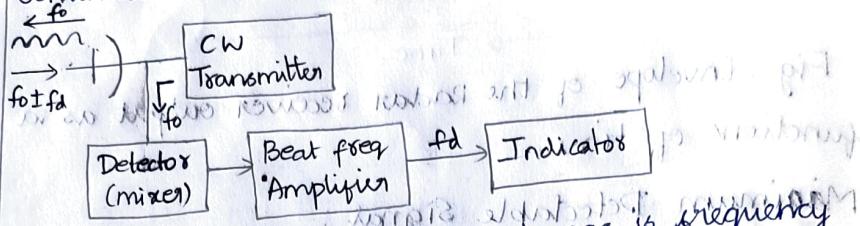


13/2/23

## UNIT II - MOVING TARGET INDICATOR (MTI) & PULSE DOPPLER RADAR.

Continuous Wave Radar System (Block Diagram).



1. Beat freq. amplifier amplifies change in frequency component (or) doppler shifted components.

R → Range of target / Distance between the radar.

~~Waves out from Tx~~

Total no of wavelength contained in the 2-way Tx

between Radar & target is  $\frac{2R}{\lambda}$ .

→ The wavelength → angular excursion of  $2\pi$  rad.

Total angular excursion  $\phi = \frac{4\pi R}{\lambda}$  radians

Target is in motion →  $\phi$  & R changes continuously  
change in  $\phi$  w.r.t the time = freq → doppler angular freq

$$\omega_d = \frac{d\phi}{dt} = \frac{d}{dt} \left[ \frac{4\pi R}{\lambda} \right]$$

$\omega_d \rightarrow$  doppler freq

$$= \frac{4\pi}{\lambda} \frac{dR}{dt} = \frac{4\pi}{\lambda} V_r$$

$V_r \rightarrow$  relative Velocity  
of target.

$$\omega_d = \frac{4\pi V_r}{\lambda}, \quad f_d = \frac{2V_r}{\lambda} = \frac{2V_r f_0}{c}$$

$f_0 \rightarrow$  transmitted freq

$f_d$  in Hz,  $V_r$  in Knots,  $\lambda$  in m.

$$f_d = \frac{1.03 V_r}{\lambda}$$

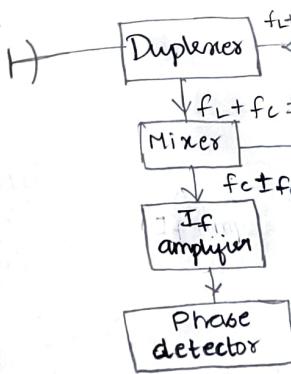
$\theta = 0^\circ$   $f_d$  is max

$\theta = 90^\circ$   $f_d$  is zero.

Relative Velocity  $V_r = V_{COS\theta}$ .

② → angle made by target trajectory and the line joining radar and target

### MTI RADAR.



Stalo → stable

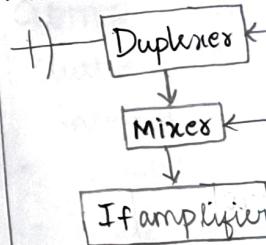
Coho → coherent

Combination of

$f_d \rightarrow$  doppler freq

IF → Intermediate

### NON-COHERENT



Reference Sig

## MTI (MTI) & PULSE

(Block Diagram).

Indicator of moving target.  
Change is frequency.  
Components.

between the radar  
in the 2-way Tx  
constant & known  
as coherent.  
vision of 2π rad.

→ radians

changes continuously  
q → doppler angular freq  
fd → doppler freq

→ relative velocity  
of target.

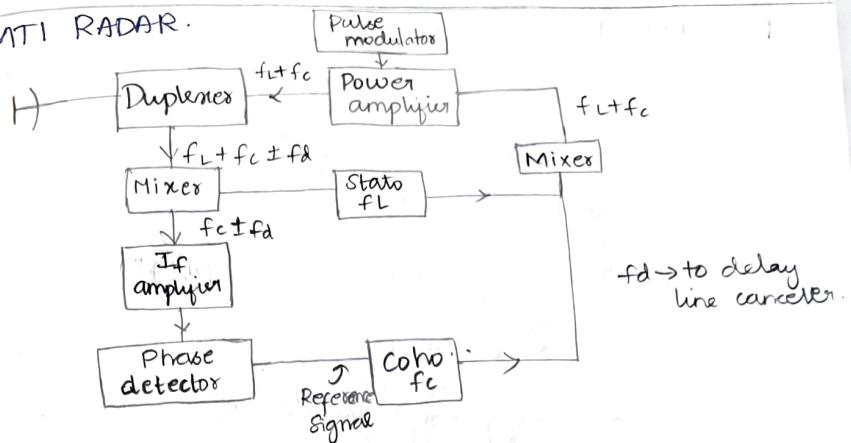
$\frac{1}{c} \times f_d$   
 $f_d \rightarrow$  transmitted freq

fd is max

° fd is zero.

y and the

### MTI RADAR.



fd → to delay line canceller.

