# Discussion 5

EE599: Deep Learning Keith M. Chugg Spring 2020

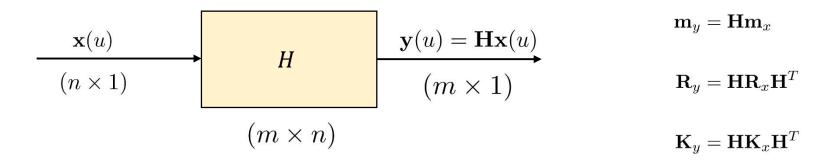


# Outline

Principal Component Analysis

• Filter Design

## Random Vectors



#### Correlation

$$\mathbf{R}_{y} = \mathbb{E} \big[ \mathbf{y}(u) \mathbf{y}^{T}(u) \big]$$

$$= \mathbb{E} \big[ \mathbf{H}(\mathbf{x}(u)) (\mathbf{H}\mathbf{x}(u))^{T} \big]$$

$$= \mathbb{E} \big[ \mathbf{H}\mathbf{x}(u) \mathbf{x}(u)^{T} \mathbf{H}^{T} \big]$$

$$= \mathbf{H} \mathbb{E} \big[ \mathbf{x}(u) \mathbf{x}(u)^{T} \big] \mathbf{H}^{T}$$

$$= \mathbf{H} \mathbf{R}_{x} \mathbf{H}^{T}$$

### Special case

$$y(u) = \mathbf{b}^T \mathbf{x}(u)$$
  $(1 \times 1)$   
 $m_y = \mathbf{b}^T \mathbf{m}_x$   
 $\mathbb{E}[y(u)^2] = \mathbf{b}^T \mathbf{R}_x \mathbf{b}$   
 $\sigma_y^2 = \mathbf{b}^T \mathbf{K}_x \mathbf{b}$ 

Note that covariance / correlation matrices are symmetric, non-negative definite.

## Karhunen-Loéve (KL) Expansion

Can always find orthonormal set of eigen-vectors of **K**.

These are an alternate coordinate systems (rotations, reflections)

in this eigen-coordinate system, the components are uncorrelated

(principle components)

Eigen-values are the variant (energy) in each of these principle directions

(can be used to reduce dimensions by throwing out components with lowenergy)

## **KL-Expansion**

$$\mathbf{K}_{x}\mathbf{e}_{k} = \lambda_{k}\mathbf{e}_{k} \qquad k = 0, 1, ...., N - 1 \qquad \text{(Eigen equation)}$$

$$\mathbf{e}_{k}^{T}\mathbf{e}_{l} = \delta[k - l] \qquad \lambda_{k} \geq 0 \qquad \text{(Orthonormal eigen-vectors)}$$

$$\mathbf{x}(u) = \sum_{k=0}^{N-1} X_{k}(u)\mathbf{e}_{k} \qquad \text{(Change of coordinates)}$$

$$X_{k}(u) = \mathbf{e}_{k}^{T}\mathbf{x}(u)$$

$$\mathbb{E}[X_{k}(u)X_{l}(u)] = \mathbf{e}_{k}^{T}\mathbf{K}_{x}\mathbf{e}_{l} = \lambda_{k}\delta(k - l) \qquad \text{(Uncorrelated components)}$$

$$\mathbf{K}_{x} = \sum_{k=0}^{N-1} \lambda_{k}\mathbf{e}_{k}\mathbf{e}_{l}^{T} \qquad \text{(Mercer's Theorem)}$$

$$\mathbb{E}[||\mathbf{x}(u)||^{2}] = \operatorname{tr}(\mathbf{K}_{x}) = \sum_{k=0}^{N-1} \lambda_{k} \qquad \text{(Total Energy)}$$

