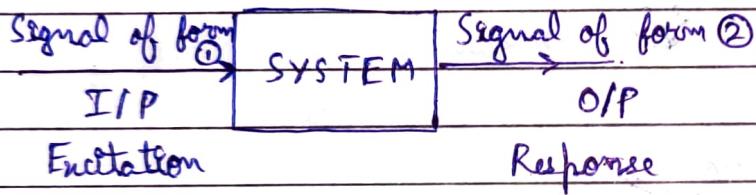
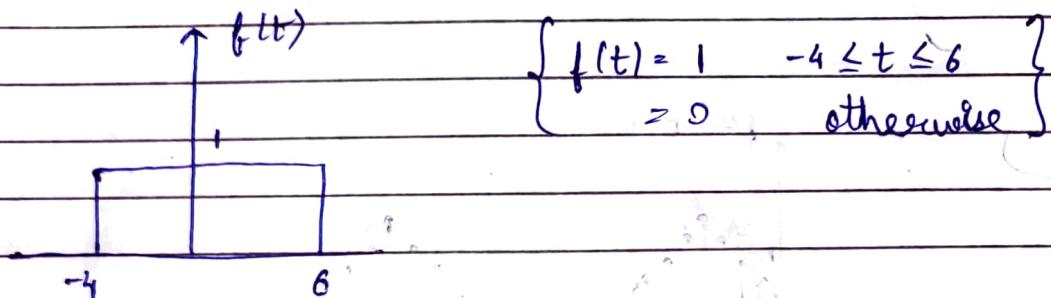


## → Intro to Signals & Systems

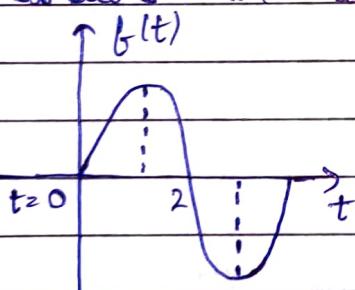
Physical quantity that contains information

Signals are expressed mathematically as function of indep. variable which is usually 'time'.

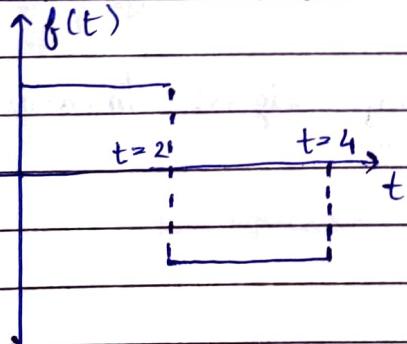


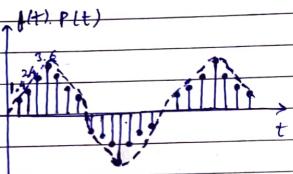
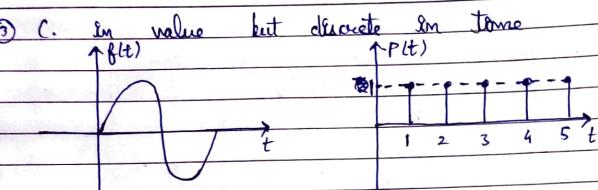
## → Basic types of signals

① Continuous in time & continuous in value signals

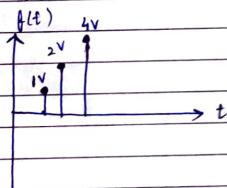


② C. in time but discrete in value



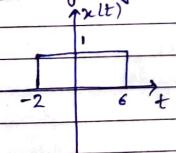


④ D in time & D in value



### → Operations on signals

① Time shifting



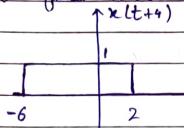
To find  $x(t \pm t_0)$

- Advance  $\rightarrow$  Any const.
- Delay  $\rightarrow$  Any const.

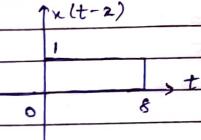
Delay: Shift signal towards right  
Advance: " " " " left

$$t_0$$

⑤ To find  $x(t+t_0)$



⑥ To find  $x(t-t_0)$



② Time scaling operation

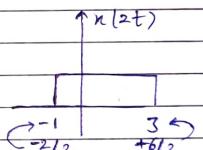
To find  $x(\alpha t)$

Scaling factor

$\alpha > 1 \rightarrow$  Signal compression

$\alpha < 1 \rightarrow$  Signal expansion

Divide the existing limits by ' $\alpha$ '

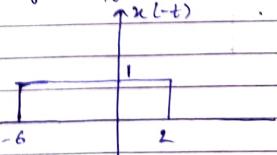


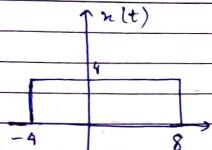
$x(t/2)$



③ Time reversal / folding operation

To find  $x(-t)$

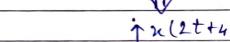
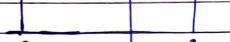
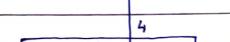




Found  $x(-2t+4)$

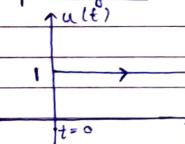
M-1: Moving right to left

$x(t+4)$



→ Elementary signals

① Unit step signal (heaviside step function).

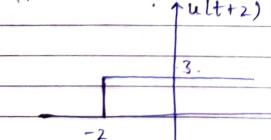


$$u(t) = \begin{cases} 1 & t > 0 \\ 0 & t < 0 \end{cases}$$

q)  $u(t-4)$



q)  $3u(t+2)$



q)  $u(t)$

$u(t)$

q)  $u(2t) = u(3t) = u(4t)$

NO EFFECT

q)  $u(-2t+4)$

$u(t+4)$

$u(2t+4)$

$u(-2t+4)$

q)  $u(-t-5)$

$u(t-5)$

$u(t-5)$

Signals

Causal

$x(t=0)$  for  $t < 0$

$x(t)$

2 6

Non-Causal

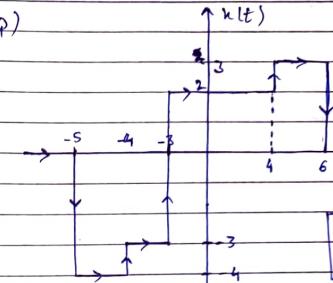
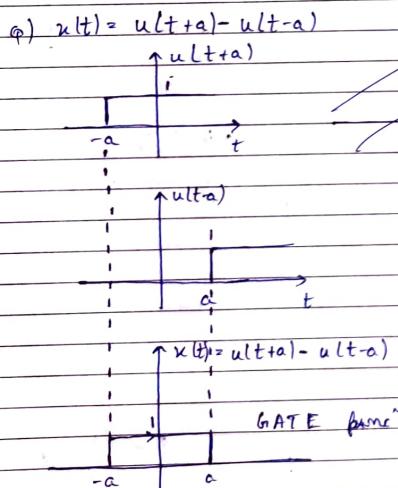
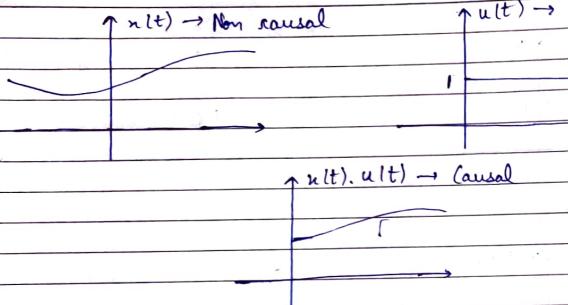
5

-9 3

Anti-Causal

10

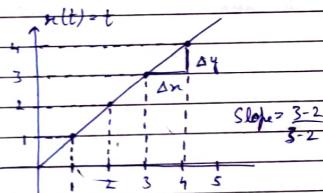
-7 -2



$$x(t) = -4u(t+5) + u(t+4) + 5u(t+3) + u(t-4) - 3u(t-6)$$

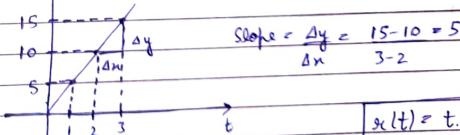
② Ramp signal

$$\begin{cases} u(t) = t & t \geq 0 \\ = 0 & \text{otherwise} \end{cases}$$

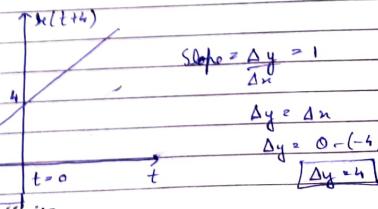


$$x(t) = A u(t) = At \quad \text{for } t \geq 0$$

$$x(t) = Ax(t) = 5t$$



$$x(t) = t \cdot u(t)$$



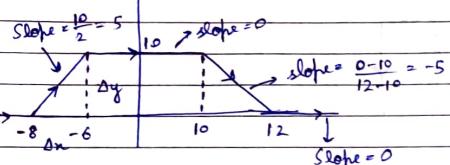
$$\text{Slope} = \frac{\Delta y}{\Delta x} = 1$$

$$\Delta y = \Delta x$$

$$\Delta y = 0 - (-4)$$

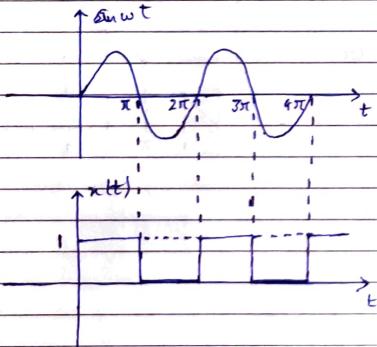
$$|\Delta y = 4|$$

$$x(t) = 5x(t+8) - 5x(t+6) - 5x(t-10) + 5x(t-12)$$



(a)  $u(t) = 1 \quad t > 0$   
 $= 0 \quad t \leq 0$

To draw  $x(t) = u(\sin \omega t) = 1 \quad \sin \omega t > 0$   
 $= 0 \quad \sin \omega t \leq 0$



### → Basic Signals

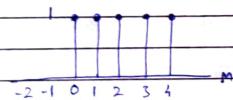
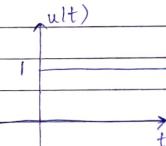
- ① Unit step signal
- ② Impulse function
- ③ Sigmoid func<sup>n</sup>
- ④ Exponential func<sup>n</sup>
- ⑤ Unit Ramp signal
- ⑥ Parabolic signal
- ⑦ Rectangular signal
- ⑧ Triangular signal
- ⑨ Sinusoidal
- ⑩ Sinc func<sup>n</sup>
- ⑪ Sampling func<sup>n</sup>

