

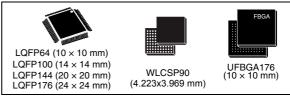
STM32F415xx STM32F417xx

ARM Cortex-M4 32b MCU+FPU, 210DMIPS, up to 1MB Flash/192+4KB RAM, crypto, USB OTG HS/FS, Ethernet, 17 TIMs, 3 ADCs, 15 comm. interfaces & camera

Datasheet - production data

Features

- Core: ARM[®] 32-bit Cortex[®]-M4 CPU with FPU, Adaptive real-time accelerator (ART Accelerator[™]) allowing 0-wait state execution from Flash memory, frequency up to 168 MHz, memory protection unit, 210 DMIPS/ 1.25 DMIPS/MHz (Dhrystone 2.1), and DSP instructions
- Memories
 - Up to 1 Mbyte of Flash memory
 - Up to 192+4 Kbytes of SRAM including 64-Kbyte of CCM (core coupled memory) data RAM
 - Flexible static memory controller supporting Compact Flash, SRAM, PSRAM, NOR and NAND memories
- LCD parallel interface, 8080/6800 modes
- · Clock, reset and supply management
 - 1.8 V to 3.6 V application supply and I/Os
 - POR, PDR, PVD and BOR
 - 4-to-26 MHz crystal oscillator
 - Internal 16 MHz factory-trimmed RC (1% accuracy)
 - 32 kHz oscillator for RTC with calibration
 - Internal 32 kHz RC with calibration
- Low-power operation
 - Sleep, Stop and Standby modes
 - V_{BAT} supply for RTC, 20×32 bit backup registers + optional 4 KB backup SRAM
- 3×12-bit, 2.4 MSPS A/D converters: up to 24 channels and 7.2 MSPS in triple interleaved mode
- 2×12-bit D/A converters
- General-purpose DMA: 16-stream DMA controller with FIFOs and burst support
- Up to 17 timers: up to twelve 16-bit and two 32-bit timers up to 168 MHz, each with up to 4 IC/OC/PWM or pulse counter and quadrature (incremental) encoder input
- · Debug mode
 - Serial wire debug (SWD) & JTAG interfaces
 - Cortex-M4 Embedded Trace Macrocell™



- Up to 140 I/O ports with interrupt capability
 - Up to 136 fast I/Os up to 84 MHz
 - Up to 138 5 V-tolerant I/Os
- Up to 15 communication interfaces
 - Up to 3 × I²C interfaces (SMBus/PMBus)
 - Up to 4 USARTs/2 UARTs (10.5 Mbit/s, ISO 7816 interface, LIN, IrDA, modem control)
 - Up to 3 SPIs (42 Mbits/s), 2 with muxed full-duplex I²S to achieve audio class accuracy via internal audio PLL or external clock
 - 2 × CAN interfaces (2.0B Active)
 - SDIO interface
- Advanced connectivity
 - USB 2.0 full-speed device/host/OTG controller with on-chip PHY
 - USB 2.0 high-speed/full-speed device/host/OTG controller with dedicated DMA, on-chip full-speed PHY and ULPI
 - 10/100 Ethernet MAC with dedicated DMA: supports IEEE 1588v2 hardware, MII/RMII
- 8- to 14-bit parallel camera interface up to 54 Mbytes/s
- Cryptographic acceleration: hardware acceleration for AES 128, 192, 256, Triple DES, HASH (MD5, SHA-1), and HMAC
- True random number generator
- CRC calculation unit
- 96-bit unique ID
- RTC: subsecond accuracy, hardware calendar

Table 1. Device summary

Reference	Part number
STM32F415xx	STM32F415RG, STM32F415VG, STM32F415ZG, STM32F415OG
STM32F417xx	STM32F417VG, STM32F417IG, STM32F417ZG, STM32F417VE, STM32F417ZE, STM32F417IE

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1 Introduction

This datasheet provides the description of the STM32F415xx and STM32F417xx lines of microcontrollers. For more details on the whole STMicroelectronics STM32[™] family, please refer to Section 2.1: Full compatibility throughout the family.

The STM32F415xx and STM32F417xx datasheet should be read in conjunction with the STM32F4xx reference manual which is available from the STMicroelectronics website www.st.com.

For information on the Cortex[®]-M4 core, please refer to the Cortex[®]-M4 programming manual (PM0214) available from *www.st.com*.

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2 Description

The STM32F415xx and STM32F417xx family is based on the high-performance ARM[®] Cortex[®]-M4 32-bit RISC core operating at a frequency of up to 168 MHz. The Cortex-M4 core features a Floating point unit (FPU) single precision which supports all ARM single-precision data-processing instructions and data types. It also implements a full set of DSP instructions and a memory protection unit (MPU) which enhances application security.

The STM32F415xx and STM32F417xx family incorporates high-speed embedded memories (Flash memory up to 1 Mbyte, up to 192 Kbytes of SRAM), up to 4 Kbytes of backup SRAM, and an extensive range of enhanced I/Os and peripherals connected to two APB buses, three AHB buses and a 32-bit multi-AHB bus matrix.

All devices offer three 12-bit ADCs, two DACs, a low-power RTC, twelve general-purpose 16-bit timers including two PWM timers for motor control, two general-purpose 32-bit timers. a true random number generator (RNG), and a cryptographic acceleration cell. They also feature standard and advanced communication interfaces.

- Up to three I²Cs
- Three SPIs, two I²Ss full duplex. To achieve audio class accuracy, the I2S peripherals
 can be clocked via a dedicated internal audio PLL or via an external clock to allow
 synchronization.
- Four USARTs plus two UARTs
- An USB OTG full-speed and a USB OTG high-speed with full-speed capability (with the ULPI),
- Two CANs
- An SDIO/MMC interface
- Ethernet and the camera interface available on STM32F417xx devices only.

New advanced peripherals include an SDIO, an enhanced flexible static memory control (FSMC) interface (for devices offered in packages of 100 pins and more), a camera interface for CMOS sensors and a cryptographic acceleration cell. Refer to *Table 2:* STM32F415xx and STM32F417xx: features and peripheral counts for the list of peripherals available on each part number.

The STM32F415xx and STM32F417xx family operates in the –40 to +105 °C temperature range from a 1.8 to 3.6 V power supply. The supply voltage can drop to 1.7 V when the device operates in the 0 to 70 °C temperature range using an external power supply supervisor: refer to *Section : Internal reset OFF*. A comprehensive set of power-saving mode allows the design of low-power applications.

The STM32F415xx and STM32F417xx family offers devices in various packages ranging from 64 pins to 176 pins. The set of included peripherals changes with the device chosen.

These features make the STM32F415xx and STM32F417xx microcontroller family suitable for a wide range of applications:

- Motor drive and application control
- Medical equipment
- Industrial applications: PLC, inverters, circuit breakers
- Printers, and scanners
- Alarm systems, video intercom, and HVAC
- Home audio appliances

Figure 5 shows the general block diagram of the device family.

Table 2. STM32F415xx and STM32F417xx: features and peripheral counts

Dorin	horala	STM32F415RG	STM32F415OG	STM32F415VG	STM32F415ZG	Ī		CTM22	F417Zx	CTM2	2F417lx
Peripherals		31W32F413RG			31W32F4152G		T				T
Flash memory in Kbytes			10.	24		512	1024	512	1024	512	1024
SRAM in	System				192(112+16+64)						
Kbytes	Backup				4						
FSMC memory	controller	No			Yes	(1)					
Ethernet			N	0				Υe	es		
	General- purpose				10						
	Advanced- control		2								
Timers	Basic		2								
	IWDG	Yes									
	WWDG	Yes									
	RTC	Yes									
Random number	er generator	Yes									
	SPI / I ² S	3/2 (full duplex) ⁽²⁾									
	I ² C	3									
	USART/UART	4/2									
Communicatio n interfaces	USB OTG FS		Yes								
II IIILEIIACES	USB OTG HS		Yes								
	CAN		2								
	SDIO	Yes									
Camera interfac	ce	No Yes									
Cryptography			Yes								





Table 2. STM32F415xx and STM32F417xx: features and peripheral counts

Peripherals	STM32F415RG	STM32F415OG	STM32F415VG	STM32F415ZG	STM32F417Vx	STM32F417Zx	STM32F417Ix	
GPIOs	51	72	82	114	82	114	140	
12-bit ADC Number of channels	3							
	16	13	16	24	16	24	24	
12-bit DAC Number of channels	Yes 2							
Maximum CPU frequency	168 MHz							
Operating voltage	1.8 to 3.6 V ⁽³⁾							
Operating temperatures	Ambient temperatures: –40 to +85 °C /–40 to +105 °C							
	Junction temperature: -40 to + 125 °C							
Package	LQFP64	WLCSP90	LQFP100	LQFP144	LQFP100	LQFP144	UFBGA176 LQFP176	

^{1.} For the LQFP100 and WLCSP90 packages, only FSMC Bank1 or Bank2 are available. Bank1 can only support a multiplexed NOR/PSRAM memory using the NE1 Chip Select. Bank2 can only support a 16- or 8-bit NAND Flash memory using the NCE2 Chip Select. The interrupt line cannot be used since Port G is not available in this package.

^{2.} The SPI2 and SPI3 interfaces give the flexibility to work in an exclusive way in either the SPI mode or the I2S audio mode.

^{3.} V_{DD}/V_{DDA} minimum value of 1.7 V is obtained when the device operates in reduced temperature range, and with the use of an external power supply supervisor (refer to *Section : Internal reset OFF*).

2.1 Full compatibility throughout the family

The STM32F415xx and STM32F417xx are part of the STM32F4 family. They are fully pinto-pin, software and feature compatible with the STM32F2xx devices, allowing the user to try different memory densities, peripherals, and performances (FPU, higher frequency) for a greater degree of freedom during the development cycle.

The STM32F415xx and STM32F417xx devices maintain a close compatibility with the whole STM32F10xxx family. All functional pins are pin-to-pin compatible. The STM32F415xx and STM32F417xx, however, are not drop-in replacements for the STM32F10xxx devices: the two families do not have the same power scheme, and so their power pins are different. Nonetheless, transition from the STM32F10xxx to the STM32F41xxx family remains simple as only a few pins are impacted.

Figure 4, *Figure 3*, *Figure 2*, and *Figure 1* give compatible board designs between the STM32F41xxx, STM32F2, and STM32F10xxx families.

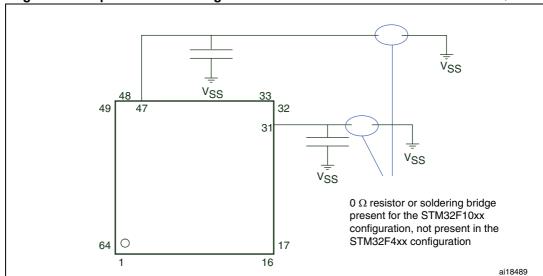


Figure 1. Compatible board design between STM32F10xx/STM32F41xxx for LQFP64

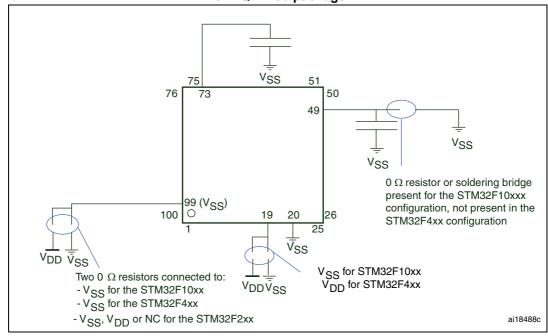
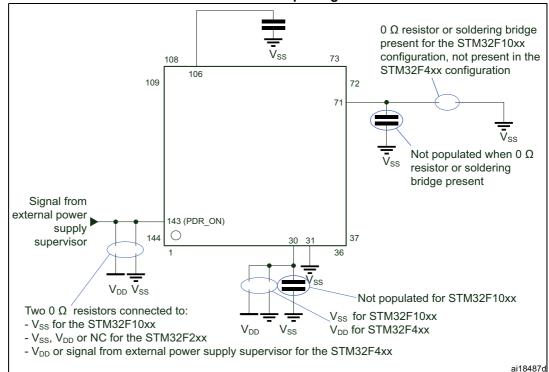


Figure 2. Compatible board design STM32F10xx/STM32F2/STM32F41xxx for LQFP100 package

Figure 3. Compatible board design between STM32F10xx/STM32F2/STM32F41xxx for LQFP144 package



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Signal from external power supply supervisor 171 (PDR_ON) 171 (PDR_ON) $170 \text{ Two } 0 \Omega \text{ resistors connected to:}$ $- \text{V}_{SS}, \text{V}_{DD} \text{ or NC for the STM32F2xx}$ $- \text{V}_{DD} \text{ or signal from external power supply supervisor for the STM32F4xx}$ MS19919V3

Figure 4. Compatible board design between STM32F2 and STM32F41xxx for LQFP176 and BGA176 packages

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2.2 Device overview

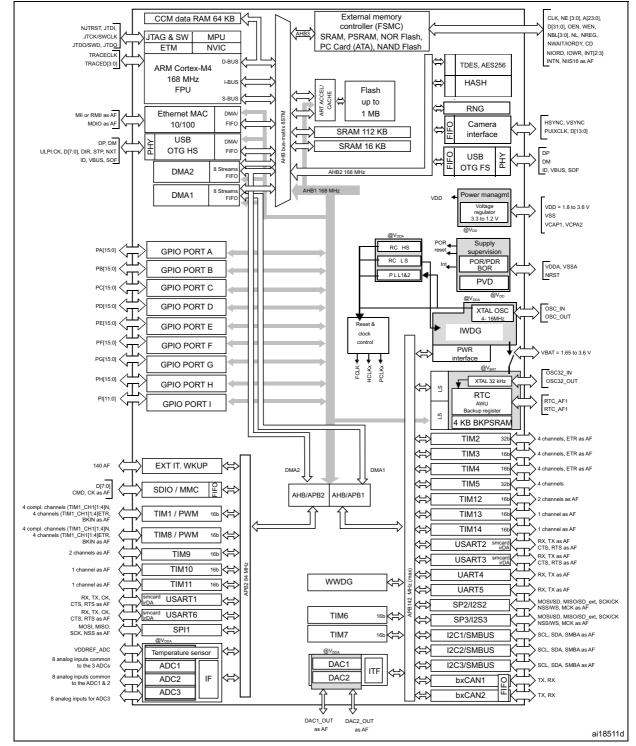


Figure 5. STM32F41xxx block diagram

The timers connected to APB2 are clocked from TIMxCLK up to 168 MHz, while the timers connected to APB1 are clocked from TIMxCLK either up to 84 MHz or 168 MHz, depending on TIMPRE bit configuration in the RCC_DCKCFGR register.

^{2.} The camera interface and ethernet are available only on STM32F417xx devices.

2.2.1 ARM® Cortex®-M4 core with FPU and embedded Flash and SRAM

The ARM Cortex-M4 processor with FPU is the latest generation of ARM processors for embedded systems. It was developed to provide a low-cost platform that meets the needs of MCU implementation, with a reduced pin count and low-power consumption, while delivering outstanding computational performance and an advanced response to interrupts.

The ARM Cortex-M4 32-bit RISC processor with FPU features exceptional code-efficiency, delivering the high-performance expected from an ARM core in the memory size usually associated with 8- and 16-bit devices.

The processor supports a set of DSP instructions which allow efficient signal processing and complex algorithm execution.

Its single precision FPU (floating point unit) speeds up software development by using metalanguage development tools, while avoiding saturation.

The STM32F415xx and STM32F417xx family is compatible with all ARM tools and software.

Figure 5 shows the general block diagram of the STM32F41xxx family.

Note: Cortex-M4 with FPU is binary compatible with Cortex-M3.

2.2.2 Adaptive real-time memory accelerator (ART Accelerator™)

The ART Accelerator™ is a memory accelerator which is optimized for STM32 industry-standard ARM® Cortex®-M4 with FPU processors. It balances the inherent performance advantage of the ARM Cortex-M4 with FPU over Flash memory technologies, which normally requires the processor to wait for the Flash memory at higher frequencies.

To release the processor full 210 DMIPS performance at this frequency, the accelerator implements an instruction prefetch queue and branch cache, which increases program execution speed from the 128-bit Flash memory. Based on CoreMark benchmark, the performance achieved thanks to the ART accelerator is equivalent to 0 wait state program execution from Flash memory at a CPU frequency up to 168 MHz.

2.2.3 Memory protection unit

The memory protection unit (MPU) is used to manage the CPU accesses to memory to prevent one task to accidentally corrupt the memory or resources used by any other active task. This memory area is organized into up to 8 protected areas that can in turn be divided up into 8 subareas. The protection area sizes are between 32 bytes and the whole 4 gigabytes of addressable memory.

The MPU is especially helpful for applications where some critical or certified code has to be protected against the misbehavior of other tasks. It is usually managed by an RTOS (real-time operating system). If a program accesses a memory location that is prohibited by the MPU, the RTOS can detect it and take action. In an RTOS environment, the kernel can dynamically update the MPU area setting, based on the process to be executed.

The MPU is optional and can be bypassed for applications that do not need it.

2.2.4 Embedded Flash memory

The STM32F41xxx devices embed a Flash memory of 512 Kbytes or 1 Mbytes available for storing programs and data.

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2.2.5 CRC (cyclic redundancy check) calculation unit

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code from a 32-bit data word and a fixed generator polynomial.

Among other applications, CRC-based techniques are used to verify data transmission or storage integrity. In the scope of the EN/IEC 60335-1 standard, they offer a means of verifying the Flash memory integrity. The CRC calculation unit helps compute a software signature during runtime, to be compared with a reference signature generated at link-time and stored at a given memory location.

2.2.6 Embedded SRAM

All STM32F41xxx products embed:

- Up to 192 Kbytes of system SRAM including 64 Kbytes of CCM (core coupled memory) data RAM
 - RAM memory is accessed (read/write) at CPU clock speed with 0 wait states.
- 4 Kbytes of backup SRAM

This area is accessible only from the CPU. Its content is protected against possible unwanted write accesses, and is retained in Standby or VBAT mode.

2.2.7 Multi-AHB bus matrix

The 32-bit multi-AHB bus matrix interconnects all the masters (CPU, DMAs, Ethernet, USB HS) and the slaves (Flash memory, RAM, FSMC, AHB and APB peripherals) and ensures a seamless and efficient operation even when several high-speed peripherals work simultaneously.

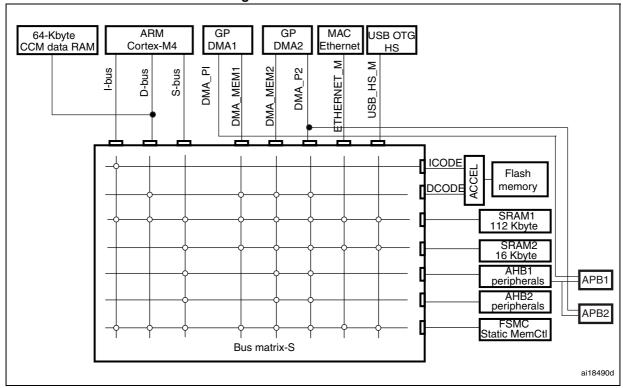


Figure 6. Multi-AHB matrix

2.2.8 DMA controller (DMA)

The devices feature two general-purpose dual-port DMAs (DMA1 and DMA2) with 8 streams each. They are able to manage memory-to-memory, peripheral-to-memory and memory-to-peripheral transfers. They feature dedicated FIFOs for APB/AHB peripherals, support burst transfer and are designed to provide the maximum peripheral bandwidth (AHB/APB).

The two DMA controllers support circular buffer management, so that no specific code is needed when the controller reaches the end of the buffer. The two DMA controllers also have a double buffering feature, which automates the use and switching of two memory buffers without requiring any special code.

Each stream is connected to dedicated hardware DMA requests, with support for software trigger on each stream. Configuration is made by software and transfer sizes between source and destination are independent.

The DMA can be used with the main peripherals:

- SPI and I²S
- I²C
- USART
- General-purpose, basic and advanced-control timers TIMx
- DAC
- SDIO
- Cryptographic acceleration
- Camera interface (DCMI)
- ADC.

2.2.9 Flexible static memory controller (FSMC)

The FSMC is embedded in the STM32F415xx and STM32F417xx family. It has four Chip Select outputs supporting the following modes: PCCard/Compact Flash, SRAM, PSRAM, NOR Flash and NAND Flash.

Functionality overview:

- Write FIFO
- Maximum FSMC CLK frequency for synchronous accesses is 60 MHz.

LCD parallel interface

The FSMC can be configured to interface seamlessly with most graphic LCD controllers. It supports the Intel 8080 and Motorola 6800 modes, and is flexible enough to adapt to specific LCD interfaces. This LCD parallel interface capability makes it easy to build cost-effective graphic applications using LCD modules with embedded controllers or high performance solutions using external controllers with dedicated acceleration.

2.2.10 Nested vectored interrupt controller (NVIC)

The STM32F415xx and STM32F417xx embed a nested vectored interrupt controller able to manage 16 priority levels, and handle up to 82 maskable interrupt channels plus the 16 interrupt lines of the Cortex[®]-M4 with FPU core.

- Closely coupled NVIC gives low-latency interrupt processing
- Interrupt entry vector table address passed directly to the core
- Allows early processing of interrupts
- Processing of late arriving, higher-priority interrupts
- Support tail chaining
- Processor state automatically saved
- Interrupt entry restored on interrupt exit with no instruction overhead

This hardware block provides flexible interrupt management features with minimum interrupt latency.

2.2.11 External interrupt/event controller (EXTI)

The external interrupt/event controller consists of 23 edge-detector lines used to generate interrupt/event requests. Each line can be independently configured to select the trigger

event (rising edge, falling edge, both) and can be masked independently. A pending register maintains the status of the interrupt requests. The EXTI can detect an external line with a pulse width shorter than the Internal APB2 clock period. Up to 140 GPIOs can be connected to the 16 external interrupt lines.

2.2.12 Clocks and startup

On reset the 16 MHz internal RC oscillator is selected as the default CPU clock. The 16 MHz internal RC oscillator is factory-trimmed to offer 1% accuracy over the full temperature range. The application can then select as system clock either the RC oscillator or an external 4-26 MHz clock source. This clock can be monitored for failure. If a failure is detected, the system automatically switches back to the internal RC oscillator and a software interrupt is generated (if enabled). This clock source is input to a PLL thus allowing to increase the frequency up to 168 MHz. Similarly, full interrupt management of the PLL clock entry is available when necessary (for example if an indirectly used external oscillator fails).

Several prescalers allow the configuration of the three AHB buses, the high-speed APB (APB2) and the low-speed APB (APB1) domains. The maximum frequency of the three AHB buses is 168 MHz while the maximum frequency of the high-speed APB domains is 84 MHz. The maximum allowed frequency of the low-speed APB domain is 42 MHz.

The devices embed a dedicated PLL (PLLI2S) which allows to achieve audio class performance. In this case, the I²S master clock can generate all standard sampling frequencies from 8 kHz to 192 kHz.

2.2.13 Boot modes

At startup, boot pins are used to select one out of three boot options:

- Boot from user Flash
- Boot from system memory
- Boot from embedded SRAM

The boot loader is located in system memory. It is used to reprogram the Flash memory by using USART1 (PA9/PA10), USART3 (PC10/PC11 or PB10/PB11), CAN2 (PB5/PB13), USB OTG FS in Device mode (PA11/PA12) through DFU (device firmware upgrade).

2.2.14 Power supply schemes

- V_{DD} = 1.8 to 3.6 V: external power supply for I/Os and the internal regulator (when enabled), provided externally through V_{DD} pins.
- V_{SSA}, V_{DDA} = 1.8 to 3.6 V: external analog power supplies for ADC, DAC, Reset blocks, RCs and PLL. V_{DDA} and V_{SSA} must be connected to V_{DD} and V_{SS}, respectively.
- V_{BAT} = 1.65 to 3.6 V: power supply for RTC, external clock 32 kHz oscillator and backup registers (through power switch) when V_{DD} is not present.

Refer to Figure 21: Power supply scheme for more details.

Note:

 V_{DD}/V_{DDA} minimum value of 1.7 V is obtained when the device operates in reduced temperature range, and with the use of an external power supply supervisor (refer to Section: Internal reset OFF).

Refer to Table 2 in order to identify the packages supporting this option.

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2.2.15 Power supply supervisor

Internal reset ON

On packages embedding the PDR_ON pin, the power supply supervisor is enabled by holding PDR_ON high. On all other packages, the power supply supervisor is always enabled.

The device has an integrated power-on reset (POR) / power-down reset (PDR) circuitry coupled with a Brownout reset (BOR) circuitry. At power-on, POR/PDR is always active and ensures proper operation starting from 1.8 V. After the 1.8 V POR threshold level is reached, the option byte loading process starts, either to confirm or modify default BOR threshold levels, or to disable BOR permanently. Three BOR thresholds are available through option bytes. The device remains in reset mode when $V_{\rm DD}$ is below a specified threshold, $V_{\rm POR/PDR}$ or $V_{\rm BOR}$, without the need for an external reset circuit.

The device also features an embedded programmable voltage detector (PVD) that monitors the V_{DD}/V_{DDA} power supply and compares it to the V_{PVD} threshold. An interrupt can be generated when V_{DD}/V_{DDA} drops below the V_{PVD} threshold and/or when V_{DD}/V_{DDA} is higher than the V_{PVD} threshold. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state. The PVD is enabled by software.

Internal reset OFF

This feature is available only on packages featuring the PDR_ON pin. The internal power-on reset (POR) / power-down reset (PDR) circuitry is disabled with the PDR_ON pin.

An external power supply supervisor should monitor V_{DD} and should maintain the device in reset mode as long as V_{DD} is below a specified threshold. PDR_ON should be connected to this external power supply supervisor. Refer to Figure 7: Power supply supervisor interconnection with internal reset OFF.

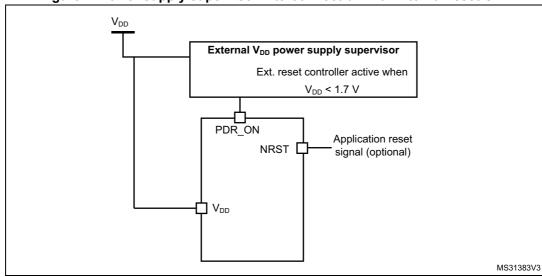


Figure 7. Power supply supervisor interconnection with internal reset OFF

1. PDR = 1.7 V for reduce temperature range; PDR = 1.8 V for all temperature range.

The V_{DD} specified threshold, below which the device must be maintained under reset, is 1.8 V (see *Figure 7*). This supply voltage can drop to 1.7 V when the device operates in the 0 to 70 °C temperature range.

A comprehensive set of power-saving mode allows to design low-power applications.

When the internal reset is OFF, the following integrated features are no more supported:

- The integrated power-on reset (POR) / power-down reset (PDR) circuitry is disabled
- The brownout reset (BOR) circuitry is disabled
- The embedded programmable voltage detector (PVD) is disabled
- V_{BAT} functionality is no more available and V_{BAT} pin should be connected to V_{DD}

All packages, except for the LQFP64 and LQFP100, allow to disable the internal reset through the PDR_ON signal.

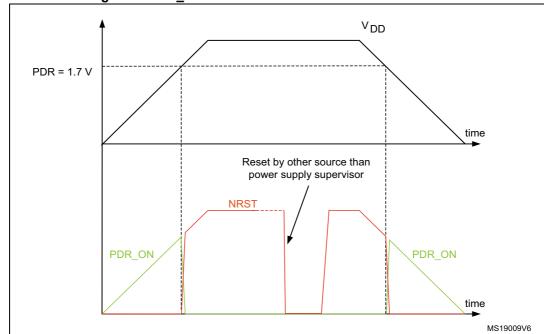


Figure 8. PDR_ON and NRST control with internal reset OFF

1. PDR = 1.7 V for reduce temperature range; PDR = 1.8 V for all temperature range.

2.2.16 Voltage regulator

The regulator has four operating modes:

- Regulator ON
 - Main regulator mode (MR)
 - Low-power regulator (LPR)
 - Power-down
- Regulator OFF

Regulator ON

On packages embedding the BYPASS_REG pin, the regulator is enabled by holding BYPASS_REG low. On all other packages, the regulator is always enabled.

There are three power modes configured by software when regulator is ON:

- MR is used in the nominal regulation mode (With different voltage scaling in Run)
 In Main regulator mode (MR mode), different voltage scaling are provided to reach the best compromise between maximum frequency and dynamic power consumption.
 Refer to Table 14: General operating conditions.
- LPR is used in the Stop modes
 The LP regulator mode is configured by software when entering Stop mode.
- Power-down is used in Standby mode.
 The Power-down mode is activated only when entering in Standby mode. The regulator output is in high impedance and the kernel circuitry is powered down, inducing zero consumption. The contents of the registers and SRAM are lost)

Two external ceramic capacitors should be connected on V_{CAP_1} & V_{CAP_2} pin. Refer to Figure 21: Power supply scheme and Figure 16: VCAP_1/VCAP_2 operating conditions.

All packages have regulator ON feature.

Regulator OFF

This feature is available only on packages featuring the BYPASS_REG pin. The regulator is disabled by holding BYPASS_REG high. The regulator OFF mode allows to supply externally a V_{12} voltage source through $V_{CAP\ 1}$ and $V_{CAP\ 2}$ pins.

Since the internal voltage scaling is not manage internally, the external voltage value must be aligned with the targeted maximum frequency. Refer to *Table 14: General operating conditions*.

The two 2.2 μ F ceramic capacitors should be replaced by two 100 nF decoupling capacitors.

Refer to Figure 21: Power supply scheme

When the regulator is OFF, there is no more internal monitoring on V_{12} . An external power supply supervisor should be used to monitor the V_{12} of the logic power domain. PA0 pin should be used for this purpose, and act as power-on reset on V_{12} power domain.

In regulator OFF mode the following features are no more supported:

- PA0 cannot be used as a GPIO pin since it allows to reset a part of the V₁₂ logic power domain which is not reset by the NRST pin.
- As long as PA0 is kept low, the debug mode cannot be used under power-on reset. As a consequence, PA0 and NRST pins must be managed separately if the debug connection under reset or pre-reset is required.
- The standby mode is not available

External V_{CAP_1/2} power supply supervisor Ext. reset controller active when V_{CAP_1/2} < Min V₁₂

PA0 NRST

DD PA0 NRST

V_{CAP_1}

V_{CAP_1}

V_{CAP_2}

ai18498V4

Figure 9. Regulator OFF

The following conditions must be respected:

- V_{DD} should always be higher than V_{CAP_1} and V_{CAP_2} to avoid current injection between power domains.
- If the time for V_{CAP_1} and V_{CAP_2} to reach V_{12} minimum value is faster than the time for V_{DD} to reach 1.8 V, then PA0 should be kept low to cover both conditions: until V_{CAP_1} and V_{CAP_2} reach V_{12} minimum value and until V_{DD} reaches 1.8 V (see *Figure 10*).
- Otherwise, if the time for V_{CAP_1} and V_{CAP_2} to reach V₁₂ minimum value is slower than the time for V_{DD} to reach 1.8 V, then PA0 could be asserted low externally (see Figure 11).
- If V_{CAP_1} and V_{CAP_2} go below V₁₂ minimum value and V_{DD} is higher than 1.8 V, then
 a reset must be asserted on PA0 pin.

Note: The minimum value of V₁₂ depends on the maximum frequency targeted in the application (see Table 14: General operating conditions).

PDR = 1.7 V or 1.8 V (2)

V_{CAP_1}/V_{CAP_2}

V_{CAP_1}/V_{CAP_2}

Win V₁₂

NRST

time

ai18491e

Figure 10. Startup in regulator OFF mode: slow V_{DD} slope - power-down reset risen after $V_{CAP\ 1}/V_{CAP\ 2}$ stabilization

- 1. This figure is valid both whatever the internal reset mode (ON or OFF).
- 2. PDR = 1.7 V for reduced temperature range; PDR = 1.8 V for all temperature ranges.

PDR = 1.7 V or 1.8 V (2)

V_{CAP_1}/V_{CAP_2}

V_{CAP_1}/V_{CAP_2}

Win V₁₂

NRST

PA0 asserted externally

Figure 11. Startup in regulator OFF mode: fast V_{DD} slope - power-down reset risen before $V_{CAP\ 1}/V_{CAP\ 2}$ stabilization

- 1. This figure is valid both whatever the internal reset mode (ON or OFF).
- 2. PDR = 1.7 V for a reduced temperature range; PDR = 1.8 V for all temperature ranges.

2.2.17 Regulator ON/OFF and internal reset ON/OFF availability

	Regulator ON	Regulator OFF	Internal reset ON	Internal reset OFF
LQFP64 LQFP100	Yes	No	Yes	No
LQFP144				Yes
WLCSP90 UFBGA176 LQFP176	Yes BYPASS_REG set to V _{SS}	Yes BYPASS_REG set to V _{DD}	Yes PDR_ON set to V _{DD}	PDR_ON connected to an external power supply supervisor

Table 3. Regulator ON/OFF and internal reset ON/OFF availability

2.2.18 Real-time clock (RTC), backup SRAM and backup registers

The backup domain of the STM32F415xx and STM32F417xx includes:

- The real-time clock (RTC)
- 4 Kbytes of backup SRAM
- 20 backup registers

The real-time clock (RTC) is an independent BCD timer/counter. Dedicated registers contain the second, minute, hour (in 12/24 hour), week day, date, month, year, in BCD (binary-coded decimal) format. Correction for 28, 29 (leap year), 30, and 31 day of the month are performed automatically. The RTC provides a programmable alarm and programmable periodic interrupts with wakeup from Stop and Standby modes. The sub-seconds value is also available in binary format.

It is clocked by a 32.768 kHz external crystal, resonator or oscillator, the internal low-power RC oscillator or the high-speed external clock divided by 128. The internal low-speed RC has a typical frequency of 32 kHz. The RTC can be calibrated using an external 512 Hz output to compensate for any natural quartz deviation.

Two alarm registers are used to generate an alarm at a specific time and calendar fields can be independently masked for alarm comparison. To generate a periodic interrupt, a 16-bit programmable binary auto-reload downcounter with programmable resolution is available and allows automatic wakeup and periodic alarms from every 120 µs to every 36 hours.

A 20-bit prescaler is used for the time base clock. It is by default configured to generate a time base of 1 second from a clock at 32.768 kHz.

The 4-Kbyte backup SRAM is an EEPROM-like memory area. It can be used to store data which need to be retained in V_{BAT} and standby mode. This memory area is disabled by default to minimize power consumption (see *Section 2.2.19: Low-power modes*). It can be enabled by software.

The backup registers are 32-bit registers used to store 80 bytes of user application data when V_{DD} power is not present. Backup registers are not reset by a system, a power reset, or when the device wakes up from the Standby mode (see Section 2.2.19: Low-power modes).

Additional 32-bit registers contain the programmable alarm subseconds, seconds, minutes, hours, day, and date.

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Like backup SRAM, the RTC and backup registers are supplied through a switch that is powered either from the V_{DD} supply when present or from the V_{BAT} pin.

2.2.19 Low-power modes

The STM32F415xx and STM32F417xx support three low-power modes to achieve the best compromise between low-power consumption, short startup time and available wakeup sources:

Sleep mode

In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs.

Stop mode

The Stop mode achieves the lowest power consumption while retaining the contents of SRAM and registers. All clocks in the V_{12} domain are stopped, the PLL, the HSI RC and the HSE crystal oscillators are disabled. The voltage regulator can also be put either in normal or in low-power mode.

The device can be woken up from the Stop mode by any of the EXTI line (the EXTI line source can be one of the 16 external lines, the PVD output, the RTC alarm / wakeup / tamper / time stamp events, the USB OTG FS/HS wakeup or the Ethernet wakeup).

Standby mode

The Standby mode is used to achieve the lowest power consumption. The internal voltage regulator is switched off so that the entire V_{12} domain is powered off. The PLL, the HSI RC and the HSE crystal oscillators are also switched off. After entering Standby mode, the SRAM and register contents are lost except for registers in the backup domain and the backup SRAM when selected.

The device exits the Standby mode when an external reset (NRST pin), an IWDG reset, a rising edge on the WKUP pin, or an RTC alarm / wakeup / tamper /time stamp event occurs.

The standby mode is not supported when the embedded voltage regulator is bypassed and the V_{12} domain is controlled by an external power.

2.2.20 V_{BAT} operation

The V_{BAT} pin allows to power the device V_{BAT} domain from an external battery, an external supercapacitor, or from V_{DD} when no external battery and an external supercapacitor are present.

V_{BAT} operation is activated when V_{DD} is not present.

The V_{BAT} pin supplies the RTC, the backup registers and the backup SRAM.

Note: When the microcontroller is supplied from V_{BAT} , external interrupts and RTC alarm/events do not exit it from V_{BAT} operation.

When PDR_ON pin is not connected to V_{DD} (internal reset OFF), the V_{BAT} functionality is no more available and V_{BAT} pin should be connected to V_{DD} .

2.2.21 Timers and watchdogs

The STM32F415xx and STM32F417xx devices include two advanced-control timers, eight general-purpose timers, two basic timers and two watchdog timers.

All timer counters can be frozen in debug mode.

Table 4 compares the features of the advanced-control, general-purpose and basic timers.

Table 4. Timer feature comparison

Table 4. Timer readure comparison										
Timer type	Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/ compare channels	Complementary output	Max interface clock (MHz)	Max timer clock (MHz)	
Advanced -control	TIM1, TIM8	16-bit	Up, Down, Up/down	Any integer between 1 and 65536	Yes	4	Yes	84	168	
General purpose T	TIM2, TIM5	32-bit	Up, Down, Up/down	Any integer between 1 and 65536	Yes	4	No	42	84	
	TIM3, TIM4	16-bit	Up, Down, Up/down	Any integer between 1 and 65536	Yes	4	No	42	84	
	TIM9	16-bit	Up	Any integer between 1 and 65536	No	2	No	84	168	
	TIM10 , TIM11	16-bit	Up	Any integer between 1 and 65536	No	1	No	84	168	
	TIM12	16-bit	Up	Any integer between 1 and 65536	No	2	No	42	84	
	TIM13 , TIM14	16-bit	Up	Any integer between 1 and 65536	No	1	No	42	84	
Basic	TIM6, TIM7	16-bit	Up	Any integer between 1 and 65536	Yes	0	No	42	84	

Advanced-control timers (TIM1, TIM8)

The advanced-control timers (TIM1, TIM8) can be seen as three-phase PWM generators multiplexed on 6 channels. They have complementary PWM outputs with programmable inserted dead times. They can also be considered as complete general-purpose timers. Their 4 independent channels can be used for:

- Input capture
- Output compare
- PWM generation (edge- or center-aligned modes)
- · One-pulse mode output

If configured as standard 16-bit timers, they have the same features as the general-purpose TIMx timers. If configured as 16-bit PWM generators, they have full modulation capability (0-100%).

The advanced-control timer can work together with the TIMx timers via the Timer Link feature for synchronization or event chaining.

TIM1 and TIM8 support independent DMA request generation.

General-purpose timers (TIMx)

There are ten synchronizable general-purpose timers embedded in the STM32F41xxx devices (see *Table 4* for differences).

• TIM2, TIM3, TIM4, TIM5

The STM32F41xxx include 4 full-featured general-purpose timers: TIM2, TIM5, TIM3, and TIM4. The TIM2 and TIM5 timers are based on a 32-bit auto-reload up/downcounter and a 16-bit prescaler. The TIM3 and TIM4 timers are based on a 16-bit auto-reload up/downcounter and a 16-bit prescaler. They all feature 4 independent channels for input capture/output compare, PWM or one-pulse mode output. This gives up to 16 input capture/output compare/PWMs on the largest packages.

The TIM2, TIM3, TIM4, TIM5 general-purpose timers can work together, or with the other general-purpose timers and the advanced-control timers TIM1 and TIM8 via the Timer Link feature for synchronization or event chaining.

Any of these general-purpose timers can be used to generate PWM outputs.

TIM2, TIM3, TIM4, TIM5 all have independent DMA request generation. They are capable of handling quadrature (incremental) encoder signals and the digital outputs from 1 to 4 hall-effect sensors.

TIM9, TIM10, TIM11, TIM12, TIM13, and TIM14

These timers are based on a 16-bit auto-reload upcounter and a 16-bit prescaler. TIM10, TIM11, TIM13, and TIM14 feature one independent channel, whereas TIM9 and TIM12 have two independent channels for input capture/output compare, PWM or one-pulse mode output. They can be synchronized with the TIM2, TIM3, TIM4, TIM5 full-featured general-purpose timers. They can also be used as simple time bases.

Basic timers TIM6 and TIM7

These timers are mainly used for DAC trigger and waveform generation. They can also be used as a generic 16-bit time base.

