

AN4507 Application note

PWM resolution enhancement through dithering technique for STM32 advanced-configuration, general-purpose and lite timers

Introduction

Nowadays power switching electronics are exhibiting remarkable performance enhancements. The switching frequency of power switching electronics is constantly increasing to meet the modern power conversion systems requirements. This is becoming challenging in what concerns control techniques. For digitally-controlled Pulse Width Modulation generators, there is a trade-off between the switching frequency and the duty-cycle fine tuning. Achieving at once high resolution and high switching frequency implies that the control circuitries operate at high frequencies.

Within this application note we are presenting the dithering technique which enhances the PWM resolution while keeping the same switching frequency and the same operating frequency for the control circuitry.

This application note supports the demonstration of the PWM dithering technique using one of the STM32 General Purpose Timers (a 16-bit Timer). The implementation and the provided STSW-STM32151 firmware are targeted at STM32F3DISCOVERY discovery kit but can be tailored easily to any of the STM32 MCUs, regardless of the hardware development environment.

Table 1. Applicable firmware and tools

| Туре | Part numbers | | | | |
|----------|-------------------|--|--|--|--|
| Firmware | STSW-STM32151. | | | | |
| Tools | STM32F3DISCOVERY. | | | | |

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1 PWM dithering technique concept

1.1 Trade-off between switching frequency and PWM resolution

There is a tight relationship between the switching frequency which is equivalent to the PWM frequency and the PWM resolution. Increasing the PWM resolution implies to decrease the PWM frequency while maintaining a constant timer clock frequency. For a general purpose STM32 timer, the minimal clock frequency required to clock the timer to achieve a given PWM frequency and resolution is given by the formula below:

Timer clock = PWM frequency \times 2^{PWM resolution}

The table below provides some examples based on the above formula.

STM32 16-bit timer **PWM** resolution **PWM** frequency 72 MHz 16 bit ~1.1 kHz 72 MHz 14 bit ~4.4 kHz 72 MHz 12 bit ~17.5 kHz 72 MHz 10 bit ~70 kHz 72 MHz 8 bit ~281 kHz 72 MHz 6 bit ~1.125 MHz 72 MHz 4 bit ~4.5 MHz

Table 2. PWM frequency versus PWM Resolution

1.2 PWM dithering

1.2.1 Introduction

To overcome the trade-off presented above, the PWM dithering technique proposes to enhance the PWM resolution while maintaining a constant PWM frequency and without any need to increase the timer clock frequency.

Theoretically, any desired number of added resolution bits is possible. Practically, the added resolution bits are limited to a few ones, in particular due to the firmware footprint exponential increase (See Section 3.2.3: Firmware Footprint).

1.2.2 Concept

The PWM dithering is done by making the PWM duty-cycle not constant anymore. The idea is to adjust the duty-cycle by one LSB with a repetitive pattern over a given number of consecutive PWM periods. An external low-pass filter is used to cut the extra switching components and the result consists in an added DC offset with a resolution less than one



duty cycle LSB. The duty-cycle adjustment can be made following two possible approaches. The first approach consists in subtracting one LSB from the periods where a duty-cycle adjustment should be made. In this case the DC offset is going to be negative. The other way is by adding one LSB to the periods where a duty-cycle adjustment should be made. In this case the DC offset is going to be positive.

The added DC offset is proportional to the ratio between the number of PWM periods where duty-cycle adjustment is made, divided by the number of consecutive PWM periods making one PWM dithering pattern. Whatever the duty-cycle adjustment pattern, the obtained ratio is always a positive sub-one fractional value. With this dithering technique, external devices can be controlled with a resolution higher than the original PWM resolution.

The resulting PWM resolution is given by the below formula:

The PWM_{Dither Resolution} is given by the below formula

$$N_{PWM_Adjustment_Periods} = 2^{PWM_{Dither_Resolution}}$$

where N_{PWM_Adjustment_Periods} is the number of PWM periods required to constitute one duty-cycle adjustment pattern (also called, PWM dithering pattern).

From the precedent formula it is obvious that the required number of periods for a complete duty-cycle adjustment pattern is a power-of-2 of the number of resolution bits, added with the dithering technique.

Just for simplicity sake, the below example is given for 1-bit enhanced PWM resolution. The concept remains the same for larger numbers of added resolution bits.

