

## **CG2023** Selected Tutorial Questions

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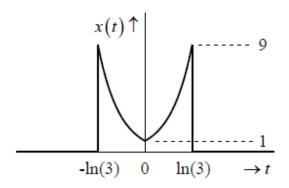
# Tutorial 1

### T1 Q4

- Q.4 A signal y(t) is formed by replicating another signal  $x(t) = e^{2|t|} \operatorname{rect}\left(\frac{t}{2\ln(3)}\right)$  at uniform intervals of 4 seconds.
  - (a) Sketch x(t).
  - (b) Find an expression for y(t) and sketch it.
  - (c) Find the total energy of x(t) and the average power of y(t).
  - (d) Using the results in part (c), determine the total energy of w(t) = 3x(-10t + 8) and average power of  $z(t) = \frac{1}{2}y(-\frac{t}{3} + 5)$ .

(a) 
$$x(t) = e^{2|t|} \operatorname{rect}\left(\frac{t}{2\ln(3)}\right)$$

$$T = 2 \ln(3)$$



• The Dirac-
$$\delta$$
 Comb function:

$$comb_{\lambda}(t) = \sum_{n} \delta(t - n\lambda)$$

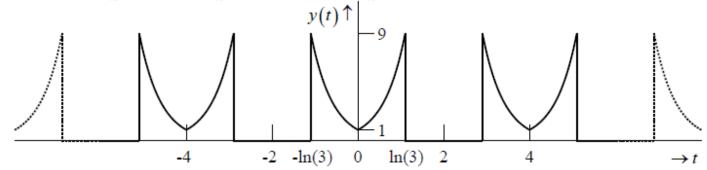
Model a periodic signal:

$$x_p(t) = x(t) * \sum_{n} \delta(t - n\lambda)$$

Model a sampled signal:

$$x_s(t) = x(t) \times \sum_{n} \delta(t - n\lambda)$$

(b) 
$$y(t) = x(t) * \sum_{n} \delta(t - 4n) = \sum_{n} x(t) * \delta(t - 4n) = \sum_{n} x(t - 4n)$$



(c) Total energy of x(t):

$$E_{x} = \int_{-\infty}^{\infty} |x(t)|^{2} dt = \int_{-\ln(3)}^{\ln(3)} e^{4|t|} dt = 2 \int_{0}^{\ln(3)} e^{4t} dt = 2 \cdot \frac{e^{4t}}{4} \Big|_{0}^{\ln(3)} = \frac{1}{2} \left(3^{4} - 1\right) = 40 \text{ Joules}$$

Average power of y(t):

$$P_{y} = \lim_{t \to \infty} \frac{1}{\tau} \int_{-\tau/2}^{\tau/2} |y(t)|^{2} dt = \frac{1}{4} \int_{-2}^{2} |y(t)|^{2} dt = \frac{1}{4} \int_{-\ln(3)}^{\ln(3)} e^{4|t|} dt = \frac{E_{x}}{4} = \frac{40}{4} = 10 \text{ Watts}$$
averaged over 1 period because  $x(t)$ 
is periodic
averaged over one

original definition of average power signals
$$\lim_{\tau \to \infty} \frac{1}{2\tau} \int_{-\tau}^{\tau} |x(t)|^2 dt = \frac{1}{T} \int_{0}^{T} |x(t)|^2 dt$$

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(d) w(t) = 3x(-10(t-0.8)) is equal to x(t) amplified by 3 times, time-compressed by 10 times, time-delayed by 0.8 seconds, and time-reversed. Therefore,

Total Energy 
$$E = \int_{-\infty}^{\infty} |x(t)|^2 dt$$
,  $0 < E < \infty$ 

$$E = \int_{-\infty}^{\infty} |w(t)|^2 dt = \int_{-\infty}^{\infty} |3x(-10(t-0.8))|^2 dt = 3^2 \int_{-\infty}^{\infty} |x(-10(t-0.8))|^2 dt$$

We define u=-10(t-0.8), and obtain  $\frac{du}{dt}=-10 \rightarrow dt=-\frac{1}{10}du$ . The limits of the integration will be  $\begin{cases} u \rightarrow -\infty, & \text{if } t \rightarrow \infty \\ u \rightarrow \infty, & \text{if } t \rightarrow -\infty \end{cases}$ 

$$E = 3^{2} \int_{-\infty}^{\infty} |x(u)|^{2} dt = 3^{2} \int_{\infty}^{-\infty} |x(u)|^{2} (\frac{-1}{10}) du = \frac{3^{2}}{10} \int_{\infty}^{-\infty} |x(u)|^{2} du$$

Total energy of 
$$w(t) = \frac{3^2}{10} \times \left[ \text{total energy of } x(t) \right] = \frac{9}{10} \times 40 = 36 \text{ Joules}$$

Time-shifting or time- reversal of an energy signal will not affect its total energy.

delayed by 15 seconds, and time-reversed. Therefore,

Assume the period of 
$$y(t)$$
 is  $T$ , and  $z(t)$  is equal to  $y(t)$ 

where a constraint of the period of  $y(t)$  is  $y(t)$  in  $y(t)$  and  $y(t)$  is  $y(t)$  in  $y(t)$  and  $y(t)$  is  $y(t)$  in  $y(t)$ 