TIM6 and TIM7 support independent DMA request generation.

Independent watchdog

The independent watchdog is based on a 12-bit downcounter and 8-bit prescaler. It is clocked from an independent 32 kHz internal RC and as it operates independently from the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free-running timer for application timeout management. It is hardware- or software-configurable through the option bytes.

Window watchdog

The window watchdog is based on a 7-bit downcounter that can be set as free-running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from



the main clock. It has an early warning interrupt capability and the counter can be frozen in debug mode.

SysTick timer

This timer is dedicated to real-time operating systems, but could also be used as a standard downcounter. It features:

- A 24-bit downcounter
- Autoreload capability
- Maskable system interrupt generation when the counter reaches 0
- Programmable clock source.

2.2.22 Inter-integrated circuit interface (I²C)

Up to three I²C bus interfaces can operate in multimaster and slave modes. They can support the Standard-mode (up to 100 kHz) and Fast-mode (up to 400 kHz). They support the 7/10-bit addressing mode and the 7-bit dual addressing mode (as slave). A hardware CRC generation/verification is embedded.

They can be served by DMA and they support SMBus 2.0/PMBus.

2.2.23 Universal synchronous/asynchronous receiver transmitters (USART)

The STM32F415xx and STM32F417xx embed four universal synchronous/asynchronous receiver transmitters (USART1, USART2, USART3 and USART6) and two universal asynchronous receiver transmitters (UART4 and UART5).

These six interfaces provide asynchronous communication, IrDA SIR ENDEC support, multiprocessor communication mode, single-wire half-duplex communication mode and have LIN Master/Slave capability. The USART1 and USART6 interfaces are able to communicate at speeds of up to 10.5 Mbit/s. The other available interfaces communicate at up to 5.25 Mbit/s.

USART1, USART2, USART3 and USART6 also provide hardware management of the CTS and RTS signals, Smart Card mode (ISO 7816 compliant) and SPI-like communication capability. All interfaces can be served by the DMA controller.

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Max. baud rate Max. baud rate Modem **USART** Standard SPI **Smartcard** in Mbit/s in Mbit/s **APB** irDA (RTS/ LIN (ISO 7816) name features master (oversampling (oversampling mapping CTS) by 16) by 8) APB2 **USART1** Χ Χ Χ Х Х Χ 5.25 10.5 (max. 84 MHz) APB1 **USART2** Х Χ Х Х Х Х 2.62 5.25 (max. 42 MHz) APB1 USART3 Х Χ Х Х Х Χ 2.62 5.25 (max. 42 MHz) APB1 UART4 Х Χ Х 2.62 5.25 (max. 42 MHz) APB1 **UART5** Χ Χ Χ 2.62 5.25 (max. 42 MHz) APB2 **USART6** Χ Χ Χ Χ Χ Χ 5.25 10.5 (max. 84 MHz)

Table 5. USART feature comparison

2.2.24 Serial peripheral interface (SPI)

The STM32F41xxx feature up to three SPIs in slave and master modes in full-duplex and simplex communication modes. SPI1 can communicate at up to 42 Mbits/s, SPI2 and SPI3 can communicate at up to 21 Mbit/s. The 3-bit prescaler gives 8 master mode frequencies and the frame is configurable to 8 bits or 16 bits. The hardware CRC generation/verification supports basic SD Card/MMC modes. All SPIs can be served by the DMA controller.

The SPI interface can be configured to operate in TI mode for communications in master mode and slave mode.

2.2.25 Inter-integrated sound (I²S)

Two standard I²S interfaces (multiplexed with SPI2 and SPI3) are available. They can be operated in master or slave mode, in full duplex and half-duplex communication modes, and can be configured to operate with a 16-/32-bit resolution as an input or output channel. Audio sampling frequencies from 8 kHz up to 192 kHz are supported. When either or both of the I²S interfaces is/are configured in master mode, the master clock can be output to the external DAC/CODEC at 256 times the sampling frequency.

All I²Sx can be served by the DMA controller.

2.2.26 Audio PLL (PLLI2S)

The devices feature an additional dedicated PLL for audio I²S application. It allows to achieve error-free I²S sampling clock accuracy without compromising on the CPU performance, while using USB peripherals.

The PLLI2S configuration can be modified to manage an I²S sample rate change without disabling the main PLL (PLL) used for CPU, USB and Ethernet interfaces.

The audio PLL can be programmed with very low error to obtain sampling rates ranging from 8 KHz to 192 KHz.

In addition to the audio PLL, a master clock input pin can be used to synchronize the I²S flow with an external PLL (or Codec output).

2.2.27 Secure digital input/output interface (SDIO)

An SD/SDIO/MMC host interface is available, that supports MultiMediaCard System Specification Version 4.2 in three different databus modes: 1-bit (default), 4-bit and 8-bit.

The interface allows data transfer at up to 48 MHz, and is compliant with the SD Memory Card Specification Version 2.0.

The SDIO Card Specification Version 2.0 is also supported with two different databus modes: 1-bit (default) and 4-bit.

The current version supports only one SD/SDIO/MMC4.2 card at any one time and a stack of MMC4.1 or previous.

In addition to SD/SDIO/MMC, this interface is fully compliant with the CE-ATA digital protocol Rev1.1.

2.2.28 Ethernet MAC interface with dedicated DMA and IEEE 1588 support

Peripheral available only on the STM32F417xx devices.

The STM32F417xx devices provide an IEEE-802.3-2002-compliant media access controller (MAC) for ethernet LAN communications through an industry-standard medium-independent interface (MII) or a reduced medium-independent interface (RMII). The STM32F417xx requires an external physical interface device (PHY) to connect to the physical LAN bus (twisted-pair, fiber, etc.). the PHY is connected to the STM32F417xx MII port using 17 signals for MII or 9 signals for RMII, and can be clocked using the 25 MHz (MII) from the STM32F417xx.

The STM32F417xx includes the following features:

- Supports 10 and 100 Mbit/s rates
- Dedicated DMA controller allowing high-speed transfers between the dedicated SRAM and the descriptors (see the STM32F40xxx/41xxx reference manual for details)
- Tagged MAC frame support (VLAN support)
- Half-duplex (CSMA/CD) and full-duplex operation
- MAC control sublayer (control frames) support
- 32-bit CRC generation and removal
- Several address filtering modes for physical and multicast address (multicast and group addresses)
- 32-bit status code for each transmitted or received frame
- Internal FIFOs to buffer transmit and receive frames. The transmit FIFO and the receive FIFO are both 2 Kbytes.
- Supports hardware PTP (precision time protocol) in accordance with IEEE 1588 2008 (PTP V2) with the time stamp comparator connected to the TIM2 input
- Triggers interrupt when system time becomes greater than target time

2.2.29 Controller area network (bxCAN)

The two CANs are compliant with the 2.0A and B (active) specifications with a bitrate up to 1 Mbit/s. They can receive and transmit standard frames with 11-bit identifiers as well as extended frames with 29-bit identifiers. Each CAN has three transmit mailboxes, two receive FIFOS with 3 stages and 28 shared scalable filter banks (all of them can be used even if one CAN is used). 256 bytes of SRAM are allocated for each CAN.

2.2.30 Universal serial bus on-the-go full-speed (OTG_FS)

The STM32F415xx and STM32F417xx embed an USB OTG full-speed device/host/OTG peripheral with integrated transceivers. The USB OTG FS peripheral is compliant with the USB 2.0 specification and with the OTG 1.0 specification. It has software-configurable endpoint setting and supports suspend/resume. The USB OTG full-speed controller requires a dedicated 48 MHz clock that is generated by a PLL connected to the HSE oscillator. The major features are:

- Combined Rx and Tx FIFO size of 320 × 35 bits with dynamic FIFO sizing
- Supports the session request protocol (SRP) and host negotiation protocol (HNP)
- 4 bidirectional endpoints
- 8 host channels with periodic OUT support
- HNP/SNP/IP inside (no need for any external resistor)
- For OTG/Host modes, a power switch is needed in case bus-powered devices are connected

2.2.31 Universal serial bus on-the-go high-speed (OTG_HS)

The STM32F415xx and STM32F417xx devices embed a USB OTG high-speed (up to 480 Mb/s) device/host/OTG peripheral. The USB OTG HS supports both full-speed and high-speed operations. It integrates the transceivers for full-speed operation (12 MB/s) and features a UTMI low-pin interface (ULPI) for high-speed operation (480 MB/s). When using the USB OTG HS in HS mode, an external PHY device connected to the ULPI is required.



The USB OTG HS peripheral is compliant with the USB 2.0 specification and with the OTG 1.0 specification. It has software-configurable endpoint setting and supports suspend/resume. The USB OTG full-speed controller requires a dedicated 48 MHz clock that is generated by a PLL connected to the HSE oscillator.

The major features are:

- Combined Rx and Tx FIFO size of 1 Kbit × 35 with dynamic FIFO sizing
- Supports the session request protocol (SRP) and host negotiation protocol (HNP)
- 6 bidirectional endpoints
- 12 host channels with periodic OUT support
- Internal FS OTG PHY support
- External HS or HS OTG operation supporting ULPI in SDR mode. The OTG PHY is connected to the microcontroller ULPI port through 12 signals. It can be clocked using the 60 MHz output.
- Internal USB DMA
- HNP/SNP/IP inside (no need for any external resistor)
- for OTG/Host modes, a power switch is needed in case bus-powered devices are connected

2.2.32 Digital camera interface (DCMI)

The camera interface is *not* available in STM32F415xx devices.

STM32F417xx products embed a camera interface that can connect with camera modules and CMOS sensors through an 8-bit to 14-bit parallel interface, to receive video data. The camera interface can sustain a data transfer rate up to 54 Mbyte/s at 54 MHz. It features:

- Programmable polarity for the input pixel clock and synchronization signals
- Parallel data communication can be 8-, 10-, 12- or 14-bit
- Supports 8-bit progressive video monochrome or raw bayer format, YCbCr 4:2:2 progressive video, RGB 565 progressive video or compressed data (like JPEG)
- Supports continuous mode or snapshot (a single frame) mode
- Capability to automatically crop the image

2.2.33 Cryptographic acceleration

The STM32F415xx and STM32F417xx devices embed a cryptographic accelerator. This cryptographic accelerator provides a set of hardware acceleration for the advanced cryptographic algorithms usually needed to provide confidentiality, authentication, data integrity and non repudiation when exchanging messages with a peer.

These algorithms consists of:

Encryption/Decryption

- DES/TDES (data encryption standard/triple data encryption standard): ECB (electronic codebook) and CBC (cipher block chaining) chaining algorithms, 64-, 128- or 192-bit key
- AES (advanced encryption standard): ECB, CBC and CTR (counter mode) chaining algorithms, 128, 192 or 256-bit key

Universal hash

- SHA-1 (secure hash algorithm)
- MD5
- HMAC

The cryptographic accelerator supports DMA request generation.

2.2.34 Random number generator (RNG)

All STM32F415xx and STM32F417xx products embed an RNG that delivers 32-bit random numbers generated by an integrated analog circuit.

2.2.35 General-purpose input/outputs (GPIOs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain, with or without pull-up or pull-down), as input (floating, with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions. All GPIOs are high-current-capable and have speed selection to better manage internal noise, power consumption and electromagnetic emission.

The I/O configuration can be locked if needed by following a specific sequence in order to avoid spurious writing to the I/Os registers.

Fast I/O handling allowing maximum I/O toggling up to 84 MHz.

2.2.36 Analog-to-digital converters (ADCs)

Three 12-bit analog-to-digital converters are embedded and each ADC shares up to 16 external channels, performing conversions in the single-shot or scan mode. In scan mode, automatic conversion is performed on a selected group of analog inputs.

Additional logic functions embedded in the ADC interface allow:

- Simultaneous sample and hold
- Interleaved sample and hold

The ADC can be served by the DMA controller. An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all selected channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

To synchronize A/D conversion and timers, the ADCs could be triggered by any of TIM1, TIM2, TIM3, TIM4, TIM5, or TIM8 timer.

2.2.37 Temperature sensor

The temperature sensor has to generate a voltage that varies linearly with temperature. The conversion range is between 1.8 V and 3.6 V. The temperature sensor is internally

connected to the ADC1_IN16 input channel which is used to convert the sensor output voltage into a digital value.

As the offset of the temperature sensor varies from chip to chip due to process variation, the internal temperature sensor is mainly suitable for applications that detect temperature changes instead of absolute temperatures. If an accurate temperature reading is needed, then an external temperature sensor part should be used.

2.2.38 Digital-to-analog converter (DAC)

The two 12-bit buffered DAC channels can be used to convert two digital signals into two analog voltage signal outputs.

This dual digital Interface supports the following features:

- two DAC converters: one for each output channel
- 8-bit or 12-bit monotonic output
- left or right data alignment in 12-bit mode
- synchronized update capability
- noise-wave generation
- triangular-wave generation
- dual DAC channel independent or simultaneous conversions
- DMA capability for each channel
- external triggers for conversion
- input voltage reference V_{REF+}

Eight DAC trigger inputs are used in the device. The DAC channels are triggered through the timer update outputs that are also connected to different DMA streams.

2.2.39 Serial wire JTAG debug port (SWJ-DP)

The ARM SWJ-DP interface is embedded, and is a combined JTAG and serial wire debug port that enables either a serial wire debug or a JTAG probe to be connected to the target.

Debug is performed using 2 pins only instead of 5 required by the JTAG (JTAG pins could be re-use as GPIO with alternate function): the JTAG TMS and TCK pins are shared with SWDIO and SWCLK, respectively, and a specific sequence on the TMS pin is used to switch between JTAG-DP and SW-DP.

2.2.40 Embedded Trace Macrocell™

The ARM Embedded Trace Macrocell provides a greater visibility of the instruction and data flow inside the CPU core by streaming compressed data at a very high rate from the STM32F41xxx through a small number of ETM pins to an external hardware trace port analyser (TPA) device. The TPA is connected to a host computer using USB, Ethernet, or any other high-speed channel. Real-time instruction and data flow activity can be recorded and then formatted for display on the host computer that runs the debugger software. TPA hardware is commercially available from common development tool vendors.

The Embedded Trace Macrocell operates with third party debugger software tools.

33 PB12

ai18493b

Pinouts and pin description 3

PC13 d 47 UCAP_2 46 PA13 45 PA12 44 PA11 43 PA10 PC14 🖂 3 PH1 □ 42 PA9 NRST PA8 PC0 d 8 LQFP64 40 PC9 PC1 🗖 9 39 PC8 38 PC7 VSSA 12 37 PC6 36 PB15 VDDA | PA0_WKUP | 35 PB14 34 PB13

Figure 12. STM32F41xxx LQFP64 pinout

1. The above figure shows the package top view.

PA1

15 PA2

PE2口 1 75 D VDD PE3 2 PE4 3 74 □ vss 73 UCAP_2 PE4U 3 PE5U 4 PE6U 5 VBATU 6 PC13 U 7 PC14 U 8 PC15 U 9 72 PA13 71 PA12 70 PA 11 69 PA10 68 PA9 67 PA8 VSS [10 VDD [11 66 PC9 65 PC8 64 PC7 63 PC6 62 PD15 LQFP100 61 PD14 60 PD13 59 PD12 58 | PD11 VDD | 19 57 PD10 56 PD9 55 PD8 VSSA 20 VSSA ☐ 20 VREF+☐ 21 VDDA☐ 22 PA0 ☐ 23 54 PB15 53 PB14 PA1 24 PA2 25 52 PB13 51 PB12 ai18495c

Figure 13. STM32F41xxx LQFP100 pinout

1. The above figure shows the package top view.

8 PE2 | 1 $\vdash \mathsf{A}^{\mathsf{DD}}$ 108 107 | V_{SS} 106 | V_{CAP} 105 | PA13 PE3 🗖 2 PE4 🗆 3 PE5 🗆 4 PE6 4 5 104 🗖 PA 12 VBAT ☐ 6 103 PA 11 PC13 🗖 7 102 PA 10 101 PA9 100 PA8 PC14 🗖 8 PC15 □ 9 PF0 □ 10 PF1 □ 11 99 | PC9 98 🗖 PC8 PF2 🗖 12 97 🗖 PC7 PF3 🗖 13 96 □ PC6 95 þ v_{DD} PF4 🗖 14 PF5 d 15 94 | V_{SS} 93 | PG8 V_{SS} □ 16 V_{DD} □ 17 PF6 □ 18 92 | PG7 91 □ PG6 LQFP144 PF7 🗖 19 90 PG5 PF8 20 89 🗖 PG4 PF9 🗆 21 88 🗖 PG3 PF10 22 87 🗖 PG2 PH0 🗖 23 86 PD15 PH1 🗖 24 85 PD14 84 $\not\vdash$ V_{DD} NRST 25 83 \(\subset V_{SS}^{\(\)} \) PC0 ☐ 26 PC1 27 82 | PD13 PC2 ☐ 28 81 □PD12 PC3 ☐ 29 80 PD11 V_{DD} □ 30 V_{SSA} □ 31 79 PD10 78 PD9 V_{REF+}□ 32 77 | PD8 V_{DDA} 33 PA 0 34 76 □PB15 75 PB14 74 PB13 73 PB12 PA 1 🗆 35 PA2□ 36 ai18496b

Figure 14. STM32F41xxx LQFP144 pinout

1. The above figure shows the package top view.

176 D P17
175 D P16
177 D P16
177 D P16
177 D P17
170 D P17
171 D P17
170 D P17
170 D P17
170 D P18
160 D P18
160 D P18
160 D P18
160 D P18
161 D P18
162 D P18
163 D P18
174 D P18
175 D PE2 ☐ 1 PE3 ☐ 2 132 🗖 PI1 131 □ PI0 PE4 3 130 PH15 PE5 PH14 129 PE6 5 V_{BAT} ☐ 6 126 | V_{SS} 125 | V_{CAP_2} PC13 🗆 8 PC14 🗆 9 124 PA13 PC15 | 10 123 🗖 PA12 PI9 🗖 11 122 □ PA11 PI10 | 12 PI11 | 13 121 PA10 120 🗖 PA9 119 PA8 118 PC9 PC8 PF0 16 117 PF1 🗖 17 116 115 PC6 PF2 | 18 PF3 🗆 19 114 🗖 V_{DD} PF4 🗖 20 113 □ V_{SS} PF5 21 112 PG8 LQFP176 V_{SS} 22 111 □ PG7 V_{DD} ☐ 23 110 PG6 PF6 24 109 F PG5 108 PG4 107 PG3 PF7 25 PF8 26 106 PG2 105 PD15 PF9 27 PF10 28 PH0 🗆 29 104 PD14 PH1 30 NRST 31 103 □ V_{DD} 102 Vss PC0 = 32 PC1 = 33 □ PD13 100 PD12 99 | PD11 98 | PD10 PC2 34 PC3 35 97 PD9 96 PD8 V_{DD} □ 36 V_{SSA} □ 37 95 | PB15 94 | PB14 V_{REF+} □ 38 V_{DDA} □ 39 PA0 □ 40 94 93 | PB13 PA1 41 PA2 42 PH2 43 92 PB12 91 | V_{DD} 90 | V_{SS} 89 | PH12 PH3 44

Figure 15. STM32F41xxx LQFP176 pinout

1. The above figure shows the package top view.

MS19916V4

Figure 16. STM32F41xxx UFBGA176 ballout

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15
Α	PE3	PE2	PE1	PE0	PB8	PB5	PG14	PG13	PB4	PB3	PD7	PC12	PA15	PA14	PA13
В	PE4	PE5	PE6	PB9	PB7	PB6	PG15	PG12	PG11	PG10	PD6	PD0	PC11	PC10	PA12
С	VBAT	PI7	PI6	PI5	VDD	PDR_ON	VDD	VDD	VDD	PG9	PD5	PD1	PI3	PI2	PA11
D	PC13	PI8	PI9	PI4	VSS	воото	VSS	VSS	VSS	PD4	PD3	PD2	PH15	PI1	PA10
Е	PC14	PF0	PI10	PI11		•						PH13	PH14	PI0	PA9
F	PC15	VSS	VDD	PH2		VSS	VSS	VSS	VSS	VSS		VSS	VCAP_2	PC9	PA8
G	PH0	VSS	VDD	PH3		VSS	VSS	VSS	VSS	VSS		VSS	VDD	PC8	PC7
Н	PH1	PF2	PF1	PH4		VSS	VSS	VSS	VSS	VSS		vss	VDD	PG8	PC6
J	NRST	PF3	PF4	PH5		VSS	VSS	VSS	VSS	VSS		VDD	VDD	PG7	PG6
K	PF7	PF6	PF5	VDD		VSS	VSS	VSS	VSS	VSS		PH12	PG5	PG4	PG3
L	PF10	PF9	PF8	BYPASS_ REG								PH11	PH10	PD15	PG2
М	VSSA	PC0	PC1	PC2	PC3	PB2	PG1	VSS	VSS	VCAP_1	PH6	PH8	PH9	PD14	PD13
N	VREF-	PA 1	PA0	PA4	PC4	PF13	PG0	VDD	VDD	VDD	PE13	PH7	PD12	PD11	PD10
Р	VREF+	PA2	PA 6	PA5	PC5	PF12	PF15	PE8	PE9	PE11	PE14	PB12	PB13	PD9	PD8
R	VDDA	PA3	PA7	PB1	PB0	PF11	PF14	PE7	PE10	PE12	PE15	PB10	PB11	PB14	PB15

^{1.} This figure shows the package top view.

10 2 8 7 PA14 VBAT PC13 PDR_ON воото PB4 PD7 PD4 PC12 VDD PC14 PI1 VCAP_2 PC15 VDD PB3 PD6 PD2 PA15 PB7 В С PA0 VSS PB9 PB6 PD5 PD1 PC11 PI0 PA12 PA11 BYPASS REG PA13 PB8 PD0 PC10 PA10 PA8 PC3 Е PC0 VSS VDD VSS VDD PC9 PC8 PC7 VSS F PH0 PH1 VDD PE10 PE14 VCAP_1 PD14 PD15 PC6 PA1 G NRST VDDA PB0 PE7 PE13 PE15 PD10 PD12 PD11 Н VSSA PA6 PB1 PE12 PB10 PD9 PB15 PA7 PB2 PE9 PE11 PB11 PB12 PB14 PB13 MS30402V1

Figure 17. STM32F41xxx WLCSP90 ballout

1. This figure shows the package bump view.

Table 6. Legend/abbreviations used in the pinout table

Name	Abbreviation	Definition						
Pin name		specified in brackets below the pin name, the pin function during and after as the actual pin name						
	S	Supply pin						
Pin type	I	Input only pin						
	I/O	Input / output pin						
	FT	5 V tolerant I/O						
I/O structure	TTa	3.3 V tolerant I/O directly connected to ADC						
i/O structure	В	Dedicated BOOT0 pin						
	RST	Bidirectional reset pin with embedded weak pull-up resistor						
Notes	Unless otherwise specified by a note, all I/Os are set as floating inputs during and after res							
Alternate functions	Functions selected through GPIOx_AFR registers							
Additional functions	Functions directly selected/enabled through peripheral registers							

Table 7. STM32F41xxx pin and ball definitions

	F	Pin r	numb	er		7. 01111021		•			
LQFP64	WLCSP90	LQFP100	LQFP144	UFBGA176	LQFP176	Pin name (function after reset) ⁽¹⁾	Pin type	I / O structure	Notes	Alternate functions	Additional functions
-	-	1	1	A2	1	PE2	I/O	FT	-	TRACECLK/ FSMC_A23 / ETH_MII_TXD3 / EVENTOUT	-
-	-	2	2	A1	2	PE3	I/O	FT	-	TRACED0/FSMC_A19 / EVENTOUT	-
-	-	3	3	B1	3	PE4	I/O	FT	-	TRACED1/FSMC_A20 / DCMI_D4/ EVENTOUT	-
-	-	4	4	B2	4	PE5	I/O	FT	-	TRACED2 / FSMC_A21 / TIM9_CH1 / DCMI_D6 / EVENTOUT	-
-	-	5	5	ВЗ	5	PE6	I/O	FT	-	TRACED3 / FSMC_A22 / TIM9_CH2 / DCMI_D7 / EVENTOUT	-
1	A10	6	6	C1	6	V _{BAT}	S	-	-	-	-
-	-	1	1	D2	7	PI8	I/O	FT	(2)(3)	EVENTOUT	RTC_TAMP1, RTC_TAMP2, RTC_TS
2	A9	7	7	D1	8	PC13	I/O	FT	(2) (3)	EVENTOUT	RTC_OUT, RTC_TAMP1, RTC_TS
3	B10	8	8	E1	9	PC14/OSC32_IN (PC14)	I/O	FT	(2)(3)	EVENTOUT	OSC32_IN ⁽⁴⁾
4	В9	9	9	F1	10	PC15/ OSC32_OUT (PC15)	I/O	FT	(2)(3)	EVENTOUT	OSC32_OUT ⁽⁴⁾
-	-	-	-	D3	11	PI9	I/O	FT	-	CAN1_RX / EVENTOUT	-
-	-	1	-	E3	12	PI10	I/O	FT	-	ETH_MII_RX_ER / EVENTOUT	-
-	-	1	1	E4	13	PI11	I/O	FT	-	OTG_HS_ULPI_DIR / EVENTOUT	-
-	-	-	-	F2	14	V _{SS}	S	-	-	-	-
-	-	1	ı	F3	15	V_{DD}	S	-	-	-	-
-	-	-	10	E2	16	PF0	I/O	FT	-	FSMC_A0 / I2C2_SDA / EVENTOUT	-

Table 7. STM32F41xxx pin and ball definitions (continued)

	F	Pin r	numb							definitions (continued)	
LQFP64	WLCSP90	LQFP100	LQFP144	UFBGA176	LQFP176	Pin name (function after reset) ⁽¹⁾	Pin type	I / O structure	Notes	Alternate functions	Additional functions
-	-	-	11	НЗ	17	PF1	I/O	FT	-	FSMC_A1 / I2C2_SCL / EVENTOUT	-
-	-	-	12	H2	18	PF2	I/O	FT	-	FSMC_A2 / I2C2_SMBA / EVENTOUT	-
-	-	-	13	J2	19	PF3	I/O	FT	(4)	FSMC_A3/EVENTOUT	ADC3_IN9
-	-	-	14	J3	20	PF4	I/O	FT	(4)	FSMC_A4/EVENTOUT	ADC3_IN14
-	-	-	15	K3	21	PF5	I/O	FT	(4)	FSMC_A5/EVENTOUT	ADC3_IN15
-	C9	10	16	G2	22	V _{SS}	S	-	-	-	-
-	B8	11	17	G3	23	V_{DD}	S	-	-	-	-
-	-	-	18	K2	24	PF6	I/O	FT	(4)	TIM10_CH1 / FSMC_NIORD/ EVENTOUT	ADC3_IN4
-	-	-	19	K1	25	PF7	I/O	FT	(4)	TIM11_CH1/FSMC_NREG/ EVENTOUT	ADC3_IN5
-	-	-	20	L3	26	PF8	I/O	FT	(4)	TIM13_CH1 / FSMC_NIOWR/ EVENTOUT	ADC3_IN6
-	-	-	21	L2	27	PF9	I/O	FT	(4)	TIM14_CH1 / FSMC_CD/ EVENTOUT	ADC3_IN7
-	-	-	22	L1	28	PF10	I/O	FT	(4)	FSMC_INTR/ EVENTOUT	ADC3_IN8
5	F10	12	23	G1	29	PH0/OSC_IN (PH0)	I/O	FT	-	EVENTOUT	OSC_IN ⁽⁴⁾
6	F9	13	24	H1	30	PH1/OSC_OUT (PH1)	I/O	FT	-	EVENTOUT	OSC_OUT ⁽⁴⁾
7	G10	14	25	J1	31	NRST	I/O	RST	-	-	-
8	E10	15	26	M2	32	PC0	I/O	FT	(4)	OTG_HS_ULPI_STP/ EVENTOUT	ADC123_IN10
9	-	16	27	МЗ	33	PC1	I/O	FT	(4)	ETH_MDC/ EVENTOUT	ADC123_IN11
10	D10	17	28	M4	34	PC2	I/O	FT	(4)	SPI2_MISO / OTG_HS_ULPI_DIR / ETH_MII_TXD2 /I2S2ext_SD/ EVENTOUT	ADC123_IN12

Table 7. STM32F41xxx pin and ball definitions (continued)

	ı	Pin r	numb	er							
LQFP64	WLCSP90	LQFP100	LQFP144	UFBGA176	LQFP176	Pin name (function after reset) ⁽¹⁾	Pin type	I / O structure	Notes	Alternate functions	Additional functions
11	E9	18	29	M5	35	PC3	I/O	FT	(4)	SPI2_MOSI / I2S2_SD / OTG_HS_ULPI_NXT / ETH_MII_TX_CLK/ EVENTOUT	ADC123_IN13
-	·	19	30	ı	36	V_{DD}	S	-	-	-	-
12	H10	20	31	M1	37	V _{SSA}	S	ı	-	-	-
-	-	-	-	N1	-	V_{REF-}	S	ı	-	-	-
-	-	21	32	P1	38	V _{REF+}	S	-	-	-	-
13	G9	22	33	R1	39	V_{DDA}	S	-	-	-	-
14	C10	23	34	N3	40	PA0/WKUP (PA0)	I/O	FT	(5)	USART2_CTS/ UART4_TX/ ETH_MII_CRS / TIM2_CH1_ETR/ TIM5_CH1 / TIM8_ETR/ EVENTOUT	ADC123_IN0/WKU P ⁽⁴⁾
15	F8	24	35	N2	41	PA1	I/O	FT	(4)	USART2_RTS / UART4_RX/ ETH_RMII_REF_CLK / ETH_MII_RX_CLK / TIM5_CH2 / TIM2_CH2/ EVENTOUT	ADC123_IN1
16	J10	25	36	P2	42	PA2	I/O	FT	(4)	USART2_TX/TIM5_CH3 / TIM9_CH1 / TIM2_CH3 / ETH_MDIO/ EVENTOUT	ADC123_IN2
-	-	-	ı	F4	43	PH2	I/O	FT	-	ETH_MII_CRS/EVENTOUT	-
-	-	-	1	G4	44	PH3	I/O	FT	-	ETH_MII_COL/EVENTOUT	-
-	-	-	1	H4	45	PH4	I/O	FT	-	I2C2_SCL / OTG_HS_ULPI_NXT/ EVENTOUT	-
-	-	-	1	J4	46	PH5	I/O	FT	-	I2C2_SDA/ EVENTOUT	-
17	H9	26	37	R2	47	PA3	I/O	FT	(4)	USART2_RX/TIM5_CH4 / TIM9_CH2 / TIM2_CH4 / OTG_HS_ULPI_D0 / ETH_MII_COL/ EVENTOUT	ADC123_IN3
18	E5	27	38	-	-	V_{SS}	S	-	-	-	-



Table 7. STM32F41xxx pin and ball definitions (continued)

	ı	Pin r	numb				İ			deminions (continued)	
LQFP64	WLCSP90	LQFP100	LQFP144	UFBGA176	LQFP176	Pin name (function after reset) ⁽¹⁾	Pin type	I / O structure	Notes	Alternate functions	Additional functions
	D9			L4	48	BYPASS_REG	I	FT	-	-	-
19	E4	28	39	K4	49	V_{DD}	S	-	-	-	-
20	J9	29	40	N4	50	PA4	I/O	ТТа	(4)	SPI1_NSS / SPI3_NSS / USART2_CK / DCMI_HSYNC / OTG_HS_SOF/ I2S3_WS/ EVENTOUT	ADC12_IN4 /DAC_OUT1
21	G8	30	41	P4	51	PA5	I/O	ТТа	(4)	SPI1_SCK/ OTG_HS_ULPI_CK / TIM2_CH1_ETR/ TIM8_CH1N/ EVENTOUT	ADC12_IN5/DAC_ OUT2
22	Н8	31	42	P3	52	PA6	I/O	FT	(4)	SPI1_MISO / TIM8_BKIN/TIM13_CH1 / DCMI_PIXCLK / TIM3_CH1 / TIM1_BKIN/ EVENTOUT	ADC12_IN6
23	J8	32	43	R3	53	PA7	I/O	FT	(4)	SPI1_MOSI/TIM8_CH1N / TIM14_CH1/TIM3_CH2/ ETH_MII_RX_DV / TIM1_CH1N / ETH_RMII_CRS_DV/ EVENTOUT	ADC12_IN7
24	-	33	44	N5	54	PC4	I/O	FT	(4)	ETH_RMII_RX_D0 / ETH_MII_RX_D0/ EVENTOUT	ADC12_IN14
25	-	34	45	P5	55	PC5	I/O	FT	(4)	ETH_RMII_RX_D1 / ETH_MII_RX_D1/ EVENTOUT	ADC12_IN15
26	G7	35	46	R5	56	PB0	I/O	FT	(4)	TIM3_CH3 / TIM8_CH2N/ OTG_HS_ULPI_D1/ ETH_MII_RXD2 / TIM1_CH2N/ EVENTOUT	ADC12_IN8
27	H7	36	47	R4	57	PB1	I/O	FT	(4)	TIM3_CH4 / TIM8_CH3N/ OTG_HS_ULPI_D2/ ETH_MII_RXD3 / TIM1_CH3N/ EVENTOUT	ADC12_IN9
28	J7	37	48	M6	58	PB2/BOOT1 (PB2)	I/O	FT	-	EVENTOUT	-

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Table 7. STM32F41xxx pin and ball definitions (continued)

	l	Pin r	numb	er							
LQFP64	WLCSP90	LQFP100	LQFP144	UFBGA176	LQFP176	Pin name (function after reset) ⁽¹⁾	Pin type	I / O structure	Notes	Alternate functions	Additional functions
-	-	-	49	R6	59	PF11	I/O	FT	-	DCMI_D12/ EVENTOUT	-
-	-	-	50	P6	60	PF12	I/O	FT	-	FSMC_A6/ EVENTOUT	-
-	ı	-	51	M8	61	V _{SS}	S	ı	-	-	-
-	ı	-	52	N8	62	V_{DD}	S	ı	-	-	-
-	ı	-	53	N6	63	PF13	I/O	FT	-	FSMC_A7/ EVENTOUT	-
-	ı	-	54	R7	64	PF14	I/O	FT	-	FSMC_A8/ EVENTOUT	-
-	ı	-	55	P7	65	PF15	I/O	FT	-	FSMC_A9/ EVENTOUT	-
-	ı	-	56	N7	66	PG0	I/O	FT	-	FSMC_A10/ EVENTOUT	-
-	1	-	57	M7	67	PG1	I/O	FT	-	FSMC_A11/ EVENTOUT	-
-	G6	38	58	R8	68	PE7	I/O	FT	-	FSMC_D4/TIM1_ETR/ EVENTOUT	-
-	H6	39	59	P8	69	PE8	I/O	FT	1	FSMC_D5/ TIM1_CH1N/ EVENTOUT	-
-	J6	40	60	P9	70	PE9	I/O	FT	-	FSMC_D6/TIM1_CH1/ EVENTOUT	-
-	-	-	61	M9	71	V _{SS}	S	-	-	-	-
-	1	-	62	N9	72	V_{DD}	S	-	-	-	-
-	F6	41	63	R9	73	PE10	I/O	FT	-	FSMC_D7/TIM1_CH2N/ EVENTOUT	-
-	J5	42	64	P10	74	PE11	I/O	FT	-	FSMC_D8/TIM1_CH2/ EVENTOUT	-
-	H5	43	65	R10	75	PE12	I/O	FT	-	FSMC_D9/TIM1_CH3N/ EVENTOUT	-
-	G5	44	66	N11	76	PE13	I/O	FT	-	FSMC_D10/TIM1_CH3/ EVENTOUT	-
-	F5	45	67	P11	77	PE14	I/O	FT	-	FSMC_D11/TIM1_CH4/ EVENTOUT	-
-	G4	46	68	R11	78	PE15	I/O	FT	-	FSMC_D12/TIM1_BKIN/ EVENTOUT	-

Table 7. STM32F41xxx pin and ball definitions (continued)

	ı	Pin r	numb		<u>ubio</u>	1101111021 4177	, p	una i		definitions (continued)	
LQFP64	WLCSP90	LQFP100	LQFP144	UFBGA176	LQFP176	Pin name (function after reset) ⁽¹⁾	Pin type	I / O structure	Notes	Alternate functions	Additional functions
29	H4	47	69	R12	79	PB10	I/O	FT	-	SPI2_SCK / I2S2_CK / I2C2_SCL/ USART3_TX / OTG_HS_ULPI_D3 / ETH_MII_RX_ER / TIM2_CH3/ EVENTOUT	-
30	J4	48	70	R13	80	PB11	I/O	FT	-	I2C2_SDA/USART3_RX/ OTG_HS_ULPI_D4 / ETH_RMII_TX_EN/ ETH_MII_TX_EN / TIM2_CH4/ EVENTOUT	-
31	F4	49	71	M10	81	V _{CAP_1}	S		-	-	-
32	ı	50	72	N10	82	V_{DD}	S		-	-	-
-	-	-	-	M11	83	PH6	I/O	FT	-	I2C2_SMBA / TIM12_CH1 / ETH_MII_RXD2/ EVENTOUT	-
-	-	-	-	N12	84	PH7	I/O	FT	-	I2C3_SCL / ETH_MII_RXD3/ EVENTOUT	-
-	-	-	ı	M12	85	PH8	I/O	FT	-	I2C3_SDA / DCMI_HSYNC/ EVENTOUT	-
-	ı	-	1	M13	86	PH9	I/O	FT	-	I2C3_SMBA / TIM12_CH2/ DCMI_D0/ EVENTOUT	-
-	-	-	ı	L13	87	PH10	I/O	FT	-	TIM5_CH1 / DCMI_D1/ EVENTOUT	-
-	-	-	-	L12	88	PH11	I/O	FT	-	TIM5_CH2 / DCMI_D2/ EVENTOUT	-
-	-	-	ı	K12	89	PH12	I/O	FT	-	TIM5_CH3 / DCMI_D3/ EVENTOUT	-
	-	-	ı	H12	90	V _{SS}	S	-	-	-	-
-	-	-	-	J12	91	V_{DD}	S	-	-	-	-

Table 7. STM32F41xxx pin and ball definitions (continued)

	ı	Pin r	numb	er							
LQFP64	WLCSP90	LQFP100	LQFP144	UFBGA176	LQFP176	Pin name (function after reset) ⁽¹⁾	Pin type	I / O structure	Notes	Alternate functions	Additional functions
33	J3	51	73	P12	92	PB12	I/O	FT	1	SPI2_NSS / I2S2_WS / I2C2_SMBA/ USART3_CK/TIM1_BKIN / CAN2_RX / OTG_HS_ULPI_D5/ ETH_RMII_TXD0 / ETH_MII_TXD0/ OTG_HS_ID/ EVENTOUT	-
34	J1	52	74	P13	93	PB13	I/O	FT	-	SPI2_SCK / I2S2_CK / USART3_CTS/ TIM1_CH1N /CAN2_TX / OTG_HS_ULPI_D6 / ETH_RMII_TXD1 / ETH_MII_TXD1/ EVENTOUT	OTG_HS_VBUS
35	J2	53	75	R14	94	PB14	I/O	FT	1	SPI2_MISO/ TIM1_CH2N / TIM12_CH1 / OTG_HS_DM/ USART3_RTS / TIM8_CH2N/I2S2ext_SD/ EVENTOUT	-
36	H1	54	76	R15	95	PB15	I/O	FT	-	SPI2_MOSI / I2S2_SD/ TIM1_CH3N / TIM8_CH3N / TIM12_CH2 / OTG_HS_DP/ EVENTOUT	RTC_REFIN
-	H2	55	77	P15	96	PD8	I/O	FT	-	FSMC_D13/USART3_TX/ EVENTOUT	-
-	НЗ	56	78	P14	97	PD9	I/O	FT	-	FSMC_D14/USART3_RX/ EVENTOUT	-
-	G3	57	79	N15	98	PD10	I/O	FT	-	FSMC_D15/USART3_CK/ EVENTOUT	-
-	G1	58	80	N14	99	PD11	I/O	FT	-	FSMC_CLE / FSMC_A16/USART3_CTS/ EVENTOUT	-
-	G2	59	81	N13	100	PD12	I/O	FT	-	FSMC_ALE/ FSMC_A17/TIM4_CH1 / USART3_RTS/ EVENTOUT	-

Table 7. STM32F41xxx pin and ball definitions (continued)

