```
Modelling CS u: control input, y: plant output State variable CS is in state variable form if
 State variable CS is in state variable form if  \begin{aligned} & x_1 = f_1(t,x_1,\dots,x_n,u),\dots,x_n = f_n(t,x_1,\dots,x_n,u) \\ & y = h(t,x_1,\dots,x_n,u) \text{ is a collection of } n \text{ 1st order ODEs.} \\ & \text{Time-Invariant (TI) CS is TI if } f_i(\cdot) \text{ does not depend on } t. \\ & \text{State space (SS) TI CS is in SS form if } x = f(x,u), y = h(x,u) \\ & \text{where } x(t) \in \mathbb{R}^n \text{ is called the state.} \\ & \text{Single-input-single-output (SISO) CS is SISO if } u(t), y(t) \in \mathbb{R}. \\ & \text{LTI CS in SS form is LTI if } \dot{x} = Ax + Bu, \ y = Cx + Du \\ & A \in \mathbb{R}^{n \times n}, \ B \in \mathbb{R}^{n \times m}, \ C \in \mathbb{R}^{p \times n}, \ D \in \mathbb{R}^{p \times m} \\ & \text{where } x(t) \in \mathbb{R}^n, \ u(t) \in \mathbb{R}^m, \ y(t) \in \mathbb{R}^p. \\ & \text{Input-Output (IO) LTI CS is in IO form if} \\ & \frac{d^ny}{dt^n} + a_{n-1} \frac{d^{n-1}y}{dt^{n-1}} + \cdot + a_1 \frac{dy}{dt} + a_0y = b_m \frac{d^m u}{dt^m} + \cdot + b_1 \frac{du}{dt} + b_0u \\ & \text{where } m \leq n \text{ (causality)} \end{aligned}
     IO to SS Model 1. Define x s.t. highest order derivative in \dot{x}
    2. Write x=Ax+Bu=f(x,u) by isolating for components of x 3. Write y=Cx+Du=h(x,u) by setting measurement output y to component of x
    g to component of x . Equilibria y_d (steady state) b/c if y(0)=y_d at t=0, then y(t)=y_d \ \forall t\geq 0.
  Equilibrium pair Consider the system x=f(x,u). The pair (\bar{x},\bar{u}) is an equilibrium pair if f(\bar{x},\bar{u})=0. Equilibrium point \bar{x} is an equilibrium point w/ control u=\bar{u}. If u=\bar{u} and x(0)=\bar{x} then x(t)=\bar{x} \forall t\geq 0 (i.e. a system that starts at equilibrium remains at equilibrium). Find Equilibrium Pair/Point 1. Set f(x,u)=0 2. Solve f(x,u)=0 to find (x,u)=(\bar{x},\bar{u}). 3. If specific u=\bar{u}, then find x=\bar{x} by solving f(x,\bar{u})=0.
  \begin{array}{l} \delta x = x - \bar{x}, \, \delta u = u - \bar{u}, \, \delta y = y - h(\bar{x}, \bar{u}) \, \delta \dot{x} = \dot{x} - f(\bar{x}, \bar{u}) \, w/\\ \delta \dot{x} = A \delta x + B \delta u, \, A = \frac{\partial f(\bar{x}, \bar{u})}{\partial x} \in \mathbb{R}^{n_1 \times n_1}, \, B = \frac{\partial f(\bar{x}, \bar{u})}{\partial u} \in \mathbb{R}^{n_1}, \\ \frac{\partial f(\bar{x}, \bar{u})}{\partial x} = \frac{\partial f(\bar{x}, \bar{u})}{\partial x} + \frac
  \delta y = C\delta x + D\delta u, \ C = \frac{\partial \underline{h}}{\partial \underline{x}}(\bar{x}, \bar{u}) \in \mathbb{R}^{1 \times n_1}, \ D = \frac{\partial h(\bar{x}, \bar{u})}{\partial u} \in \mathbb{R}
    *Only valid at equ. pairs.
                                               Linear Approx. Given a diff. fcn. f: \mathbb{R} \to \mathbb{R}, its linear approx at \bar{x} is f_{\lim} = f(\bar{x}) + f'(\bar{x})(x - \bar{x}).
     *Remainder Thm: f(x) = f_{\text{lin}} + r(x) where \lim_{x \to \bar{x}} \frac{r(x)}{x - \bar{x}} = 0.
       *Note: Can provide a good approx. near \bar{x} but not globally.
