

# Multi-carrier Modulation and OFDM

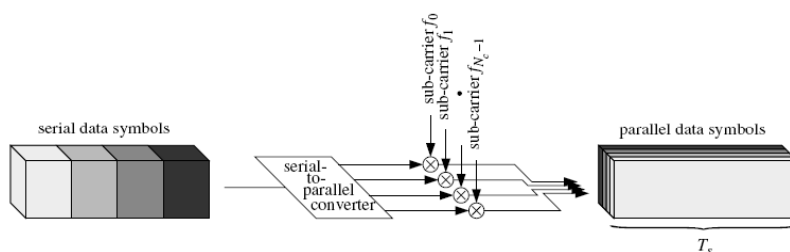
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## **Multi-carrier systems: basic idea**

- Typical mobile radio channel is a fading channel that is *flat* or *frequency selective*
- For high bandwidth applications channel is frequency selective and delay spread dictates throughput
- Multicarrier modulation is a technique where multiple low data rate carriers are combined by a transmitter to form a composite high data rate transmission
- In a classic multi-carrier system, the available spectrum is split into several non-overlapping frequency sub channels. The individual data elements are modulated into these sub channels and are thus frequency multiplexed

### Multi-carrier transmission

- Converts a high-data rate bit stream into multiple lower-data rate substreams
- Each substream is modulated onto a different carrier



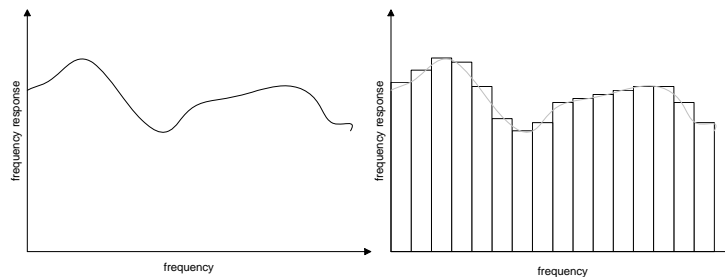
(Fazel and Kaiser, 2003)

### Advantage

- Since symbols are transmitted at a lower rate, the effects of delay spread are reduced
  - Reduced inter-symbol interference
- This in turn reduces the complexity of the equalizer

### Robustness to delay spread

- MC-modulation increases the symbol time by modulating into narrow sub-channels



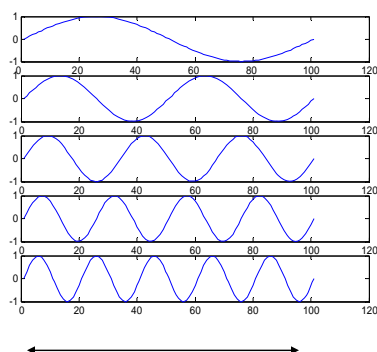
Channel frequency responses for a single carrier and multicarrier system. In the multicarrier system each sub channel only undergoes slight distortion

### Orthogonal Frequency Division Multiplexing

- In classic multicarrier system guard bands have to be inserted, resulting in poor spectral efficiency
- A more efficient approach is to allow the spectra of individual subcarriers to overlap
- Problem: If individual subcarriers are overlapping isn't there interference between carriers?
- Answer: No! If subcarrier tones are separated by the inverse of the signaling symbol duration, independent separation of frequency multiplexed tones is possible
  - This ensures that the spectra of individual sub channels are zeros at other subcarrier frequencies

### OFDM carrier

- Orthogonal waveforms are generated by using signals that have integer number of cycles in the duration  $T_s$


 $T_s$ 

Subcarriers in OFDM

### OFDM symbols

- Consider  $N_c$  complex-valued source symbols:  
 $S_n, n = 0, 1, \dots, N_c - 1$
- These symbols are transmitted in parallel using  $N_c$  sub-carriers
  - All of these symbols combined are referred to as an OFDM symbol
- If the source symbol duration is  $T_d$ , then the OFDM symbol duration is  $T_s = N_c T_d$
- The  $N_c$  sub-carriers have a spacing of

$$f = \frac{1}{T_s}$$

### OFDM carriers

- The baseband information of the  $k^{\text{th}}$  carrier can be expressed as

$$\underbrace{(x_k + jy_k)}_{\text{data symbol}} \underbrace{(\cos 2\pi kft + j \sin 2\pi kf)}_{k^{\text{th}} \text{ carrier}}$$