	ı	Pin r	numb							definitions (continued)	
LQFP64	WLCSP90	LQFP100	LQFP144	UFBGA176	LQFP176	Pin name (function after reset) ⁽¹⁾	Pin type	I / O structure	Notes	Alternate functions	Additional functions
-	-	60	82	M15	101	PD13	I/O	FT	-	FSMC_A18/TIM4_CH2/ EVENTOUT	-
-	-	-	83	-	102	V _{SS}	S		-	-	-
-	ı	-	84	J13	103	V_{DD}	S		-	-	-
-	F2	61	85	M14	104	PD14	I/O	FT	-	FSMC_D0/TIM4_CH3/ EVENTOUT/ EVENTOUT	-
-	F1	62	86	L14	105	PD15	I/O	FT	-	FSMC_D1/TIM4_CH4/ EVENTOUT	-
-	-	-	87	L15	106	PG2	I/O	FT	-	FSMC_A12/ EVENTOUT	-
-	-	-	88	K15	107	PG3	I/O	FT	-	FSMC_A13/ EVENTOUT	-
-	ı	-	89	K14	108	PG4	I/O	FT	-	FSMC_A14/ EVENTOUT	-
-	-	-	90	K13	109	PG5	I/O	FT	-	FSMC_A15/ EVENTOUT	-
-	-	-	91	J15	110	PG6	I/O	FT	-	FSMC_INT2/ EVENTOUT	-
-	ı	-	92	J14	111	PG7	I/O	FT	-	FSMC_INT3/USART6_CK/ EVENTOUT	-
-	-	-	93	H14	112	PG8	I/O	FT	-	USART6_RTS / ETH_PPS_OUT/ EVENTOUT	-
-	-	-	94	G12	113	V _{SS}	S		-	-	-
-	-	-	95	H13	114	V _{DD}	S		-	-	-
37	F3	63	96	H15	115	PC6	I/O	FT	-	I2S2_MCK / TIM8_CH1/SDIO_D6 / USART6_TX / DCMI_D0/TIM3_CH1/ EVENTOUT	-
38	E1	64	97	G15	116	PC7	I/O	FT	-	I2S3_MCK / TIM8_CH2/SDIO_D7 / USART6_RX / DCMI_D1/TIM3_CH2/ EVENTOUT	-
39	E2	65	98	G14	117	PC8	I/O	FT	-	TIM8_CH3/SDIO_D0 /TIM3_CH3/ USART6_CK / DCMI_D2/ EVENTOUT	-

Table 7. STM32F41xxx pin and ball definitions (continued)

	ı	Pin r	numb	er							
LQFP64	WLCSP90	LQFP100	LQFP144	UFBGA176	LQFP176	Pin name (function after reset) ⁽¹⁾	Pin type	I / O structure	Notes	Alternate functions	Additional functions
40	E3	66	99	F14	118	PC9	I/O	FT	1	I2S_CKIN/ MCO2 / TIM8_CH4/SDIO_D1 / /I2C3_SDA / DCMI_D3 / TIM3_CH4/ EVENTOUT	-
41	D1	67	100	F15	119	PA8	I/O	FT	-	MCO1 / USART1_CK/ TIM1_CH1/ I2C3_SCL/ OTG_FS_SOF/ EVENTOUT	-
42	D2	68	101	E15	120	PA9	I/O	FT	-	USART1_TX/ TIM1_CH2 / I2C3_SMBA / DCMI_D0/ EVENTOUT	OTG_FS_VBUS
43	D3	69	102	D15	121	PA10	I/O	FT	-	USART1_RX/ TIM1_CH3/ OTG_FS_ID/DCMI_D1/ EVENTOUT	-
44	C1	70	103	C15	122	PA11	I/O	FT	-	USART1_CTS / CAN1_RX / TIM1_CH4 / OTG_FS_DM/ EVENTOUT	-
45	C2	71	104	B15	123	PA12	I/O	FT	-	USART1_RTS / CAN1_TX/ TIM1_ETR/ OTG_FS_DP/ EVENTOUT	-
46	D4	72	105	A15	124	PA13 (JTMS-SWDIO)	I/O	FT	ı	JTMS-SWDIO/ EVENTOUT	-
47	B1	73	106	F13	125	V _{CAP_2}	S	-	-	-	-
-	E7	74	107	F12	126	V _{SS}	S	-	-	-	-
48	E6	75	108	G13	127	V_{DD}	S	-	-	-	-
-	1	-	-	E12	128	PH13	I/O	FT	1	TIM8_CH1N / CAN1_TX/ EVENTOUT	-
-	1	-	-	E13	129	PH14	I/O	FT	1	TIM8_CH2N / DCMI_D4/ EVENTOUT	-
-	-	-	-	D13	130	PH15	I/O	FT	-	TIM8_CH3N / DCMI_D11/ EVENTOUT	-
-	C3	-	1	E14	131	PI0	I/O	FT	-	TIM5_CH4 / SPI2_NSS / I2S2_WS / DCMI_D13/ EVENTOUT	-
-	B2	-	-	D14	132	PI1	I/O	FT	-	SPI2_SCK / I2S2_CK / DCMI_D8/ EVENTOUT	-



Table 7. STM32F41xxx pin and ball definitions (continued)

	ı	Pin r	numb							definitions (continued)	
LQFP64	WLCSP90	LQFP100	LQFP144	UFBGA176	LQFP176	Pin name (function after reset) ⁽¹⁾	Pin type	I / O structure	Notes	Alternate functions	Additional functions
-	-	-	-	C14	133	Pl2	I/O	FT	-	TIM8_CH4 /SPI2_MISO / DCMI_D9 / I2S2ext_SD/ EVENTOUT	-
-	-	-	ı	C13	134	PI3	I/O	FT		TIM8_ETR / SPI2_MOSI / I2S2_SD / DCMI_D10/ EVENTOUT	-
-	-	-	-	D9	135	V_{SS}	S	ı	-	-	-
-	-	-	-	C9	136	V_{DD}	S	-	-	-	-
49	A2	76	109	A14	137	PA14 (JTCK/SWCLK)	I/O	FT	-	JTCK-SWCLK/ EVENTOUT	-
50	В3	77	110	A13	138	PA15 (JTDI)	1/0	FT	-	JTDI/ SPI3_NSS/ I2S3_WS/TIM2_CH1_ETR / SPI1_NSS / EVENTOUT	-
51	D5	78	111	B14	139	PC10	I/O	FT	-	SPI3_SCK / I2S3_CK/ UART4_TX/SDIO_D2 / DCMI_D8 / USART3_TX/ EVENTOUT	-
52	C4	79	112	B13	140	PC11	I/O	FT	-	UART4_RX/ SPI3_MISO / SDIO_D3 / DCMI_D4/USART3_RX / I2S3ext_SD/ EVENTOUT	-
53	А3	80	113	A12	141	PC12	I/O	FT	-	UART5_TX/SDIO_CK / DCMI_D9 / SPI3_MOSI /I2S3_SD / USART3_CK/ EVENTOUT	-
-	D6	81	114	B12	142	PD0	I/O	FT	-	FSMC_D2/CAN1_RX/ EVENTOUT	-
-	C5	82	115	C12	143	PD1	I/O	FT	-	FSMC_D3 / CAN1_TX/ EVENTOUT	-
54	B4	83	116	D12	144	PD2	I/O	FT	-	TIM3_ETR/UART5_RX/ SDIO_CMD / DCMI_D11/ EVENTOUT	-
-	-	84	117	D11	145	PD3	I/O	FT	-	FSMC_CLK/ USART2_CTS/ EVENTOUT	-

Table 7. STM32F41xxx pin and ball definitions (continued)

	I	Pin r	numb	er						,	
LQFP64	WLCSP90	LQFP100	LQFP144	UFBGA176	LQFP176	Pin name (function after reset) ⁽¹⁾	Pin type	I / O structure	Notes	Alternate functions	Additional functions
-	A4	85	118	D10	146	PD4	I/O	FT	-	FSMC_NOE/ USART2_RTS/ EVENTOUT	-
-	C6	86	119	C11	147	PD5	I/O	FT	-	FSMC_NWE/USART2_TX/ EVENTOUT	-
-	-	-	120	D8	148	V_{SS}	S	-	-	-	-
-	-	-	121	C8	149	V_{DD}	S	-	-	-	-
-	B5	87	122	B11	150	PD6	I/O	FT	-	FSMC_NWAIT/ USART2_RX/ EVENTOUT	-
-	A5	88	123	A11	151	PD7	I/O	FT	-	USART2_CK/FSMC_NE1/ FSMC_NCE2/ EVENTOUT	-
-	-	-	124	C10	152	PG9	I/O	FT	-	USART6_RX / FSMC_NE2/FSMC_NCE3/ EVENTOUT	-
-	-	-	125	B10	153	PG10	I/O	FT	-	FSMC_NCE4_1/ FSMC_NE3/ EVENTOUT	-
-	-	-	126	В9	154	PG11	I/O	FT	-	FSMC_NCE4_2 / ETH_MII_TX_EN/ ETH_RMII_TX_EN/ EVENTOUT	-
-	-	-	127	В8	155	PG12	I/O	FT	-	FSMC_NE4 / USART6_RTS/ EVENTOUT	-
-	-	-	128	A8	156	PG13	I/O	FT	-	FSMC_A24 / USART6_CTS /ETH_MII_TXD0/ ETH_RMII_TXD0/ EVENTOUT	-
-	-	-	129	A7	157	PG14	I/O	FT	-	FSMC_A25 / USART6_TX /ETH_MII_TXD1/ ETH_RMII_TXD1/ EVENTOUT	-
-	E8	-	130	D7	158	3 V _{SS}		-	-	-	-
-	F7	-	131	C7	159	V_{DD}	S	-	-	-	-
-	-	-	132	В7	160	PG15	I/O	FT	-	USART6_CTS / DCMI_D13/ EVENTOUT	-

Table 7. STM32F41xxx pin and ball definitions (continued)

	ı	Pin r	numb							deminions (continueu)	
LQFP64	WLCSP90	LQFP100	LQFP144	UFBGA176	LQFP176	Pin name (function after reset) ⁽¹⁾	Pin type	I / O structure	Notes	Alternate functions	Additional functions
55	В6	89	133	A10	161	PB3 (JTDO/ TRACESWO)	I/O	FT	-	JTDO/ TRACESWO/ SPI3_SCK / I2S3_CK / TIM2_CH2 / SPI1_SCK/ EVENTOUT	-
56	A6	90	134	A9	162	PB4 (NJTRST)	I/O	FT	-	NJTRST/ SPI3_MISO / TIM3_CH1 / SPI1_MISO / I2S3ext_SD/ EVENTOUT	-
57	D7	91	135	A6	163	PB5	I/O	FT	-	I2C1_SMBA/ CAN2_RX / OTG_HS_ULPI_D7 / ETH_PPS_OUT/TIM3_CH2 / SPI1_MOSI/ SPI3_MOSI / DCMI_D10 / I2S3_SD/ EVENTOUT	-
58	C7	92	136	В6	164	PB6	I/O	FT	-	I2C1_SCL/ TIM4_CH1 / CAN2_TX / DCMI_D5/USART1_TX/ EVENTOUT	-
59	В7	93	137	В5	165	PB7	I/O	FT	-	I2C1_SDA / FSMC_NL / DCMI_VSYNC / USART1_RX/ TIM4_CH2/ EVENTOUT	-
60	A7	94	138	D6	166	BOOT0	I	В	-	-	V _{PP}
61	D8	95	139	A5	167	PB8	I/O	FT	-	TIM4_CH3/SDIO_D4/ TIM10_CH1 / DCMI_D6 / ETH_MII_TXD3 / I2C1_SCL/ CAN1_RX/ EVENTOUT	-
62	C8	96	140	B4	168	PB9	I/O	FT	-	SPI2_NSS/ I2S2_WS / TIM4_CH4/ TIM11_CH1/ SDIO_D5 / DCMI_D7 / I2C1_SDA / CAN1_TX/ EVENTOUT	-
-	-	97	141	A4	169	PE0	I/O	FT	-	TIM4_ETR / FSMC_NBL0 / DCMI_D2/ EVENTOUT	-
-	ı	98	142	A3	170	PE1	I/O	FT	-	FSMC_NBL1 / DCMI_D3/ EVENTOUT	-
63	-	99	-	D5	-	V_{SS}	S	ı	-	-	-

Pin number / O structure Pin type Pin name Notes **UFBGA176** Additional LQFP176 WLCSP90 LQFP100 LQFP144 LQFP64 (function after **Alternate functions** functions reset)(1) 143 C6 171 PDR_ON I FT **A8** -10 64 A1 144 C5 172 S V_{DD} 0 TIM8_BKIN / DCMI_D5/ PI4 I/O FT Π4 173 **EVENTOUT** TIM8_CH1/ 174 PI5 I/O DCMI VSYNC/ C4 FT **EVENTOUT** TIM8 CH2 / DCMI D6/ C3 175 PI6 I/O FT **EVENTOUT** TIM8_CH3 / DCMI_D7/ C2 176 PI7 I/O FT **EVENTOUT**

Table 7. STM32F41xxx pin and ball definitions (continued)

- 1. Function availability depends on the chosen device.
- PC13, PC14, PC15 and PI8 are supplied through the power switch. Since the switch only sinks a limited amount of current (3 mA), the use of GPIOs PC13 to PC15 and PI8 in output mode is limited:
 - The speed should not exceed 2 MHz with a maximum load of 30 pF.

 - These I/Os must not be used as a current source (e.g. to drive an LED).
- 3. Main function after the first backup domain power-up. Later on, it depends on the contents of the RTC registers even after reset (because these registers are not reset by the main reset). For details on how to manage these I/Os, refer to the RTC register description sections in the STM32F4xx reference manual, available from the STMicroelectronics website: www.st.com.
- 4. FT = 5 V tolerant except when in analog mode or oscillator mode (for PC14, PC15, PH0 and PH1).
- If the device is delivered in an UFBGA176 or WLCSP90 and the BYPASS_REG pin is set to VDD (Regulator off/internal reset ON mode), then PA0 is used as an internal Reset (active low).

Table 8. FSMC pin definition

			FSMC			WLCSP90
Pins ⁽¹⁾	CF	NOR/PSRAM/ SRAM	NOR/PSRAM Mux		LQFP100 ⁽²⁾	(2)
PE2	-	A23	A23	-	Yes	-
PE3	-	A19	A19	-	Yes	-
PE4	-	A20	A20	-	Yes	-
PE5	-	A21	A21	-	Yes	-
PE6	ı	A22	A22	-	Yes	-
PF0	A0	A0	-	-	-	-



Table 8. FSMC pin definition (continued)

			FSMC pin definition			
Pins ⁽¹⁾	CF	NOR/PSRAM/ SRAM	NOR/PSRAM Mux	NAND 16 bit	LQFP100 ⁽²⁾	WLCSP90 (2)
PF1	A1	A1	-	-	-	-
PF2	A2	A2	-	-	-	-
PF3	A3	A3	-	-	-	-
PF4	A4	A4	-	-	-	-
PF5	A5	A5	-	-	-	-
PF6	NIORD	-	-	-	-	-
PF7	NREG	-	-	-	-	-
PF8	NIOWR	-	-	-	-	-
PF9	CD	-	-	-	-	-
PF10	INTR	-	-	-	-	-
PF12	A6	A6	-	-	-	-
PF13	A7	A7	-	-	-	-
PF14	A8	A8	-	-	-	-
PF15	A9	A9	-	-	-	-
PG0	A10	A10	-	-	-	-
PG1		A11	-	-	-	-
PE7	D4	D4	DA4	D4	Yes	Yes
PE8	D5	D5	DA5	D5	Yes	Yes
PE9	D6	D6	DA6	D6	Yes	Yes
PE10	D7	D7	DA7	D7	Yes	Yes
PE11	D8	D8	DA8	D8	Yes	Yes
PE12	D9	D9	DA9	D9	Yes	Yes
PE13	D10	D10	DA10	D10	Yes	Yes
PE14	D11	D11	DA11	D11	Yes	Yes
PE15	D12	D12	DA12	D12	Yes	Yes
PD8	D13	D13	DA13	D13	Yes	Yes
PD9	D14	D14	DA14	D14	Yes	Yes
PD10	D15	D15	DA15	D15	Yes	Yes
PD11	-	A16	A16	CLE	Yes	Yes
PD12	-	A17	A17	ALE	Yes	Yes
PD13	-	A18	A18	-	Yes	-
PD14	D0	D0	DA0	D0	Yes	Yes
PD15	D1	D1	DA1	D1	Yes	Yes

Table 8. FSMC pin definition (continued)

			FSMC	•		WI CCDOO
Pins ⁽¹⁾	CF	NOR/PSRAM/ SRAM	NOR/PSRAM Mux	NAND 16 bit	LQFP100 ⁽²⁾	WLCSP90 (2)
PG2	-	A12	-	-	-	-
PG3	-	A13	-	-	-	-
PG4	-	A14	-	-	-	-
PG5	-	A15	-	-	-	-
PG6	-	-	-	INT2	-	-
PG7	-	-	-	INT3	-	-
PD0	D2	D2	DA2	D2	Yes	Yes
PD1	D3	D3	DA3	D3	Yes	Yes
PD3	-	CLK	CLK	-	Yes	-
PD4	NOE	NOE	NOE	NOE	Yes	Yes
PD5	NWE	NWE	NWE	NWE	Yes	Yes
PD6	NWAIT	NWAIT	NWAIT	NWAIT	Yes	Yes
PD7	-	NE1	NE1	NCE2	Yes	Yes
PG9	-	NE2	NE2	NCE3	-	-
PG10	NCE4_1	NE3	NE3	-	-	-
PG11	NCE4_2	-	-	-	-	-
PG12	-	NE4	NE4	-	-	-
PG13	-	A24	A24	-	-	-
PG14	-	A25	A25	-	-	-
PB7	-	NADV	NADV	-	Yes	Yes
PE0	-	NBL0	NBL0	-	Yes	-
PE1	-	NBL1	NBL1	-	Yes	-

Full FSMC features are available on LQFP144, LQFP176, and UFBGA176. The features available on smaller packages are given in the dedicated package column.

^{2.} Ports F and G are not available in devices delivered in 100-pin packages.

Table	9. Alterr	nate func	tion	map	ping

		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13		
P	ort	sys	TIM1/2	TIM3/4/5	TIM8/9/10 /11	I2C1/2/3	SPI1/SPI2/ I2S2/I2S2e xt	SPI3/I2Sext /I2S3	USART1/2/3/ I2S3ext	UART4/5/ USART6	CAN1/2 TIM12/13/ 14	OTG_FS/ OTG_HS	ЕТН	FSMC/SDIO /OTG_FS	DCMI	AF14	AF15
	PA0	-	TIM2_CH1_ ETR	TIM 5_CH1	TIM8_ETR	-	-	-	USART2_CTS	UART4_TX	-	-	ETH_MII_CRS	-	-	-	EVENTOUT
	PA1	-	TIM2_CH2	TIM5_CH2	-	-	-	-	USART2_RTS	UART4_RX	-	-	ETH_MII _RX_CLK ETH_RMIIREF _CLK	-	-	-	EVENTOUT
	PA2	-	TIM2_CH3	TIM5_CH3	TIM9_CH1	-	-	-	USART2_TX	-	-	-	ETH_MDIO	-	-	-	EVENTOUT
	PA3		TIM2_CH4	TIM5_CH4	TIM9_CH2	-	-	-	USART2_RX	-	-	OTG_HS_ULPI_ D0	ETH _MII_COL	-	-	-	EVENTOUT
	PA4		-	-	-	-	SPI1_NSS	SPI3_NSS I2S3_WS	USART2_CK	-	-	-	-	OTG_HS_SOF	DCMI_ HSYNC	-	EVENTOUT
	PA5	-	TIM2_CH1_ ETR	-	TIM8_CH1N	-	SPI1_SCK	-	-	-	-	OTG_HS_ULPI_ CK	-	-	-	-	EVENTOUT
	PA6	-	TIM1_BKIN	TIM3_CH1	TIM8_BKIN	-	SPI1_MISO	-	-	=	TIM13_CH1	-	-	-	DCMI_PIXCK	-	EVENTOUT
Port A	PA7	-	TIM1_CH1N	TIM3_CH2	TIM8_CH1N	=	SPI1_MOSI	-	-	=	TIM14_CH1	-	ETH_MII_RX_DV ETH_RMII _CRS_DV	-	-	-	EVENTOUT
	PA8	MCO1	TIM1_CH1	-	-	I2C3_SCL	=	=	USART1_CK	=	-	OTG_FS_SOF	=	-	-	-	EVENTOUT
	PA9	=	TIM1_CH2	-	-	I2C3_ SMBA	=	=	USART1_TX	Ξ	-	-	-	-	DCMI_D0	-	EVENTOUT
	PA10	=	TIM1_CH3		-	-	=	=	USART1_RX	=	-	OTG_FS_ID	=	-	DCMI_D1	-	EVENTOUT
	PA11	-	TIM1_CH4	-	-	-	-	-	USART1_CTS	-	CAN1_RX	OTG_FS_DM	-	-	-	-	EVENTOUT
	PA12	-	TIM1_ETR	-	-	-	-	-	USART1_RTS	-	CAN1_TX	OTG_FS_DP	-	-	-	-	EVENTOUT
	PA13	JTMS- SWDIO	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENTOUT
	PA14	JTCK- SWCLK	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENTOUT
	PA15	JTDI	TIM 2_CH1 TIM 2_ETR	-	-	-	SPI1_NSS	SPI3_NSS/ I2S3_WS	-	-	-	-	-	-	-	-	EVENTOUT





Table 9. Alternate function mapping (continued)

			1		1				· · · · · · · · · · · · · · · · · · ·	<u> </u>	<u>`</u>		1		1		ı — —
		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13		
Po	ort	sys	TIM1/2	TIM3/4/5	TIM8/9/10 /11	I2C1/2/3	SPI1/SPI2/ I2S2/I2S2e xt	SPI3/I2Sext /I2S3	USART1/2/3/ I2S3ext	UART4/5/ USART6	CAN1/2 TIM12/13/ 14	OTG_FS/ OTG_HS	ЕТН	FSMC/SDIO /OTG_FS	DCMI	AF14	AF15
	PB0	ı	TIM1_CH2N	TIM3_CH3	TIM8_CH2N	-	-	-	-	-	-	OTG_HS_ULPI_ D1	ETH _MII_RXD2	-	-	ı	EVENTOUT
	PB1	-	TIM1_CH3N	TIM3_CH4	TIM8_CH3N		-	-	-	-	-	OTG_HS_ULPI_ D2	ETH _MII_RXD3	-	-	-	EVENTOUT
	PB2	ii ii	=	ē	-	=	=	=	=	÷	-	=	=	=	-	=	EVENTOUT
	PB3	JTDO/ TRACES WO	TIM2_CH2	-	-	-	SPI1_SCK	SPI3_SCK I2S3_CK	-	-	-	-	-	-	-	=	EVENTOUT
	PB4	NJTRST	=	TIM3_CH1		=	SPI1_MISO	SPI3_MISO	I2S3ext_SD	=	-	=	=	=	-	-	EVENTOUT
	PB5	-	=	TIM3_CH2		I2C1_SMB A	SPI1_MOSI	SPI3_MOSI I2S3_SD		=	CAN2_RX	OTG_HS_ULPI_ D7	ETH_PPS_OUT	=	DCMI_D10	-	EVENTOUT
	PB6	-	-	TIM4_CH1		I2C1_SCL	-	-	USART1_TX	-	CAN2_TX	-	-	-	DCMI_D5	-	EVENTOUT
	PB7	-	-	TIM4_CH2		I2C1_SDA	-	-	USART1_RX	-	-	-	-	FSMC_NL	DCMI_VSYN C	-	EVENTOUT
Port B	PB8	-	-	TIM4_CH3	TIM10_CH1	I2C1_SCL	-	-	-	-	CAN1_RX	-	ETH _MII_TXD3	SDIO_D4	DCMI_D6	-	EVENTOUT
	PB9	-	-	TIM4_CH4	TIM11_CH1	I2C1_SDA	SPI2_NSS I2S2_WS	-	-	-	CAN1_TX	-	-	SDIO_D5	DCMI_D7	ı	EVENTOUT
	PB10	1	TIM2_CH3	-	-	I2C2_SCL	SPI2_SCK I2S2_CK	-	USART3_TX	-	-	OTG_HS_ULPI_ D3	ETH_MII_RX_ER	-	-	ı	EVENTOUT
	PB11	÷	TIM2_CH4	-	-	I2C2_SDA	-	-	USART3_RX	=	-	OTG_HS_ULPI_ D4	ETH _MII_TX_EN ETH _RMII_TX_EN	=	-	-	EVENTOUT
	PB12	=	TIM1_BKIN	-	-	I2C2_ SMBA	SPI2_NSS I2S2_WS	-	USART3_CK	=	CAN2_RX	OTG_HS_ULPI_ D5	ETH _MII_TXD0 ETH _RMII_TXD0	OTG_HS_ID	-	=	EVENTOUT
	PB13	=	TIM1_CH1N	-	-	=	SPI2_SCK I2S2_CK	-	USART3_CTS	=	CAN2_TX	OTG_HS_ULPI_ D6	ETH _MII_TXD1 ETH _RMII_TXD1	-	-	=	EVENTOUT
	PB14	÷	TIM1_CH2N	-	TIM8_CH2N	÷	SPI2_MISO	I2S2ext_SD	USART3_RTS	-	TIM12_CH1	-	-	OTG_HS_DM	-	-	EVENTOUT
	PB15	RTC_ REFIN	TIM1_CH3N	-	TIM8_CH3N	-	SPI2_MOSI I2S2_SD	-	-	-	TIM12_CH2	-	-	OTG_HS_DP	-	-	EVENTOUT

Table 9. Alternate function mapping (continued)

										~PP9	(,					
		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13		
Po	ort	sys	TIM1/2	TIM3/4/5	TIM8/9/10 /11	I2C1/2/3	SPI1/SPI2/ I2S2/I2S2e xt	SPI3/I2Sext /I2S3	USART1/2/3/ I2S3ext	UART4/5/ USART6	CAN1/2 TIM12/13/ 14	OTG_FS/ OTG_HS	ETH	FSMC/SDIO /OTG_FS	DCMI	AF14	AF15
	PC0	-	=	-	-	-	-	=	-	=	-	OTG_HS_ULPI_ STP	-	=	=	-	EVENTOUT
	PC1	-	-	-	-	-	-	-	-	-	-	-	ETH_MDC	-	-	-	EVENTOUT
	PC2	-	-	-	-	-	SPI2_MISO	I2S2ext_SD	-	-	-	OTG_HS_ULPI_ DIR	ETH _MII_TXD2	-	-	-	EVENTOUT
	PC3	=	-	-	-	=	SPI2_MOSI I2S2_SD	-	-	=	-	OTG_HS_ULPI_ NXT	ETH _MII_TX_CLK	=	=	-	EVENTOUT
	PC4	=	-	-	-	=	-	=	-	=	-	-	ETH_MII_RXD0 ETH_RMII_RXD0	=	=	-	EVENTOUT
	PC5	=	-	-	-	=	-	=	-	=	-	-	ETH _MII_RXD1 ETH _RMII_RXD1	=	=	-	EVENTOUT
	PC6			TIM3_CH1	TIM8_CH1		I2S2_MCK		-	USART6_TX	-	-	-	SDIO_D6	DCMI_D0	-	EVENTOUT
Port C	PC7			TIM3_CH2	TIM8_CH2	-	-	I2S3_MCK	-	USART6_RX	-	-	-	SDIO_D7	DCMI_D1	-	EVENTOUT
	PC8	ì	-	TIM3_CH3	TIM8_CH3	-	-	-	1	USART6_CK	-	-	1	SDIO_D0	DCMI_D2	1	EVENTOUT
	PC9	MCO2	-	TIM3_CH4	TIM8_CH4	I2C3_SDA	I2S_CKIN	-	-	-	-	-	-	SDIO_D1	DCMI_D3	-	EVENTOUT
	PC10	i i	=	=	-	-	-	SPI3_SCK/ I2S3_CK	USART3_TX/	UART4_TX	=	=	-	SDIO_D2	DCMI_D8		EVENTOUT
	PC11	i	=	-	-	=	I2S3ext_SD	SPI3_MISO/	USART3_RX	UART4_RX	=	=	=	SDIO_D3	DCMI_D4	-	EVENTOUT
	PC12	ı	-	-	-	-	-	SPI3_MOSI I2S3_SD	USART3_CK	UART5_TX	-	-	-	SDIO_CK	DCMI_D9	-	EVENTOUT
	PC13	ii ii	=·	-	=	=	=	=	=	=	=	=	=	=	=	=	EVENTOUT
	PC14	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENTOUT
	PC15	i		-	-	-	-	-	-	-	-	-	-	-	-	-	EVENTOUT



Table 9. Alternate function mapping (continued)

											(,					
		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13		
P	ort	sys	TIM1/2	TIM3/4/5	TIM8/9/10 /11	I2C1/2/3	SPI1/SPI2/ I2S2/I2S2e xt	SPI3/I2Sext /I2S3	USART1/2/3/ I2S3ext	UART4/5/ USART6	CAN1/2 TIM12/13/ 14	OTG_FS/ OTG_HS	ETH	FSMC/SDIO /OTG_FS	DCMI	AF14	AF15
	PD0	÷	-	-	-	-	-	-	-	-	CAN1_RX	-	-	FSMC_D2	-	-	EVENTOUT
	PD1	-	-	-	-	-	-	-	-	-	CAN1_TX	-	-	FSMC_D3	-	-	EVENTOUT
	PD2	=	=	TIM3_ETR	-	-	=	=	=	UART5_RX	-	=	=	SDIO_CMD	DCMI_D11	-	EVENTOUT
	PD3	=	=	-	-	-	-	=	USART2_CTS	=	-	ē	=	FSMC_CLK	=	-	EVENTOUT
	PD4	-	-	-	-	-	-	-	USART2_RTS	=	-	=	-	FSMC_NOE	=	-	EVENTOUT
	PD5	-	-	-	-	-	-	-	USART2_TX	=	-	=	-	FSMC_NWE	=	-	EVENTOUT
	PD6	-	-	-	-	-	-	-	USART2_RX	-	-	-	-	FSMC_NWAIT	-	-	EVENTOUT
Port D	PD7	-	=	-	-	-	-	=	USART2_CK	=	-	-	-	FSMC_NE1/ FSMC_NCE2	=	-	EVENTOUT
	PD8	=	=	-	-	-	-	=	USART3_TX	=	-	ē	=	FSMC_D13	=	-	EVENTOUT
	PD9	-	-	-	-	-	-	-	USART3_RX	=	-	=	-	FSMC_D14	=	-	EVENTOUT
	PD10	-	-	-	-	-	-	-	USART3_CK	-	-	-	-	FSMC_D15	-	-	EVENTOUT
	PD11	-	-	-	-	-	-	-	USART3_CTS	-	-	-	-	FSMC_A16	-	-	EVENTOUT
	PD12	-		TIM4_CH1	-	-	-	-	USART3_RTS		-	-	-	FSMC_A17	-	-	EVENTOUT
	PD13	=	-	TIM4_CH2	-	-	-	-	=	=	-	=	=	FSMC_A18	=	-	EVENTOUT
	PD14	-		TIM4_CH3	-	-	-	-	-	-	-	-	-	FSMC_D0	-	-	EVENTOUT
	PD15	=	=	TIM4_CH4	-	-	-	-	=	=	-	=	=	FSMC_D1	=	-	EVENTOUT

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Table 9. Alternate function mapping (continued)

										~PP9	•	,					
		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13		
Pe	ort	sys	TIM1/2	TIM3/4/5	TIM8/9/10 /11	I2C1/2/3	SPI1/SPI2/ I2S2/I2S2e xt	SPI3/I2Sext /I2S3	USART1/2/3/ I2S3ext	UART4/5/ USART6	CAN1/2 TIM12/13/ 14	OTG_FS/ OTG_HS	ЕТН	FSMC/SDIO /OTG_FS	DCMI	AF14	AF15
	PE0	-	=	TIM4_ETR	-	-	-	-	-	-	-	E	=	FSMC_NBL0	DCMI_D2	-	EVENTOUT
	PE1	-	=	=	-	-	=	-	=	-	-	i i	=	FSMC_NBL1	DCMI_D3	-	EVENTOUT
	PE2	TRACECL K	-	-	-	=	-	-	-	-	-	=	ETH _MII_TXD3	FSMC_A23	=	=	EVENTOUT
	PE3	TRACED0	-	-	-	-	-	-	-	-	-	=	-	FSMC_A19	-	-	EVENTOUT
	PE4	TRACED1	-	-	-	-	-	-	-	-	-	-	-	FSMC_A20	DCMI_D4	-	EVENTOUT
	PE5	TRACED2	-	-	TIM9_CH1	-	-	-	-	-	-	=	-	FSMC_A21	DCMI_D6	-	EVENTOUT
	PE6	TRACED3	-	-	TIM9_CH2	-	-	-	-	-	-	=	-	FSMC_A22	DCMI_D7	-	EVENTOUT
Port E	PE7	-	TIM1_ETR	-	-	-	-	-	-	-	-	=	-	FSMC_D4	-	-	EVENTOUT
	PE8	-	TIM1_CH1N	-	-	-	-	-	-	-	-	=	-	FSMC_D5	-	-	EVENTOUT
	PE9	-	TIM1_CH1	-	-	-	-	-	-	-	-	=	-	FSMC_D6	=	-	EVENTOUT
	PE10	-	TIM1_CH2N	-	-	-	-	-	-	-	-	=	-	FSMC_D7	=	-	EVENTOUT
	PE11	-	TIM1_CH2	-	-	-	-	-	-	-	-	=	-	FSMC_D8	=	-	EVENTOUT
	PE12	-	TIM1_CH3N	-	-	-	-	-	-	-	-	=	-	FSMC_D9	=	-	EVENTOUT
	PE13	-	TIM1_CH3	-	-	-	-	-	-	-	-	=	-	FSMC_D10	=	-	EVENTOUT
	PE14	-	TIM1_CH4	=	-	=	=	-	=	-	-	=	=	FSMC_D11	=	=	EVENTOUT
	PE15	-	TIM1_BKIN	-	-	-	-	-	-	-	-	-	-	FSMC_D12	-	-	EVENTOUT



Table 9. Alternate function mapping (continued)

									i .		(00						
		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13		
Po	ort	sys	TIM1/2	TIM3/4/5	TIM8/9/10 /11	I2C1/2/3	SPI1/SPI2/ I2S2/I2S2e xt	SPI3/I2Sext /I2S3	USART1/2/3/ I2S3ext	UART4/5/ USART6	CAN1/2 TIM12/13/ 14	OTG_FS/ OTG_HS	ETH	FSMC/SDIO /OTG_FS	DCMI	AF14	AF15
	PF0	=	=	=	=	I2C2_SDA	=	=	=	=	-	=	=	FSMC_A0	=	-	EVENTOUT
	PF1	÷	Е	T.	=	I2C2_SCL	е	TI.	i i	Ü	-	Ē	i i	FSMC_A1	ii.	-	EVENTOUT
	PF2	-	-	-	-	I2C2_ SMBA	-	-	-	-	-	-	-	FSMC_A2	-	-	EVENTOUT
	PF3	=	=	=	=	=	=	i i	ē	ii.	-	=	-	FSMC_A3	=	-	EVENTOUT
	PF4	=	=	=	=	=	=	=	=	-	-	=	=	FSMC_A4	₽	-	EVENTOUT
	PF5	i	-	ī	-	-	-	1	i	ı	-	-	1	FSMC_A5	i	-	EVENTOUT
	PF6	-	-	-	TIM10_CH1	-	-	1	i	1	-	-	1	FSMC_NIORD	-	-	EVENTOUT
Port F	PF7	-	-	-	TIM11_CH1	-	-	-	-	-	-	-	-	FSMC_NREG	-	-	EVENTOUT
roiti	PF8	-	-	-	-	-	-	-	-	-	TIM13_CH1	-	-	FSMC_ NIOWR	-	-	EVENTOUT
	PF9	=	=	=	-	-	=	=	E	=	TIM14_CH1	=	=	FSMC_CD	=	-	EVENTOUT
	PF10	i	-	ı	-	1	1	ı	ì	ī	-	-	ı	FSMC_INTR	i	-	EVENTOUT
	PF11	i	-	ī	-	-	-	1	i	ı	-	-	1		DCMI_D12	-	EVENTOUT
	PF12	-	-	-	-	-	-	-	-	-	-	-	-	FSMC_A6	-	-	EVENTOUT
	PF13	-	-	-	-	-	-	-	-	-	-	-	-	FSMC_A7	-	-	EVENTOUT
	PF14	-	-	-	-	-	-	-	-	-	-	-	-	FSMC_A8	-	-	EVENTOUT
	PF15	-	-	-	-	-	-	-	-	-	-	-	-	FSMC_A9	-	-	EVENTOUT

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Tab	le 9. Alt	ernate fu	ınction m	apping	(contin	ued)
AF4	AF5	AF6	AF7	AF8	AF9	AF1

		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13		
P	ort	sys	TIM1/2	TIM3/4/5	TIM8/9/10 /11	I2C1/2/3	SPI1/SPI2/ I2S2/I2S2e xt	SPI3/I2Sext /I2S3	USART1/2/3/ I2S3ext	UART4/5/ USART6	CAN1/2 TIM12/13/ 14	OTG_FS/ OTG_HS	ETH	FSMC/SDIO /OTG_FS	DCMI	AF14	AF15
	PG0	-	-	-	-	-	-	-	-	-	-	-	-	FSMC_A10	-	-	EVENTOUT
	PG1	-	-	-	-	-	-	-		-	-	-	-	FSMC_A11	-	-	EVENTOUT
	PG2	-	-	-	-	-	-	-	-	-	-	-	-	FSMC_A12	-	-	EVENTOUT
	PG3	-	-	-	-	-	-	-	-	-	-	-	-	FSMC_A13	-	-	EVENTOUT
	PG4	-	-	-	-	-	-	-	-	-	-	-	-	FSMC_A14	-	-	EVENTOUT
	PG5	-	-	-	-	-	-	-	-	-	-	-	-	FSMC_A15	-	-	EVENTOUT
	PG6	-	-	-	-	-	-	-	-	-	-	-	-	FSMC_INT2	-	-	EVENTOUT
	PG7	-	-	-	-	-	-	-	-	USART6_CK	-	-	-	FSMC_INT3	-	-	EVENTOUT
	PG8	-	-	-	-	-	-	-	-	USART6_ RTS	-	-	ETH _PPS_OUT	-	-	-	EVENTOUT
Port G	PG9	=	=	-	=	=	=	=	-	USART6_RX	-	-	-	FSMC_NE2/ FSMC_NCE3	-	-	EVENTOUT
	PG10	Ü	=	÷	=	=	-	=	=	-	-	-	-	FSMC_ NCE4_1/ FSMC_NE3	-	-	EVENTOUT
	PG11	-	-	-	-	-	-	-	-	-	-	-	ETH _MII_TX_EN ETH _RMII_ TX_EN	FSMC_NCE4_ 2	-	-	EVENTOUT
	PG12	=	=	-	-	-	-	=	=	USART6_ RTS	-	-	-	FSMC_NE4	=	-	EVENTOUT
	PG13	-	-	-	-	-	-	-	-	UART6_CTS	-	-	ETH _MII_TXD0 ETH _RMII_TXD0	FSMC_A24	-	-	EVENTOUT
	PG14	-	-	-	-	-	-	-	-	USART6_TX	-	-	ETH _MII_TXD1 ETH _RMII_TXD1	FSMC_A25	-	-	EVENTOUT
	PG15	-	-	-	-	-	-	-	-	USART6_ CTS	-	-	-	-	DCMI_D13	-	EVENTOUT





Table 9. Alternate function mapping (continued)

		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13			
P	Port	sys	sys	TIM1/2	TIM3/4/5	TIM8/9/10 /11	I2C1/2/3	SPI1/SPI2/ I2S2/I2S2e xt	SPI3/I2Sext /I2S3	USART1/2/3/ I2S3ext	UART4/5/ USART6	CAN1/2 TIM12/13/ 14	OTG_FS/ OTG_HS	ETH	FSMC/SDIO /OTG_FS	DCMI	AF14	AF15
	PH0	-	-	=	-	=	-	=	=	=	-	ē	=	=	Ξ	-	EVENTOUT	
	PH1	1	-	=	-	ı	-	=	=	Ш	=	i)	The state of the s	e	Ξ	-	EVENTOUT	
	PH2	1	-	-	-	1	-	-	-	1	-	1	ETH _MII_CRS	-	-	-	EVENTOUT	
	PH3	1	-	-	-	1	-	-	-	1	-	1	ETH _MII_COL	-	-	-	EVENTOUT	
	PH4	-	-	-	-	I2C2_SCL	-	-	-	-	-	OTG_HS_ULPI_ NXT	-	-	-	-	EVENTOUT	
	PH5	-	-	-	-	I2C2_SDA	-	=	=	=	-	=	≡	=	=	-	EVENTOUT	
	PH6	-	-	-	-	I2C2_ SMBA	-	-	-	-	TIM12_CH1	=	ETH _MII_RXD2	-	-	=	EVENTOUT	
Port H	PH7	-	-	-	-	I2C3_SCL	-	-	-	1	-	-	ETH _MII_RXD3	-	-	-	EVENTOUT	
FOILH	PH8	i	-	-	-	I2C3_SDA	-	-	-	ı	-	=	=	-	DCMI_ HSYNC	=	EVENTOUT	
	PH9	-	-	-	-	I2C3_ SMBA	-	=	-	=	TIM12_CH2	=	=	=	DCMI_D0	=	EVENTOUT	
	PH10	-	-	TIM5_CH1	-	-	-	-	-	1	-	-	-	-	DCMI_D1	-	EVENTOUT	
	PH11	1	-	TIM5_CH2	-	1	-	-	-	1	-	1	1	-	DCMI_D2	-	EVENTOUT	
	PH12	1	-	TIM5_CH3	-	-	-	-	-	-	-	-	-	-	DCMI_D3	-	EVENTOUT	
	PH13	-	-	-	TIM8_CH1N	-	-	-	-	-	CAN1_TX	-	-	-	-	-	EVENTOUT	
	PH14	-	-	-	TIM8_CH2N	-	-	-	-	-	-	-	-	-	DCMI_D4	-	EVENTOUT	
	PH15	-	-	-	TIM8_CH3N	=	-	-	-	-	-	-	-	-	DCMI_D11	-	EVENTOUT	

Table 9. Alternate function mapping (continued)

	Table of Automate Table in approp						(commutation)										
		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13		
Pe	ort	sys	TIM1/2	TIM3/4/5	TIM8/9/10 /11	I2C1/2/3	SPI1/SPI2/ I2S2/I2S2e xt	SPI3/I2Sext /I2S3	USART1/2/3/ I2S3ext	UART4/5/ USART6	CAN1/2 TIM12/13/ 14	OTG_FS/ OTG_HS	ETH	FSMC/SDIO /OTG_FS	DCMI	AF14	AF15
	PI0	i	-	TIM5_CH4	-	-	SPI2_NSS I2S2_WS	-	=	-	-	-	=	-	DCMI_D13	ı	EVENTOUT
	PI1	-	-	-	-	-	SPI2_SCK I2S2_CK	-	-	=	-	-	-	-	DCMI_D8	-	EVENTOUT
	PI2	=	-	-	TIM8_CH4	=	SPI2_MISO	I2S2ext_SD	=	=	=	=	=	=	DCMI_D9	=	EVENTOUT
	PI3	÷.	-	=	TIM8_ETR	=	SPI2_MOSI I2S2_SD	=	=	=	-	=	=	n.	DCMI_D10	=	EVENTOUT
	PI4	=	-	-	TIM8_BKIN	-	-	-	=	-	-	-	-	-	DCMI_D5	-	EVENTOUT
Port I	PI5	-	-	-	TIM8_CH1	-	-	-	-	-	-	-	-	-	DCMI_ VSYNC	-	EVENTOUT
	PI6	-	-	-	TIM8_CH2	-	-	-	-	-	-	-	-	-	DCMI_D6	-	EVENTOUT
	PI7	-	-	-	TIM8_CH3	-	-	-	-	-	-	-	-	-	DCMI_D7	-	EVENTOUT
	PI8	-	-	-	-	-	-	-	-	-	-	-	-	-	1	-	EVENTOUT
	PI9	-	-	-	-	-	-	-	-	-	CAN1_RX	-	-	-	-	-	EVENTOUT
	PI10	i	-	-	-	-	-	-	-	-	-	-	ETH _MII_RX_ER	-	-	i	EVENTOUT
	PI11	i	-	-	-	-	-	-	1	-	-	OTG_HS_ULPI_ DIR	-	-	1	i	EVENTOUT

4 Memory mapping

The memory map is shown in Figure 18.