 $z(t) = \frac{1}{2}y(-\frac{1}{3}(t-15))$  is equal to y(t) amplified by 0.5 time, time-expanded by 3 times, time-

Assume the period of y(t) is T, and z(t) is equal to y(t) $\lim_{\tau \to \infty} \frac{1}{2\tau} \int_{-\tau}^{\tau} \left| x(t) \right|^2 dt = \frac{1}{T} \int_{0}^{T} \left| x(t) \right|^2 dt$ time-expanded by 3 times, therefore the period of z(t) is 3T

time-expanded by 3 times, therefore the period of 
$$Z(t)$$
 is 31
$$P = \frac{1}{3T} \int_{-3T/2}^{3T/2} \left| \frac{1}{2} y \left( -\frac{1}{3} (t - 15) \right) \right|^2 dt = \frac{1}{2^2} \frac{1}{3T} \int_{-3T/2}^{3T/2} \left| y \left( -\frac{1}{3} (t - 15) \right) \right|^2 dt$$

**Average** 

We define 
$$u=-\frac{1}{3}(t-15)$$
, and obtain  $\frac{du}{dt}=-\frac{1}{3}\to dt=-3du$ .

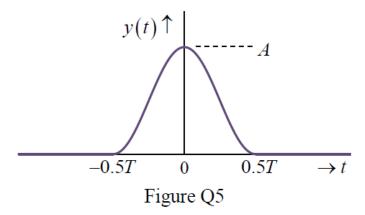
The limits of the integration will be 
$$\begin{cases} u = -\frac{T}{2} + 5, & \text{if } t = \frac{3T}{2} \\ u = \frac{T}{2} + 5, & \text{if } t = -\frac{3T}{2} \end{cases}$$
 Average power of  $x(t)$  
$$P = \frac{1}{2^2} \frac{1}{3T} \int_0^{3T} \left| y \left( -\frac{1}{3} (t - 15) \right) \right|^2 dt = \frac{1}{2^2} \frac{1}{3T} \int_{T+5}^{-\frac{T}{2} + 5} |y(u)|^2 (-3) du = \frac{1}{2^2} \frac{3}{3} \left| \frac{1}{T} \int_{-\frac{T}{2}}^{\frac{T}{2}} |y(u)|^2 du$$

Average power of  $z(t) = \frac{1}{2^2} \times \left[ \text{average power of } y(t) \right] = \frac{1}{4} \times 10 = \frac{5}{2} \text{ Watt}$ 

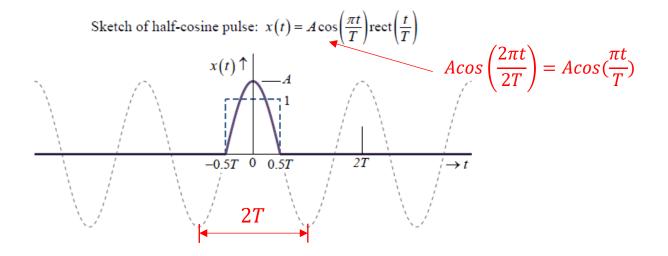
Time-scaling, time-shifting or time-reversal of a periodic signal will not affect its average power.

### T1 Q5

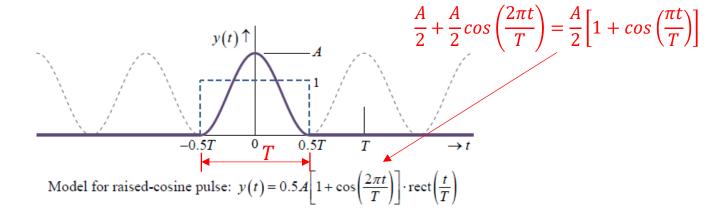
- Q.5 In digital communications, half-cosine or raised-cosine pulses are sometimes used to pulse shape a binary waveform so as to reduce intersymbol interference.
  - (a) Sketch and label the half-cosine pulse  $x(t) = A\cos\left(\frac{\pi t}{T}\right)\operatorname{rect}\left(\frac{t}{T}\right)$  where A and T are positive constants.
  - (b) Derive the mathematical model for the raised-cosine pulse, y(t), shown in Figure Q5.











# **Tutorial 2**

### T2 Q1

Q.1 The discrete-frequency spectrum of a signal x(t) is shown in Figure Q1.

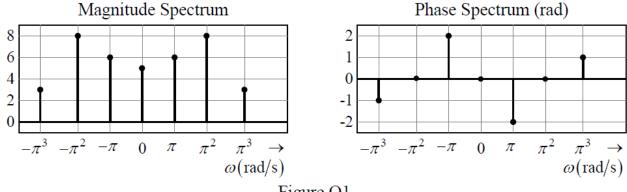


Figure Q1

- (a) Express x(t) as a sum of complex sinusoids and show whether or not x(t) is a real signal.
- (b) Find the DC value and average power of x(t)?
- (c) Do the frequencies of the sinusoidal components of x(t) have common factor(s). Hence, what conclusion can you draw about the Fourier series expansion of x(t)?

