# **Simulation Example**

# 1 Proposed Method: Simulation and Analysis

This section introduces a simulation example that convincingly demonstrates the advantages of the proposed method. Accessible via https://github.com/lahe1/SMC\_simulation, the simulation example highlights the proposed method's broader applicability when compared to the method presented in [23], [24], and showcases superior precision in contrast to the open-loop state estimation methods detailed in [14]–[16]. Specifically, the simulation comprises five parts, as illustrated in Fig. 1. We proceed to introduce these parts as follows.

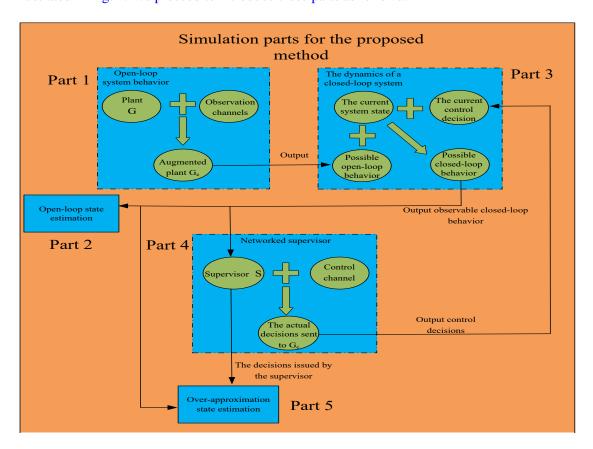


Figure 1: The simulation parts.

#### Part 1. Modeling open-loop system behavior:

• Given an NTDES  $\Re = (G, oc, cc)$ , this part augments plant G by incorporating the dynamics of the observation channels oc into an augmented plant  $G_e$ , which models the open-loop system behavior. The augmented plant is defined in Definition 4 on Page 5 of the paper.

## Part 2. Computing open-loop state estimates:

• When an observable event occurs, this part computes the open-loop state estimate and then uses it to highlight the precision of the proposed approach.

## Part 3. Modeling closed-loop system dynamics:

• This part models the dynamics of a closed-loop system  $S/\Re$  controlled by a supervisor S. First, compute the active events of the current system state based on the augmented plant  $G_e$ , which represents the possible open-loop behavior at that state. Next, determine which of those active events can occur at the current state given the current control decision; these represent the possible closed-loop behavior at that state. Finally, from among those events, we randomly select one to occur.

### Part 4. Modeling a networked supervisor:

- This part models a networked supervisor consisting of a supervisor S and a control channel with an upper bound cc on control delays and losses. Once a control decision is issued by S, it is first used to compute the current over-approximation state estimate. Subsequently, this decision is transmitted through the control channel. The control channel models the transmission of these decisions, accounting for delays and losses, and then outputs these decisions to the augmented plant. The entire process simulates the dynamics of the networked supervisor.
- To highlight the broader applicability of the proposed method, we conduct simulations for a non-FIFO control channel (see 1.1) and a FIFO one (see 1.2).

#### Part 5. Implementing the over-approximation state estimation method:

• In this part, the developed method, as detailed in Algorithm 1 on page 11 of the revised paper, is implemented to compute current over-approximation state estimates upon the occurrence of observable events and the issuance of control decisions.

For the simulation, we consider an NTDES  $\mathfrak{N}=(G,oc,cc)$ . The plant G is depicted in Fig. 2. The observation channels  $oc=(oc_1,oc_2)$  satisfies  $oc_1=(\Sigma_{o,1}=\{a\},\Sigma_{l,1}=\emptyset,N_{d,1}=1)$  and  $oc_2=(\Sigma_{o,2}=\{b,c\},\Sigma_{l,2}=\{c\},N_{d,2}=1)$ . The value of bound cc is set to 0 and 1, respectively. If the reviewer wishes to test larger values of cc, they can do so using the executable program named SMC\_simulation.exe, accessed at https://github.com/lahe1/SMC\_simulation. However, to reduce the number of pages in this response letter, we will not consider larger values of cc here.

