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In it, I give references and citations whenever I refer to or use the published, or unpublished, work of others. I am aware that this course is bound by penalties as set out in the College examination offenses policy.

Signed: _____

RTDSP Lab 3

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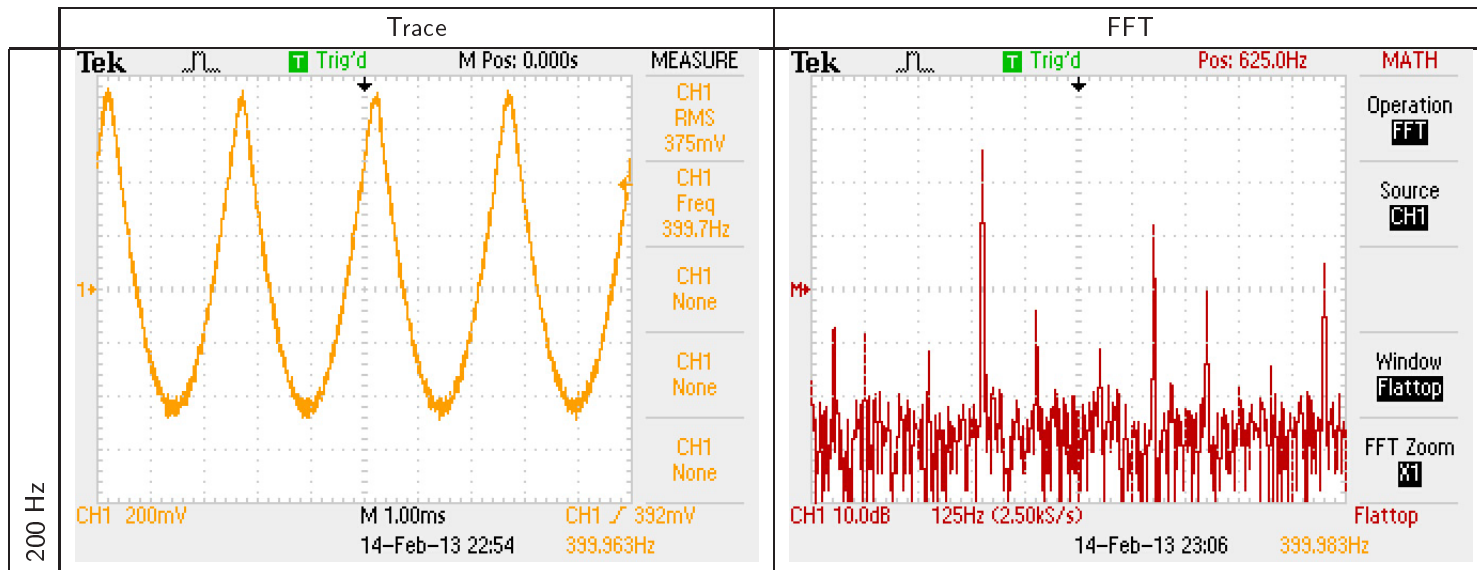
1 Exercise 1 - Input Sine Wave Rectification