(a) By inspection of the discrete-frequency spectrum of x(t):

$$x(t) = 3e^{-j}e^{-j\pi^3t} + 8e^{-j\pi^2t} + 6e^{j2}e^{-j\pi t} + 5 + 6e^{-j2}e^{j\pi t} + 8e^{j\pi^2t} + 3e^{j}e^{j\pi^3t}$$
$$= 3e^{-j(\pi^3t+1)} + 8e^{-j\pi^2t} + 6e^{-j(\pi t-2)} + 5 + 6e^{j(\pi t-2)} + 8e^{j\pi^2t} + 3e^{j(\pi^3t+1)}$$

Combining complex conjugate terms, we get

$$x(t) = 5 + 12\cos(\pi t - 2) + 16\cos(\pi^2 t) + 6\cos(\pi^3 t + 1)$$

which shows that x(t) is a real signal.

$$Z(t) = a(t) + jb(t)$$

$$Z^*(t) = a(t) - jb(t)$$

$$Z(t) + Z^*(t) = 2a(t)$$
 A real signal

(b) Find the DC value and average power of x(t)?

$$x(t) = 5 + 12\cos(\pi t - 2) + 16\cos(\pi^2 t) + 6\cos(\pi^3 t + 1)$$

The **DC value (Direct Current value)** refers to the constant or average component of a signal over time. It represents the non-varying part of a signal, in contrast to the **AC (Alternating Current) component**, which fluctuates.

For periodic signals:

DC Value of 
$$x(t) = c_0 = \frac{1}{T_p} \int_0^{T_p} x(t) dt$$

For aperiodic signals, the DC value is given by the Fourier transform evaluated at f=0

DC Value of 
$$x(t) = X(f) \Big|_{f=0} = \int_{-\infty}^{\infty} x(t) e^{-j2\pi(0)t} dt = \int_{-\infty}^{\infty} x(t) dt$$

DC value of 
$$x(t) = \left[ \underbrace{\frac{5}{\text{DC value of 5}}}_{\text{DC value of }} + \underbrace{\frac{0}{\text{DC value of }}}_{\text{12}\cos(\pi t - 2)} + \underbrace{\frac{0}{\text{DC value of }}}_{\text{16}\cos(\pi^2 t)} + \underbrace{\frac{0}{\text{DC value of 6}}}_{\text{6}\cos(\pi^3 t + 1)} \right] = 5$$

(b) Find the DC value and average power of x(t)?

#### Method I

For a general sinusoid signal  $\alpha \cos\left(2\pi\frac{t}{\tau}\right)$ , the average power can be obtained as

$$P = \frac{1}{T} \int_0^T |x(t)|^2 dt = \frac{1}{T} \int_0^T \alpha^2 \cos^2 \left( 2\pi \frac{t}{T} \right) dt = \frac{\alpha^2}{T} \int_0^T \frac{1}{2} \left( 1 + \cos \left( 4\pi \frac{t}{T} \right) \right) dt$$
$$= \frac{\alpha^2}{2T} \left[ \int_0^T 1 \, dt + \int_0^T \cos \left( 4\pi \frac{t}{T} \right) dt \right] = \frac{\alpha^2}{2T} \times T = \frac{\alpha^2}{2}$$

Avg Pow of 
$$x(t) = \left[ \underbrace{\frac{5^2}{\text{Avg Pow of 5}}}_{\text{12}\cos(\pi t - 2)} + \underbrace{\frac{16^2/2}{\text{Avg Pow of 6}}}_{\text{16}\cos(\pi^2 t)} + \underbrace{\frac{6^2/2}{\text{Avg Pow of 6}}}_{\text{6}\cos(\pi^3 t + 1)} \right] = 243 \text{ W}$$

Power Spectral Density of  $x_p(t)$ :

$$P_x(f) = \sum_{k=-\infty}^{\infty} |c_k|^2 \delta(f - kf_p)$$

 $P_{x}(f) = \sum_{k=-\infty}^{\infty} |c_{k}|^{2} \delta(f - kf_{p})$   $\begin{cases} |c_{-3}|^{2} |c_{-1}|^{2} & |c_{0}|^{2} \\ |c_{-3}|^{2} |c_{-1}|^{2} & |c_{1}|^{2} \\ |c_{-3}|^{2} |c_{-2}|^{2} & |c_{-1}|^{2} \\ |c_{-3}|^{2} |c_{-2}|^{2} & |c_{-1}|^{2} \\ |c_{-3}|^{2} & |c_{-2}|^{2} \\ |c_{-3}|^{2} & |c_{-2}|^{2} \\ |c_{-3}|^{2} & |c_{-3}|^{2} \\ |c_{-3}|^{2} & |c_{-2}|^{2} \\ |c_{-3}|^{2} & |c_{-2}|^{2} \\ |c_{-3}|^{2} & |c_{-3}|^{2} \\ |c_{-3}|$ 

$$P = \int_{-\infty}^{\infty} P_x(f) df = \sum_{\tilde{k}=-\infty}^{\infty} |c_k|^2$$
(5.13)

