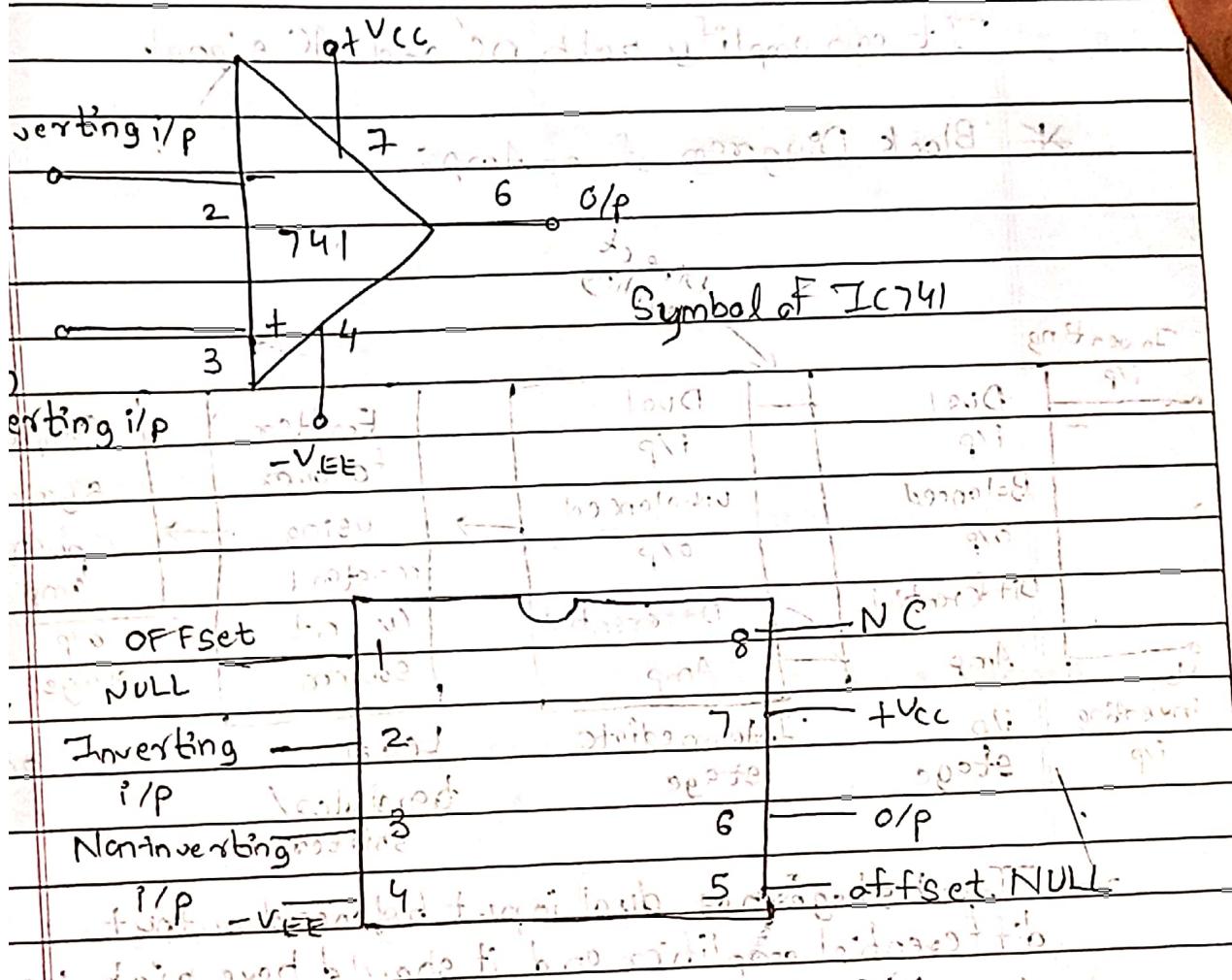


Module 1

Basics of Operational Amp (opamp and IC741)

Opamp - operational amplifier (performs mathematical operations). It is a voltage-controlled voltage source.