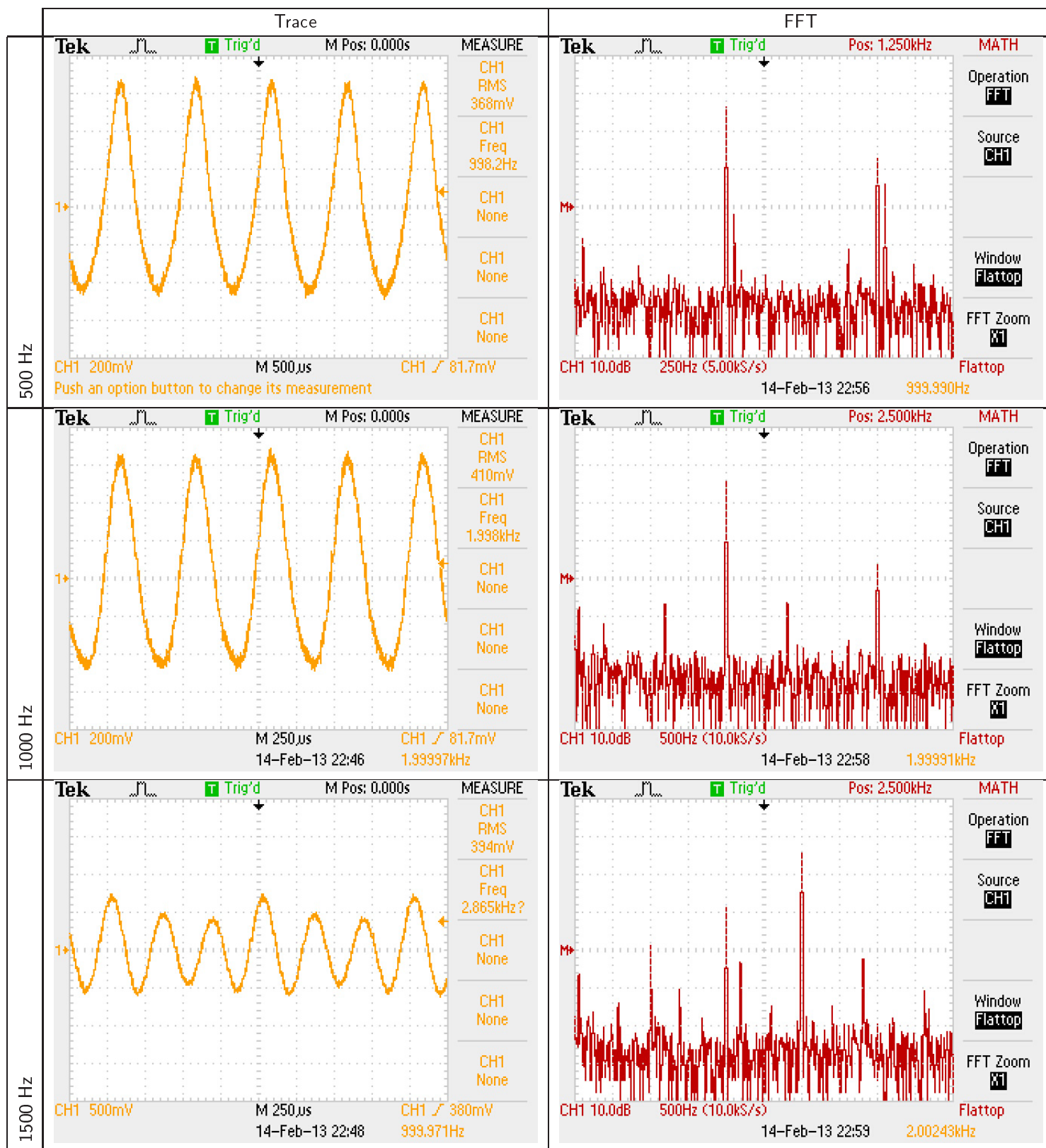
1.1 Reason for 0 V Centre

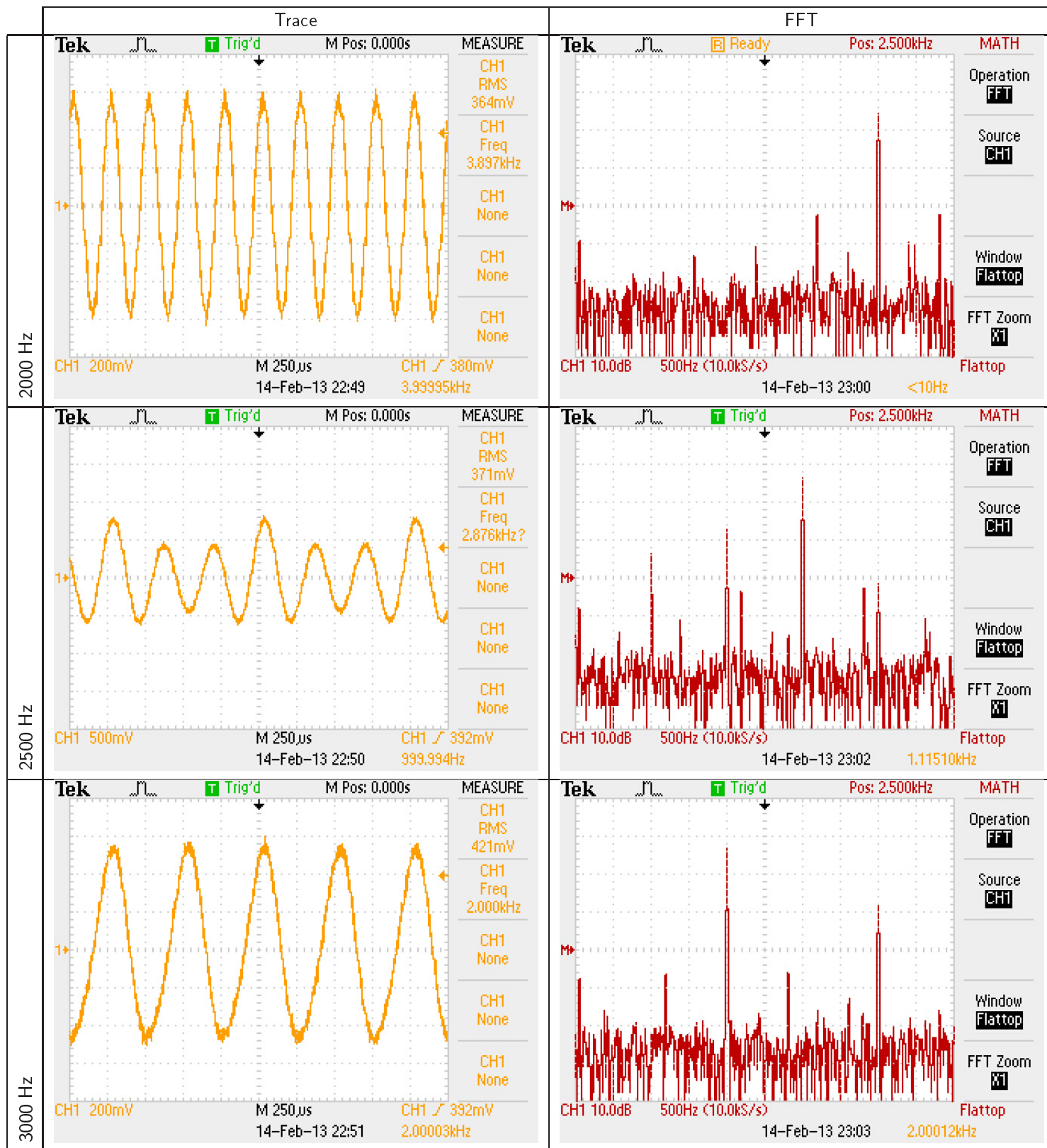
The capacitor at the line out port blocks the DC component of the output signal. This is because the impedance of the capacitor is given by $\frac{1}{j\omega C}$. When the frequency $\omega = 0$, the capacitor has infinite impedance. Thus, it acts to block the DC component of the output signal.