By inspection of the discrete-frequency spectrum of x(t):

$$x(t) = 3e^{-j}e^{-j\pi^3t} + 8e^{-j\pi^2t} + 6e^{j2}e^{-j\pi t} + 5 + 6e^{-j2}e^{j\pi t} + 8e^{j\pi^2t} + 3e^{j}e^{j\pi^3t}$$
$$= 3e^{-j(\pi^3t+1)} + 8e^{-j\pi^2t} + 6e^{-j(\pi t-2)} + 5 + 6e^{j(\pi t-2)} + 8e^{j\pi^2t} + 3e^{j(\pi^3t+1)}$$

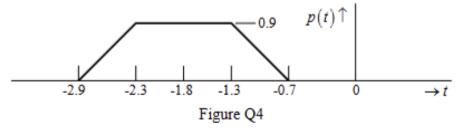
Average Power of 
$$x(t) = 3^2 + 8^2 + 6^2 + 5^2 + 6^2 + 8^2 + 3^2 = 243 W$$

(c) The frequencies  $\{\pi^3, \pi^2, \pi, 0, \pi, \pi^2, \pi^3\}$  do not have a common factor, indicating that x(t) is APERIODIC. Therefore, x(t) do not have a Fourier series expansion.

**REMARKS:** Summing sinusoids does not necessarily lead to a periodic signal unless the frequencies of the sinusoids are harmonically related, i.e. they have at least one common factor.

### T2 Q4

Q.4 Determine the Fourier series coefficients,  $c_k$ , of  $x(t) = \sum_{n=-\infty}^{\infty} 2p(t-1.6n)$  where p(t) is shown in Figure Q4.



$$x(t) = \sum_{n = -\infty}^{\infty} 2p(t - 1.6n) = p(t) * \sum_{n = -\infty}^{\infty} 2\delta(t - 1.6n)$$

$$= \cdots + \underbrace{2p(t + 1.6)}_{n = -1} + \underbrace{2p(t)}_{n = 0} + \underbrace{2p(t - 1.6)}_{n = 1} + \underbrace{2p(t - 3.2)}_{n = 2} + \underbrace{2p(t - 4.8)}_{n = 3} + \cdots$$

$$p(t)$$

$$\sum_{n = -1}^{\infty} 2b(t - 1.6n)$$

$$\sum_{$$

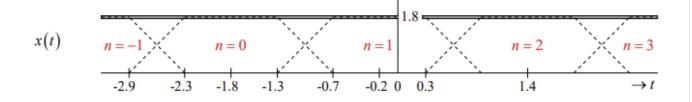
-0.2 0

-0.7

-1.8

 $\rightarrow t$ 

1.4



#### From the above figure, observe that x(t) = 1.8

Based on Fourier series expansion

$$x_p(t) = \sum_{k=-\infty}^{\infty} c_k e^{j2\pi kt/T_p}$$

We can rewrite x(t) = 1.8 as

$$x(t) = 1.8e^{j2\pi(0)t/T_p} + \sum_{k=-\infty}^{-1} 0e^{j2\pi kt/T_p} + \sum_{k=1}^{\infty} 0e^{j2\pi kt/T_p}$$

$$c_0$$

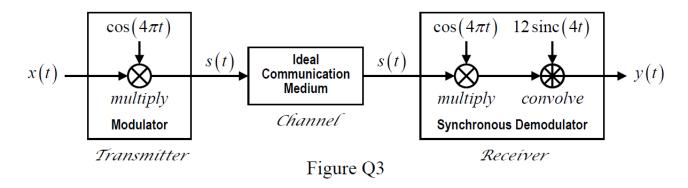
#### Therefore,

- x(t) has one zero-frequency component of value 1.8, which implies that  $c_0 = 1.8$ .
- x(t) has no non-zero frequency components, which implies that  $c_k = 0, \forall k \neq 0$

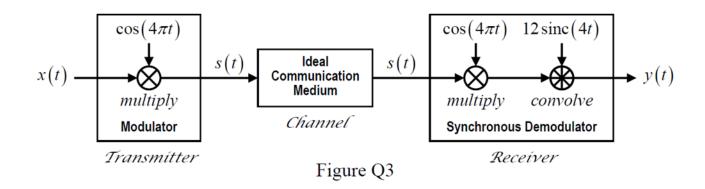
# **Tutorial 3**

### T3 Q3

Q.3 A communication system is shown in Figure Q3 where x(t) is the transmitter input signal and y(t) is the receiver output signal.



