

Lecture 2

VE 311 Analog Circuits

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Recap of Last Lecture



- Logistics
- Analog Circuits
- Moore's Law

To Be Covered In This Lecture



- Scaling
- Review of 215 (Thevenin)
- Semiconductor Basics

RF, mm-Wave, and THz chips





Imaging



Communication



Biomedical

- Now we can interact with higher and higher frequencies.
- What my lab explores.



Apple A-series As an Example





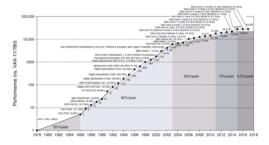
Alternative Processing Power (A9-A14)

- 16 core NPU (11 TOPS)
- ML accelerators
- GPU and Rendering
- Facial recognition
- Speech recognition
- Augmented reality
- ullet Emergence of dedicated AI accelerator ASICs are 10 imes more efficient than CPU.
- Historically CPU doubles every 24 mos., it's now down to 30%
- We're now seeing 100%+ processing power improvements for SoC



CPU Performance Trends





Performance of CPU over the years (Source: HENNESSY,2017)

- Dennard scaling in CPU has come to an end.
- Now transistors can interact with 100 GHz, why the clock is still 4 GHz?



Dennard Scaling



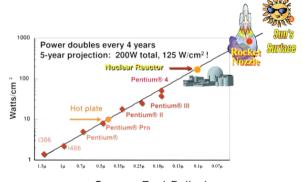
Transistor properties	dennardian	post dennard
Quantity	s^2	s^2
Frequency	s	s
Capacitance	1/s	1/s
V_{dd}^2	$1/s^2$	1
$power = V_{dd}^2 \cdot F \cdot C \cdot Q$	1	s^2

- As the size of the transistors shrunk, and the voltage was reduced, circuits could operate at higher frequencies at the same power.
- Dennard scaling ignored the leakage current, saturation velocity, and threshold voltage, which establish a baseline of power per transistor.



CPU Performance Trends



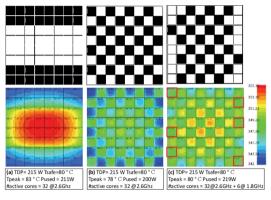


Source: Fred Pollack

Power density increases exponentially.



Dark Silicon



Heat map of multi-core processors

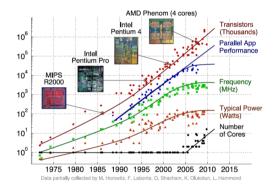


Solutions:

- Thermal management
- Power management
- Multi-thread mapping
- Degree of-parallelism
- Frequency/Supply (Dim)

In a 10 nm process, dark silicon can be 50% $\,\,80\%$

Power and Performance



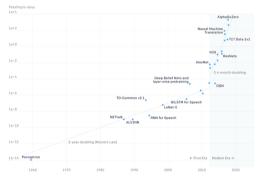
Performance of CPU over the years



- These created a "Power Wall" that has limited practical processor frequency to around 4 GHz since 2006.
- Improvement from multi-core and accelerators, but not frequency.

Ever-increasing Need





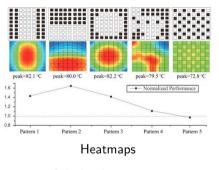
Requirement of computational power over algorithms

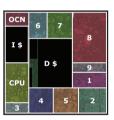
• The development of AI algorithms requires increasing computational power.



Area versus Power







ASIC Cores

- Because of dark silicon, power instead of area is the biggest challenge.
- In RISC, many power are wasted for reconfigurability, such as instruction fetching.

RF, mm-Wave, and THz chips





Samsung 9810



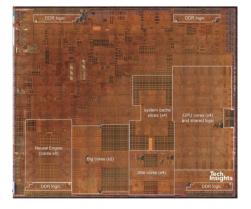
Apple A11 Bionic



Kirin 980

• Integration of CPU, GPU, and XPU on the same die is possible.