- The OFDM signal is the sum of all the signals in each of its subcarriers which can be written as (usually implemented using IFFT)

$$s(t) = \sum_{k=0}^{N_c-1} (x_k + jy_k)(\cos 2\pi kft + j \sin 2\pi kf)$$

### Recovering the individual symbols

- The individual modulated symbols at the receiver are recovered using the FFT
- The  $k$ th output from the FFT is:

$$z_k = \int_0^{T_s} s(t)(\cos 2\pi kft - j \sin 2\pi kf)dt = \sum_{n=0}^{N_c-1} \left\{ \int_0^{T_s} (x_n + jy_n) \cos 2\pi nf(\cos 2\pi kft - j \sin 2\pi kf)dt + j \int_0^{T_s} (x_n + jy_n) \sin 2\pi nf(\cos 2\pi kft - j \sin 2\pi kf)dt \right\}$$

### Solving these integrals...

- Trigonometry reminder:

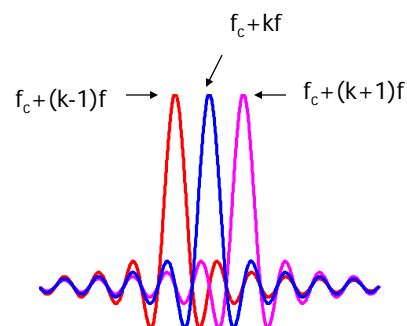
$$2 \cos A \cos B = \cos(A - B) + \cos(A + B)$$

$$2 \sin A \sin B = \cos(A - B) - \cos(A + B)$$

$$2 \sin A \cos B = \sin(A - B) + \sin(A + B)$$

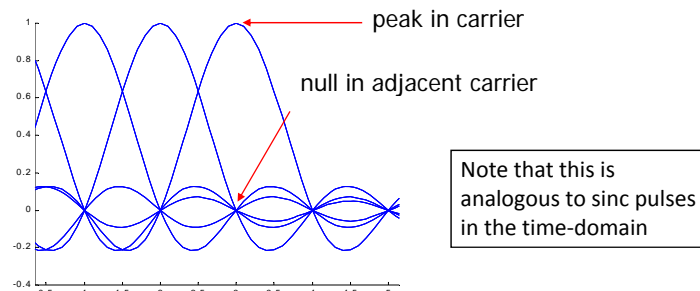
### OFDM spectrum

- The individual spectra of the subcarriers are sinc functions
- Zero crossings occur at every integer multiple of  $f$  and hence no Inter-Carrier Interference occurs in the frequency domain
- Note the analogy with time-domain sinc pulses

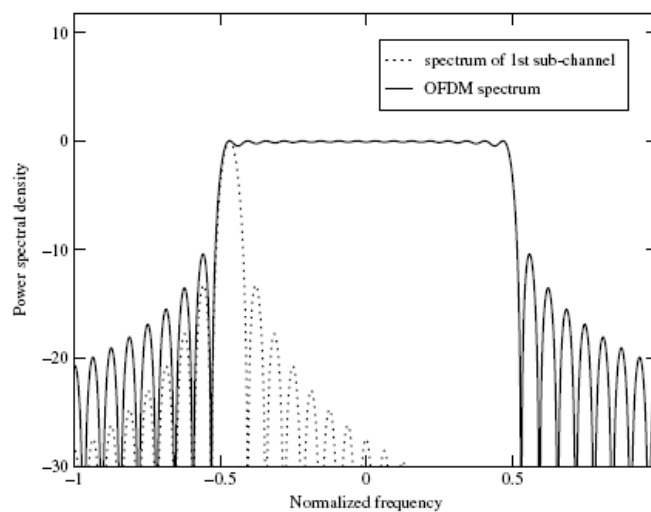


### Spectral efficiency

- For  $N$  sub-carriers, the bandwidth of conventional FDM is  $2N/T$  while that of OFDM is  $(N+1)/T$ . By allowing the sub-carrier spectra to overlap, OFDM improves the spectral efficiency.



