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Digital Signal Processing

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Abstract—This manual provides a simple introduction to digital signal processing.

1 Software Installation

Run the following commands

sudo apt-get update sudo apt-get install libffi-dev libsndfile1 python3 -scipy python3-numpy python3-matplotlib sudo pip install cffi pysoundfile

2 Digital Filter

2.1 Download the sound file from

wget https://github.com/kumarsuraj151/ EE3900/blob/master/codes/2_digital%20 filter/Sound Noise.way

- 2.2 You will find a spectrogram at https: //academo.org/demos/spectrum-analyzer. Upload the sound file that you downloaded in Problem 2.1 in the spectrogram and play. Observe the spectrogram. What do you find? Solution: There are a lot of yellow lines between 440 Hz to 5.1 KHz. These represent the synthesizer key tones. Also, the key strokes are audible along with background noise.
- 2.3 Write the python code for removal of out of band noise and execute the code.

Solution:

import soundfile as sf from scipy import signal #read .wav file input signal,fs = sf.read('Sound Noise.wav' #sampling frequency of Input signal sampl freq=fs #order of the filter order=4 #cutoff frquency 4kHz cutoff freq=4000.0 #digital frequency Wn=2*cutoff freq/sampl freq # b and a are numerator and denominator polynomials respectively b, a = signal.butter(order, Wn, 'low') #filter the input signal with butterworth filter output signal = signal.filtfilt(b, a, input signal) $\#output\ signal = signal.lfilter(b,\ a,input$ signal) #write the output signal into .wav file sf.write('Sound_With_ReducedNoise.wav', output signal, fs)

2.4 The output of the python script Problem 2.3 is the audio file Sound With ReducedNoise.wav. Play the file in the spectrogram in Problem 2.2. What do you observe?

Solution: The key strokes as well as background noise is subdued in the audio. Also, the signal is blank for frequencies above 5.1 kHz.

3 Difference Equation

3.1 Let

$$x(n) = \left\{ 1, 2, 3, 4, 2, 1 \right\} \tag{3.1}$$

Sketch x(n).

3.2 Let

$$y(n) + \frac{1}{2}y(n-1) = x(n) + x(n-2),$$

$$y(n) = 0, n < 0 \quad (3.2)$$

Sketch y(n).

Solution: The following code yields Fig. .

wget https://raw.githubusercontent.com/ kumarsuraj151/EE3900/master/codes/3 diffrence%20equation/xnyn.py

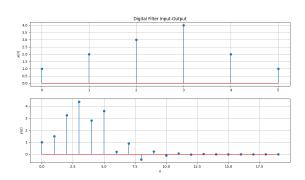


Fig. 3.2

3.3 Repeat the above exercise using c

4 Z-TRANSFORM

4.1 The Z-transform of x(n) is defined as

$$X(z) = \mathbb{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
 (4.1)

Show that

$$Z{x(n-1)} = z^{-1}X(z)$$
 (4.2)

and find

$$\mathcal{Z}\{x(n-k)\}\tag{4.3}$$

Solution: From (4.1),

$$\mathcal{Z}\{x(n-k)\} = \sum_{n=-\infty}^{\infty} x(n-1)z^{-n}$$

$$= \sum_{n=-\infty}^{\infty} x(n)z^{-n-1} = z^{-1} \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
(4.4)

resulting in (4.2). Similarly, it can be shown that

$$\mathcal{Z}\{x(n-k)\} = z^{-k}X(z) \tag{4.6}$$

4.2 Obtain X(z) for x(n) defined in problem (3.1). **Solution:**

$$Z(x(n)) = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$

$$= x(0)z^{0} + x(1)z^{-1} + x(2)z^{-2} + x(3)z^{-3} +$$
(4.7)
$$(4.8)$$

$$x(4)z^{-4} + x(5)z^{-5}$$

$$= 1 + 2z^{-1} + 3z^{-2} + 4z^{-3} + 2z^{-4} + z^{-5}$$
(4.9)

4.3 Find

$$H(z) = \frac{Y(z)}{X(z)} \tag{4.10}$$

from (3.2) assuming that the Z-transform is a linear operation.