Find 
$$y(t)$$
 if  $x(t) = 4\operatorname{sinc}^2(t)$ .



$$x(t) = 4\operatorname{sinc}^2(t)$$

$$y(t) = \left[ \left[ x(t) \times \cos(4\pi t) \right] \times \cos(4\pi t) \right] + 12 \operatorname{sinc}(4t)$$

$$= \left[ 4 \operatorname{sinc}^2(t) \times \cos^2(4\pi t) \right] + 12 \operatorname{sinc}(4t)$$

$$= \left[ 4 \operatorname{sinc}^2(t) \times \frac{1}{2} \left[ 1 + \cos(8\pi t) \right] \right] + 12 \operatorname{sinc}(4t)$$

$$\cos^2(\theta) = 0.5[1 + \cos(2\theta)]$$

#### Method I

$$Y(f) = \Im\{y(t)\} = \left[\Im\{4\operatorname{sinc}^{2}(t)\} * \Im\{\frac{1}{2}[1 + \cos(8\pi t)]\}\right] \times \Im\{12\operatorname{sinc}(4t)\}$$

$$= \left[4\operatorname{tri}(f) * \left[\frac{1}{2}\delta(f) + \frac{1}{4}\delta(f+4) + \frac{1}{4}\delta(f-4)\right]\right] \times \operatorname{3rect}\left(\frac{f}{4}\right)$$

$$= \left[2\operatorname{tri}(f) + \operatorname{tri}(f+4) + \operatorname{tri}(f-4)\right] \times \operatorname{3rect}\left(\frac{f}{4}\right)$$

$$\cos(2\pi f_0 t) \leq \frac{1}{2} [\delta(f - f_0) + \delta(f + f_0)]$$

Replication: 
$$x(t) * \delta(t - \lambda) = x(t - \lambda)$$

$$\mathfrak{I}\{K\delta(t)\} = K$$

$$\mathfrak{I}\{K\} = K\delta(f)$$

$$\mathfrak{I}\{Ke^{j2\pi f_0}\} = K\delta(f - f_0)$$

$$Y(f) = \Im\{y(t)\} = \left[\Im\{4sinc^{2}(t)\} * \Im\left\{\frac{1}{2}[1 + \cos(8\pi t)]\right\}\right] \times \Im\{12sinc(4t)\}$$

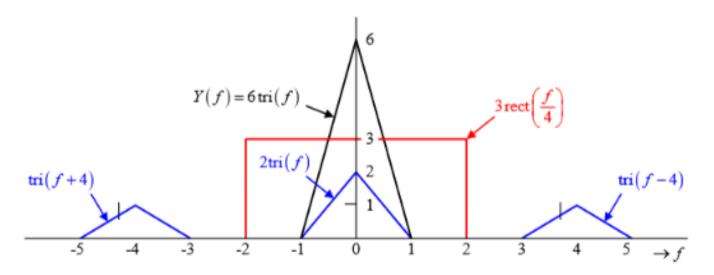
$$=\left[4tri(f)*\Im\left\{\frac{1}{2}\left[1+\frac{1}{2}e^{j8\pi t}-\frac{1}{2}e^{-j8\pi t}\right]\right\}\right]\times 3rect\left(\frac{t}{4}\right)=\left[4tri(f)*\Im\left\{\frac{1}{2}\left[1+\frac{1}{2}e^{j8\pi t}-\frac{1}{2}e^{-j8\pi t}\right]\right\}\right]\times 3rect\left(\frac{t}{4}\right)$$

$$= \left[4tri(f) * \left[\frac{1}{2}\delta(f) + \frac{1}{4}\delta(f+4) + \frac{1}{4}\delta(f-4)\right]\right] \times 3rect\left(\frac{t}{4}\right) = \left[2tri(f) + tri(f+4) + tri(f-4)\right] \times 3rect\left(\frac{t}{4}\right)$$

$$Y(f) = \left[2tri(f) + tri(f+4) + tri(f-4)\right] \times 3rect\left(\frac{f}{4}\right)$$

... by inspection of the sketch below

$$= 6 tri(f)$$



Therefore, 
$$y(t) = \Im^{-1} \{ 6 \operatorname{tri}(f) \} = 6 \operatorname{sinc}^{2}(t) = \frac{3}{2} x(t)$$
.

### T3 Q4

- Q.4 (a) Find the Fourier transform of  $x(t) = \cos^3(2\pi t)$ .
  - (b) The spectrum X(f) of a periodic signal x(t) is shown in Figure Q4. Express x(t) as a function of real sinusoids.

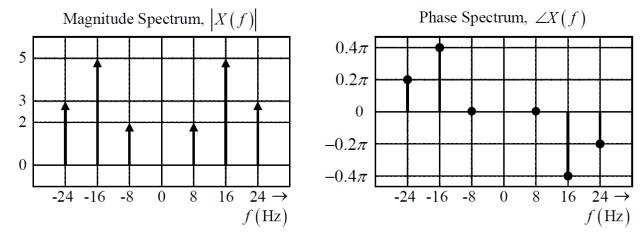


Figure Q4

**Convolution** is **Commutative** : x(t)\*y(t) = y(t)\*x(t).

$$(v) = y(v) + x(v).$$

• Convolution is **Associative** : 
$$[w(t)*x(t)]*y(t) = w(t)*[x(t)*y(t)]$$

**4** Convolution is **Distributive** : 
$$w(t)*[x(t)+y(t)]=w(t)*x(t)+w(t)*y(t)$$

Solution to Q.4

(a) 
$$\Im\{\cos(2\pi t)\} = \Im\{\frac{1}{2}e^{j2\pi t} + \frac{1}{2}e^{-j2\pi t}\} = \frac{1}{2}\delta(f-1) + \frac{1}{2}\delta(f+1)$$