Within this example we want to add only one extra bit of resolution using the dithering technique. The duty-cycle adjustment pattern will occupy 2 PWM periods. A sequence of PWM periods is presented in *Figure 1* below. The PWM dithering technique consists in adjusting the duty-cycle of two consecutive PWM periods. The pattern could be the same for the next two PWM periods to maintain the same output value or it can be changed to a new one.



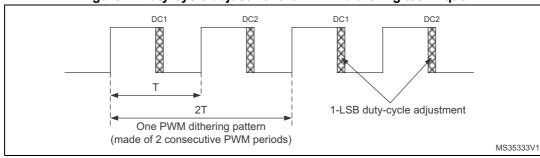


Figure 1. Duty-cycle adjustment for PWM dithering technique

For 1-added-resolution-bit PWM dithering, four patterns are possible. These patterns are listed in *Table 3* below.

| Tallot of a cooling patterns for a later and a later and | | | | | | | | |
|--|-------|----------------------|--|--|--|--|--|--|
| DC1 | DC2 | (DC1 + DC2)/2 | | | | | | |
| 0 LSB | 0 LSB | 0 LSB ⁽¹⁾ | | | | | | |
| 0 LSB | 1 LSB | ½ LSB ⁽²⁾ | | | | | | |
| 1 LSB | 0 LSB | ½ LSB ⁽²⁾ | | | | | | |
| 1 LSB | 1 LSB | 1 LSB ⁽¹⁾ | | | | | | |

Table 3. Possible patterns for 1-bit PWM dithering

- 1. Not of interest because achievable by hardware.
- 2. PWM dithering effect: for 1-bit PWM dithering, the fine tuning step is ½ LSB.

The first and the last patterns give a fine tuning of one full LSB. This is not of interest because this is already achievable by hardware. The second and the third patterns are the ones targeted by the PWM dithering technique. These patterns make possible to control the PWM resolution by a smaller step ($\frac{1}{2}$ LSB). For 1-added-resolution-bit PWM dithering, the two dithering-effect patterns are similar. Any one of them can be used for achieving the $\frac{1}{2}$ LSB fine-tuning step.

1.2.3 N-added-resolution-bit PWM dithering

For larger numbers of added resolution bits, the same concept as described above remains applicable. For N-added-resolution-bit PWM dithering, 2^N patterns are possible. As it is the case for the 1-added-resolution-bit PWM dithering, the first and the last patterns (those which have a fine tuning step of 1 LSB) are not of interest. All the remaining patterns present the dithering effect. When choosing which pattern to use, two considerations must be taken into account:

- The first one, which is intuitive, is the desired dither value.
- The second one is to choose the pattern that generates as low as possible ripple while keeping the required dither effect.

Let's consider a 3-added-resolution-bit PWM dithering example. In this case, each PWM dithering pattern is composed of 2^3 = 8 PWM periods and the fine tuning step is 1/8 LSB. Let's assume the desired average dither effect should be 1/2 LSB. *Table 4* and *Table 5* provide such pattern examples producing a ½ LSB fine-tuning. The patterns in *Table 4* are characterized by a high ripple magnitude whereas the pattern in *Table 5* is showing a minimal output ripple.



Table 4. PWM dithering pattern for generating 1/2 LSB dither effect (highest ripple magnitude)

| Average value | DC1 | DC2 | DC3 | DC4 | DC5 | DC6 | DC7 | DC8 |
|---------------|-----|-----|-----|-----|-----|-----|-----|-----|
| 4/8 = ½ LSB | 1 | 1 | 1 | 1 | 0 | 0 | 0 | 0 |
| 4/8 = ½ LSB | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 |

Table 5. PWM dithering pattern for generating 1/2 LSB dither effect (lowest ripple magnitude)

| Average value | DC1 | DC2 | DC3 | DC4 | DC5 | DC6 | DC7 | DC8 |
|---------------|-----|-----|-----|-----|-----|-----|-----|-----|
| 4/8 = ½ LSB | 1 | 0 | 1 | 0 | 1 | 0 | 1 | 0 |

Carefully selecting the PWM dithering pattern helps reducing the output ripple induced by the PWM dithering technique. Nevertheless, a slight degradation of the output ripple margin is unavoidable. *Table* 6 lists the PWM dithering patterns that produce the lowest output ripple for each average value achievable with a 3-added-resolution-bit PWM dithering implementation.