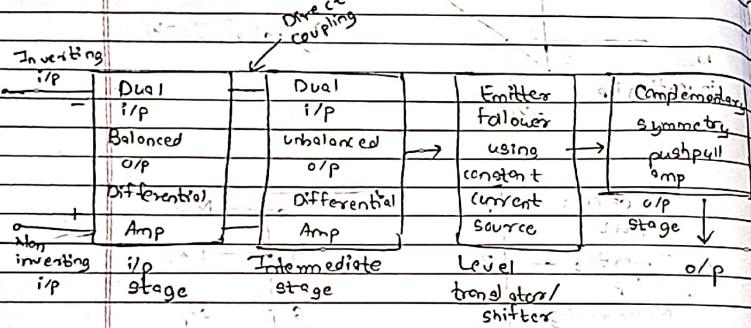
→ OOPAmp is Direct-Coupled High-Gain Amplifier
 consist of one or more differential amplifiers
 followed by level translator and output stage.
 ② Output stage is generally push-pull, complementary
 symmetry pair

different coupling techniques used
 consisting of two diodes connected to negative
 rail through diode + receiving at

(1) Why opAmp is known as operational Amplifier?
 → With the help of opAmp we can perform mathematical functions like addition, subtraction, multiplication, logarithmic, integration, differentiation etc.

→ It can amplify both DC and AC signals.

* Block Diagram of opAmp:



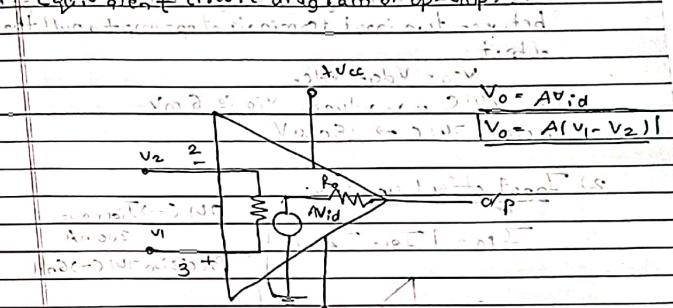
→ Input stage has dual input balanced output differential amplifier and it should have high input impedance.

→ Input stage is directly coupled with dual input unbalance output diff. amplifier which forms intermediate stage.

→ Third stage consists of level shifter (generally emitter follower circuit) which shifts the signal to ground.

OTE 8.1-3: A complementary symmetry pushpull amp forms the output stage which should have low output impedance to give a ~ high voltage gain.

* Equivalent circuit diagram of op-amp:

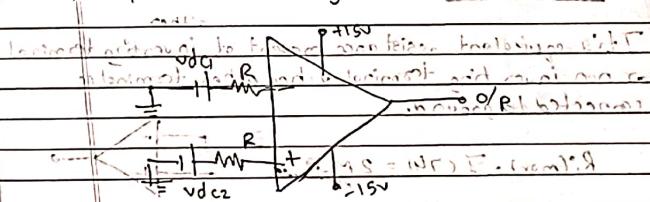


* Features of IC 741:-

- (1) No external frequency compensation required.
- (2) Short circuit protection.
- (3) Offset null capability.
- (4) Large common mode and differential voltage ranges.
- (5) Low power consumption.
- (6) No latchup problems.

* Different parameters of op-amp-

1) Input offset voltage = (S) non-inverting input



2) Commercial temperature range
(0 to 70°C)

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Input offset voltage is voltage that must be applied between two input terminals of op-amp to null the output.

$$V_{IO} = V_{DC1} - V_{DC2}$$

For 741C max value of V_{IO} is 6 mV

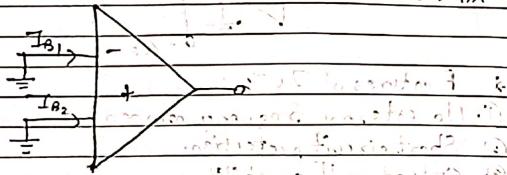
Precision 741C $\rightarrow 150 \mu V$

2) Input offset current -

$$741C \rightarrow I_{IO(max)} = 200 \text{ nA}$$

$$I_{IO} = |I_{B1} - I_{B2}|$$

Precision 741C $\rightarrow 6 \text{ nA}$



3) Input bias current -

$$I_B = I_{B1} + I_{B2} \quad I_B = 500 \text{ nA.} \quad 741C$$

$I_B = \pm 7 \text{ nA} = \text{precision}$

4) Input Resistance (R_i) -

This equivalent resistance present at inverting terminal or non-inverting terminal when other terminal is connected to ground.