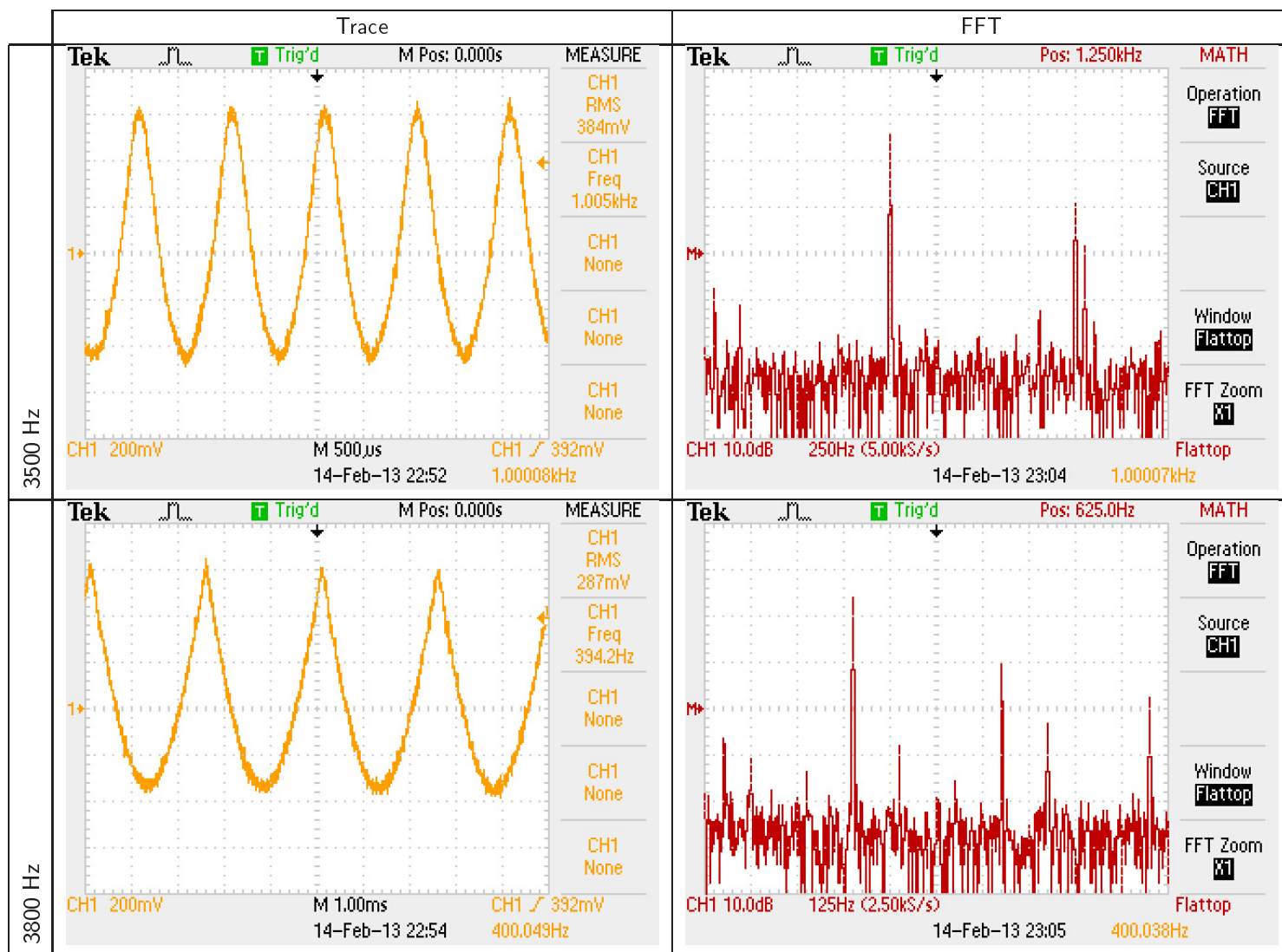
1.2 Oscilloscope Traces

The traces for the various input frequencies and their corresponding FFT is given in the table below. This table will be used to answer the second question regarding the output frequency.









1.3 On Frequency

When a signal is input into the system, the output effectively has twice the frequency of the input due to the rectification done on the input signal. This is because when a sine wave is rectified, its original frequency component is removed, and a DC component along with decreasing frequency components that are even multiples (i.e. two times, four times, etc.) of the original frequency component (see figure 1). This is observable from the FFT traces in in section 1.2. Thus, subject to the Nyquist Sampling Frequency limitation, the output cannot be more than 4 KHz (for a sampling frequency of 8KHz), and this translates to having an input frequency that, theoretically, can have a maximum of 2 KHz (which will give an output of 4 KHz) before aliasing kicks in.

When the input frequency goes beyond 2 KHz, aliasing takes place and folding can be observed in the output. This is evident if one compares the trace and FFT output (in section 1.2) of input frequencies of 200 Hz and 3800 Hz. It can be seen that these two input frequencies both give an output frequency of 400 Hz and have the same peak at 400 Hz in their FFT. This is due to folding being observed as a result in aliasing. Folding is the “wrapping” around of the frequency output due to aliasing. When a signal is sampled, copies of its frequency spectrum is made in the frequency domain. If the signal has frequency content higher than half that of the sampling frequency, these copies of the frequency content overlaps. The overlapped frequency content will cause an apparently lower frequency signal to be produced. This is the folding effect of

