[2 points]. Prove that the eigenvalues of a real symmetric matrix $A \in \mathbb{R}^{N \times N}$, where $A = A^T$, are real-valued (*i.e.*, not complex-valued). Prove also that the eigenvectors of A corresponding to different eigenvalues are orthogonal to each other.

First part:

Let $\lambda \in \mathcal{C}$ be the eigenvalue of A => $Av = \lambda v$, v is the eigenvector,

$$(Av)^{T} = (\lambda v)^{T}$$

$$v^{T}A^{T} = \bar{\lambda}v^{T}$$

$$v^{T}A = \bar{\lambda}v^{T}$$

$$v^{T}Av = \bar{\lambda}v^{T}v$$

$$v^{T}(\lambda v) = \bar{\lambda}v^{T}v$$

$$\lambda v^{T}v = \bar{\lambda}v^{T}v$$

$$\lambda = \bar{\lambda}$$

As the conjugate of λ is same as λ , the eigenvalues are real valued.

Second part:

Let λ and μ be different eigenvalues of A and v and w are the corresponding eigenvectors

$$\lambda < v, w > = < \lambda v, w >$$
 $= < Av, w >$
 $= < V, A^{T}w >$
 $= < v, Aw >$
 $= < v, \mu w >$
 $= \mu < v, w >$
 $= > (\lambda - \mu) < v, w > = 0$

as λ and μ are different eigenvalues, two eigenvectors v and w are orthogonal.

[5 points]. Derive the K-th largest direction of variance in principal component analysis (PCA).

[5 points]. 1) Suppose that a discrete-time linear system has outputs y[n] for the given inputs x[n], as shown in Fig. 1. Determine the response $y_4[n]$ when the input is as shown in Fig. 2.

- a) [1 point]. Express $x_4[n]$ as a linear combination of $x_1[n]$, $x_2[n]$, and $x_3[n]$.
- b) [1 point]. Using the fact that the system is linear, determine $y_4[n]$, the response to $x_4[n]$.
- c) [1 point]. From the input-output pairs in Fig. 1, determine whether the system is time-invariant.
- 2) Determine the discrete-time convolution of x[n] and h[n] for the following two cases.
- a) [1 point]. As shown in Fig. 3.
- b) [1 point]. As shown in Fig. 4.

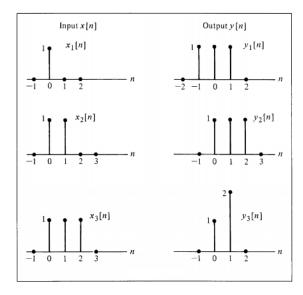


Figure 1:

Figure 2:

a)
$$x_4(n) = 2x_1(n) - 2x_2(n) + x_3(n)$$

b) As the system is linear ,
$$y_4(n)=y_1(n)-2y_2(n)+y_3(n)$$
 $y_4(n)=2\{0\ 1\ 1\ 1\ 0\ 0\}-2\{0\ 0\ 1\ 1\ 1\ 0\}+\{0\ 0\ 1\ 2\ 0\ 0\}=\{0\ 2\ 1\ 2\ -2\ 0\}$

c) The system is not time invariant, as we get $y_2[n] = x_2[n] + \delta(n+2)$, if we input n as delayed input n_0 , the output will not be the same.

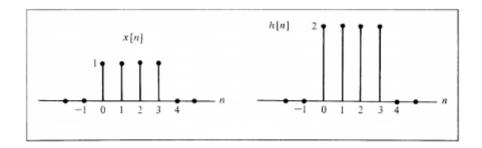


Figure 3:

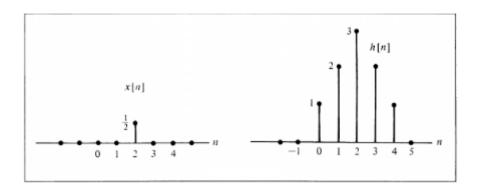


Figure 4:

2)

a)

h\x	0	1	1	1	1	0
0	0	0	0	0	0	0
2	0	2	2	2	2	0
2	0	2	2	2	2	0
2	0	2	2	2	2	0
2	0	2	2	2	2	0
0	0	0	0	0	0	0

Summing up diagonally, we get $y[n] = \{0 \ 0 \ 2 \ 4 \ 6 \ 8 \ 6 \ 4 \ 2 \ 0 \ 0\}$

b)

h\x	0	0	0.5	0	0
1	0	0	0.5	0	0
2	0	0	1	0	0
3	0	0	1.5	0	0
2	0	0	1	0	0
1	0	0	0.5	0	0

Summing up diagonally, we get $y[n] = \{0 \ 0 \ 0.5 \ 1 \ 1.5 \ 1 \ 0.5 \ 0 \ 0\}$

[3 points]. From the lecture slides, we know that the convolution of discrete-time signals x[n] and y[n], for $n \in [-\infty, +\infty]$, is defined as

$$z[n] = (x * y)[n] = \sum_{m = -\infty}^{\infty} x[m]y[n - m] = \sum_{m = -\infty}^{\infty} x[n - m]y[m].$$
 (1)

Here we introduce the Discrete-time Fourier Transform (DTFT) of a discrete-time signal x[n]:

$$X(\omega) = \text{DTFT}(x) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}.$$
 (2)

Try to prove the Convolution Theorem:

$$Z(\omega) = \text{DTFT}(x * y) = \text{DTFT}(x) \cdot \text{DTFT}(y) = X(\omega) \cdot Y(\omega). \tag{3}$$

$$\begin{split} &Z(w) = DTFT(x \cdot y) \\ &= \sum_{n = -\infty}^{\infty} (x \cdot y)[n]e^{j\Omega n} \\ &= \sum_{n = -\infty}^{\infty} \sum_{k = -\infty}^{\infty} x[k]y[n-k]e^{j\Omega n} \\ &= \sum_{k = -\infty}^{\infty} x[k] \sum_{n = -\infty}^{\infty} y[n-k]e^{j\Omega n} \\ &= \lim_{k = -\infty} x[k] \sum_{n = -\infty}^{\infty} y[n-k]e^{j\Omega n} \\ &= \lim_{k = -\infty} x[k] \sum_{m = -\infty}^{\infty} y[m]e^{-j\Omega (m+k)} \\ &= \sum_{k = -\infty}^{\infty} x[k] \sum_{m = -\infty}^{\infty} y[m]e^{-j\Omega m}e^{-j\Omega k} \\ &= \sum_{k = -\infty}^{\infty} x[k]e^{-j\Omega k} \sum_{m = -\infty}^{\infty} y[m]e^{-j\Omega m} \\ &= X(w) \cdot Y(w) \end{split}$$