Always exists because **K** is positive semi-definite (PSD) and symmetric

## **KL-Expansion**

$$d_k(u) = \mathbf{e}_k^T \mathbf{x}(u) \qquad k = 0, 1, ...., D - 1$$

$$\mathbf{d}(u) = \mathbf{E}^T \mathbf{x}(u)$$

$$\mathbf{K}_d = \mathbf{E}^T \mathbf{K}_x \mathbf{E}$$

$$= \mathbf{E}^T (\mathbf{E} \mathbf{\Lambda} \mathbf{E}^T) \mathbf{E}$$

$$= \mathbf{\Lambda} = \mathbf{diag}(\lambda_k)$$

Multiplying by  $\mathbf{E}^T$  makes the component uncorrelated

$$\mathbf{E} = \left[ egin{array}{c|c} \mathbf{e}_0 & \mathbf{e}_1 & \mathbf{e}_2 & \dots & \mathbf{e}_{D-1} \end{array} 
ight]$$

## **KL-Expansion: Relation to Whitening**

$$w_k(u) = \frac{X_k(u)}{\sqrt{\lambda_k}} = \frac{\mathbf{e}_k^T \mathbf{x}(u)}{\sqrt{\lambda_k}}$$
  $k = 0, 1, ..., D - 1$ 

$$\mathbf{w}(u) = \mathbf{\Lambda}^{-1/2} \mathbf{E}^T \mathbf{x}(u)$$

$$\mathbf{K}_w = \mathbf{\Lambda}^{-1/2} \mathbf{E}^T \mathbf{K}_x \mathbf{E} \mathbf{\Lambda}^{-1/2}$$

$$= \mathbf{\Lambda}^{-1/2} \mathbf{\Lambda} \mathbf{\Lambda}^{-1/2}$$

$$= \mathbf{I}$$

For any orthogonal matrix **U**, this whitening matrix also works:

$$\mathbf{G} = \mathbf{U} \mathbf{\Lambda}^{-1/2} \mathbf{E}^T$$

## KL- Expansion: Relation to PCA

$$\tilde{\mathbf{x}}_k(u) = \mathbf{e}_k^K \mathbf{x}(u) \qquad k = 0, 1, \dots, K - 1$$

$$\tilde{\mathbf{x}}(u) = \mathbf{E}_{[:K]} \mathbf{x}(u) \qquad \text{First } K \text{ components}$$

$$\mathbf{K}_{\tilde{\mathbf{x}}} = \mathbf{\Lambda}_{[:K]} \qquad \text{Assumes ordered eigen-values: } \lambda_0, \lambda_1, \dots, \lambda_{D-1}$$

$$\mathbb{E}[||\tilde{\mathbf{x}}(u)||^2] = \sum_{k=0}^{K-1} \lambda_k$$

$$\mathbb{E}[||\mathbf{x}(u) - \tilde{\mathbf{x}}(u)||^2] = \sum_{k=K}^{D-1} \lambda_k \qquad \text{minimizes approximation error (lossy compression)}$$

**PCA** is simply taking only the K most important Eigen-directions or principal components

$$\mathbf{E}_{[:K]} = \left[ egin{array}{c|c} \mathbf{e}_0 & \mathbf{e}_1 & \mathbf{e}_2 & \ldots & \mathbf{e}_{K-1} \end{array} 
ight]$$

## KL- Expansion / PCA for Data

Everything is the same, except that we use the data averaging instead of the expectation  $\mathbb{E}[\cdot]$ 

$$\hat{\mathbf{R}}_{x} = \langle \mathbf{x} \mathbf{x}^{T} \rangle_{\mathcal{D}}$$

$$= \frac{1}{N} \sum_{n=0}^{N-1} \mathbf{x}_{n} \mathbf{x}_{n}^{T}$$

$$= \frac{1}{N} \mathbf{X}^{T} \mathbf{X}$$

Both KL and PCA can be applied to  $\bf R$  or  $\bf K$ . Center x if you want to use  $\bf K$ 

$$x \leftarrow x - m$$

(same if mean is zero)

$$\mathbf{X} = \left[egin{array}{c} \mathbf{x}_0^T \ \mathbf{x}_1^T \ dots \ \mathbf{x}_{N-1}^T \end{array}
ight]$$

$$\mathbf{X}^T = \begin{bmatrix} \mathbf{x}_0 & \mathbf{x}_1 \dots & \mathbf{x}_{N-1} \end{bmatrix}$$

