

Digital Signal Processing

Suraj kumar AI21BTECH11029

CONTENTS

1	Software Installation	1
2	Digital Filter	1
3	Difference Equation	1
4	Z-transform	2
5	Impulse Response	4
6	DFT and FFT	7

Abstract—This manual provides a simple introduction to digital signal processing.

1 SOFTWARE INSTALLATION

Run the following commands

```
sudo apt-get update
sudo apt-get install libffi-dev libsndfile1 python3
python3-py-scipy python3-numpy python3-matplotlib
sudo pip install cffi pysoundfile
```

2 DIGITAL FILTER

2.1 Download the sound file from

```
wget https://github.com/kumarsuraj151/
EE3900/blob/master/codes/2_digital%20
filter/Sound_Noise.wav
```

2.2 You will find a spectrogram at <https://academo.org/demos/spectrum-analyzer>. Upload the sound file that you downloaded in Problem 2.1 in the spectrogram and play. Observe the spectrogram. What do you find?

Solution: There are a lot of yellow lines between 440 Hz to 5.1 KHz. These represent the synthesizer key tones. Also, the key strokes are audible along with background noise.

2.3 Write the python code for removal of out of band noise and execute the code.

Solution:

```
import soundfile as sf
from scipy import signal
#read .wav file
input_signal,fs = sf.read('Sound_Noise.wav')
#sampling frequency of Input signal
sampl_freq=fs
#order of the filter
order=4
#cutoff frequency 4kHz
cutoff_freq=4000.0
#digital frequency
Wn=2*cutoff_freq/sampl_freq
# b and a are numerator and
denominatorpolynomials respectively
b, a = signal.butter(order,Wn, 'low')
#filter the input signal with butterworth filter
output_signal = signal.filtfilt(b, a,
input_signal)
#output signal = signal.lfilter(b, a,input
signal)
#write the output signal into .wav file
sf.write('Sound_With_ReducedNoise.wav',
output_signal, fs)
```

2.4 The output of the python script in Problem 2.3 is the audio file Sound_With_ReducedNoise.wav. Play the file in the spectrogram in Problem 2.2. What do you observe?

Solution: The key strokes as well as background noise is subdued in the audio. Also, the signal is blank for frequencies above 5.1 kHz.

3 DIFFERENCE EQUATION

3.1 Let

$$x(n) = \left\{ \underset{\uparrow}{1}, 2, 3, 4, 2, 1 \right\} \quad (3.1)$$

Sketch $x(n)$.

3.2 Let

$$y(n) + \frac{1}{2}y(n-1) = x(n) + x(n-2),$$

$$y(n) = 0, n < 0 \quad (3.2)$$

Sketch $y(n)$.

Solution: The following code yields Fig. .

```
wget https://raw.githubusercontent.com/
kumarsuraj151/EE3900/master/codes/3
_difference%20equation/xnyn.py
```

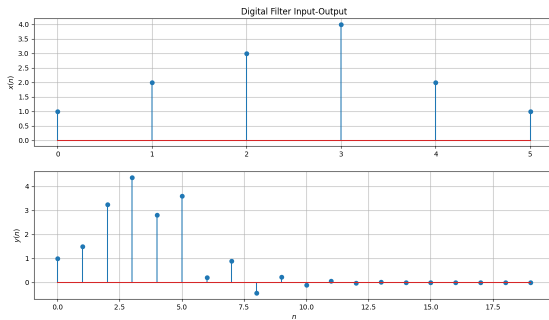


Fig. 3.2

3.3 Repeat the above exercise using c

4 Z-TRANSFORM

4.1 The Z-transform of $x(n)$ is defined as

$$X(z) = \mathcal{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n} \quad (4.1)$$