Replication: 
$$x(t) * \delta(t - \lambda) = x(t - \lambda)$$

$$\Im\{\cos^{3}(2\pi t)\} = \Im\{\cos(2\pi t)\} * \Im\{\cos(2\pi t)\} * \Im\{\cos(2\pi t)\}$$

$$= \frac{1}{2} \Big[\delta(f-1) + \delta(f+1)\Big] * \frac{1}{2} \Big[\delta(f-1) + \delta(f+1)\Big] * \frac{1}{2} \Big[\delta(f-1) + \delta(f+1)\Big]$$

$$= \frac{1}{4} \Big[\delta(f-2) + \delta(f) + \delta(f) + \delta(f+2)\Big] * \frac{1}{2} \Big[\delta(f-1) + \delta(f+1)\Big]$$

$$= \frac{1}{8} \Big[\delta(f-1-2) + \delta(f-1) + \delta(f-1) + \delta(f-1+2)\Big]$$

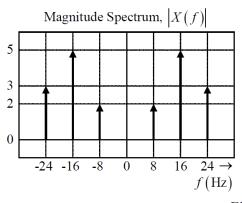
$$= \frac{1}{8} \Big[\delta(f-1-2) + \delta(f+1) + \delta(f+1) + \delta(f+1+2)\Big]$$

$$= \frac{1}{8} \Big[\delta(f-3) + 3\delta(f-1) + 3\delta(f+1) + \delta(f+3)\Big] \qquad \cos(2\pi f_0 t) \iff \frac{1}{2} \Big[\delta(f-f_0) + \delta(f+f_0)\Big]$$

Remarks: Another approach is to use trigonometric identities to first convert  $\cos^3(2\pi t)$  into

$$\left[\frac{3}{4}\cos(2\pi t) + \frac{1}{4}\cos(6\pi t)\right]$$
 then apply Fourier transform. This will do away with the multiple convolution operations.

(b) The spectrum X(f) of a periodic signal x(t) is shown in Figure Q4. Express x(t) as a function of real sinusoids.



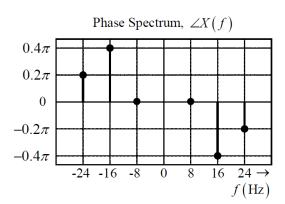


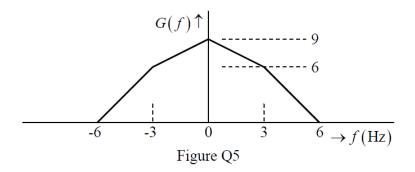
Figure Q4

$$\begin{split} X(f) &= 2 \Big[ \delta(f-8) + \delta(f+8) \Big] + 5 \Big[ e^{-j0.4\pi} \delta(f-16) + e^{j0.4\pi} \delta(f+16) \Big] \\ &+ 3 \Big[ e^{-j0.2\pi} \delta(f-24) + e^{j0.2\pi} \delta(f+24) \Big] \end{split}$$

$$x(t) = 2\left[e^{j2\pi(8)t} + e^{j2\pi(-8)t}\right] + 5\left[e^{-j0.4\pi}e^{j2\pi(16)t} + e^{j0.4\pi}e^{j2\pi(-16)t}\right]$$
$$+3\left[e^{-j0.2\pi}e^{j2\pi(24)t} + e^{j0.2\pi}e^{j2\pi(-24)t}\right]$$
$$= 4\cos(16\pi t) + 10\cos(32\pi t - 0.4\pi) + 6\cos(48\pi t - 0.2\pi)$$

### T3 Q5

Q.5 A function g(t) is replicated at intervals of 0.5 seconds to generate the periodic signal x(t). The Fourier transform, G(f), of g(t) is shown in Figure Q5.



- (a) Derive the Fourier transform, X(f), of x(t) and sketch the continuous-frequency spectrum of x(t).
- (b) Derive the Fourier series coefficients,  $c_k$ , of x(t) and sketch the discrete-frequency spectrum of x(t).
- (c) Show that the inverse Fourier transform of the X(f) found in part (a) is the equal to the Fourier series expansion of x(t).

$$\Im\{comb_{\lambda}(t)\} = \frac{1}{\lambda} \sum_{k} c_{k} \delta(f - k/\lambda)$$

(a) Fourier transform, X(f), of x(t):

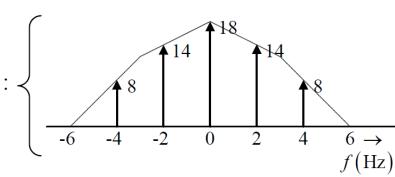
$$x(t) = g(t) * \sum_{n} \delta(t - 0.5n)$$

$$X(f) = \Im\{g(t) * \sum_{n} \delta(t - 0.5n)\} = \Im\{g(t)\} \times \Im\{\sum_{n} \delta(t - 0.5n)\}$$

$$= G(f) \times 2\sum_{k} \delta(f - 2k) = \sum_{k} 2G(2k)\delta(f - 2k)$$

$$= 8\delta(f + 4) + 14\delta(f + 2) + 18\delta(f) + 14\delta(f - 2) + 8\delta(f - 4)$$

Continuous-frequency spectrum, X(f):  $\prec$ 



$$X_{p}(f) = \sum_{k=-\infty}^{\infty} c_{k} \delta\left(f - \frac{k}{T_{p}}\right) = \sum_{k=-\infty}^{\infty} c_{k} \delta\left(f - kf_{p}\right)$$

(4.16)

(b) Fourier series coefficients,  $c_k$ , of x(t):

From part (a): 
$$X(f) = \sum_{k} 2G(2k)\delta(f - 2k) \qquad (1)$$

But, X(f) is also given by:

$$X(f) = \sum_{k} c_k \delta(f - 2k) \qquad (2)$$

Comparing (1) and (2), we obtain  $c_k = 2G(2k)$ .