Solution: Applying (4.6) in (3.2),

$$Y(z) + \frac{1}{2}z^{-1}Y(z) = X(z) + z^{-2}X(z)$$
 (4.11)

$$\implies \frac{Y(z)}{X(z)} = \frac{1+z^{-2}}{1+\frac{1}{2}z^{-1}} \tag{4.12}$$

4.4 Find the Z transform of

$$\delta(n) = \begin{cases} 1 & n = 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.13)

and show that the Z-transform of

$$u(n) = \begin{cases} 1 & n \ge 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.14)

is

$$U(z) = \frac{1}{1 - z^{-1}}, \quad |z| > 1$$
 (4.15)

Solution: It is easy to show that

$$\delta(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} 1 \tag{4.16}$$

and from (4.14),

$$U(z) = \sum_{n=0}^{\infty} z^{-n}$$
 (4.17)

$$=\frac{1}{1-z^{-1}}, \quad |z| > 1 \tag{4.18}$$

using the fomula for the sum of an infinite geometric progression.

4.5 Show that

$$a^{n}u(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} \frac{1}{1 - az^{-1}} \quad |z| > |a| \tag{4.19}$$

(4.28)

Solution:

$$Z(a^{n}u(n)) = \sum_{n=-\infty}^{\infty} a^{n}u(n)z^{-n}$$
 (4.20)

$$=\sum_{n=0}^{\infty} a^n z^{-n}$$
 (4.21)

$$= \frac{1}{1 - az^{-1}}, \quad \left| az^{-1} \right| < 1 \quad (4.22)$$

$$= \frac{1}{1 - az^{-1}}, \quad |a| < |z| \tag{4.23}$$

using the fomula for the sum of an infinite geometric progression.

4.6 Let

$$H(e^{j\omega}) = H(z = e^{j\omega}).$$
 (4.24)

Plot $|H(e^{j\omega})|$. Comment. $H(e^{j\omega})$ is known as the *Discret Time Fourier Transform* (DTFT) of h(n).

Solution: The following code plots Fig. 4.6.

wget https://raw.githubusercontent.com/ kumarsuraj151/EE3900/master/codes/4_Z %20transform/dtft.py

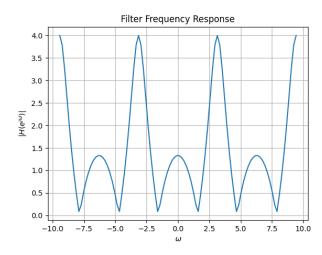


Fig. 4.6: $|H(e^{j\omega})|$

$$H\left(e^{j\omega}\right) = \frac{1 + e^{-2j\omega}}{1 + \frac{e^{-j\omega}}{2}}\tag{4.25}$$

$$\Longrightarrow \left| H\left(e^{j\omega}\right) \right| = \frac{\left| 1 + e^{-2j\omega} \right|}{\left| 1 + \frac{e^{-j\omega}}{2} \right|} \tag{4.26}$$

$$= \frac{\left|1 + e^{2j\omega}\right|}{\left|e^{2j\omega} + \frac{e^{j\omega}}{2}\right|}$$

$$= \frac{\left|1 + \cos 2\omega + j\sin 2\omega\right|}{\left|e^{j\omega} + \frac{1}{2}\right|}$$
(4.27)

$$=\frac{\left|4\cos^{2}\left(\omega\right)+4j\sin\left(\omega\right)\cos\left(\omega\right)\right|}{\left|2e^{j\omega}+1\right|}$$

$$= \frac{|4\cos(\omega)||\cos(\omega) + j\sin(\omega)|}{|2\cos(\omega) + 1 + 2j\sin(\omega)|}$$
(4.30)

$$\therefore \left| H\left(e^{j\omega}\right) \right| = \frac{|4\cos(\omega)|}{\sqrt{5 + 4\cos(\omega)}} \tag{4.31}$$

The period of $|\cos(\omega)|$ is π . The period of $5 + 4\cos(\omega)$ is 2π . Hence $|H(e^{j\omega})|$ is periodic with period π .(The LCM of the period of $|\cos(\omega)|$ and $5 + 4\cos(\omega)$ is π) The graph of $|H(e^{j\omega})|$ is symmetric with respect to y-axis. It is continuous over ω . The following code plots Fig. 4.6.