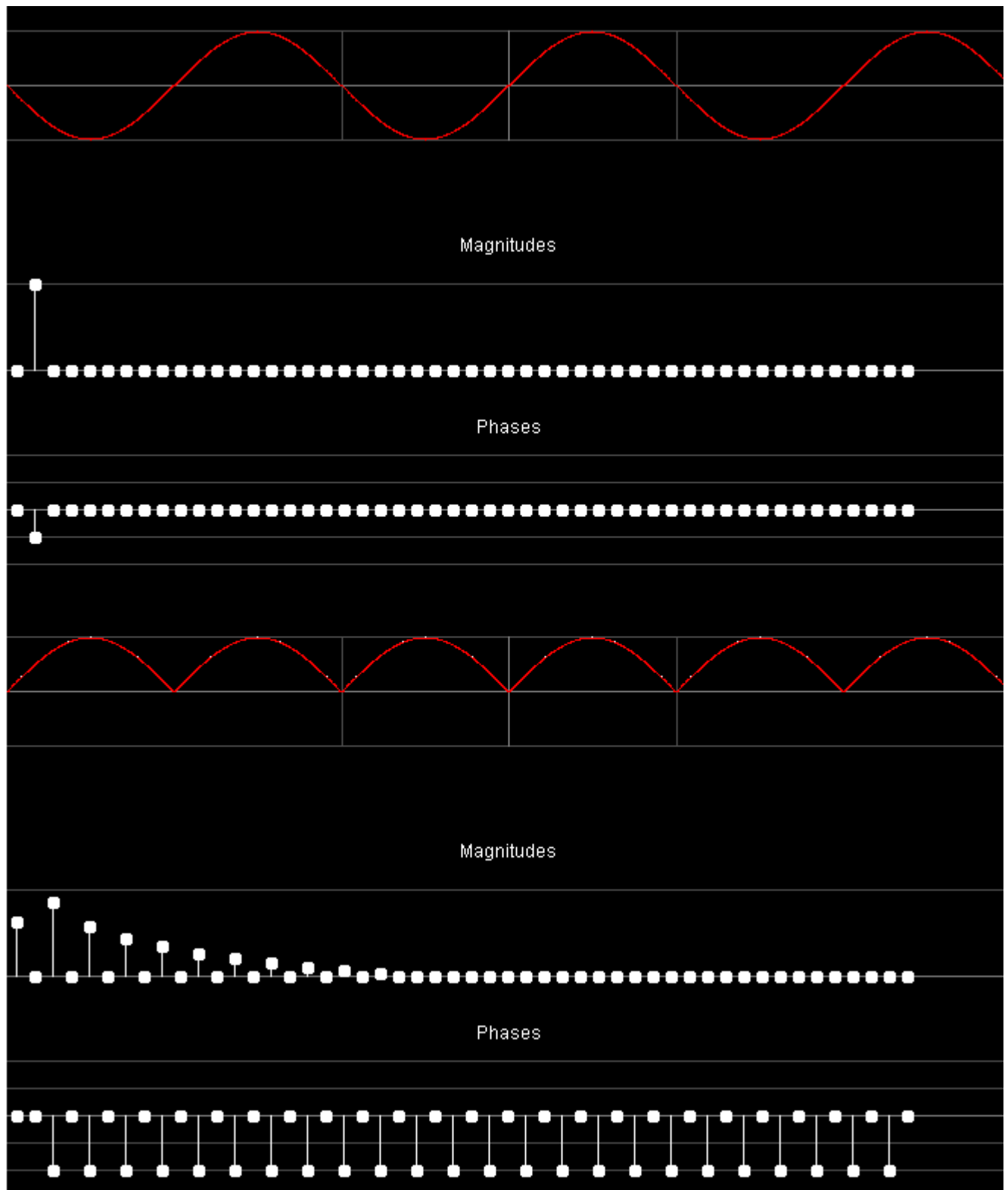


Figure 1: The Fourier transform of a sine wave (above) and its fully rectified version (below). Taken from an applet at <http://www.falstad.com/fourier/e-fullrect.html>

aliasing. Let f_a be the apparent frequency of the sampled frequency, f be the (supposed) frequency of the sinusoid, and f_s be the sampling frequency. Then the folding effect is given by

$$|f_a| = |f - nf_s| \quad (1)$$

where $|f_a| \leq \frac{f_s}{2}$ and $n \in \mathbb{Z}$.

It can be observed that when the input frequency is at 3800 Hz, the output frequency, f , should be at 7600 Hz. Then according to the equation (1), $n = 1$ and $f_a = 400$, which is what can be observed.

A similar observation can be seen in the output traces for inputs of 2.5 KHz, 3 KHz, and 3.5 KHz. They give rise to apparent frequencies of 3 KHz, 2 KHz, and 1 KHz respectively. According to equation (1), they would all have $n = 1$.

Amplitude modulation can also be observed. This is due to the rectification of the sine wave, giving rise to additional harmonics, as described in the first paragraph. Take, for example, the FFT of the input frequency of 1.5 KHz. There are peaks at 1 KHz, due to the wrap around effect of the duplicated even harmonics that is supposed to be at 6 KHz (as described by equation (1)), and another at 2 KHz due to “a copy” made by the sampling process. Also, from the FFT traces, it can be seen that some of frequency components above 4 KHz are not properly filtered out, due to the imperfect anti-aliasing filter. This also contributes to the distortion of the output signal.

1.4 Code Operation

Rectification is done every time a sample is sent to the codec. The function to service the interrupt is given below:

```

1 void ISR_AIC(void){
2     int sample = mono_read_16Bit(); // read
3     sample = abs(sample); // rectify
4     mono_write_16Bit((Int16) sample); // write
5 }
```

The code first reads the sample from the input, calculates its absolute value and then write it out to the output port. Note the cast to Int16 as the output port is operating with 16 bits of data as per the configuration.

2 Exercise 2 - Interrupt generated sine wave

2.1 Code Operation

The code to generate the sine wave is similar to the one used in Lab 2. Firstly, a lookup table is prepared during the initialisation stages to prepare 256 values for one quarter of a sine wave. The table is prepared using the following function (which is the same as Lab 2):

```

1 void sine_init(void){
2     int i;
3     for (i = 0; i < SINE_TABLE_SIZE; ++i)
4         table[i] = sin(i * ( (2*PI) / (RES_MULTIPLIER*SINE_TABLE_SIZE) ));
5 }
```

The macro RES_MULTIPLIER is set to a value of '4'. The service interrupt routine as described in section 1.4 is then replaced with essentially the same function as in Lab 2.