The augmented plant  $G_e = (Q_e, \Sigma_e, \delta_e, q_{0_e})$  for  $\Re$  is shown in Fig. 3. Detailed information on the states of  $G_e$  is presented in Table 2, which tracks the states of  $G_e$  along with the corresponding global observation channel configurations. The set  $\Gamma$  of control decisions that a supervisor  $S_e$  can issue is displayed in Table 2. Note that the system state is initialized as  $q_{0e}$ , and the initial control decision used by  $G_e$  is randomly chosen from  $\Gamma$  to better demonstrate that the proposed method can deal with different supervisors.

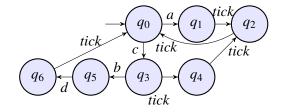


Figure 2: A TDES *G*:  $\Sigma_o = \{a, b, c, tick\}, \Sigma_{uo} = \{d\}, \Sigma_c = \{a, b\}, \Sigma_{uc} = \{c, d\}, \text{ and } \Sigma_{for} = \{b\}.$ 

Table 1: Control decision set Γ.
$\gamma_1 = (\{c, d, a_o, b_o, c_o, c_l\}, \emptyset)$
$\gamma_2 = (\{c, d, a_o, b_o, c_o, c_l, a\}, \emptyset)$
$\gamma_3 = (\{c, d, a_o, b_o, c_o, c_l, b\}, \emptyset)$
$\gamma_4 = (\{c, d, a_o, b_o, c_o, c_l, b\}, \{b\})$
$\gamma_5 = (\{c, d, a_o, b_o, c_o, c_l, a, b\}, \emptyset)$
$\gamma_6 = (\{c, d, a_o, b_o, c_o, c_l, a, b\}, \{b\})$

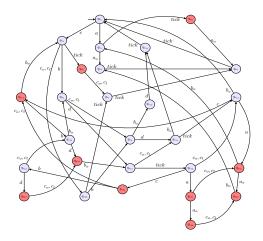


Figure 3: The augmented plant  $G_e$  for the NTDES.

Table 2: States of the augmented plant  $G_e$ .

$q_{e_1} = (q_1, ((a, 1), \varepsilon))$
$q_{e_3} = (q_1, (\varepsilon, \varepsilon))$
$q_{e_5} = (q_3, (\varepsilon, (c, 1)))$
$q_{e_7} = (q_4, (\varepsilon, \varepsilon))$
$q_{e_9} = (q_5, (\varepsilon, (b, 1)))$
$q_{e_{11}} = (q_6, (\varepsilon, (b, 1)))$
$q_{e_{13}} = (q_0, (\varepsilon, (b, 0)))$
$q_{e_{15}} = (q_5, (\varepsilon, (b, 0)(c, 1)(b, 1)))$
$q_{e_{17}} = (q_6, (\varepsilon, (c, 1)(b, 1)))$
$q_{e_{19}} = (q_1, ((a, 1), (b, 0)))$
$q_{e_{21}} = (q_1, ((a, 1), (c, 0)(b, 0))$
$q_{e_{23}} = (q_3, (\varepsilon, (c, 0)(b, 0)(c, 1)))$
$q_{e_{25}} = (q_6, (\varepsilon, (c, 0)(b, 0)(c, 1)(b, 1)))$

## 1.1 Proposed Method: Simulation and Analysis for non-FIFO Control Channels

The C++ code for the simulation can be accessed at https://github.com/lahe1/SMC\_simulation/blob/master/SMC\_simulation/SMC\_simulation.cpp, and the code's structure specifically for the non-FIFO control channel is illustrated in the flowchart shown in Fig. 4.