$$R_i^{(max)} = I_{C741} = 2 \text{ M}\Omega$$



5) Input Capacitance -

This equivalent capacitance measured at either non-inverting terminal or inverting terminal with other terminal connected to ground.

$$I_{C741C} = 1.4 \text{ pF.} \leftarrow \text{Should be as low as possible.}$$

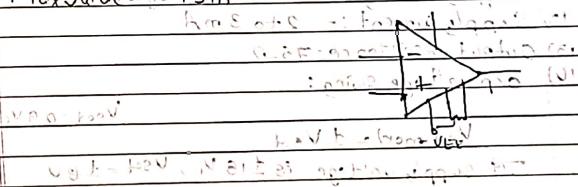
6) Input voltage Range:-

When we apply same input signal at both terminals then it is known as common mode signal and input voltage range is defined under this condition. Max value for 74741 $\pm 13 \text{ V.}$

7) Offset Voltage Adjustment Range:-

This capability to null offset voltage. For this potentiometer of 10 kΩ is connected to offset null pins 1 and 5 and variable terminal is connected to $-VEE$. By using a potentiometer we can reduce offset voltage range.

$$\text{Max value} = \pm 15 \text{ mV}$$



9) Common Mode Rejection Ratio (CMRR)

$$\text{CMRR} = \frac{A_d}{A_c} = \frac{\text{differential mode gain}}{\text{common mode gain}}$$

↑ opamp
CMRR = ∞

- It indicates ability to reject common mode signal
- It should be higher as possible.
- Max. value 90 dB

9) Supply Voltage Rejection Ratio (SVRR)

$$\text{SVRR} = \frac{AV_{io}}{AV_s} = 1 \text{ V/V}$$

This ratio of change
in output voltage with respect to
change in supply voltage

10) Large signal voltage gain:

$$A_d = \frac{V_o}{V_s} = \frac{\text{maximum differential voltage}}{\text{maximum input voltage}}$$

Max. value = 2×10^5

(11) Supply voltage :- Maximum range = $\pm 15 \text{ V}$

(12) Supply current :- 2 to 3 mA

(13) Output resistance : 75Ω

(14) O/p voltage swing :

$$V_{sat} = 0.9 \text{ Vcc}$$

$$V_o(\text{max}) = \pm V_{sat}$$

If supply voltage is $\pm 15 \text{ V}_i$, $V_{sat} = \pm 9 \text{ V}$

15) Slow Rate

$$SR = \frac{dV_o}{dt} \text{ volt/sec} = 0.5 \text{ V/usec}$$

It is defined as maximum rate of change of output voltage with respect to time. It decides capability of opamp to change its output rapidly.

(16) Bandwidth: It is the frequency range of an amplifier over which the signal frequencies are amplified equally.

$$\text{Bandwidth of } 741C = 1 \text{ MHz.}$$

(17) Gain bandwidth product:- Bandwidth of op amp when gain of amplifier is 1.

(18) Output short circuit current :- Max value : 25 mA

Ideal op-amp Practical value

R_i	∞	$10^12 \Omega$
R_o	0	75Ω
A_v	∞	2×10^5
B.W.	∞	1 MHz
CMRR	∞	90 dB
Slew rate	0.0	0.5 V/ μ s
I/p offset voltage	0	6 mV
SVRR	0	150 nV/V
I/p bias current	0	500 nA
i/p offset current	0	200 nA

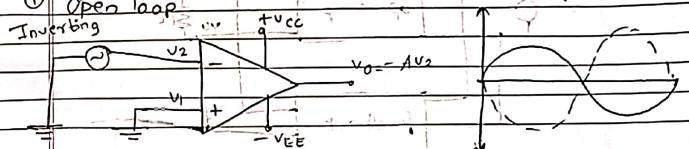
Op-amp configuration

- Open loop
- Inverting
- Non-inverting
- Differential
- Closed loop
- Inverting
- Non-inverting
- Differential