The routine, by keeping track of the “phase” of the current wave, will output the value to allow for the generation of the sine wave in the appropriate frequency. It keeps track of the “phase” of the wave using a static variable `index`. The static variables `prev_freq` and `prev_sample` are used to check for changes in those settings. If either are changed, the `index` is reset to zero, and the values described below will be recalculated.

The first value determines the number of samples necessary to generate the entire wave using the following line:

```
1 | double cycleSampleCount = (double) sampling_freq / (double) sine_freq;
```

The number of entries in the sine table to skip each time a sample is required is then calculated using the line, along with an increment of the `index`:

```
1 | step = (double) (SINE_TABLE_SIZE*RES_MULTIPPLIER)/(double) cycleSampleCount;
2 | // ... other code ...
3 | index += step;
```

To ensure that we do not exceed the number of entries in the table and cause a segmentation fault, the following line will reset the `index` with the necessary offset for the next cycle to ensure a smoother wave and allows the generation of “odd” frequencies:

```
1 | // Check that index is in range
2 | if ((int) index >= SINE_TABLE_SIZE*RES_MULTIPPLIER)
3 |     index -= (SINE_TABLE_SIZE*RES_MULTIPPLIER);
```

The table is then retrieved from the table, multiplied by an appropriate gain and sent to the output.

```
1 | sample = sine_value((int) round(index)); // get value from table
2 | output = (Int16) fabs(sample*gain); // output saved to a variable for ease of debugging
3 | mono_write_16Bit(output); // write to port
```

The `sine_value` function works by first determining the quadrant of a sine wave in which the `index` we want to retrieve is at, and also calculates the “progress” in that specific quadrant:

```
1 | int quadrant = index/SINE_TABLE_SIZE; // the quadrant in which the cycle is in
2 | int modulo = index % SINE_TABLE_SIZE; // the modulo
```

Then, according to the quadrant the `index` is in, the appropriate value is read from the table array and adjusted accordingly.

```
1 | if (quadrant == 0)
2 |     value = table[index];
3 | else if (quadrant == 1)
4 |     value = table[SINE_TABLE_SIZE-modulo-1];
5 | else if (quadrant == 2)
6 |     value = table[modulo]*-1;
7 | else if (quadrant == 3)
8 |     value = table[SINE_TABLE_SIZE-modulo-1]*-1;
9 | else
10 |     value = 0;
```


2.2 Oscilloscope Traces

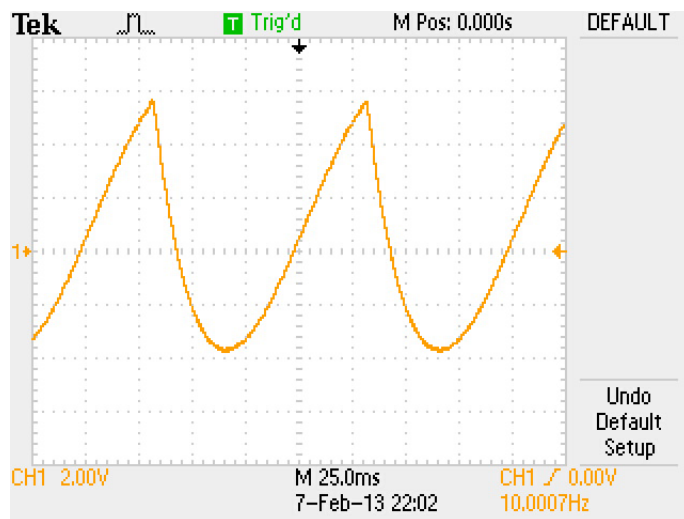


Figure 2: 5 Hz input sine wave rectified to 10 Hz

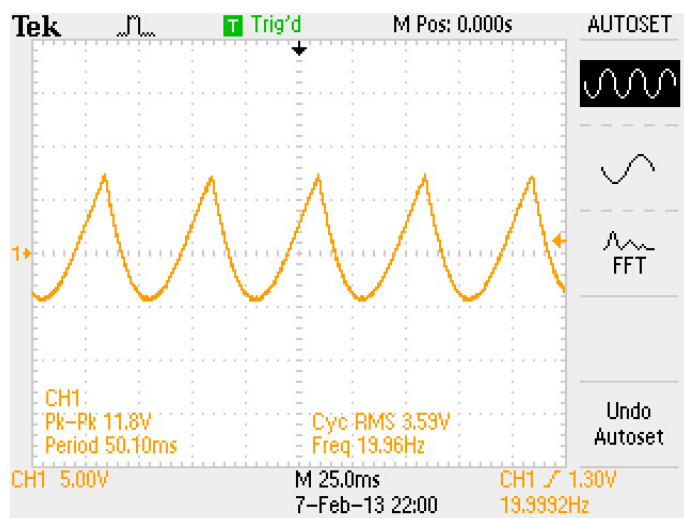


Figure 3: 10 Hz input sine wave rectified to 20 Hz.