### → Unit Step Signal

$$u(t) \& u(m)$$

$$u(t) = \begin{cases} 1 & t \geq 0 \\ 0 & t < 0 \end{cases}$$

$$u(m) = \begin{cases} 1 & m \geq 0 \\ 0 & m < 0 \end{cases}$$



### Properties

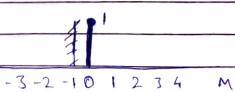
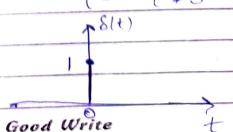
- 1)  $[u(t)]^m = u(t)$
- 2)  $u(at) = u(t)$   
 $u(a(t - t_0)) = u[a(t - t_0/a)]$   
 $= u(t - t_0/a)$

$$\begin{aligned} u(2t-4) &= u(t-4/2) \\ &= u(t-2) \end{aligned}$$

### → Impulse function ( $\delta(t)$ )

$$\delta(t) = \begin{cases} 1 & t = 0 \\ 0 & t \neq 0 \end{cases}$$

$$\delta(m) = \begin{cases} 1 & m = 0 \\ 0 & m \neq 0 \end{cases}$$



Properties

$$1) \int_{-\infty}^{\infty} S(t) dt = 1$$

$$2) S(m-k) = \begin{cases} 1 & m=k \\ 0 & m \neq k \end{cases}$$

$$3) S(m) = u(m) - u(m-1)$$

$$4) f(t) S(t) = f(0)$$

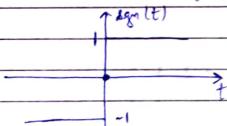
$$5) S(t-t_0) f(t) = f(t_0)$$

$$6) S(kt) = \frac{1}{k!} S(t)$$

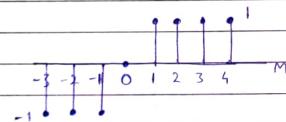
$$7) S(-t) = S(t) \rightarrow \text{Even function}$$

$\rightarrow$  Signum function ( $\text{sgn}(t)$ ,  $\text{sgn}(m)$ )

$$\text{sgn}(t) = \begin{cases} 1 & t > 0 \\ 0 & t = 0 \\ -1 & t < 0 \end{cases}$$



$$\text{sgn}(m) = \begin{cases} 1 & m > 0 \\ 0 & m = 0 \\ -1 & m < 0 \end{cases}$$



Relationship b/w  $u(t)$  &  $\text{sgn}(t)$

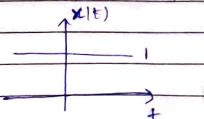
$$[\text{sgn}(t) = 2u(t) - 1]$$

$\rightarrow$  Exponential signal [ $x(t) = e^{xt}$ ]

Shape of exponential depends upon  $x$

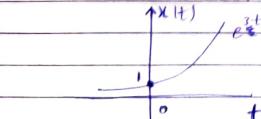
i) When  $x = 0$

$$u(t) = e^{0 \cdot t} = 1$$



ii) When  $x > 0$

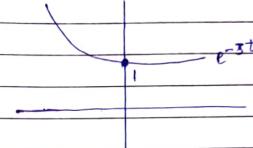
$$x = 3 \Rightarrow x(t) = e^{3t}$$



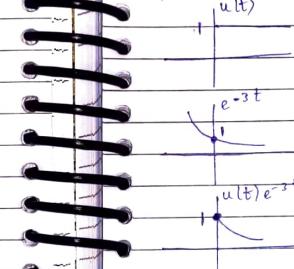
Good Write

iii) When  $x < 0$

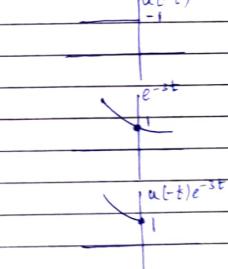
$$x = -3 \Rightarrow x(t) = e^{-3t}$$



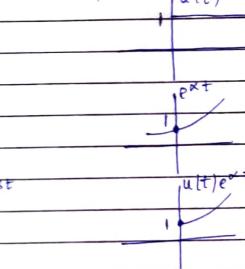
$$a) u(t) e^{-3t}$$



$$b) u(t) e^{-3(t)}$$



$$c) u(t) e^{xt}$$



$\rightarrow$  Unit Ramp signal [ $x(t)$  or  $x(m)$ ]

$$x(t) = \begin{cases} 1 & t \geq 0 \\ 0 & t < 0 \end{cases}$$

$$x(m) = \begin{cases} 1 & m \geq 0 \\ 0 & m < 0 \end{cases}$$

$$u(t)$$



$$[u(t) = x(t)]$$

$$[S(t) = u(t)]$$

$$u(t) = \frac{d}{dt} x(t)$$

$$[S(t) = x(t)]$$

Good Write

Good Write

$$\delta(t) = \frac{d^2 x(t)}{dt^2}$$

→ Unit Parabolic signal

$$x(t) = \begin{cases} t^2 & t \geq 0 \\ 0 & t < 0 \end{cases}$$

$$x(m) = \begin{cases} \frac{m^2}{2} & m \geq 0 \\ 0 & m < 0 \end{cases}$$



$$\int x(t) dt = \int t^2 dt = \frac{t^3}{3} = x(t)$$

$$\int \int u(t) dt = \frac{t^2}{2} = x(t)$$

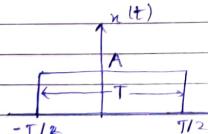
$$\int \int \delta(t) = x(t)$$

$$\int x(t) dt = \int u(t) dt = \int \int \delta(t)$$

$$u(t) = \frac{d}{dt} x(t) \quad u(t) = \frac{d}{dt} \int u(t) dt \quad \delta(t) = \frac{d^2}{dt^2} x(t)$$

→ Rectangular pulse [x(t)]

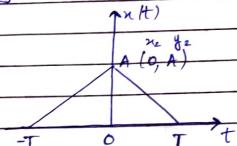
$$x(t) = A \text{rect}\left(\frac{t}{T}\right)$$



$A$  = amplitude of rect.  
 $T$  = period of rect.

→ Triangular signal [x(t)]

$$x(t) = A \left(1 - \left|\frac{t}{T}\right|\right)$$



$$y - y_1 = \frac{y_2 - y_1}{x_2 - x_1} (x - x_1)$$

 $(T, 0)$  $x_1, y_1$  $(T, 0)$  $x_1, y_1$ 

$$y - 0 = \frac{A - 0}{0 - (-T)} (x - (-T))$$

$$y = \frac{A}{T} (x + T)$$

$$x(t) = \frac{A}{T} (t + T) = \frac{A}{T} \left(T \left(1 + \frac{t}{T}\right)\right)$$

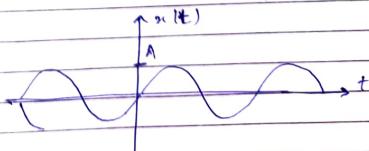
$$x(t) = A \left(1 + \frac{t}{T}\right)$$

$$\text{Similarly } x(t) = A \left(1 - \frac{t}{T}\right)$$

→ Sinusoidal Signal [x(t) & x(m)]

$$x(t) = A \cos(\omega_0 t + \phi) \text{ OR } A \sin(\omega_0 t + \phi)$$

$\downarrow$   
Time period → Phase shift  
Amplitude Fundamental freq.



$$\omega_0 = 2\pi f$$

$$f = \frac{1}{2\pi} \omega_0$$

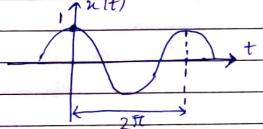
$$T = \frac{1}{f} = \frac{2\pi}{\omega_0}$$

q)  $x(t) = \cos t$

$$A=1, \omega_0 = 1$$

$$T = 2\pi = 2\pi$$

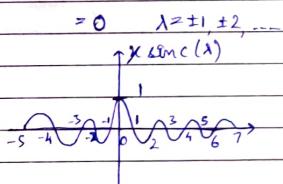
$$\omega_0 = \frac{2\pi}{T}$$



→ Sinc & sampling functions

Sinc

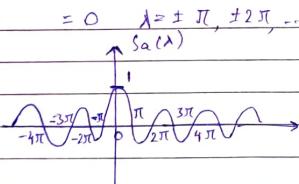
- Denoted with  $\text{sinc}(\lambda)$
- Normalized func<sup>n</sup>
- $\text{sinc}(\lambda) = \frac{\sin \lambda}{\lambda}$



$$\text{sinc}(0) = \text{sinc}(0) = 1$$

Sampling

- Denoted with  $\text{Sa}(\lambda)$
- Unnormalized func<sup>n</sup> sinc
- $\text{Sa}(\lambda) = \frac{\sin \lambda}{\lambda}$



→ Classification of Signals

- 1) Continuous time & discrete time signals
- 2) Even & Odd signals
- 3) Periodic & Aperiodic signals
- 4) Energy signals or power signals.
- 5) Deterministic & Non-deterministic signals (Random)
- 6) Real & Imaginary signals

• For discrete time signals time is discrete amplitude is continuous

• For digital signal both amplitude & time are discrete

→ Even & Odd signals

$$x(t) = x(-t) \quad x(t) = -x(-t)$$

$$\downarrow \quad \downarrow$$

Even      Odd

$$x(-t) = x_e(t) - x_o(t)$$

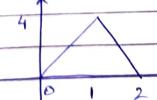
$$x(t) + x(-t) = 2x_e(t)$$

$$\boxed{x_e(t) = \frac{x(t) + x(-t)}{2}}$$

$$x(t) - x(-t) = 2x_o(t)$$

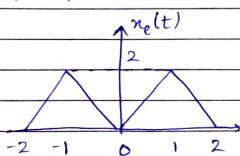
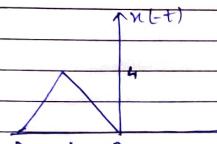
$$\boxed{x_o(t) = \frac{x(t) - x(-t)}{2}}$$

p) Find  $x_e(t)$  &  $x_o(t)$  of signal.