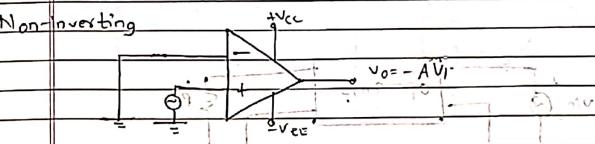
$$V_o = A(V_1 - V_2)$$

= $-A V_2$

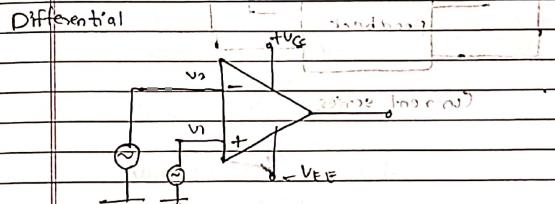
① Open loop

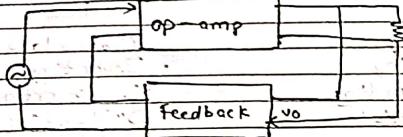


Non-inverting

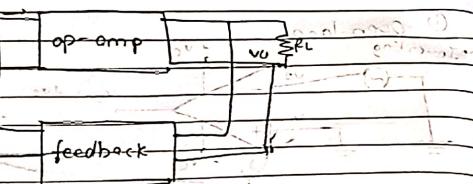


Differential

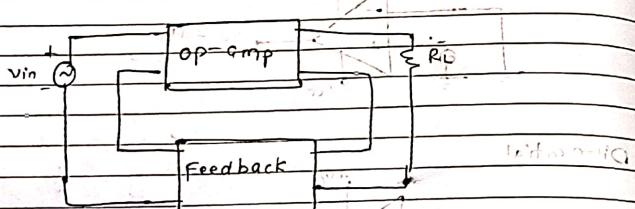




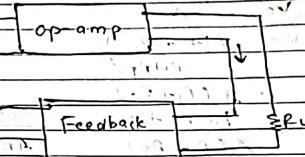
voltage series feedback



voltage shunt

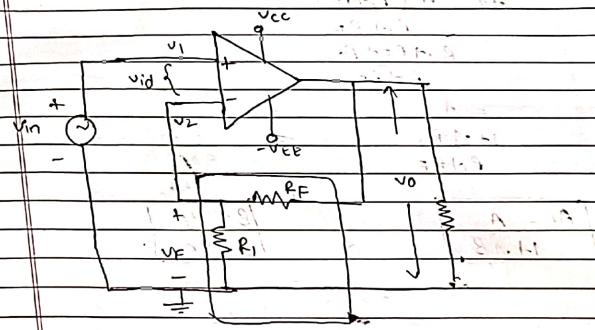


Current series



current shunt

**K Voltage Series
(Non inverting amp with Feedback)**



$$V_o = A(V_1 - V_2)$$

$$V_1 = V_{in}$$

$$V_2 = V_f = \frac{R_1}{R_1 + R_f} V_o$$

$$V_o = A(V_{in} - \frac{R_1}{R_1 + R_f} V_o)$$

$$V_o \left(1 + \frac{A R_1}{R_1 + R_f} \right) = A V_{in}$$

$$\frac{V_o}{V_{in}} = \frac{A}{R_i + R_F}$$

$$= A \left(\frac{R_i + R_F}{R_i + R_F + A R_i} \right)$$

$$= A \left(\frac{R_i + R_F}{R_i + R_F + A R_i} \right) \quad | A R_i \gg R_i + R_F$$

$$A_f = \frac{V_o}{V_{in}} = \frac{1 + R_F}{R_i}$$

$$A_f = A \left(\frac{R_i + R_F}{R_i + R_F + A R_i} \right)$$

$$R_i + R_F$$

$$R_i + R_F + A R_i$$

$$F_H, R_F$$

$$= A$$

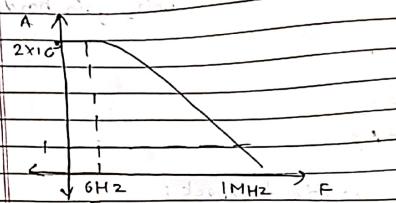
$$1 + A R_i$$

$$R_i + R_F$$

$$A_f = A \quad | \quad \beta = \frac{U_f}{U_o}$$

REDMI NOTE 8 PRO
KARAN

* Voltage gain of feedback amplifier (Non inverting)



$$U_{GB} = A \sqrt{f_0}$$

$$U_{GB} = A_f F_f$$

$$A \sqrt{f_0} = A_f F_f$$

$$F_f = A \sqrt{f_0}$$

$$F_f = A F_0 = \frac{f_0 (1 + A \beta)}{A / HAB}$$

$$A_f = 1 + \frac{R_F}{R_i}$$

$$A_f = A \frac{1 + A \beta}{1 + A \beta}$$

$$R_{i,f} = R_i (1 + A \beta)$$

$$R_{o,f} = R_o \frac{1 + A \beta}{1 + A \beta}$$

* Break frequency : Frequency at which gain is 3dB down from its value known as break frequency (f_0).