     *Gen. f: \mathbb{R}^{n_1} \to \mathbb{R}^{n_2}, f(x) = f(\bar{x}) + \frac{\partial f}{\partial x}(\bar{x})(x - \bar{x}) + R(x)
  *Jacobian: \frac{\partial f}{\partial x}(\bar{x}) = \left[\frac{\partial f}{\partial x_1}(\bar{x}) \cdots \frac{\partial f}{\partial x_{n_1}}(\bar{x})\right] \in \mathbb{R}^{n_2 \times n_1}
Linearization Steps I. Find equ. pair (\bar{x}, \bar{u})
2. Derive A, B, C, D and then evaluate at (\bar{x}, \bar{u})
  3. Write \dot{\delta x} = A\delta x + B\delta u and \delta y = C\delta x + D\delta u
    Laplace Transform Given a fcn f: \mathbb{R}_+ = [0, \infty) \rightarrow \mathbb{R}^n, its
 Laplace transform is F(s) = \mathcal{L}\{f(t)\}:=\int_{0^{-}}^{\infty} f(t)e^{-st} dt, s \in \mathbb{C}.
*$\mathcal{L}: f(t) \rightarrow F(s), t \in \mathbb{R}_+ \text{ (time dom.) & } s \in \mathbb{C} \text{ (Laplace dom.).}
    "L:f(t) \mapsto F(s), t \in \mathbb{R}_+ (time dom.), \alpha s \in \mathfrak{C} (Laplace dom.), \mathbb{R}_+ D.W. CTS: A fon f : \mathbb{R}_+ \to \mathbb{R}^n is \mathfrak{p}.\mathfrak{w}. cts if on every finite interval of \mathbb{R}, f(t) has at most a finite # of discontinuity points
    (t_i) and the limits \lim_{t\to t_i^+} f(t), \lim_{t\to t_i^-} f(t) are finite
    Exp. Order A function f: \mathbb{R}_+ \to \mathbb{R}^n is of exp. order if \exists
 constants K, \rho, T > 0 s.t. \|f(t)\| \le Ke^{\rho t}, \ \forall t \ge T. Existence of LT Thm If f(t) is p.w. cts and of exp. order w/ constants K, \rho, T > 0, then F(\cdot) exists and is defined \forall s \in D := \{s \in C : \operatorname{Re}(s) > \rho\} and F(\cdot) is analytic on D. *Analytic fon iff differentiable fcn. *D: Region of convergence (ROC), open half plane.
                                                                                                                                       D Re(6)
Unit Step 1(t) := \begin{cases} 1, & \text{if } t \geq 0 \\ 0, & \text{otherwise} \end{cases}
Table of Common Laplace Transforms: f(t) \mid F(s)
1(t) \mapsto \frac{1}{s} \quad t1(t) \mapsto \frac{1}{s^2} \quad t^k \ 1(t) \mapsto \frac{k!}{s^{k+1}} \quad e^{at} \ 1(t) \mapsto \frac{1}{s-a}
t^n e^{at} \ 1(t) \mapsto \frac{n!}{(s-a)^{n+1}} \quad \sin(at) \ 1(t) \mapsto \frac{a}{s^2+a^2}
\cos(at) \ 1(t) \mapsto \frac{s}{s^2+a^2}
Prop. of Laplace Transform Linearity: \mathcal{L}\{cf(t) + g(t)\} = c\mathcal{L}\{f(t)\} + \mathcal{L}\{g(t)\}, c \sim \text{constant}.
Differentiation: If the Laplace transform of f'(t) exists, then \mathcal{L}\{f'(t)\} = s\mathcal{L}\{f(t)\} - f(0^-).
     \mathcal{L}\{f'(t)\} = s\mathcal{L}\{f(t)\} - f(0^{-}).
 If the Laplace transform of f^{(n)}(t) := \frac{d^n f}{dt^n}(t) exists, then \mathcal{L}\{f^{(n)}(t)\} = s^n \mathcal{L}\{f(t)\} - \sum_{i=1}^n s^{n-i} f^{(i-1)}(0^-). Integration: \mathcal{L}\left\{\int_0^t f(\tau) d\tau\right\} = \frac{1}{s} \mathcal{L}\{f(t)\}.
  Convolution: Let (f*g)(t) := \int_t^t f(\tau)g(t-\tau)d\tau = \int_0^t f(t-\tau)g(\tau)d\tau, then \mathcal{L}\{(f*g)(t)\} := \mathcal{L}\{f(t)\}\mathcal{L}\{g(t)\}. Time Delay: \mathcal{L}\{f(t-T)1(t-T)\} = e^{-Ts}\mathcal{L}\{f(t)\}, T \geq 0. Multiplication by t: \mathcal{L}\{tf(t)\} := -\frac{d}{ds}[\mathcal{L}\{f(t)\}].