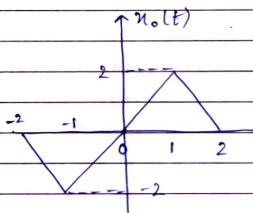


$$x_e(t) = \frac{x(t) + x(-t)}{2}$$

$$x_o(t) = \frac{x(t) - x(-t)}{2}$$



$$x_e(t) = \frac{1}{2} [x(t) + x(-t)]$$



$$x_o(t) = [x(t) - x(-t)]$$

- Q) Trace the even & odd component of signal  
 $x(t) = \cos t + \sin t + \cos 2\pi t + \sin^2 t$
- $\begin{matrix} \cos t & \sin t & \cos 2\pi t & \sin^2 t \\ E & O & O & E \end{matrix}$
- $x_e(t) = \frac{x(t) + x(-t)}{2}$        $x_o(t) = \cos t + \sin^2 t$
- $x_o(t) = \frac{x(t) - x(-t)}{2}$        $x_o(t) = \sin t + \cos 2\pi t$

→ Checking of periodicity

Method 1: Ratio

Consider two Signals  $x_1(t)$  &  $x_2(t)$  with periods  $T_1$  &  $T_2$  resp. when they are summed the resultant signal is said to be periodic when  $T_1 = \text{Rational multiple of } T_2$

Q)  $x(t) = \cos 4t + \sin \pi t$

$$\omega_1 = 4 \quad \omega_2 = \pi$$

$$T_1 = \frac{2\pi}{4} \quad T_2 = \frac{2\pi}{\pi}$$

$$T_1 = \frac{\pi}{2} \quad T_2 = 2$$

$$\frac{T_1}{T_2} = \frac{\pi/2}{2} = \frac{\pi}{4}$$

→ Aperiodic

Q)  $x(t) = \cos 2\pi t + \sin 4\pi t$

$$\omega_1 = 2\pi \quad \omega_2 = 4\pi$$

$$T_1 = \frac{2\pi}{2\pi} \quad T_2 = \frac{2\pi}{4\pi}$$

$$T_1 = 1 \quad T_2 = \frac{1}{2}$$

$$\frac{T_1}{T_2} = \frac{1}{2}$$

→ Periodic

Method 2: GCD

If GCD of freq. is possible then it is said to be periodic otherwise aperiodic.

## → Energy & Power Signals

Total energy transmitted to load  $E = \int_{-T}^T x^2(t) dt$

A signal is said to be an energy / Energy signal if the total Energy / power transmitted is finite.

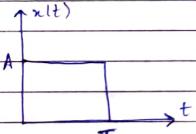
$$0 < E < \infty \text{ OR } 0 < P < \infty$$

Average Power  $P = \frac{1}{T} \int_{-T/2}^{T/2} x^2(t) dt$  (Finite duration)

$$P = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} x^2(t) dt \quad (\text{Infinite })$$

Power of Energy signal = 0  
Energy power =  $\infty$

q)  $x(t) = \begin{cases} A & 0 < t < T \\ 0 & \text{else} \end{cases}$



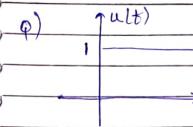
$$E = \int_{-T}^T x^2(t) dt$$

$$= \int_0^T A^2 dt$$

$$= A^2 [T - 0]$$

$$= A^2 T \text{ Joules} \rightarrow \text{Finite}$$

Energy signal ✓



$$E = \int_{-\infty}^{\infty} x^2(t) dt = \int_{-\infty}^{\infty} 1^2 dt$$

$$= \lim_{T \rightarrow \infty} \int_{-T}^T 1^2 dt = \lim_{T \rightarrow \infty} T = \infty \rightarrow \text{Infinite}$$

Not an energy signal

$$P = \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T x^2(t) dt$$

$$= \lim_{T \rightarrow \infty} \frac{1}{2T} \int_0^T 1 dt$$

$$= \lim_{T \rightarrow \infty} \frac{1}{2T} [t]_0^T$$

$$= \lim_{T \rightarrow \infty} \frac{1}{2T} [T - 0]$$

$$= \frac{1}{2} = 0.5 \text{ W}$$

Power signal ✓

q)  $x(t) = e^{-at} u(t)$



$$E = \int_{-\infty}^{\infty} x^2(t) dt$$

$$= \int_{-\infty}^{\infty} [e^{-at} u(t)]^2 dt$$

$$= \int_{-\infty}^{\infty} [e^{-2at}] dt$$

$$= \lim_{T \rightarrow \infty} \int_0^T [e^{-2at}] dt$$

$$= \frac{1}{-2a} e^{-2at} \Big|_0^{\infty}$$

$$= \frac{e^{-\infty}}{-2a} + \frac{e^0}{-2a} = \frac{1}{2a} \rightarrow \text{Finite}$$

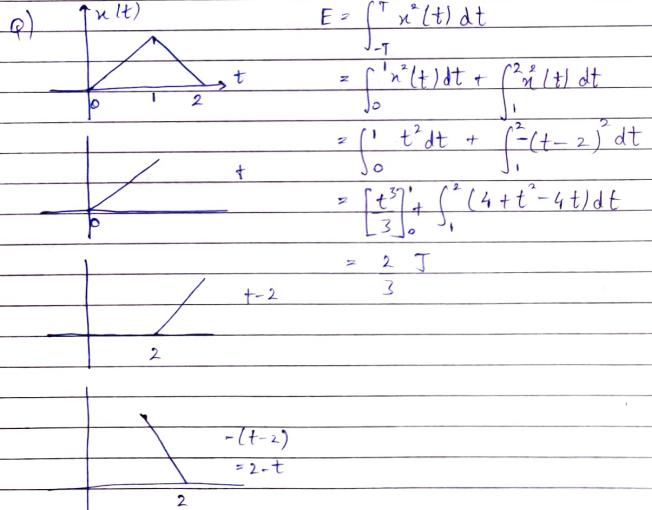
Energy signal ✓

Q) If energy of given signal  $x(t) = e^{-st} u(t)$  is  $\frac{1}{10}$  J. Then energy of time scaled version of signal  $x(2t)$  is

If then Energy of  $x(t) = E$   
 $x(at) = \frac{E}{a}$

$$\text{Energy of } x(t) = \frac{1}{10} \text{ J}$$

$$\text{Energy of } x(2t) = \frac{1}{10} = \frac{1}{20} \text{ J}$$



Q)

$$E = \int_{-T}^T x^2(t) dt$$

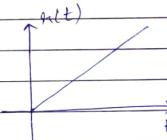
$$= \int_{-2}^2 n^2(t) dt$$

$$= \int_{-2}^{-1} (-t+2)^2 dt + \int_{-1}^1 1^2 dt + \int_1^2 (2-t)^2 dt$$

$$= 8 \text{ J}$$

Q) Check energy & power of stamp signal.

$$u(t) = \begin{cases} t & t \geq 0 \\ 0 & t < 0 \end{cases}$$



$$E = \lim_{T \rightarrow \infty} \int_{-T}^T u^2(t) dt$$

$$= \lim_{T \rightarrow \infty} \int_0^T t^2 dt$$

$$= \lim_{T \rightarrow \infty} \frac{T^3}{3}$$

$$= \infty$$

$$E = \infty$$

$$\text{Energy } X$$

$$P = \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T x^2(t) dt$$

$$= \lim_{T \rightarrow \infty} \frac{1}{2P} \times \frac{T^3}{3}$$

$$P = \infty$$

Neither energy nor power signal

## Classification of signals

- 1) Linear & Non-Linear system
- 2) Time Variant & Time invariant system
- 3) LTI LTV & LTI systems
- 4) State & Dynamic systems
- 5) Causal & Non-Causal
- 6) Invertible & Non-invertible systems
- 7) Stable & Unstable system.

## Linear & Non-Linear system

A system is said to be linear if it satisfies the superposition principle. Consider a system with input  $x_1(t)$ ,  $x_2(t)$  & output  $y_1(t)$  &  $y_2(t)$

$$T[a_1x_1(t) + a_2x_2(t)] = a_1T[x_1(t)] + a_2T[x_2(t)]$$

$$q) y(t) = x^2(t)$$

$$\begin{aligned} T[x_1(t)] &= x_1^2(t) \quad \rightarrow a_1x_1^2(t) + a_2x_2^2(t) \quad (\text{Independent response}) \\ T[x_2(t)] &= x_2^2(t) \quad || \\ T[a_1x_1(t) + a_2x_2(t)] &= [a_1x_1(t) + a_2x_2(t)]^2 \quad (\text{Common response}) \end{aligned}$$

Non-Linear system

$$q) y(t) = x(t)$$

$$\begin{aligned} T[x_1(t)] &= x_1(t) \quad \rightarrow a_1x_1(t) + a_2x_2(t) \\ T[x_2(t)] &= x_2(t) \quad || \\ T[a_1x_1(t) + a_2x_2(t)] &= a_1x_1(t) + a_2x_2(t) \end{aligned}$$

Linear system

## Time Variant & Time Invariant

A system is said to be time variant if its input, output characteristics changes with time otherwise said to be time invariant  
Condition for time invariance is  $y(m, k) = y(m-k)$   
where  $y(m, k) = T[x(m-k)]$

$$q) y(m) = x(m) + x(2m) x(m-2)$$

$$y(m, k) = T[x(m-k)] = x(m-k) + x(m-2-k)$$

$$y(m-k) = x(m-k) + x(m-k-2)$$

$$y(m, k) = y(m-k)$$

$\therefore$  Time invariant

$$q) y(m) = x(m) + m(x(m-3))$$

$$y(m, k) = T[x(m-k)] = x(m-k) + \frac{m}{m-k} x(m-3-k)$$

$$y(m-k) = x(m-k) + (m-k)x(m-3-k)$$

$$y(m, k) \neq y(m-k)$$

Time variant

## Linear time variant & Linear time invariant

A system is said to be LTV when it satisfies both linearity & time variance.