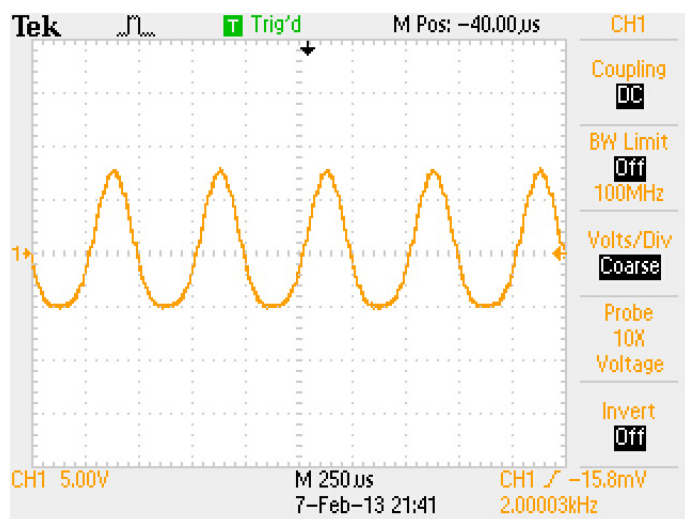


Figure 4: 1 KHz sine wave rectified to 2 KHz

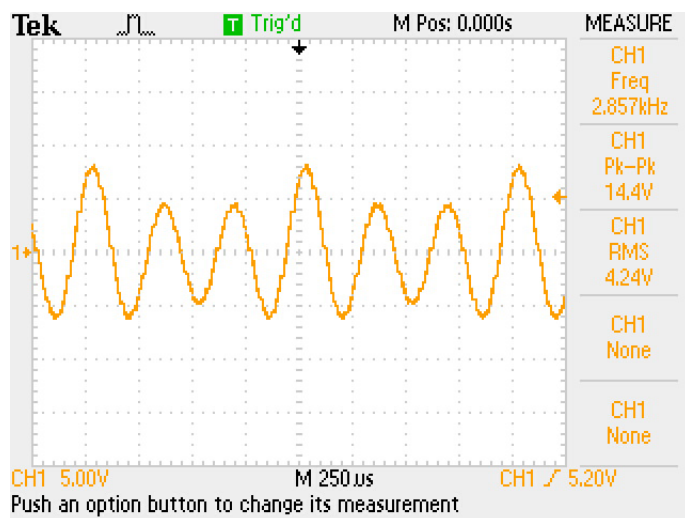


Figure 5: 1.5 KHz input rectified to 3 KHz. Notice the varying amplitude due to amplitude modulating effects

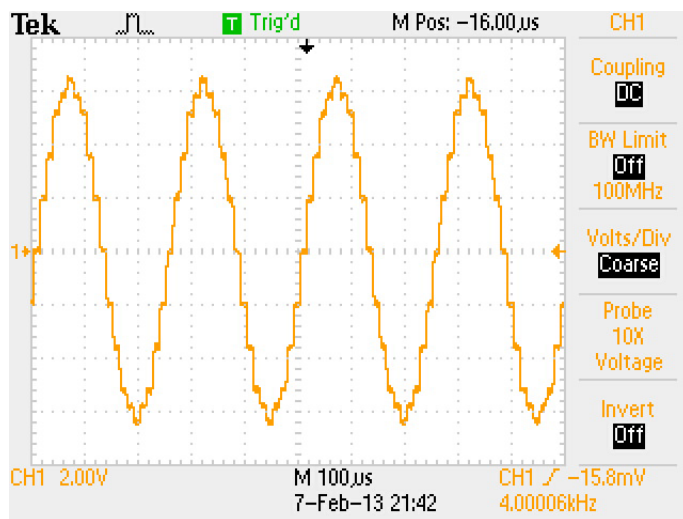


Figure 6: 2 KHz input rectified to 4 KHz. The Nyquist Frequency

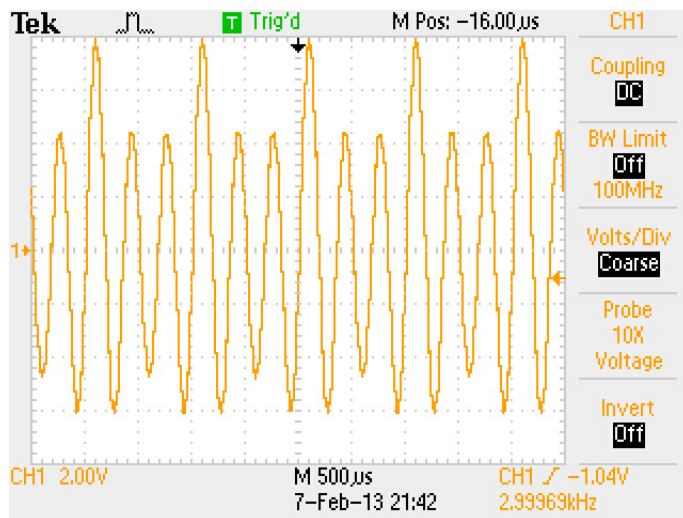


Figure 7: 2.5 KHz input that is supposed to give a 5 KHz. Due to aliasing giving rise to folding, the output is at 3 KHz.