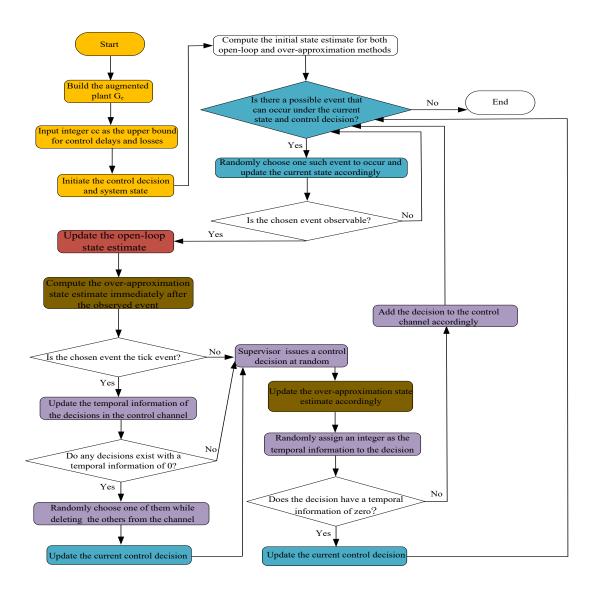


Figure 4: The algorithm flowchart for the simulation with non-FIFO control channel.

First, we construct the augmented plant  $G_e$  and set the initial system state to  $q_{0e}$ . Then, we input a control and delay bound cc. These steps are detailed in lines 1549–1938 of the provided C++ code. Next, we randomly select an initial control decision, implemented in line 718.

Subsequently, we compute the initial open-loop state estimate and the initial over-approximation state estimate, associated with lines 691 and 734, respectively. Then, we execute line 774 to ascertain the existence of a closed-loop behavior that can occur in the current state and control decision. If such behavior is not present, we stop the simulation. If it is present, we randomly select a feasible event to occur using line 796 and update the current state accordingly using line 804. Determining whether the selected event is observable is implemented in line 829.

After selecting an observable event, we accordingly compute the open-loop state estimate (line

- 850), followed by the computation of the over-approximation state estimate (line 855). The latter computation is based on the operator  $O_r$  as defined in Eqs. (11)–(13) of the revised paper.
- (1) If the selected event is not *tick* (as determined in line 885), we randomly select a control decision from the control decision set  $\Gamma$  to model the supervisor's issuing behavior (line 1006). Then, we incorporate the newly issued control decision to update the over-approximation state estimate, using the operator  $U_r$  as defined in Eqs. (9) and (10).

To model control delays, we assign a random integer value between 0 and cc (inclusive) as the temporal information to any control decision issued (line 1009). This temporal information indicates the number of tick events that must occur before the completion of the decision's transmission. If the assigned value for a control decision is zero, it indicates the absence of control delay and loss. As a result, the transmission of the decision through the channel occurs instantaneously, and we can immediately apply this decision to update the current control decision. After that, we ascertain the existence of a closed-loop behavior that can occur in the current state and control decision. The implementation of the decision handling without delays and losses can be found in lines 1020–1026. On the other hand, if the assigned value is greater than zero, indicating a control delay, the decision is sent to the control channel. The current control decision remain unchanged. Here, too, we check for the possibility of closed-loop behavior based on the current state and control decision (lines 1027–1036). The entire process for modeling control delays is illustrated in Fig. 5.

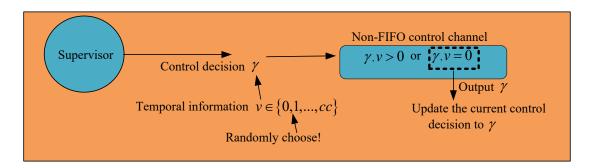


Figure 5: An illustration of modeling control delays.

(2) If the selected event is *tick*, we will update the temporal information for each control decision in the control channel by decreasing it by one (lines 885–893). After that, the control channel outputs all control decisions with temporal information valued at zero. To simulate control losses bounded by *cc*, we randomly select only one control decision from the output and use it to update the current control decision while discarding the others (lines 907–976). Subsequently, we randomly select a control decision to model the supervisor's issuing behavior after the occurrence of *tick*, and then execute the same steps as mentioned earlier. The entire process for modeling control losses is illustrated in Fig. 6.

Since the temporal information values we assign to control decisions are randomly selected from the range of 0 to cc, the decisions in the control channel are not required to follow a FIFO order when being output. Therefore, we have successfully simulated the delays and losses in a non-FIFO control channel, as desired.

Now, we present the simulation result. For brevity and to conserve the number of pages of the response letter, the tables below present data from the first ten observable events for each cc value tested. It should be noted that, as mentioned earlier, we use a random approach to simulate system dynamics, control decision issuance, and control delays and losses. Hence, even for the same cc, the data obtained from the simulation do not need to be the same.