State of the Art SOCs



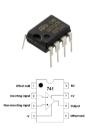


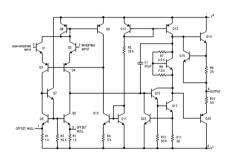
- 2x Large CPUs
- 4x Small CPUs
- GPUs
- Neural processing unit (NPU)
- Lots of memory
- DDR memory interfaces

Course Outcome



Understanding



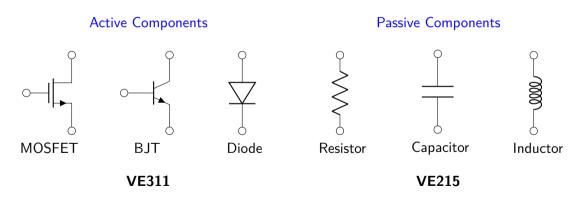


Innovation

- You are only bounded by your imagination.
- Welcome to our lab's weekly group meeting.
- Propose project ideas and seek guidance.
- Take upper-level courses.

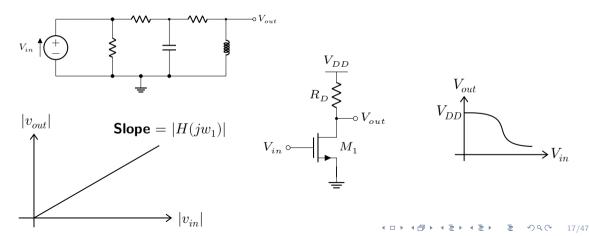
Active vs Passive Components





Linear vs Nonlinear Circuit





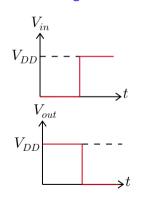
Analog vs Digital

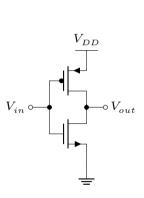


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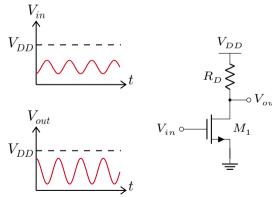
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VE312: Digital



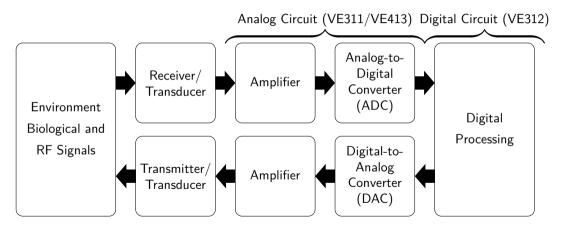


VE311: Analog



Analog Circuit in IC





IC Design Process



Hand calculations on paper based on proper approximations.



Pre-simulation: Schematic design and simulation on Spice.

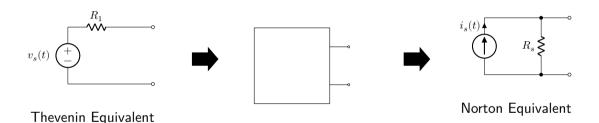


Post-simulation: Layout drawing, simulation and design rule check on Cadence.

Tapeout: Layout design sent to IC manufacturers.

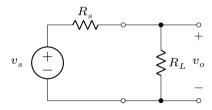






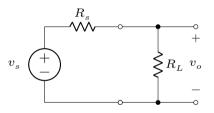
• Read Sedra and Smith, Section 1.1





- What is the voltage at the output?
- We should know Thevenin and Norton Equivalents of a network.

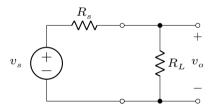






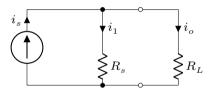
• If the resistance is replaced with an impedance, the same principle shall hold.





 \bullet Do we want larger or smaller R_s to get the largest output voltage?



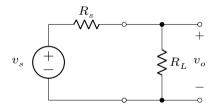


$$i_o = i_s \frac{R_s}{R_L + R_s}$$

Norton Equivalents

 \bullet From a current perspective, do we want large or small $R_s \mbox{?}$





- What is the condition for maximum power transfer to the load?
- Why is it important?

Power Calculation



Why is power $\frac{1}{2}\operatorname{Re}\left\{VI^{*}\right\}$

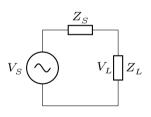
$$s(t) = v(t) \cdot i(t) = V_M \cdot \cos(\omega t + \phi_V) \cdot I_M \cdot \cos(\omega t + \phi_I)$$
 (1)

$$= \frac{V_M I_M}{2} \cdot \left[\cos\left(\phi_V - \phi_I\right) + \cos\left(2\omega t + \phi_V + \phi_I\right)\right] \tag{2}$$

So the average power is nothing more than the non-time varying component.