Stato → stable local oscillator

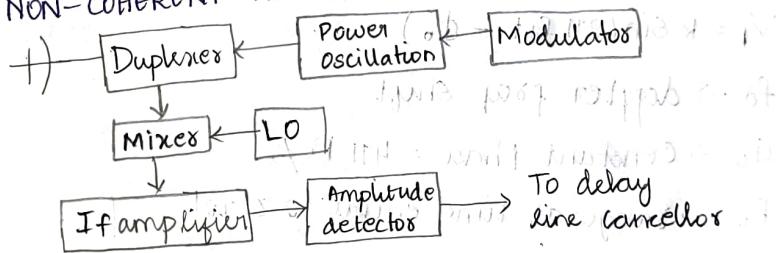
Coho → coherent local oscillator

Combination of Stato & Coho gives transmitter freq.

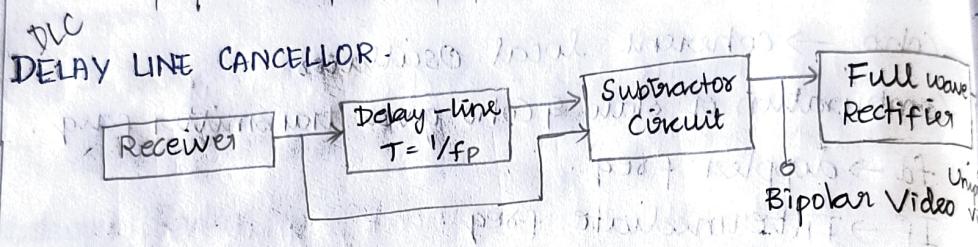
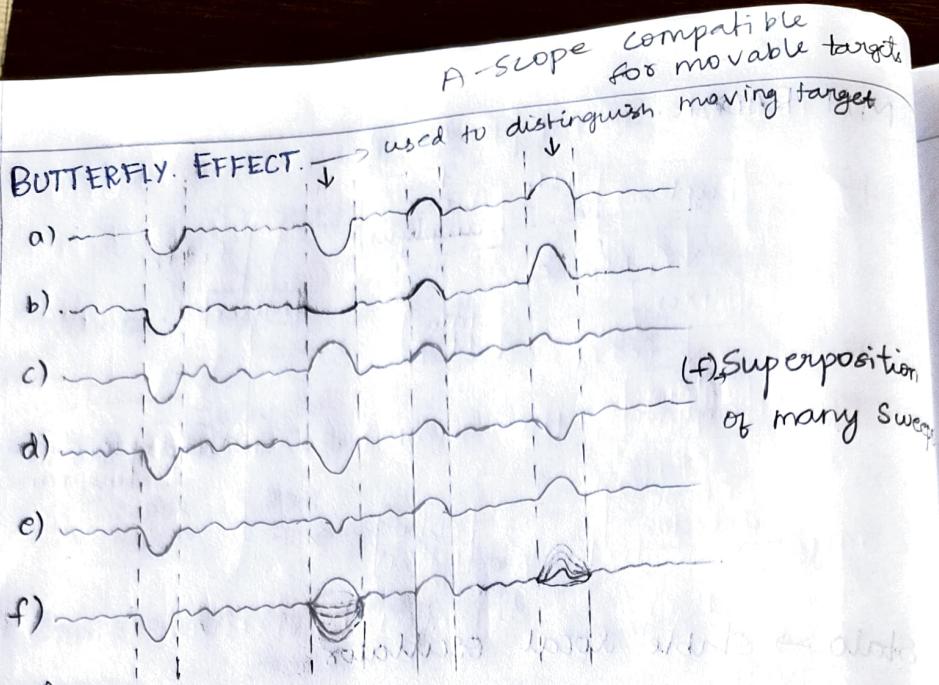
fd → doppler freq.