We first provide simulation data with cc = 0, indicating no control delays and losses, as

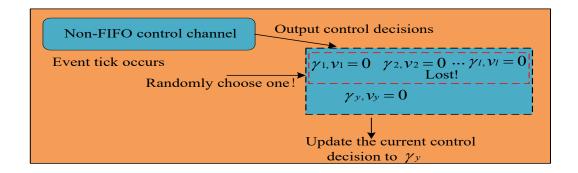


Figure 6: An illustration of modeling control losses.

illustrated in Table 3. Table 4 is presented to clarify the meanings of each entry listed in Table 3. For instance, let us sequentially introduce the entries of the forth row in Table 3. (1) *tick* represents the event observed by a supervisor. (2)  $\gamma_1$  is the control decision issued by the supervisor after observing *tick*. (3)  $q_{e4}$  represents a possible system's state after the occurrence of *tick*, estimating under the open-loop state estimation method. (4)  $\gamma_1$  denotes the control decision in use by the plant  $G_e$ . (5) "No" signifies the absence of control delays after the occurrence of *tick*. (6)  $q_{e4}$  (in the sixth column) represents a possible system's state after the occurrence of *tick* and the issuance of the decision  $\gamma_1$ , estimating under the over-approximation state estimation method. (7)  $q_{e4}$  is the actual state in which the system has reached until the next observable event.

Table 3: Simulation result: Non-FIFO and cc = 0

Table 3: Simulation result: Non-FIFO and $cc = 0$ .							
$\Sigma_{n,o}$	S	$E_{ol}$	Γ	Losses	$E_{\Gamma_S}$	$G_e$	
ε	$\gamma_1$	$q_{0e},q_{e1},q_{e5},q_{e8},q_{e26},q_{e9},q_{e17},q_{e11}$	$\gamma_1$	No	$q_{0e}, q_{e5}, q_{e26}$	$q_{0e}, q_{e5}$	
tick	$\gamma_3$	$q_{e2}, q_{e6}, q_{e7}, q_{e18}, q_{e13}, q_{e21}, q_{e23},$	$\gamma_3$	No	$q_{e6},q_{e7}$	$q_{e6},q_{e7}$	
		$q_{e14}, q_{e19}, q_{e24}, q_{e15}, q_{e25}, q_{e16}$					
tick	$\gamma_1$	$q_{e4}$	$\gamma_1$	No	$q_{e4}$	$q_{e4}$	
tick	$\gamma_4$	$q_{0e}, q_{e1}, q_{e5}, q_{e8},$	$\gamma_4$	No	$q_{0e}, q_{e5}, q_{e8}, q_{e26},$	$q_{0e}, q_{e5},$	
		$q_{e26}, q_{e9}, q_{e17}, q_{e11}$			$q_{e9}, q_{e17}, q_{e11}$	$q_{e26}, q_{e9}$	
$b_o$	$\gamma_4$	$q_{e10},q_{e12}$	$\gamma_4$	No	$q_{e10}, q_{e12}$	$q_{e10},q_{e12}$	
tick	$\gamma_2$	$q_{0e}, q_{e1}, q_{e5}, q_{e8}, q_{e26}, q_{e9}, q_{e17}, q_{e11}$	$\gamma_2$	No	$q_{0e}, q_{e1}, q_{e5}, q_{e26}$	$q_{0e},q_{e5}$	
tick	γ5	$q_{e2}, q_{e6}, q_{e7}, q_{e18}, q_{e13}, q_{e21}, q_{e23},$	γ5	No	$q_{e2}, q_{e6}, q_{e7}$	$q_{e6}$	
		$q_{e14}, q_{e19}, q_{e24}, q_{e15}, q_{e25}, q_{e16}$					
$c_o$	$\gamma_4$	$q_{e7}, q_{e13}, q_{e19}, q_{e14}, q_{e15}, q_{e16}$	$\gamma_4$	No	$q_{e7}$	$q_{e7}$	
tick	$\gamma_1$	$q_{e4}$	$\gamma_1$	No	$q_{e4}$	$q_{e4}$	
tick	$\gamma_4$	$q_{0e}, q_{e1}, q_{e5}, q_{e8}, q_{e26}, q_{e9}, q_{e17}, q_{e11}$	$\gamma_4$	No	$q_{0e}, q_{e5}, q_{e8}, q_{e26},$	$q_{0e}, q_{e5},$	
					$q_{e9}, q_{e17}, q_{e11}$	$q_{e8},q_{e17}$	
tick	$\gamma_1$	$q_{e2}, q_{e6}, q_{e7}, q_{e18}, q_{e13}, q_{e21}, q_{e23},$	$\gamma_1$	No	$q_{e18}, q_{e13}, q_{e23}, q_{e14}$	$q_{e18}$	
		$q_{e14}, q_{e19}, q_{e24}, q_{e15}, q_{e25}, q_{e16}$					
$c_o$	$\gamma_2$	$q_{e7}, q_{e13}, q_{e19}, q_{e14}, q_{e15}, q_{e16}$	$\gamma_2$	No	$q_{e13}, q_{e14}, q_{e19}$	$q_{e13}$	