Table 6. PWM dithering patterns, 3-added-resolution-bit PWM (lowest output ripple)

| Average value | DC1 | DC2 | DC3 | DC4 | DC5 | DC6 | DC7 | DC8 | Ripple |
|---------------|-----|-----|-----|-----|-----|-----|-----|-----|----------|
| 0 LSB | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | None |
| 1/8 LSB | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | Highest |
| 2/8 LSB | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 1 | |
| 3/8 LSB | 0 | 0 | 0 | 1 | 0 | 1 | 0 | 1 | \ |
| 4/8 LSB | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 | Lowest |
| 5/8 LSB | 0 | 1 | 0 | 1 | 1 | 0 | 1 | 1 | ^ |
| 6/8 LSB | 0 | 1 | 1 | 1 | 0 | 1 | 1 | 1 |] |
| 7/8 LSB | 0 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | Highest |
| 1 LSB | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | None |

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2 PWM dithering technique implementation for STM32 Timers

2.1 STM32 timers features for PWM dithering implementation

The PWM dithering technique can be applied to any of the STM32 advanced configuration, general purpose and lite timers, regardless of their resolution: 16-bit or 32-bit timers.

The STM32 timers present a set of common features that make the dithering technique implementation relatively easy and with a small impact on the overall system performance. Below is a brief presentation of the STM32 timer features that are needed for the dithering technique implementation. For further details on these features or any other STM32 timer features, refer to the corresponding microcontroller reference manual.

2.1.1 STM32 timer time-base

The time-base unit is a basic component of the STM32 timers. It is included in every STM32 timer. The time-base unit contains the counter register, the auto-reload register and several other components. Depending on the configured counting direction, the counter is either incrementing from zero to the auto-reload register value or is decrementing from the auto-reload register value to zero. The time-base unit is used to define the timer counting periodicity.

2.1.2 STM32 timer channels

STM32 timers implement different number of channels ranging from 0 to 6. The number of channels depends on the type of timer and the product into which it has been embedded. The timers with zero channel (only a time-base unit) are called basic timers. These timers are not concerned by this application note because at least one channel (one timer output) is required by the PWM dithering technique.

STM32 timer channels can be used as input channels or as output channels. When configured in output mode, STM32 channels can be configured to output PWM signals. The PWM output configuration is the one used by the dithering technique.

When configured in PWM mode, the duty-cycle of the output waveform is controlled by the counter register value and the channel register value. The timer is continuously carrying on a comparison between the counter value and the timer channel value. Depending on the comparison result and the configured PWM mode (PWM1 or PWM2 mode), the channel output is either high or low. The resulting signal, as shown in *Figure 2*, is a PWM signal with a period duration controlled by the auto-reload register value and a duty-cycle controlled by the channel register value.



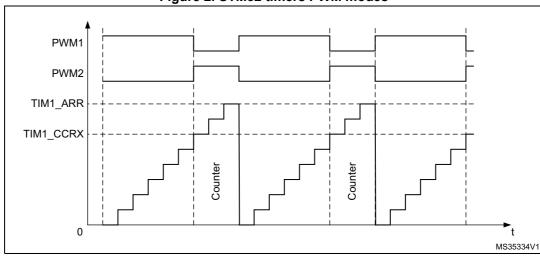


Figure 2. STM32 timers PWM modes

2.1.3 Channel register preload feature

As the channel output level depends on the continuous comparison between the counter register value and the timer channel register value, any change to the channel register may induce immediate channel output level change. As a consequence, writing directly to the channel register in the middle of a PWM period may generate spurious waveforms. To overcome this problem, the STM32 timer channels embed two channel registers: an active channel register and a shadow channel register. These two channel registers are controlled by the preload feature. When preload feature is disabled, any write operation is made directly into the active channel register used for carrying the comparison operation. As explained above, this configuration may generates spurious waveforms if the write access is not well synchronized with the counter value (this is practically difficult to achieve). When preload is activated, any write access is made to the shadow channel register while the channel active register remains unchanged (no perturbation on the ongoing PWM period). As soon as the timer counter reaches its counting boundary an update event is generated by the timer and the shadow register content is transferred into the active register which will be used during the next PWM period for carrying the comparison operation and define the duty-cycle ratio.

In summary, the preload feature guaranties that no spurious waveforms are generated while accessing the channel register. The channel register write access can be performed anytime within a PWM period.