A system is said to be LTI when it satisfies both linearity & time invariance.

$$q) y(m) = mx^2(m)$$

$$y_1(t) = T[x_1(t)] = m x_1^2(t) \quad \rightarrow a_1mx_1^2(t) + a_2mx_2^2(t)$$

$$y_2(t) = T[x_2(t)] = m x_2^2(t) \quad ||$$

$$T[a_1x_1(t) + a_2x_2(t)] = m[a_1x_1(t) + a_2x_2(t)]^2$$

Non-Linear

## → Static & Dynamic system

q)  $y(m) = u(m)$  → Present  
 $y(0) = u(0)$   
 Static system

q)  $y(m) = 2u^2(t)$  → Present  
 $y(-1) = 2(-2)^2 = 2u^2(-2)$   
 Static

q)  $y(m) = u(m) + u(m-1)$   
 $y(1) = u(1) + u(1-1)$   
 $= u(1) + u(0)$   
 Dynamic      Past

q)  $y(t) = u(t) + u(t+3)$   
 $y(-1) = u(-1) + u(-1+3)$   
 $y(-1) = u(-1) + u(2)$   
 Present      Future  
 Dynamic

## → Causal & Non-Causal system

A system dependent on present & past inputs &  
 not on future input → Causal

A system dependent on present, past & future  
 inputs → Non Causal

q)  $y(m) = u(m) + \frac{1}{u(m-1)}$   
 $y(1) = u(1) + \frac{1}{u(0)}$   
 Present      Past  
Causal

q)  $y(t) = 2u(t) + \frac{1}{u^2(t)}$   
 $y(0) = 2u(0) + \frac{1}{u^2(0)}$   
Causal

q)  $y(m) = u(m) + \frac{1}{2u(m+1)}$   
 $y(0) = u(0) + \frac{1}{2u(1)}$   
Non-Causal

All static systems also  
 causal  
 All causal systems may  
 not be static

All non-causal systems are dynamic  
 Good Write Vice-Versa is not true

## → Immobile & Non-immobile system

If input appears at the output → Immobile  
 If  $y(t) \neq u(t)$  → Non-immobile

$\xrightarrow{h(t)} y(t)$

$y(t) = u(t) \cdot h(t)$

## → Stable & Unstable system

When system produces bounded output for a bounded  
 input → Stable

q)  $y(m) = u^2(m)$

$x(m) \xrightarrow{\quad} y(h) = u^2(m)$

Let  $u(m) = u(m)$

$y(m) = [u(m)]^2 = u(m) \rightarrow \text{Stable}$

q)  $y(t) = 2u^2(t)$

Let  $y(t) = u(t)$

$y(t) = 2u^2(t) = 2u(t) \rightarrow \text{Stable}$

q)  $y(t) = \int u(t)$

Let  $u(t) = u(t)$

$y(t) = \int u(t) = u(t) \rightarrow \text{Unstable}$

$$q) x(t) = 4 \cos \pi t + 5 \sin 2\pi t + 2 \sin 3\pi t$$

$$\omega_1 = \pi$$

$$\omega_2 = 2\pi$$

$$\omega_3 = 3\pi$$

$$T_1 = \frac{2\pi}{\pi} = 2$$

$$T_2 = 1$$

$$T_3 = \frac{2}{3}$$

$$\frac{T_L}{T_2} = \frac{2\pi}{\pi} = 2 \quad T_1 = \frac{2}{1} = 2 \quad T_2 = \frac{1}{\frac{2}{3}} = \frac{3}{2}$$

$\therefore x(t)$  is periodic

T = LCM of  $(T_1, T_2, T_3)$

$$= \text{LCM of } \left( \frac{2}{1}, \frac{1}{1}, \frac{2}{3} \right) \quad \left[ \begin{array}{l} \text{LCM of Num.} \\ \text{HCF of Dem.} \end{array} \right]$$

$$= \frac{2}{1} = 2$$

$$q) x(t) = 4 + \cos^2 4\pi t$$

$$x(t) = 4 + \left[ \frac{1 + \cos 8\pi t}{2} \right]$$

$$= \frac{4+1}{2} + \frac{1}{2} \cos 8\pi t$$

$$= \frac{9}{2} + \frac{1}{2} \cos 8\pi t$$

$$\omega_0 = 8\pi$$

$$T = \frac{2\pi}{8\pi} = \frac{1}{4} = 0.25 \text{ s}$$

→ Effect of time scaling on periodicity & period.

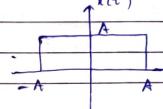
If  $x(t)$ ,  $T = T_1$   
 $x(at)$ ,  $T = T_1$

→ Effect of time scaling on energy of a signal

Consider an energy signal existing for  $-4 \leq t \leq 4$  with  $E_x = 100 \text{ J}$

If  $x_1(t) = x(st)$ , find  $E_{x_1}(t) = ?$

$$E_x(t) = \int_{-4}^4 A^2 dt = 100 \text{ (Given)}$$



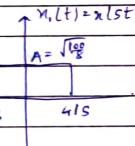
$$= A^2 \times 8 = 100$$

$$\Rightarrow A = \sqrt{\frac{100}{8}}$$

$$x_1(t) = x(st)$$

$$E_{x_1}(t) = \int_{-4/s}^{4/s} \left( \sqrt{\frac{100}{8}} \right)^2 dt$$

$$= 20 \text{ J}$$



Energy of $x(t) = E$
Energy of $x(st) = E$
X

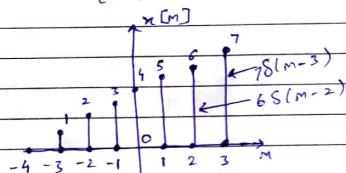
→ Effect of time scaling on power of a signal

No effect

→ Representation of discrete time signals

1) Tabular form

$$x[m] = \{1, 2, 3, 4, 5, 6, 7\}$$



2) Impulse representation

$$x[m] = \{1, 2, 3, 4, 5, 6, 7\}$$

$$\begin{aligned} x[m] &= 8(m+3) + 28(m+2) \\ &\quad + 38(m+1) + 48(m) + 58(m-1) \\ &\quad + 68(m-2) + 78(m-3) \end{aligned}$$

→ Time scaling on discrete time signals

To find  $x[\alpha m]$

Scaling factor

$\alpha > 1 \rightarrow$  Decimation in time

$\alpha < 1 \rightarrow$  Imp Interpolation

$$q) x[m] = \{1, 2, 3, 4, 5, 6, 7\}$$

To find  $x_1[m] = x[2m]$

$$x_1[m] = x[2m] = \{9, 14, 20, 8, 2\} \quad (2 \text{ values skipped})$$

$$x_1[m] = x[3m] = \{8, 5, 9, 18\} \quad (3 \text{ " " })$$

$x_1[0] \rightarrow$

Interpolation [ $\alpha < 1$ ]

$$x[m] = \{1, 2, 3, 4, 5, 6, 7\}$$

$$x_1[m] = x[m]$$

$(\alpha-1)$  zeros will be appended b/w each two values  
Good Write of  $x[m]$

$$x[m] = \{5, 9, 3, 12, 14, 18\}$$

$$x[m] = \{5, 0, 0, 9, 0, 0, 3, 0, 0, 12, 0, 0, 14, 0, 0, 18\}$$

$$q) x[m] = \{3, 5, 6, 8, 9\}$$

$$\text{To find } x_1[m] = x\left[\frac{2m}{3}\right]$$



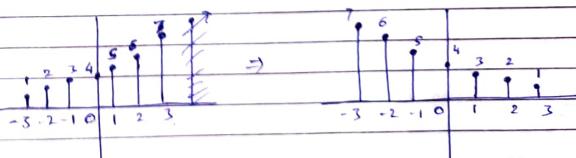
$$x_1[2m] = \{5, 8\}$$

$$x_1\left[\frac{2m}{3}\right] = \{5, 0, 0, 8\} = x_1[m]$$

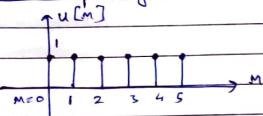
→ Time reversal operation

$$x[m] = \{1, 2, 3, 4, 5, 6, 7\}$$

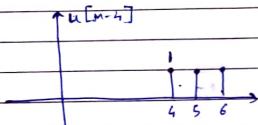
$$x[-m] = \{7, 6, 5, 4, 3, 2, 1\}$$



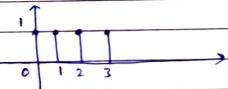
→ Unit step signal



→ Unit Impulse / Sample / Delta function



$$\begin{aligned} \frac{d}{dt} u(t) &= S(t) \\ \int_{-\infty}^t S(\tau) d\tau &= u(t) \\ \int_{-\infty}^{+\infty} S(\tau) d\tau &= 1 \end{aligned}$$



→ Relation b/w impulse & step signal

$$\textcircled{1} \quad u[m] - u[m-1] = \delta[m]$$

$$\textcircled{2} \quad \sum_{k=-\infty}^{\infty} \delta[k] = u[m]$$

$$\text{for } m < 0 \quad \sum_{k=-\infty}^m \delta[k] = 0$$

$$\text{for } m > 0 \quad \sum_{k=-\infty}^m \delta[k] = 1$$

→ Properties of delta function

$$\textcircled{1} \quad \delta[\alpha m] = \delta[m]$$

$$\textcircled{2} \quad n[m], \delta[m] = n[0], \delta[m]$$
  
$$n[m] \cdot \delta[m-m_0] = n[m_0] \delta[m-m_0]$$

$$\textcircled{3} \quad \sum_{m=-\infty}^{\infty} n[m] \cdot \delta[m] = n[0]$$

$$\sum_{m=-\infty}^{\infty} n[m] \cdot \delta[m-m_0] = n[m_0]$$

Good Write

Even & Odd sequences

$$n[-m] = n[m] \rightarrow \text{Even}$$

$$n[-m] = -n[m] \rightarrow \text{Odd}$$

$$n[m] = \{2, 2, 1, 3, 2\} \rightarrow \text{Even}$$

$$n[m] = \{-2, -3, 0, 3, 2\} \rightarrow \text{Odd}$$

$$n[-m] = \{2, 3, 0, -3, -2\} \rightarrow \text{Odd}$$

Amy P.T sequence can never be odd if it has having any non-zero value at origin

$$n_e[m] = n[m] + n[-m]$$

$$n_o[m] = n[m] - n[-m]$$

$$\text{Q) } y(m) = n(m) \cdot n(m-1)$$

For S/p  $n(m)$

$$y(m) = n(m) \cdot n(m-1)$$

For S/p  $-n(m)$

$$\begin{aligned} y(m) &= \{-n(m)\} \{-n(m-1)\} \\ &= n(m) \cdot n(m-1) \end{aligned}$$

∴ Non-invertible

$$\text{Q) } y(m) = \begin{cases} n(m+2) & m \geq 0 \\ n(m) & m < -1 \end{cases}$$

For  $n(m) = \delta(m)$

$$\begin{aligned} y(m) &= \delta(m+2) \quad \text{for } m \geq 0 \\ &= 0 \end{aligned}$$

For  $n(m) = 2\delta(m)$

$$y(m) = 2 \cdot \delta(m+2) \quad \text{for } m \geq 0$$

Good Write = 0

∴ Non-invertible

Q)  $y(t) = \int_{-\infty}^t u(t) \cdot \sin 2t \, dt$

$$x(t) = \sin 2t$$

$$y(t) = \int_{-\infty}^t \sin^2 t \, dt = \int_{-\infty}^t [1 - \cos 2t] \, dt$$

$$= \frac{1}{2} \left[ \int_{-\infty}^t dt - \int_{-\infty}^t \cos 2t \, dt \right]$$