Figure 18. STM32F41xxx memory map

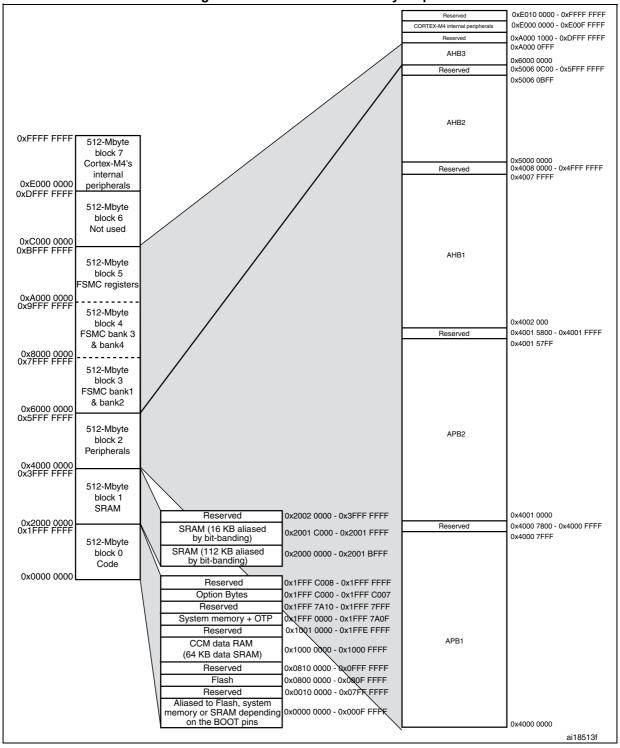


Table 10. STM32F41x register boundary addresses

Bus	Boundary address	Peripheral
	0xE00F FFFF - 0xFFFF FFFF	Reserved
Cortex-M4	0xE000 0000 - 0xE00F FFFF	Cortex-M4 internal peripherals
	0xA000 1000 - 0xDFFF FFFF	Reserved
	0xA000 0000 - 0xA000 0FFF	FSMC control register
	0x9000 0000 - 0x9FFF FFFF	FSMC bank 4
AHB3	0x8000 0000 - 0x8FFF FFFF	FSMC bank 3
	0x7000 0000 - 0x7FFF FFFF	FSMC bank 2
	0x6000 0000 - 0x6FFF FFFF	FSMC bank 1
	0x5006 0C00- 0x5FFF FFFF	Reserved
	0x5006 0800 - 0x5006 0BFF	RNG
	0x5006 0400 - 0x5006 07FF	HASH
	0x5006 0000 - 0x5006 03FF	CRYP
AHB2	0x5005 0400 - 0x5005 FFFF	Reserved
	0x5005 0000 - 0x5005 03FF	DCMI
	0x5004 0000- 0x5004 FFFF	Reserved
	0x5000 0000 - 0x5003 FFFF	USB OTG FS
	0x4008 0000- 0x4FFF FFFF	Reserved

Table 10. STM32F41x register boundary addresses (continued)

Bus	Boundary address	Peripheral
	0x4004 0000 - 0x4007 FFFF	USB OTG HS
	0x4002 9400 - 0x4003 FFFF	Reserved
	0x4002 9000 - 0x4002 93FF	
	0x4002 8C00 - 0x4002 8FFF	
	0x4002 8800 - 0x4002 8BFF	ETHERNET MAC
	0x4002 8400 - 0x4002 87FF	
	0x4002 8000 - 0x4002 83FF	
	0x4002 6800 - 0x4002 7FFF	Reserved
	0x4002 6400 - 0x4002 67FF	DMA2
	0x4002 6000 - 0x4002 63FF	DMA1
	0x4002 5000 - 0x4002 5FFF	Reserved
	0x4002 4000 - 0x4002 4FFF	BKPSRAM
AHB1	0x4002 3C00 - 0x4002 3FFF	Flash interface register
ALIDI	0x4002 3800 - 0x4002 3BFF	RCC
	0x4002 3400 - 0x4002 37FF	Reserved
	0x4002 3000 - 0x4002 33FF	CRC
	0x4002 2400 - 0x4002 2FFF	Reserved
	0x4002 2000 - 0x4002 23FF	GPIOI
	0x4002 1C00 - 0x4002 1FFF	GPIOH
	0x4002 1800 - 0x4002 1BFF	GPIOG
	0x4002 1400 - 0x4002 17FF	GPIOF
	0x4002 1000 - 0x4002 13FF	GPIOE
	0x4002 0C00 - 0x4002 0FFF	GPIOD
	0x4002 0800 - 0x4002 0BFF	GPIOC
	0x4002 0400 - 0x4002 07FF	GPIOB
	0x4002 0000 - 0x4002 03FF	GPIOA
	0x4001 5800- 0x4001 FFFF	Reserved

Table 10. STM32F41x register boundary addresses (continued)

Bus	Boundary address	Peripheral
	0x4001 4C00 - 0x4001 57FF	Reserved
	0x4001 4800 - 0x4001 4BFF	TIM11
	0x4001 4400 - 0x4001 47FF	TIM10
	0x4001 4000 - 0x4001 43FF	TIM9
	0x4001 3C00 - 0x4001 3FFF	EXTI
	0x4001 3800 - 0x4001 3BFF	SYSCFG
	0x4001 3400 - 0x4001 37FF	Reserved
	0x4001 3000 - 0x4001 33FF	SPI1
APB2	0x4001 2C00 - 0x4001 2FFF	SDIO
	0x4001 2400 - 0x4001 2BFF	Reserved
	0x4001 2000 - 0x4001 23FF	ADC1 - ADC2 - ADC3
	0x4001 1800 - 0x4001 1FFF	Reserved
	0x4001 1400 - 0x4001 17FF	USART6
	0x4001 1000 - 0x4001 13FF	USART1
	0x4001 0800 - 0x4001 0FFF	Reserved
	0x4001 0400 - 0x4001 07FF	TIM8
	0x4001 0000 - 0x4001 03FF	TIM1
	0x4000 7800- 0x4000 FFFF	Reserved

Table 10. STM32F41x register boundary addresses (continued)

Bus	Boundary address	Peripheral
	0x4000 7800 - 0x4000 7FFF	Reserved
	0x4000 7400 - 0x4000 77FF	DAC
	0x4000 7000 - 0x4000 73FF	PWR
	0x4000 6C00 - 0x4000 6FFF	Reserved
	0x4000 6800 - 0x4000 6BFF	CAN2
	0x4000 6400 - 0x4000 67FF	CAN1
	0x4000 6000 - 0x4000 63FF	Reserved
	0x4000 5C00 - 0x4000 5FFF	I2C3
	0x4000 5800 - 0x4000 5BFF	I2C2
	0x4000 5400 - 0x4000 57FF	I2C1
	0x4000 5000 - 0x4000 53FF	UART5
	0x4000 4C00 - 0x4000 4FFF	UART4
	0x4000 4800 - 0x4000 4BFF	USART3
	0x4000 4400 - 0x4000 47FF	USART2
	0x4000 4000 - 0x4000 43FF	I2S3ext
APB1	0x4000 3C00 - 0x4000 3FFF	SPI3 / I2S3
	0x4000 3800 - 0x4000 3BFF	SPI2 / I2S2
	0x4000 3400 - 0x4000 37FF	I2S2ext
	0x4000 3000 - 0x4000 33FF	IWDG
	0x4000 2C00 - 0x4000 2FFF	WWDG
	0x4000 2800 - 0x4000 2BFF	RTC & BKP Registers
	0x4000 2400 - 0x4000 27FF	Reserved
	0x4000 2000 - 0x4000 23FF	TIM14
	0x4000 1C00 - 0x4000 1FFF	TIM13
	0x4000 1800 - 0x4000 1BFF	TIM12
	0x4000 1400 - 0x4000 17FF	TIM7
	0x4000 1000 - 0x4000 13FF	TIM6
	0x4000 0C00 - 0x4000 0FFF	TIM5
	0x4000 0800 - 0x4000 0BFF	TIM4
	0x4000 0400 - 0x4000 07FF	TIM3
	0x4000 0000 - 0x4000 03FF	TIM2

5 Electrical characteristics

5.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to V_{SS}.

5.1.1 Minimum and maximum values

Unless otherwise specified the minimum and maximum values are guaranteed in the worst conditions of ambient temperature, supply voltage and frequencies by tests in production on 100% of the devices with an ambient temperature at $T_A = 25$ °C and $T_A = T_A$ max (given by the selected temperature range).

Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes and are not tested in production. Based on characterization, the minimum and maximum values refer to sample tests and represent the mean value plus or minus three times the standard deviation (mean $\pm 3\Sigma$).

5.1.2 Typical values

Unless otherwise specified, typical data are based on T_A = 25 °C, V_{DD} = 3.3 V (for the 1.8 V \leq V_{DD} \leq 3.6 V voltage range). They are given only as design guidelines and are not tested.

Typical ADC accuracy values are determined by characterization of a batch of samples from a standard diffusion lot over the full temperature range, where 95% of the devices have an error less than or equal to the value indicated (mean $\pm 2\Sigma$).

5.1.3 Typical curves

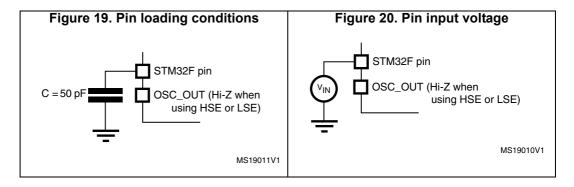
Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

5.1.4 Loading capacitor

The loading conditions used for pin parameter measurement are shown in Figure 19.

5.1.5 Pin input voltage

The input voltage measurement on a pin of the device is described in *Figure 20*.



5.1.6 Power supply scheme

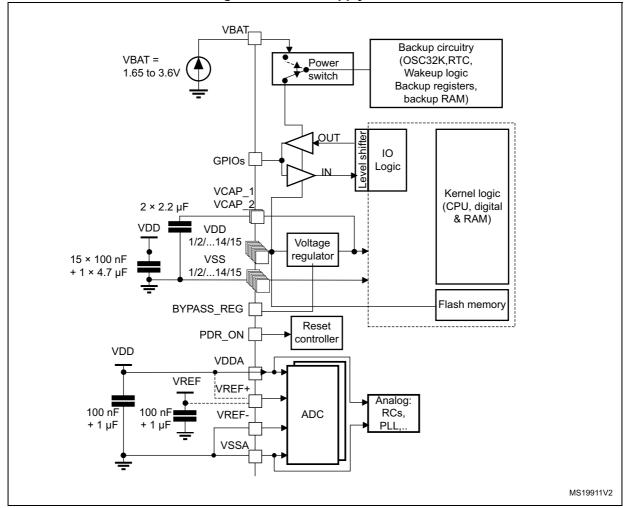


Figure 21. Power supply scheme

- Each power supply pair must be decoupled with filtering ceramic capacitors as shown above. These capacitors must be
 placed as close as possible to, or below, the appropriate pins on the underside of the PCB to ensure the good functionality
 of the device.
- To connect BYPASS_REG and PDR_ON pins, refer to Section 2.2.16: Voltage regulator and Table 2.2.15: Power supply supervisor
- The two 2.2 μF ceramic capacitors should be replaced by two 100 nF decoupling capacitors when the voltage regulator is OFF.
- 4. The 4.7 μ F ceramic capacitor must be connected to one of the V_{DD} pin.
- 5. $V_{DDA}=V_{DD}$ and $V_{SSA}=V_{SS}$.

5.1.7 Current consumption measurement

IDD_VBAT VBAT VDDA

Figure 22. Current consumption measurement scheme

5.2 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in *Table 11: Voltage characteristics*, *Table 12: Current characteristics*, and *Table 13: Thermal characteristics* may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Symbol	Ratings	Min	Max	Unit
V_{DD} – V_{SS}	External main supply voltage (including V_{DDA} , V_{DD}) ⁽¹⁾	-0.3	4.0	
V	Input voltage on five-volt tolerant pin ⁽²⁾	V _{SS} -0.3	V _{DD} +4	V
V _{IN}	Input voltage on any other pin	V _{SS} -0.3	4.0	
$ \Delta V_{DDx} $	Variations between different V _{DD} power pins	-	50	mV
V _{SSX} - V _{SS}	Variations between all the different ground pins	-	50	IIIV
V _{ESD(HBM)}	Electrostatic discharge voltage (human body model)	see Section 5.3.14: Absolute maximum ratings (electrical sensitivity)		

Table 11. Voltage characteristics

- All main power (V_{DD}, V_{DDA}) and ground (V_{SS}, V_{SSA}) pins must always be connected to the external power supply, in the permitted range.
- V_{IN} maximum value must always be respected. Refer to Table 12 for the values of the maximum allowed injected current.

Symbol Ratings Max. Unit Total current into V_{DD} power lines (source)⁽¹⁾ 240 I_{VDD} Total current out of V_{SS} ground lines (sink)⁽¹⁾ I_{VSS} 240 Output current sunk by any I/O and control pin I_{IO} Output current source by any I/Os and control pin 25 mΑ Injected current on five-volt tolerant I/O(3) -5/+0 I_{INJ(PIN)} (2) Injected current on any other pin⁽⁴⁾ ±5 $\Sigma I_{\text{INJ(PIN)}}^{(4)}$ Total injected current (sum of all I/O and control pins)⁽⁵⁾ ±25

Table 12. Current characteristics

- All main power (V_{DD}, V_{DDA}) and ground (V_{SS}, V_{SSA}) pins must always be connected to the external power supply, in the permitted range.
- 2. Negative injection disturbs the analog performance of the device. See note in Section 5.3.21: 12-bit ADC characteristics
- 3. Positive injection is not possible on these I/Os. A negative injection is induced by $V_{IN} < V_{SS}$. $I_{INJ(PIN)}$ must never be exceeded. Refer to *Table 11* for the values of the maximum allowed input voltage.
- 4. A positive injection is induced by $V_{IN} > V_{DD}$ while a negative injection is induced by $V_{IN} < V_{SS}$. $I_{INJ(PIN)}$ must never be exceeded. Refer to *Table 11* for the values of the maximum allowed input voltage.
- 5. When several inputs are submitted to a current injection, the maximum ΣI_{INJ(PIN)} is the absolute sum of the positive and negative injected currents (instantaneous values).

Table 13. Thermal characteristics

Symbol	Ratings	Value	Unit
T _{STG}	Storage temperature range	-65 to +150	°C
T _J	Maximum junction temperature	125	°C

5.3 Operating conditions

5.3.1 General operating conditions

Table 14. General operating conditions

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f	Internal AHB clock frequency	VOS bit in PWR_CR register = 0 ⁽¹⁾	0	-	144	
f _{HCLK}	internal Arib clock frequency	VOS bit in PWR_CR register= 1	0	-	168	MHz
f _{PCLK1}	Internal APB1 clock frequency	-	0	-	42	IVITZ
f _{PCLK2}	Internal APB2 clock frequency	-	0	-	84	
V _{DD}	Standard operating voltage	-	1.8 ⁽²⁾	-	3.6	V
\/ (3)(4)	Analog operating voltage (ADC limited to 1.2 M samples)	Must be the same potential as	1.8 ⁽²⁾	-	2.4	V
V _{DDA} ⁽³⁾⁽⁴⁾	Analog operating voltage (ADC limited to 1.4 M samples)	V _{DD} ⁽⁵⁾	2.4	ı	3.6	V
V _{BAT}	Backup operating voltage	-	1.65	-	3.6	V



Table 14. General operating conditions (continued)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
	Regulator ON:	VOS bit in PWR_CR register = 0 ⁽¹⁾ Max frequency 144MHz	1.08	1.14	1.20	V
V ₁₂	1.2 V internal voltage on V _{CAP_1} /V _{CAP_2} pins	VOS bit in PWR_CR register= 1 Max frequency 168MHz	1.20	1.26	1.32	V
12	Regulator OFF:	Max frequency 144MHz	1.10	1.14	1.20	V
	1.2 V external voltage must be supplied from external regulator on V _{CAP_1} /V _{CAP_2} pins	Max frequency 168MHz	1.20	1.26	1.30	V
	Input voltage on RST and FT	$2 \text{ V} \le \text{V}_{DD} \le 3.6 \text{ V}$	-0.3	-	5.5	
	pins ⁽⁶⁾	V _{DD} ≤ 2 V	-0.3	-	5.2	
V_{IN}	Input voltage on TTa pins	-	-0.3	-	V _{DDA} + 0.3	
	Input voltage on B pin	-	-	-	5.5	
	Power dissipation at $T_A = 85$ °C	LQFP64	-	-	435	mW
		LQFP100	-	-	465	
D		LQFP144	-	-	500	
P_{D}	for suffix 6 or $T_A = 105$ °C for suffix $7^{(7)}$	LQFP176	-	-	526	
		UFBGA176	-	-	513	
		WLCSP90	-	-	543	
	Ambient temperature for 6 suffix	Maximum power dissipation	-40	-	85	°C
TA	version	Low-power dissipation ⁽⁸⁾	-40	-	105	
IA	Ambient temperature for 7 suffix	Maximum power dissipation	-40	-	105	°C
	version	Low-power dissipation ⁽⁸⁾	-40	-	125	
TJ	Junction temperature range	6 suffix version	-40	-	105	°C
IJ	Junction temperature range	7 suffix version	-40	-	125	

The average expected gain in power consumption when VOS = 0 compared to VOS = 1 is around 10% for the whole temperature range, when the system clock frequency is between 30 and 144 MHz.

8. In low-power dissipation state, T_A can be extended to this range as long as T_J does not exceed T_{Jmax} .

V_{DD}/V_{DDA} minimum value of 1.7 V is obtained when the device operates in reduced temperature range, and with the use of an external power supply supervisor (refer to Section: Internal reset OFF).

^{3.} When the ADC is used, refer to Table 67: ADC characteristics.

^{4.} If V_{REF+} pin is present, it must respect the following condition: V_{DDA} - V_{REF+} < 1.2 V.

^{5.} It is recommended to power V_{DD} and V_{DDA} from the same source. A maximum difference of 300 mV between V_{DD} and V_{DDA} can be tolerated during power-up and power-down operation.

^{6.} To sustain a voltage higher than V_{DD} +0.3, the internal pull-up and pull-down resistors must be disabled.

^{7.} If T_A is lower, higher P_D values are allowed as long as T_J does not exceed T_{Jmax} .

Table 15. Limitations depending on the operating power supply range

Operating power supply range	ADC operation	Maximum Flash memory access frequency with no wait state (f _{Flashmax})	Maximum Flash memory access frequency with wait states ⁽¹⁾ (2)	I/O operation	Clock output Frequency on I/O pins	Possible Flash memory operations
V _{DD} =1.8 to 2.1 V ⁽³⁾	Conversion time up to 1.2 Msps	20 MHz ⁽⁴⁾	160 MHz with 7 wait states	Degraded speed performanceNo I/O compensation	up to 30 MHz	8-bit erase and program operations only
V _{DD} = 2.1 to 2.4 V	Conversion time up to 1.2 Msps	22 MHz	168 MHz with 7 wait states	Degraded speed performanceNo I/O compensation	up to 30 MHz	16-bit erase and program operations
V _{DD} = 2.4 to 2.7 V	Conversion time up to 2.4 Msps	24 MHz	168 MHz with 6 wait states	Degraded speed performanceI/O compensation works	up to 48 MHz	16-bit erase and program operations
$V_{DD} = 2.7 \text{ to}$ 3.6 $V^{(5)}$	Conversion time up to 2.4 Msps	30 MHz	168 MHz with 5 wait states	Full-speed operationI/O compensation works	 up to 60 MHz when V_{DD} = 3.0 to 3.6 V up to 48 MHz when V_{DD} = 2.7 to 3.0 V 	32-bit erase and program operations

^{1.} It applies only when code executed from Flash memory access, when code executed from RAM, no wait state is required.

^{2.} Thanks to the ART accelerator and the 128-bit Flash memory, the number of wait states given here does not impact the execution speed from Flash memory since the ART accelerator allows to achieve a performance equivalent to 0 wait state program execution.

^{3.} V_{DD}/VDDA minimum value of 1.7 V is obtained when the device operates in reduced temperature range, and with the use of an external power supply supervisor (refer to Section : Internal reset OFF).

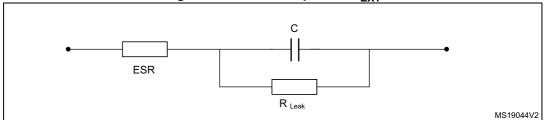
^{4.} Prefetch is not available. Refer to AN3430 application note for details on how to adjust performance and power.

^{5.} The voltage range for OTG USB FS can drop down to 2.7 V. However it is degraded between 2.7 and 3 V.

5.3.2 V_{CAP_1}/V_{CAP_2} external capacitor

Stabilization for the main regulator is achieved by connecting an external capacitor C_{EXT} to the $V_{CAP\ 1}/V_{CAP\ 2}$ pins. C_{EXT} is specified in *Table 16*.

Figure 23. External capacitor C_{EXT}



1. Legend: ESR is the equivalent series resistance.

Table 16. V_{CAP 1}/V_{CAP 2} operating conditions⁽¹⁾

Symbol	Parameter	Conditions
CEXT	Capacitance of external capacitor	2.2 μF
ESR	ESR of external capacitor	< 2 Ω

^{1.} When bypassing the voltage regulator, the two 2.2 μ F V_{CAP} capacitors are not required and should be replaced by two 100 nF decoupling capacitors.

5.3.3 Operating conditions at power-up / power-down (regulator ON)

Subject to general operating conditions for T_A.

Table 17. Operating conditions at power-up / power-down (regulator ON)

Symbol	Parameter	Min	Max	Unit
	V _{DD} rise time rate	20	8	µs/V
^t ∨DD	V _{DD} fall time rate	20	8	μ5/ ν

5.3.4 Operating conditions at power-up / power-down (regulator OFF)

Subject to general operating conditions for T_A.

Table 18. Operating conditions at power-up / power-down (regulator OFF)⁽¹⁾

Symbol	Parameter	Conditions	Min	Max	Unit
+	V _{DD} rise time rate	Power-up	20	8	
t _{VDD}	V _{DD} fall time rate	Power-down	20	8	
t	V _{CAP_1} and V _{CAP_2} rise time rate	Power-up	20	8	μs/V
t _{VCAP}	V_{CAP_1} and V_{CAP_2} fall time rate	Power-down	20		

To reset the internal logic at power-down, a reset must be applied on pin PA0 when V_{DD} reach below minimum value of V₁₂.

5.3.5 Embedded reset and power control block characteristics

The parameters given in *Table 19* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 14*.

Table 19. Embedded reset and power control block characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
		PLS[2:0]=000 (rising edge)	2.09	2.14	2.19	V
		PLS[2:0]=000 (falling edge)	1.98	2.04	2.08	>
		PLS[2:0]=001 (rising edge)	2.23	2.30	2.37	V
		PLS[2:0]=001 (falling edge)	2.13	2.19	2.25	>
		PLS[2:0]=010 (rising edge)	2.39	2.45	2.51	>
		PLS[2:0]=010 (falling edge)	2.29	2.35	2.39	>
		PLS[2:0]=011 (rising edge)	2.54	2.60	2.65	V
V _{PVD}	Programmable voltage detector level selection	PLS[2:0]=011 (falling edge)	2.44	2.51	2.56	٧
		PLS[2:0]=100 (rising edge)	2.70	2.76	2.82	V
		PLS[2:0]=100 (falling edge)	2.59	2.66	2.71	V
		PLS[2:0]=101 (rising edge)	2.86	2.93	2.99	V
		PLS[2:0]=101 (falling edge)	2.65	2.84	2.92	V
		PLS[2:0]=110 (rising edge)	2.96	3.03	3.10	V
		PLS[2:0]=110 (falling edge)	2.85	2.93	2.99	V
		PLS[2:0]=111 (rising edge)	3.07	3.14	3.21	V
		PLS[2:0]=111 (falling edge)	2.95	3.03	3.09	V
V _{PVDhyst} ⁽¹⁾	PVD hysteresis	-	-	100	-	mV
V	Power-on/power-down	Falling edge	1.60	1.68	1.76	V
$V_{POR/PDR}$	reset threshold	Rising edge	1.64	1.72	1.80	V
V _{PDRhyst} ⁽¹⁾	PDR hysteresis	-	-	40	-	mV
V _{BOR1}	Brownout level 1	Falling edge	2.13	2.19	2.24	V
*BOR1	threshold	Rising edge	2.23	2.29	2.33	V

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V	Brownout level 2	Falling edge	2.44	2.50	2.56	V
V _{BOR2}	threshold	Rising edge	2.53	2.59	2.63	V
V	Brownout level 3	Falling edge	2.75	2.83	2.88	V
V _{BOR3}	threshold	Rising edge	2.85	2.92	2.97	V
V _{BORhyst} ⁽¹⁾	BOR hysteresis	-	-	100	-	mV
T _{RSTTEMPO} ⁽¹⁾⁽²⁾	Reset temporization	-	0.5	1.5	3.0	ms
I _{RUSH} ⁽¹⁾	InRush current on voltage regulator power-on (POR or wakeup from Standby)	-	-	160	200	mA
E _{RUSH} ⁽¹⁾	InRush energy on voltage regulator power-on (POR or wakeup from Standby)	V _{DD} = 1.8 V, T _A = 105 °C, I _{RUSH} = 171 mA for 31 μs	-	-	5.4	μC

Table 19. Embedded reset and power control block characteristics (continued)

5.3.6 Supply current characteristics

The current consumption is a function of several parameters and factors such as the operating voltage, ambient temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code.

The current consumption is measured as described in *Figure 22: Current consumption measurement scheme*.

All Run mode current consumption measurements given in this section are performed using a CoreMark-compliant code.

Typical and maximum current consumption

The MCU is placed under the following conditions:

- At startup, all I/O pins are configured as analog inputs by firmware.
- All peripherals are disabled except if it is explicitly mentioned.
- The Flash memory access time is adjusted to f_{HCLK} frequency (0 wait state from 0 to 30 MHz, 1 wait state from 30 to 60 MHz, 2 wait states from 60 to 90 MHz, 3 wait states from 90 to 120 MHz, 4 wait states from 120 to 150 MHz, and 5 wait states from 150 to 168 MHz).
- When the peripherals are enabled HCLK is the system clock, f_{PCLK1} = f_{HCLK}/4, and f_{PCLK2} = f_{HCLK}/2, except is explicitly mentioned.
- The maximum values are obtained for $V_{DD} = 3.6 \text{ V}$ and maximum ambient temperature (T_A) , and the typical values for $T_A = 25 \,^{\circ}\text{C}$ and $V_{DD} = 3.3 \,^{\circ}\text{V}$ unless otherwise specified.

^{1.} Guaranteed by design.

^{2.} The reset temporization is measured from the power-on (POR reset or wakeup from V_{BAT}) to the instant when first instruction is read by the user application code.

Table 20. Typical and maximum current consumption in Run mode, code with data processing running from Flash memory (ART accelerator enabled) or RAM ⁽¹⁾

				Тур	Ма	x ⁽²⁾	
Symbol	Parameter	Conditions	f _{HCLK}	T _A = 25 °C	T _A = 85 °C	T _A = 105 °C	Unit
			168 MHz	87	102	109	
			144 MHz	67	80	86	
			120 MHz	56	69	75	
			90 MHz	44	56	62	
		(2)	60 MHz	30	42	49	
		External clock ⁽³⁾ , all peripherals enabled ⁽⁴⁾⁽⁵⁾	30 MHz	16	28	35	
		-	25 MHz	12	24	31	
			16 MHz ⁽⁶⁾	9	20	28	mA
			8 MHz	5	17	24	
			4 MHz	3	15	22	
	Supply current in		2 MHz	2	14	21	
I _{DD}	Run mode		168 MHz	40	54	61	
			144 MHz	31	43	50	
			120 MHz	26	38	45	
			90 MHz	20	32	39	
		(2)	60 MHz	14	26	33	
		External clock ⁽³⁾ , all peripherals disabled ⁽⁴⁾⁽⁵⁾	30 MHz	8	20	27	
		poripriorale aleastea	25 MHz	6	18	25	
			16 MHz ⁽⁶⁾	5	16	24	
			8 MHz	3	15	22	
			4 MHz	2	14	21	
			2 MHz	2	14	21	

- 1. Code and data processing running from SRAM1 using boot pins.
- 2. Guaranteed by characterization, tested in production at V_{DD} max and f_{HCLK} max with peripherals enabled.
- 3. External clock is 4 MHz and PLL is on when $f_{HCLK} > 25$ MHz.
- 4. When the ADC is ON (ADON bit set in the ADC_CR2 register), add an additional power consumption of 1.6 mA per ADC for the analog part.
- 5. When analog peripheral blocks such as ADCs, DACs, HSE, LSE, HSI, or LSI are ON, an additional power consumption should be considered.
- 6. In this case HCLK = system clock/2.

Table 21. Typical and maximum current consumption in Run mode, code with data processing running from Flash memory (ART accelerator disabled)

Symala al	Parameter	Conditions	£	Тур	Ma	ax ⁽¹⁾	Unit	
Symbol	Parameter	Conditions	f _{HCLK}	T _A = 25 °C	T _A = 85 °C	T _A = 105 °C		
			168 MHz	93	109	117		
			144 MHz	76	89	96		
			120 MHz	67	79	86		
		External clock ⁽²⁾ , all peripherals enabled ⁽³⁾⁽⁴⁾	90 MHz	53	65	73		
			60 MHz	37	49	56		
			30 MHz	20	32	39		
			25 MHz	16	27	35		
			16 MHz	11	23	30	· mA	
			8 MHz	6	18	25		
			4 MHz	4	16	23		
	Supply current		2 MHz	3	15	22		
I _{DD}	in Run mode		168 MHz	46	61	5 23 5 22 1 69 2 60	IIIA	
			16 MHz 11 23 3 8 MHz 6 18 2 4 MHz 4 16 2 2 MHz 3 15 2 168 MHz 46 61 6 144 MHz 40 52 6 120 MHz 37 48 5 90 MHz 30 42 5	60				
			120 MHz	37	48	56		
			90 MHz	30	42	50		
		External clock ⁽²⁾ ,	60 MHz	22	33	41		
		all peripherals disabled ⁽³⁾⁽⁴⁾	30 MHz	12	24	31		
		disabled ⁽³⁾⁽⁴⁾	25 MHz	10	21	29	-	
			16 MHz	7	19	26		
			8 MHz	4	16	23		
			4 MHz	3	15	22		
			2 MHz	2	14	21		

^{1.} Guaranteed by characterization, tested in production at V_{DD} max and f_{HCLK} max with peripherals enabled.

^{2.} External clock is 4 MHz and PLL is on when f_{HCLK} > 25 MHz.

^{3.} When analog peripheral blocks such as (ADCs, DACs, HSE, LSE, HSI,LSI) are on, an additional power consumption should be considered

^{4.} When the ADC is ON (ADON bit set in the ADC_CR2 register), add an additional power consumption of 1.6 mA per ADC for the analog part.

Figure 24. Typical current consumption versus temperature, Run mode, code with data processing running from Flash (ART accelerator ON) or RAM, and peripherals OFF

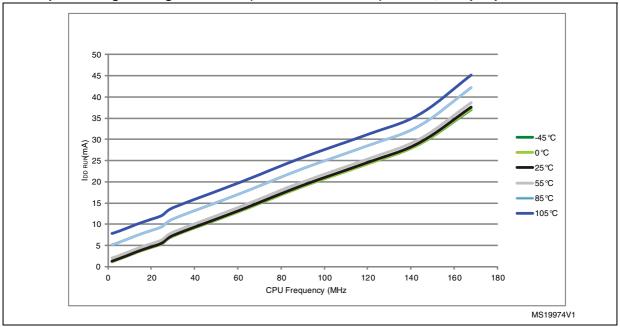


Figure 25. Typical current consumption versus temperature, Run mode, code with data processing running from Flash (ART accelerator ON) or RAM, and peripherals ON

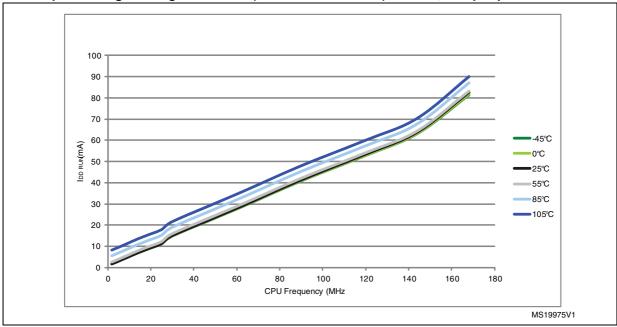


Figure 26. Typical current consumption versus temperature, Run mode, code with data processing running from Flash (ART accelerator OFF) or RAM, and peripherals OFF

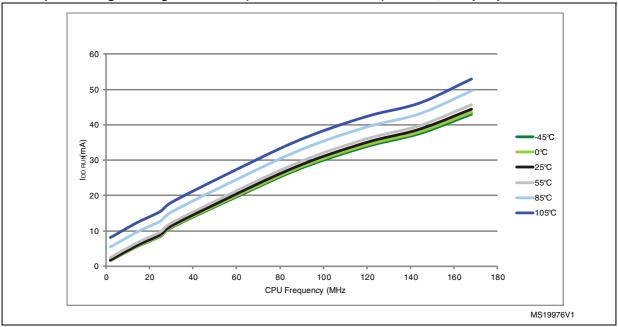


Figure 27. Typical current consumption versus temperature, Run mode, code with data processing running from Flash (ART accelerator OFF) or RAM, and peripherals ON

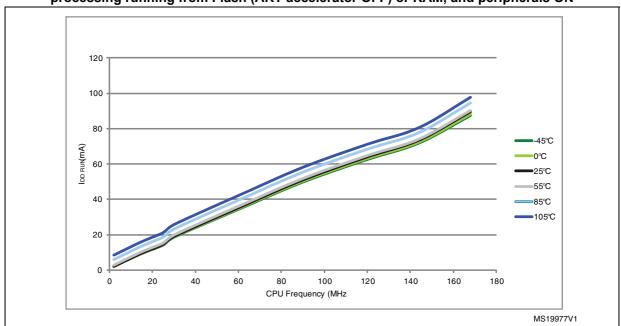


Table 22. Typical and maximum current consumption in Sleep mode

		ypiour una maximum our		Тур	Max		
Symbol	Parameter	Conditions	f _{HCLK}	T _A = 25 °C	T _A = 85 °C	T _A = 105 °C	Unit
			168 MHz	59	77	84	
			144 MHz	46	61	67	
		120 MHz	38	53	60		
	External clock ⁽²⁾ , all peripherals enabled ⁽³⁾	90 MHz	30	44	51		
		60 MHz	20	34	41		
		30 MHz	11	24	31		
		25 MHz	8	21	28		
		16 MHz	6	18	25		
			8 MHz	3	16	23	- mA
			4 MHz	2	15	22	
.	Supply current in		2 MHz	2	14	21	
I _{DD}	Sleep mode		168 MHz	12	27	35	
			144 MHz	9	22	29	
			120 MHz	8	20	28	
			90 MHz	7	19	26	
		(0)	60 MHz	5	17	24	1
		External clock ⁽²⁾ , all peripherals disabled	30 MHz	3	16	23	
		periprierale aleasiea	25 MHz	2	15	22	-
			16 MHz	2	14	21	
			8 MHz	1	14	21	
			4 MHz	1	13	21	
			2 MHz	1	13	21	

^{1.} Guaranteed by characterization, tested in production at V_{DD} max and f_{HCLK} max with peripherals enabled.

^{2.} External clock is 4 MHz and PLL is on when f_{HCLK} > 25 MHz.

^{3.} Add an additional power consumption of 1.6 mA per ADC for the analog part. In applications, this consumption occurs only while the ADC is ON (ADON bit is set in the ADC_CR2 register).