If → Intermediate freq.

### NON-COHERENT MTI RADARS.



Reference S/g → echo S/g received from the stationary field.



Freq. Response of Delay line canceller.

$$V_r = K \sin(2\pi f_d t - \phi_0)$$

$f_d$  → doppler freq shift

$$\phi_0 \rightarrow \text{constant phase} = 4\pi R_0 / \lambda$$

$R_0 \rightarrow \text{range at time equal to zero}$

$\lambda \rightarrow \text{wavelength}$

$K \rightarrow \text{amplitude}$

Signal from previous transmission.

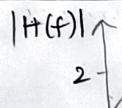
$$V_2 = K \sin[2\pi f_d(t - T_p) - \phi_0]$$

Output of DLC

$$V = V_1 - V_2$$

$$= 2K \sin(\pi f_d T_p) \cos[2\pi f_d(t - T_p) - \phi_0]$$

$$|H(f)| = 2 \sin(\pi f_d T_p)$$



Blind Speed.

$$f_d = \frac{2V}{\lambda}$$

velocity  $v$  with

METHODS TO

- Operate
- Operate
- Operate
- Operate

CLUTTER AT

Clutter Sp

1. Internal
2. Instabil
3. finite

Single D

CA

MTI Im

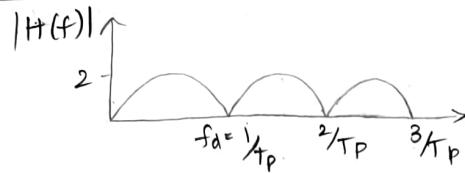
c

Clutte

$$\sin a - \sin b$$

$$= 2 \sin \left( \frac{a-b}{2} \right) \cos \left( \frac{a+b}{2} \right)$$

target



$$n = 1, 2, 3, \dots$$

position  
any sweeps

Blind Speed.

$$f_d = \frac{2V_s}{\lambda} = n f_p = n/T_p$$

$n \rightarrow$  number blind speed

velocity of the target that travels,  $V_n = \frac{n\lambda}{2T_p} = \frac{n\lambda f_p}{2}$   
with blind speed

METHODS TO ELIMINATE BLIND SPEED.

- Operate the radar at high long wavelength.
- Operate the radar at high PRF
- Operate the radar with more than one PRF
- Operate the radar at RF frequency.

Stable local oscillator & Coho  $\rightarrow$  Present in MTI

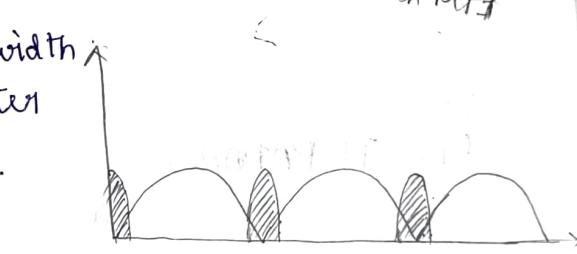
CLOUTER ATTENUATION.

Clutter Spectrum has finite width

1. Internal motion of the clutter
2. Instabilities of stato & cohoh.
3. finite Signal duration

Single DLC

$$CA = \frac{f_p^2 \lambda^2}{T_b \pi^2 \sigma_v^2}$$



2-DLC.

$$CA = \frac{f_p^4 \lambda^4}{T_b \pi^4 \sigma_v^4}$$

MTI Improvement factor =  $\frac{SNR \text{ at the o/p of clutter filter}}{\text{Signal to clutter ratio of the o/p of the clutter filter}}$

Clutter  
interference signal which has no information

$$If = \frac{(Signal / clutter) o/p}{(Signal / clutter) o/p} / f_d$$

$$= \frac{C_{in}}{C_{out}} \times \frac{S_{out}}{S_{in}} / f_d = CA \times \text{any power}$$

Full wave  
Rectifier

Video  
Uridole  
Video

MTI

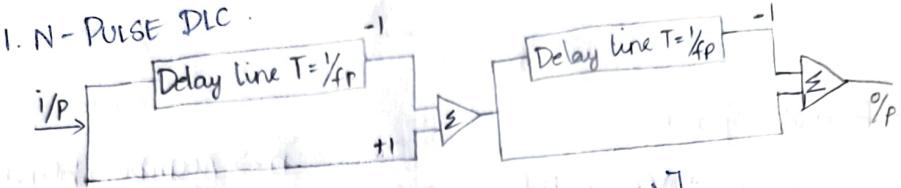
$OS(\frac{1+b}{2})$

J

Low noise AMP.