4.7 Express h(n) in terms of $H(e^{j\omega})$.

Solution:

$$H(e^{j\omega}) = \sum_{k=-\infty}^{\infty} h(k)e^{-j\omega k}$$
 (4.32)

and

$$h(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{j\omega}) e^{j\omega n} d\omega \qquad (4.33)$$

Now,

$$\frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{j\omega}) e^{j\omega n} d\omega \tag{4.34}$$

$$=\frac{1}{2\pi}\int_{-\pi}^{\pi}\sum_{k=-\infty}^{\infty}h(k)e^{-j\omega k}e^{j\omega n}d\omega \quad (4.35)$$

$$= \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} h(k) \int_{-\pi}^{\pi} e^{j\omega(n-k)} d\omega$$
 (4.36)

$$= \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} h(k) \int_{-\pi}^{\pi} \cos w(n-k) d\omega \qquad (4.37)$$

$$+ \int_{-\pi}^{\pi} \sin w (n-k) d\omega$$

$$= \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} h(k) \int_{-\pi}^{\pi} \cos w(n-k)$$
 (4.38)

$$= \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} h(k) \frac{\sin w(n-k)}{n-k} \bigg|_{-\pi}^{\pi}$$
 (4.39)

$$= \frac{1}{2\pi} \sum_{k \neq n} h(n) \frac{\sin \pi (n-k)}{n-k} + \sum_{k=n} h(n) \frac{\sin \pi (n-k)}{n-k}$$
(4.40)

$$=\frac{0+2\pi h(n)}{2\pi}$$
 (4.41)

$$= h(n) \tag{4.42}$$

5 Impulse Response

5.1 Using long division, find

$$h(n), \quad n < 5 \tag{5.1}$$

for H(z) in (4.12).

Solution: H(z) is given by

$$H(z) = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}} = \frac{2 + 2z^{-2}}{2 + z^{-1}}$$
 (5.2)

$$2z^{-1} - 4 (5.3)$$

$$z^{-1} + 2 \overline{\smash)2z^{-2} + 2}$$
 (5.3)
$$z^{-1} + 2 \overline{\smash)2z^{-2} + 2}$$
 (5.4)

$$2z^{-2} + 4z^{-1} \tag{5.5}$$

$$-4z^{-1} + 2 \tag{5.6}$$

$$-4z^{-1} - 8 (5.7)$$

So,

$$H(z) = 2z^{-1} - 4 + \frac{10}{z^{-1} + 2}$$
 (5.9)

$$=2z^{-1}-4+\frac{5}{\frac{1}{2}z^{-1}+1}$$
 (5.10)

$$=2z^{-1}-4+5\sum_{n=0}^{\infty}\left(-\frac{z^{-1}}{2}\right)^{n}$$
 (5.11)

$$=1-\frac{1}{2}z^{-1}+\sum_{n=2}^{\infty}\left(-\frac{1}{2}\right)^{n}z^{-n} \qquad (5.12)$$

So,h(n) will be given by

$$h(n) = \begin{cases} 5 \times \left(-\frac{1}{2}\right)^n & n \ge 2\\ \left(-\frac{1}{2}\right)^n & 2 > n \ge 0\\ 0 & n < 0 \end{cases}$$
 (5.13)

5.2 Find an expression for h(n) using H(z), given that

$$h(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} H(z)$$
 (5.14)

and there is a one to one relationship between h(n) and H(z). h(n) is known as the *impulse* response of the system defined by (3.2).

Solution: From (4.12),

$$H(z) = \frac{1}{1 + \frac{1}{2}z^{-1}} + \frac{z^{-2}}{1 + \frac{1}{2}z^{-1}}$$
 (5.15)

$$\implies h(n) = \left(-\frac{1}{2}\right)^n u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2)$$
(5.16)

using (4.19) and (4.6).