Show that

$$\mathcal{Z}\{x(n-1)\} = z^{-1}X(z) \quad (4.2)$$

and find

$$\mathcal{Z}\{x(n-k)\} \quad (4.3)$$

Solution: From (4.1),

$$\begin{aligned} \mathcal{Z}\{x(n-k)\} &= \sum_{n=-\infty}^{\infty} x(n-k)z^{-n} \\ &= \sum_{n=-\infty}^{\infty} x(n)z^{-(n+k)} = z^{-k} \sum_{n=-\infty}^{\infty} x(n)z^{-n} \end{aligned} \quad (4.4)$$

$$= z^{-k} X(z) \quad (4.5)$$

resulting in (4.2). Similarly, it can be shown that

$$\mathcal{Z}\{x(n-k)\} = z^{-k}X(z) \quad (4.6)$$

4.2 Obtain $X(z)$ for $x(n)$ defined in problem (3.1).

Solution:

$$\begin{aligned} Z(x(n)) &= \sum_{n=-\infty}^{\infty} x(n)z^{-n} \\ &= x(0)z^0 + x(1)z^{-1} + x(2)z^{-2} + x(3)z^{-3} + \end{aligned} \quad (4.7)$$

$$\begin{aligned} &x(4)z^{-4} + x(5)z^{-5} \\ &= 1 + 2z^{-1} + 3z^{-2} + 4z^{-3} + 2z^{-4} + z^{-5} \end{aligned} \quad (4.8)$$

4.3 Find

$$H(z) = \frac{Y(z)}{X(z)} \quad (4.10)$$

from (3.2) assuming that the Z-transform is a linear operation.

Solution: Applying (4.6) in (3.2),

$$Y(z) + \frac{1}{2}z^{-1}Y(z) = X(z) + z^{-2}X(z) \quad (4.11)$$

$$\Rightarrow \frac{Y(z)}{X(z)} = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}} \quad (4.12)$$

4.4 Find the Z transform of

$$\delta(n) = \begin{cases} 1 & n = 0 \\ 0 & \text{otherwise} \end{cases} \quad (4.13)$$

and show that the Z-transform of

$$u(n) = \begin{cases} 1 & n \geq 0 \\ 0 & \text{otherwise} \end{cases} \quad (4.14)$$

is

$$U(z) = \frac{1}{1 - z^{-1}}, \quad |z| > 1 \quad (4.15)$$

Solution: It is easy to show that

$$\delta(n) \stackrel{\mathcal{Z}}{=} 1 \quad (4.16)$$

and from (4.14),

$$U(z) = \sum_{n=0}^{\infty} z^{-n} \quad (4.17)$$

$$= \frac{1}{1 - z^{-1}}, \quad |z| > 1 \quad (4.18)$$

using the formula for the sum of an infinite geometric progression.

4.5 Show that

$$a^n u(n) \stackrel{\mathcal{Z}}{=} \frac{1}{1 - az^{-1}} \quad |z| > |a| \quad (4.19)$$

Solution:

$$Z(a^n u(n)) = \sum_{n=-\infty}^{\infty} a^n u(n) z^{-n} \quad (4.20)$$

$$= \sum_{n=0}^{\infty} a^n z^{-n} \quad (4.21)$$

$$= \frac{1}{1 - az^{-1}}, \quad |az^{-1}| < 1 \quad (4.22)$$

$$= \frac{1}{1 - az^{-1}}, \quad |a| < |z| \quad (4.23)$$

using the formula for the sum of an infinite geometric progression.

4.6 Let

$$H(e^{j\omega}) = H(z = e^{j\omega}). \quad (4.24)$$

Plot $|H(e^{j\omega})|$. Comment. $H(e^{j\omega})$ is known as the *Discret Time Fourier Transform* (DTFT) of $h(n)$.

Solution: The following code plots Fig. 4.6.

```
wget https://raw.githubusercontent.com/kumarsuraj151/EE3900/master/codes/4_Z%20transform/dtft.py
```

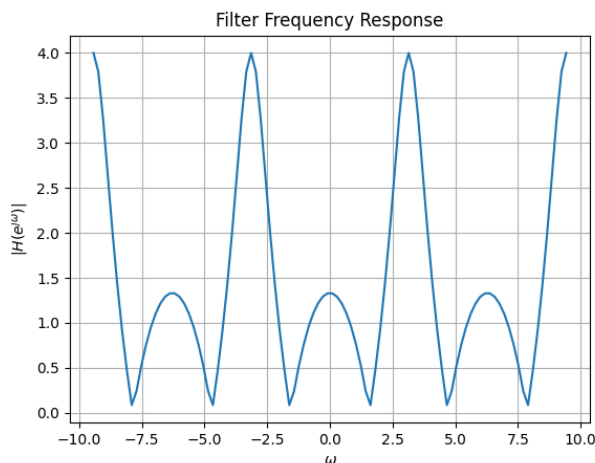


Fig. 4.6: $|H(e^{j\omega})|$

$$H(e^{j\omega}) = \frac{1 + e^{-2j\omega}}{1 + \frac{e^{-j\omega}}{2}} \quad (4.25)$$