Notice that every entry in the second column of Table 3 is identical to the corresponding entry in the fourth column of the same row. Additionally, all entries in the fifth column are 'no.' The two observations indicate that the simulation has successfully implemented the scenario where cc = 0, i.e., the control channel has no delays or losses.

According to Table 3, the states that the system has actually reached are encompassed by both the open-loop and the over-approximation state estimates. Therefore, both methods can estimate the plant states. However, the proposed method is significantly more precise than the open-loop

Table 4. Explanation of Table 3.					
Column $\Sigma_{n,o}$ : The events observed by the supervisor					
Column S: The control decisions issued by the supervisor					
Column $E_{ol}$ : The open-loop state estimates after observing the events listed in the same rows					
Column Γ: The control decisions currently in use					
Column Losses: The control decisions currently lost					
Column $E_{\Gamma_S}$ : The most recently over-approximation state estimates					
Column $G_e$ : The actual states for the system					

method. For example, in the third row, the proposed method accurately estimates the actual system states, whereas the open-loop method estimates a large number of states that the system will not actually reach. We next provide simulation data with cc = 1, indicating that the control channel has control delays and losses bounded by 1, as illustrated in Table 5.

Table 5: Simulation result: Non-FIFO and cc = 1.

$\Sigma_{n,o}$	S	$\mid E_{ol} \mid$	Γ	Losses	$\mid E_{\Gamma_S} \mid$	$G_e$
ε	$\gamma_1$	$q_{0e}, q_{e1}, q_{e5}, q_{e8}, q_{e26}, q_{e9}, q_{e17}, q_{e11}$	$\gamma_1$	No	$q_{0e}, q_{e5}, q_{e26},$	$q_{0e}, q_{e5}$
tick	$\gamma_2$	$q_{e2}, q_{e6}, q_{e7}, q_{e18}, q_{e13}, q_{e21}, q_{e23},$	$\gamma_1$	No	$q_{e6}, q_{e7}$	$q_{e6}$
		$q_{e14}, q_{e19}, q_{e24}, q_{e15}, q_{e25}, q_{e16}$				
$c_o$	$\gamma_2$	$q_{e7}, q_{e13}, q_{e19}, q_{e14}, q_{e15}, q_{e16}$	$\gamma_1$	No	$q_{e7}$	$q_{e7}$
tick	$\gamma_1$	$q_{e4}$	$\gamma_1$	$\gamma_2$	$q_{e7}$	$q_{e7}$
tick	$\gamma_4$	$q_{0e}, q_{e1}, q_{e5}, q_{e8}, q_{e26}, q_{e9}, q_{e17}, q_{e11}$	$\gamma_1$	No	$q_{0e}, q_{e5}, q_{e8}, q_{e26}, q_{e9}, q_{e17}, q_{e11}$	$q_{0e},q_{e5}$
$c_o$	<b>γ</b> 3	$q_{e26}, q_{e9}, q_{e11}$	$\gamma_1$	No	$q_{e26}, q_{e9}, q_{e11}$	$q_{e26}$
tick	γ4	$q_{e7}, q_{e13}, q_{e14}, q_{e19}, q_{e15}, q_{e16}$	γ4	No	$q_{e7}, q_{e13}, q_{e14}, q_{e15}, q_{e16}$	$q_{e7}$
tick	$\gamma_4$	$q_{e4}$	$\gamma_4$	No	$q_{e4}$	$q_{e4}$
tick	<b>γ</b> 1	$q_{0e}, q_{e1}, q_{e5}, q_{e8}, q_{e26}, q_{e9}, q_{e17}, q_{e11}$	γ4	No	$q_{0e}, q_{e5}, q_{e8}, q_{e26}, q_{e9}, q_{e17}, q_{e11}$	$q_{0e}, q_{e5}, q_{e26}, q_{e9}$
$b_o$	γ5	$q_{e10}, q_{e12}$	γ <sub>5</sub>	No	$q_{e10}, q_{e12}$	$q_{e10}, q_{e12}$
tick	γ1	$q_{0e}, q_{e1}, q_{e5}, q_{e8}, q_{e26}, q_{e9}, q_{e17}, q_{e11}$	γ1	No	$q_{0e}, q_{e1}, q_{e5}, q_{e8},$ $q_{e26}, q_{e9}, q_{e17}, q_{e11}$	$q_{0e},q_{e5}$
$c_o$	$\gamma_2$	$q_{e26}, q_{e9}, q_{e11}$	$\gamma_1$	No	$q_{e26}, q_{e9}, q_{e11}$	$q_{e26}$