Therefore, we create the complex power notion S=P+jQ





What is the power delivered to the load?

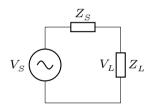
source impedance

$$Z_s = R_s + jX_s \tag{3}$$

load impedance

$$Z_L = R_L + jX_L \tag{4}$$





$$P = \frac{1}{2}V_L^2 \operatorname{Re}\left(\frac{1}{Z_L}\right) = \frac{1}{2}V_S^2 \left|\frac{Z_L}{Z_S + Z_L}\right|^2 \operatorname{Re}\left(\frac{1}{Z_L}\right)$$
 (5)

$$P = \frac{1}{2} V_S^2 \frac{R_L}{(R_S + R_L)^2 + (X_S + X_L)^2}$$
 (6)

power delivered to load as function of circuit parameters



$$P = \frac{1}{2} V_S^2 \frac{R_L}{(R_S + R_L)^2 + (X_S + X_L)^2} \tag{7}$$

- To find the maximum, we can take partial derivative and set them to be zeros
- We can do $\frac{\partial P}{\partial R_L}=0$ and $\frac{\partial P}{\partial X_L}=0$

$$\begin{cases} R_S^2 - R_L^2 + (X_L + X_S)^2 = 0\\ X_L (X_L + X_S) = 0 \end{cases}$$
 (8)

$$\begin{cases}
R_S = R_L \\
X_L = -X_S
\end{cases}$$
(9)



$$P = \frac{1}{2}V_S^2 \frac{R_L}{(R_S + R_L)^2 + (X_S + X_L)^2} \tag{10}$$

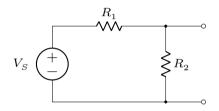
Now we plug it back into the equation

$$P = \frac{V_S^2}{8R_S} \tag{11}$$

• It represents the highest power that can be extracted from the source.

Finding Thevenin Equivalent Voltage





To find V_{TH}

- Open circuit output terminals
- ullet Find V_{OC}

$$V_{TH} = V_{OC} \tag{12}$$



Finding Thevenin Equivalent Resistance



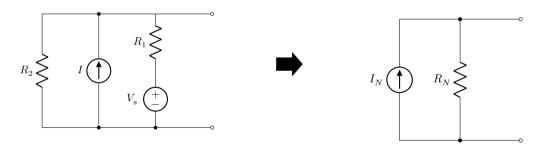
To find R_{TH}

- Zero out independent sources
- Current sources are replaced with an open circuit and voltage sources are replaced with a short circuit.
- Determine the equivalent resistance seen through the port by applying a test source $(V_T \ {\rm or} \ I_T)$

$$R_{TH} = V_T / I_T \tag{13}$$

Practice: Finding Norton Equivalent

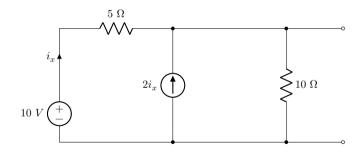




• Make sure you know this by heart, practice Thevenin model as well.

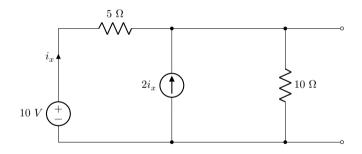
Thevenin Voltage of Dependent Sources



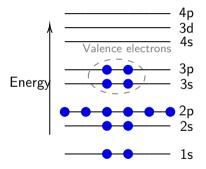


Thevenin Resistance of Dependent Sources



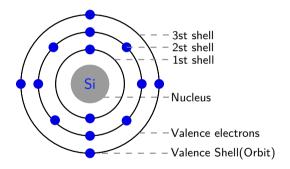






• Pauli exclusion principle requires that each electron must have a distinct energy state defined by a unique set of quantum numbers.



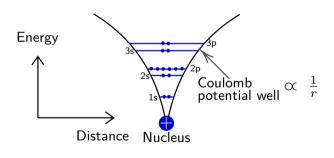


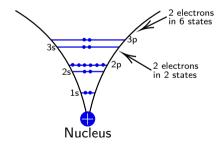
- Silicon has its inner shell (1s, 2s and 2p orbitals) totally filled with electrons.
- It's outer shell has 2 valence electrons in 3s orbital and 2 valence electrons in 3p orbital.