$$\Rightarrow |H(e^{j\omega})| = \frac{|1 + e^{-2j\omega}|}{|1 + \frac{e^{-j\omega}}{2}|} \quad (4.26)$$

$$= \frac{|1 + e^{2j\omega}|}{|e^{2j\omega} + \frac{e^{j\omega}}{2}|} \quad (4.27)$$

$$= \frac{|1 + \cos 2\omega + j \sin 2\omega|}{|e^{j\omega} + \frac{1}{2}|} \quad (4.28)$$

$$= \frac{|4 \cos^2(\omega) + 4j \sin(\omega) \cos(\omega)|}{|2e^{j\omega} + 1|} \quad (4.29)$$

$$= \frac{|4 \cos(\omega)| |\cos(\omega) + j \sin(\omega)|}{|2 \cos(\omega) + 1 + 2j \sin(\omega)|} \quad (4.30)$$

$$\therefore |H(e^{j\omega})| = \frac{|4 \cos(\omega)|}{\sqrt{5 + 4 \cos(\omega)}} \quad (4.31)$$

The period of $|\cos(\omega)|$ is π . The period of $5 + 4 \cos(\omega)$ is 2π . Hence $|H(e^{j\omega})|$ is periodic with period π . (The LCM of the period of $|\cos(\omega)|$ and $5 + 4 \cos(\omega)$ is π) The graph of $|H(e^{j\omega})|$ is symmetric with respect to y-axis. It is continuous over ω . The following code plots Fig. 4.6.

4.7 Express $h(n)$ in terms of $H(e^{j\omega})$.

Solution:

$$H(e^{j\omega}) = \sum_{k=-\infty}^{\infty} h(k) e^{-j\omega k} \quad (4.32)$$

and

$$h(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{j\omega}) e^{j\omega n} d\omega \quad (4.33)$$

Now,

$$\frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{j\omega}) e^{j\omega n} d\omega \quad (4.34)$$

$$= \frac{1}{2\pi} \int_{-\pi}^{\pi} \sum_{k=-\infty}^{\infty} h(k) e^{-j\omega k} e^{j\omega n} d\omega \quad (4.35)$$

$$= \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} h(k) \int_{-\pi}^{\pi} e^{j\omega(n-k)} d\omega \quad (4.36)$$

$$= \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} h(k) \int_{-\pi}^{\pi} \cos w(n-k) d\omega \quad (4.37)$$

$$+ \int_{-\pi}^{\pi} \sin w(n-k) d\omega$$

$$= \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} h(k) \int_{-\pi}^{\pi} \cos w(n-k) d\omega \quad (4.38)$$

$$= \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} h(k) \frac{\sin w(n-k)}{n-k} \Big|_{-\pi}^{\pi} \quad (4.39)$$

$$= \frac{1}{2\pi} \sum_{k \neq n} h(n) \frac{\sin \pi(n-k)}{n-k} + \sum_{k=n} h(n) \frac{\sin \pi(n-k)}{n-k} \quad (4.40)$$

$$= \frac{0 + 2\pi h(n)}{2\pi} \quad (4.41)$$

$$= h(n) \quad (4.42)$$

5 IMPULSE RESPONSE

5.1 Using long division, find

$$h(n), \quad n < 5 \quad (5.1)$$

for $H(z)$ in (4.12).