"stacked" data matrix

## KL- Expansion / PCA for Data

### PCA for data

$$\tilde{\mathbf{x}}_n = \mathbf{E}_{[:K]}^T \mathbf{x}_n$$

First K components

$$\mathbf{E}_{[:K]} = \left[ egin{array}{c|c} \mathbf{e}_0 & \mathbf{e}_1 & \mathbf{e}_2 & \ldots & \mathbf{e}_{K-1} \end{array} 
ight]$$

### Apply to the "stacked" data matrix

$$ilde{\mathbf{X}} = \left[egin{array}{c} (\mathbf{E}_{[:K]}^T \mathbf{x}_0)^T \ (\mathbf{E}_{[:K]}^T \mathbf{x}_1)^T \ dots \ (\mathbf{E}_{[:K]}^T \mathbf{x}_{N-1})^T \end{array}
ight] = \mathbf{X} \mathbf{E}_{[:K]}$$

$$egin{array}{lll} ilde{\mathbf{X}} &=& \mathbf{X} & \mathbf{E}_{[:K]} \ _{N imes D} & _{D imes K} \end{array}$$

Dimension reduced from *D* to *K* 

$$\tilde{\mathbf{X}}^T = \begin{bmatrix} \mathbf{E}_{[:K]}^T \mathbf{x}_0 & \mathbf{E}_{[:K]}^T \mathbf{x}_1 \dots & \mathbf{E}_{[:K]}^T \mathbf{x}_{N-1} \end{bmatrix} = \mathbf{E}_{[:K]}^T \mathbf{X}^T$$

PCA reduces the number of features from *D* to *K* 

# KL / PCA for Data: Singular Value Decomposition (SVD)

#### **SVD** for an arbitrary matrix **A**

$$\mathbf{A}_{m \times n} = \mathbf{U}_{m \times m} \sum_{m \times n} \mathbf{V}^T$$

The **SVD** for A provides the KL factorization for the non-negative definition, symmetric  $A^TA$ 

### Use **SVD** to expand matrix $A^TA$

$$\mathbf{A}^{T}\mathbf{A} = (\mathbf{U}\boldsymbol{\Sigma}\mathbf{V})^{T}\mathbf{U}\boldsymbol{\Sigma}\mathbf{V}$$

$$= \mathbf{V}\boldsymbol{\Sigma}^{T}\mathbf{U}^{T}\mathbf{U}\boldsymbol{\Sigma}\mathbf{V}$$

$$= \mathbf{V}\sum_{n \times n} \sum_{n \times n} \mathbf{\Sigma}^{T} \quad \mathbf{V}^{T}$$

$$= \mathbf{E}\boldsymbol{\Lambda}\mathbf{E}^{T}$$

 ${\bf U}$  and  ${\bf V}$  orthogonal matrices,  ${\bf \Sigma}$  is the diagonal with singular values on diagonal

Note that this is also the **SVD** for  $A^TA$ 

## KL-Expansion / PCA for Data: Singular Value Decomposition (SVD)

#### **SVD** for a stacked data matrix **X**

$$egin{array}{lll} \mathbf{X} &=& \mathbf{U} & \mathbf{\Sigma} & \mathbf{V}^T \ _{N imes D} &=& \mathbf{V} & \mathbf{\Sigma} \mathbf{\Sigma}^T & \mathbf{V}^T \ _{D imes D} &=& \mathbf{V} & \mathbf{\Sigma} \mathbf{\Sigma}^T & \mathbf{V}^T \ _{D imes D} &=& \mathbf{E} \mathbf{\Lambda} \mathbf{E}^T \end{array}$$