### TYPES OF DLC

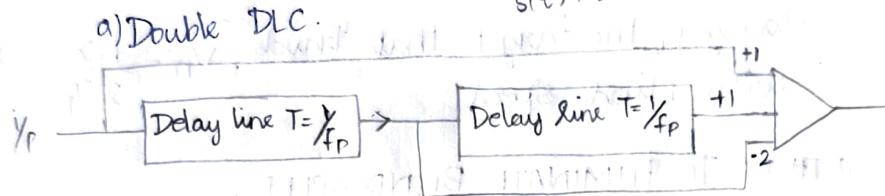
#### 1. N-PULSE DLC



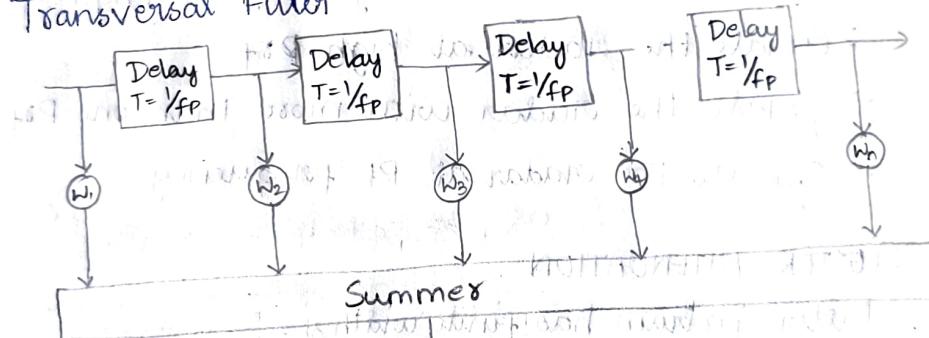
$$S(t) - S(t+T_p) - [S(t+T_p) - S(t+2T_p)]$$

$$S(t) + 2S(t+2T_p) - 2S(t+T_p)$$

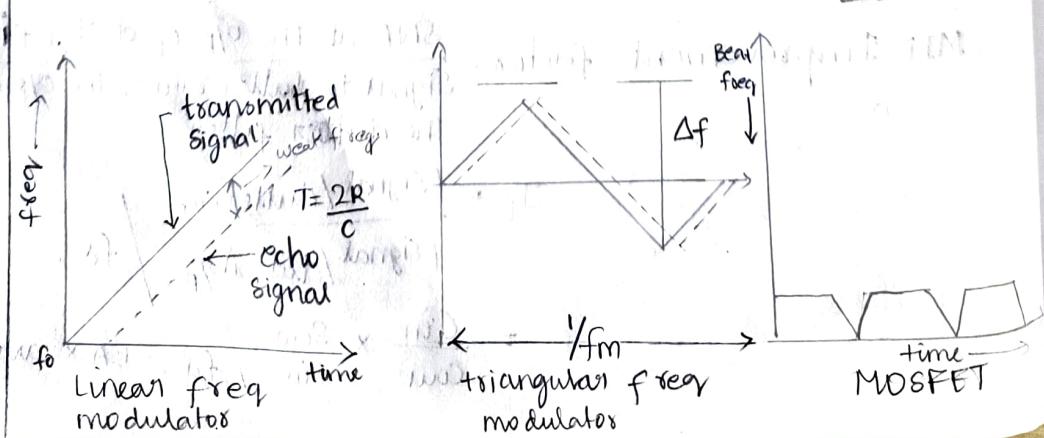
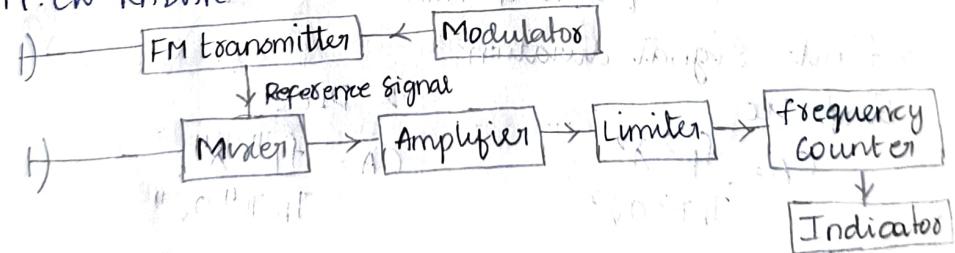
#### a) Double DLC