```

Trig. Id. $2\sin(2t)=2\sin(t)\cos(t)$, $\sin(a-b)=\sin(a)\cos(b)-\cos(a)\sin(b)$, $\cos(a-b)=\cos(a)\cos(b)+\sin(a)\sin(b)$ Complete the Square: $ax^2+bx+c=a(x+\frac{b}{2a})^2-\frac{b^2}{4a}+c$ LT Steps: 1. Write f(t) as a sum and use linearity *Trig. id. may be useful. 2. Use prop. of LT and common LT to find F(s)Inverse Laplace Transform Given F(s), its inverse LT is $f(t)=\mathcal{L}^{-1}\{F(s)\}:=\frac{1}{2\pi}\int_{c-j\infty}^{c+j\infty}F(s)\varepsilon^{st}ds$

Shift in s: $\mathcal{L}\lbrace e^{at}f(t)\rbrace = \mathcal{L}\lbrace f(t)\rbrace \Big|_{s\to s-a}^{-1} = F(s-a)$, where

 $F(s) = \mathcal{L}\{f(t)\} \& a \text{ const.}$

```
=\lim_{w\to\infty}\frac{1}{2\pi}\int_{c-jw}^{c+jw}F(s)e^{st}ds,\ c\in\mathbb{C}\ \text{is selected s.t. the line}\\ L:=\{s\in\mathbb{C}:s=c+j\omega,\omega\in\mathbb{R}\}\ \text{is inside the ROC of}\ F(s).\\ \textbf{Zero:}\ z\in\mathbb{C}\ \text{is a zero of}\ F(s)\ \text{if}\ F(z)=0.
  Pole: p \in \mathcal{C} is a pole of F(s) if \frac{f}{F(p)} = 0.

Cauchy's Residue THM If F(s) is analytic (complex diff.) everywhere except at isolated poles \{p_1, \dots, p_N\}, then \mathcal{L}^{-1}\{F(s)\} = \sum_{i=1}^{N} \operatorname{Res}\left[F(s)e^{st}, s = p_i\right] \mathbf{1}(t),
 \mathcal{L}^{-1}\{F(s)\} = \sum_{i=1}^{N} \operatorname{Res} \left[F(s)e^{st}, s = p_i\right] \mathbf{1}(t),
*Res[F(s)e^{st}, s = p_i]: Residue of F(s)e^{st} at s = p_i.
Residue Computation Let G(s) be a complex analytic fcn w/ a pole at s = p, r be the multiplicity of the pole p. Then \operatorname{Res}[G(s), s = p] = \frac{1}{(r-1)!} \lim_{s \to p} \frac{d^{r-1}}{ds^{r-1}} \left[G(s)(s-p)^r\right].
Inv. LT Partial Frac.: 1. Factorize F(s) into partial fractions. 2. Find coefficients and use LT table to find inverse LT.
**Complete the square.
Inv. LT Residue: 1. Find poles of F(s) and their residues. 2. Use Cauchy's Residue THM to find inverse LT.
**Transfer Function: Consider a CS in IO form. Assume zero initial conds. y(0) = \cdots = \frac{d(n-1)}{dt(n-1)}(0) = 0 and
  u(0) = \cdots = \frac{d^{(m-1)}u}{dt^{(m-1)}}(0) = 0. \text{ Then the TF from } u \text{ to } y \text{ is}
G(s) := \frac{y(s)}{U(s)} = \frac{b_m s^m + \cdots + b_0}{s^n + a_{n-1} s^{n-1} + \cdots + a_0}.
*0 Ini. Conds.: y_0(s) = G(s)u(s)
    *\emptyset Ini. Conds.: y_{\emptyset}(s) = y_0(s) + \frac{\text{poly. based on initial conds.}}{s^n + a_{n-1}s^{n-1} + \dots + a_0}
 TF Steps (IO to TF): 1. Given IO form of CS, assume zero initial conds.  
2. Find G(s) by taking LT of IO form and forming Y(s)/U(s).  
*Careful: Y(s)/U(s) = G(s) not U(s)/Y(s) = G(s).  
Impulse Response: Given CS modeled by TF G(s), its IR is g(t) := \mathcal{L}^{-1}\{G(s)\}.  
*\mathcal{L}\{\delta(t)\} = 1, then if u(t) = \delta(t), then Y(s) = U(s)G(s) = G(s).  
$S to TF: G(s) = C(sI - A)^{-1}B + D s.t. y(s) = G(s)U(s).  
*Assume x(0) = 0 \in \mathbb{R}^n (zero initial conds.).  
*LTI: G(s) of an LTI system is always a rational fcn.  