3 Code Listing

3.1 Exercise 1

```
1  /*****
2      DEPARTMENT OF ELECTRICAL AND ELECTRONIC ENGINEERING
3      IMPERIAL COLLEGE LONDON
4
5      EE 3.19: Real Time Digital Signal Processing
6      Dr Paul Mitcheson and Daniel Harvey
7
8      LAB 3: Interrupt I/O Exercise 1
9      *****/
10
11 /***** Pre-processor statements *****/
12
13 #include <stdlib.h>
14 #include <stdio.h>
15 // Included so program can make use of DSP/BIOS configuration tool.
16 #include "dsp_bios_cfg.h"
17
18 /* The file dsk6713.h must be included in every program that uses the BSL. This
19    example also includes dsk6713_aic23.h because it uses the
20    AIC23 codec module (audio interface). */
21 #include "dsk6713.h"
22 #include "dsk6713_aic23.h"
23
24 // math library (trig functions)
25 #include <math.h>
26
27 // Some functions to help with writing/reading the audio ports when using interrupts.
28 #include <helper_functions_ISR.h>
29
30 /***** Global declarations *****/
31
32 /* Audio port configuration settings: these values set registers in the AIC23 audio
33    interface to configure it. See TI doc SLWS106D 3-3 to 3-10 for more info. */
34 DSK6713_AIC23_Config Config = { \
35     /*****
36     /* REGISTER          FUNCTION          SETTINGS          */
37     /***** */\
38     0x0017, /* 0 LEFTINVOL   Left line input channel volume  0dB          */\
39     0x0017, /* 1 RIGHTINVOL  Right line input channel volume  0dB          */\
40     0x01f9, /* 2 LEFTHPVOL   Left channel headphone volume   0dB          */\
41     0x01f9, /* 3 RIGHTHPVOL  Right channel headphone volume   0dB          */\
42     0x0011, /* 4 ANAPATH     Analog audio path control        DAC on, Mic boost 20dB */\
43     0x0000, /* 5 DIGPATH     Digital audio path control        All Filters off      */\
44     0x0000, /* 6 DPOWERDOWN  Power down control                All Hardware on       */\
45     0x0043, /* 7 DIGIF       Digital audio interface format    16 bit              */\
46     0x008d, /* 8 SAMPLERATE  Sample rate control                8 KHZ                */\
47     0x0001, /* 9 DIGACT      Digital interface activation       On                   */\
48     /*****
49     */\
50 };
51
52 // Codec handle:— a variable used to identify audio interface
```

```

53 DSK6713_AIC23_CodecHandle H_Codec;
54
55 /***** Function prototypes *****/
56 void init_hardware(void);
57 void init_HWI(void);
58 void ISR_AIC(void);
59 /***** Main routine *****/
60 void main(){
61
62
63     // initialize board and the audio port
64     init_hardware();
65
66     /* initialize hardware interrupts */
67     init_HWI();
68
69     /* loop indefinitely , waiting for interrupts */
70     while(1)
71     {
72
73     }
74
75 /***** init_hardware() *****/
76 void init_hardware()
77 {
78     // Initialize the board support library , must be called first
79     DSK6713_init();
80
81     // Start the AIC23 codec using the settings defined above in config
82     H_Codec = DSK6713_AIC23_openCodec(0, &Config);
83
84     /* Function below sets the number of bits in word used by MSBSP (serial port) for
85     receives from AIC23 (audio port). We are using a 32 bit packet containing two
86     16 bit numbers hence 32BIT is set for receive */
87     MCBSP_FSETS(RCR1, RWDLEN1, 32BIT);
88
89     /* Configures interrupt to activate on each consecutive available 32 bits
90     from Audio port hence an interrupt is generated for each L & R sample pair */
91     MCBSP_FSETS(SPCR1, RINTM, FRM);
92
93     /* These commands do the same thing as above but applied to data transfers to
94     the audio port */
95     MCBSP_FSETS(XCR1, XWDLEN1, 32BIT);
96     MCBSP_FSETS(SPCR1, XINTM, FRM);
97
98
99 }
100
101 /***** init_HWI() *****/
102 void init_HWI(void)
103 {
104     IRQ_globalDisable(); // Globally disables interrupts
105     IRQ_nmiEnable(); // Enables the NMI interrupt (used by the debugger)
106     IRQ_map(IRQ_EVT_RINT1, 4); // Maps an event to a physical interrupt
107     IRQ_enable(IRQ_EVT_RINT1); // Enables the event
108     IRQ_globalEnable(); // Globally enables interrupts
109

```

```

110 }
111
112 /***** WRITE YOUR INTERRUPT SERVICE ROUTINE HERE*****/
113
114 void ISR_AIC(void){
115     int sample = mono_read_16Bit(); // read
116     sample = abs(sample); // rectify
117     mono_write_16Bit((Int16) sample); // write
118 }

```

3.2 Exercise 2

```

1  /*****
2      DEPARTMENT OF ELECTRICAL AND ELECTRONIC ENGINEERING
3      IMPERIAL COLLEGE LONDON
4
5      EE 3.19: Real Time Digital Signal Processing
6      Dr Paul Mitcheson and Daniel Harvey
7
8      LAB 3: Interrupt I/O Exercise 2
9      *****/
10 /***** Pre-processor statements *****/
11
12 #include <stdlib.h>
13 #include <stdio.h>
14 // Included so program can make use of DSP/BIOS configuration tool.
15 #include "dsp_bios_cfg.h"
16
17 /* The file dsk6713.h must be included in every program that uses the BSL. This
18    example also includes dsk6713_aic23.h because it uses the
19    AIC23 codec module (audio interface). */
20 #include "dsk6713.h"
21 #include "dsk6713_aic23.h"
22
23 // math library (trig functions)
24 #include <math.h>
25
26 // Some functions to help with writing/reading the audio ports when using interrupts.
27 #include <helper_functions_ISR.h>
28
29 // PI defined here for use in your code
30 #define PI 3.141592653589793
31
32 // The number of entries in the sine lookup table
33 #define SINE_TABLE_SIZE 256
34
35 // Multiplier depending on whether INCREASE_RES is set
36 #define RES_MULTIPLIER 4
37
38 /***** Global declarations *****/
39
40 /* Audio port configuration settings: these values set registers in the AIC23 audio
41    interface to configure it. See TI doc SLWS106D 3-3 to 3-10 for more info. */
42 DSK6713_AIC23_Config Config = { \
43     /*****
44     /* REGISTER          FUNCTION          SETTINGS          */
45     /*****