### Transversal Filter



### FM. CW RADAR



$f_r \rightarrow$  Beat

$$f_r = f_0 T$$

Beat  $f_r$

$$f_r =$$

$$f_{b u p}$$

$$f_{b d}$$

### TRACKING R

→  
→  
→

### AMPLITUDE -

Antenna  
feeds

$f_r \rightarrow$  Beat freq due to target

$$f_r = f_0 T_{\text{transit time}} = \frac{2R}{c} f_0$$

Beat freq modulated at rate  $f_m$ .

$$f_r = \frac{2R}{c} 2 f_m \\ = \frac{4R f_m}{c}$$

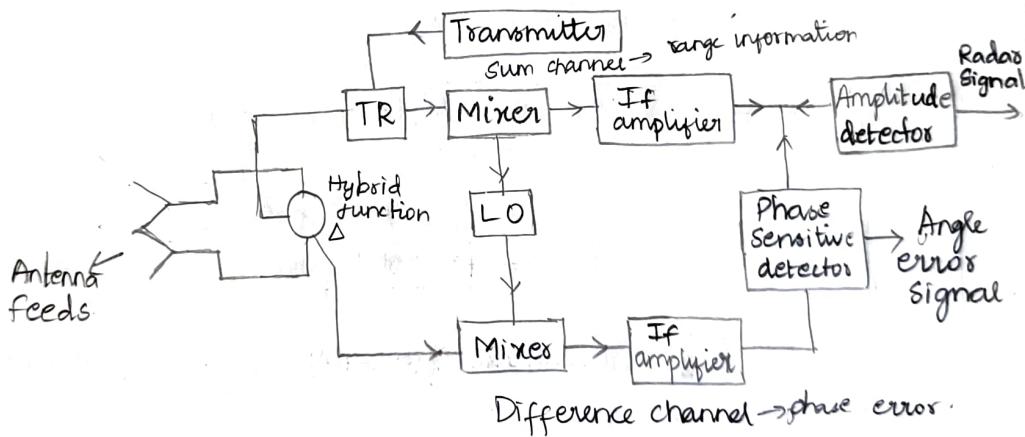
$f_{b\text{up}} = f_r - f_d \rightarrow$  approaching the target

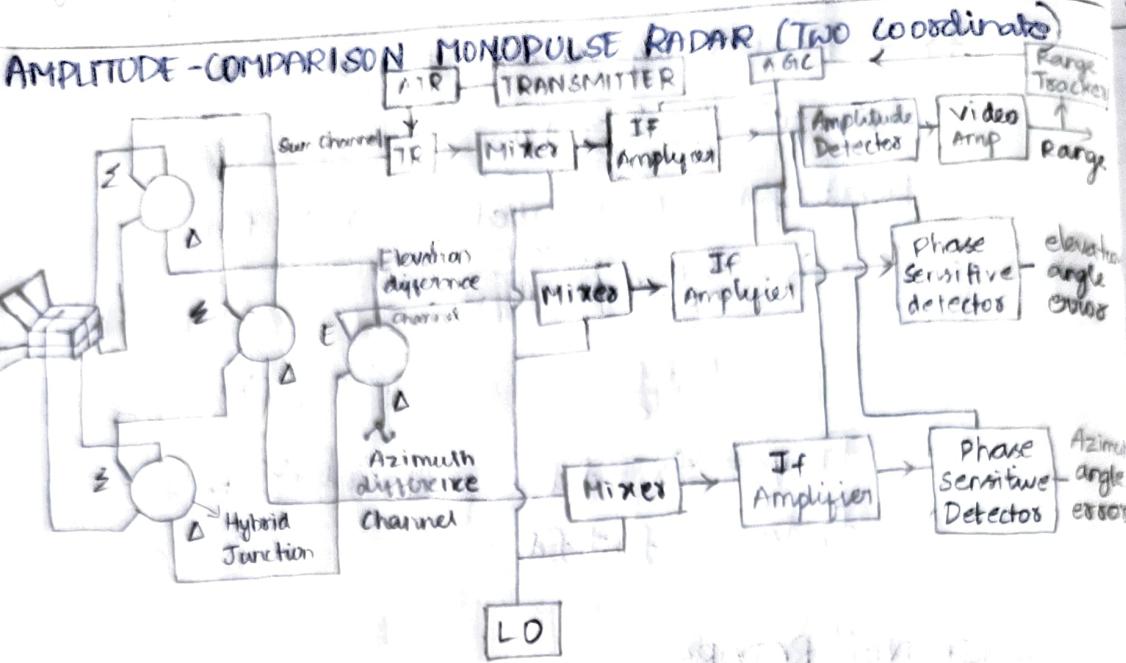
$f_{b\text{down}} = f_r + f_d \rightarrow$  away from the target