Discrete-frequency spectrum, 
$$c_k$$
:  $c_0 = 18, c_{\pm 1} = 14, c_{\pm 2} = 8$   $c_k = 0; \forall |k| \ge 3$ 

x(t)

From part (a):

expansion of x(t) is

(c)

Periodic Signal

 $X(f) = 8\delta(f+4) + 14\delta(f+2) + 18\delta(f) + 14\delta(f-2) + 8\delta(f-4)$ 

 $x(t) = \Im^{-1}\left\{X(f)\right\} = 8e^{-j8\pi t} + 14e^{-j4\pi t} + 18 + 14e^{j4\pi t} + 8e^{j8\pi t}$ 

With the period of x(t) equal to 0.5 seconds and the  $c_k$ 's found in part (b), the Fourier series

 $x(t) = \sum_{k} c_k e^{j4\pi kt} = 8e^{-j8\pi t} + 14e^{-j4\pi t} + 18 + 14e^{j4\pi t} + 8e^{j8\pi t}$ 

Fourier Series Coefficients / Discretefrequency Spectrum

Continuous-frequency Spectrum  $X_p(f) = \sum_{k} c_k \delta\left(f - \frac{k}{T_p}\right)$ 

> Inverse Fourier Transform •  $\Im\{Ke^{j2\pi f_0 t}\}=K\delta(f-f_0)$

 $\Im\{K\delta(t)\}=K$ 

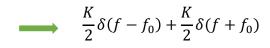
 $\Im\{K\} = K\delta(f)$ 

 $\Im\{Ke^{j2\pi f_0}\}=K\delta(f-f_0)$ 

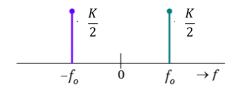
### Solution Approach

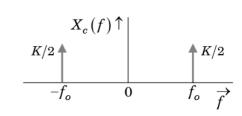
Discrete-frequency Spectrum

$$Kcos(2\pi f_0 t) = \frac{K}{2} e^{j2\pi f_0 t} + \frac{K}{2} e^{-j2\pi f_0 t} \longrightarrow \frac{K}{2} \delta(f - f_0) + \frac{K}{2} \delta(f + f_0)$$



Magnitude Spectrum of  $x_c(t)$ 





- Properties of  $\delta(t)$ 
  - $\circ$  Symmetry :  $\delta(t) = \delta(-t)$
  - $\circ$  Sampling:  $x(t)\delta(t-\lambda) =$  $x(\lambda)\delta(t-\lambda)$
  - $\circ$  Sifting:  $\int_{-\infty}^{\infty} x(t)\delta(t-\lambda)dt = x(\lambda)$
  - $\circ$  Replication:  $x(t) * \delta(t \lambda) =$  $x(t-\lambda)$
- The Dirac- $\delta$  Comb function:

$$comb_{\lambda}(t) = \sum_{i} \delta(t - n\lambda)$$

Model a periodic signal:

$$x_p(t) = x(t) * \sum_{n} \delta(t - n\lambda)$$

Model a sampled signal:

$$x_s(t) = x(t) \times \sum_{i} \delta(t - n\lambda)$$

# **Tutorial 4**

### T4 Q3

Q.3 The spectrum of a lowpass energy signal x(t) is given by  $X(f) = 10^{-0.1|f|}$ . Find the 90% energy containment bandwidth of x(t). Fact:  $10^{\beta} = e^{\beta \ln(10)}$ 

 $\uparrow E_x(f)$ 

The M% energy containment bandwidth, B, of a real energy signal is the smallest bandwidth that

 $\int_{f_c - 0.5B}^{f_c + 0.5B} E_x(f) df = \frac{M}{100} \times \frac{E}{2}$ 

Spectrum of x(t):  $X(f) = 10^{-0.1|f|} = e^{-\lfloor 0.1 \ln(10) \rfloor |f|}$ 

ESD of 
$$x(t)$$
:  $E_x(f) = |X(f)|^2 = e^{-[0.2\ln(10)]\cdot |f|}$ 

Energy of x(t) contained within a bandwidth of B:

$$e(B) = \int_{-B}^{B} E_{x}(f) df = 2 \int_{0}^{B} e^{-\left[0.2 \ln(10)\right] \cdot f} df = 2 \left[ \frac{e^{-\left[0.2 \ln(10)\right] \cdot f}}{-0.2 \ln(10)} \right]_{0}^{B}$$
$$= \frac{10}{\ln(10)} \left[ 1 - e^{-\left[0.2 \ln(10)\right] \cdot B} \right] \qquad (1)$$

