_0} = |\eta| \right)$$

$$\vec{P} = E_0 H_0 e^{-2\alpha z} \cos(\omega t - \beta z) \cos(\omega t - \beta z - \theta_n) \hat{z}$$

$$\therefore \vec{P} = \frac{|E_0|^2}{\eta} e^{-2\alpha z} \cos(\omega t - \beta z) \cos(\omega t - \beta z - \theta_n) \hat{z}$$

$$\vec{P}_{avg} = \frac{1}{T} \int_0^T \vec{P} \cdot dt$$

$$= \frac{1}{T} \int_0^T \frac{|E_0|^2}{\eta} e^{-2\alpha z} \cos(\omega t - \beta z) \cos(\omega t - \beta z - \theta_n) \hat{z} dt$$

$$= \frac{1}{2T} \int_0^T \frac{|E_0|^2}{\eta} e^{-2\alpha z} \left[\cos(2\omega t - 2\beta z - \theta_n) + \cos(\theta_n) \right] dt \hat{z}$$

$$= \frac{1}{2T} \frac{|E_0|^2}{\eta} e^{-2\alpha z} \left[\left(\frac{\sin(2\omega t - 2\beta z - \theta_n)}{2\omega} \right) \Big|_0^T + T \cos \theta_n \right] \hat{z}$$

$$\therefore \vec{P}_{avg} = \frac{|E_0|^2}{2\eta} e^{-2\alpha z} \cos \theta_n \hat{z} \text{ W}/m^2$$

$$\therefore \text{Time avg power} = \vec{P}_{avg} \cdot \vec{\text{Area}}$$

$$= \int_S \vec{P}_{avg} \cdot d\vec{s}$$

Generalised equation for time-average power is,

$$\underline{\underline{\vec{P}_{avg} = \frac{1}{2} \operatorname{Re}(\vec{E} \times \vec{H}^*)}}$$

eg: Magnetic field of EM wave is given by

$$\vec{H} = \frac{5\sqrt{3}}{\eta} \cos(\omega t - \beta z) \hat{a}_x + \frac{5}{\eta} \sin(\omega t - \beta z + \pi/2) \hat{a}_y \text{ A/m}$$

Determine the time-average power density.

Sol:- $\vec{E} = \eta \cdot \vec{H}$

$$= 5\sqrt{3} \cos(\omega t - \beta z) \hat{a}_x + 5 \sin(\omega t - \beta z + \pi/2) \hat{a}_y$$

$$\vec{E} = \eta \cdot \vec{H}$$

$$\vec{E} = 5\sqrt{3} \cos(\omega t - \beta z) (-\hat{a}_y) + 5 \sin(\omega t - \beta z + \pi/2) \hat{a}_x$$

$$\vec{P} = \vec{E} \times \vec{H}$$

$$= \left(\frac{(5\sqrt{3})^2}{\eta} \hat{a}_z + \frac{(5)^2}{\eta} \right) \times \frac{1}{2} \cdot \hat{a}_z$$