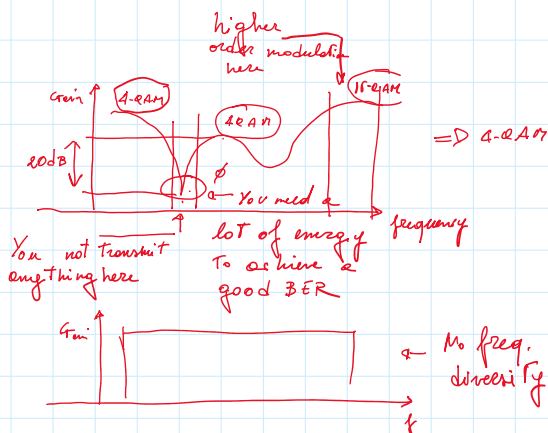
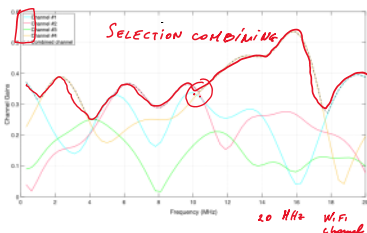
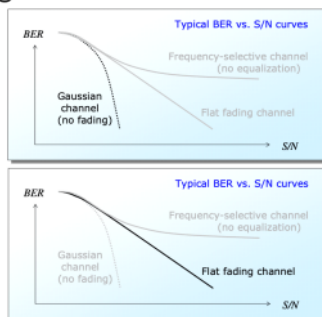
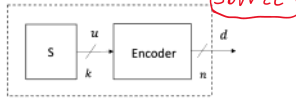


ELECTRONICS AND COMMUNICATIONS SYSTEMS
COMPUTER ENGINEERING



Time diversity: interleaving and coding

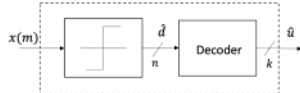
- The main idea was formalized by Claude Shannon in 1948.
- Channel coding introduces some redundancy in the transmitted bits to either
 - Detect errors at the receiver
 - Improve the bit error probability at the receiver.



- The redundancy is measured by the code rate $R = \frac{k}{n} < 1$, the ratio between k , the number of bits at the input of the encoder and n , the number of bits at the output of the encoder.

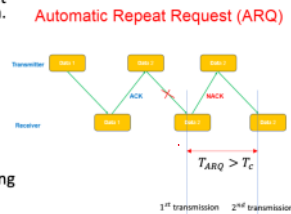
Error detection coding

- Very simple technique: one or more bits of parity are added at the end of a word.
- The receiver computes the parity check applying the same algorithm implemented at the receiver:
 - If the result computed at the receiver matches the parity check bits, the parity bits are discarded and the received bits are considered error-free.
 - If the result does not match the parity check bits, there is one or more errors in the string of received bits and the receiver requests a re-transmission.



Data retransmission

- The receiver feeds back an ACK for a correct reception and a NACK for a faulty reception.
 - After receiving a NACK, the transmitter resends the data packet
 - ARQ exploits the *time diversity* of the channel by retransmitting the data after T_{ARQ} , a time interval, longer than the channel coherence time T_c .
 - The new transmission will experience a different and hopefully better channel
 - Advanced receivers are capable of combining the two received messages to further improve the chances of a successful reception.
-
- Automatic Repeat Request (ARQ)**
- The diagram illustrates the ARQ process. A Transmitter (green) sends packets to a Receiver (blue). Packet 1 is received correctly, resulting in an ACK. Packet 2 is received with an error, resulting in a NACK. The transmitter then resends packet 2. A time interval T_{ARQ} is shown between the first and second transmissions of packet 2, which is longer than the channel coherence time T_c . The first transmission of packet 2 is labeled '1st transmission' and the second is labeled '2nd transmission'.



Error correction coding and channel capacity

- Channel codes can be employed to *correct* the errors introduced by the channel.
- Given a communication channel of bandwidth B , Shannon proved that the channel capacity can be computed as

$$C = B \log_2(1 + \text{SNR}) \text{ b/s}$$
 - For any transmission with rate $R < C$ and an arbitrarily small ϵ , it is possible to find an error correction code such that the error probability is $P_e < \epsilon$.
 - On the contrary, if $R > C$ it is not possible to find any code that can make the probability of error of the transmission over the channel arbitrarily small.