### Equivalent approaches:

- 1. Find the **SVD** of **X**. Take **V**
- 2. Find the Eigen decomposition of  $X^TX$ . Take E = V
- 3. Find the **SVD** of  $X^TX$ . Take V = U = E

$$ilde{\mathbf{X}} = \mathbf{X} \quad \mathbf{V}_{[:K]}$$

## KL-Expansion / PCA for Data: Singular Value Decomposition (SVD)

#### Equivalent approaches:

- 1. Find the **SVD** of **X**. Take **V**
- 2. Find the Eigen decomposition of  $\mathbf{X}^{\mathbf{T}}\mathbf{X}$ . Take  $\mathbf{E} = \mathbf{V}$
- 3. Find the **SVD** of  $X^TX$ . Take V = U = E

May want to use method 3, with numpy.linalg.svd, instead of method 2, with numpy.linalg.eig, since the **SVD** returns the Eigen-vectors in sorted order and eig does not.

$$ilde{\mathbf{X}} = \mathbf{X} \quad \mathbf{V}_{[:K]}$$
 $N \times K \quad N \times D \quad D \times K$ 

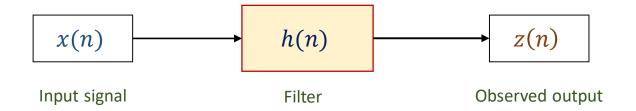
# Outline

Principal Component Analysis

• Filter Design

## Filter Design

This involves creating a signal processing system that satisfies a pre-defined set of requirements.



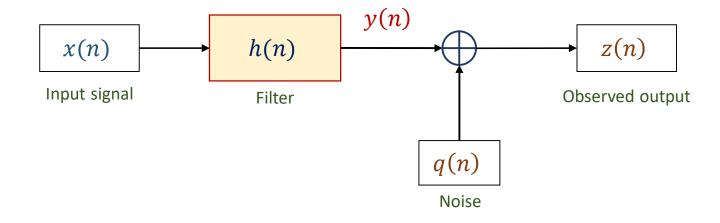
### **Example of Design requirements**

- Frequency Response: Low-pass, high-pass, band-pass, or band-stop
- Impulse Response : Finite (FIR) or infinite (IIR)
- Causality: Causal or non-causal
- Phase and group delay

The goal is to find h(n) that minimizes the error between the observed output z(n) and the expected output y(n)

## Homework 2: Filter Design for Additive Noise

Input signal x(n) goes through a linear system and is observed in additive Gaussian noise.



$$z(n) = y(n) + q(n)$$

$$y(n) = h_0 x(n) + h_1 x(n-1) + h_2 x(n-2)$$

We need a system that minimize the mean squared error between y(n) and z(n)

## Wiener Filter

This is a *linear* and *time invariant* filter that separates original signal  $y_n(u)$  from a corrupted signal.



### **Assumptions:**

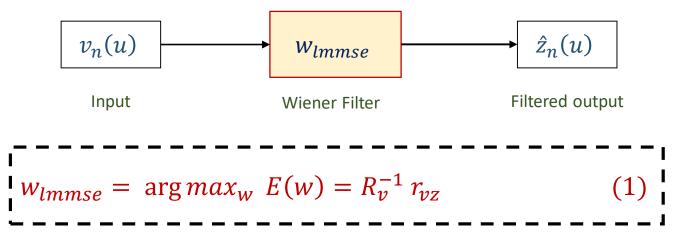
- Original signal and (additive) noise are stationary linear stochastic processes.
- Known spectral characteristics or known <u>autocorrelation</u> and <u>cross-correlation</u>.
- It is a minimum <u>mean-square error</u> (MMSE) estimator.