For further details regarding the preload feature, refer to the corresponding microcontroller reference manual.

2.1.4 Timer DMA burst feature

In order to apply dithering effect on several timer outputs simultaneously (it is of high interest when designing multi-phase power converters), it is required to update several channel registers using the same DMA channel (on some products this is also called stream). The DMA burst feature built into the STM32 timers makes possible to update the content of several timer registers following a single event generated by the timer. In particular, this makes possible the update of several channel registers simultaneously, following the timer update event.

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2.2 System-level features and peripherals used in PWM dithering implementation

PWM dithering technique consists in refreshing the duty-cycle value for each PWM period. In other terms, it consists in refreshing the channel register value on each PWM period.

To offload the CPU, the STM32 DMA peripheral can be used to perform the timer channel register refresh thanks to the features described in Section 2.2.1 and Section 2.2.2.

2.2.1 Timer update-event-triggered DMA transfer

The STM32 timers feature the capability to trigger DMA transfers following the occurrence of particular timer events. For PWM dithering implementation, the timer update event can be used to start a DMA transfer. The timer update event is a periodic event that occurs each time the timer counter reaches its counting boundary. This event is used to trigger the channel register refresh at the start of each new PWM period.

2.2.2 DMA half-transfer interrupt

As explained above, in order to generate the PWM dithering effect, the duty-cycle should be updated at each PWM period for a given range of periods and following a given pattern. The pattern is a series of successive values that will be transferred one by one into the channel register at each new PWM period. For a given dither value, a dedicated pattern should be calculated. As a consequence, when a pattern is being transferred, the next one should be calculated. One way to do this is by splitting the pattern table into two halves. When DMA is transferring values from one part of the table, the firmware can calculate the other part of the table and prepare the pattern for the next PWM period series (the duty-cycle adjustment pattern) corresponding to the desired dither value.

DMA half-transfer interrupt and transfer-complete interrupt can be used for implementing this mechanism. When the half-transfer interrupt occurs this means that the DMA has completed the transfer of the values from the first half of the table and is going to transfer values corresponding to the next series of PWM periods from the second half of the table. When transfer-complete interrupt occurs this means that DMA has completed the transfer of values from the second part of the table and is going to switch back to the first half of the table for transferring pattern values for the next series of PWM periods.

2.3 PWM dithering implementation using STM32 MCUs

This section explains how to assemble the above described features in one application in order to enhance the resolution of one timer channel configured in PWM mode.

First of all, the programmer should reserve the needed space in the SRAM for the table of patterns. The size of the table of patterns can be calculated using the below formula:

Table_size =
$$2 \times 2^N$$



where N is the PWM dithering effect added resolution bits. When using a 16-bit timer, the table of patterns should be a table of half-words, otherwise if using a 32-bit timer, the table of patterns should be a table of words.

The firmware routines that initialize the peripherals needed for PWM dithering should respect the following recommendations.

- DMA should be configured to transfer values from the table of patterns located in SRAM memory into the timer channel register. The DMA source address increment should be configured in circular mode in order to roll over when it reaches the end of the table of patterns and points again on the table top. The DMA destination address should be fixed. The DMA half-transfer interrupt and transfer complete interrupt should be enabled for the relevant channel (on some products it is called stream). The firmware example provided with this application note presents a proposal on how to configure the DMA for implementing the PWM dithering technique. For further details about DMA modes and configurations, refer to the microcontroller corresponding reference manual.
- The timer should have its time-base initialized with appropriate values in order to generate the required PWM frequency. The "preload" function should be activated for the timer channel register. The timer should be configured to send DMA requests following each update event.
- The dithering patterns should be calculated based on the desired duty-cycle. If the timer possible hardware resolution for duty-cycle is x-bit length and the desired dithereffect added resolution is y-bit length, the duty-cycle passed to the dithering routines should be (x+y)-bit length. The DMA half-transfer and transfer-complete interrupts can be used as triggers for dithering patterns calculation routines. These DMA interrupts are useful to know which part of the table of patterns is used by DMA and which one is free and can be updated by the dithering patterns generation routines. The demonstration firmware provided with this application note shows an example of possible implementation of dithering patterns generation routines. This example, while being functional, might be further enhanced in order to improve performance (e.g. reduce footprint, reduce CPU load, etc.). Figure 3 illustrates how to assemble the STM32 MCU features together in order to implement the required timer PWM output resolution.