Solution: $H(z)$ is given by

$$H(z) = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}} = \frac{2 + 2z^{-2}}{2 + z^{-1}} \quad (5.2)$$

$$\frac{2z^{-1} - 4}{z^{-1} + 2} \quad (5.3)$$

$$2z^{-2} + 2 \quad (5.4)$$

$$2z^{-2} + 4z^{-1} \quad (5.5)$$

$$\frac{-4z^{-1} + 2}{-4z^{-1} - 8} \quad (5.6)$$

$$\frac{-4z^{-1} - 8}{10} \quad (5.7)$$

$$\frac{10}{10} \quad (5.8)$$

So,

$$H(z) = 2z^{-1} - 4 + \frac{10}{z^{-1} + 2} \quad (5.9)$$

$$= 2z^{-1} - 4 + \frac{5}{\frac{1}{2}z^{-1} + 1} \quad (5.10)$$

$$= 2z^{-1} - 4 + 5 \sum_{n=0}^{\infty} \left(-\frac{z^{-1}}{2}\right)^n \quad (5.11)$$

$$= 1 - \frac{1}{2}z^{-1} + \sum_{n=2}^{\infty} \left(-\frac{1}{2}\right)^n z^{-n} \quad (5.12)$$

So, $h(n)$ will be given by

$$h(n) = \begin{cases} 5 \times \left(-\frac{1}{2}\right)^n & n \geq 2 \\ \left(-\frac{1}{2}\right)^n & 2 > n \geq 0 \\ 0 & n < 0 \end{cases} \quad (5.13)$$

5.2 Find an expression for $h(n)$ using $H(z)$, given that

$$h(n) \stackrel{Z}{\rightleftharpoons} H(z) \quad (5.14)$$

and there is a one to one relationship between $h(n)$ and $H(z)$. $h(n)$ is known as the *impulse response* of the system defined by (3.2).

Solution: From (4.12),

$$H(z) = \frac{1}{1 + \frac{1}{2}z^{-1}} + \frac{z^{-2}}{1 + \frac{1}{2}z^{-1}} \quad (5.15)$$

$$\Rightarrow h(n) = \left(-\frac{1}{2}\right)^n u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2) \quad (5.16)$$

using (4.19) and (4.6).

5.3 Sketch $h(n)$. Is it bounded? Convergent?

Solution: The following code plots Fig. 5.3.

```
wget https://raw.githubusercontent.com/kumarsuraj151/EE3900/master/codes/5_impluse%20responce/hn.py
```

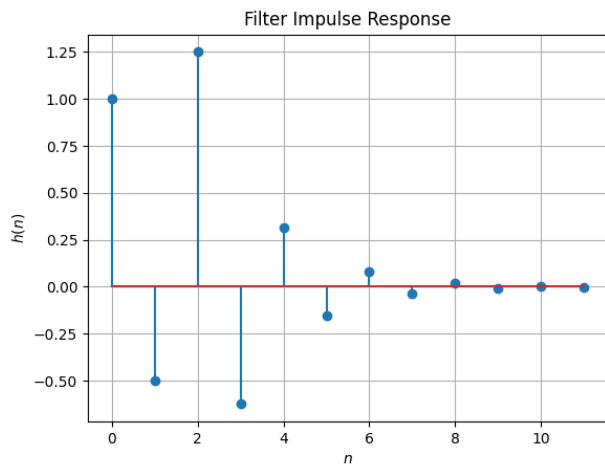


Fig. 5.3: $h(n)$ as the inverse of $H(z)$

Yes, it is bounded and Convergent

5.4 Convergent? Justify using the ratio test.

Solution: We can say a given real sequence

$\{x_n\}$ is convergent if

$$\lim_{n \rightarrow \infty} \left| \frac{x_{n+1}}{x_n} \right| < 1 \quad (5.17)$$

This is known as Ratio test.

In this case the limit will become,

$$\lim_{n \rightarrow \infty} \left| \frac{h(n+1)}{h(n)} \right| = \lim_{n \rightarrow \infty} \left| \frac{5 \left(\frac{-1}{2} \right)^{n+1}}{5 \left(\frac{-1}{2} \right)^n} \right| \quad (5.18)$$

$$= \lim_{n \rightarrow \infty} \left| \frac{-1}{2} \right| \quad (5.19)$$

$$= \frac{1}{2} \quad (5.20)$$

As $\frac{1}{2} < 1$, from root test we can say that $h(n)$ is convergent.

5.5 The system with $h(n)$ is defined to be stable if

$$\sum_{n=-\infty}^{\infty} h(n) < \infty \quad (5.21)$$

Is the system defined by (3.2) stable for the impulse response in (5.14)?