$\therefore$  Unstable

→ Imreversible & Non-invertible systems

If s/p produces distinct o/p  $\rightarrow$  Imreversible

If s/p can be determined by observing the o/p

Q)  $y(t) = 4x(t)$

x(t)	$y(t) = 4x(t)$
0	0
-1	-4
+1	+4

$\therefore$  Imreversible system

Q)  $y(t) = x^2(t)$

x(t)	$y = x^2(t)$
0	0
+1	1
-1	1
2	4
-2	4

$\therefore$  Non-invertible system

### → Test Signals

$$u(t) \quad u(m)$$

$$-u(t) \quad -u(m)$$

$$\delta(t) \quad \delta(m)$$

$$-\delta(t) \quad -\delta(m)$$

$$2u(t) \quad 2u(m)$$

$$2\delta(t) \quad 2\delta(m)$$

Q)  $y(t) = \sin t \cdot x(t)$

For  $x(t) = \delta(t)$

$$y(t) = \sin t \cdot \delta(t)$$

$$= \sin 0 \cdot \delta(t) = 0$$

$\therefore$  Non-invertible system

For  $x(t) = 4\delta(t)$

$$y(t) = \sin t \cdot 4\delta(t) = 0$$

$$[x(t) \cdot \delta(t) = x(0) \cdot \delta(t)]$$

Q)  $y(m) = x(m) \cdot n(m-1)$

For s/p  $u(m)$

$$y(m) = u(m) \cdot u(m-1)$$

For s/p  $-u(m)$

$$y(m) = -u(m) \cdot [-u(m-1)]$$

$$= u(m) \cdot u(m-1)$$

$\therefore$  Non-invertible system

Q)  $y(m) = \begin{cases} x(m+2) & m \geq 0 \\ x(m) & m < -1 \end{cases}$

For  $x(m) = \delta(m)$

$$y(m) = \delta(m+2) \text{ for } m \geq 0$$

$$= 0$$

For  $x(m) = 2\delta(m)$

$$y(m) = 2\delta(m+2) \text{ for } m \geq 0$$

$$= 0$$

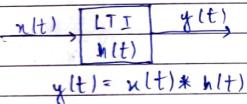
$\therefore$  Non-invertible system

Convolution

Continuous TS  
Convolution integral

Discrete TS  
Convolution summation

It is used to find o/p of any LTI system for any arbitrary s/p provided impulse response of the system is known.

Continuous time Convolution

$$y(t) = x(t) * h(t)$$

$$y(t) = \int_{-\infty}^{\infty} x(\tau) \cdot h(t-\tau) d\tau$$

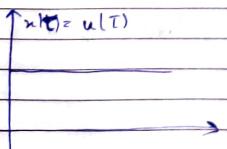
$$h(t-\tau) = h(-\tau+t)$$

Q)  $x(t) = u(t)$ ,  $h(t) = u(t)$

Find o/p  $y(t) = u(t) * h(t)$

$$y(t) = x(t) * h(t) = \int_{-\infty}^{\infty} x(\tau) \cdot h(t-\tau) d\tau$$

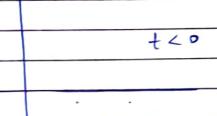
$$x(t) = u(t) \Rightarrow x(\tau) = u(\tau)$$



$$h(t) = u(t) \Rightarrow h(\tau) = u(\tau)$$

$$h(t-\tau) = u(t-\tau) = u(-\tau+t)$$

$$\uparrow h(t-\tau) = u(t-\tau)$$



$$\uparrow h(t-\tau) = u(t-\tau)$$



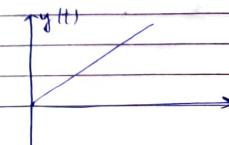
$$y(t) = \int_{-\infty}^{\infty} u(\tau) u(t-\tau)$$

$y(t) = 0$  for  $t < 0$  as no common area exists

For  $t \geq 0$

$$y(t) = \int_0^t 1 \cdot 1 d\tau = [1]_0^t = t$$

$$\begin{aligned} y(t) &= 0 & t < 0 \\ y(t) &= t & t \geq 0 \\ y(t) &= t \cdot u(t) = x(t) \end{aligned}$$



$$[u(t) * u(t)] = x(t)$$

Q)  $x(t) = u(t)$   $\boxed{h(t) = s(t)}$   $\rightarrow y(t)$

$$y(t) = x(t) * h(t) = \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau$$

$$x(t) = u(t) \quad h(t) = s(t)$$

$$x(\tau) = u(\tau) \quad h(\tau) h(t-\tau) = s(t-\tau)$$

$$y(t) = \int_{-\infty}^{\infty} u(\tau) s(t-\tau) d\tau = 0$$

$$= u(t) \Big|_{\tau=t} = u(t)$$

$$[u(t) * s(t)] = u(t)$$

Q)  $x(t) = e^t u(-t) \quad h(t) = u(t)$

$$y(t) = \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau$$

$$x(t) = e^t u(-t)$$

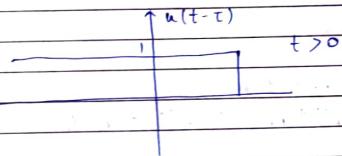
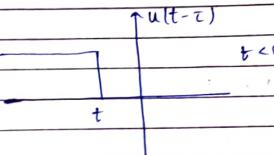
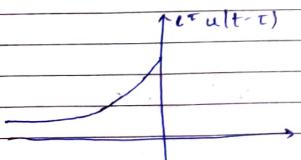
$$h(t) = u(t)$$

$$x(\tau) = e^\tau u(-\tau)$$

$$h(t-\tau) = u(t-\tau)$$

$$y(t) = \int_{-\infty}^{\infty} e^\tau u(-\tau) \cdot u(t-\tau) d\tau$$

$$e^t u(t)$$



For  $t > 0$

$$y(t) = \int_{-\infty}^t e^\tau \cdot d\tau = [e^\tau]_{-\infty}^t = e^t - 0 = e^t$$

For  $t < 0$

$$y(t) = \int_{-\infty}^0 e^\tau \cdot d\tau = [e^\tau]_{-\infty}^0 = 1 - 0 = 1$$

$$y(t) = \begin{cases} e^t & t < 0 \\ 1 & t \geq 0 \end{cases}$$

→ Properties of Convolution

(i) Commutative Property

$$\begin{aligned} y(t) &= x(t) * h(t) = h(t) * x(t) \\ &= \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau = \int_{-\infty}^{\infty} x(t-\tau) h(\tau) d\tau \end{aligned}$$

## 2) Associative property

$$y(t) = \{x_1(t) * x_2(t)\} * x_3(t)$$

$$= x_1(t) * [x_2(t) * x_3(t)]$$

## 3) Distributive property

$$\begin{aligned} y(t) &= x(t) * [h_1(t) \pm h_2(t)] \\ &= x(t) * h_1(t) \pm x(t) * h_2(t) \end{aligned}$$

## 4) Property based on linearity

$$\begin{aligned} y(t) &= x(t) * h(t) \\ x'(t) * h(t) &= y'(t) \\ x(t) * h'(t) &= y'(t) \\ x(t) * h'(t) &= y''(t) \end{aligned}$$

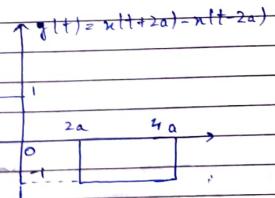
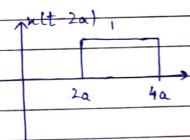
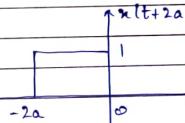
## 5) Property based on time invariance

$$\begin{aligned} x(t) * h(t) &= y(t) \\ x(t-t_0) * h(t) &= y(t-t_0) \\ x(t) * h(t-t_0) &= y(t-t_0) \\ x(t-t_0) * h(t-t_0) &= y(t-2t_0) \end{aligned}$$

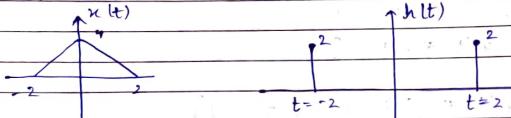
$$\begin{aligned} \text{Ex - } u(t) * u(t-4) &= x(t-t-4) \\ u(t-2) * u(t-4) &= x(t-7) \\ u(t-3) * \delta(t+4) &= u(t+1) \end{aligned}$$

(a)  $y(t) = 1 \quad 0 \leq t \leq 2a$   
 $h(t) = \delta(t+2a) - \delta(t-2a)$   
 $y(t) = u(t) * h(t) = ?$

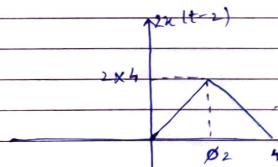
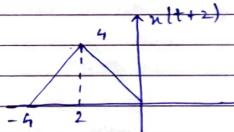
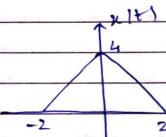
$$\begin{aligned} y(t) &= u(t) * [\delta(t+2a) - \delta(t-2a)] \\ &= [u(t) * \delta(t+2a)] - [u(t) * \delta(t-2a)] \\ y(t) &= u(t+2a) - u(t-2a) \end{aligned}$$



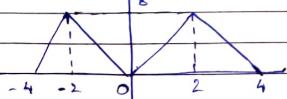
$$q) x(t) = 4 \delta(t) \quad h(t) = 2\delta(t+2) + 2\delta(t-2)$$