Table 23. Typical and maximum current consumptions in Stop mode

			Тур		Max		
Symbol	Parameter	Conditions	T _A = 25 °C	T _A = 25 °C	T _A = 85 °C	T _A = 105 °C	Unit
I _{DD_STOP}	Supply current in Stop mode with main regulator in Run mode	Flash in Stop mode, low-speed and high- speed internal RC oscillators and high-speed oscillator OFF (no independent watchdog)	0.45	1.5	11.00	20.00	
		Flash in Deep power-down mode, low-speed and high-speed internal RC oscillators and high-speed oscillator OFF (no independent watchdog)	0.40	1.5	11.00	20.00	mA
	Supply current in Stop mode with main regulator in Low-power mode	Flash in Stop mode, low-speed and high- speed internal RC oscillators and high-speed oscillator OFF (no independent watchdog)	0.31	1.1	8.00	15.00	IIIA
		Flash in Deep power-down mode, low-speed and high-speed internal RC oscillators and high-speed oscillator OFF (no independent watchdog)	0.28	1.1	8.00	15.00	

Table 24. Typical and maximum current consumptions in Standby mode

			Тур			Ма		
Symbol	Parameter	Conditions	T _A = 25 °C			T _A = 85 °C	T _A = 105 °C	Unit
			V _{DD} = 1.8 V	V _{DD} = 2.4 V	V _{DD} = 3.3 V	V _{DD} = 3.6 V		
		Backup SRAM ON, low- speed oscillator and RTC ON	3.0	3.4	4.0	20	36	
	Supply current	Backup SRAM OFF, low- speed oscillator and RTC ON	2.4	2.7	3.3	16	32	
I _{DD_STBY} in Standby mode		Backup SRAM ON, RTC OFF	2.4	2.6	3.0	12.5	24.8	μА
		Backup SRAM OFF, RTC OFF	1.7	1.9	2.2	9.8	19.2	

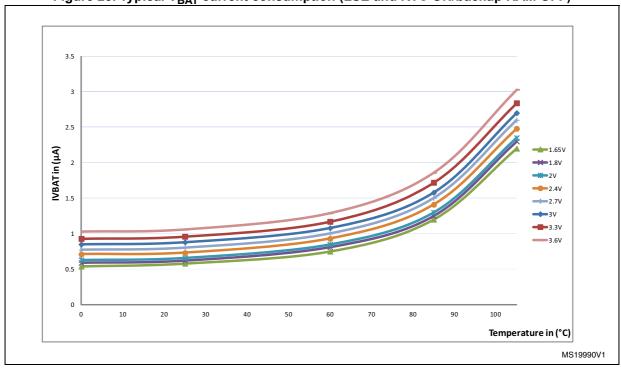
^{1.} Guaranteed by characterization.

Table 25. Typical and maximum current consumptions in V_{BAT} mode

Symbol Parameter				Тур			Max ⁽¹⁾		
	Parameter	Conditions	T _A = 25 °C			T _A = 85 °C	T _A = 105 °C	Unit	
			V _{BAT} = 1.8 V	V _{BAT} = 2.4 V	V _{BAT} = 3.3 V	V _{BAT} = 3.6 V			
I _{DD_VBA} domain T supply current	Backup SRAM ON, low-speed oscillator and RTC ON	1.29	1.42	1.68	6	11			
	Backup SRAM OFF, low-speed oscillator and RTC ON	0.62	0.73	0.96	3	5	μΑ		
	Backup SRAM ON, RTC OFF	0.79	0.81	0.86	5	10			
	Backup SRAM OFF, RTC OFF	0.10	0.10	0.10	2	4			

^{1.} Guaranteed by characterization.

Figure 28. Typical V_{BAT} current consumption (LSE and RTC ON/backup RAM OFF)



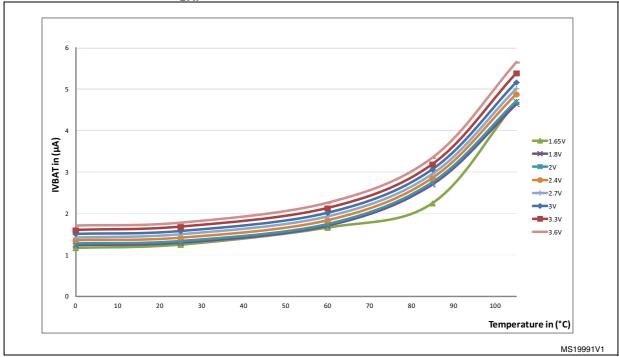


Figure 29. Typical V_{BAT} current consumption (LSE and RTC ON/backup RAM ON)

Additional current consumption

The MCU is placed under the following conditions:

- All I/O pins are configured in analog mode.
- The Flash memory access time is adjusted to f_{HCLK} frequency.
- The voltage scaling is adjusted to f_{HCLK} frequency as follows:
 - Scale 2 for f_{HCLK} ≤ 144 MHz
 - Scale 1 for 144 MHz < f_{HCLK} ≤ 168 MHz.
- The system clock is HCLK, f_{PCLK1} = f_{HCLK}/4, and f_{PCLK2} = f_{HCLK}/2.
- The HSE crystal clock frequency is 25 MHz.
- T_Δ= 25 °C.

Table 26. Typical current consumption in Run mode, code with data processing running from Flash memory, regulator ON (ART accelerator enabled except prefetch), $V_{DD} = 1.8 \text{ V}^{(1)}$

except prefetcing, V _{DD} = 1.0 V								
Symbol	Parameter	Conditions	f _{HCLK} (MHz)	Typ. at T _A = 25 °C	Unit			
			160	36.2				
IDD		All peripheral disabled	144	29.3	mA			
			120	24.7				
			90	19.3				
			60	13.4				
			30	7.7				
			25	6.0				

When peripherals are enabled, the power consumption corresponding to the analog part of the peripherals (such as ADC or DAC) is not included.

I/O system current consumption

The current consumption of the I/O system has two components: static and dynamic.

I/O static current consumption

All the I/Os used as inputs with pull-up generate current consumption when the pin is externally held low. The value of this current consumption can be simply computed by using the pull-up/pull-down resistors values given in *Table 48: I/O static characteristics*.

For the output pins, any external pull-down or external load must also be considered to estimate the current consumption.

Additional I/O current consumption is due to I/Os configured as inputs if an intermediate voltage level is externally applied. This current consumption is caused by the input Schmitt trigger circuits used to discriminate the input value. Unless this specific configuration is required by the application, this supply current consumption can be avoided by configuring these I/Os in analog mode. This is notably the case of ADC input pins which should be configured as analog inputs.

Caution:

Any floating input pin can also settle to an intermediate voltage level or switch inadvertently, as a result of external electromagnetic noise. To avoid current consumption related to

floating pins, they must either be configured in analog mode, or forced internally to a definite digital value. This can be done either by using pull-up/down resistors or by configuring the pins in output mode.

I/O dynamic current consumption

In addition to the internal peripheral current consumption measured previously (see *Table 28: Peripheral current consumption*), the I/Os used by an application also contribute to the current consumption. When an I/O pin switches, it uses the current from the MCU supply voltage to supply the I/O pin circuitry and to charge/discharge the capacitive load (internal or external) connected to the pin:

$$I_{SW} = V_{DD} \times f_{SW} \times C$$

where

 I_{SW} is the current sunk by a switching I/O to charge/discharge the capacitive load

V_{DD} is the MCU supply voltage

 f_{SW} is the I/O switching frequency

C is the total capacitance seen by the I/O pin: $C = C_{INT} + C_{EXT}$

The test pin is configured in push-pull output mode and is toggled by software at a fixed frequency.

Table 27. Switching output I/O current consumption

Symbol	Parameter	Conditions ⁽¹⁾	I/O toggling frequency (f _{SW})	Тур	Unit
			2 MHz	0.02	
		$V_{DD} = 3.3 V^{(2)}$	8 MHz	0.14	
		$C = C_{INT}$	25 MHz	0.51	
			50 MHz	0.86	
			60 MHz	1.30	
			2 MHz	0.10	
		V _{DD} = 3.3 V	8 MHz	0.38	
		C _{EXT} = 0 pF	25 MHz	1.18	
		$C = C_{INT} + C_{EXT} + C_{S}$	50 MHz	2.47	mA
			60 MHz	2.86	
		V_{DD} = 3.3 V C_{EXT} = 10 pF C = C_{INT} + C_{EXT} + C_{S}	2 MHz	0.17	
	I/O switching current		8 MHz	0.66	
I _{DDIO}			25 MHz	1.70	
			50 MHz	2.65	
			60 MHz	3.48	
			2 MHz	0.23	
		V _{DD} = 3.3 V	8 MHz	0.95	
		C _{EXT} = 22 pF	25 MHz	3.20	
		$C = C_{INT} + C_{EXT} + C_{S}$	50 MHz	4.69	
			60 MHz	8.06	
			2 MHz	0.30	
		V _{DD} = 3.3 V	8 MHz	1.22	
		C _{EXT} = 33 pF	25 MHz	3.90	
		$C = C_{INT} + C_{EXT} + C_{S}$	50 MHz	8.82	
			60 MHz	_(3)	

^{1.} C_S is the PCB board capacitance including the pad pin. C_S = 7 pF (estimated value).

^{2.} This test is performed by cutting the LQFP package pin (pad removal).

^{3.} At 60 MHz, C maximum load is specified 30 pF.

On-chip peripheral current consumption

The current consumption of the on-chip peripherals is given in *Table 28*. The MCU is placed under the following conditions:

- At startup, all I/O pins are configured as analog pins by firmware.
- All peripherals are disabled unless otherwise mentioned
- The code is running from Flash memory and the Flash memory access time is equal to 5 wait states at 168 MHz.
- The code is running from Flash memory and the Flash memory access time is equal to 4 wait states at 144 MHz, and the power scale mode is set to 2.
- The ART accelerator is ON.
- The given value is calculated by measuring the difference of current consumption
 - with all peripherals clocked off
 - with one peripheral clocked on (with only the clock applied)
- When the peripherals are enabled: HCLK is the system clock, f_{PCLK1} = f_{HCLK}/4, and f_{PCLK2} = f_{HCLK}/2.
- The typical values are obtained for V_{DD} = 3.3 V and T_A = 25 °C, unless otherwise specified.

Table 28. Peripheral current consumption

		I _{DD} (7	Гур) ⁽¹⁾	
Perip	pheral	Scale1 (up t 168 MHz)	Scale2 (up to 144 MHz)	Unit
	GPIOA	2.70	2.40	
	GPIOB	2.50	2.22	
	GPIOC	2.54	2.28	1
	GPIOD	2.55	2.28	1
	GPIOE	2.68	2.40	1
	GPIOF	2.53	2.28	
	GPIOG	2.51	2.22	1
	GPIOH	2.51	2.22	1
AHB1	GPIOI	2.50	2.22	μΑ/MHz
(up to 168 MHz)	OTG_HS+ULPI	28.33	25.38	1
	CRC	0.41	0.40	1
	BKPSRAM	0.63	0.58	
	DMA1	37.44	33.58	1
	DMA2	37.69	33.93	1
	ETH_MAC ETH_MAC_TX ETH_MAC_RX ETH_MAC_PTP	20.43	18.39	

Table 28. Peripheral current consumption (continued)

		I _{DD} (T		
Peripheral		Scale1 (up t 168 MHz)	Scale2 (up to 144 MHz)	Unit
	OTG_FS	26.45	26.67	
	DCMI	5.87	5.35	
AHB2 (up to 168 MHz)	RNG	1.50	1.67	μΑ/MHz
(4) 10 100 11112)	Hash	9.73	8.86	
	Crypto	2.23	2.08	
AHB3 (up to 168 MHz)	FSMC	12.46	11.31	μΑ/MHz
Bus matrix ⁽²⁾		13.10	11.81	μΑ/MHz

Table 28. Peripheral current consumption (continued)

	•	I _{DD} (1	Гур) ⁽¹⁾	
Perip	heral	Scale1 (up t 168 MHz)	Scale2 (up to 144 MHz)	Unit
	TIM2	16.71	16.50	
	TIM3	12.33	11.94	
	TIM4	13.45	12.92	
	TIM5	17.14	16.58	
	TIM6	2.43	3.06	
	TIM7	2.43	2.22	
	TIM12	6.62	6.83	
	TIM13	5.05	5.47	
	TIM14	5.26	5.61	
	PWR	1.00	0.56	
	USART2	2.69	2.78	
	USART3	2.74	2.78	1
APB1 (up to 42 MHz)	UART4	3.24	3.33	μA/MHz
(45 to 12 Wi12)	UART5	2.69	2.78	1
	I2C1	2.67	2.50	
	I2C2	2.83	2.78	1
	I2C3	2.81	2.78	1
	SPI2	2.43	2.22]
	SPI3	2.43	2.22	
	I2S2 ⁽³⁾	2.43	2.22	
	I2S3 ⁽³⁾	2.26	2.22	
	CAN1	5.12	5.56]
	CAN2	4.81	5.28	
	DAC ⁽⁴⁾	1.67	1.67	
	WWDG	1.00	0.83	

 $I_{DD}(Typ)^{(1)}$ **Peripheral** Unit Scale1 Scale2 (up t 168 MHz) (up to 144 MHz) **SDIO** 7.08 7.92 16.79 TIM1 15.51 TIM8 17.88 16.53 7.28 TIM9 7.64 TIM₁₀ 4.89 4.82 TIM11 5.19 4.82 APB2 ADC1⁽⁵⁾ 4.67 4.58 µA/MHz (up to 84 MHz) ADC2⁽⁵⁾ 4.67 4.58 ADC3⁽⁵⁾ 4.43 4.44 SPI1 1.32 1.39 **USART1** 3.51 3.72 **USART6** 3.55 3.75 **SYSCFG** 0.74 0.56

Table 28. Peripheral current consumption (continued)

- 1. When the I/O compensation cell is ON, I_{DD} typical value increases by 0.22 mA.
- 2. The BusMatrix is automatically active when at least one master is ON.
- 3. To enable an I2S peripheral, first set the I2SMOD bit and then the I2SE bit in the SPI_I2SCFGR register.
- 4. When the DAC is ON and EN1/2 bits are set in DAC_CR register, add an additional power consumption of 0.8 mA per DAC channel for the analog part.
- 5. When the ADC is ON (ADON bit set in the ADC_CR2 register), add an additional power consumption of 1.6 mA per ADC for the analog part.

5.3.7 Wakeup time from low-power mode

The wakeup times given in *Table 29* is measured on a wakeup phase with a 16 MHz HSI RC oscillator. The clock source used to wake up the device depends from the current operating mode:

- Stop or Standby mode: the clock source is the RC oscillator
- Sleep mode: the clock source is the clock that was set before entering Sleep mode.

All timings are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 14*.

Symbol	Parameter	Min ⁽¹⁾	Typ ⁽¹⁾	Max ⁽¹⁾	Unit
t _{WUSLEEP} ⁽²⁾ Wakeup from Sleep mode		1	5	-	CPU clock cycle
	Wakeup from Stop mode (regulator in Run mode and Flash memory in Stop mode)	-	13	-	
t _{WUSTOP} ⁽²⁾	Wakeup from Stop mode (regulator in low-power mode and Flash memory in Stop mode)	-	17	40	116
WUSTOP` 7	Wakeup from Stop mode (regulator in Run mode and Flash memory in Deep power-down mode)	-	105	-	μs
	Wakeup from Stop mode (regulator in low-power mode and Flash memory in Deep power-down mode)	-	110	-	
t _{WUSTDBY} (2)(3)	Wakeup from Standby mode	260	375	480	μs

Table 29. Low-power mode wakeup timings

5.3.8 External clock source characteristics

High-speed external user clock generated from an external source

The characteristics given in *Table 30* result from tests performed using an high-speed external clock source, and under ambient temperature and supply voltage conditions summarized in *Table 14*.

Table 30. High-speed external user clock characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f _{HSE_ext}	External user clock source frequency ⁽¹⁾		1	-	50	MHz
V _{HSEH}	OSC_IN input pin high level voltage		0.7V _{DD}	-	V_{DD}	V
V _{HSEL}	OSC_IN input pin low level voltage	_	V_{SS}	ı	0.3V _{DD}	٧
$t_{w(HSE)} \ t_{w(HSE)}$	OSC_IN high or low time ⁽¹⁾		5	ı	-	ns
$t_{r(HSE)} \ t_{f(HSE)}$	OSC_IN rise or fall time ⁽¹⁾		-	-	10	113
C _{in(HSE)}	OSC_IN input capacitance ⁽¹⁾	-	-	5	-	pF
DuCy _(HSE)	Duty cycle	-	45		55	%
ال	OSC_IN Input leakage current	$V_{SS} \le V_{IN} \le V_{DD}$	-	-	±1	μΑ

^{1.} Guaranteed by design.

^{1.} Guaranteed by characterization.

^{2.} The wakeup times are measured from the wakeup event to the point in which the application code reads the first instruction.

^{3.} $t_{WUSTDBY}$ minimum and maximum values are given at 105 °C and –45 °C, respectively.

Low-speed external user clock generated from an external source

The characteristics given in Table 31 result from tests performed using an low-speed external clock source, and under ambient temperature and supply voltage conditions summarized in Table 14.

Table 31. Low-speed external user clock characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f _{LSE_ext}	User External clock source frequency ⁽¹⁾		-	32.768	1000	kHz
V _{LSEH}	OSC32_IN input pin high level voltage		0.7V _{DD}	-	V _{DD}	V
V _{LSEL}	OSC32_IN input pin low level voltage	-	V _{SS}	-	0.3V _{DD}	
$t_{w(LSE)} \ t_{f(LSE)}$	OSC32_IN high or low time ⁽¹⁾		450	-	-	ns
t _{r(LSE)}			-	-	50	113
C _{in(LSE)}	OSC32_IN input capacitance ⁽¹⁾	-	-	5	-	pF
DuCy _(LSE)	Duty cycle	-	30	-	70	%
ΙL	OSC32_IN Input leakage current	$V_{SS}\!\leq\!V_{IN}\!\leq\!V_{DD}$	-	-	±1	μΑ

^{1.} Guaranteed by design.

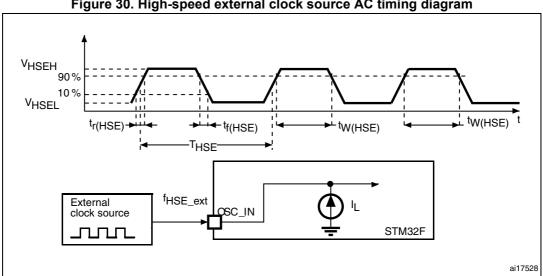


Figure 30. High-speed external clock source AC timing diagram

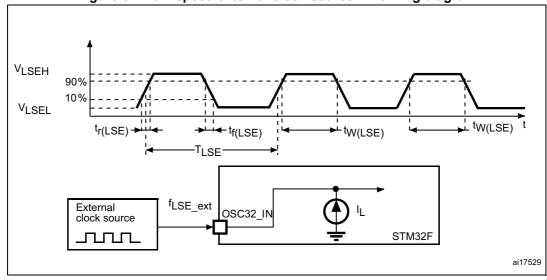


Figure 31. Low-speed external clock source AC timing diagram

High-speed external clock generated from a crystal/ceramic resonator

The high-speed external (HSE) clock can be supplied with a 4 to 26 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in *Table 32*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f _{OSC_IN}	Oscillator frequency	-	4	-	26	MHz
R _F	Feedback resistor	-	-	200	-	kΩ
G _m	Oscillator transconductance	Startun	5	-	-	mA/V
G _{mcritmax}	Maximum critical crystal G _m	Startup	-	-	1	IIIAV V
t _{SU(HSE)} ⁽²⁾	Startup time	V _{DD} is stabilized	-	2	-	ms

Table 32. HSE 4-26 MHz oscillator characteristics (1)

For C_{L1} and C_{L2} , it is recommended to use high-quality external ceramic capacitors in the 5 pF to 25 pF range (typ.), designed for high-frequency applications, and selected to match the requirements of the crystal or resonator (see *Figure 32*). C_{L1} and C_{L2} are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of C_{L1} and C_{L2} . PCB and MCU pin capacitance must be included (10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing C_{L1} and C_{L2} .

^{1.} Guaranteed by design.

Guaranteed by characterization. t_{SU(HSE)} is the startup time measured from the moment it is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is measured for a standard crystal resonator and can vary significantly with the crystal manufacturer

Note:

For information on electing the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

Resonator with integrated capacitors

CL1

STM32F

RESONATOR WITH INTEGRATION OF THE PROPERTY
Figure 32. Typical application with an 8 MHz crystal

R_{EXT} value depends on the crystal characteristics.

Low-speed external clock generated from a crystal/ceramic resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in *Table 33*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

	Table 66: Lot 656matel characteristics (ILSE 62:1 66 KHZ)									
Symbol	Parameter	Conditions	Min	Тур	Max	Unit				
f _{OSC_IN}	Oscillator frequency	-	-	32.768	-	MHz				
R _F	Feedback resistor	-	-	18.4	-	ΜΩ				
I _{DD}	LSE current consumption	-	-	-	1	μA				
G _m	Oscillator transconductance	Startup	2.8	-	-	μΑ/V				
G _{mcritmax}	Maximum critical crystal G _m	Startup	-	-	0.56	μΑVV				
t _{SU(LSE)} ⁽²⁾	startup time	V _{DD} is stabilized	-	2	-	s				

Table 33. LSE oscillator characteristics ($f_{LSE} = 32.768 \text{ kHz}$) (1)

Note:

For information on electing the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

^{1.} Guaranteed by design.

Guaranteed by characterization. t_{SU(LSE)} is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768 kHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer

Resonator with integrated capacitors

CL1
OSC32
N
Bias controlled gain
STM32F
ai17531

Figure 33. Typical application with a 32.768 kHz crystal

5.3.9 Internal clock source characteristics

The parameters given in *Table 34* and *Table 35* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 14*.

High-speed internal (HSI) RC oscillator

Table 34. HSI oscillator characteristics (1)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f _{HSI}	Frequency	-	-	16	-	MHz
ACC	HSI user trimming step ⁽²⁾	-	-	-	1	%
	Accuracy of the HSI oscillator	$T_A = -40 \text{ to } 105 ^{\circ}\text{C}^{(3)}$	-8	-	4.5	%
ACC _{HSI}		$T_A = -10 \text{ to } 85 ^{\circ}\text{C}^{(3)}$	-4	-	4	%
		T _A = 25 °C ⁽⁴⁾	-1	-	1	%
t _{su(HSI)} (2)	HSI oscillator startup time	-	-	2.2	4	μs
I _{DD(HSI)} ⁽²⁾	HSI oscillator power consumption	-	-	60	80	μΑ

- 1. V_{DD} = 3.3 V, T_A = -40 to 105 °C unless otherwise specified.
- 2. Guaranteed by design.
- 3. Guaranteed by characterization.
- 4. Factory calibrated, parts not soldered.

Low-speed internal (LSI) RC oscillator

Table 35. LSI oscillator characteristics (1)

Symbol	Parameter	Min	Тур	Max	Unit
f _{LSI} ⁽²⁾	Frequency	17	32	47	kHz
t _{su(LSI)} (3)	LSI oscillator startup time	-	15	40	μs
I _{DD(LSI)} ⁽³⁾	LSI oscillator power consumption	ı	0.4	0.6	μA

- 1. V_{DD} = 3 V, T_A = -40 to 105 °C unless otherwise specified.
- 2. Guaranteed by characterization.
- 3. Guaranteed by design.

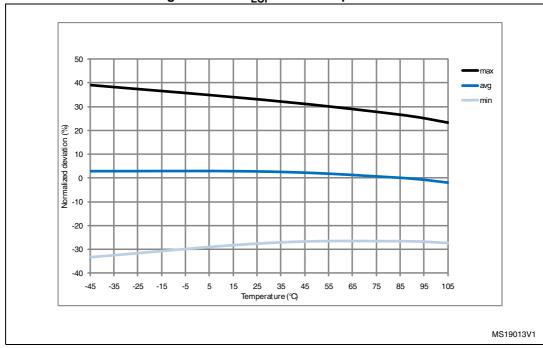


Figure 34. ACC_{LSI} versus temperature

5.3.10 PLL characteristics

The parameters given in *Table 36* and *Table 37* are derived from tests performed under temperature and V_{DD} supply voltage conditions summarized in *Table 14*.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f _{PLL_IN}	PLL input clock ⁽¹⁾	-	0.95 ⁽²⁾	1	2.10	MHz
f _{PLL_OUT}	PLL multiplier output clock	-	24	-	168	MHz
f _{PLL48_OUT}	48 MHz PLL multiplier output clock	-	-	48	75	MHz
f _{VCO_OUT}	PLL VCO output	-	100	-	432	MHz
+	PLL lock time	VCO freq = 100 MHz	75	-	200	116
^t LOCK	PLL IOCK (IIIIE	VCO freq = 432 MHz	100	-	300	μs

Table 36. Main PLL characteristics

Table 36. Main PLL characteristics (continued)

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
			RMS	-	25	-	
	Cycle-to-cycle jitter	System clock	peak to peak	1	±150	-	
		120 MHz	RMS	-	15	-	
Jitter ⁽³⁾	Period Jitter	t.	peak to peak	-	±200	-	ps
	Main clock output (MCO) for RMII Ethernet	Cycle to cycle at 50 MHz on 1000 samples		-	32	-	
	Main clock output (MCO) for MII Ethernet	Cycle to cycle at 29 on 1000 samples	-	40	-		
	Bit Time CAN jitter	Cycle to cycle at 1 MHz on 1000 samples		-	330	-	
I _{DD(PLL)} ⁽⁴⁾	PLL power consumption on VDD VCO freq = 100 MHz	·		0.15	-	0.40	mA
DD(FLL)	VCO freq = 432 MHz		0.45		0.75		
I _{DDA(PLL)} ⁽⁴⁾	PLL power consumption on VDDA	VCO freq = 100 MI VCO freq = 432 MI		0.30 0.55	-	0.40 0.85	mA

Take care of using the appropriate division factor M to obtain the specified PLL input clock values. The M factor is shared between PLL and PLLI2S.

Table 37. PLLI2S (audio PLL) characteristics

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
f _{PLLI2S_IN}	PLLI2S input clock ⁽¹⁾	-		0.95 ⁽²⁾	1	2.10	MHz
f _{PLLI2S_OUT}	PLLI2S multiplier output clock	-		-	-	216	MHz
f _{VCO_OUT}	PLLI2S VCO output	-		100	-	432	MHz
	PLLI2S lock time	VCO freq = 100 MHz	Z	75	-	200	
t _{LOCK}	PLLIZS IOCK LITTLE	VCO freq = 432 MHz			-	300	μs
	Master I ² S clock jitter	Cycle to cycle at	RMS	-	90	-	
		48KHz period,	peak to peak	-	±280	-	ps
Jitter ⁽³⁾	Master 1 3 Clock Jitter	Average frequency of 12.288 MHz N = 432, R = 5 on 1000 samples		-	90	-	ps
	WS I ² S clock jitter	Cycle to cycle at 48 KHz on 1000 samples		-	400	-	ps

^{2.} Guaranteed by design.

^{3.} The use of 2 PLLs in parallel could degraded the Jitter up to +30%.

^{4.} Guaranteed by characterization.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
I _{DD(PLLI2S)} (4)	PLLI2S power consumption on V _{DD}	VCO freq = 100 MHz VCO freq = 432 MHz	0.15 0.45	-	0.40 0.75	mA
I _{DDA(PLLI2S)} (4)	PLLI2S power consumption on V _{DDA}	VCO freq = 100 MHz VCO freq = 432 MHz	0.30 0.55	-	0.40 0.85	mA

Table 37. PLLI2S (audio PLL) characteristics (continued)

- 1. Take care of using the appropriate division factor M to have the specified PLL input clock values.
- 2. Guaranteed by design.
- 3. Value given with main PLL running.
- 4. Guaranteed by characterization.

5.3.11 PLL spread spectrum clock generation (SSCG) characteristics

The spread spectrum clock generation (SSCG) feature allows to reduce electromagnetic interferences (see *Table 44: EMI characteristics*). It is available only on the main PLL.

Table 38. SSCG parameters constraint

Symbol	Parameter	Min	Тур	Max ⁽¹⁾	Unit
f _{Mod}	Modulation frequency	-	-	10	KHz
md	Peak modulation depth	0.25	-	2	%
MODEPER * INCSTEP		-	-	2 ¹⁵ –1	-

^{1.} Guaranteed by design.

Equation 1

The frequency modulation period (MODEPER) is given by the equation below:

$$\texttt{MODEPER} = \mathsf{round}[f_{\mathsf{PLL}\ \mathsf{IN}}/(4 \times f_{\mathsf{Mod}})]$$

 $f_{PLL\ IN}$ and f_{Mod} must be expressed in Hz.

As an example:

If f_{PLL_IN} = 1 MHz, and f_{MOD} = 1 kHz, the modulation depth (MODEPER) is given by equation 1:

MODEPER = round
$$[10^6/(4 \times 10^3)] = 250$$

Equation 2

Equation 2 allows to calculate the increment step (INCSTEP):

INCSTEP = round[
$$((2^{15}-1) \times md \times PLLN)/(100 \times 5 \times MODEPER)$$
]

f_{VCO OUT} must be expressed in MHz.

With a modulation depth (md) = ± 2 % (4 % peak to peak), and PLLN = 240 (in MHz):

INCSTEP = round[
$$((2^{15}-1)\times2\times240)/(100\times5\times250)$$
] = 126md(quantitazed)%

An amplitude quantization error may be generated because the linear modulation profile is obtained by taking the quantized values (rounded to the nearest integer) of MODPER and INCSTEP. As a result, the achieved modulation depth is quantized. The percentage quantized modulation depth is given by the following formula:

$$\text{md}_{\text{quantized}}\% = (\text{MODEPER} \times \text{INCSTEP} \times 100 \times 5) / ((2^{15} - 1) \times \text{PLLN})$$

As a result:

$$md_{quantized}\% = (250 \times 126 \times 100 \times 5)/((2^{15} - 1) \times 240) = 2,002\%$$
(peak)

Figure 35 and Figure 36 show the main PLL output clock waveforms in center spread and down spread modes, where:

F0 is $f_{PLL\ OUT}$ nominal.

T_{mode} is the modulation period.

md is the modulation depth.

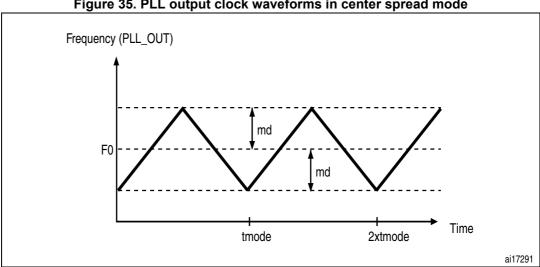


Figure 35. PLL output clock waveforms in center spread mode

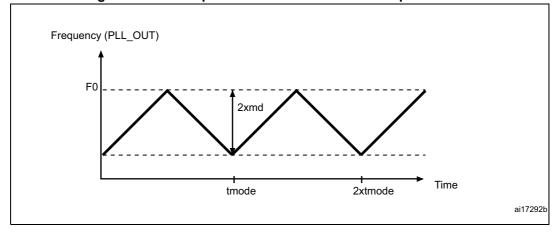


Figure 36. PLL output clock waveforms in down spread mode

5.3.12 Memory characteristics

Flash memory

The characteristics are given at T_A = -40 to 105 °C unless otherwise specified.

The devices are shipped to customers with the Flash memory erased.

Table 39. Flash memory characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
		Write / Erase 8-bit mode, V _{DD} = 1.8 V	-	5	-	
I_{DD}	I _{DD} Supply current	Write / Erase 16-bit mode, V_{DD} = 2.1 V	-	8	-	mA
		Write / Erase 32-bit mode, V_{DD} = 3.3 V	-	12	-	

Table 40. Flash memory programming

Symbol	Parameter	Conditions	Min ⁽¹⁾	Тур	Max ⁽¹⁾	Unit
t _{prog}	Word programming time	Program/erase parallelism (PSIZE) = x 8/16/32	-	16	100 ⁽²⁾	μs
		Program/erase parallelism (PSIZE) = x 8	-	400	800	
t _{ERASE16KB}	t _{ERASE16KB} Sector (16 KB) erase time	Program/erase parallelism (PSIZE) = x 16	-	300	600	ms
		Program/erase parallelism (PSIZE) = x 32	-	250	500	
		Program/erase parallelism (PSIZE) = x 8	-	1200	2400	
t _{ERASE64KB}	Sector (64 KB) erase time	Program/erase parallelism (PSIZE) = x 16	-	700	1400	ms
		Program/erase parallelism (PSIZE) = x 32	-	550	1100	

Table 40. Flash memory programming (continued)

Symbol	Parameter	Conditions	Min ⁽¹⁾	Тур	Max ⁽¹⁾	Unit
	Sector (128 KB) erase time t _{ME} Mass erase time	Program/erase parallelism (PSIZE) = x 8	-	2	4	
t _{ERASE128KB}		Program/erase parallelism (PSIZE) = x 16	-	1.3	2.6	s
		Program/erase parallelism (PSIZE) = x 32	-	1	2	
		Program/erase parallelism (PSIZE) = x 8	-	16	32	
t _{ME}		Program/erase parallelism (PSIZE) = x 16	-	11	22	S
V _{prog} Programming voltage		Program/erase parallelism (PSIZE) = x 32	-	8	16	
	32-bit program operation	2.7	-	3.6	V	
	Programming voltage	16-bit program operation	2.1	ı	3.6	٧
		8-bit program operation	1.8		3.6	٧

^{1.} Guaranteed by characterization.

^{2.} The maximum programming time is measured after 100K erase operations.

Symbol	Parameter	Conditions	Min ⁽¹⁾	Тур	Max ⁽¹⁾	Unit
t _{prog}	Double word programming		-	16	100 ⁽²⁾	μs
t _{ERASE16KB}	Sector (16 KB) erase time	T _A = 0 to +40 °C	-	230	-	
t _{ERASE64KB}	Sector (64 KB) erase time	V _{DD} = 3.3 V	-	490	-	ms
t _{ERASE128KB}	Sector (128 KB) erase time	V _{PP} = 8.5 V	-	875	-	
t _{ME}	Mass erase time		-	6.9	-	S
V _{prog}	Programming voltage	-	2.7	-	3.6	V
V _{PP}	V _{PP} voltage range	-	7	-	9	V
I _{PP}	Minimum current sunk on the V _{PP} pin	-	10	-	-	mA
t _{VPP} (3)	Cumulative time during which V _{PP} is applied	-	-	-	1	hour

Table 41. Flash memory programming with V_{PP}

Table 42. Flash memory endurance and data retention

Symbol Parameter		Conditions	Value	Unit
			Min ⁽¹⁾	
N _{END}	⊢nduranca	$T_A = -40 \text{ to } +85 ^{\circ}\text{C} \text{ (6 suffix versions)}$ $T_A = -40 \text{ to } +105 ^{\circ}\text{C} \text{ (7 suffix versions)}$	10	kcycles
		1 kcycle ⁽²⁾ at T _A = 85 °C	30	
t _{RET}	Data retention	1 kcycle ⁽²⁾ at T _A = 105 °C	10	Years
		10 kcycles ⁽²⁾ at T _A = 55 °C	20	

^{1.} Guaranteed by characterization.

5.3.13 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports). the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- **Electrostatic discharge (ESD)** (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- FTB: A burst of fast transient voltage (positive and negative) is applied to V_{DD} and V_{SS} through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.



^{1.} Guaranteed by design.

^{2.} The maximum programming time is measured after 100K erase operations.

^{3.} V_{PP} should only be connected during programming/erasing.

^{2.} Cycling performed over the whole temperature range.

A device reset allows normal operations to be resumed.

The test results are given in *Table 43*. They are based on the EMS levels and classes defined in application note AN1709.

Table 43. EMS characteristics

Symbol	Parameter	Conditions	Level/ Class
V _{FESD}	Voltage limits to be applied on any I/O pin to induce a functional disturbance	V_{DD} = 3.3 V, LQFP176, T_{A} = +25 °C, f_{HCLK} = 168 MHz, conforms to IEC 61000-4-2	2B
V _{EFTB}	Fast transient voltage burst limits to be applied through 100 pF on V _{DD} and V _{SS} pins to induce a functional disturbance	V _{DD} = 3.3 V, LQFP176, T _A = +25 °C, f _{HCLK} = 168 MHz, conforms to IEC 61000-4-2	4A

Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.

Software recommendations

The software flowchart must include the management of runaway conditions such as:

- Corrupted program counter
- Unexpected reset
- Critical Data corruption (control registers...)

Prequalification trials

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the Oscillator pins for 1 second.

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

Electromagnetic Interference (EMI)

The electromagnetic field emitted by the device are monitored while a simple application, executing EEMBC? code, is running. This emission test is compliant with SAE IEC61967-2 standard which specifies the test board and the pin loading.

Table 44. EMI characteristics

Symbol	Parameter	Conditions	Monitored frequency band	Max vs. [f _{HSE} /f _{CPU}]	Unit
			nequency band	25/168 MHz	
		V - 2.2 V T - 25 °C LOED176	0.1 to 30 MHz	32	
	package, conformation EEMBC, code reaction ART accelerate	V _{DD} = 3.3 V, T _A = 25 °C, LQFP176 package, conforming to SAE J1752/3	30 to 130 MHz	25	dΒμV
		EEMBC, code running from Flash with ART accelerator enabled	130 MHz to 1GHz	29	
c		Peak level $V_{DD} = 3.3 \text{ V, } T_{A} = 25 \text{ °C, LQFP176}$ package, conforming to SAE J1752/3	SAE EMI Level	4	-
S _{EMI}			0.1 to 30 MHz	19	
			30 to 130 MHz	16	dΒμV
		EEMBC, code running from Flash with ART accelerator and PLL spread	130 MHz to 1GHz	18	
		spectrum enabled	SAE EMI level	3.5	-

5.3.14 Absolute maximum ratings (electrical sensitivity)

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts × (n+1) supply pins). This test conforms to the JESD22-A114/C101 standard.

Table 45. ESD absolute maximum ratings

Symbol	Ratings	ngs Conditions (Maximum value ⁽¹⁾	Unit
V _{ESD(HBM)}	Electrostatic discharge voltage (human body model)	T _A = +25 °C conforming to JESD22-A114	2	2000 ⁽²⁾	V
V _{ESD(CDM)}	Electrostatic discharge voltage (charge device model)	T _A = +25 °C conforming to ANSI/ESD STM5.3.1	II	500	V

^{1.} Guaranteed by characterization.

^{2.} On V_{BAT} pin, $V_{ESD(HBM)}$ is limited to 1000 V.

Static latchup

Two complementary static tests are required on six parts to assess the latchup performance:

- A supply overvoltage is applied to each power supply pin
- A current injection is applied to each input, output and configurable I/O pin

These tests are compliant with EIA/JESD 78A IC latchup standard.

Table 46. Electrical sensitivities

	Symbol	Parameter	Conditions	Class
ĺ	LU	Static latch-up class	T _A = +105 °C conforming to JESD78A	II level A

5.3.15 I/O current injection characteristics

As a general rule, current injection to the I/O pins, due to external voltage below V_{SS} or above V_{DD} (for standard, 3 V-capable I/O pins) should be avoided during normal product operation. However, in order to give an indication of the robustness of the microcontroller in cases when abnormal injection accidentally happens, susceptibility tests are performed on a sample basis during device characterization.

Functional susceptibilty to I/O current injection

While a simple application is executed on the device, the device is stressed by injecting current into the I/O pins programmed in floating input mode. While current is injected into the I/O pin, one at a time, the device is checked for functional failures.

The failure is indicated by an out of range parameter: ADC error above a certain limit (>5 LSB TUE), out of conventional limits of induced leakage current on adjacent pins (out of $5 \mu A/+0 \mu A$ range), or other functional failure (for example reset, oscillator frequency deviation).

Negative induced leakage current is caused by negative injection and positive induced leakage current by positive injection.

The test results are given in *Table 47*.

Functional susceptibility Symbol Description Unit **Negative Positive** injection injection Injected current on BOOT0 pin - 0 NA Injected current on NRST pin -0NA Injected current on PE2, PE3, PE4, PE5, PE6, PI8, PC13, PC14, PC15, PI9, PI10, PI11, PF0, $I_{\text{INJ}}^{\,(1)}$ PF1, PF2, PF3, PF4, PF5, PF10, PH0/OSC IN, NA -0mΑ PH1/OSC OUT, PC0, PC1, PC2, PC3, PB6, PB7, PB8, PB9, PE0, PE1, PI4, PI5, PI6, PI7, PDR ON, BYPASS REG Injected current on all FT pins - 5 NA Injected current on any other pin - 5 +5

Table 47. I/O current injection susceptibility

5.3.16 I/O port characteristics

General input/output characteristics

Unless otherwise specified, the parameters given in *Table 48* are derived from tests performed under the conditions summarized in *Table 14*. All I/Os are CMOS and TTL compliant.