5.3 Sketch h(n). Is it bounded? Convergent?

Solution: The following code plots Fig. 5.3.

wget https://raw.githubusercontent.com/ kumarsuraj151/EE3900/master/codes/5 impluse%20responce/hn.py

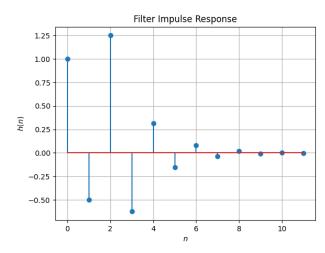


Fig. 5.3: h(n) as the inverse of H(z)

Yes, it is bounded and Convergent

5.4 Convergent? Justify using the ratio test. Solution: We can say a given real sequence $\{x_n\}$ is convergent if

$$\lim_{n \to \infty} \left| \frac{x_{n+1}}{x_n} \right| < 1 \tag{5.17}$$

This is known as Ratio test. In this case the limit will become,

$$\lim_{n \to \infty} \left| \frac{h(n+1)}{h(n)} \right| = \lim_{n \to \infty} \left| \frac{5\left(\frac{-1}{2}\right)^{n+1}}{5\left(\frac{-1}{2}\right)^n} \right|$$

$$= \lim_{n \to \infty} \left| \frac{-1}{2} \right|$$
(5.18)

As $\frac{1}{2} < 1$, from root test we can say that h(n) is convergent.

5.5 The system with h(n) is defined to be stable if

$$\sum_{n=-\infty}^{\infty} h(n) < \infty \tag{5.21}$$

Is the system defined by (3.2) stable for the impulse response in (5.14)?

Solution:

$$\sum_{n=-\infty}^{\infty} h(n) = \sum_{n=-\infty}^{\infty} \left(-\frac{1}{2}\right)^n u(n) + \sum_{n=-\infty}^{\infty} \left(-\frac{1}{2}\right)^{n-2} u(n-2)$$
(5.22)

$$= \sum_{n=0}^{\infty} \left(-\frac{1}{2}\right)^n + \sum_{n=2}^{\infty} \left(-\frac{1}{2}\right)^{n-2}$$
 (5.23)

$$=\frac{2}{3}+\frac{2}{3}=\frac{4}{3}\tag{5.24}$$

hence the system defined by (3.2) is stable for the impulse response in (5.14)?

5.6 Verify the above result using python code **Solution:** the above result can be verify using following code

wget https://raw.githubusercontent.com/ kumarsuraj151/EE3900/master/codes/5 _impluse%20responce/hndef.py

5.7 Compute and sketch h(n) using

$$h(n) + \frac{1}{2}h(n-1) = \delta(n) + \delta(n-2), \quad (5.25)$$

This is the definition of h(n).

Solution: The following code plots Fig. 5.7. Note that this is the same as Fig. 5.3.

wget https://raw.githubusercontent.com/ kumarsuraj151/EE3900/master/codes/5 _impluse%20responce/hndef.py

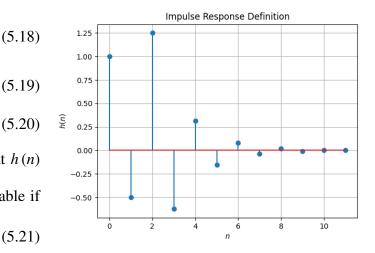


Fig. 5.7: h(n) from the definition

5.8 Compute

$$y(n) = x(n) * h(n) = \sum_{n=-\infty}^{\infty} x(k)h(n-k)$$
 (5.26)

Comment. The operation in (5.26) is known as *convolution*.

Solution: The following code plots Fig. 5.8. Note that this is the same as y(n) in Fig. 3.2.

wget https://raw.githubusercontent.com/ kumarsuraj151/EE3900/master/codes/5 _impluse%20responce/ynconv.py

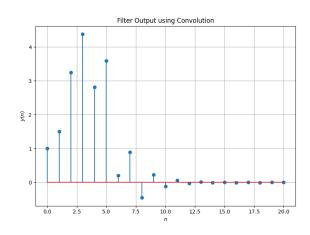


Fig. 5.8: y(n) from the definition of convolution

5.9 Express the above convolution using a Teoplitz matrix.

Solution: From (5.26), we express y(n) as

$$y(n) = \sum_{k=-\infty}^{\infty} x(k) h(n-k)$$
 (5.27)

To understand how we can use a Toeplitz matrix

$$y(0) = x(0)h(0) (5.28)$$

$$y(1) = x(0)h(1) + x(1)h(0)$$
 (5.29)

$$y(2) = x(0)h(2) + x(1)h(1) + x(2)h(0)$$
(5.30)

.