→ Unity Gain Bandwidth: The frequency at which frequency gain = 1 is known as unity gain bandwidth.

$$F_U = F_0 (1 + A\beta)$$

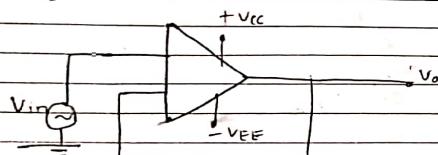
$$V_{out} = \frac{\pm V_{sat}}{1 + A\beta}$$

Advantages of negative feedback:

- 1) Input impedance will increase
- 2) Output impedance will reduce
- 3) Bandwidth will increase
- 4) Effect of noise will be minimum
- 5) Output offset voltage will reduce
- 6) Temperature variation effects will be minimum.

* Applications of Non-inverting amp:

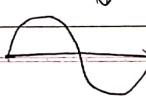
→ Voltage Follower:



$$Av = 1 + \frac{R_F}{R_1} \quad \therefore R_F = 0 \quad \therefore Av = 1$$

$$\beta = \frac{R_F}{R_1 + R_F}$$

$$\beta = 1 \quad (\because R_F = 0)$$



Voltage follower acts as a buffer

$$R_{if} = A R_i$$

$$R_{of} = R_o$$

$$A$$

$$F_F = A F_0$$

$$V_{out} = \frac{\pm V_{sat}}{A}$$

Q. 741C op-Amp having following parameters $R_i = 10k\Omega$ and $R_o = 1k\Omega$, supply voltage $\pm 15V$. Compute A , R_{if} , R_{of} , F_F , Total output offset voltage with feedback.

$$\rightarrow A_F = 1 + \frac{R_F}{R_i} = 1 + \frac{10}{1} = 11$$

$$\begin{aligned} R_{if} &= R_i (1 + A\beta) = 2 \left(1 + 2 \times 10^5 \left(\frac{1}{11} \right) \right) \\ &= 2 \left(1 + 2 \times 10^5 \right), \\ &\approx 33.07 \cdot 78. \end{aligned}$$

$$\begin{aligned} R_{of} &= R_o = \frac{R_o}{1 + A\beta} = \frac{18182.82}{1 + 2 \times 10^5} \\ &\approx 75, \\ &= 18182.82 \\ &= 4.12 \times 10^{-3} \\ &\approx 4.12 \text{ m}\Omega. \end{aligned}$$

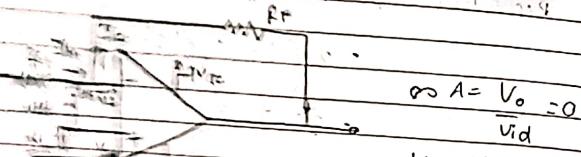
$$= 5 \left(\frac{1 + 10^5}{11} \right)$$

$$= 5 \left(\frac{1 + 10^5}{11} \right) = 90 \text{ mHz}$$

$$\omega_{\text{c}} = \frac{V_{\text{out}}}{I \cdot A_B}$$

* Voltage subtract ampl. (Inverting amplifier)

Voltage Output Ideally difference between two input voltage should be zero means voltage at non-inverting terminal should be equal to voltage at inverting terminal. If one terminal is connected to ground then the other terminal is connected to the ground.



$$\text{as } A = \frac{V_o}{V_{\text{in}}} = 0$$

$$V_1 = V_2$$

$$T_{\text{in}} = T_B + T_F$$

$$\frac{V_o - V_2}{R_1} = \frac{V_2 - V_o}{R_F}$$

$$\frac{V_o}{R_1} = -V_o$$

$$\frac{V_o}{R_1} = -\frac{R_F}{R_1}$$

$$\frac{V_{\text{in}} - V_2}{R_1} = \frac{V_2 - V_o}{R_F} \quad (\text{assuming } V_1 = V_2)$$

$$\frac{V_{\text{in}} - V_2}{R_1} = \frac{V_o}{R_F}$$

$$V_2 = -V_o$$

$$\frac{V_{\text{in}} + V_o}{R_1} = -\frac{V_o}{A_B} - V_o$$

$$\frac{V_{\text{in}}}{R_1} = -A_B V_o$$

$$\boxed{\frac{V_o}{V_{\text{in}}} = -A_B \frac{R_F}{R_1 + R_F + A_B R_1}}$$

* Divide Numerator and Denominator with $R_1 + R_F$.