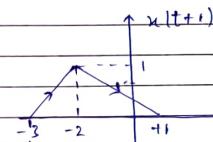
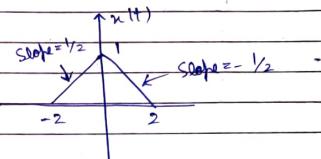
$$\begin{aligned} y(t) &= x(t) * h(t) \\ &= x(t) * [2\delta(t+2) + 2\delta(t-2)] \\ &= 2x(t+2) + 2x(t-2) \end{aligned}$$



$$y(t) = 2x(t+2) + 2x(t-2)$$



$$q) y(t) = x(t+1) + x(t-1)$$



$$\text{Find } y(t) \Big|_{t=1.5}$$

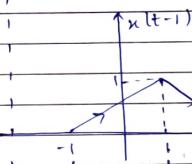
$$\Delta x = 1.5 - 1$$

$$\Delta y = ?$$

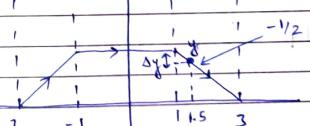
$$\text{Slope}_r = \frac{\Delta y}{\Delta x} = -\frac{1}{2}$$

$$\Delta y = -1 = -0.25$$

$$\begin{aligned} y &= 1/(-0.25) \\ y &= 0.75 \end{aligned}$$



$$y(t) = x(t+1) + x(t-1)$$



Q)  $u(t) = u(t+a) - u(t-a)$

$$h(t) = \text{sgn}(t)$$

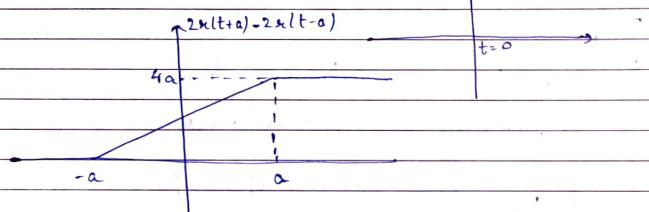
$$x(t) * h(t) = y(t)$$

$$x(t) * h'(t) = y'(t)$$

$$y'(t) = [u(t+a) - u(t-a)] * 2S(t)$$

$$y'(t) = 2u(t+a) - 2u(t-a)$$

$$[y(t) = 2u(t+a) - 2u(t-a)]$$



$$p(t) = \frac{t^2}{2} u(t)$$

$$\frac{d}{dt} [\text{Parabola}] = \text{Ramp}$$

$$u(t) = t \cdot u(t)$$

$$\frac{d}{dt} [\text{Ramp}] = \text{Step}$$

$$u(t) = 1 \cdot u(t)$$

$$\frac{d}{dt} [\text{Step}] = \text{Impulse}$$

Q)  $y(t) = u(t-2) * \delta(4-t)$   
 $= u(t-2) * \delta(t-4)$

Q)  $y(t) = u(t) * \delta(3-t)$   
 $= \delta(t-4) * \delta(4-t) = \delta(4t-3)$   
 $= \delta\left(\frac{1}{4}(t-3)\right)$

$$= \frac{1}{4} \delta\left(t-\frac{3}{4}\right)$$

$$= u(t) * \frac{1}{4} \delta(t-\frac{3}{4})$$

$$y(t) = \frac{1}{4} u(t - \frac{3}{4})$$

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→ Discrete time convolution

Q)  $x[n] = u[n]$

$h[n] = u[n]$

$y[n] = x[n] * h[n]$

$n[m] = u[m] \Rightarrow x[k] = u[k]$

$h[m] = u[m] \Rightarrow h[m-k] = u[m-k]$   
 $= u[-k+m]$

$y[m] = 0, m < 0$

$y[m] = m+1, m \geq 0$

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$x[k] = u[k]$

1

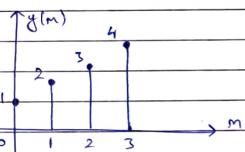
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3

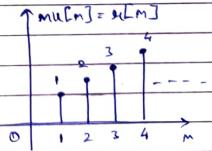
4

...

m



$y[m] = (m+1)u[m] \quad : \quad u[m] * u[m] = (m+1)u[m]$   
 $= u[m+1]$



Good Write

Good Write

$$1) x[m] * s[m] = x[m]$$

$$2) u[m-1] * s[m+3] = u[m+2]$$

$$3) \{u[m-1] - u[m-2]\} * \{s[m-1] - u[m-2]\}$$

$$\dots \cdot s[m-1] * s[m-1]$$

$$s[m-2]$$

$$q) x[m] = (0.5)^m u[m]$$

$$h[m] = s[m+2] + 0.5 s[m+1]$$

$$y[m] = (0.5)^m u[m] * \{s[m+2] + 0.5 s[m+1]\}$$

$$= (0.5)^{m+2} u[m+2] + 0.5 (0.5)^{m+1} u[m+1]$$

$$= (0.5)^{m+2} u[m+2] + 0.5^{m+2} u[m+1]$$

$$= (0.5)^{m+2} \{u[m+2] + u[m+1]\}$$

$$q) n[m] = \left(\frac{1}{2}\right)^m u[m]$$

$$h[n] = s[m] - \frac{1}{2} s[m-1]$$

$$y[n] = n[n] * \{s[m] - \frac{1}{2} s[m-1]\}$$

$$= x[m] - \frac{1}{2} x[m-1]$$

$$= \left(\frac{1}{2}\right)^m u[m] - \frac{1}{2} \left(\frac{1}{2}\right)^{m-1} u[m-1]$$

$$= \left(\frac{1}{2}\right)^m \{u[m] - u[m-1]\}$$

$$= \left(\frac{1}{2}\right)^m \delta[m]$$

$$= \left(\frac{1}{2}\right)^0 \cdot \delta[n]$$

$$[y[n] = s[n]]$$

a) These LTI systems are described by their impulse responses as

$$h_1[m] = \left(\frac{1}{2}\right)^m u[m]$$

$$h_2[m] = u[m+3]$$

$$h_3[m] = s[m] - s[m-1]$$

If the systems are connected in cascade, then the impulse response of equivalent system will be

$$[h_1[m] = \left(\frac{1}{2}\right)^m u[m]] \quad [h_2[m] = u[m+3]] \quad [s[m] = s[m+6]]$$

$$h[m] = h_1[m] * \{h_2[m] * h_3[m]\} \rightarrow 0$$

$$h_2[m] * h_3[m] = u[m+3] * \{s[m] - s[m-1]\}$$

$$= u[m+3] - u[m+2] = s[m+3]$$

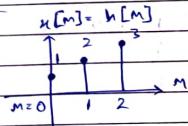
$$h[m] = \left(\frac{1}{2}\right)^m u[m] * s[m+3]$$

$$h[m] = \left(\frac{1}{2}\right)^{m+3} u[m+3]$$

→ Convolution of sequences

$$q) x[m] = \{1, 2, 3\}$$

$$h[m] = \{1, 2, 3\}$$



$$y[m] = \sum_{k=-\infty}^{\infty} x[k] h[m-k]$$

$$x[k] = \{1, 2, 3\}$$

$$h[k] = \{1, 2, 3\}$$

$$h[-k] = \{3, 2, 1\}, \{0, 0, 0\}$$

$$y[0] = \sum_{k=-\infty}^{\infty} x[k] \cdot h[-k] = 1$$

$$y[n] = \sum_{k=-\infty}^{\infty} x[k] h[1-k]$$

$$x[k] = \{0, 1, 2, 3\}$$

$$h[1-k] = \{3, 2, 1, 0\}$$

$$y[1] = 0 + 2 + 2 + 0 = 4$$

$$y[2] = \sum_{k=-\infty}^{\infty} x[k] h[2-k]$$

$$x[k] = \{1, 2, 3\}$$

$$h[2-k] = \{3, 2, 1\}$$

$$y[2] = 3 + 4 + 3 = 10$$

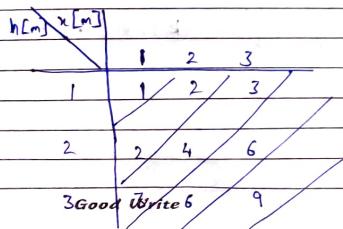
$$y[-1] = \sum_{k=-\infty}^{\infty} x[k] h[-1-k]$$

$$x[k] = \{1, 2, 3\}$$

$$h[-1-k] = \{3, 2, 1, 0, 0\}$$

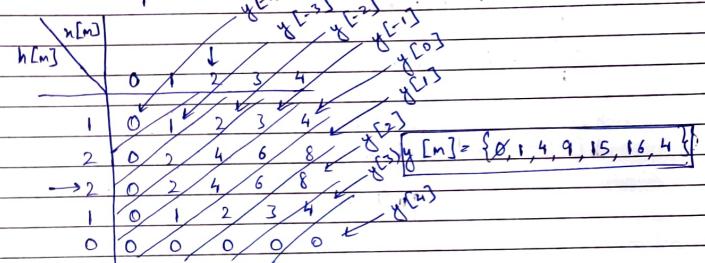
$$y[-1] = 0$$

### Motzkin method



$$q) x[m] = \{1, 2, 3, 4\}$$

$$h[m] = \{1, 2, 2, 1\}$$



$$p) y[n] = g[m] * h[n]$$

$$h[n] = \left(\frac{1}{2}\right)^n u[n] = \{1, \frac{1}{2}, \frac{1}{4}, \frac{1}{8}, \dots\}$$

$$g[m] \rightarrow \text{(causal sequence)} = \{g(0), g(1), g(2), \dots\}$$

$$g[0] = 1, \quad g[1] = \frac{1}{2}$$

$$g[1] = ?$$

$$\begin{aligned}
 & g[m] \quad | \\
 & \downarrow \quad 1 \quad \frac{1}{2} \quad \frac{1}{4} \quad \frac{1}{8} \quad \dots \\
 \rightarrow g(0) & \quad g(0) \quad \frac{1}{2}g(0) \quad \frac{1}{4}g(0) \quad \frac{1}{8}g(0) \\
 g(1) & \quad g(1) \quad \frac{1}{2}g(1) \quad \frac{1}{4}g(1) \quad \frac{1}{8}g(1) \\
 g(2) & \quad g(2) \quad \frac{1}{2}g(2) \quad \frac{1}{4}g(2) \quad \frac{1}{8}g(2) \\
 & \vdots \quad \vdots \quad \vdots \quad \vdots
 \end{aligned}$$

$y[0] = g[0] = 1$   
 $y[1] = g[1] + \frac{1}{2}g[0] = \frac{3}{2}$   
 $\frac{1}{2} = g[1] + \frac{1}{2} \times 1$   
 $\boxed{g[1] = 0}$

$$q) x[n] = \alpha^n u[n]$$

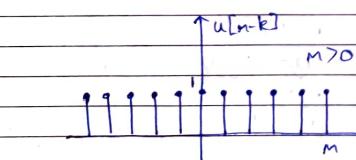
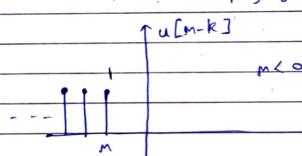
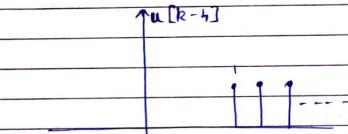
$$h[n] = \left(\frac{-1}{2}\right)^n u[n-4]$$

$$y[n] = \sum_{k=-\infty}^{\infty} u[k] h[n-k]$$

$$= \sum_{k=-\infty}^{\infty} h[k] u[n-k]$$

$$u[n-k] = (\alpha)^{m-k} u[m-k]$$

$$h[k] = \left(\frac{-1}{2}\right)^k u[k-4]$$



$$y[n] = \sum_{k=-\infty}^{\infty} \left(\frac{-1}{2}\right)^k u[k-4] \cdot (\alpha)^{m-k} \cdot u[m-k]$$

$$y[n] = 0 \quad m < 0$$

$$\begin{aligned} y[n] &= \sum_{k=4}^{\infty} \binom{n}{2} \cdot (\alpha)^n \cdot (\alpha)^k \\ &= (\alpha)^n \sum_{k=4}^{\infty} \binom{-1}{2}^k \end{aligned}$$

$$\text{Let } k-4 = m$$

$$k = m+4$$

$$\text{When } k=4, m=0$$

$$k=m, m=k-4$$

$$(\alpha)^n \sum_{m=0}^{n-4} \binom{-1}{2}^m$$

$$= (\alpha)^n \cdot \left(\frac{-1}{2}\right)^4 \sum_{m=0}^{n-4} \binom{-1}{2}^m$$

$$= (\alpha)^n \cdot \left(\frac{-1}{2}\right)^4 \left[ \frac{1 - (-1/2\alpha)^{m-4+1}}{1 - (-1/2\alpha)} \right]$$

$$= \frac{1}{16} (\alpha)^n \left[ \frac{1 - (-1/2\alpha)^{m-3}}{1 + 1/2\alpha} \right]$$