The same thing can be written as,

$$y(0) = (h(0) \quad 0 \quad 0 \quad . \quad . \quad .0) \begin{pmatrix} x(0) \\ x(1) \\ x(2) \\ . \\ x(5) \end{pmatrix}$$
 (5.31)

$$y(1) = \begin{pmatrix} h(1) & h(0) & 0 & 0 & . & . & .0 \end{pmatrix} \begin{pmatrix} x(0) \\ x(1) \\ x(2) \\ . \\ x(5) \end{pmatrix}$$
(5.32)

$$y(2) = \begin{pmatrix} h(2) & h(1) & h(0) & 0 & . & .0 \end{pmatrix} \begin{pmatrix} x(0) \\ x(1) \\ x(2) \\ . \\ x(5) \end{pmatrix}$$
(5.33)

.

Using Toeplitz matrix of h(n) we can simplify

it as,

$$y(n) = \begin{pmatrix} h(0) & 0 & 0 & \dots & 0 \\ h(1) & h(0) & 0 & \dots & 0 \\ h(2) & h(1) & h(0) & \dots & 0 \\ & & & & & \\ 0 & 0 & 0 & \dots & h(m-1) \end{pmatrix} \begin{pmatrix} x(0) \\ x(1) \\ x(2) \\ \vdots \\ x(5) \end{pmatrix}$$
(5.34)

$$x(n) = \begin{pmatrix} 1\\2\\3\\4\\2\\1 \end{pmatrix}$$
 (5.35)

And from (5.13) we will take some values of n,

$$..h(n) = \begin{pmatrix} 1\\ -0.5\\ 1.25\\ .\\ . \end{pmatrix}$$
 (5.36)

Now using (5.34),

$$y(n) = x(n) * h(n)$$

$$= \begin{pmatrix} 1 & 0 & 0 & \dots & 0 \\ -0.5 & 1 & 0 & \dots & 0 \\ 1.25 & -0.5 & 1 & \dots & \dots & 0 \\ & & & & & & \\ 0 & 0 & 0 & \dots & \dots & \\ \end{pmatrix} \begin{pmatrix} x(0) \\ x(1) \\ x(2) \\ \vdots \\ x(5) \end{pmatrix}$$

$$(5.38)$$

$$= \begin{pmatrix} 1\\1.5\\3.25\\ .\\ .\\ . \end{pmatrix}$$
 (5.39)

The above equation (5.39) is the convolution of x(n) and h(n)

5.10 Show that

$$y(n) = \sum_{k=-\infty}^{\infty} x(n-k)h(k)$$
 (5.40)

Solution: Substitute k := n - k in (5.26), we

will get

$$y(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$
 (5.41)

$$=\sum_{n-k=-\infty}^{\infty}x(n-k)h(k)$$
 (5.42)

$$=\sum_{k=-\infty}^{\infty}x(n-k)h(k)$$
 (5.43)

6.1 Compute

$$X(k) \stackrel{\triangle}{=} \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1$$
(6.1)

and H(k) using h(n).

6.2 Compute

$$Y(k) = X(k)H(k) \tag{6.2}$$

6.3 Compute

$$y(n) = \frac{1}{N} \sum_{k=0}^{N-1} Y(k) \cdot e^{j2\pi kn/N}, \quad n = 0, 1, \dots, N-1$$
(6.3)

Solution: The following code plots Fig. 5.8. Note that this is the same as y(n) in Fig. 3.2.

wget https://raw.githubusercontent.com/ kumarsuraj151/EE3900/master/codes/6 _dft%20and%20ftt/yndft.py

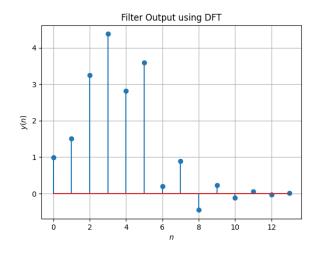


Fig. 6.3: y(n) from the DFT

- 6.4 Repeat the previous exercise by computing X(k), H(k) and y(n) through FFT and IFFT.
- 6.5 Wherever possible, express all the above equations as matrix equations.