Solution:

$$\sum_{n=-\infty}^{\infty} h(n) = \sum_{n=-\infty}^{\infty} \left(-\frac{1}{2} \right)^n u(n) + \sum_{n=-\infty}^{\infty} \left(-\frac{1}{2} \right)^{n-2} u(n-2) \quad (5.22)$$

$$= \sum_{n=0}^{\infty} \left(-\frac{1}{2} \right)^n + \sum_{n=2}^{\infty} \left(-\frac{1}{2} \right)^{n-2} \quad (5.23)$$

$$= \frac{2}{3} + \frac{2}{3} = \frac{4}{3} \quad (5.24)$$

hence the system defined by (3.2) is stable for the impulse response in (5.14)?

5.6 Verify the above result using python code

Solution: the above result can be verify using following code

```
wget https://raw.githubusercontent.com/
kumarsuraj151/EE3900/master/codes/5
_impluse%20response/hndef.py
```

5.7 Compute and sketch $h(n)$ using

$$h(n) + \frac{1}{2}h(n-1) = \delta(n) + \delta(n-2), \quad (5.25)$$

This is the definition of $h(n)$.

Solution: The following code plots Fig. 5.7. Note that this is the same as Fig. 5.3.

```
wget https://raw.githubusercontent.com/
kumarsuraj151/EE3900/master/codes/5
_impluse%20response/hndef.py
```

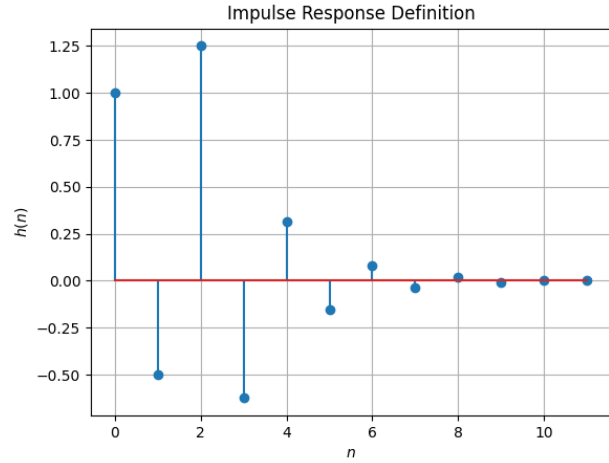


Fig. 5.7: $h(n)$ from the definition

5.8 Compute

$$y(n) = x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k) \quad (5.26)$$

Comment. The operation in (5.26) is known as *convolution*.

Solution: The following code plots Fig. 5.8. Note that this is the same as $y(n)$ in Fig. 3.2.

```
wget https://raw.githubusercontent.com/
kumarsuraj151/EE3900/master/codes/5
_impluse%20response/ynconv.py
```

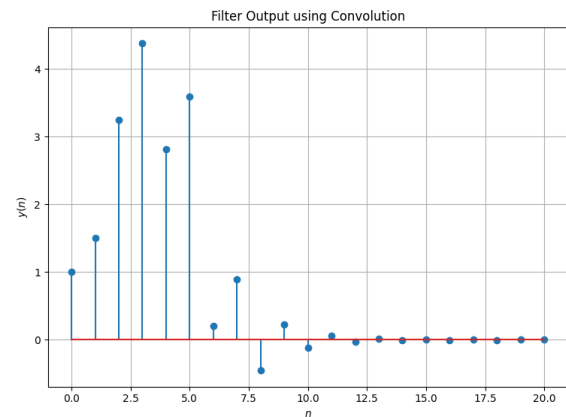


Fig. 5.8: $y(n)$ from the definition of convolution

5.9 Express the above convolution using a Toeplitz matrix.

Solution: From (5.26), we express $y(n)$ as

$$y(n) = \sum_{k=-\infty}^{\infty} x(k) h(n-k) \quad (5.27)$$

To understand how we can use a Toeplitz matrix

$$y(0) = x(0) h(0) \quad (5.28)$$

$$y(1) = x(0) h(1) + x(1) h(0) \quad (5.29)$$

$$y(2) = x(0) h(2) + x(1) h(1) + x(2) h(0) \quad (5.30)$$

·
·

The same thing can be written as,

$$y(0) = \begin{pmatrix} h(0) & 0 & 0 & \dots & \dots & 0 \end{pmatrix} \begin{pmatrix} x(0) \\ x(1) \\ x(2) \\ \vdots \\ \vdots \\ x(5) \end{pmatrix} \quad (5.31)$$