→ Continuous Fourier Time Series

↳ Only for periodic signals

$$\begin{array}{l} e^{j\omega_0 t} \\ e^{j2\omega_0 t} \\ e^{j3\omega_0 t} \end{array}$$

Freq. Freq. Freq.

$$x(t) = A_1 e^{j\omega_0 t} + A_2 e^{j2\omega_0 t} + A_3 e^{j3\omega_0 t} + \dots$$

Synthesis equation

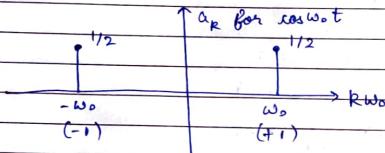
$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t}$$

Analysis equation

$$a_k = \frac{1}{T} \int_0^T x(t) \cdot e^{-jk\omega_0 t} dt$$

$$x(t) = \cos \omega_0 t = \frac{e^{j\omega_0 t} + e^{-j\omega_0 t}}{2} = \frac{1}{2} e^{j\omega_0 t} + \frac{1}{2} e^{-j\omega_0 t}$$

$$a_0 = \frac{1}{2}, \quad a_1 = \frac{1}{2}$$



$$x(t) = \sin \omega_0 t = \frac{e^{j\omega_0 t} - e^{-j\omega_0 t}}{2j} = \frac{1}{2j} e^{j\omega_0 t} - \frac{1}{2j} e^{-j\omega_0 t}$$

$$a_0 = \frac{1}{2j}, \quad a_1 = -\frac{1}{2j}$$

q) Find magnitude & phase spectra of

$$x(t) = \underbrace{4 + 2 \cos 2\pi t}_{T_1} + \underbrace{4 \sin 5\pi t}_{T_2}$$

$$\omega_0 = \frac{2\pi}{3}$$

$$\omega_0 = \frac{5\pi}{3}$$

$$T_1 = \frac{2\pi}{(2\pi/3)} = 3$$

$$T_2 = \frac{2\pi}{(5\pi/3)} = \frac{6}{5}$$

$$T_1 = 3$$

$$T_2 = \frac{6}{5}$$

$$T = \text{LCM} \left( \frac{3}{1}, \frac{6}{5} \right)$$

$$= 6$$

$$[T = 6]$$

Good Write

$$\omega_0 = \frac{2\pi}{T} = \frac{\pi}{3} \text{ rad/s}$$

$2 \cos \frac{2\pi}{3} t \rightarrow$  Represents 2<sup>nd</sup> harmonic

$4 \sin \frac{5\pi}{3} t \rightarrow$  " 5<sup>th</sup> harmonic "

$$\begin{aligned} x(t) &= 4 \cdot e^{j0 \cdot \frac{2\pi}{3} t} + 2 \left[ e^{\frac{j2 \times \pi/3 t}{2}} + e^{-\frac{j2 \times \pi/3 t}{2}} \right] \\ &\quad + 4 \left[ e^{\frac{j5 \times \pi/3 t}{2}} + e^{-\frac{j5 \times \pi/3 t}{2}} \right] \\ &= 4 \cdot e^{j0 \cdot \frac{2\pi}{3} t} + e^{j2 \times \frac{\pi}{3} t} + e^{-j2 \times \frac{\pi}{3} t} + 2 e^{\frac{j5\pi/3 t}{2}} - 2 e^{-\frac{j5\pi/3 t}{2}} \end{aligned}$$

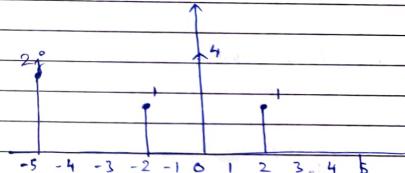
$$a_0 = 4$$

$$a_2 = 1$$

$$a_{-2} = 1$$

$$a_5 = -2j$$

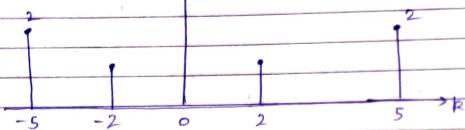
$$a_{-5} = +2j$$



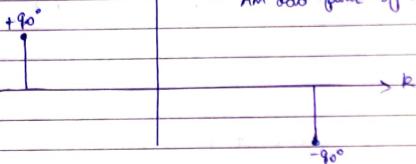
$a_k$	Values	$ a_k $	Magnitude	Phase angles
$a_0$	4	4	4	$0^\circ$
$a_2$	1	1	1	$0^\circ$
$a_{-2}$	1	1	1	$0^\circ$
$a_5$	$-2j$	2	2	$-90^\circ$
$a_{-5}$	$2j$	2	2	$+90^\circ$

Good Write

$|a_k| \leftarrow$  Magnitude spectrum  
Always an even frame of  $k$



$\angle a_k \leftarrow$  Phase spectrum  
An odd frame of  $k$



Q) Consider a periodic signal with  $T=8$  sec & Fourier series coefficients given as:

$$a_0 = 2$$

$$a_3 = 8j$$

$$a_2 = 2$$

$$a_{-3} = -8j$$

Find the signal  $x(t)$

Synthesis equation  $x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\omega t}$

$$T = 8 \text{ sec.}$$

$$\omega_0 = \frac{2\pi}{T} = \frac{\pi}{4} \text{ rad/s}$$

$$x(t) = a_0 e^{j\omega_0 t} + a_2 e^{j2\omega_0 t} + a_3 e^{j3\omega_0 t} + a_{-3} e^{-j3\omega_0 t}$$

$$= 2e^{j\pi/2 t} + 2e^{-j\pi/2 t} + 8j e^{j3\pi/4 t} + 8j e^{-j3\pi/4 t}$$

$$= 2\sqrt{2} \left[ e^{j\pi/2 t} + 2e^{-j\pi/2 t} \right] + 8j \left[ e^{j3\pi/4 t} - e^{-j3\pi/4 t} \right]$$

Good Write

$$x(t) = 4 \cos \frac{\pi}{2} t - 16 \sin \frac{3\pi}{4} t$$

→ Properties of Fourier series

① Linearity property

$$x(t) \xrightarrow{\text{F.S}} a_k$$

$$y(t) \xrightarrow{\text{F.S}} b_k$$

$$x(t) = \alpha x(t) + \beta y(t) \xrightarrow{\text{F.S}} c_k$$

$$c_k = \alpha \cdot a_k + \beta \cdot b_k$$

$$x(t) = n(t) = \cos \omega_0 t$$

$$a_0 = \frac{1}{2}, \quad a_{-1} = \frac{1}{2}$$

$$y(t) = m(t) = \sin \omega_0 t$$

$$b_1 = \frac{1}{2j}, \quad b_2 = -\frac{1}{2j}$$

$$z(t) = n(t) + y(t) = \cos \omega_0 t + i \sin \omega_0 t$$

$$= \frac{1}{2} e^{j\omega_0 t} + \frac{1}{2} e^{-j\omega_0 t} + \frac{1}{2j} e^{j\omega_0 t} - \frac{1}{2j} e^{-j\omega_0 t}$$

$$= \underbrace{\left[ \frac{1}{2} + \frac{1}{2j} \right] e^{j\omega_0 t}}_{c_1 = a_0 + b_1} + \underbrace{\left[ \frac{1}{2} - \frac{1}{2j} \right] e^{-j\omega_0 t}}_{c_{-1} = a_{-1} + b_{-1}}$$

② Time shifting property

$$x(t) \xrightarrow{\text{F.S}} a_k$$

$$x(t-t_0) \xrightarrow{\text{F.S}} b_k$$

$$b_k = e^{-j\omega_0 t_0} a_k$$

Good Write

$$x(t+t_0) \longleftrightarrow e^{jk\omega_0 t_0} \cdot a_k$$

$$b_k = e^{-jk\omega_0 t_0} \cdot a_k$$

$$b_k = |e^{-jk\omega_0 t_0}| \cdot |a_k|$$

$$|b_k| = |a_k|$$

$$b_k = |e^{-jk\omega_0 t_0} + j a_k|$$

$$|b_k| = |a_k - k\omega_0|$$

Q)  $x(t) \longleftrightarrow a_k$

$$x(t) = x(t+t_0) + x(t-t_0) \longleftrightarrow b_k$$

Find  $b_k$  in terms of  $a_k$

$$x(t) \longleftrightarrow a_k$$

$$x(t+t_0) \longleftrightarrow e^{jk\omega_0 t_0} \cdot a_k$$

$$x(t-t_0) \longleftrightarrow e^{-jk\omega_0 t_0} \cdot a_k$$

∴ using linearity property

$$\begin{aligned} b_k &= e^{jk\omega_0 t_0} \cdot a_k + e^{-jk\omega_0 t_0} \cdot a_k \\ &= a_k [e^{jk\omega_0 t_0} + e^{-jk\omega_0 t_0}] \times 2 \end{aligned}$$