### Power spectral density example ( $N_c = 16$ )



### Guard interval

- As  $N_c$  increases, the OFDM symbol duration  $T_s$  becomes large as compared to the duration of the impulse response  $\tau_{\max}$  of the channel
- To completely eliminate ISI, must add a guard interval  $T_g \geq \tau_{\max}$
- The new duration of the OFDM symbol is then

$$T'_s = T_s + T_g$$

### Sampled sequence with cyclic guard extension

- The length of the guard interval  $L_g$  must be

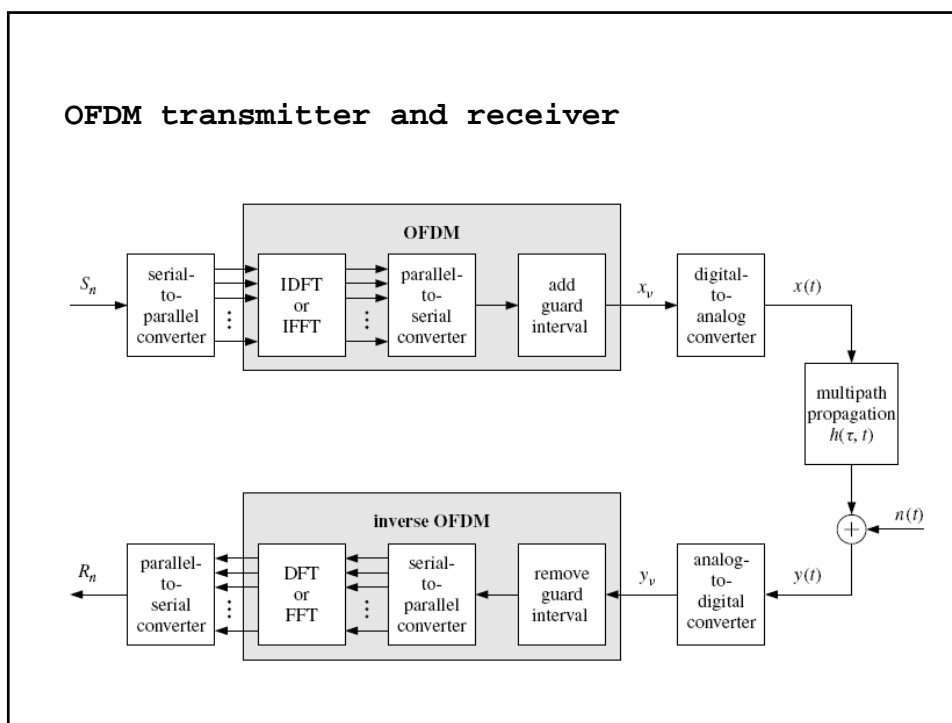
$$L_g \geq \left\lceil \frac{\tau_{\max} N_c}{T_s} \right\rceil$$

- The sampled signal with the guard extension becomes

$$x_v = \frac{1}{N_c} \sum_{n=0}^{N_c-1} S_n e^{j2\pi n v / N_c}, v = -L_g, \dots, N_c - 1$$



### OFDM transmitter and receiver



### Matrix notation

- Complex-valued source symbols, transmitted in parallel as an OFDM symbol

$$\mathbf{s} = (S_0 \quad S_1 \quad \dots \quad S_{N_c-1})^T$$

- $N_c \times N_c$  channel matrix

$$\mathbf{H} = \begin{pmatrix} H_{0,0} & 0 & \dots & 0 \\ 0 & H_{1,1} & & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \dots & H_{N_c-1,N_c-1} \end{pmatrix}$$

(Why is this a diagonal matrix?)