Conditions Symbol Unit **Parameter** Min Max Typ 0.3V_{DD}-0.04⁽¹⁾ FT, TTa and NRST I/O input low $1.7 \ V \le V_{DD} \le 3.6 \ V$ level voltage $0.3V_{DD}^{(2)}$ $1.75 \text{ V} \le \text{V}_{DD} \le 3.6 \text{ V}$ V_{IL} $-40 \, ^{\circ}C \le T_A \le 105 \, ^{\circ}C$ BOOT0 I/O input low level 0.1V_{DD}-+0.1⁽¹⁾ voltage $1.7 \ V \le V_{DD} \le 3.6 \ V$ $0~^{\circ}C \le T_A \le 105~^{\circ}C$ ٧ $0.45V_{DD} + 0.3^{(1)}$ FT, TTa and NRST I/O input low $1.7 \text{ V} \le \text{V}_{DD} \le 3.6 \text{ V}$ level voltage $0.7V_{DD}^{(2)}$ $1.75 \text{ V} \le \text{V}_{DD} \le 3.6 \text{ V}$ V_{IH} $-40 \, ^{\circ}\text{C} \le T_A \le 105 \, ^{\circ}\text{C}$ BOOT0 I/O input low level 0.17V_{DD}+0.7⁽¹⁾ voltage $1.7 \text{ V} \le \text{V}_{DD} \le 3.6 \text{ V}$ $0 \, ^{\circ}C \le T_A \le 105 \, ^{\circ}C$

Table 48. I/O static characteristics

^{1.} It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative currents.

Symbol	Paran	neter	Conditions	Min	Тур	Max	Unit
	FT, TTa and NRST I/O input hysteresis		1.7 V ≤ V _{DD} ≤ 3.6 V	10%V _{DD} ⁽³⁾	-	-	
V _{HYS}	POOTO I/O input	t hystorosis	$1.75 \text{ V} \le \text{V}_{DD} \le 3.6 \text{ V}$ -40 °C \le T_A \le 105 °C	0.1			V
	BOOT0 I/O input hysteresis		$1.7 \text{ V} \le \text{V}_{DD} \le 3.6 \text{ V}$ $0 \text{ °C} \le \text{T}_{A} \le 105 \text{ °C}$	0.1	-	-	
	I/O input leakage	e current ⁽⁴⁾	$V_{SS} \le V_{IN} \le V_{DD}$	-	-	±1	
I _{lkg}	I/O FT input leakage current (5)		V _{IN} = 5 V	-	-	3	μA
R _{PU}	Weak pull-up equivalent resistor ⁽⁶⁾	All pins except for PA10 and PB12 (OTG_FS_ID, OTG_HS_ID)	$V_{IN} = V_{SS}$	30	40	50	
		PA10 and PB12 (OTG_FS_ID, OTG_HS_ID)	-	7	10	14	kΩ
R _{PD}	Weak pull-down equivalent resistor ⁽⁷⁾	All pins except for PA10 and PB12	$V_{IN} = V_{DD}$	30	40	50	
	16919101	PA10 and PB12	-	7	10	14	
C _{IO} ⁽⁸⁾	I/O pin capacitance			-	5	-	pF

Table 48. I/O static characteristics (continued)

All I/Os are CMOS and TTL compliant (no software configuration required). Their characteristics cover more than the strict CMOS-technology or TTL parameters.

^{1.} Guaranteed by design.

^{2.} Tested in production.

^{3.} With a minimum of 200 mV.

^{4.} Leakage could be higher than the maximum value, if negative current is injected on adjacent pins.Refer to Table 47: I/O current injection susceptibility

To sustain a voltage higher than V_{DD} + 0.3 V, the internal pull-up/pull-down resistors must be disabled. Leakage could be higher than the maximum value, if negative current is injected on adjacent pins. Refer to *Table 47: I/O current injection* susceptibility.

^{6.} Pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance is minimum (~10% order).

^{7.} Pull-up and pull-down resistors are designed with a true resistance in series with a switchable NMOS. This NMOS contribution to the series resistance is minimum (~10% order).

^{8.} Hysteresis voltage between Schmitt trigger switching levels. Guaranteed by characterization.

Output driving current

The GPIOs (general purpose input/outputs) can sink or source up to ± 8 mA, and sink or source up to ± 20 mA (with a relaxed V_{OL}/V_{OH}) except PC13, PC14 and PC15 which can sink or source up to ± 3 mA. When using the PC13 to PC15 GPIOs in output mode, the speed should not exceed 2 MHz with a maximum load of 30 pF.

In the user application, the number of I/O pins which can drive current must be limited to respect the absolute maximum rating specified in *Section 5.2*. In particular:

- The sum of the currents sourced by all the I/Os on V_{DD}, plus the maximum Run consumption of the MCU sourced on V_{DD}, cannot exceed the absolute maximum rating I_{VDD} (see *Table 12*).
- The sum of the currents sunk by all the I/Os on V_{SS} plus the maximum Run consumption of the MCU sunk on V_{SS} cannot exceed the absolute maximum rating I_{VSS} (see *Table 12*).

Output voltage levels

Unless otherwise specified, the parameters given in *Table 49* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 14*. All I/Os are CMOS and TTL compliant.

		,	_		
Symbol	Parameter	Conditions	Min	Max	Unit
V _{OL} ⁽²⁾	Output low level voltage	CMOS port	-	0.4	
V _{OH} ⁽³⁾	Output high level voltage	I _{IO} = +8 mA 2.7 V < V _{DD} < 3.6 V	V _{DD} -0.4	-	V
V _{OL} (2)	Output low level voltage	TTL port	-	0.4	
V _{OH} ⁽³⁾	Output high level voltage	I _{IO} =+ 8mA 2.7 V < V _{DD} < 3.6 V	2.4	-	V
V _{OL} ⁽²⁾⁽⁴⁾	Output low level voltage	I _{IO} = +20 mA	-	1.3	V
V _{OH} ⁽³⁾⁽⁴⁾	Output high level voltage	2.7 V < V _{DD} < 3.6 V	V _{DD} -1.3	-	\ \ \
V _{OL} ⁽²⁾⁽⁴⁾		I _{IO} = +6 mA	-	0.4	V
V _{OH} ⁽³⁾⁽⁴⁾	Output high level voltage	2 V < V _{DD} < 2.7 V	V _{DD} -0.4	-]

Table 49. Output voltage characteristics⁽¹⁾

- PC13, PC14, PC15 and PI8 are supplied through the power switch. Since the switch only sinks a limited amount of current (3 mA), the use of GPIOs PC13 to PC15 and PI8 in output mode is limited: the speed should not exceed 2 MHz with a maximum load of 30 pF and these I/Os must not be used as a current source (e.g. to drive an LED).
- 2. The $I_{|O}$ current sunk by the device must always respect the absolute maximum rating specified in *Table 12* and the sum of $I_{|O}$ (I/O ports and control pins) must not exceed I_{VSS} .
- 3. The I_{IO} current sourced by the device must always respect the absolute maximum rating specified in *Table 12* and the sum of I_{IO} (I/O ports and control pins) must not exceed I_{VDD} .
- 4. Guaranteed by characterization.

Input/output AC characteristics

The definition and values of input/output AC characteristics are given in *Figure 37* and *Table 50*, respectively.

Unless otherwise specified, the parameters given in *Table 50* are derived from tests performed under the ambient temperature and V_{DD} supply voltage conditions summarized in *Table 14*.

Table 50. I/O AC characteristics⁽¹⁾⁽²⁾

OSPEEDRy [1:0] bit value ⁽¹⁾	Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
				$C_L = 50 \text{ pF}, V_{DD} > 2.70 \text{ V}$	-	-	4	
	f 400 4	Maximum frequency ⁽³⁾	$C_L = 50 \text{ pF}, V_{DD} > 1.8 \text{ V}$	-	-	2	MHz	
00	'max(IO)out	maximum requericy	$C_L = 10 \text{ pF}, V_{DD} > 2.70 \text{ V}$	-	-	8	101112	
00			$C_L = 10 \text{ pF}, V_{DD} > 1.8 \text{ V}$	-	-	4		
	t _{f(IO)out} / t _{r(IO)out}	Output high to low level fall time and output low to high level rise time	C _L = 50 pF, V _{DD} = 1.8 V to 3.6 V	-	-	100	ns	
			C _L = 50 pF, V _{DD >} 2.70 V	-	-	25		
	f	(IO)out Maximum frequency ⁽³⁾	C _L = 50 pF, V _{DD >} 1.8 V	-	-	12.5	MHz	
	'max(IO)out	iwaxiinuin irequency.	$C_L = 10 \text{ pF}, V_{DD} > 2.70 \text{ V}$	-	-	50 ⁽⁴⁾		
01			C _L = 10 pF, V _{DD >} 1.8 V	-	-	20		
01	t _{f(IO)out} / t _{r(IO)out}		C _L = 50 pF, V _{DD} >2.7 V	-	-	10	ns	
		Output high to low level fall time and output low to high	$C_L = 50 \text{ pF}, V_{DD} > 1.8 \text{ V}$	-	-	20		
		tinos (level rise time	C _L = 10 pF, V _{DD >} 2.70 V	-	-	6	115
			C _L = 10 pF, V _{DD >} 1.8 V	-	-	10		
			C _L = 40 pF, V _{DD >} 2.70 V	-	-	50 ⁽⁴⁾		
	f	Maximum fraguanay(3)	C _L = 40 pF, V _{DD >} 1.8 V	-	-	25	MHz	
	^I max(IO)out	Maximum frequency ⁽³⁾	C _L = 10 pF, V _{DD >} 2.70 V	-	-	100 ⁽⁴⁾	IVITZ	
10			C _L = 10 pF, V _{DD >} 1.8 V	-	-	50 ⁽⁴⁾		
			C _L = 40 pF, V _{DD >} 2.70 V	-	-	6		
		Output high to low level fall	C _L = 40 pF, V _{DD >} 1.8 V	-	-	10	ne	
	t _{r(IO)out}	. Y Turre and output low to mun	C _L = 10 pF, V _{DD >} 2.70 V	-	-	4	ns	
			C _L = 10 pF, V _{DD >} 1.8 V	-	-	6		

OSPEEDRy [1:0] bit value ⁽¹⁾	Symbol	Parameter	Conditions	Min	Тур	Max	Unit
			$C_L = 30 \text{ pF}, V_{DD} > 2.70 \text{ V}$	-	-	100 ⁽⁴⁾	
11	_	Maximum frequency ⁽³⁾	$C_L = 30 \text{ pF}, V_{DD} > 1.8 \text{ V}$	-	-	50 ⁽⁴⁾	MHz
	max(IO)out		C _L = 10 pF, V _{DD >} 2.70 V	-	-	180 ⁽⁴⁾	(4)
			C _L = 10 pF, V _{DD >} 1.8 V	-	-	100 ⁽⁴⁾	
''		Output high to low level fall	C _L = 30 pF, V _{DD >} 2.70 V	-	-	4	
	t _{f(IO)out} /		C _L = 30 pF, V _{DD >} 1.8 V	-	-	6	l no
	t _{r(IO)out}	time and output low to high level rise time	C _L = 10 pF, V _{DD >} 2.70 V	-	-	2.5	ns
			C _L = 10 pF, V _{DD >} 1.8 V	-	-	4	
-	t _{EXTIpw}	Pulse width of external signals detected by the EXTI controller		10	-	-	ns

Table 50. I/O AC characteristics⁽¹⁾⁽²⁾ (continued)

- The maximum frequency is defined in Figure 37.
- 4. For maximum frequencies above 50 MHz, the compensation cell should be used.

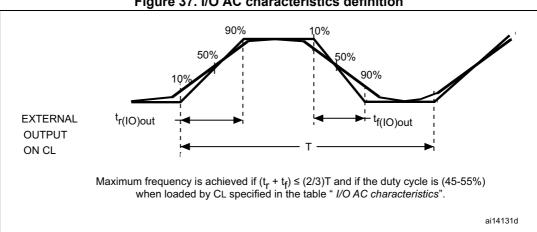


Figure 37. I/O AC characteristics definition

Guaranteed by characterization.

The I/O speed is configured using the OSPEEDRy[1:0] bits. Refer to the STM32F4xx reference manual for a description of the GPIOx_SPEEDR GPIO port output speed register.

5.3.17 NRST pin characteristics

The NRST pin input driver uses CMOS technology. It is connected to a permanent pull-up resistor, R_{PLI} (see *Table 48*).

Unless otherwise specified, the parameters given in *Table 51* are derived from tests performed under the ambient temperature and V_{DD} supply voltage conditions summarized in *Table 14*.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{IL(NRST)} ⁽¹⁾	NRST Input low level voltage	TTL ports	-	-	0.8	
V _{IH(NRST)} ⁽¹⁾	NRST Input high level voltage	2.7 V ≤ V _{DD} ≤ 3.6 V	2	ı	-	V
V _{IL(NRST)} ⁽¹⁾	NRST Input low level voltage	CMOS ports	ı	ı	0.3V _{DD}	V
V _{IH(NRST)} ⁽¹⁾	NRST Input high level voltage	1.8 V ≤ V _{DD} ≤ 3.6 V	0.7V _{DD}	ı	ı	
V _{hys(NRST)}	NRST Schmitt trigger voltage hysteresis	-	-	200	-	mV
R _{PU}	Weak pull-up equivalent resistor ⁽²⁾	$V_{IN} = V_{SS}$	30	40	50	kΩ
V _{F(NRST)} ⁽¹⁾	NRST Input filtered pulse		-	-	100	ns
V _{NF(NRST)} ⁽¹⁾	NRST Input not filtered pulse	V _{DD} > 2.7 V	300		-	ns
T _{NRST_OUT}	Generated reset pulse duration	Internal Reset source	20	-	-	μs

Table 51. NRST pin characteristics

^{2.} The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance must be minimum (~10% order).

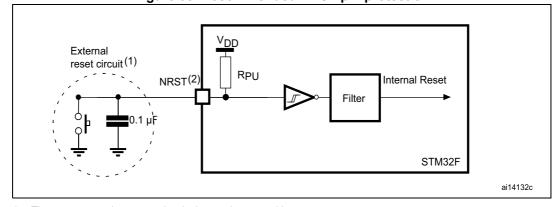


Figure 38. Recommended NRST pin protection

- 1. The reset network protects the device against parasitic resets.
- The user must ensure that the level on the NRST pin can go below the V_{IL(NRST)} max level specified in Table 51. Otherwise the reset is not taken into account by the device.

^{1.} Guaranteed by design.

5.3.18 TIM timer characteristics

The parameters given in *Table 52* and *Table 53* are guaranteed by design.

Refer to *Section 5.3.16: I/O port characteristics* for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

Table 52. Characteristics of TIMx connected to the APB1 domain⁽¹⁾

Symbol	Parameter	Conditions	Min	Max	Unit
		AHB/APB1	1	-	t _{TIMxCLK}
t _{res(TIM)}	Timer resolution time	prescaler distinct from 1, f _{TIMxCLK} = 84 MHz	11.9	-	ns
		AHB/APB1	1	-	t _{TIMxCLK}
		prescaler = 1, f _{TIMxCLK} = 42 MHz	23.8	-	ns
f	Timer external clock		0	f _{TIMxCLK} /2	MHz
f _{EXT}	frequency on CH1 to CH4		0	42	MHz
Res _{TIM}	Timer resolution		-	16/32	bit
	16-bit counter clock		1	65536	t _{TIMxCLK}
t	period when internal clock is selected	f _{TIMxCLK} = 84 MHz APB1= 42 MHz	0.0119	780	μs
^t COUNTER	32-bit counter clock		1	-	t _{TIMxCLK}
	period when internal clock is selected		0.0119	51130563	μs
4	Maximum possible sount		-	65536 × 65536	t _{TIMxCLK}
t _{MAX_COUNT}	Maximum possible count		-	51.1	s

^{1.} TIMx is used as a general term to refer to the TIM2, TIM3, TIM4, TIM5, TIM6, TIM7, and TIM12 timers.

Symbol	Parameter	Conditions	Min	Max	Unit
		AHB/APB2	1	-	t _{TIMxCLK}
t _{res(TIM)}	Timer resolution time	prescaler distinct from 1, f _{TIMxCLK} = 168 MHz	5.95	-	ns
		AHB/APB2	1	-	t _{TIMxCLK}
		prescaler = 1, f _{TIMxCLK} = 84 MHz	11.9	-	ns
	Timer external clock		0	f _{TIMxCLK} /2	MHz
f _{EXT}	frequency on CH1 to CH4		0	84	MHz
Res _{TIM}	Timer resolution	f _{TIMxCLK} = 168 MHz	ı	16	bit
t _{COUNTER}	16-bit counter clock period when internal clock is selected	APB2 = 84 MHz	1	65536	t _{TIMxCLK}
t _{MAX_COUNT}	Maximum possible count		-	32768	t _{TIMxCLK}

Table 53. Characteristics of TIMx connected to the APB2 domain⁽¹⁾

5.3.19 Communications interfaces

I²C interface characteristics

The I²C interface meets the timings requirements of the I²C-bus specification and user manual rev. 03 for:

- Standard-mode (Sm): with a bit rate up to 100 kbit/s
- Fast-mode (Fm): with a bit rate up to 400 kbit/s.

The I²C timings requirements are guaranteed by design when the I2C peripheral is properly configured (refer to RM0090 reference manual).

The SDA and SCL I/O requirements are met with the following restrictions: the SDA and SCL I/O pins are not "true" open-drain. When configured as open-drain, the PMOS connected between the I/O pin and V_{DD} is disabled, but is still present. Refer to Section 5.3.16: I/O port characteristics for more details on the I²C I/O characteristics.

All I²C SDA and SCL I/Os embed an analog filter. Refer to the table below for the analog filter characteristics:

Symbol Parameter Min Max Unit

Maximum pulse width of spikes that are suppressed by the analog filter 50⁽²⁾ 260⁽³⁾ ns

Table 54. I2C analog filter characteristics⁽¹⁾

- 1. Guaranteed by design.
- 2. Spikes with widths below $t_{AF(min)}$ are filtered.
- 3. Spikes with widths above $t_{AF(max)}$ are not filtered

^{1.} TIMx is used as a general term to refer to the TIM1, TIM8, TIM9, TIM10, and TIM11 timers.

SPI interface characteristics

Unless otherwise specified, the parameters given in *Table 55* for SPI are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and V_{DD} supply voltage conditions summarized in *Table 14* with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: 0.5 V_{DD}

Refer to Section 5.3.16: I/O port characteristics for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO).

Table 55. SPI dynamic characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f		Master mode, SPI1, 2.7V < V _{DD} < 3.6V			42	
SDI clock froguency	SDI glock fraguancy	Slave mode, SPI1, 2.7V < V _{DD} < 3.6V	-	_	42	MU
1/4	SPI clock frequency	Master mode, SPI1/2/3, 1.7V < V _{DD} < 3.6V			21	MHz
1/t _{c(SCK)}		Slave mode, SPI1/2/3, 1.7V < V _{DD} < 3.6V	-	-	21	
Duty(SCK)	Duty cycle of SPI clock frequency	Slave mode	30	50	70	%

Table 55. SPI dynamic characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
t _{w(SCKH)}	SCK high and low time	Master mode, SPI presc = 2, 2.7V < V _{DD} < 3.6V	T _{PCLK} -0.5	T _{PCLK}	T _{PCLK} +0.5	
t _{w(SCKL)}	SCK high and low time	Master mode, SPI presc = 2, 1.7V < V _{DD} < 3.6V	T _{PCLK} -2	T _{PCLK}	T _{PCLK} +2	
t _{su(NSS)}	NSS setup time	Slave mode, SPI presc = 2	4 x T _{PCLK}			
t _{h(NSS)}	NSS hold time	Slave mode, SPI presc = 2	2 x T _{PCLK}	-	-	
t _{su(MI)}	Data input actus time	Master mode	6.5	-	-	
t _{su(SI)}	Data input setup time	Slave mode	2.5	-	-	
t _{h(MI)}	Data input hold time	Master mode	2.5	-	-	
t _{h(SI)}	- Data input hold time	Slave mode	4	-	-	
t _{a(SO)} ⁽²⁾	Data output access time	Slave mode, SPI presc = 2	0	-	4 x T _{PCLK}	
	Data culturi dicable time	Slave mode, SPI1, 2.7V < V _{DD} < 3.6V	0	-	7.5	
t _{dis(SO)} (3)	Data output disable time	Slave mode, SPI1/2/3 1.7V < V _{DD} < 3.6V	0	-	16.5	ns
		Slave mode (after enable edge), SPI1, 2.7V < V _{DD} < 3.6V	-	11	13	
t _{v(SO)}	Data output valid/hold time	Slave mode (after enable edge), SPI2/3, 2.7V < V _{DD} < 3.6V	-	12	16.5	
t _{h(SO)}	Data output valid/hold time	Slave mode (after enable edge), SPI1, 1.7V < V _{DD} < 3.6V	-	15.5	19	
		Slave mode (after enable edge), SPI2/3, 1.7V < V _{DD} < 3.6V	-	18	20.5	
+	Data output valid time	Master mode (after enable edge), SPI1, 2.7V < V _{DD} < 3.6V	-	-	2.5	
t _{v(MO)}	Data output valid time	Master mode (after enable edge), SPI1/2/3, 1.7V < V _{DD} < 3.6V	-	-	4.5	
t _{h(MO)}	Data output hold time	Master mode (after enable edge)	0	-	-	

^{1.} Guaranteed by characterization.

^{2.} Min time is for the minimum time to drive the output and the max time is for the maximum time to validate the data.

^{3.} Min time is for the minimum time to invalidate the output and the max time is for the maximum time to put the data in Hi-Z.

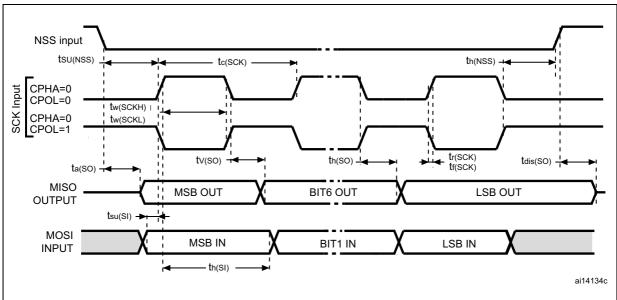
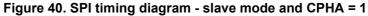
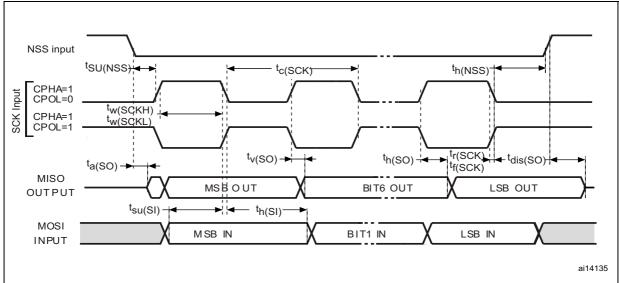


Figure 39. SPI timing diagram - slave mode and CPHA = 0





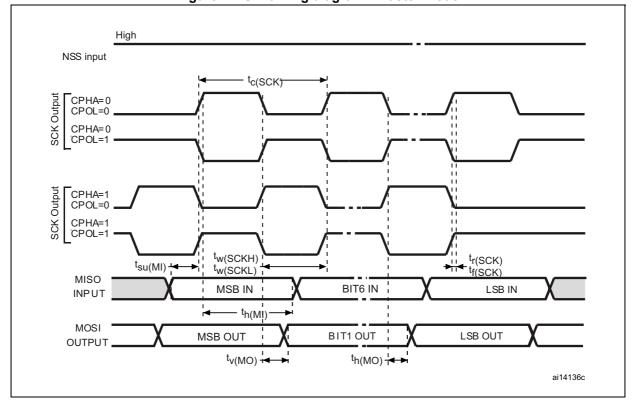


Figure 41. SPI timing diagram - master mode

I²S interface characteristics

Unless otherwise specified, the parameters given in *Table 56* for the i^2S interface are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and V_{DD} supply voltage conditions summarized in *Table 14*, with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: 0.5 V_{DD}

Refer to Section 5.3.16: I/O port characteristics for more details on the input/output alternate function characteristics (CK, SD, WS).

Table 56. I²S dynamic characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Max	Unit
f _{MCK}	I ² S main clock output	-	256 x 8K	256 x F _S ⁽²⁾	MHz
f.	I ² S clock frequency	Master data: 32 bits	-	64 x F _S	MHz
f _{CK}	1 3 clock frequency	Slave data: 32 bits	-	64 x F _S	IVII IZ
D _{CK}	I ² S clock frequency duty cycle	Slave receiver	30	70	%
t _{v(WS)}	WS valid time	Master mode	0	6	
t _{h(WS)}	WS hold time	Master mode	0	-	
t _{su(WS)}	WS setup time	Slave mode	1	-	
t _{h(WS)}	WS hold time	Slave mode	0	-	
t _{su(SD_MR)}	Data input actual time	Master receiver	7.5	-	
t _{su(SD_SR)}	Data input setup time	Slave receiver	2	-	ns
t _{h(SD_MR)}	Data input hold time	Master receiver	0	-	110
t _{h(SD_SR)}	Data input hold time	Slave receiver	0	-	
$t_{v(SD_ST)}$ $t_{h(SD_ST)}$	Data output valid time	Slave transmitter (after enable edge)	-	27	
t _{v(SD_MT)}		Master transmitter (after enable edge)	-	20	
t _{h(SD_MT)}	Data output hold time	Master transmitter (after enable edge)	2.5	-	

^{1.} Guaranteed by characterization.

Note:

Refer to the I^2S section of RM0090 reference manual for more details on the sampling frequency (F_S) . f_{MCK} , f_{CK} , and D_{CK} values reflect only the digital peripheral behavior. The value of these parameters might be slightly impacted by the source clock accuracy. D_{CK} depends mainly on the value of ODD bit. The digital contribution leads to a minimum value of $I2SDIV / (2 \times I2SDIV + ODD)$ and a maximum value of $I2SDIV + ODD) / (2 \times I2SDIV + ODD)$. F_S maximum value is supported for each mode/condition.

^{2.} The maximum value of 256 x F_S is 42 MHz (APB1 maximum frequency).

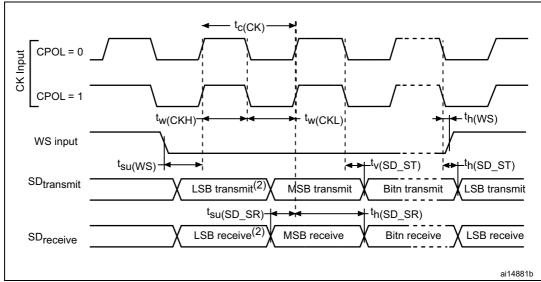


Figure 42. I²S slave timing diagram (Philips protocol)

LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first byte.

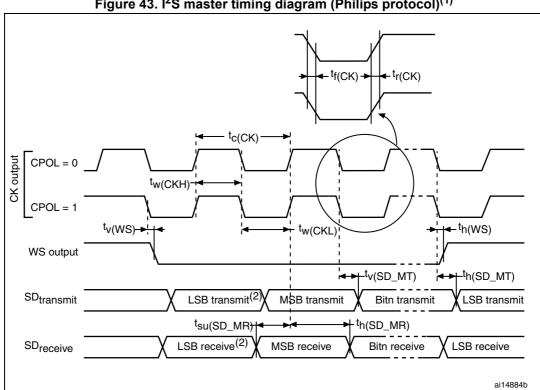


Figure 43. I²S master timing diagram (Philips protocol)⁽¹⁾

- Guaranteed by characterization.
- LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first

USB OTG FS characteristics

This interface is present in both the USB OTG HS and USB OTG FS controllers.

Table 57. USB OTG FS startup time

Symbol	Parameter	Max	Unit
t _{STARTUP} ⁽¹⁾	USB OTG FS transceiver startup time	1	μs

^{1.} Guaranteed by design.

Table 58. USB OTG FS DC electrical characteristics

Sym	bol	Parameter	Conditions	Min. ⁽¹⁾	Тур.	Max. ⁽¹⁾	Unit
	V_{DD}	USB OTG FS operating voltage	-	3.0 ⁽²⁾	-	3.6	V
Input	V _{DI} ⁽³⁾	Differential input sensitivity	I(USB_FS_DP/DM, USB_HS_DP/DM)	0.2	-	-	
levels	V _{CM} ⁽³⁾	Differential common mode range	Includes V _{DI} range	0.8	-	2.5	V
	V _{SE} ⁽³⁾	Single ended receiver threshold	-	1.3	-	2.0	
Output	V_{OL}	Static output level low	R_L of 1.5 k Ω to 3.6 $V^{(4)}$	-	-	0.3	V
levels	V _{OH}	Static output level high	R_L of 15 k Ω to $V_{SS}^{(4)}$	2.8	-	3.6	V
D		PA11, PA12, PB14, PB15 (USB_FS_DP/DM, USB_HS_DP/DM)	V _{IN} = V _{DD}	17	21	24	
R _{PD}		PA9, PB13 (OTG_FS_VBUS, OTG_HS_VBUS)	VIN - VDD	0.65	1.1	2.0	kΩ
		PA12, PB15 (USB_FS_DP, USB_HS_DP)	V _{IN} = V _{SS}	1.5	1.8	2.1	
R _F	PU	PA9, PB13 (OTG_FS_VBUS, OTG_HS_VBUS)	$V_{IN} = V_{SS}$	0.25	0.37	0.55	

^{1.} All the voltages are measured from the local ground potential.

^{2.} The STM32F415xx and STM32F417xx USB OTG FS functionality is ensured down to 2.7 V but not the full USB OTG FS electrical characteristics which are degraded in the 2.7-to-3.0 V $\rm V_{DD}$ voltage range.

^{3.} Guaranteed by design.

^{4.} R_L is the load connected on the USB OTG FS drivers

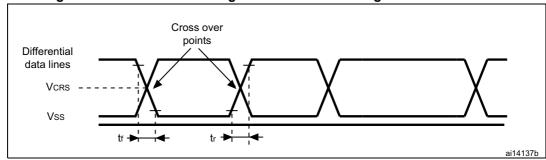


Figure 44. USB OTG FS timings: definition of data signal rise and fall time

Table 59. USB OTG FS electrical characteristics⁽¹⁾

	Driver characteristics								
Symbol Parameter Conditions Min Max Unit									
t _r	Rise time ⁽²⁾	C _L = 50 pF	4	20	ns				
t _f	Fall time ⁽²⁾	C _L = 50 pF	4	20	ns				
t _{rfm}	Rise/ fall time matching	t _r /t _f	90	110	%				
V _{CRS}	Output signal crossover voltage	-	1.3	2.0	V				

^{1.} Guaranteed by design.

USB HS characteristics

Unless otherwise specified, the parameters given in *Table 62* for ULPI are derived from tests performed under the ambient temperature, f_{HCLK} frequency summarized in *Table 61* and V_{DD} supply voltage conditions summarized in *Table 60*, with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: 0.5V_{DD}.

Refer to Section Section 5.3.16: I/O port characteristics for more details on the input/output characteristics.

Table 60. USB HS DC electrical characteristics

Syml	ool	Parameter	Min. ⁽¹⁾	Max. ⁽¹⁾	Unit
Input level	V_{DD}	USB OTG HS operating voltage	2.7	3.6	V

^{1.} All the voltages are measured from the local ground potential.

Table 61. USB HS clock timing parameters⁽¹⁾

Parameter		Symbol	Min	Nominal	Max	Unit
f _{HCLK} value to guarantee prope USB HS interface	er operation of	-	30	-	-	MHz
Frequency (first transition) 8-bit ±10%		F _{START_8BIT}	54	60	66	MHz

Measured from 10% to 90% of the data signal. For more detailed informations, please refer to USB Specification - Chapter 7 (version 2.0).

Symbol Min **Nominal** Unit **Parameter** Max Frequency (steady state) ±500 ppm 59.97 60 60.03 MHz $\mathsf{F}_{\mathsf{STEADY}}$ Duty cycle (first transition) 8-bit ±10% 40 50 60 % D_{START_8BIT} Duty cycle (steady state) ±500 ppm $\mathsf{D}_{\mathsf{STEADY}}$ 49.975 50 50.025 % Time to reach the steady state frequency and 1.4 ms T_{STEADY} duty cycle after the first transition Peripheral 5.6 _ T_{START_DEV} Clock startup time after the ms de-assertion of SuspendM Host T_{START_HOST} PHY preparation time after the first transition T_{PREP} μs of the input clock

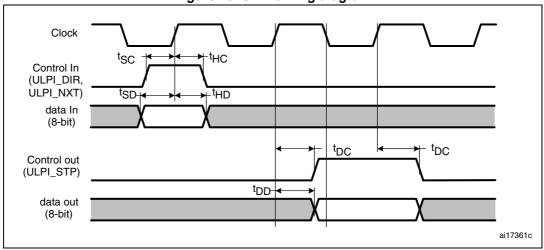
Table 61. USB HS clock timing parameters⁽¹⁾

Table 62. ULPI timing

Parameter	Cymhal	Valu	Unit	
Farameter	Symbol	Min.	Max.	Oill
Control in (ULPI_DIR) setup time	+	-	2.0	
Control in (ULPI_NXT) setup time	t _{SC}	-	1.5	
Control in (ULPI_DIR, ULPI_NXT) hold time	t _{HC}	0	-	
Data in setup time	t _{SD}	-	2.0	ns
Data in hold time	t _{HD}	0	-	
Control out (ULPI_STP) setup time and hold time	t _{DC}	-	9.2	
Data out available from clock rising edge	t _{DD}	-	10.7	

^{1.} V_{DD} = 2.7 V to 3.6 V and T_A = -40 to 85 °C.

Figure 45. ULPI timing diagram



^{1.} Guaranteed by design.

Ethernet characteristics

Unless otherwise specified, the parameters given in Table 64, Table 65 and Table 66 for SMI, RMII and MII are derived from tests performed under the ambient temperature, f_{HCLK} frequency summarized in Table 14 and VDD supply voltage conditions summarized in *Table 63*, with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: 0.5V_{DD}.

Refer to Section 5.3.16: I/O port characteristics for more details on the input/output characteristics.

Table 63. Ethernet DC electrical characteristics

	Symb	ol	Parameter	Min. ⁽¹⁾	Max. ⁽¹⁾	Unit
Ī	Input level	V_{DD}	Ethernet operating voltage	2.7	3.6	V

^{1.} All the voltages are measured from the local ground potential.

Table 64 gives the list of Ethernet MAC signals for the SMI (station management interface) and Figure 46 shows the corresponding timing diagram.

tMDC ETH_MDC td(MDIO) → ETH_MDIO(O) tsu(MDIO) th(MDIO) ETH_MDIO(I) MS31384V1

Figure 46. Ethernet SMI timing diagram

Table 64. Dynamic characteristics: Eternity MAC signals for SMI⁽¹⁾

Symbol	Parameter	Min	Тур	Max	Unit
t _{MDC}	MDC cycle time(2.38 MHz)	411	420	425	
T _{d(MDIO)}	Write data valid time	6	10	13	ns
t _{su(MDIO)}	Read data setup time	12	-	-	113
t _{h(MDIO)}	Read data hold time	0	-	-	

^{1.} Guaranteed by characterization.

Table 65 gives the list of Ethernet MAC signals for the RMII and Figure 47 shows the corresponding timing diagram.

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RMII_REF_CLK

RMII_TX_EN
RMII_TXD[1:0]

tsu(RXD)
tsu(CRS)

RMII_RXD[1:0]
RMII_CRS_D

ai15667

Figure 47. Ethernet RMII timing diagram

Table 65. Dynamic characteristics: Ethernet MAC signals for RMII

Symbol	Rating	Min	Тур	Max	Unit
t _{su(RXD)}	Receive data setup time	2	-	-	ns
t _{ih(RXD)}	Receive data hold time	1	-	-	ns
t _{su(CRS)}	Carrier sense set-up time	0.5	-	-	ns
t _{ih(CRS)}	Carrier sense hold time	2	-	-	ns
t _{d(TXEN)}	Transmit enable valid delay time	8	9.5	11	ns
t _{d(TXD)}	Transmit data valid delay time	8.5	10	11.5	ns

Table 66 gives the list of Ethernet MAC signals for MII and *Figure 47* shows the corresponding timing diagram.

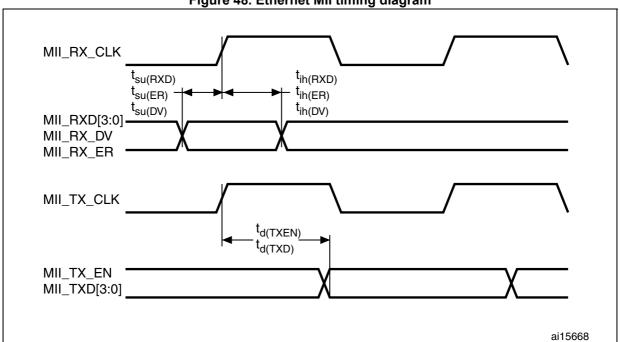


Figure 48. Ethernet MII timing diagram

Symbol	Parameter	Min	Тур	Max	Unit
t _{su(RXD)}	Receive data setup time	9		-	
t _{ih(RXD)}	Receive data hold time	10		-	
t _{su(DV)}	Data valid setup time	9		-	
t _{ih(DV)}	Data valid hold time	8		-	200
t _{su(ER)}	Error setup time	6		-	ns
t _{ih(ER)}	Error hold time	8		-	
t _{d(TXEN)}	Transmit enable valid delay time	0	10	14	
t _{d(TXD)}	Transmit data valid delay time	0	10	15	

Table 66. Dynamic characteristics: Ethernet MAC signals for MII⁽¹⁾

5.3.20 CAN (controller area network) interface

Refer to Section 5.3.16: I/O port characteristics for more details on the input/output alternate function characteristics (CANTX and CANRX).

5.3.21 12-bit ADC characteristics

Unless otherwise specified, the parameters given in *Table 67* are derived from tests performed under the ambient temperature, f_{PCLK2} frequency and V_{DDA} supply voltage conditions summarized in *Table 14*.

Symbol Parameter Conditions Unit Min Max Typ $1.8^{(1)}$ V_{DDA} Power supply 3.6 ٧ 1.8(1)(2)(3) V_{REF} V Positive reference voltage V_{DDA} $V_{DDA} = 1.8^{(1)(3)}$ to 2.4 V 0.6 15 18 MHz ADC clock frequency f_{ADC} $V_{DDA} = 2.4 \text{ to } 3.6 \text{ V}^{(3)}$ 0.6 30 36 MHz $f_{ADC} = 30 \text{ MHz},$ 1764 kHz 12-bit resolution $f_{TRIG}^{(4)}$ External trigger frequency 17 1/f_{ADC} 0 (V_{SSA} or V_{REF-} Conversion voltage range⁽⁵⁾ V V_{AIN} V_{REF+} tied to ground) See Equation 1 for $R_{AIN}^{(4)}$ External input impedance 50 κΩ details $R_{ADC}^{\overline{(4)(6)}}$ Sampling switch resistance 6 ĸΟ Internal sample and hold $C_{ADC}^{(4)}$ 4 pF capacitor f_{ADC} = 30 MHz 0.100 μs Injection trigger conversion $t_{lat}^{(4)}$ latency $3^{(7)}$ 1/f_{ADC}

Table 67. ADC characteristics

^{1.} Guaranteed by characterization.

Table 67. ADC characteristics (continued)						
Symbol	Parameter	Conditions	Min	Тур	Max	Unit
t _{latr} ⁽⁴⁾	Regular trigger conversion	f _{ADC} = 30 MHz	-	-	0.067	μs
latr` ′	latency		-	-	2 ⁽⁷⁾	1/f _{ADC}
ts ⁽⁴⁾	Sampling time	f _{ADC} = 30 MHz	0.100	-	16	μs
ıs. ,	Sampling time	-	3	-	480	1/f _{ADC}
t _{STAB} ⁽⁴⁾	Power-up time	-	-	2	3	μs
		f _{ADC} = 30 MHz 12-bit resolution	0.50	-	16.40	μs
		f _{ADC} = 30 MHz 10-bit resolution	0.43	-	16.34	μs
t _{CONV} ⁽⁴⁾	Total conversion time (including sampling time)	f _{ADC} = 30 MHz 8-bit resolution	0.37	-	16.27	μs
		f _{ADC} = 30 MHz 6-bit resolution	0.30	-	16.20	μs
		9 to 492 (t _S for sampling approximation)	+n-bit resolution f	or succe	ssive	1/f _{ADC}
		12-bit resolution Single ADC	-	-	2	Msps
f _S ⁽⁴⁾	Sampling rate (f _{ADC} = 30 MHz, and t _S = 3 ADC cycles)	12-bit resolution Interleave Dual ADC mode	-	-	3.75	Msps
	3	12-bit resolution Interleave Triple ADC mode	-	-	6	Msps
I _{VREF+} (4)	ADC V _{REF} DC current consumption in conversion	-	-	300	500	μA

Table 67. ADC characteristics (continued)

- 2. It is recommended to maintain the voltage difference between $V_{\text{REF+}}$ and V_{DDA} below 1.8 V.
- 3. $V_{DDA} V_{REF+} < 1.2 \text{ V}.$

4. Guaranteed by characterization.

mode

ADC $V_{\rm DDA}$ DC current

consumption in conversion

- 5. V_{REF+} is internally connected to V_{DDA} and V_{REF-} is internally connected to V_{SSA} .
- 6. R_{ADC} maximum value is given for V_{DD} =1.8 V, and minimum value for V_{DD} =3.3 V.
- 7. For external triggers, a delay of 1/f_{PCLK2} must be added to the latency specified in *Table* 67.