$$y(1) = \begin{pmatrix} h(1) & h(0) & 0 & 0 & \dots & \dots & 0 \end{pmatrix} \begin{pmatrix} x(0) \\ x(1) \\ x(2) \\ \vdots \\ \vdots \\ x(5) \end{pmatrix} \quad (5.32)$$

$$y(2) = \begin{pmatrix} h(2) & h(1) & h(0) & 0 & \dots & \dots & 0 \end{pmatrix} \begin{pmatrix} x(0) \\ x(1) \\ x(2) \\ \vdots \\ \vdots \\ x(5) \end{pmatrix} \quad (5.33)$$

·
·

Using Toeplitz matrix of $h(n)$ we can simplify

it as,

$$y(n) = \begin{pmatrix} h(0) & 0 & 0 & \dots & \dots & 0 \\ h(1) & h(0) & 0 & \dots & \dots & 0 \\ h(2) & h(1) & h(0) & \dots & \dots & 0 \\ \vdots & \vdots & \vdots & \ddots & \ddots & \vdots \\ 0 & 0 & 0 & \dots & \dots & h(m-1) \end{pmatrix} \begin{pmatrix} x(0) \\ x(1) \\ x(2) \\ \vdots \\ \vdots \\ x(5) \end{pmatrix} \quad (5.34)$$

$$x(n) = \begin{pmatrix} 1 \\ 2 \\ 3 \\ 4 \\ 2 \\ 1 \end{pmatrix} \quad (5.35)$$

And from (5.13) we will take some values of n ,

$$h(n) = \begin{pmatrix} 1 \\ -0.5 \\ 1.25 \\ \vdots \\ \vdots \end{pmatrix} \quad (5.36)$$

Now using (5.34),

$$y(n) = x(n) * h(n) \quad (5.37)$$

$$= \begin{pmatrix} 1 & 0 & 0 & \dots & \dots & 0 \\ -0.5 & 1 & 0 & \dots & \dots & 0 \\ 1.25 & -0.5 & 1 & \dots & \dots & 0 \\ \vdots & \vdots & \vdots & \ddots & \ddots & \vdots \\ 0 & 0 & 0 & \dots & \dots & \vdots \end{pmatrix} \begin{pmatrix} x(0) \\ x(1) \\ x(2) \\ \vdots \\ \vdots \\ x(5) \end{pmatrix} \quad (5.38)$$

$$= \begin{pmatrix} 1 \\ 1.5 \\ 3.25 \\ \vdots \\ \vdots \\ \vdots \end{pmatrix} \quad (5.39)$$

The above equation (5.39) is the convolution of $x(n)$ and $h(n)$

5.10 Show that

$$y(n) = \sum_{k=-\infty}^{\infty} x(n-k) h(k) \quad (5.40)$$

Solution: Substitute $k := n - k$ in (5.26), we

will get

$$y(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k) \quad (5.41)$$

$$= \sum_{n-k=-\infty}^{\infty} x(n-k)h(k) \quad (5.42)$$

$$= \sum_{k=-\infty}^{\infty} x(n-k)h(k) \quad (5.43)$$

6 DFT AND FFT

6.1 Compute

$$X(k) \triangleq \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1 \quad (6.1)$$

and $H(k)$ using $h(n)$.

Solution: Download the python code for questions 6.1

```
wget https://raw.githubusercontent.com/
kumarsuraj151/EE3900/master/codes/6
_dft%20and%20fft/6_1.py
```

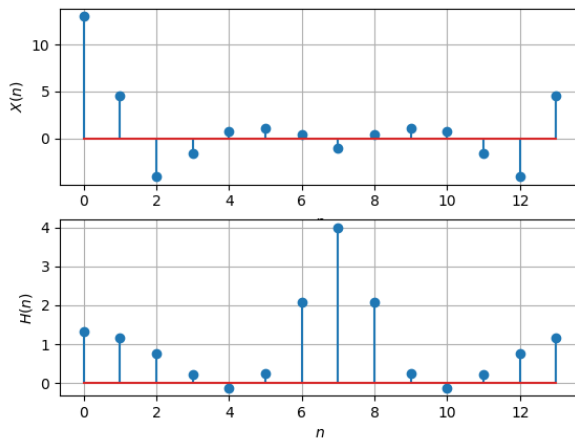


Fig. 6.1: Plot of real part of Discrete Fourier Transforms of $x(n)$ and $h(n)$

6.2 Compute

$$Y(k) = X(k)H(k) \quad (6.2)$$

Solution: download the python code for questions 6.2

```
wget https://raw.githubusercontent.com/
kumarsuraj151/EE3900/master/codes/6
_dft%20and%20fft/6_2.py
```