**Matrix notation (cont'd)**

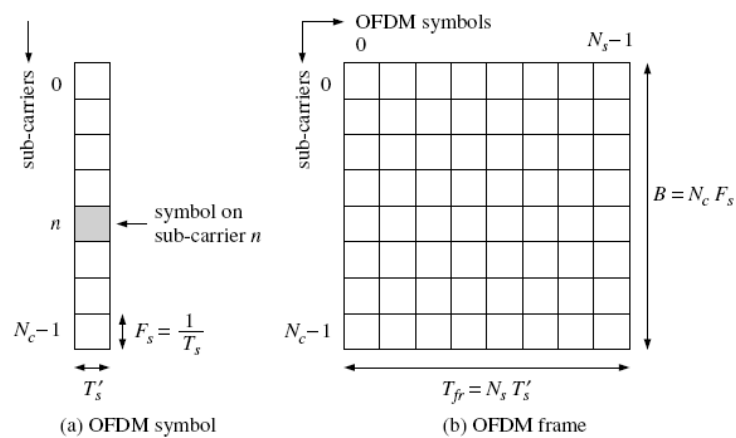
- Additive noise

$$\mathbf{n} = (N_0 \quad N_1 \quad \dots \quad N_{N_c-1})^T$$

- Received signals

$$\mathbf{r} = (R_0 \quad R_1 \quad \dots \quad R_{N_c-1})^T$$

$$\mathbf{r} = \mathbf{H}\mathbf{s} + \mathbf{n}$$

**An OFDM frame**

**OFDM advantages**

- High spectral efficiency for large number of subcarriers: nearly rectangular frequency-domain representation of signal
- Low-complexity receivers: due to low ICI and ISI if guard interval is long enough
- Flexible spectrum adaptation: good for DSA
- Different modulation can be applied to different subcarriers to suit the transmission conditions on each subcarrier

**OFDM disadvantages**

- High peak-to-average power ratio (PAPR): requires highly linear power amplifiers
- Some loss of spectral efficiency due to guard interval
- Average frequency and time synchronization is required
- More sensitive to Doppler spread than single-carrier systems

**OFDM example: DVB-T**

Bandwidth	8 MHz	
# of carriers	1705 (2k FFT)	6817 (8k FFT)
Symbol duration $T_s$	224 $\mu$ s	896 $\mu$ s
Carrier spacing $F_s$	4.464 kHz	1.116 kHz
Guard time $T_g$	$T_s/32, T_s/16, T_s/8, T_s/4$	
Modulation	QPSK, 16-QAM, 64-QAM	
FEC coding	Reed Solomon + convolutional with code rate $\frac{1}{2}$ up to $\frac{7}{8}$	
Max. data rate	31.7 Mbps	

**OFDM example: IEEE 802.11a**

Bandwidth	20 MHz
# of carriers	52 (64 FFT)
Symbol duration $T_s$	4 $\mu$ s
Carrier spacing $F_s$	312.5 kHz
Guard time $T_g$	0.8 $\mu$ s
Modulation	BPSK, QPSK, 16-QAM, 64-QAM
FEC coding	Convolutional code with rate $\frac{1}{2}$ up to $\frac{3}{4}$
Max. data rate	54 Mbps

**OFDM example: IEEE 802.16a**

Bandwidth	From 1.5 to 28 MHz
# of carriers	256
Symbol duration $T_s$	from 8 to 125 $\mu$ s (depending on the bandwidth)
Guard time $T_g$	from 1/32 up to 1/4 of $T_s$
Modulation	QPSK, 16-QAM, 64-QAM
FEC coding	Reed Solomon + convolutional code with rate 1/2 up to 5/6

**OFDM applications**

- Wireline
  - Asymmetric Digital Subscriber Loop (ADSL)
- Wireless
  - Digital Audio Broadcasting (DAB)
  - Digital Video Broadcasting-Terrestrial (DVB-T)
  - Integrated Services Digital Broadcasting-Terrestrial (ISDB-T)
  - Wireless LAN (IEEE 802.11(a), HiperLAN/2)
  - Wireless MAN (IEEE 802.16 a/b)