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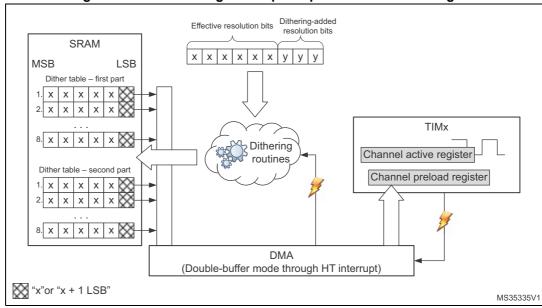


Figure 3. PWM dithering technique implementation bloc diagram



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3 Demonstration

This section details a 3-added-resolution-bit PWM dithering implementation. It provides insights about the firmware architecture used for the implementation. The demonstration example is given for 3 added resolution bits, but can be easily extended for different number of added resolution bits. The provided demonstration firmware is targeting the STM32F3 series of products. It can be easily migrated to any other product in the STM32 microcontrollers family.

The demonstration firmware generates a triangular waveform. The idea is to demonstrate the PWM dithering effect on the stairs-like rising slop of the generated waveform while keeping the falling one as it is. The lower the PWM resolution, the higher the rising stair-like steps. Applying the dithering technique on the PWM signal used to generate the triangular waveform adds a certain number of extra resolution bits. If properly implemented, the PWM dithering technique should make the output steps smaller (dithering effect magnitude depends on the number of added resolution bits). The dithering effect should be observed at the output of the external low pass filter (see Section 3.3.2: Low-pass filter dimensioning considerations), rather than directly at the output of the timer.

3.1 Demonstration hardware environment

The demonstration firmware provided with this application note is intended to be run on STM32F3 Discovery board. The STM32F3DISCOVERY discovery kit user manual and related documentation are available on STMicroelectronics website. Before it is used for the demonstration, the Discovery board needs to be slightly reworked to mount the low-pass filter on the timer output used for the demonstration.



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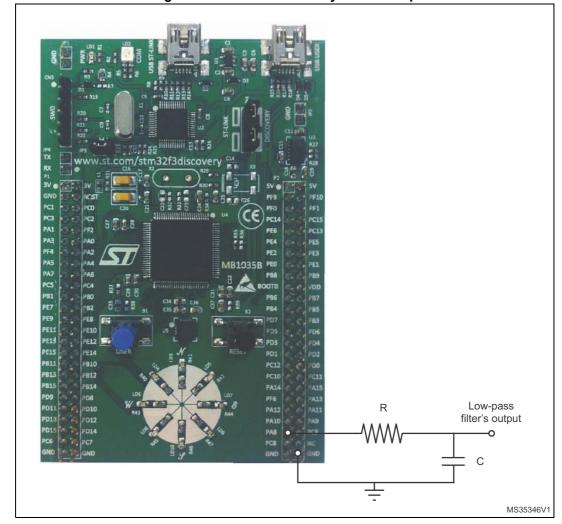


Figure 4. STM32F3 Discovery board setup

The low-pass filter cut-off frequency should be calculated based on several parameters including the PWM period. For this demonstration example the cut-off frequency is around 160 kHz (R = 1 kOhm, C = 1 nF). Our PWM switching frequency is around 1.125 MHz which is definitely in the rejected bandwidth of the low-pass filter. The resistor and the capacitor components should be mounted as close as possible to the timer output pin on the Discovery board. If the low-pass filter is placed on a daughter board, the connections between the daughter board and the Discovery board should be as short as possible in order to avoid distortion of the filter parameters and the resulting waveform (e.g. switching noise may become more important due to the inductance of the connection wires).

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3.2 Demonstration firmware architecture

3.2.1 Utilization of the microcontroller hardware resources

This demonstration firmware example is developed for the STM32F303VCT6 microcontroller present on the STM32F3Discovery board. The list of hardware resources used for the PWM dithering implementation are listed below:

- DMA1 channel 5 is used for timer channel register refresh
- Timer1 channel1 is used for the PWM resolution enhancement through dithering technique
- Timer1 channel1 output is configured to be outputted on the pin PA.09 that belongs to the GPIO port A.