According to Table 5, the simulation has successfully modeled control delays and losses bounded by cc = 1 in the control channel. Moreover, the non-FIFO property of the control channel has also been well simulated. For instance, let us observe rows 10 and 11. Control decision  $\gamma_5$  is received by the plant  $G_e$  first, despite  $\gamma_1$  entering the control channel earlier.

By setting *cc* to 0 and 1, the simulation confirmed that the proposed method can be effectively applied to the non-FIFO control channel for estimating system states. It also showed that the proposed method is more precise than the open-loop approach.

### 1.2 Proposed Method: Simulation and Analysis for FIFO Control Channels

The simulation code's structure specifically for the FIFO control channel is illustrated in the flowchart shown in Fig. 7. Specifically, apart from the steps enclosed within the red dashed box, which are specific to simulating the properties of a FIFO control channel and differ from those in Fig. 4 for simulating a non-FIFO channel, all other steps are the same.

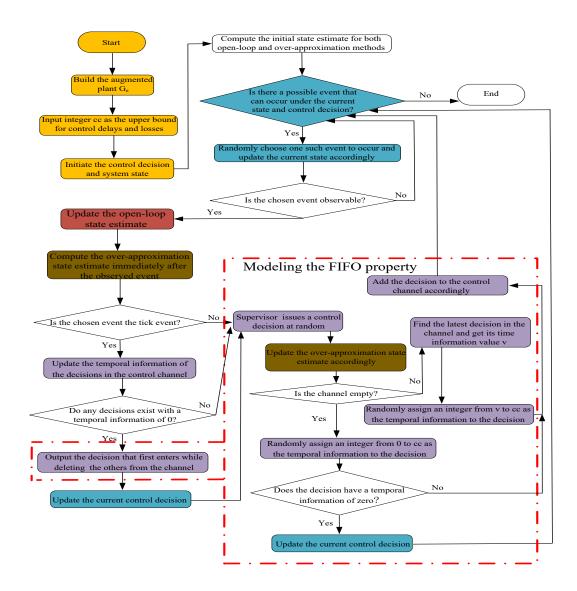


Figure 7: The algorithm flowchart for the simulation with FIFO control channel.

Let us consider the steps outlined within the red dashed box. When a supervisor issues a control decision  $\gamma$ , to model control delays, we initially check whether the control channel is empty (implemented in line 1440 of the C++ code available at https://github.com/lahe1/SMC\_simulation/blob/master/SMC\_simulation/SMC\_simulation.cpp).