Equation 1: R_{AIN} max formula

$$R_{AIN} = \frac{(k-0.5)}{f_{ADC} \times C_{ADC} \times ln(2^{N+2})} - R_{ADC}$$



1.6

1.8

mΑ

V_{DD}/V_{DDA} minimum value of 1.7 V is obtained when the device operates in reduced temperature range, and with the use of an external power supply supervisor (refer to Section: Internal reset OFF).

The formula above (*Equation 1*) is used to determine the maximum external impedance allowed for an error below 1/4 of LSB. N = 12 (from 12-bit resolution) and k is the number of sampling periods defined in the ADC_SMPR1 register.

Table 68. ADC accuracy at f_{ADC} = 30 MHz

Symbol	Parameter	Test conditions	Тур	Max ⁽¹⁾	Unit
ET	Total unadjusted error		±2	±5	
EO	Offset error	 f _{PCLK2}	±1.5	±2.5	
EG	Gain error	f_{ADC} = 30 MHz, R_{AIN} < 10 kΩ, V_{DDA} = 1.8 ⁽²⁾ to 3.6 V	±1.5	±3	LSB
ED	Differential linearity error	$V_{DDA} = 1.8^{(2)} \text{ to } 3.6 \text{ V}$	±1	±2	
EL	Integral linearity error		±1.5	±3	

^{1.} Guaranteed by characterization.

Note:

ADC accuracy vs. negative injection current: injecting a negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative currents. Any positive injection current within the limits specified for $I_{INJ(PIN)}$ and $SI_{INJ(PIN)}$ in Section 5.3.16 does not affect the ADC accuracy.

V_{DD}/V_{DDA} minimum value of 1.7 V is obtained when the device operates in reduced temperature range, and with the use of an external power supply supervisor (refer to Section: Internal reset OFF).

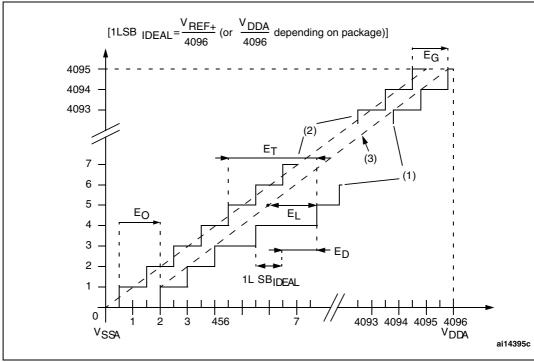


Figure 49. ADC accuracy characteristics

- See also Table 68.
- 2. Example of an actual transfer curve.
- Ideal transfer curve.
- End point correlation line.
- E_T = Total Unadjusted Error: maximum deviation between the actual and the ideal transfer curves.
 - EO = Offset Error: deviation between the first actual transition and the first ideal one. EG = Gain Error: deviation between the last ideal transition and the last actual one.
 - ED = Differential Linearity Error: maximum deviation between actual steps and the ideal one.
 - EL = Integral Linearity Error: maximum deviation between any actual transition and the end point correlation line.

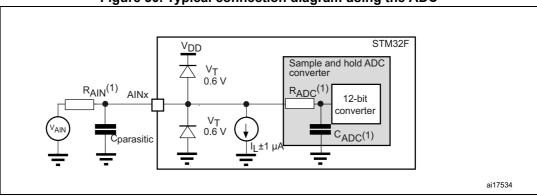
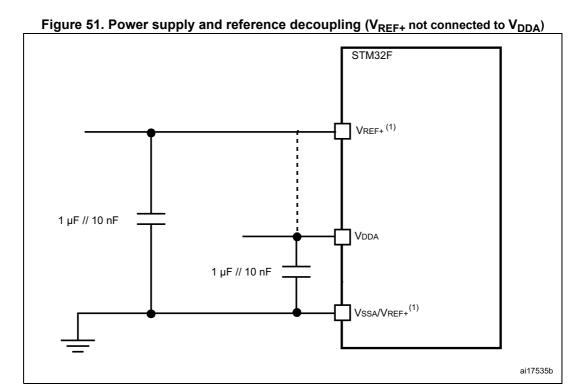


Figure 50. Typical connection diagram using the ADC

- Refer to Table 67 for the values of $R_{AIN},\,R_{ADC}$ and C_{ADC}
- $C_{parasitic}$ represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (roughly 5 pF). A high $C_{parasitic}$ value downgrades conversion accuracy. To remedy this, f_{ADC} should be reduced.

General PCB design guidelines

Power supply decoupling should be performed as shown in *Figure 51* or *Figure 52*, depending on whether V_{REF+} is connected to V_{DDA} or not. The 10 nF capacitors should be ceramic (good quality). They should be placed them as close as possible to the chip.



V_{REF+} and V_{REF} inputs are both available on UFBGA176. V_{REF+} is also available on LQFP100, LQFP144, and LQFP176. When V_{REF+} and V_{REF} are not available, they are internally connected to V_{DDA} and V_{SSA}.

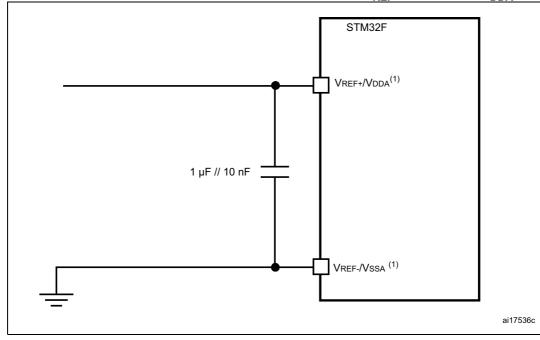


Figure 52. Power supply and reference decoupling (V_{REF+} connected to V_{DDA})

V_{REF+} and V_{REF-} inputs are both available on UFBGA176. V_{REF+} is also available on LQFP100, LQFP144, and LQFP176. When V_{REF+} and V_{REF-} are not available, they are internally connected to V_{DDA} and V_{SSA}.

5.3.22 Temperature sensor characteristics

Table 69. Temperature sensor characteristics

Symbol	Parameter	Min	Тур	Max	Unit
T _L ⁽¹⁾	V _{SENSE} linearity with temperature	-	±1	±2	°C
Avg_Slope ⁽¹⁾	Average slope	-	2.5		mV/°C
V ₂₅ ⁽¹⁾	Voltage at 25 °C	-	0.76		V
t _{START} (2)	Startup time	-	6	10	μs
T _{S_temp} ⁽²⁾	ADC sampling time when reading the temperature (1 °C accuracy)	10	-	-	μs

^{1.} Guaranteed by characterization.

Table 70. Temperature sensor calibration values

Symbol	Parameter	Memory address
TS_CAL1	TS ADC raw data acquired at temperature of 30 °C, V _{DDA} =3.3 V	0x1FFF 7A2C - 0x1FFF 7A2D
TS_CAL2	TS ADC raw data acquired at temperature of 110 °C, V _{DDA} =3.3 V	0x1FFF 7A2E - 0x1FFF 7A2F

^{2.} Guaranteed by design.

5.3.23 V_{BAT} monitoring characteristics

Table 71. V_{BAT} monitoring characteristics

Symbol	Parameter	Min	Тур	Max	Unit
R	Resistor bridge for V _{BAT}	-	50	-	KΩ
Q	Ratio on V _{BAT} measurement	-	2	-	
Er ⁽¹⁾	Error on Q	-1	-	+1	%
T _{S_vbat} ⁽²⁾⁽²⁾	ADC sampling time when reading the V _{BAT} 1 mV accuracy	5	-	-	μs

^{1.} Guaranteed by design.

5.3.24 Embedded reference voltage

The parameters given in *Table 72* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 14*.

Table 72. Embedded internal reference voltage

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{REFINT}	Internal reference voltage	-40 °C < T _A < +105 °C	1.18	1.21	1.24	V
T _{S_vrefint} ⁽¹⁾	ADC sampling time when reading the internal reference voltage	-	10	-	-	μs
V _{RERINT_s} ⁽²⁾	Internal reference voltage spread over the temperature range	V _{DD} = 3 V	i	3	5	mV
T _{Coeff} ⁽²⁾	Temperature coefficient	-	-	30	50	ppm/°C
t _{START} (2)	Startup time	-	-	6	10	μs

^{1.} Shortest sampling time can be determined in the application by multiple iterations.

Table 73. Internal reference voltage calibration values

Symbol	Parameter	Memory address
V _{REFIN_CAL}	Raw data acquired at temperature of 30 °C, V _{DDA} =3.3 V	0x1FFF 7A2A - 0x1FFF 7A2B

5.3.25 DAC electrical characteristics

Table 74. DAC characteristics

Symbol	Parameter	Min	Тур	Max	Unit	Comments
V_{DDA}	Analog supply voltage	1.8 ⁽¹⁾	-	3.6	V	
V _{REF+}	Reference supply voltage	1.8 ⁽¹⁾	-	3.6	V	$V_{REF+} \le V_{DDA}$
V _{SSA}	Ground	0	-	0	V	

^{2.} Shortest sampling time can be determined in the application by multiple iterations.

^{2.} Guaranteed by design.

Table 74. DAC characteristics (continued)

Symbol	Parameter	Min	Тур	Max	Unit	Comments		
R _{LOAD} ⁽²⁾	Resistive load with buffer ON	5	-	-	kΩ			
R _O ⁽²⁾	Impedance output with buffer OFF	-	-	15	kΩ	When the buffer is OFF, the Minimum resistive load between DAC_OUT and V_{SS} to have a 1% accuracy is 1.5 M Ω		
C _{LOAD} ⁽²⁾	Capacitive load	i	-	50	pF	Maximum capacitive load at DAC_OUT pin (when the buffer is ON).		
DAC_OUT min ⁽²⁾	Lower DAC_OUT voltage with buffer ON	0.2	-	-	٧	It gives the maximum output excursion of the DAC. It corresponds to 12-bit input code		
DAC_OUT max ⁽²⁾	Higher DAC_OUT voltage with buffer ON	-	-	V _{DDA} – 0.2	٧	$(0x0E0)$ to $(0xF1C)$ at V_{REF+} = 3.6 V and $(0x1C7)$ to $(0xE38)$ at V_{REF+} = 1.8 V		
DAC_OUT min ⁽²⁾	Lower DAC_OUT voltage with buffer OFF	-	0.5	-	mV	It gives the maximum output		
DAC_OUT max ⁽²⁾	Higher DAC_OUT voltage with buffer OFF	-	-	V _{REF+} – 1LSB	٧	excursion of the DAC.		
(4)	DAC DC V _{REF} current consumption in quiescent mode (Standby mode)	-	170	240	μΑ	With no load, worst code (0x800) at V_{REF+} = 3.6 V in terms of DC consumption on the inputs		
I _{VREF+} ⁽⁴⁾		-	50	75		With no load, worst code (0xF1C) at V _{REF+} = 3.6 V in terms of DC consumption on the inputs		
	DAC DC VDDA current consumption in quiescent mode ⁽³⁾	-	280	380	μA	With no load, middle code (0x800) on the inputs		
I _{DDA} ⁽⁴⁾		-	475	625	μA	With no load, worst code (0xF1C) at V _{REF+} = 3.6 V in terms of DC consumption on the inputs		
DNL ⁽⁴⁾	Differential non linearity Difference between two consecutive code-1LSB)	-	-	±0.5	LSB	Given for the DAC in 10-bit configuration.		
		-	-	±2	LSB	Given for the DAC in 12-bit configuration.		
INL ⁽⁴⁾	Integral non linearity (difference between measured value at Code i and the value at Code i on a line drawn between Code 0 and last Code 1023)	-	-	±1	LSB	Given for the DAC in 10-bit configuration.		
		-	-	±4	LSB	Given for the DAC in 12-bit configuration.		



Table 74. DAC characteristics (continued)

Symbol	Parameter	Min	Тур	Max	Unit	Comments
Offset ⁽⁴⁾	Offset error (difference between measured value at Code (0x800) and the ideal value = V _{REF+} /2)	-	-	±10	mV	Given for the DAC in 12-bit configuration
		-	-	±3	LSB	Given for the DAC in 10-bit at V _{REF+} = 3.6 V
		-	-	±12	LSB	Given for the DAC in 12-bit at V _{REF+} = 3.6 V
Gain error ⁽⁴⁾	Gain error	-	-	±0.5	%	Given for the DAC in 12-bit configuration
t _{SETTLING} ⁽⁴⁾	Settling time (full scale: for a 10-bit input code transition between the lowest and the highest input codes when DAC_OUT reaches final value ±4LSB	-	3	6	μs	$C_{LOAD} \le 50 \text{ pF},$ $R_{LOAD} \ge 5 \text{ k}\Omega$
THD ⁽⁴⁾	Total Harmonic Distortion Buffer ON	-	-	-	dB	$\begin{aligned} &C_{LOAD} \leq \ 50 \ pF, \\ &R_{LOAD} \geq 5 \ k\Omega \end{aligned}$
Update rate ⁽²⁾	Max frequency for a correct DAC_OUT change when small variation in the input code (from code i to i+1LSB)	-	-	1	MS/s	$C_{LOAD} \le 50 \text{ pF},$ $R_{LOAD} \ge 5 \text{ k}\Omega$
t _{WAKEUP} (4)	Wakeup time from off state (Setting the ENx bit in the DAC Control register)	-	6.5	10	μs	$C_{LOAD} \le 50$ pF, $R_{LOAD} \ge 5$ k Ω input code between lowest and highest possible ones.
PSRR+ (2)	Power supply rejection ratio (to V _{DDA}) (static DC measurement)	-	-67	-40	dB	No R _{LOAD} , C _{LOAD} = 50 pF

V_{DD}/V_{DDA} minimum value of 1.7 V is obtained when the device operates in reduced temperature range, and with the use of an external power supply supervisor (refer to Section: Internal reset OFF).

^{2.} Guaranteed by design.

^{3.} The quiescent mode corresponds to a state where the DAC maintains a stable output level to ensure that no dynamic consumption occurs.

^{4.} Guaranteed by characterization.

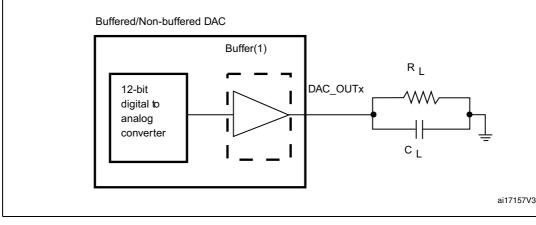


Figure 53. 12-bit buffered /non-buffered DAC

 The DAC integrates an output buffer that can be used to reduce the output impedance and to drive external loads directly without the use of an external operational amplifier. The buffer can be bypassed by configuring the BOFFx bit in the DAC_CR register.

5.3.26 FSMC characteristics

Unless otherwise specified, the parameters given in *Table 75* to *Table 86* for the FSMC interface are derived from tests performed under the ambient temperature, f_{HCLK} frequency and V_{DD} supply voltage conditions summarized in *Table 14*, with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: 0.5V_{DD}

Refer to Section Section 5.3.16: I/O port characteristics for more details on the input/output characteristics.

Asynchronous waveforms and timings

Figure 54 through Figure 57 represent asynchronous waveforms and Table 75 through Table 78 provide the corresponding timings. The results shown in these tables are obtained with the following FSMC configuration:

- AddressSetupTime = 1
- AddressHoldTime = 0x1
- DataSetupTime = 0x1
- BusTurnAroundDuration = 0x0

In all timing tables, the T_{HCLK} is the HCLK clock period.

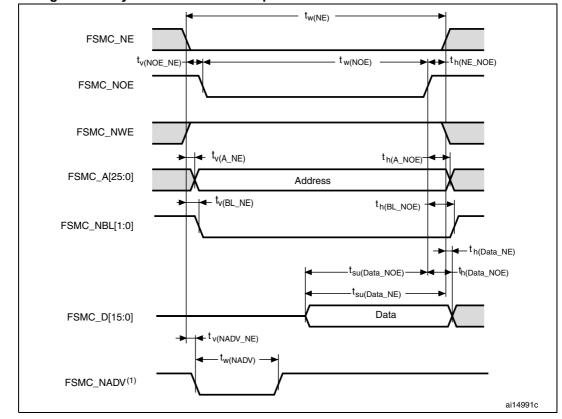


Figure 54. Asynchronous non-multiplexed SRAM/PSRAM/NOR read waveforms

1. Mode 2/B, C and D only. In Mode 1, FSMC_NADV is not used.

Table 75. Asynchronous non-multiplexed SRAM/PSRAM/NOR read timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{w(NE)}	FSMC_NE low time	2T _{HCLK} -0.5	2 T _{HCLK} +1	ns
t _{v(NOE_NE)}	FSMC_NEx low to FSMC_NOE low	0.5	3	ns
t _{w(NOE)}	FSMC_NOE low time	2T _{HCLK} -2	2T _{HCLK} + 2	ns
t _{h(NE_NOE)}	FSMC_NOE high to FSMC_NE high hold time	0	-	ns
t _{v(A_NE)}	FSMC_NEx low to FSMC_A valid	-	4.5	ns
t _{h(A_NOE)}	Address hold time after FSMC_NOE high	4	-	ns
t _{v(BL_NE)}	FSMC_NEx low to FSMC_BL valid	-	1.5	ns
t _{h(BL_NOE)}	FSMC_BL hold time after FSMC_NOE high	0	-	ns
t _{su(Data_NE)}	Data to FSMC_NEx high setup time	T _{HCLK} +4	-	ns
t _{su(Data_NOE)}	Data to FSMC_NOEx high setup time	T _{HCLK} +4	-	ns
t _{h(Data_NOE)}	Data hold time after FSMC_NOE high	0	-	ns
t _{h(Data_NE)}	Data hold time after FSMC_NEx high	0	-	ns
t _{v(NADV_NE)}	FSMC_NEx low to FSMC_NADV low		2	ns
t _{w(NADV)}	FSMC_NADV low time	-	T _{HCLK}	ns

^{1.} $C_L = 30 pF$.

^{2.} Guaranteed by characterization.

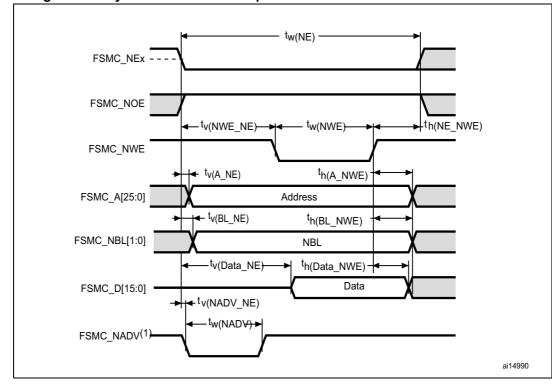


Figure 55. Asynchronous non-multiplexed SRAM/PSRAM/NOR write waveforms

1. Mode 2/B, C and D only. In Mode 1, FSMC_NADV is not used.

Table 76. Asynchronous non-multiplexed SRAM/PSRAM/NOR write timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit		
t _{w(NE)}	FSMC_NE low time	3T _{HCLK}	3T _{HCLK} + 4	ns		
t _{v(NWE_NE)}	FSMC_NEx low to FSMC_NWE low	T _{HCLK} -0.5	T _{HCLK} +0.5	ns		
t _{w(NWE)}	FSMC_NWE low time	T _{HCLK} -1	T _{HCLK} +2	ns		
t _{h(NE_NWE)}	FSMC_NWE high to FSMC_NE high hold time	T _{HCLK} -1	-	ns		
t _{v(A_NE)}	FSMC_NEx low to FSMC_A valid	-	0	ns		
t _{h(A_NWE)}	Address hold time after FSMC_NWE high	T _{HCLK} -2	T _{HCLK} -2	T _{HCLK} – 2	-	ns
t _{v(BL_NE)}	FSMC_NEx low to FSMC_BL valid	-	1.5	ns		
t _{h(BL_NWE)}	FSMC_BL hold time after FSMC_NWE high	T _{HCLK} -1	-	ns		
t _{v(Data_NE)}	Data to FSMC_NEx low to Data valid	-	T _{HCLK} +3	ns		
t _{h(Data_NWE)}	Data hold time after FSMC_NWE high	T _{HCLK} -1	-	ns		
t _{v(NADV_NE)}	FSMC_NEx low to FSMC_NADV low	-	2	ns		
t _{w(NADV)}	FSMC_NADV low time	-	T _{HCLK} +0.5	ns		

^{1.} $C_L = 30 pF$.

^{2.} Guaranteed by characterization.

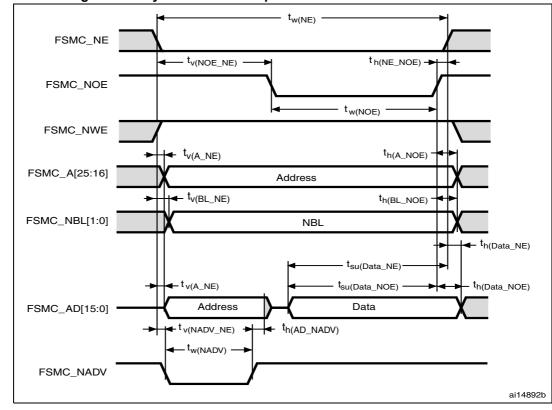


Figure 56. Asynchronous multiplexed PSRAM/NOR read waveforms

Table 77. Asynchronous multiplexed PSRAM/NOR read timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{w(NE)}	FSMC_NE low time	3T _{HCLK} -1	3T _{HCLK} +1	ns
t _{v(NOE_NE)}	FSMC_NEx low to FSMC_NOE low	2T _{HCLK} -0.5	2T _{HCLK} +0.5	ns
t _{w(NOE)}	FSMC_NOE low time	T _{HCLK} -1	T _{HCLK} +1	ns
t _{h(NE_NOE)}	FSMC_NOE high to FSMC_NE high hold time	0	-	ns
t _{v(A_NE)}	FSMC_NEx low to FSMC_A valid	-	3	ns
t _{v(NADV_NE)}	FSMC_NEx low to FSMC_NADV low	1	2	ns
t _{w(NADV)}	FSMC_NADV low time	T _{HCLK} -2	T _{HCLK} +1	ns
t _{h(AD_NADV)}	FSMC_AD(adress) valid hold time after FSMC_NADV high)	T _{HCLK}	-	ns
t _{h(A_NOE)}	Address hold time after FSMC_NOE high	T _{HCLK} -1	-	ns
t _{h(BL_NOE)}	FSMC_BL time after FSMC_NOE high	0	-	ns
t _{v(BL_NE)}	FSMC_NEx low to FSMC_BL valid	-	2	ns
t _{su(Data_NE)}	Data to FSMC_NEx high setup time	T _{HCLK} +4	-	ns
t _{su(Data_NOE)}	Data to FSMC_NOE high setup time	T _{HCLK} +4	-	ns
t _{h(Data_NE)}	Data hold time after FSMC_NEx high	0	-	ns
t _{h(Data_NOE)}	Data hold time after FSMC_NOE high	0	-	ns

^{1.} $C_1 = 30 \text{ pF}.$

^{2.} Guaranteed by characterization.

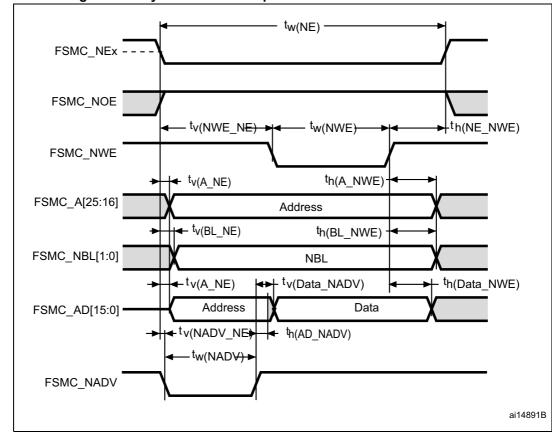


Figure 57. Asynchronous multiplexed PSRAM/NOR write waveforms

Table 78. Asynchronous multiplexed PSRAM/NOR write timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{w(NE)}	FSMC_NE low time	4T _{HCLK} -0.5	4T _{HCLK} +3	ns
t _{v(NWE_NE)}	FSMC_NEx low to FSMC_NWE low	T _{HCLK} -0.5	T _{HCLK} -0.5	ns
t _{w(NWE)}	FSMC_NWE low tim e	2T _{HCLK} -0.5	2T _{HCLK} +3	ns
t _{h(NE_NWE)}	FSMC_NWE high to FSMC_NE high hold time	T _{HCLK}	1	ns
t _{v(A_NE)}	FSMC_NEx low to FSMC_A valid	-	0	ns
t _{v(NADV_NE)}	FSMC_NEx low to FSMC_NADV low	1	2	ns
t _{w(NADV)}	FSMC_NADV low time	T _{HCLK} -2	T _{HCLK} + 1	ns
t _{h(AD_NADV)}	FSMC_AD(address) valid hold time after FSMC_NADV high)	T _{HCLK} -2	-	ns
t _{h(A_NWE)}	Address hold time after FSMC_NWE high	T _{HCLK}	-	ns
t _{h(BL_NWE)}	FSMC_BL hold time after FSMC_NWE high	T _{HCLK} -2	-	ns
t _{v(BL_NE)}	FSMC_NEx low to FSMC_BL valid		1.5	ns
t _{v(Data_NADV)}	FSMC_NADV high to Data valid		T _{HCLK} -0.5	ns
t _{h(Data_NWE)}	Data hold time after FSMC_NWE high	T _{HCLK}	-	ns

^{1.} $C_L = 30 pF$.



2. Guaranteed by characterization.

Synchronous waveforms and timings

Figure 58 through Figure 61 represent synchronous waveforms and Table 80 through Table 82 provide the corresponding timings. The results shown in these tables are obtained with the following FSMC configuration:

- BurstAccessMode = FSMC_BurstAccessMode_Enable;
- MemoryType = FSMC_MemoryType_CRAM;
- WriteBurst = FSMC_WriteBurst_Enable;
- CLKDivision = 1; (0 is not supported, see the STM32F40xxx/41xxx reference manual)
- DataLatency = 1 for NOR Flash; DataLatency = 0 for PSRAM

In all timing tables, the T_{HCLK} is the HCLK clock period (with maximum FSMC_CLK = 60 MHz).

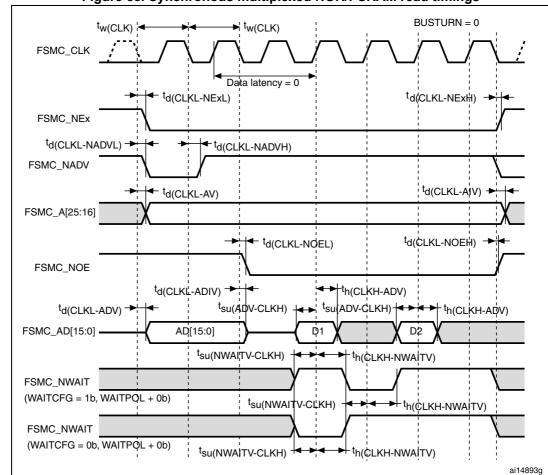


Figure 58. Synchronous multiplexed NOR/PSRAM read timings

Table 79. Synchronous multiplexed NOR/PSRAM read timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{w(CLK)}	FSMC_CLK period	2T _{HCLK}	-	ns
t _{d(CLKL-NExL)}	FSMC_CLK low to FSMC_NEx low (x=02)	-	0	ns
t _{d(CLKL-NExH)}	FSMC_CLK low to FSMC_NEx high (x= 02)	2	-	ns
t _{d(CLKL-NADVL)}	FSMC_CLK low to FSMC_NADV low	-	2	ns
t _{d(CLKL-NADVH)}	FSMC_CLK low to FSMC_NADV high	2	-	ns
t _{d(CLKL-AV)}	FSMC_CLK low to FSMC_Ax valid (x=1625)	-	0	ns
t _{d(CLKL-AIV)}	FSMC_CLK low to FSMC_Ax invalid (x=1625)	0	-	ns
t _{d(CLKL-NOEL)}	FSMC_CLK low to FSMC_NOE low	-	0	ns
t _{d(CLKL-NOEH)}	FSMC_CLK low to FSMC_NOE high	2	-	ns
t _{d(CLKL-ADV)}	FSMC_CLK low to FSMC_AD[15:0] valid	-	4.5	ns
t _{d(CLKL-ADIV)}	FSMC_CLK low to FSMC_AD[15:0] invalid	0	-	ns
t _{su(ADV-CLKH)}	FSMC_A/D[15:0] valid data before FSMC_CLK high	6	-	ns
t _{h(CLKH-ADV)}	FSMC_A/D[15:0] valid data after FSMC_CLK high	0	-	ns
t _{su(NWAIT-CLKH)}	FSMC_NWAIT valid before FSMC_CLK high	4	-	ns
t _{h(CLKH-NWAIT)}	FSMC_NWAIT valid after FSMC_CLK high	0	-	ns

^{1.} $C_L = 30 \text{ pF}.$

^{2.} Guaranteed by characterization.

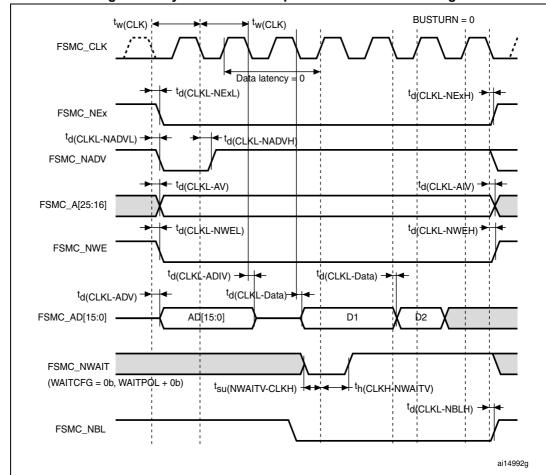


Figure 59. Synchronous multiplexed PSRAM write timings

Table 80. Synchronous multiplexed PSRAM write timings⁽¹⁾⁽²⁾

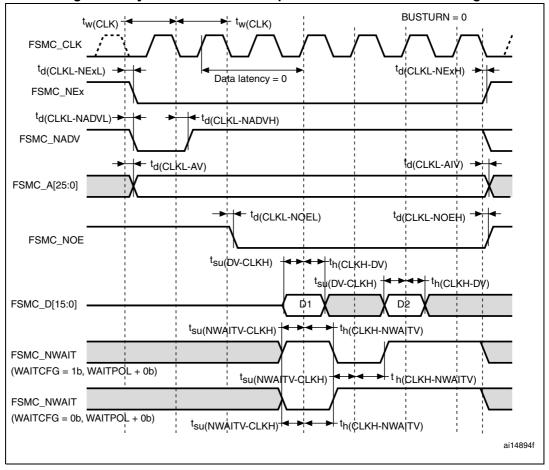
Symbol	Parameter	Min	Max	Unit
			-	
t _{w(CLK)}	FSMC_CLK period	2T _{HCLK}	-	ns
t _{d(CLKL-NExL)}	FSMC_CLK low to FSMC_NEx low (x=02)	-	1	ns
t _{d(CLKL-NExH)}	FSMC_CLK low to FSMC_NEx high (x= 02)	1	-	ns
t _{d(CLKL-NADVL)}	FSMC_CLK low to FSMC_NADV low	1	0	ns
t _{d(CLKL-} NADVH)	FSMC_CLK low to FSMC_NADV high	0	ı	ns
t _{d(CLKL-AV)}	FSMC_CLK low to FSMC_Ax valid (x=1625)	-	0	ns
t _{d(CLKL-AIV)}	FSMC_CLK low to FSMC_Ax invalid (x=1625)	8	ı	ns
t _{d(CLKL-NWEL)}	FSMC_CLK low to FSMC_NWE low	-	0.5	ns
t _{d(CLKL-NWEH)}	FSMC_CLK low to FSMC_NWE high	0	-	ns
t _{d(CLKL-ADIV)}	FSMC_CLK low to FSMC_AD[15:0] invalid	0	-	ns
t _{d(CLKL-DATA)}	FSMC_A/D[15:0] valid data after FSMC_CLK low	-	3	ns

Table 80. Synchronous multiplexed PSRAM write timings⁽¹⁾⁽²⁾ (continued)

Symbol	Parameter	Min	Max	Unit
t _{d(CLKL-NBLH)}	FSMC_CLK low to FSMC_NBL high	0	-	ns
t _{su(NWAIT} - CLKH)	FSMC_NWAIT valid before FSMC_CLK high	4	1	ns
t _{h(CLKH-NWAIT)}	FSMC_NWAIT valid after FSMC_CLK high	0	-	ns

^{1.} $C_L = 30 pF$.

Figure 60. Synchronous non-multiplexed NOR/PSRAM read timings



^{2.} Guaranteed by characterization.

Table 81. Synchronous non-multiplexed NOR/PSRAM read timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{w(CLK)}	FSMC_CLK period	2T _{HCLK} -0.5	-	ns
t _{d(CLKL-NExL)}	FSMC_CLK low to FSMC_NEx low (x=02)	-	0.5	ns
t _{d(CLKL-NExH)}	FSMC_CLK low to FSMC_NEx high (x= 02)	0	-	ns
t _{d(CLKL-NADVL)}	FSMC_CLK low to FSMC_NADV low	-	2	ns
t _{d(CLKL-NADVH)}	FSMC_CLK low to FSMC_NADV high	3	-	ns
t _{d(CLKL-AV)}	FSMC_CLK low to FSMC_Ax valid (x=1625)	-	0	ns
t _{d(CLKL-AIV)}	FSMC_CLK low to FSMC_Ax invalid (x=1625)	2	-	ns
t _{d(CLKL-NOEL)}	FSMC_CLK low to FSMC_NOE low	-	0.5	ns
t _{d(CLKL-NOEH)}	FSMC_CLK low to FSMC_NOE high	1.5	-	ns
t _{su(DV-CLKH)}	FSMC_D[15:0] valid data before FSMC_CLK high	6	-	ns
t _{h(CLKH-DV)}	FSMC_D[15:0] valid data after FSMC_CLK high	3	-	ns
t _{su(NWAIT-CLKH)}	FSMC_NWAIT valid before FSMC_CLK high	4	-	ns
t _{h(CLKH-NWAIT)}	FSMC_NWAIT valid after FSMC_CLK high	0	-	ns

^{1.} $C_L = 30 pF$.

^{2.} Guaranteed by characterization.

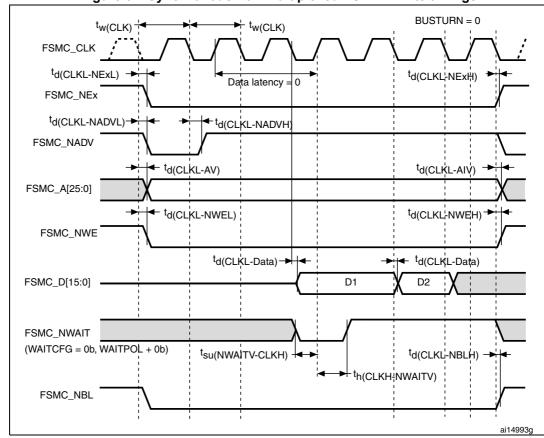


Figure 61. Synchronous non-multiplexed PSRAM write timings

Table 82. Synchronous non-multiplexed PSRAM write timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{w(CLK)}	FSMC_CLK period	2T _{HCLK}	-	ns
td(CLKL-NExL)	FSMC_CLK low to FSMC_NEx low (x=02)	-	1	ns
t _{d(CLKL-NExH)}	FSMC_CLK low to FSMC_NEx high (x= 02)	1	-	ns
t _{d(CLKL-NADVL)}	FSMC_CLK low to FSMC_NADV low	-	7	ns
t _{d(CLKL-NADVH)}	FSMC_CLK low to FSMC_NADV high	6	-	ns
t _{d(CLKL-AV)}	FSMC_CLK low to FSMC_Ax valid (x=1625)	-	0	ns
t _{d(CLKL-AIV)}	FSMC_CLK low to FSMC_Ax invalid (x=1625)	6	-	ns
t _{d(CLKL-NWEL)}	FSMC_CLK low to FSMC_NWE low	-	1	ns
t _{d(CLKL-NWEH)}	FSMC_CLK low to FSMC_NWE high	2	-	ns
t _{d(CLKL-Data)}	FSMC_D[15:0] valid data after FSMC_CLK low	-	3	ns
t _{d(CLKL-NBLH)}	FSMC_CLK low to FSMC_NBL high	3	-	ns
t _{su(NWAIT-CLKH)}	FSMC_NWAIT valid before FSMC_CLK high	4	-	ns
t _{h(CLKH-NWAIT)}	FSMC_NWAIT valid after FSMC_CLK high	0	-	ns

^{1.} $C_L = 30 pF$.

^{2.} Guaranteed by characterization.

PC Card/CompactFlash controller waveforms and timings

Figure 62 through *Figure 67* represent synchronous waveforms, and *Table 83* and *Table 84* provide the corresponding timings. The results shown in this table are obtained with the following FSMC configuration:

- COM.FSMC_SetupTime = 0x04;
- COM.FSMC_WaitSetupTime = 0x07;
- COM.FSMC HoldSetupTime = 0x04;
- COM.FSMC_HiZSetupTime = 0x00;
- ATT.FSMC SetupTime = 0x04;
- ATT.FSMC_WaitSetupTime = 0x07;
- ATT.FSMC_HoldSetupTime = 0x04;
- ATT.FSMC HiZSetupTime = 0x00;
- IO.FSMC SetupTime = 0x04;
- IO.FSMC_WaitSetupTime = 0x07;
- IO.FSMC HoldSetupTime = 0x04;
- IO.FSMC HiZSetupTime = 0x00;
- TCLRSetupTime = 0;
- TARSetupTime = 0.

In all timing tables, the T_{HCLK} is the HCLK clock period.