*Not Invertible: Values of s s.t. sI - A not invertible can correspond to poles of G(s).  
Inverse: 1. For A \in \mathbb{R}^{n \times n}, find [cof(A)]_{(i,j)} = (-1)^{i+j} \det(A_{(i,j)}).  
*A_{G(s,j)} : A w/ row i and col. j removed.
       TF Steps (IO to TF): 1. Given IO form of CS, assume zero
    *A_{(i,j)}: A w/ row i and col. j removed.
    2. Assemble cof(A) and find det(A) = \sum_{j=1}^{n} a_{ij} [cof(A)]_{(i,j)}
    w/ fixed i or \det(A) = \sum_{i=1}^{n} a_{ij} [\operatorname{cof}(A)]_{(i,j)} w/ fixed j
  3. Find A^{-1} = \frac{1}{\det(A)} \operatorname{adj}(A) = \frac{1}{\det(A)} [\operatorname{cof}(A)]^T.
   *2 × 2 : A^{-1} = \frac{1}{\det(A)} \begin{bmatrix} d & -b \\ -c & a \end{bmatrix}
    TF (SS to TF): 1. Given SS form, assume zero initial conds.
   2. Solve G(s) = C(sI - A)^{-1}B + D.
   *If C = \begin{bmatrix} 0 & 1_i & 0 \end{bmatrix} & B = \begin{bmatrix} 0 & 1_j & 0 \end{bmatrix}, then only need ith row
  & jth col. of \operatorname{adj}(sI-A) s.t. G(s) = \frac{\operatorname{Iadj}(sI-A)|_{(i,j)}}{\operatorname{det}(sI-A)} + D. WHAT TO DO IF MULTIPLE NON ZERO ENTRIES i and j? TF to SS: Consider G(s) = \frac{b_m s^m + \dots + b_0}{s^n + a_{n-1} s^{n-1} + \dots + a_0} = \frac{N(s)}{D(s)}
where m < n (i.e. G(s) is strictly proper). Then the SS form is  *A = \begin{bmatrix} 0 & 1 & 0 & \cdots & 0 \\ 0 & 0 & 1 & \cdots & 0 \\ \vdots & \vdots & \ddots & \ddots & \vdots \\ 0 & 0 & 0 & 0 & 1 \\ -a_0 & -a_1 & -a_2 & \cdots & -a_{n-1} \end{bmatrix}, B = \begin{bmatrix} 0 \\ \vdots \\ 0 \\ 1 \end{bmatrix}, 
 C = \begin{bmatrix} b_m & b_{m-1} & \cdots & b_1 & | & 0 & 0 & \cdots & 0 \\ 0 & 1 & 1 & 0 & 0 & \cdots & 0 \end{bmatrix}, D = 0. 
*Unique: State space of a TF is not unique. Summary:
    Block Diagram Types of Blocks:
   Cascade: y_2 = (G_1(s)G_2(s))U \stackrel{\text{SISO}}{=} y_2 = (G_2(s)G_1(s))U
                                        \cup \longrightarrow \overline{[\underline{\alpha}_1]} \xrightarrow{y_1} \overline{[\underline{\alpha}_2]} \longrightarrow y_2 \quad \equiv \quad \cup \longrightarrow \overline{[\underline{G}_1,\underline{G}_2]} \longrightarrow y_2
    Parallel y = (G_1(s) + G_2(s))U
  \begin{aligned} & \mathbf{Feedback} \ y = \left( \frac{G_1(s)}{1 + G_1(s)G_2(s)} \right) R \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & \\ & & 
  *SC: Unity Feedback Loop (UFL) if G_2(s)=1. 

Manipulations: 1. y=G(U_1-U_2)=GU_1+GU_2

2. y_1=GU y_2=U | y_1=GU y_2=G\frac{1}{G}U

3. From feedback loop to UFL.
              Find TF from Block Diagram: 1. Start from in → out, making
```

simplifications using block diagram rules.

2. Simplify until you get the form $U(s) \to G(s) \to Y(s)$. Time Response: Elementary Terms: $\mathbf{1}(t) \leftarrow \text{pole} \ @ \ 0$ th $\mathbf{1}(t) \leftarrow \text{pole} \ @ \ 0$ w/ mult. $n \mid e^{at} \mathbf{1}(t) \leftarrow \text{pole} \ @ \ a$ $\sin(\omega t + \phi) \mathbf{1}(t) \leftarrow \text{pole} \ @ \ t \neq j\omega \mid \cos(\omega t + \phi) \mathbf{1}(t) \leftarrow \text{pole} \ @ \ t \neq j\omega$