(1) If the control channel is empty, we assign a randomly selected temporal information value

to the issued decision  $\gamma$  ranging from 0 to cc (line 1442).

(2) If the channel is not empty, we identify the last decision  $\gamma'$  entered and retrieve its time information value  $\nu$ . Subsequently, as implemented in line 1467, we assign a randomly selected time information value to the issued decision  $\gamma$  within the range of  $\nu$  to cc. This approach ensures that the time value assigned to  $\gamma$  is greater than or equal to  $\nu$ , thus simulating the FIFO characteristic where  $\gamma$  completes its transmission in the channel later than  $\gamma'$ .

In summary, the method for modeling control delays when the channel is empty is the same as illustrated in Fig. 6, while the case when the channel is not empty is shown in Fig. 8.

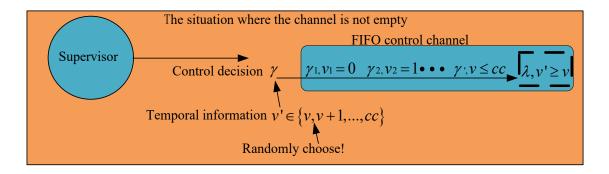


Figure 8: An illustration of modeling control delays for FIFO channel.

When the event *tick* occurs, we will update the temporal information for each control decision in the control channel (lines 1308–1324). After that, the control channel outputs all control decisions with temporal information valued at zero (line 1334). To simulate control losses, we select the control decision that first enters the channel from the output and use it to update the current control decision, while discarding the rest (lines 1353 and 1388). The entire method for modeling control losses is illustrated in Fig. 9. We finally provide simulation data with cc = 1, as illustrated in Table 6

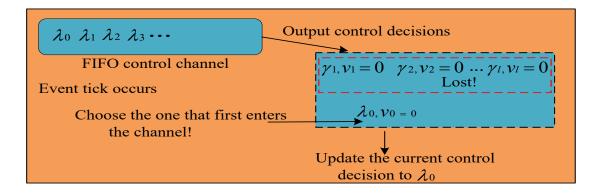


Figure 9: An illustration of modeling control losses for FIFO channel.

According to Table 6, the simulation has successfully modeled control delays and losses bounded by cc=1 in the control channel. Moreover, the FIFO characteristic of the control channel has also been well simulated, as no case has been found where a later control decision to enter the control channel completes transmission before an earlier one. Moreover, the simulation confirmed that the proposed method can be effectively applied to FIFO control channels for estimating system states. It also showed that the proposed method is more precise than the open-loop approach.

Table 6: Simulation result: FIFO and cc = 1.