## TRACKING RADARS.

- 
- 
- 

### AMPLITUDE - COMPARISON MONOPULSE RADAR (One coordinate)





Automatic detection  
 ↓  
 operators can't  
 fix problems  
 always go w/  
 go for Auto-

- Detective
1. Neyman
  2. Like
  3. Inver
  4. Segm
  - L & C

## AUTOMATION

1. Sa
2. QI
3. A-
4. Sig
5. In
6. CF
7. Cl
8. T
9. A

## MAT

Fo

H

17/2  
17/3

-CFAR (Const. false alarm rate) clutter, pulse compression

## UNIT-3: DETECTION OF SIGNALS IN NOISE.

Detection Criteria:

1. Neyman Pearson Observer.
  2. Likelihood Receiver  $\rightarrow$  statistical analysis tool
  3. Inverse Probability Receiver. Bayes Criterion
  4. Sequential Observer  $\rightarrow$  no need to track signal continuously
- $L_x(\delta) = \frac{P_{sn}}{P_n}, L_x(\delta) \uparrow \rightarrow$  likelihood Ratio
- mistake noise for signal  
Type I - error
- Signal considered as noise Type II - error.
- missed detection  
fixed  
type I  $\rightarrow$  should be decreased
- Bayes  
 $P(B) = \frac{P(B/A) \times P(A)}{P(A/B)}$

## AUTOMATIC DETECTION.

steps involved in Automatic Detection.

1. Sampling
2. Quantization
3. A-D conversion
4. Signal Processing
5. Integration
6. CFAR  $\rightarrow$  setting threshold.
7. Clutter Map.
8. Threshold detection
9. Angle & Range measurement.

Clutter  $\rightarrow$  unwanted information

## MATCHED FILTER.

Freq Response

$$H(f) = G_a S^*(f) e^{-j2\pi f t_m}.$$

$G_a \rightarrow$  Constant (Gain parameter)

$t_m \rightarrow$  time at which the o/p of the matched filter is max.

$S^*(f) \rightarrow$  Complex Conjugate of the Spectrum of the received spectrum of signal  $s(t)$ .

more details in next class

Fourier Transform of  $s(t)$

$$S(f) = \int_{-\infty}^{\infty} s(t) \exp(-j2\pi f t) dt \quad \text{magnitude of } S(f)$$

Received Signal Spectrum

$$S(f) = |S(f)| \exp[-j\phi_s(f)] \quad \text{Complex.}$$

$|S(f)| \rightarrow$  Amplitude Spectrum

$\phi_s(f) \rightarrow$  Phase Spectrum

Matched Filter Response

$$H(f) = |H(f)| \exp[-j\phi_m(f)] \quad \text{magnitude of } H(f)$$

$$G_a = 1 \quad \text{constant gain}$$

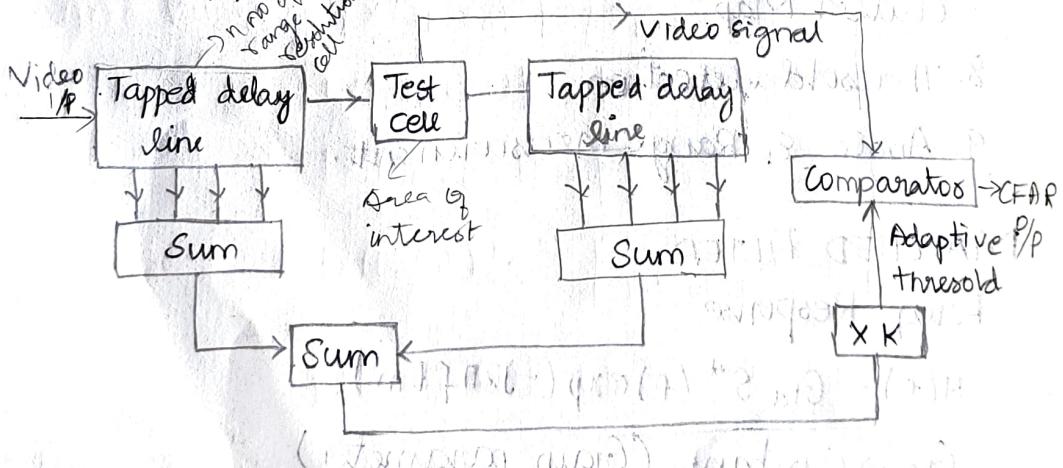
$$|H(f)| \exp[-j\phi_m(f)] = |S(f)| \exp[-j(\phi_s(f) - 2\pi f t_m)]$$