$$\hat{z}_{n}(u) = w_{lmmse}^{T} x_{n}(u)$$

$$E(w) = \mathbb{E}[\|y_{n}(u) - w^{T}x_{n}(u)\|^{2}]$$

$$w_{lmmse} = \arg\max_{w} E(w) = R_{x}^{-1} r_{xz}$$

## Homework 2: Wiener Filter



$$y(n) = h_0 x(n) + h_1 x(n-1) + h_2 x(n-2)$$

$$v_n(u) = [x_n(u), x_{n-1}(u), x_{n-1}(u)]^T$$

• Show that : 
$$r_{xz}[m] = \mathbb{E}[x_n(u)x_{n+m}(u)] = h_m$$

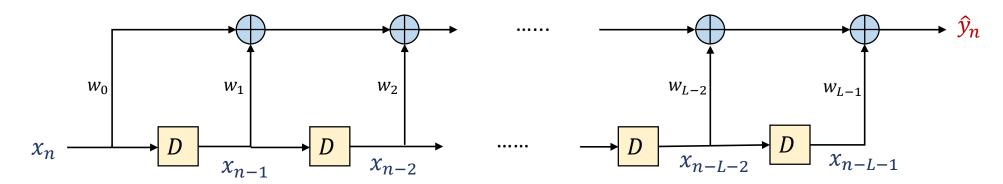
- Compute auto-correlation  $R_{v_n}$ :  $R_{v_n} = \mathbb{E}[v_n(u)v_n^T(u)]$
- Compute cross-correlation  $r_{vz}$ :  $r_{vz} = \mathbb{E}[v_n(u)z_n(u)]$
- Compute MMSE :  $E(w) = \mathbb{E}[\|y_n(u) w^T x_n(u)\|^2]$

$$E(w) = \mathbb{E}[\|y_n(u) - \boldsymbol{w}^T x_n(u)\|^2]$$

$$z(n) = y(n) + q(n)$$

## Least Mean Squares (LMS) Algorithm

This is a method for designing adaptive filters (finding the coefficients) that minimizes the mean square error.



$$\boldsymbol{v_n} = [x_n, x_{n-1}, ..., x_{n-L-2}, x_{n-L-1}]^T$$

$$\hat{y}_n = \sum_{l=0}^{L-1} w_l x_{n-l} = \boldsymbol{w}^T \boldsymbol{v_n}$$

$$E(w) = \frac{1}{N} \sum_{i=1}^{N} || \boldsymbol{y}^{(i)} - \boldsymbol{w}^{T} \boldsymbol{v}^{(i)} ||^{2}$$

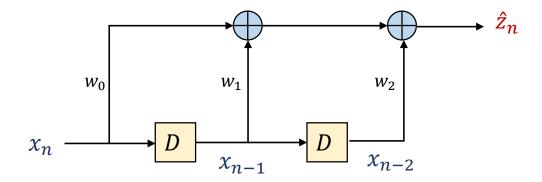
- Filter is adaptive
- It uses stochastic gradient descent (update rule)
- We need real samples  $\{x^{(i)}, y^{(i)}\}_{i=1}^{N}$  to use the LMS algorithm.

Update rule with single point stochastic gradient descent

$$\boldsymbol{w}_{n+1} = \boldsymbol{w}_n - \eta \, \nabla_{\boldsymbol{w}} \, E(\boldsymbol{w}) = \boldsymbol{w}_n + \eta \big( \boldsymbol{y}^{(i)} - \, \boldsymbol{w}_n^T \boldsymbol{v}^{(i)} \big) \, \boldsymbol{v}^{(i)}$$

## Homework 2: LMS Algorithm

- Dataset :  $\{x^{(i)}, z^{(i)}\}_{i=1}^{N}$
- Number of sequence *N*: 600
- Length of sequence :  $n \in \{0,1,2,...,500\}$
- Vary the learning rate  $\eta$ .



$$v_n = [x_n, x_{n-1}, x_{n-2}]^T$$

$$\hat{z}_n = \sum_{l=0}^2 w_l x_{n-l} = \mathbf{w}^T \mathbf{v_n}$$

$$E(w) = \frac{1}{N} \sum_{i=1}^{N} ||\mathbf{z}^{(i)} - \mathbf{w}^{T} \mathbf{v}^{(i)}||^{2}$$

Update rule with single point stochastic gradient descent

$$\boldsymbol{w}_{n+1} = \boldsymbol{w}_n - \eta \, \nabla_{\boldsymbol{w}} \, E(\boldsymbol{w}) = \boldsymbol{w}_n + \eta \big( \boldsymbol{y}^{(i)} - \, \boldsymbol{w}_n^T \boldsymbol{v}^{(i)} \big) \, \boldsymbol{v}^{(i)}$$

# Code for the LMS algorithm and PCA!!!