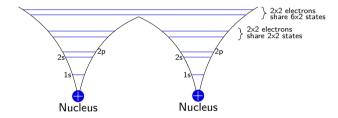


• When two silicon atoms are far away from each other, there is no interaction.







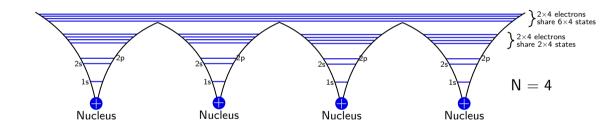


When two silicon atoms are close enough to each other:

- Wavefunctions overlap and potential wells are influenced by neighboring nucleus
- The valence electrons become delocalized (e.g. through tunneling).
- Each state splits into N substates, where N is number of atoms.



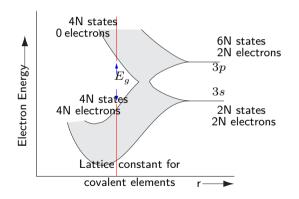




• Discrete states grow into bands when N is large.

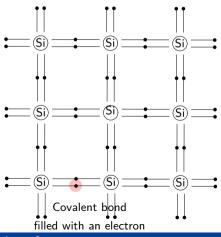






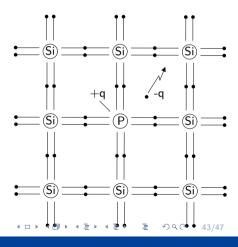
- ullet the band gap energy (1.12 $e{
 m V}$) is relatively small
- electrons with sufficient thermal energy can jump from valence band to conduction band



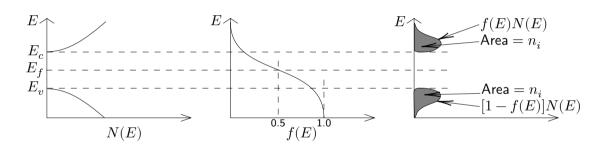




Electrons with sufficient thermal energy can jump from valence band to conduction band.







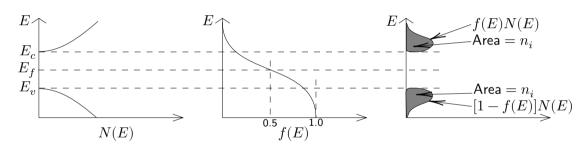
Density of States $/(\mathrm{cm}^3*\mathrm{J})$

Fermi-Dirac Distribution

$$f(E) = \frac{1}{\exp(\frac{E - E_f}{k \text{T}}) + 1}$$
 (14)



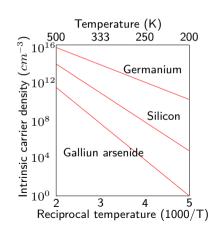




$$n = \int_{E_c}^{\infty} f(E)N(E)dE = n_i \qquad \text{(15)} \qquad p = \int_{-\infty}^{E_v} [1 - f(E)]N(E)dE = n_i \quad \text{(16)}$$







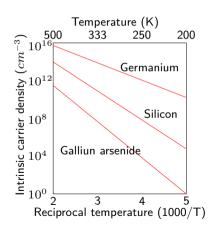
$$np = n_i^2 = BT^3 \exp(-\frac{E_G}{kT})$$
 (17)

k (Boltzmann's Constant)

$$= 1.38 \times 10^{-23} \text{ J/K} = 8.62 \times 10^{-5} \text{ eV/K}$$
 (18)

	B (K ⁻³ ·cm ⁻⁶)	E_G (eV)
Si	1.08×10^{31}	1.12
Ge	2.31×10^{30}	0.66
GaAs	1.27×10^{29}	1.42





At 300K

$$n_i^2 = (1.08 \cdot 10^{31}) \cdot 300^3 \cdot e^{\frac{-1.12}{(8.62 \cdot 10^{-5}) \cdot 300}}$$
$$= 4.52 \times 10^{19} (1/\text{cm}^6)$$
(19)

$$n_i = 6.73 \times 10^9 (1/\mathrm{cm}^3) \approx 10^{10} (1/\mathrm{cm}^3) \qquad \text{(20)}$$