Equating amplitude & phase

$$\Rightarrow |H(f)| = |S(f)|$$

$$\Rightarrow \phi_m(f) = -\phi_s(f) + 2\pi f t_m$$

CONSTANT FALSE ALARM RATE RECEIVER (CFAR)



Working Cell - Averaging CFAR (using the unit cell)

$$\text{Loss (dB)} = -\frac{10}{M} \log P_{fa}$$

M → no. of reference cell

N → no. of pulses integrated

P<sub>fa</sub> → Probability of false alarm.



CLUTTER ED

Sum up

Separate

loss → 0

CENSORED

→ To remove

much larger

→ Threshold

ORDERED

→ O/P of

Pick K<sup>th</sup> e

AMBIQUITY

represent

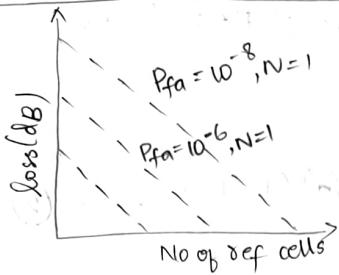
Signal of

obj & Utili

It gives

selection

O/P of



### CLUTTER EDGES

Sum up separately leading & trailing edges

Separately  $\rightarrow$  fix up  $> 0 <$  threshold ( $G_{10} - CFAR$ )  
 LOS  $\rightarrow$  0.1 to 0.3 dB.

### CENSORED MEAN-LEVEL DETECTOR (CMD)

$\rightarrow$  To remove the O/P of those reference cells that are much larger predefined no. of ref. cells are removed

$\rightarrow$  Threshold by M-T cells.

clutter map  
 $\rightarrow$  gives location to

### ORDERED STATISTICS

$\rightarrow$  O/P of ref cells put in order from smallest to largest

Pick  $k^{\text{th}}$  ordered value.

### AMBIGUITY DIAGRAM (AD)

Represents the response of the matched filter to the signal of which it is matched. It provides an indication of the utility of the particular classes of radar waveform. It gives the radar designer guidelines for the selection of suitable waveform for various applications.

$$\text{O/P of matched filter} = \int s_r(t) s^*(t - T_R) dt \quad \text{--- (1)}$$

$s_r(t) \rightarrow$  received signal

$s(t) \rightarrow$  Transmitted Signal

$s^*(t) \rightarrow$  Complex conjugate

$T_R \rightarrow$  estimate of the time-delay

+ detection criteria, CFAR, Ambiguity  
pulse compression, range cgn of surface clutter

Tx-ed Signal in complex  $j2\pi f t$  form.

$$u(t) e^{j2\pi f t}$$

The received echo signal

$$s_r(t) = u(t - T_R) e^{j2\pi(f_0 + f_d)(t - T_0)} \quad (2)$$

O/p of matched filter.

$$\begin{aligned} &= \int_{-\infty}^{\infty} u(t - T_0) e^{j2\pi(f_0 + f_d)(t - T_0)} \left[ u(t - T_R') e^{j2\pi f_0(t - T_R')} \right] dt \\ &= \int_{-\infty}^{\infty} u(t - T_0) u^*(t - T_R') e^{j2\pi(f_0 + f_d)(t - T_0) - j2\pi f_0(t - T_R')} dt \end{aligned} \quad (3)$$

$$\text{Let } T_0 = 0, f_0 = 0 \Rightarrow T_0 + T_R' = -T_R' = T_R \Rightarrow$$

O/p of matched filter

$$X(T_R, f_d) = \int_{-\infty}^{\infty} u(t) u^*(t + T_R) e^{j2\pi f d} dt \quad (4)$$

$|X(T_R, f_d)|^2 \rightarrow$  ambiguity function & plot  $\rightarrow$  ambiguity diagram.