Total energy of x(t) is equal to energy of x(t) contained within a bandwidth of  $\infty$ . Hence, substituting  $B = \infty$  into (1), we get

$$e(\infty) = \int_{-\infty}^{\infty} E_x(f) df = \frac{10}{\ln(10)}$$
 ····· TOTAL ENERGY

LOWPASS SIGNAL x(t):

contains at least M% of the total signal energy  $E=\int_{-\infty}^{\infty}E_x(f)df$  .

frequency

BANDPASS SIGNAL x(t):

$$e(B) = \frac{10}{\ln(10)} \left[ 1 - e^{-[0.2 \ln(10)]B} \right]$$

The 90% energy containment bandwidth,  $f_{90}$ , of x(t) is obtained by solving:

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### T4 Q4

Q.4 Derive and sketch the spectrum of  $72\text{sinc}^2(12t) * \sum_n \delta(t-0.5n)$  where \* denotes convolution. Hence, determine the power spectral density and the average power of x(t)?

Signal, x(t):

$$x(t) = 72 \operatorname{sinc}^{2}(12t) * \sum_{n} \delta(t - 0.5n)$$

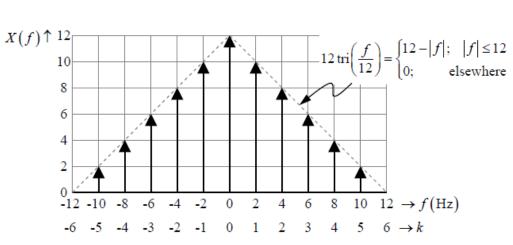
$$\Im\{comb_{\lambda}(t)\} = \frac{1}{\lambda} \sum_{k} c_{k} \delta(f - k/\lambda)$$

Spectrum, X(f):

$$X(f) = \Im\{x(t)\} = \Im\{72\operatorname{sinc}^2(12t)\} \times \Im\{\sum_n \delta(t - 0.5n)\}$$
$$= 6\operatorname{tri}\left(\frac{f}{12}\right) \times \frac{1}{0.5} \sum_k \delta\left(f - \frac{k}{0.5}\right) = 12\operatorname{tri}\left(\frac{f}{12}\right) \times \sum_k \delta\left(f - 2k\right)$$

 $\cdots$  deducing from the plot of X(f) below

$$= \sum_{k=-5}^{5} (12 - |f|) \delta(f - 2k) = \sum_{k=-5}^{5} (12 - 2|k|) \delta(f - 2k)$$



• The Dirac- $\delta$  Comb function:

$$comb_{\lambda}(t) = \sum_{n} \delta(t - n\lambda)$$

Model a periodic signal:

$$x_p(t) = x(t) * \sum \delta(t - n\lambda)$$

Model a sampled signal:

$$x_s(t) = x(t) \times \sum \delta(t - n\lambda)$$

$$P = \int_{-\infty}^{\infty} P_x(f) df = \int_{-\infty}^{\infty} \sum_{k=-\infty}^{\infty} |c_k|^2 \delta(f - kf_p) df = \sum_{k=-\infty}^{\infty} |c_k|^2 \int_{-\infty}^{\infty} \delta(f - kf_p) df = \sum_{k=-\infty}^{\infty} |c_k|^2$$
Power spectral density,  $P_x(f)$ :
$$P_x(f) = \sum_{k=-5}^{5} (12 - 2|k|)^2 \delta(f - 2k)$$

 $\int_{0}^{\infty} P(s) ds = \sum_{k=0}^{5} (12 - 2|k|)^{2} \int_{0}^{\infty} s(s-2k)^{2}$ 

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 $\int_{-\infty}^{\varepsilon} \delta(t)dt = 1, \forall \varepsilon > 0$ 

$$P = \int_{-\infty}^{\infty} P_X(f) df = \sum_{k=-5}^{5} (12 - 2|k|)^2 \underbrace{\int_{-\infty}^{\infty} \delta(f - 2k) df}_{=1} = \sum_{k=-5}^{5} (12 - 2|k|)^2$$

$$= 12^2 + 2(10^2 + 8^2 + 6^2 + 4^2 + 2^2) = 584 \text{ Watts}$$

Power Spectral Density of  $x_p(t)$ :

 $P_x(f) = \sum_{k=0}^{\infty} |c_k|^2 \delta(f - kf_p)$ 

 $P = \int_{-\infty}^{\infty} P_x(f) df = \sum_{k=-\infty}^{\infty} |c_k|^2$ 

Average power, P:

Average Power of  $x_p(t)$ :

 $f_0$ ,  $f_1$ ,  $f_2$ ,  $f_3$  and  $f_4$  have no common factor  $|\mu_1|$   $|\mu_2|$   $|\mu_3|$   $|\mu_4|$   $|\mu_4|$ Spectrum  $f_0$   $f_1$   $f_2$   $f_3$   $f_4$ Power

Spectral Density  $|C_k|^2$ 

Non-periodic signal composed of a sum of sinusoids

# Thanks!