$$b_k = 2 \cos(k\omega_0 t_0) \cdot a_k$$

③ Time reversal property

$$x(t) \xrightarrow{\text{F.S}} a_k$$

$$x(-t) \xrightarrow{\text{F.S}} b_k$$

$$b_k = a_{-k}$$

(Case-I - If  $x(t)$  is even signal

$$x(-t) = x(t)$$

$$b_k = +a_{-k} = a_k$$

(Case-II - If  $x(t)$  is odd signal

$$x(-t) = -x(t)$$

$$x(-t) = -x(t) \longleftrightarrow -a_k$$

$$b_k = a_{-k} = -a_k$$

Q)  $x(t) \longleftrightarrow a_k$

$$x(t) = x(t-t_0) + x(t-t_0) \longleftrightarrow b_k$$

Find  $b_k$  in terms of  $a_k$

$$x(t) \longleftrightarrow a_k$$

$$x(t+1) \longleftrightarrow e^{jk\omega_0 t} \cdot a_k$$

$$x(t-1) \longleftrightarrow e^{-jk\omega_0 t} \cdot a_k \quad \text{--- ①}$$

$$x(t) \longleftrightarrow a_k$$

$$x(t-1) \longleftrightarrow e^{-jk\omega_0 t} \cdot a_k \quad \text{--- ②}$$

$$b_k = e^{-jk\omega_0 t} \cdot a_{-k} + e^{-jk\omega_0 t} \cdot a_k$$

$$b_k = e^{-jk\omega_0 t} [a_{-k} + a_k] \quad \text{--- ③}$$

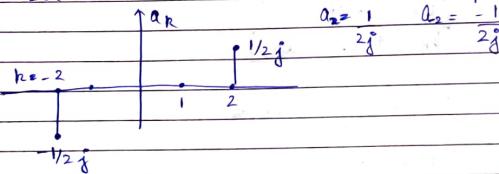
④ Time scaling property

$$x(t) \xrightarrow{\text{F.S}} a_k$$

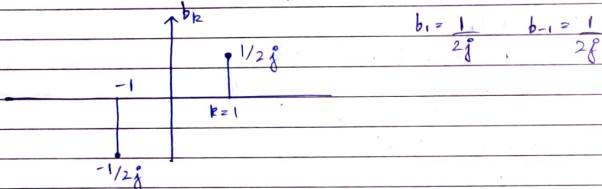
$$x(\alpha t) \xrightarrow{\text{F.S}} b_k$$

$$b_k = a_{\frac{k}{\alpha}}$$

Q)  $u_1(t) = \sin 2\omega_0 t$



$$u_1(t) = x(t|1/2) = \sin \left( 2\omega_0 \left( \frac{t}{2} \right) \right) = \sin \omega_0 t$$



④ Differentiation property

$$u(t) \xrightarrow{\text{F.S.}} a_R$$

$$\frac{d}{dt} u(t) \xrightarrow{\text{F.S.}} b_R$$

$dt$

$$[b_R = jk \omega_0 a_R]$$

$$|b_R| = |jk \omega_0| \cdot |a_R| = k \omega_0 |a_R|$$

$$\angle b_R = \angle jk \omega_0 + \angle a_R = \angle a_R + 90^\circ$$

⑤ Integration property

$$u(t) \xrightarrow{\text{F.S.}} a_R$$

$$\int_{-\infty}^t u(\tau) d\tau \xrightarrow{\text{F.S.}} b_R$$

$$[b_R = |a_R| \frac{\pi}{2}]$$

$$\boxed{b_R = \frac{1}{jk \omega_0} a_R}$$

$$\boxed{b_R = |a_R| - \frac{\pi}{2}}$$

⑥ Frequency shifting property

Integer

$$u(t) \xrightarrow{\text{F.S.}} a_R$$

$$e^{jkn\omega_0 t} \xrightarrow{\text{F.S.}} b_R$$

then  $[b_R = a_{R-m}]$

$$e^{-jkn\omega_0 t} \cdot u(t) \xleftrightarrow{\text{F.S.}} a_{R+m}$$

⑦ Multiplication property

$$u(t) \longleftrightarrow a_R$$

$$y(t) \longleftrightarrow b_R$$

$$z(t) = x(t) \cdot y(t) \longleftrightarrow c_R$$

$$[c_R = a_R * b_R]$$

Discrete convolution

→ Discrete time Fourier series

$$u(N+m) = u(m) \quad n(t) = \sum_{k=-\infty}^{\infty} e^{jkn\omega_0 t}$$

$$u[m] = \sum_{k=0}^{N-1} c(k) e^{j \frac{2\pi k m}{N}}$$

$$c_R = \frac{1}{T} \int u(t) e^{-jkn\omega_0 t} dt$$

$$(c_R) = \frac{1}{N} \sum_{m=0}^{N-1} u[m] e^{-\frac{j2\pi km}{N}}$$

Periodicity property of  $c(k)$

$$k \rightarrow k+N$$

$$c(k+N) = \frac{1}{N} \sum_{m=0}^{N-1} u[m] e^{-\frac{j2\pi (k+N)m}{N}}$$

$$\boxed{c(k+N) = c(k)}$$

Properties of DFTS① Linearity property

$$\alpha x_1[m] + \beta x_2[m] \longleftrightarrow \alpha C_1(k) + \beta C_2(k)$$

② Time shifting property

$$x[n] \longleftrightarrow C(k)$$

$$x[n-m_0] \longleftrightarrow e^{-j\frac{2\pi}{N}k_0 \cdot m_0} \cdot C(k)$$

③ Time scaling

$$x[n] \longleftrightarrow C(k)$$

$$x[\alpha n] \longleftrightarrow C(k/\alpha)$$

④ Convolution theorem

$$x_1[n] * x_2[n] \longleftrightarrow N.C_1(k).C_2(k)$$

↑  
Periodic convolution

$$y[n] = x_1[n] * x_2[n] = \sum_{m=0}^{N-1} x_1(m) x_2(m-n)$$

⑤ Multiplication theorem

$$x_1[n] \cdot x_2[n] \longleftrightarrow C_1(k) * C_2(k)$$

⑥ Frequency shifting theorem

$$x[n] \longleftrightarrow C(k)$$

$$e^{j\frac{2\pi}{N}k_0 \cdot m} \cdot x[m] \longleftrightarrow C(k-k_0)$$

⑦ Symmetric properties

$x[n]$ Real	$C(k)$ Conjugate Symmetric	$\text{Img}$	$\text{R.E}$	$\text{R.O}$	$\text{I.E}$	$\text{I.O}$
		$\text{CAS}$	$\text{RE}$	$\text{ID}$	$\text{IE}$	$\text{RP}$

$$x[n] = \cos \frac{\pi n}{3}$$

$$C(k) = \frac{1}{N} \sum_{m=0}^{N-1} x[m] \cdot e^{-j\frac{2\pi k m}{N}}$$

$$x[n] = \sum_{k=0}^{N-1} C(k) \cdot e^{j\frac{2\pi k n}{N}}$$

$$x[n] = \cos \frac{\pi n}{3}$$

$$\omega_0 = \frac{\pi}{3} \quad N = 2\pi \cdot m \quad \frac{N}{\omega_0}$$

$$= \left( \frac{2\pi}{\pi} \times 3 \right) / m$$

$$N=6$$

$$x[n] = \sum_{k=0}^2 C(k) \cdot e^{j\frac{2\pi k n}{6}}$$

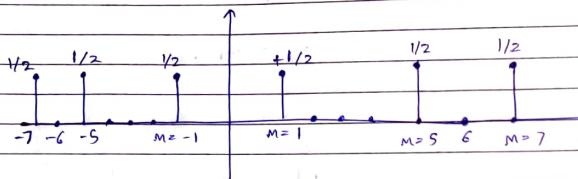
$$C(k) = (-3)e^{\frac{j2\pi(-3)}{6}} + (-2)e^{\frac{j2\pi(-2)}{6}} + (-1)e^{\frac{j2\pi(-1)}{6}}$$

$$+ (0) + (1)e^{\frac{j2\pi(1)}{6}} + (2)e^{\frac{j2\pi(2)}{6}}$$

$$x[n] = \cos \frac{\pi n}{3} = \frac{1}{2} e^{j\frac{2\pi n}{3}} + \frac{1}{2} e^{-j\frac{2\pi n}{3}}$$

Comparing ① & ②

$$c(-1) = \frac{1}{2}, \quad c(1) = \frac{1}{2}, \quad c(-3) = c(-2) = c(0) = c(2) = 0$$



→ Parseval's theorem (CTFS)

Total avg. power in any periodic signal is equal to sum of avg. power contained in all of its harmonic components.

$$P_x = \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T |x(t)|^2 dt = \sum_{k=-\infty}^{\infty} |a_k|^2$$

q)  $x(t) = \cos \omega_0 t$

$$\text{Power} = \frac{A^2}{2} = \frac{1}{2} W$$

$$a_0 = \frac{1}{2} \quad a_{-1} = \frac{1}{2}$$

$$\sum_{k=-\infty}^{\infty} |a_k|^2 = |a_0|^2 + |a_{-1}|^2 = \frac{1}{4} + \frac{1}{4} = \frac{1}{2} W$$

$$a_k = \frac{1}{T} \int_{-T/2}^{T/2} x(t) e^{-j\omega_0 t} dt$$

→ Parseval's theorem (DTFS)

$$\frac{1}{N} \sum_{m=0}^{N-1} |x(m)|^2 = \sum_{k=0}^{N-1} |c(k)|^2$$

N=4

$$c(k) = \frac{1}{4} \left[ e^{-j\pi k/2} + 2e^{-j\pi k} + 3e^{-j3\pi k/2} \right] \\ = \frac{1}{4} \left[ (e^{-j\pi k})^k + 2(e^{-j\pi})^k + 3(e^{-j3\pi/2})^k \right]$$

$$e^{-j\pi k} = \cos \pi - j \sin \pi = -1$$

$$e^{-j\pi} = \cos \pi - j \sin \pi = 1$$

$$e^{-j3\pi/2} = \cos \frac{3\pi}{2} - j \sin \frac{3\pi}{2} = +j$$

$$c(k) = \frac{1}{4} [(-j)^k + 2(-1)^k + 3(j)^k]$$

## → Impulse response

$$\delta(t) \rightarrow \boxed{\text{LTI system}} \rightarrow h(t)$$

$h(t)$  is fixed

↓ LT

$H(s)$  is fixed

$$1) y(t) = \int_{-\infty}^t u(\tau) d\tau$$

ILT ↑ ↓ LT

$$Y(s) = \frac{X(s)}{s} \Rightarrow \frac{Y(s)}{X(s)} = \frac{1}{s} = H(s)$$

↓ ILT

$$h(t) = u(t)$$

$$2) y(t) = ?$$

$$\begin{matrix} \text{LT} \\ \rightarrow h(t) = \end{matrix}$$

$$H(s) = \frac{Y(s)}{X(s)} \Rightarrow Y(s) = H(s) \cdot X(s)$$

↓ ILT

$$y(t) = h(t) * u(t)$$

↑ Convolution