SIMPLIFIED HYBRID MODEL UNIT-8

SIMPLIFIED HYBRID MODEL

In most practical cases it is appropriate to obtain approximate values of A_V , A_i etc rather than calculating exact values. How the circuit can be modified without greatly reducing the accuracy. Fig. 4 shows the CE amplifier equivalent circuit in terms of h-parameters Since 1 / h_{oe} in parallel with R_L is approximately equal to R_L if 1 / $h_{oe} >> R_L$ then h_{oe} may be neglected. Under these conditions.

$$I_c = h_{fe} I_B$$
.

$$h_{re} \ v_c = h_{re} \ I_c \ R_L = h_{re} \ h_{fe} \ I_b \ R_L$$
 .

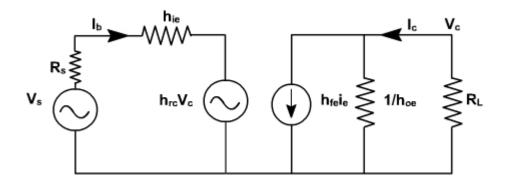


Fig. 4

Since $h_{fe}h_{re} \gg 0.01$, this voltage may be neglected in comparison with h_{ic} I_b drop across h_{ie} provided R_L is not very large. If load resistance R_L is small than hoe and h_{re} can be neglected.

$$\begin{split} A_I &= -\frac{h_{fe}}{1 + h_{oe} \ R_L} \quad \approx - h_{fe} \\ R_i &= h_{ie} \\ A_{ij} &= \frac{A_I \ R_L}{R_i} = -\frac{h_{fe} \ R_L}{h_{io}} \end{split}$$

Output impedence seems to be infinite. When $V_s=0$, and an external voltage is applied at the output we fined $I_b=0$, $I_C=0$. True value depends upon R_S and lies between 40 K and 80K.

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On the same lines, the calculations for CC and CB can be done.

CE amplifier with an emitter resistor:

The voltage gain of a CE stage depends upon h_{fe} . This transistor parameter depends upon temperature, aging and the operating point. Moreover, h_{fe} may vary widely from device to device, even for same type of transistor. To stabilize voltage gain A $_V$ of each stage, it should be independent of h_{fe} . A simple and effective way is to connect an emitter resistor R_e as shown in fig. 5. The resistor provides negative feedback and provide stabilization.

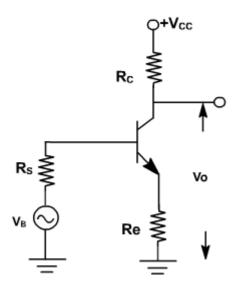


Fig. 5

An approximate analysis of the circuit can be made using the simplified model.

Current gain
$$A_i = \frac{I_L}{I_b} = -\frac{I_C}{I_b} = -\frac{h_{fe} I_b}{I_b}$$
$$= -h_{fe}$$

It is unaffected by the addition of $R_{\rm C}$.

Input resistance is given by

$$\begin{aligned} R_{i} &= \frac{V_{i}}{I_{b}} \\ &= \frac{h_{ie} I_{b} + (1 + h_{fe}) I_{b} R_{e}}{I_{b}} \\ &= h_{ie} = (1 + h_{fe}) R_{e} \end{aligned}$$

The input resistance increases by $(1+h_{\rm fe})R_{\rm e}$

$$A_v = \frac{A_i R_L}{R_i} = \frac{-h_{\text{fe}} \ R_L}{h_{ie} + (1 + h_{fe}) R_e}$$

Clearly, the addition of Re reduces the voltage gain.

If
$$(1+h_{\rm fe})R_{\rm e} >> h_{\rm ie}$$
 and $h_{\rm fe} >> 1$

then

$$A_{v} = \frac{-h_{fe} R_{L}}{(1+h_{fe})R_{e}} \approx -\frac{R_{L}}{R_{e}}$$

Subject to above approximation A_{V} is completely stable. The output resistance is infinite for the approximate model.