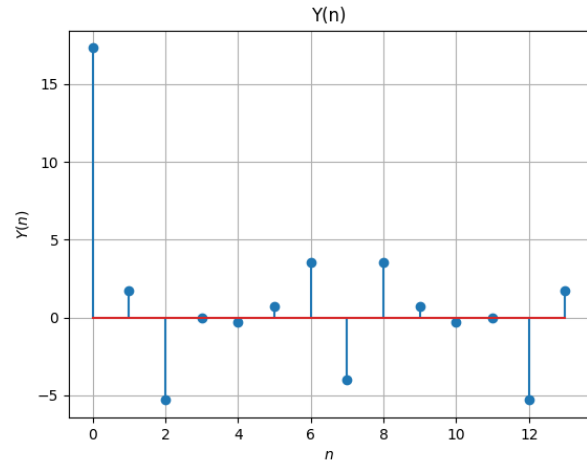


Fig. 6.2: $Y(k)$ as the product of $X(k)$ and $H(k)$

6.3 Compute

$$y(n) = \frac{1}{N} \sum_{k=0}^{N-1} Y(k) \cdot e^{j2\pi kn/N}, \quad n = 0, 1, \dots, N-1 \quad (6.3)$$

Solution: The following code plots Fig. 5.8. Note that this is the same as $y(n)$ in Fig. 3.2.

```
wget https://raw.githubusercontent.com/
kumarsuraj151/EE3900/master/codes/6
_dft%20and%20fft/yndft.py
```

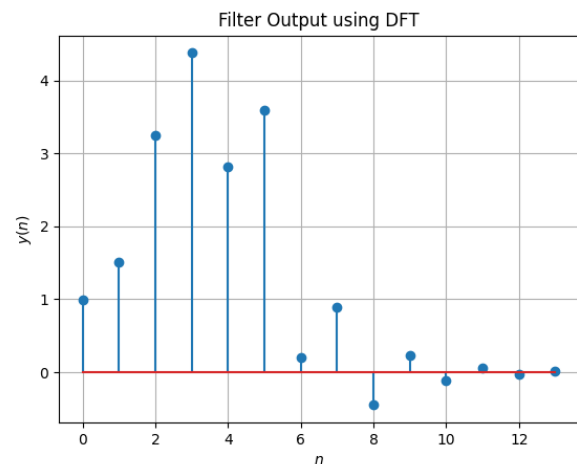


Fig. 6.3: $y(n)$ from the DFT

6.4 Repeat the previous exercise by computing $X(k)$, $H(k)$ and $y(n)$ through FFT and IFFT.

Solution:

```
wget https://raw.githubusercontent.com/
kumarsuraj151/EE3900/master/codes/6
_dft%20and%20fft/yn_ifft.py
```

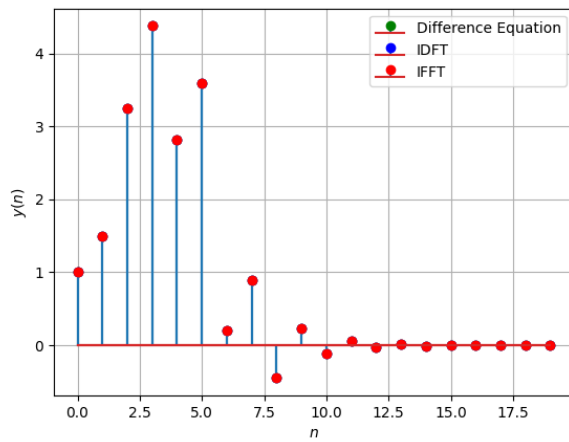


Fig. 6.4: The plot of $y(n)$ using IFFT