$$\frac{V_o}{V_{\text{in}}} = -A_B \frac{R_F}{R_1 + R_F}$$

$$\boxed{\frac{V_o}{V_{\text{in}}} = -A_B}$$

$A_f = -A_B$

$B = \text{Feedback gain}$

$k = \text{Multiplying factor}$

$$\boxed{① B = \frac{R_F}{R_1 + R_F}}$$

$$\boxed{② R_{\text{in}} = R_1 + \left(\frac{R_F \cdot R_1}{R_1 + R_F} \right)}$$

$$\boxed{③ R_{\text{out}} = R_F}$$

$$\boxed{④ F_F = A_B (1 + A_B)}$$

$$\boxed{⑤ V_{\text{out}} = \pm V_{\text{sat}}}$$

$$1 + A_B$$

* Applications of Inverting Amplifier

$$\textcircled{1} \text{ Current to Voltage Converter} = \textcircled{2} \text{ Inverter}$$

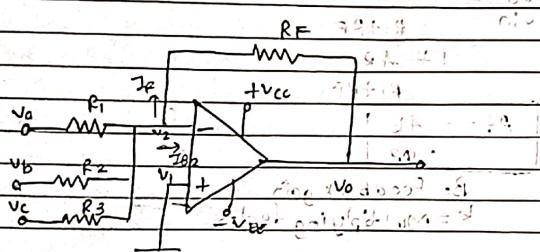
$\Rightarrow \frac{V_o}{V_{in}} = -RF$

$V_o = -\frac{RF}{R_1} V_{in}$

(For Designing
Inverter)

* Summing, Scaling and Averaging amp

① Inverting configuration -



$$\begin{aligned} I_a + I_b + I_c &= I_{B_2} + I_F & (1) \\ \therefore I_a + I_b + I_c &= I_F & (2) \end{aligned}$$

$$\frac{V_a - V_2}{R_1} + \frac{V_b - V_2}{R_2} + \frac{V_c - V_2}{R_3} = \frac{V_2 - V_o}{R_F}$$

$$V_a + V_b - V_c = -V_c$$

$$\frac{V_a + V_b - V_c}{R_1 R_2 R_3} = -V_o \quad \text{SN+}$$

$R_1 \quad R_2 \quad R_3 \quad R_{FB4.412.3} \quad .1 \text{ m}$

$$V_o = - \frac{R_F}{R_1 + R_2 + R_3} (V_a + V_b + V_c)$$

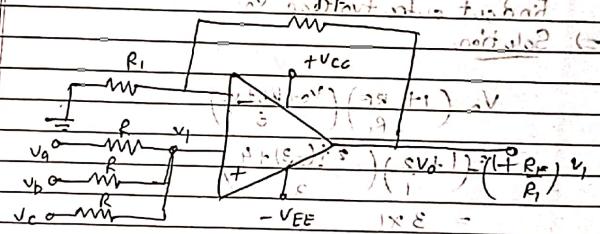
② Averaging amp

③ Scaling amp

$$R_F = R$$

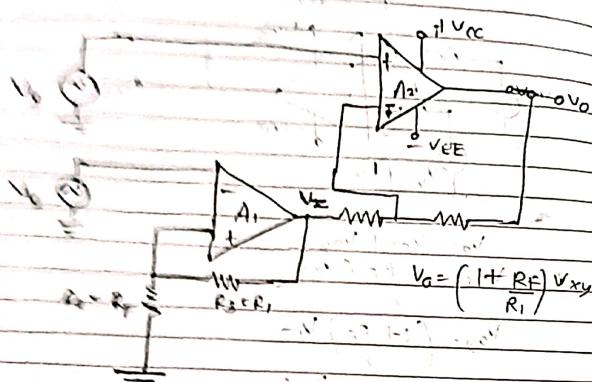
$R_1 = R_2 = R_3 \neq R$ because $1/2$ is not in $\frac{1}{2}$

* Application of Non-inverting amplifier



$$V_1 = \left(\frac{R/2}{R/2 + R} \right) V_a + \left(\frac{R/2}{R + R/2} \right) V_b + \left(\frac{R/2}{R + R/2} \right) V_c$$

using two op-amps:



$\Rightarrow A_1$ will work as a non-inverting amp

$$V_z = \left(1 + \frac{R_2}{R_1}\right) V_y$$

$$V_x = V_{ox} + V_{oz}$$

$$V_{ox} = \left(1 + \frac{R_F}{R_1}\right) V_x$$

$$V_{oz} = -R_F V_z$$

$$V_z = \left(1 + \frac{R_3}{R_2}\right) V_y$$

$$V_o = \left(1 + \frac{R_F}{R_1}\right) V_x + -\frac{R_F}{R_1} \left(1 + \frac{R_3}{R_2}\right) V_y$$

REDMI NOTE 6 PRO

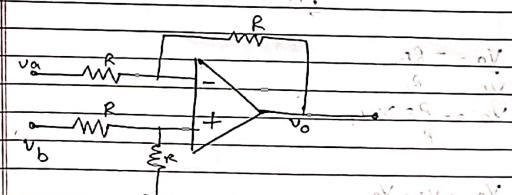
$$R_3 = R_1, R_2 = R_F$$

$$= \left(1 + \frac{R_F}{R_1}\right) V_x + -\frac{R_F}{R_1} \left(1 + \frac{R_1}{R_F}\right) V_y$$

$$= \left(1 + \frac{R_F}{R_1}\right) V_x - \frac{R_F}{R_1} V_y$$

$$= \left(1 + \frac{R_F}{R_1}\right) V_x - \frac{R_F}{R_1} V_y$$

* Differential amp as Subtractor:-



$$V_o = -\frac{R_2}{R_1} (V_a - V_b)$$

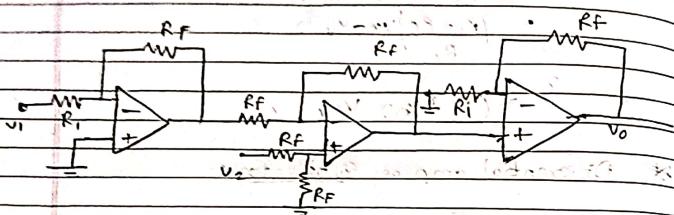
$$= -(V_a - V_b) \cdot \frac{R_2}{R_1} = -2V$$

$$V_o = V_b - V_a$$

$$= -2V$$

Ex For given circuit Define o/p voltage if

$$R_f = 10 k\Omega, R_1 = 1 k\Omega$$



$$V_{o1} = -R_f \frac{V_1}{R_1}$$

$$V_{o1} = -R_f \times V_1$$

$$V_{o2} = V_2 - V_{o1}$$

$$= V_2 + R_f \times V_1 = V_2 + 10 V_1$$

$$V_{o3} = \left(1 + R_f\right) V_{o2}$$

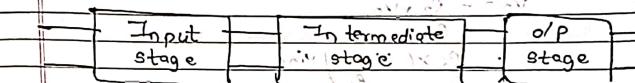
$$= \left(1 + R_f\right) \left(V_2 + R_f \times V_1\right)$$

$$= \left(1 + 10\right) \left(V_2 + 10 V_1\right)$$

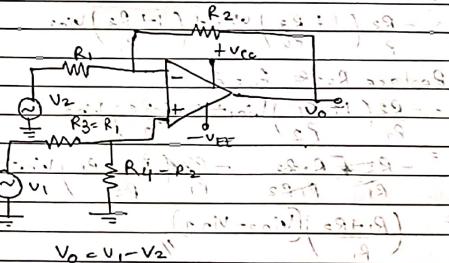
$$V_{o3} = 11 \left(V_2 + 10 V_1\right)$$

JO REDMINISTER
CO KARAN

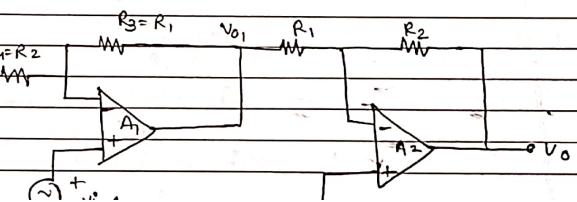
Instrumentation amplifier :- (IA)



* TA using one op-amp:



* IA Using 2 op-amps:



$$V_o = \left(\frac{R_3}{R_4} \right) V_{in_2}$$

Due to V_{in_1} ,

$$V_o' = -\frac{R_2}{R_1} \times V_{in_1}$$

$$= -\frac{R_2}{R_1} \times \left(1 + \frac{R_3}{R_4} \right) V_{in_2}$$

Due to V_{in_2} ,

$$V_o'' = \left(1 + \frac{R_3}{R_4} \right) V_{in_2}$$

$$V_o = V_o' + V_o''$$

$$= -\frac{R_2}{R_1} \left(1 + \frac{R_3}{R_4} \right) V_{in_1} + \left(1 + \frac{R_3}{R_4} \right) V_{in_2}$$

Replace $R_3 = R_1$, $R_4 = R_2$,

$$= -\frac{R_2}{R_1} \left(1 + \frac{R_1}{R_2} \right) V_{in_1} + \left(1 + \frac{R_1}{R_2} \right) V_{in_2}$$

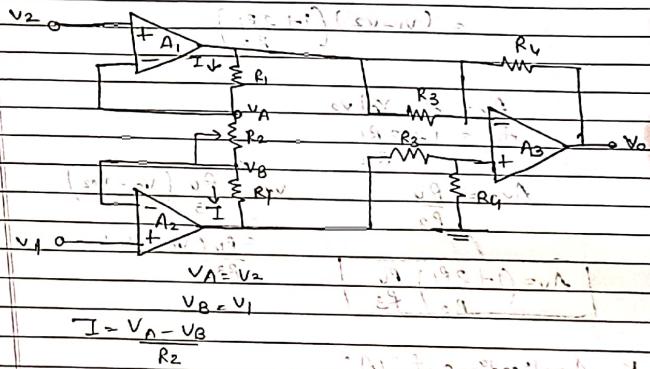
$$= -\frac{R_2}{R_1} \frac{R_1 R_2}{R_1 + R_2} \left(\frac{R_2 + R_1}{R_1} \right) V_{in_1} + \left(\frac{R_1 + R_2}{R_1} \right) V_{in_2}$$

$$= \left(\frac{R_1 + R_2}{R_1} \right) (V_{in_2} - V_{in_1})$$

* Advantages of dual Instrumentation op-amp

1. High Input Resistance
2. Low Output Resistance
3. CMRR can be maximized by variable resistors

* IA using 3A op-amp



$$V_A = V_2$$

$$V_B = V_1$$

$$I = \frac{V_A - V_B}{R_2}$$

$$V_{o1} = V_A + I R_1$$

$$V_{o2} = V_B - I R_1$$

$$V_{o1} = V_A + I R_1$$

$$= V_2 + \left(\frac{V_2 - V_1}{R_2} \right) R_1$$

$$= \frac{(R_1 + R_2) V_2 - R_1 V_1}{R_2}$$

$$V_{o2} = \frac{R_1 (R_1 + R_2) V_1 - R_1 V_2}{R_2}$$

$$V_o = V_{o2} - V_{o1}$$

EIA

$$\begin{aligned}
 & \frac{(R_1 + R_2) V_1 - R_1 V_2}{R_2} = \frac{(R_1 + R_2) V_2 - R_2 V_1}{R_1} \\
 & = \frac{R_1 V_1 + R_2 V_1 - R_1 V_2 - R_2 V_2}{R_2} = \frac{R_2 V_2 + R_1 V_1 - R_1 V_2 - R_2 V_2}{R_2} \\
 & = \frac{2 R_1 (V_1 - V_2) + R_2 (V_1 - V_2)}{R_2} \\
 & = (V_1 - V_2) \left(1 + \frac{2 R_1}{R_2} \right)
 \end{aligned}$$

$$Av = A_{V_1} \times A_{V_2}$$

$$A_{V_1} = 1 + 2 R_1$$

$$A_{V_2} = \frac{R_4}{R_3}$$

$$V_{O_2} = \frac{R_4}{R_3} (V_{O_1} - V_{O_2})$$

$$= \frac{R_4}{R_3} (V_{O_2} - V_{O_1})$$

$$Av = \left(1 + \frac{2 R_1}{R_2} \right) \frac{R_4}{R_3}$$

* Applications of TA:

e.g. 1) Temperature indicators

2) Pressure monitoring and control

3) Temperature controller

4) Light Intensity Meter

5) Electronic weighing scale.