3.2.2 Demonstration firmware description through flowcharts

After the CPU has completed the system clock configuration, it branches to the application main routine. The application main function is described by the flowchart in *Figure 5*:

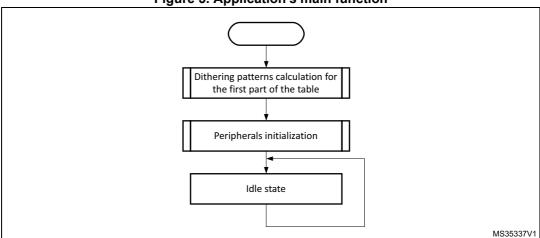


Figure 5. Application's main function

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Before initializing the timer and DMA peripherals and start refreshing the timer channel register, the first half of the table of patterns should be initialized. The following table updates are handled by the DMA interrupt handler. The peripherals initialization function is described in the flowchart below:

RCC configuration (enables DMA1, GPIOA and TIM1 clocks)

GPIOA port configuration (PA.09 is used as TIM1 channel 1 output)

DMA configuration + NVIC configuration

TIM1 configuration

Figure 6. Peripherals initialization flowchart

After completing the peripherals initialization the CPU goes into an infinite loop. Dithering patterns calculation task is handled by the DMA interrupt service routine (ISR) following each DMA half-transfer or transfer-complete interrupt. The flowchart in *Figure 7* describes the DMA interrupt handler function (ISR):

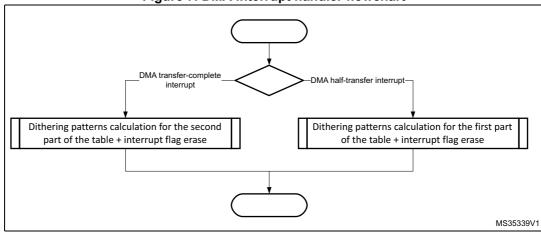


Figure 7. DMA interrupt handler flowchart

3.2.3 Firmware Footprint

The firmware routines and variables declaration associated with the PWM dithering technique implementation are located in the C source code files named "main.c" and "stm32f30x_it.c". The footprint of these two files (obtained with IAR EWARM v7.70

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toolchain) is shown in Table 7.

Table 7. Firmware Footprint

| File Name | Read-only code memory | , I | | | |
|----------------|--------------------------|----------|----------|--|--|
| main.c | 416 bytes | 44 bytes | 44 bytes | | |
| stm32f30x_it.c | 140 bytes | 0 bytes | 0 bytes | | |

3.3 Experimental results

3.3.1 Dithering effect on PWM resolution enhancement

The *Figure 8* shows the dithering effect on the rising slope of the triangular waveform. The rising slope looks more like a continuous oblique line where the falling slope still have the stairs-like shape. The falling slope has its duty-cycle controlled using a 6-bit resolution granularity with the timer clocked at 72 MHz and the PWM frequency set at 1.125 MHz (see *Table 2*, for details). When the PWM dithering is applied on the rising slope, the duty-cycle control resolution is extended by 3 bits to reach 9-bit resolution. The output control granularity is reduced by a ratio of $2^3 = 8$ and the output step becomes smaller (almost invisible on *Figure 8*).

Figure 8. Dithering effect applied on the rising slope of the triangular waveform



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3.3.2 Low-pass filter dimensioning considerations

For the example shown in *Figure 9*, the low-pass filter cut-off frequency has been deliberately altered (cut-off frequency too close to the PWM frequency divided by $2^3 = 8$). The switching added by the dithering-effect is not well averaged and the expected fine control may not be obtained. Nevertheless, *Figure 9* is interesting to better understand the effects of the dithering technique on the raw (unfiltered) PWM output signal.



Figure 9. Low-pass filter cut-off frequency effect on the dithering technique efficiency

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4 References

1. "Quantization Resolution and Limit Cycling in Digitally Controlled PWM Converters" by Angel V. Peterchev, Student Member, IEEE, and Seth R. Sanders, Member, IEEE.

- 2. STM32F30xx and STM32F31xx datasheets available on www.st.com.
- 3. STM32F30xx and STM32F31xx Reference Manual (RM0316) available on www.st.com.



AN4507 Revision history

5 Revision history

Table 8. Document revision history

| Date | Revision | Changes |
|-------------|----------|------------------|
| 25-Jun-2014 | 1 | Initial release. |

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