# ${\rm SSY130}$ - Hand-in 2 ${\rm FIR}$ Differentiator Design

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#### Question 1)

The code 1 above, shows a FIR filter that performs approximate differentiation for low frequencies up to frequency 0:05 Hz, while blocking the frequency band above 0.1 Hz to reduce the influence of the noise:

Listing 1: Low pass filter design

```
dt = 1;
                 % S
        1/dt;
                 % Hz
      = 0.05;
                 % Hz
        0.1;
                 % Hz
                   FIR filter order
                                                    % *pi [rad/sample]
H_f = [0, fcut, fstop, f_s/2] / (f_s/2);
                                                    % filter amplitude
             1,
                     0,
                             0] .*H_f*pi *f_s;
h_diff = firpm(N, H_f, H_a, 'differentiator');
```

As specified by the firpm Matlab command, the frequency vector  $\mathbf{H}_{-}\mathbf{f}$  is given as normalized frequency points, specified in the range between 0 and 1, where 1 corresponds to the Nyquist frequency  $f_s/2$ . Moreover, in order to perform properly the differentiation in the pass-band region, the amplitude vector  $\mathbf{H}_{-}\mathbf{a}$  must correspond to the value jw, where w is the frequency in rad/s. Therefore, the amplitude vector makes use of the frequency vector multiplied by  $\pi \cdot f_s$  which will result in the right unit. The ideal filter frequency response can be verified in figure 2.

### Question 2)

Figure 1 shows the designed filter impulse response and figure 2 the comparison between the desired ideal and the real FIR frequency response.

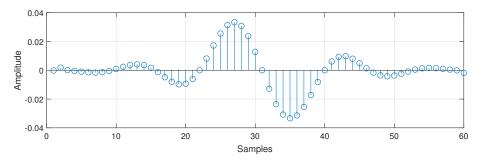


Figure 1: FIR differentiator filter impulse response

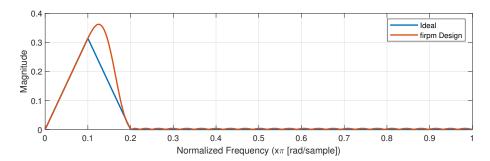


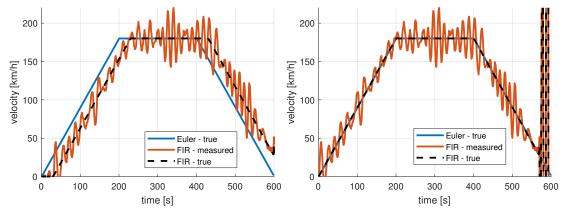
Figure 2: FIR differentiator filter frequency response

#### Question 3)

If the filter impulse response was truncated to M=61 samples, the algorithm must have introduced a delay of (M-1)/2=30 samples. In time this is equivalent to  $30 \cdot \Delta t = 30 \cdot 1s = 30s$ .

## Question 4)

Figure 3a shows the Euler and FIR filter for true and measured data. After offline delay compensation, the delay effect is totally removed, as shown in figure 3b:



(a) True and measured velocity using Euler (b) True and measured velocity using Euler and and FIR filter with off-line delay correction

At the end of the filter output very large oscillations occur due to the effect of convolution v(n) = h(n) \* s(n). If h has size N and s has size M, then the convolution will result in M+N-1 elements. It follows, that the last N-1 elements are calculated by zero-padding s(n) in N-1 elements. What happens is that the last element in s is very large and the zero-padding will artificially create a huge position displacement and consequently a huge velocity. That is why, we usually discard this part of the resulted convoluted signal.

#### Question 5)

Figure 4 reveals the result of applying a trivial Euler derivative filter in a signal corrupted by noise.

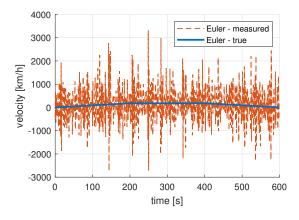


Figure 4: True and filtered measured velocity

Clearly, the obtained velocity turned to be dominated by the noise. This occurs because when taking the derivative of the signal, which is corrupted by the noise, higher frequencies will have its magnitudes largely amplified. Derivatives in time domain is equivalent to multiplying by jw in frequency domain. Therefore, noise components with high frequencies will be amplified and will dominate the final signal.

If we are able to filter the noise and then take the derivative, we decrease this noise amplification effect and the signal is cleaner. Indeed, we still see the noise in figure 3a and we could decrease the pass-band frequency to filter this visible high frequency noises. However, we assume the risk of reducing also the gain of the true signal, that are close to this frequency. Each differentiator filter must be then adjusted according to the desired frequency operation range.

## Question 6)

The maximum speed of the vehicle is 180 km/h.