Error correction codes

bits 0,1

- Linear algebraic codes are the most used type of codes:
 - Block codes
 - Convolutional codes
- All algebraic operations are performed in the $GF(2)$, the Galois field of two elements $\{0,1\}$.

+	0	1
0	0	1
1	1	0

x	0	1
0	0	0
1	0	1

- They have been initially studied for AWGN transmissions but can be employed over fading channels.

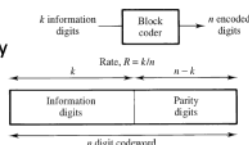
Block codes: encoder

- Block codes are most of the times in **systematic form**: the coded word is formed by k information bits and $n - k$ parity bits.
- The encoder can be represented as the code **generator matrix G** the encoder.
- The word u of k bits is encoded in the coded word d of n bits

$$d = uG$$

- The encoder add redundancy so that all coded words differ of as many bits as possible.

$$G = [I_k \mid P]$$



REPETITION CODES

$$0 \rightarrow 000$$

$$1 \rightarrow 111$$

$$\begin{matrix} 0 & 0 & 0 \\ 1 & 1 & 1 \end{matrix}$$

$$000 \rightarrow 0$$

$$010 \rightarrow 0$$

$$100 \rightarrow 0$$

$$110 \rightarrow 1$$

$$101 \rightarrow 1$$

$$011 \rightarrow 1$$

$$111 \rightarrow 1$$

MAJORITY
DECODING

Assume we
have received
 $\hat{x} = [010]$

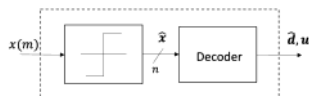
$$d(\hat{x}, 000) = 1$$

$$d(\hat{x}, 111) = 2$$

$$\begin{matrix} 1 & 1 & 1 \\ 1 & 1 & 1 \end{matrix} + \begin{matrix} 1 & 0 & 1 \\ 1 & 0 & 1 \end{matrix} = \begin{matrix} 0 & 1 & 0 \\ 0 & 1 & 0 \end{matrix}$$

Block codes: decoder

- After having received the string of n bits \hat{x} , the decoder selects the codeword \hat{d} , as the one that has minimal distance from \hat{x}
 $\hat{d} = \arg \min d(\hat{x}, d)$ **MAXIMUM LIKELIHOOD**
- The distance is computed as the number of bits that are different in the two string of bits.
- An error event occurs when, due to the noise, the received vector \hat{x} is closer to a codeword different from the transmitted one.
- Codes where the distance between words is large are more robust against noise and fading than codes where the distance is small.



block code with $K = 2^k$ $n = 30$ $R = 2/3$
what is the number of different codewords

$$K = 1, n = 3$$

$$\begin{matrix} 0 & 0 & 0 \\ 1 & 1 & 1 \end{matrix}$$

In total I have $2^1 = 2$

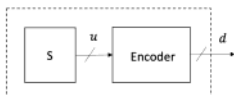
$$2^{20} = 2^{10 \cdot 2} \approx 1 \text{ Mio}$$

3 Mb/s \Rightarrow 1 codeword is 30 bits

received
100.000 of codewords per second
and for each codeword you need 1 Mio comparisons.

Convolutional codes: encoder

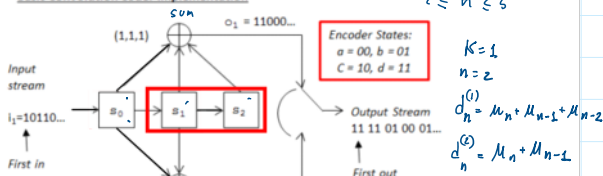
- The encoder of a k, n convolutional code works as n linear filters in $GF(2)$. Each of the n outputs of the encoder is a linear combination of the input
- The filter impulse response for the j -th output bit, the j -th code generator, is a series of 0 and 1.



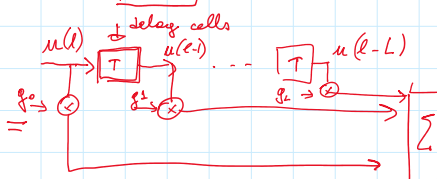
Convolutional codes: encoder

$K=1$

Basic Convolutional Code Implementation



$$x(t) \rightarrow h(t) \rightarrow y(t), y(t) = x(t) \otimes h(t)$$



We OR in $GF(2)$!

$$u(l) \xrightarrow{\text{delay cells}} u(l-1) \rightarrow u(l-L)$$