$\Sigma_{n,o}$	S	$E_{ol}$	Γ	Losses	$E_{\Gamma_S}$	$G_e$
ε	$\gamma_1$	$q_{0e}, q_{e1}, q_{e5}, q_{e8}, q_{e26}, q_{e9},$	$\gamma_1$	No	$q_{0e}, q_{e5}, q_{e26}$	$q_{0e}, q_{e5}$
		$q_{e17}, q_{e11}$				
tick	$\gamma_4$	$q_{e2}, q_{e6}, q_{e7}, q_{e18}, q_{e13}, q_{e21},$	$\gamma_4$	No	$q_{e6},q_{e7}$	$q_{e6}$
		$q_{e23}, q_{e14}, q_{e19}, q_{e24}, q_{e15},$				
		$q_{e25}, q_{e16}$				
$c_o$	$\gamma_2$	$q_{e7}, q_{e13}, q_{e19}, q_{e14}, q_{e15}, q_{e16}$	$\gamma_2$	No	$q_{e7}$	$q_{e7}$
tick	$\gamma_6$	$q_{e4}$	$\gamma_2$	No	$q_{e4}$	$q_{e4}$
tick	$\gamma_6$	$q_{0e}, q_{e1}, q_{e5}, q_{e8}, q_{e26}, q_{e9},$	γ <sub>6</sub>	No	$q_{0e}, q_{e1}, q_{e5}, q_{e8}, q_{e26},$	$q_{0e}, q_{e5}, q_{e8},$
		$q_{e17}, q_{e11}$			$q_{e9}, q_{e17}, q_{e11}$	$q_{e17}, q_{e11}$
tick	$\gamma_2$	$q_{e2}, q_{e6}, q_{e7}, q_{e18}, q_{e13}, q_{e21},$	γ <sub>6</sub>	No	$q_{e2}, q_{e18}, q_{e13}, q_{e21},$	$q_{e13}, q_{e14}$
		$q_{e23}, q_{e14}, q_{e19}, q_{e24}, q_{e15},$			$q_{e23}, q_{e14}, q_{e19}, q_{e24},$	
		$q_{e25}, q_{e16}$			$q_{e15}, q_{e25}, q_{e16}$	
$b_o$	$\gamma_3$	$q_{0e}, q_{e5}, q_{e1}, q_{e8}, q_{e17}, q_{e26},$	$\gamma_6$	No	$q_{0e}, q_{e5}, q_{e1}, q_{e8}, q_{e17},$	$q_{e5}$
		$q_{e9}, q_{e11}$			$q_{e26}, q_{e9}, q_{e11}$	
$c_o$	$\gamma_5$	$q_{e26}, q_{e9}, q_{e11}$	<b>γ</b> 6	No	$q_{e26}, q_{e9}, q_{e11}$	$q_{e26}, q_{e9}$
$b_o$	$\gamma_6$	$q_{0e},q_{e12}$	γ <sub>6</sub>	No	$q_{0e},q_{e12}$	$q_{0e},q_{e12}$
tick	$\gamma_3$	$q_{0e}, q_{e1}, q_{e5}, q_{e8}, q_{e26}, q_{e9},$	$\gamma_2$	$\gamma_3,\gamma_5,\gamma_6$	$q_{0e}, q_{e1}, q_{e5}, q_{e8}, q_{e26},$	$q_{0e}, q_{e1}, q_{e26}$
		$q_{e17}, q_{e11}$			$q_{e9}, q_{e17}, q_{e11}$	
tick	$\gamma_2$	$q_{e2}, q_{e6}, q_{e7}, q_{e18}, q_{e13}, q_{e21},$	$\gamma_2$	No	$q_{e2}, q_{e6}, q_{e7}, q_{e18}, q_{e13},$	$q_{e7}$
		$q_{e23}, q_{e14}, q_{e19}, q_{e24}, q_{e15},$			$q_{e21}, q_{e23}, q_{e14}, q_{e19},$	
		$q_{e25}, q_{e16}$			$q_{e24}, q_{e15}, q_{e25}, q_{e16}$	
tick	$\gamma_1$	$q_{e4}$	$\gamma_1$	No	$q_{e4}$	$q_{e4}$

## 1.3 Simulation Conclusion

Through the simulation, we have verified the applicability of the proposed over-approximation state estimation method to both non-FIFO (see 1.1) and FIFO (see 1.2) control channels. This demonstrates its broader potential compared to the approach presented in [23], [24], which is limited to FIFO scenarios.

Moreover, the simulation shows that the proposed method, considering the impact of control decisions on state estimation, offers greater precision than the open-loop state estimation approach proposed in [4]–[7], [14]–[16], [19].

Additionally, the proposed method is shown to effectively handle control delays and losses, taking into account the passage of time modeled by the occurrences of the event *tick*. This is crucial since real-world delays and losses are typically measured by time passage [17], [19], [20]. Therefore, the proposed method is more practical and suitable for real-world systems than the time-agnostic state estimation methods in [4]–[8], [19], [24]. To provide a concise overview and aid in the comprehension of the suitability of the various methods, we have compiled table 7.

Table 7: Overview of the existing state estimation methods.

	Citation	Timed	FIFO	non-FIFO	Control			
ſ	[4],[5],[6],[7]	Х	✓	✓	Х			
	[8]	X	✓	✓	<b>√</b>			
	[14], [15],[16]	✓	✓	✓	Х			
ſ	[19]	Х	✓	✓	Х			
ſ	[23]	✓	✓	Х	✓			
ſ	[24]	Х	✓	Х	✓			
Ī	The proposed method	✓	✓	✓	✓			

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