```

```

46     0x0017, /* 0 LEFTINVOL Left line input channel volume 0dB */\
47     0x0017, /* 1 RIGHTINVOL Right line input channel volume 0dB */\
48     0x01f9, /* 2 LEFTHPVOL Left channel headphone volume 0dB */\
49     0x01f9, /* 3 RIGHTHPVOL Right channel headphone volume 0dB */\
50     0x0011, /* 4 ANAPATH Analog audio path control DAC on, Mic boost 20dB */\
51     0x0000, /* 5 DIGPATH Digital audio path control All Filters off */\
52     0x0000, /* 6 DPOWERDOWN Power down control All Hardware on */\
53     0x0043, /* 7 DIGIF Digital audio interface format 16 bit */\
54     0x008d, /* 8 SAMPLERATE Sample rate control 8 KHZ */\
55     0x0001 /* 9 DIGACT Digital interface activation On */\
56     /****** */
57 };
58
59
60 // Codec handle:- a variable used to identify audio interface
61 DSK6713_AIC23_CodecHandle H_Codec;
62
63 /* Sampling frequency in HZ. Must only be set to 8000, 16000, 24000
64 32000, 44100 (CD standard), 48000 or 96000 */
65 int sampling_freq = 8000;
66
67 // Gain - Less than 16 bit
68 int16 gain = 30000;
69
70 /* Use this variable in your code to set the frequency of your sine wave
71 be carefull that you do not set it above the current nyquist frequency! */
72 float sine_freq = 1000.0;
73
74 /* The array to hold the values of the sine lookup table */
75 float table[SINE_TABLE_SIZE] = {0};
76
77 /****** Function prototypes *****/
78 void init_hardware(void);
79 void init_HWI(void);
80 void ISR_AIC(void);
81 void sine_init(void);
82 #ifdef INCREASE_RES
83     float sine_value(int);
84 #endif
85 /****** Main routine *****/
86 void main(){
87
88     // initialise sine table
89     sine_init();
90
91     // initialize board and the audio port
92     init_hardware();
93
94     /* initialize hardware interrupts */
95     init_HWI();
96
97     /* loop indefinitely, waiting for interrupts */
98     while(1)
99     {};
100
101 }
102

```

```

103 /***** init_hardware() *****/
104 void init_hardware()
105 {
106     // Initialize the board support library, must be called first
107     DSK6713_init();
108
109     // Start the AIC23 codec using the settings defined above in config
110     H_Codec = DSK6713_AIC23_openCodec(0, &Config);
111
112     /* Function below sets the number of bits in word used by MSBSP (serial port) for
113     receives from AIC23 (audio port). We are using a 32 bit packet containing two
114     16 bit numbers hence 32BIT is set for receive */
115     MCBSP_FSETS(RCR1, RWDLEN1, 32BIT);
116
117     /* Configures interrupt to activate on each consecutive available 32 bits
118     from Audio port hence an interrupt is generated for each L & R sample pair */
119     MCBSP_FSETS(SPCR1, RINTM, FRM);
120
121     /* These commands do the same thing as above but applied to data transfers to
122     the audio port */
123     MCBSP_FSETS(XCR1, XWDLEN1, 32BIT);
124     MCBSP_FSETS(SPCR1, XINTM, FRM);
125
126 }
127
128 /***** init_HWI() *****/
129 void init_HWI(void)
130 {
131     IRQ_globalDisable(); // Globally disables interrupts
132     IRQ_nmiEnable(); // Enables the NMI interrupt (used by the debugger)
133     IRQ_map(IRQ_EVT_XINT1, 4); // Maps an event to a physical interrupt
134     IRQ_enable(IRQ_EVT_XINT1); // Enables the event
135     IRQ_globalEnable(); // Globally enables interrupts
136
137 }
138
139 /***** Interrupt Service Routine - Generate Sine *****/
140 void ISR_AIC(void){ // code is generally similar to Lab 2
141     float sample; // the sample value
142     Int16 output; // output value
143     static double index = 0; // Store the "progress" of the sine wave generation
144     static double prev_freq = 0; // Previous frequency
145     static double prev_sample = 0; // Previous sampling frequency
146     // Based on sampling frequency and sine frequency, determine number of samples per cycle
147     static double cycleSampleCount;
148     // Determine the number of intervals to "skip" each time we proceed to the next stage of the
149     // sine wave generation
150     static double step;
151
152     if (prev_freq != sine_freq || prev_sample != sampling_freq){ // If frequency has changed,
153         // we should take note.
154         index = 0;
155         prev_freq = sine_freq;
156         prev_sample = sampling_freq;
157         cycleSampleCount = (double) sampling_freq / (double) sine_freq;

```



```

158         step = (double) (SINE_TABLE_SIZE*RES_MULTIPLIER)/(double) cycleSampleCount;
159     }
160     index += step; // "advance" to the next step
161
162     // Check that index is in range
163     if ((int) index >= SINE_TABLE_SIZE*RES_MULTIPLIER)
164         index -= (SINE_TABLE_SIZE*RES_MULTIPLIER); // Reset with an offset so that we can
165         try and generate more frequencies
166
167     sample = sine_value((int) round(index)); // get value from table
168     output = (Int16) fabs(sample*gain); // output saved to a variable for ease of debugging
169     mono_write_16Bit(output); // write to port
170 }
171 /***** sine_init() *****/
172 void sine_init(void){
173     int i;
174
175     for (i = 0; i < SINE_TABLE_SIZE; ++i)
176         table[i] = sin(i * (2*PI) / (RES_MULTIPLIER*SINE_TABLE_SIZE));
177 }
178 float sine_value(int index){
179     int quadrant = index/SINE_TABLE_SIZE; // the quadrant in which the cycle is in
180     int modulo = index % SINE_TABLE_SIZE; // the modulo
181     float value;
182     if (quadrant == 0)
183         value = table[index];
184     else if (quadrant == 1)
185         value = table[SINE_TABLE_SIZE-modulo-1];
186     else if (quadrant == 2)
187         value = table[modulo]*-1;
188     else if (quadrant == 3)
189         value = table[SINE_TABLE_SIZE-modulo-1]*-1;
190     else
191         value = 0;
192
193     return value;
194 }

```