PROPERTIES OF AMBIGUITY DIAGRAM.  $E \rightarrow$  Energy of the

max values of  $|X(T_R, f_d)|^2 = (2E)^2$   $\rightarrow$  (1) Signal.

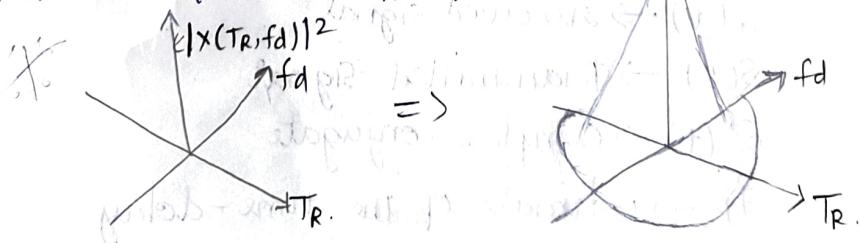
$$|X(-T_R, f_d)|^2 = |X(T_R, f_d)|^2 \quad (2)$$

$$|X(T_R, 0)|^2 = \left| \int u(t) u^*(t + T_R) dt \right|^2 \quad (3)$$

$$\text{magnitude of } |X(0, f_d)|^2 = \left| \int u^2(t) e^{j2\pi f d} dt \right|^2 \quad (4)$$

$$\text{volume of } (2E)^2 \text{ obtained from } \int |X(T_R, f_d)|^2 dT_R df_d = (2E)^2 \quad (5)$$

IDEAL AMBIGUITY DIAGRAM



1. An MTI Radar wavelength is

$$\text{Blind speed} = ?$$

$$f_p = ?$$

2. If a stat. of 6 GHz moving with

$$\text{freq., } f$$

$$fd$$

(CFAR, Ambiguity  
Range cgn of  
surface clutter)

1. An MTI Radar operates at a PRF of 1.5 kHz its operating wavelength is 3cm determine lowest blind speed.

$$\text{Blind Speed} = \frac{n\lambda f_p}{2}$$

$$f_p = 15 \text{ kHz}, \lambda = 3 \text{ cm}$$

$$= \frac{1 \times 3 \times 10^{-2} \times 15 \times 10^3}{2} \\ = \frac{45 \times 10^3}{2} = 22.5 \text{ m/sec}$$

2. If a stationary CW radar transmits at a freq. of 6 GHz, find the doppler freq. due to a target moving with a radial velocity of 200 km/hr.

$$\text{freq, } f_d = 6 \text{ GHz.}$$

$$f_d = \frac{2 V_r}{\lambda} \quad V_r = \frac{200 \times 10^3 \text{ m/sec}}{3600}$$

$$= \frac{2 \times 200 \times 10^3}{3600 \times 0.5 \times 10^9} \\ = 0.222 \times 10^3 \\ = 0.222 \text{ GHz.}$$

$$f = 6 \text{ GHz.}$$

$$\lambda = \frac{c}{f} = \frac{3 \times 10^8}{6 \times 10^9} \\ = 0.5 \times 10^{-1}$$

plot  $\rightarrow$  ambiguity diagram.

$\rightarrow$  Energy of the signal.

(2). (3)

$|t + T_p| dt|^2 - (3)$

$f_d |dt|^2 - (4)$

Volume

$E^2 - (5)$

$(2E)^2$

$f_d$   
 $T_p$



Q) MTI radar op at 50Hz with PRF of 1kHz, find 2nd and 3rd BFnd speeds of the radar.

Ans:-

\* Matched Filter receiver

\* Detection criteria

↳ threshold  $\rightarrow$  ENR<sup>0</sup>

↳ detection criteria

↳ 2 types of error  $\Phi_S$  normal person

↳ reduce F.A rate  $\rightarrow$  2nd one. (Rarely wood rad)

↳ inverse prob.  $\rightarrow$  3rd one.

↳ sequential oscillator

$\Rightarrow$  An MTI  
from a Automobile  
the radar is operating PRF  
Speed of the

SOL

\* AUTOMATIC DETECTION (UM)

↳ steps involved

↳ 8(0) 7 steps

\* CFAR

\* Ambiguity diagram.

↳ u properties.