FSMC NCE4 2(1) FSMC NCE4 1 th(NCEx-AI)- t_V(NCEx-A) FSMC_A[10:0] th(NCEx-NREG) td(NREG-NCEx) th(NCEx-NIORD) td(NIORD-NCEx) th(NCEx-NIOWR) FSMC_NREG FSMC_NIOWR FSMC_NIORD FSMC_NWE ^td(NCE4 1-NOE)✓ ^tw(NOE) FSMC_NOE t_{su(D-NOE)} th(NOE-D) FSMC D[15:0] ai14895b

Figure 62. PC Card/CompactFlash controller waveforms for common memory read access

1. FSMC_NCE4_2 remains high (inactive during 8-bit access.

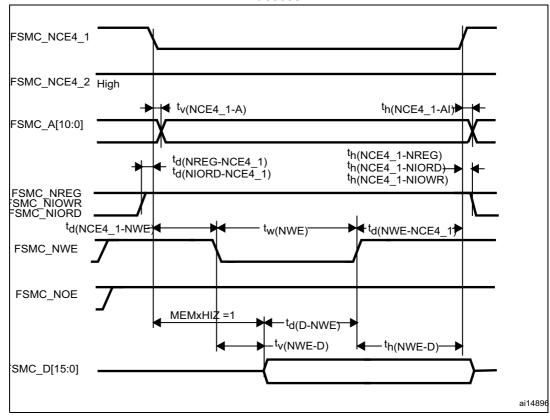


Figure 63. PC Card/CompactFlash controller waveforms for common memory write access

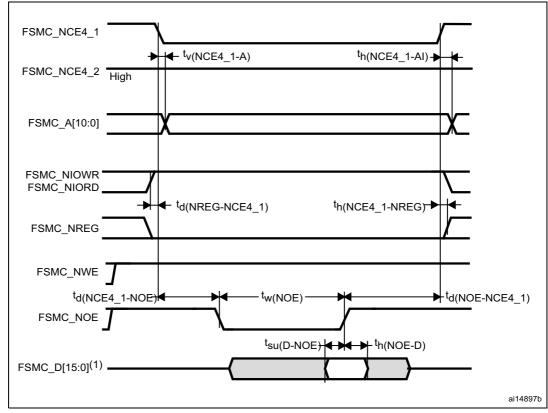


Figure 64. PC Card/CompactFlash controller waveforms for attribute memory read access

1. Only data bits 0...7 are read (bits 8...15 are disregarded).

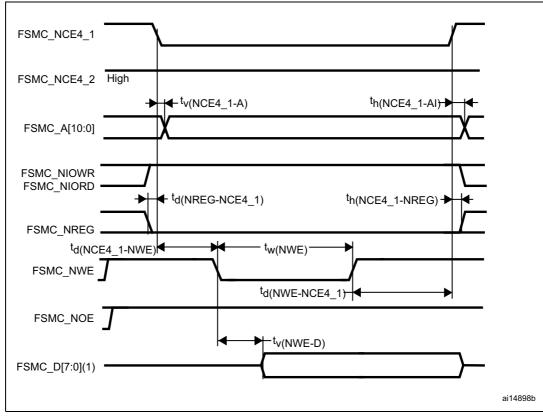


Figure 65. PC Card/CompactFlash controller waveforms for attribute memory write access

1. Only data bits 0...7 are driven (bits 8...15 remains Hi-Z).

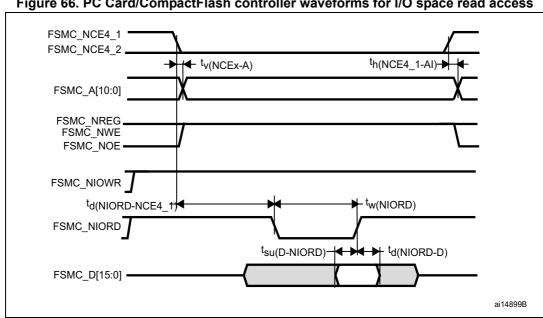


Figure 66. PC Card/CompactFlash controller waveforms for I/O space read access

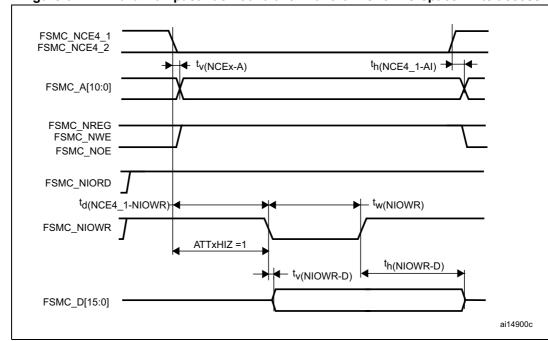


Figure 67. PC Card/CompactFlash controller waveforms for I/O space write access

Table 83. Switching characteristics for PC Card/CF read and write cycles in attribute/common space $^{(1)(2)}$

Symbol	Parameter	Min	Max	Unit
t _{v(NCEx-A)}	FSMC_Ncex low to FSMC_Ay valid	-	0	ns
t _{h(NCEx_AI)}	FSMC_NCEx high to FSMC_Ax invalid	4	-	ns
t _{d(NREG-NCEx)}	FSMC_NCEx low to FSMC_NREG valid	-	3.5	ns
t _{h(NCEx-NREG)}	FSMC_NCEx high to FSMC_NREG invalid	T _{HCLK} +4	-	ns
t _{d(NCEx-NWE)}	FSMC_NCEx low to FSMC_NWE low	-	5T _{HCLK} +0.5	ns
t _{d(NCEx-NOE)}	FSMC_NCEx low to FSMC_NOE low	-	5T _{HCLK} +0.5	ns
t _{w(NOE)}	FSMC_NOE low width	8T _{HCLK} -1	8T _{HCLK} +1	ns
t _{d(NOE_NCEx)}	FSMC_NOE high to FSMC_NCEx high	5T _{HCLK} +2.5	-	ns
t _{su (D-NOE)}	FSMC_D[15:0] valid data before FSMC_NOE high	4.5	-	ns
t _{h(N0E-D)}	FSMC_N0E high to FSMC_D[15:0] invalid	3	-	ns
t _{w(NWE)}	FSMC_NWE low width	8T _{HCLK} -0.5	8T _{HCLK} + 3	ns
t _{d(NWE_NCEx)}	FSMC_NWE high to FSMC_NCEx high	5T _{HCLK} -1	-	ns
t _{d(NCEx-NWE)}	FSMC_NCEx low to FSMC_NWE low	-	5T _{HCLK} + 1	ns
t _{v(NWE-D)}	FSMC_NWE low to FSMC_D[15:0] valid	-	0	ns
t _h (NWE-D)	FSMC_NWE high to FSMC_D[15:0] invalid	8T _{HCLK} –1	-	ns
t _d (D-NWE)	FSMC_D[15:0] valid before FSMC_NWE high	13T _{HCLK} –1	-	ns

^{1.} $C_L = 30 pF$.

^{2.} Guaranteed by characterization.

Symbol Parameter Min Max Unit FSMC NIOWR low width 8T_{HCLK}-1 ns t_{w(NIOWR)} FSMC_NIOWR low to FSMC_D[15:0] valid 5T_{HCLK}-1 ns t_{v(NIOWR-D)} FSMC_NIOWR high to FSMC_D[15:0] invalid 8T_{HCLK}-2 ns t_{h(NIOWR-D)} FSMC_NCE4_1 low to FSMC_NIOWR valid 5T_{HCLK}+ 2.5 ns t_{d(NCE4_1-NIOWR)} FSMC_NCEx high to FSMC_NIOWR invalid 5T_{HCLK}-1.5 ns th(NCEx-NIOWR) FSMC_NCEx low to FSMC_NIORD valid 5T_{HCLK}+ 2 t_{d(NIORD-NCEx)} ns FSMC NCEx high to FSMC NIORD) valid 5T_{HCLK}- 1.5 t_{h(NCEx-NIORD)} ns FSMC_NIORD low width 8T_{HCLK}-0.5 _ t_{w(NIORD)} ns FSMC D[15:0] valid before FSMC NIORD high 9 ns t_{su(D-NIORD)} FSMC_D[15:0] valid after FSMC_NIORD high 0 _ ns t_{d(NIORD-D)}

Table 84. Switching characteristics for PC Card/CF read and write cycles in I/O space⁽¹⁾⁽²⁾

NAND controller waveforms and timings

Figure 68 through *Figure 71* represent synchronous waveforms, and *Table 85* and *Table 86* provide the corresponding timings. The results shown in this table are obtained with the following FSMC configuration:

- COM.FSMC SetupTime = 0x01;
- COM.FSMC_WaitSetupTime = 0x03;
- COM.FSMC_HoldSetupTime = 0x02;
- COM.FSMC_HiZSetupTime = 0x01;
- ATT.FSMC_SetupTime = 0x01;
- ATT.FSMC_WaitSetupTime = 0x03;
- ATT.FSMC HoldSetupTime = 0x02;
- ATT.FSMC HiZSetupTime = 0x01;
- Bank = FSMC_Bank_NAND;
- MemoryDataWidth = FSMC MemoryDataWidth 16b;
- ECC = FSMC_ECC_Enable;
- ECCPageSize = FSMC_ECCPageSize_512Bytes;
- TCLRSetupTime = 0;
- TARSetupTime = 0.

In all timing tables, the T_{HCLK} is the HCLK clock period.

^{1.} $C_1 = 30 pF$.

^{2.} Guaranteed by characterization.

FSMC_NCEX

ALE (FSMC_A17)
CLE (FSMC_A16)

FSMC_NWE

FSMC_NOE (NRE)

t_d(ALE-NOE)

t_h(NOE-ALE)

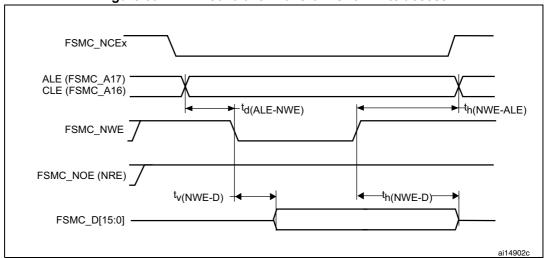
t_h(NOE-D)

FSMC_D[15:0]

ai14901c

Figure 68. NAND controller waveforms for read access





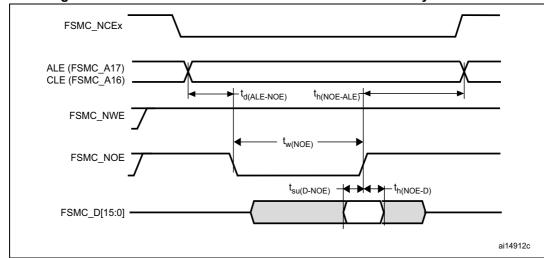


Figure 70. NAND controller waveforms for common memory read access

Figure 71. NAND controller waveforms for common memory write access

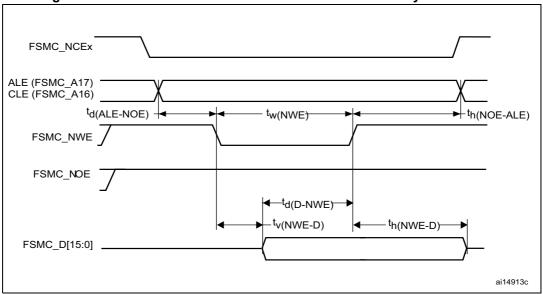


Table 85. Switching characteristics for NAND Flash read cycles⁽¹⁾

Symbol	Parameter	Min	Max	Unit
t _{w(N0E)}	FSMC_NOE low width	4T _{HCLK} - 0.5	4T _{HCLK} + 3	ns
t _{su(D-NOE)}	FSMC_D[15-0] valid data before FSMC_NOE high	10	-	ns
t _{h(NOE-D)}	FSMC_D[15-0] valid data after FSMC_NOE high	0	-	ns
t _{d(ALE-NOE)}	FSMC_ALE valid before FSMC_NOE low	-	3T _{HCLK}	ns
t _{h(NOE-ALE)}	FSMC_NWE high to FSMC_ALE invalid	3T _{HCLK} -2	-	ns

^{1.} $C_L = 30 pF$.

Symbol	Parameter	Min	Max	Unit
t _{w(NWE)}	FSMC_NWE low width	4T _{HCLK} -1	4T _{HCLK} + 3	ns
t _{v(NWE-D)}	FSMC_NWE low to FSMC_D[15-0] valid	-	0	ns
t _{h(NWE-D)}	FSMC_NWE high to FSMC_D[15-0] invalid	3T _{HCLK} –2	-	ns
t _{d(D-NWE)}	FSMC_D[15-0] valid before FSMC_NWE high	5T _{HCLK} -3	-	ns
t _{d(ALE-NWE)}	FSMC_ALE valid before FSMC_NWE low	-	3T _{HCLK}	ns
t _{h(NWE-ALE)}	FSMC_NWE high to FSMC_ALE invalid	3T _{HCLK} -2	-	ns

Table 86. Switching characteristics for NAND Flash write cycles⁽¹⁾

5.3.27 Camera interface (DCMI) timing specifications

Unless otherwise specified, the parameters given in *Table 87* for DCMI are derived from tests performed under the ambient temperature, f_{HCLK} frequency and V_{DD} supply voltage summarized in *Table 13*, with the following configuration:

PCK polarity: falling

VSYNC and HSYNC polarity: high

Data format: 14 bits

DCMI_PIXCLK

DCMI_PIXCLK

DCMI_HSYNC

DCMI_HSYNC

DCMI_VSYNC

DATA[0:13]

MS32414V2

Figure 72. DCMI timing diagram

Table 87. DCMI characteristics⁽¹⁾

Symbol	Parameter	Min	Max	Unit
	Frequency ratio DCMI_PIXCLK/f _{HCLK}	-	0.4	
DCMI_PIXCLK	Pixel clock input	-	54	MHz
D _{pixel}	Pixel clock input duty cycle	30	70	%

^{1.} $C_L = 30 pF$.

Symbol	Parameter	Min	Max	Unit
t _{su(DATA)}	Data input setup time	2.5	-	
t _{h(DATA)}	Data hold time	1	-	
t _{su(HSYNC)} , t _{su(VSYNC)}	HSYNC/VSYNC input setup time	2	-	ns
t _{h(HSYNC)} , t _{h(VSYNC)}	HSYNC/VSYNC input hold time	0.5	-	

Table 87. DCMI characteristics⁽¹⁾ (continued)

5.3.28 SD/SDIO MMC card host interface (SDIO) characteristics

Unless otherwise specified, the parameters given in *Table 88* are derived from tests performed under ambient temperature, f_{PCLK_X} frequency and V_{DD} supply voltage conditions summarized in *Table 14* with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: 0.5V_{DD}

Refer to Section 5.3.16: I/O port characteristics for more details on the input/output characteristics.

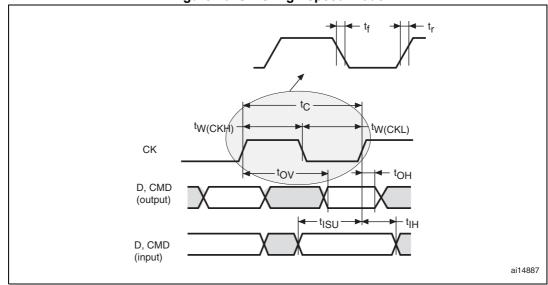


Figure 73. SDIO high-speed mode

^{1.} Guaranteed by characterization.

ai14888

CK

D, CMD
(output)

Figure 74. SD default mode

Table 88. Dynamic characteristics: SD / MMC characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
f _{PP}	Clock frequency in data transfer mode		0		48	MHz	
	SDIO_CK/f _{PCLK2} frequency ratio		-	-	8/3	ī	
t _{W(CKL)}	Clock low time	f _{PP} = 48 MHz	8.5	9	-	ns	
t _{W(CKH)}	Clock high time	f _{PP} = 48 MHz	8.3	10	-	115	
CMD, D input	ts (referenced to CK) in MMC and SD HS mode						
t _{ISU}	Input setup time HS	f _{PP} = 48 MHz	3	-	-	ns	
t _{IH}	Input hold time HS	f _{PP} = 48 MHz	0	-	-	115	
CMD, D outp	CMD, D outputs (referenced to CK) in MMC and SD HS mode						
t _{OV}	Output valid time HS	f _{PP} = 48 MHz	-	4.5	6	no	
t _{OH}	Output hold time HS	f _{PP} = 48 MHz	1	-	-	ns	
CMD, D input	ts (referenced to CK) in SD default mode						
t _{ISUD}	Input setup time SD	f _{PP} = 24 MHz	1.5	-	-	ne	
t _{IHD}	Input hold time SD	f _{PP} = 24 MHz	0.5	-	-	ns	
CMD, D outp	CMD, D outputs (referenced to CK) in SD default mode						
t _{OVD}	Output valid default time SD f _{PP} = 24 MH		-	4.5	7	ns	
t _{OHD}	Output hold default time SD	f _{PP} = 24 MHz	0.5	-	-	110	

^{1.} Guaranteed by characterization.

5.3.29 RTC characteristics

Table 89. RTC characteristics

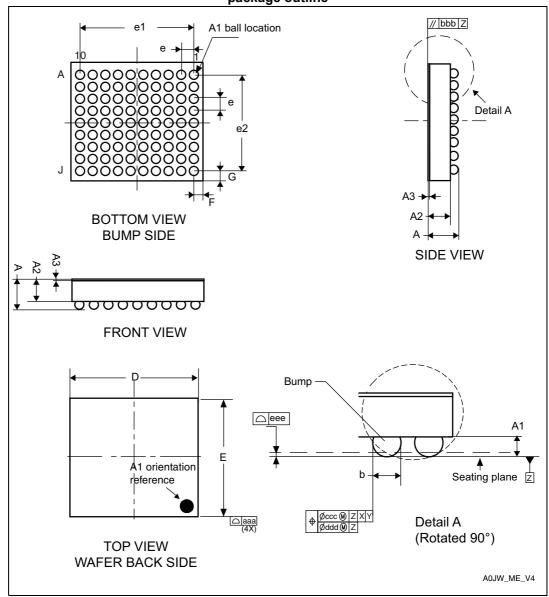
Symbol	Parameter	Conditions	Min	Max
-	f _{PCLK1} /RTCCLK frequency ratio	Any read/write operation from/to an RTC register	4	-

6 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK[®] packages, depending on their level of environmental compliance. ECOPACK[®] specifications, grade definitions and product status are available at: www.st.com. ECOPACK[®] is an ST trademark.

6.1 WLCSP90 package information

Figure 75. WLCSP90 - 4.223 x 3.969 mm, 0.400 mm pitch wafer level chip scale package outline



1. Drawing is not to scale.

Table 90. WLCSP90 - 4.223 x 3.969 mm, 0.400 mm pitch wafer level chip scale package mechanical data

Cumbal	millimeters			inches ⁽¹⁾			
Symbol	Тур	Min	Max	Тур	Min	Max	
A	0.540	0.600	0.585	0.0207	0.0219	0.0230	
A1	-	0.190	-	-	0.0069	-	
A2	-	0.380	-	-	0.0150	-	
A3 ⁽²⁾	-	0.025	-	-	0.0010	-	
b ⁽³⁾	0.240	0.270	0.300	0.0087	0.0098	0.0110	
D	4.188	4.223	4.258	0.1008	0.1022	0.1036	
Е	3.934	3.969	4.004	0.1115	0.1129	0.1143	
е	-	0.400	-	-	0.0157	-	
e1	-	3.600	-	-	0.0787	-	
e2	-	3.200	-	-	0.0787	-	
F	-	0.3115	-	-	0.0117	-	
G	-	0.3845	-	-	0.0171	-	
aaa	-	-	0.100	-	-	0.0039	
bbb	-	-	0.100	-	-	0.0039	
ccc	-	-	0.100	-	-	0.0039	
ddd	-	-	0.050	-	-	0.0020	
eee	-	-	0.050	-	-	0.0020	

- 1. Values in inches are converted from mm and rounded to 4 decimal digits.
- 2. Back side coating.
- 3. Dimension is measured at the maximum bump diameter parallel to primary datum Z.

Figure 76. WLCSP90 - 4.223 x 3.969 mm, 0.400 mm pitch wafer level chip scale recommended footprint

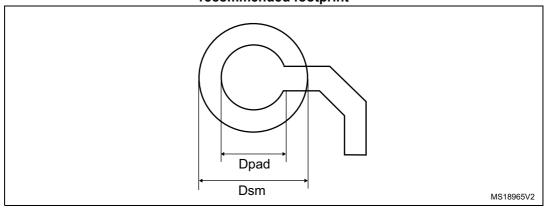


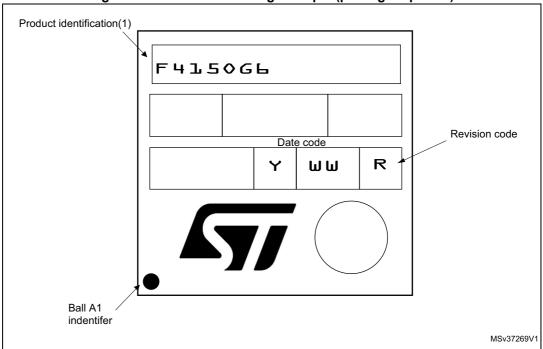
Table 91. WLCSP90 recommended PCB design rules

Dimension	Recommended values
Pitch	0.4 mm
Dpad	260 µm max. (circular) 220 µm recommended
Dsm	300 μm min. (for 260 μm diameter pad)
PCB pad design	Non-solder mask defined via underbump allowed

Device marking for WLCSP90

The following figure gives an example of topside marking and ball A1 position identifier location.

Figure 77. WLCSP90 marking example (package top view)



Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering Samples to run qualification activity.

6.2 LQFP64 package information

SEATING PLANE

O.25 mm

GAUGE PLANE

D1

D3

33

D3

TIDENTIFICATION

SW_ME_V3

Figure 78. LQFP64 - 64-pin, 10 x 10 mm low-profile quad flat package outline

1. Drawing is not to scale.

Table 92. LQFP64 – 64-pin 10 x 10 mm low-profile quad flat package mechanical data

Comphal		millimeters			inches ⁽¹⁾	
Symbol	Min	Тур	Max	Min	Тур	Max
Α	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
С	0.090	-	0.200	0.0035	-	0.0079
D	-	12.000	-	-	0.4724	-
D1	-	10.000	-	-	0.3937	-
D3	-	7.500	-	-	0.2953	-
E	-	12.000	-	-	0.4724	-
E1	-	10.000	-	-	0.3937	-

inches⁽¹⁾ millimeters **Symbol** Min Тур Max Min Тур Max E3 7.500 0.2953 0.500 е 0.0197 Κ 0° 3.5° 7° 0° 3.5° 7° L 0.450 0.600 0.750 0.0177 0.0236 0.0295 L1 1.000 0.0394 CCC 0.080 0.0031

Table 92. LQFP64 – 64-pin 10 x 10 mm low-profile quad flat package mechanical data (continued)

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.

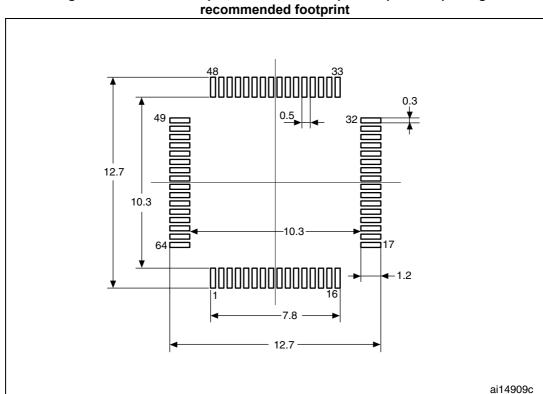


Figure 79. LQFP64 – 64-pin, 10 x 10 mm low-profile quad flat package recommended footprint

- Drawing is not to scale.
- 2. Dimensions are in millimeters.

170/203

Device marking for LQFP64

The following figure gives an example of topside marking and pin 1 position identifier location.

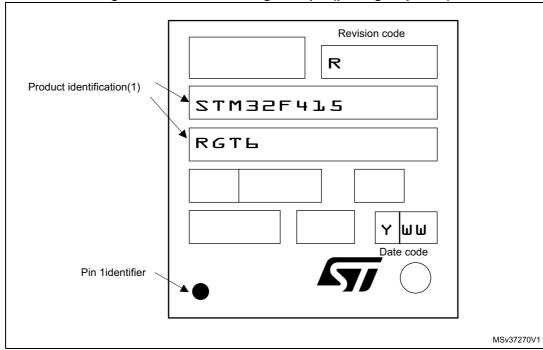


Figure 80. LPQF64 marking example (package top view)

1. Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering Samples to run qualification activity.

6.3 LQPF100 package information

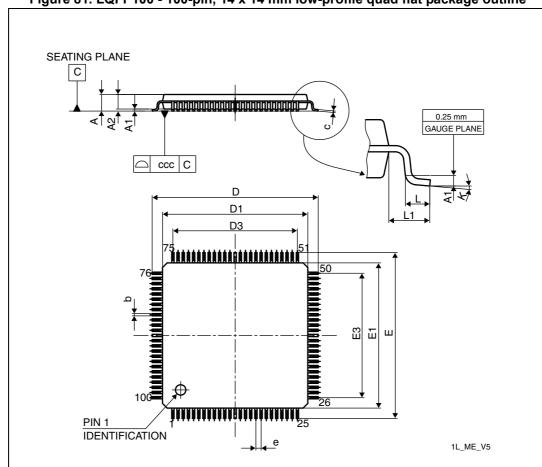


Figure 81. LQFP100 - 100-pin, 14 x 14 mm low-profile quad flat package outline

1. Drawing is not to scale.

Table 93. LQPF100 – 100-pin, 14 x 14 mm low-profile quad flat package mechanical data⁽¹⁾

			oonamoar aat	=			
Comple of		millimeters			inches		
Symbol	Min	Тур	Max	Min	Тур	Max	
Α	-	-	1.600	-	-	0.0630	
A1	0.050	-	0.150	0.0020	-	0.0059	
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571	
b	0.170	0.220	0.270	0.0067	0.0087	0.0106	
С	0.090	-	0.200	0.0035	-	0.0079	
D	15.800	16.000	16.200	0.6220	0.6299	0.6378	
D1	13.800	14.000	14.200	0.5433	0.5512	0.5591	
D3	-	12.000	-	-	0.4724	-	
E	15.80	16.000	16.200	0.6220	0.6299	0.6378	

millimeters inches **Symbol** Min Тур Max Min Тур Max E1 13.800 14.000 14.200 0.5433 0.5512 0.5591 E3 12.000 0.4724 0.500 0.0197 е 0.600 L 0.450 0.750 0.0177 0.0236 0.0295 L1 1.000 0.0394 7° 7° k 0° 3.5° 0° 3.5° CCC 0.080 0.0031

Table 93. LQPF100 – 100-pin, 14 x 14 mm low-profile quad flat package mechanical data⁽¹⁾ (continued)

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.

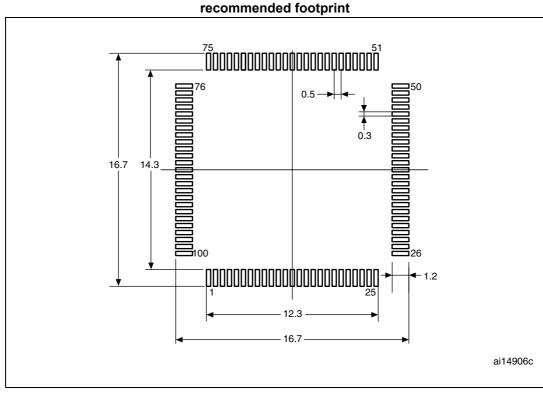


Figure 82. LQFP100 - 100-pin, 14 x 14 mm low-profile quad flat

- 1. Drawing is not to scale.
- 2. Dimensions are in millimeters.

Device marking for LFP100

The following figure gives an example of topside marking and pin 1 position identifier location.

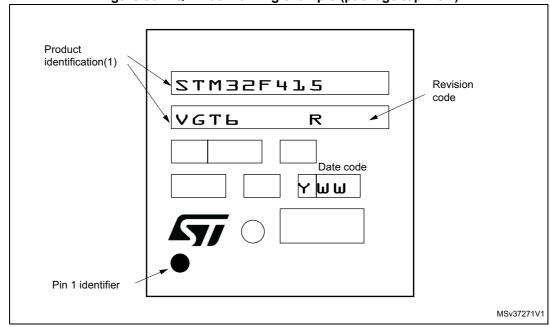


Figure 83. LQFP100 marking example (package top view)

^{1.} Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering Samples to run qualification activity.

6.4 LQFP144 package information

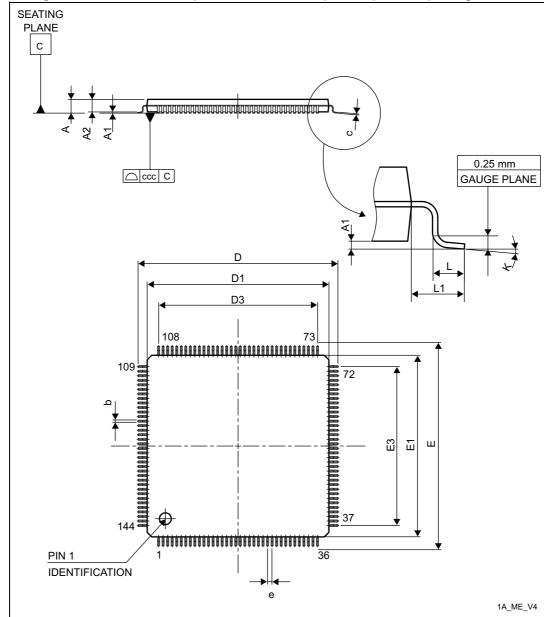


Figure 84. LQFP144 - 144-pin, 20 x 20 mm low-profile quad flat package outline

1. Drawing is not to scale.

Table 94. LQFP144 - 144-pin, 20 x 20 mm low-profile quad flat package mechanical data

Cumbal		millimeters			inches ⁽¹⁾	
Symbol	Min	Тур	Max	Min	Тур	Max
А	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
С	0.090	-	0.200	0.0035	-	0.0079
D	21.800	22.000	22.200	0.8583	0.8661	0.874
D1	19.800	20.000	20.200	0.7795	0.7874	0.7953
D3	-	17.500	-	-	0.689	-
E	21.800	22.000	22.200	0.8583	0.8661	0.8740
E1	19.800	20.000	20.200	0.7795	0.7874	0.7953
E3	-	17.500	-	-	0.6890	-
е	-	0.500	-	-	0.0197	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
k	0°	3.5°	7°	0°	3.5°	7°
ccc	-	-	0.080	-	-	0.0031

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.

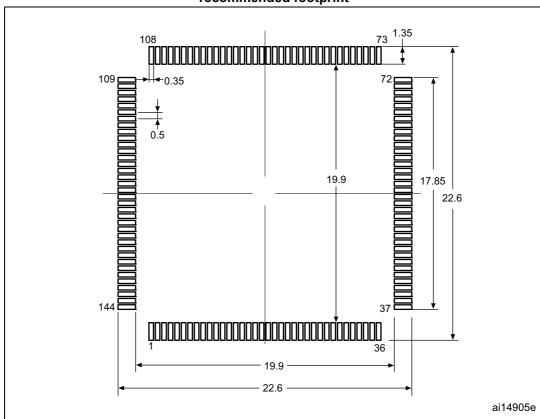


Figure 85. LQFP144 - 144-pin,20 x 20 mm low-profile quad flat package recommended footprint

- 1. Drawing is not to scale.
- 2. Dimensions are in millimeters.

Device marking for LQPF144

The following figure gives an example of topside marking and pin 1 position identifier location.

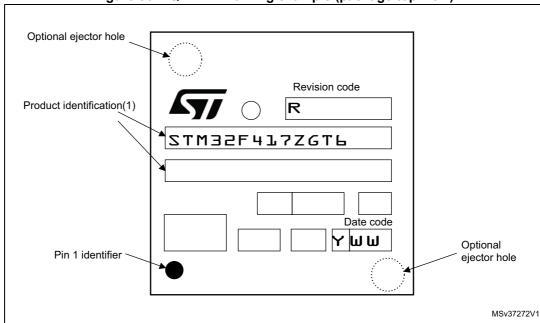
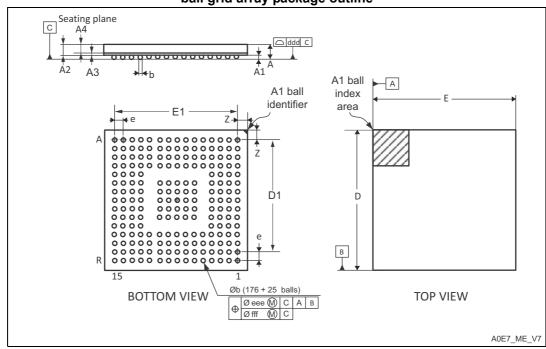


Figure 86. LQFP144 marking example (package top view)

1. Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering Samples to run qualification activity.

6.5 UFBGA176+25 package information

Figure 87. UFBGA176+25 - 201-ball, 10 x 10 mm, 0.65 mm pitch, ultra fine pitch ball grid array package outline



1. Drawing is not to scale.

Table 95. UFBGA176+25 - 201-ball, 10 × 10 × 0.65 mm pitch, ultra thin fine pitch ball grid array mechanical data

		ban gria	array meen	inicai data		
Cumbal	millimeters			inches ⁽¹⁾		
Symbol	Min	Тур	Max	Min	Тур	Max
Α	-	-	0.600	-	-	0.0236
A1	-	-	0.110	-	-	0.0043
A2	-	0.450	-	-	0.0177	-
A3	-	0.130	-	-	0.0051	0.0094
A4	-	0.320	-	-	0.0126	-
b	0.240	0.290	0.340	0.0094	0.0114	0.0134
D	9.850	10.000	10.150	0.3878	0.3937	0.3996
D1	-	9.100	-	-	0.3583	-
E	9.850	10.000	10.150	0.3878	0.3937	0.3996
E1	-	9.100	-	-	0.3583	-
е	-	0.650	-	-	0.0256	-
Z	-	0.450	-	-	0.0177	-
ddd	-	-	0.080	-	-	0.0031

Table 95. UFBGA176+25 - 201-ball, 10 × 10 × 0.65 mm pitch, ultra thin fine pitch ball grid array mechanical data (continued)

Symbol	millimeters				inches ⁽¹⁾	
Symbol	Min	Тур	Max	Min	Тур	Max
eee	-	-	0.150	-	-	0.0059
fff	-	-	0.050	-	-	0.0020

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 88. UFBGA176+25 - 201-ball, 10 x 10 mm, 0.65 mm pitch, ultra fine pitch ball grid array recommended footprint

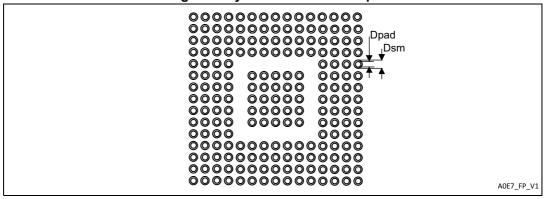


Table 96. UFBGA176+2 recommended PCB design rules (0.65 mm pitch BGA)

Dimension	Recommended values		
Pitch	0.65		
Dpad	0.300 mm		
	0.400 mm typ. (depends on the soldermask registration tolerance)		

Note: Non solder mask defined (NSMD) pads are recommended.

4 to 6 mils solder paste screen printing process.

Stencil opening is 0.300 mm.

Stencil thickness is between 0.100 mm and 0.125 mm.

Pad trace width is 0.100 mm.

Device marking for UFBGA176+25

The following figure gives an example of topside marking and ball A 1 position identifier location.

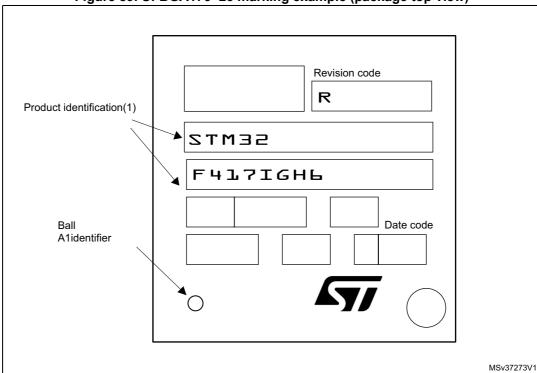


Figure 89. UFBGA176+25 marking example (package top view)

^{1.} Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering Samples to run qualification activity.

6.6 LQFP176 package information

C Seating plane

O.25 mm
gauge plane

PIN 1
IDENTIFICATION

D

TT_ME_V2

Figure 90. LQFP176 - 176-pin, 24 x 24 mm low profile quad flat package outline

1. Drawing is not to scale.

Table 97. LQFP176 - 176-pin, 24 x 24 mm low profile quad flat package mechanical data

Symbol	millimeters			inches ⁽¹⁾		
Symbol	Min	Тур	Max	Min	Тур	Max
А	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	-	1.450	0.0531	-	0.0571
b	0.170	-	0.270	0.0067	-	0.0106
С	0.090	-	0.200	0.0035	-	0.0079
D	23.900	-	24.100	0.9409	-	0.9488
HD	25.900	-	26.100	1.0197	-	1.0276

Table 97. LQFP176 - 176-pin, 24 x 24 mm low profile quad flat package mechanical data (continued)

Correla a l	millimeters			inches ⁽¹⁾		
Symbol	Min	Тур	Max	Min	Тур	Max
ZD	-	1.250	-	-	0.0492	-
Е	23.900	-	24.100	0.9409	-	0.9488
HE	25.900	-	26.100	1.0197	-	1.0276
ZE	-	1.250	-	-	0.0492	-
е	-	0.500	-	-	0.0197	-
L ⁽²⁾	0.450	-	0.750	0.0177	-	0.0295
L1	-	1.000	-	-	0.0394	-
k	0°	-	7°	0°	-	7°
ccc	-	-	0.080	-	-	0.0031

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.

^{2.} L dimension is measured at gauge plane at 0.25 mm above the seating plane.

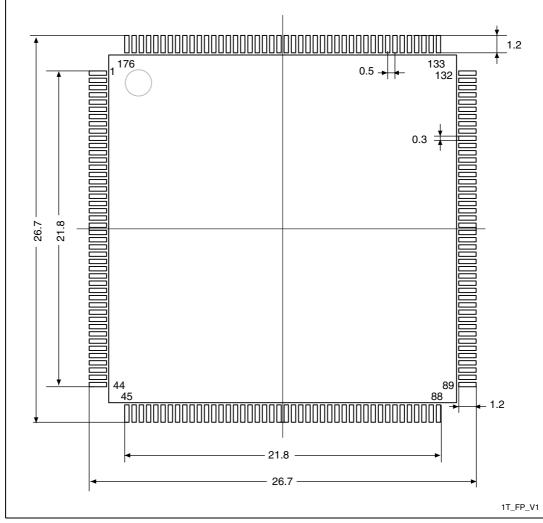


Figure 91. LQFP176 - 176-pin, 24 x 24 mm low profile quad flat recommended footprint

1. Dimensions are expressed in millimeters.

Device marking for LQFP176

The following figure gives an example of topside marking and pin 1 position identifier location.

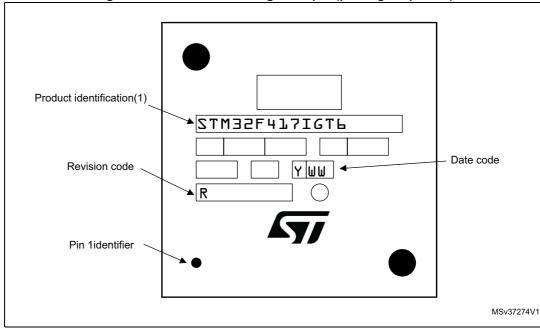


Figure 92. LQFP176 marking example (package top view)

Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet
qualified and therefore not yet ready to be used in production and any consequences deriving from such
usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering
samples in production. ST Quality has to be contacted prior to any decision to use these Engineering
Samples to run qualification activity.

6.7 Thermal characteristics

The maximum chip-junction temperature, T_J max, in degrees Celsius, may be calculated using the following equation:

 $T_J \max = T_A \max + (P_D \max x \Theta_{JA})$

Where:

- T_A max is the maximum ambient temperature in °C,
- Θ_{JA} is the package junction-to-ambient thermal resistance, in °C/W,
- P_D max is the sum of P_{INT} max and $P_{I/O}$ max (P_D max = P_{INT} max + $P_{I/O}$ max),
- P_{INT} max is the product of I_{DD} and V_{DD}, expressed in Watts. This is the maximum chip internal power.

P_{I/O} max represents the maximum power dissipation on output pins where:

$$P_{I/O} \max = \sum (V_{OL} \times I_{OL}) + \sum ((V_{DD} - V_{OH}) \times I_{OH}),$$

taking into account the actual V_{OL} / I_{OL} and V_{OH} / I_{OH} of the I/Os at low and high level in the application.

Symbol	Parameter	Value	Unit
	Thermal resistance junction-ambient LQFP64 - 10 × 10 mm / 0.5 mm pitch	46	
	Thermal resistance junction-ambient LQFP100 - 14 × 14 mm / 0.5 mm pitch	43	
Θ_{JA}	Thermal resistance junction-ambient LQFP144 - 20 × 20 mm / 0.5 mm pitch	40	°C/W
	Thermal resistance junction-ambient LQFP176 - 24 × 24 mm / 0.5 mm pitch	38	C/VV
	Thermal resistance junction-ambient UFBGA176 - 10× 10 mm / 0.65 mm pitch	39	
	Thermal resistance junction-ambient WLCSP90 - 0.400 mm pitch	38.1	

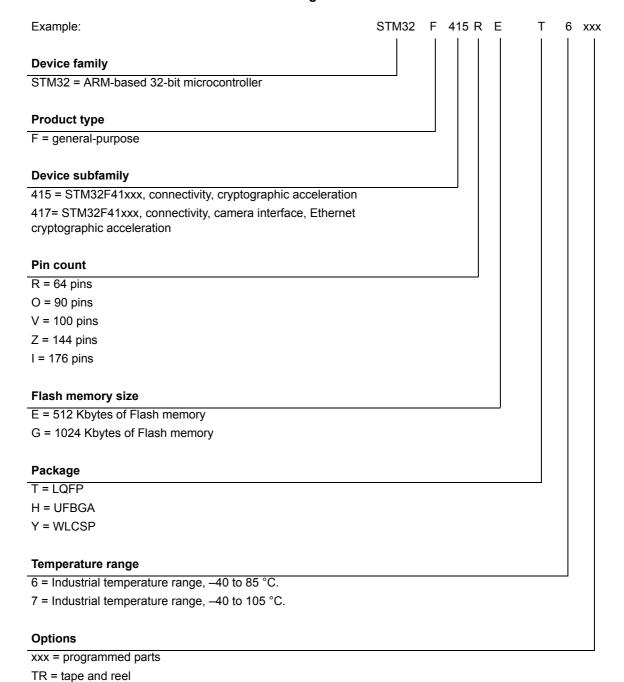
Table 98. Package thermal characteristics

Reference document

JESD51-2 Integrated Circuits Thermal Test Method Environment Conditions - Natural Convection (Still Air). Available from www.jedec.org.

7 Part numbering

Table 99. Ordering information scheme



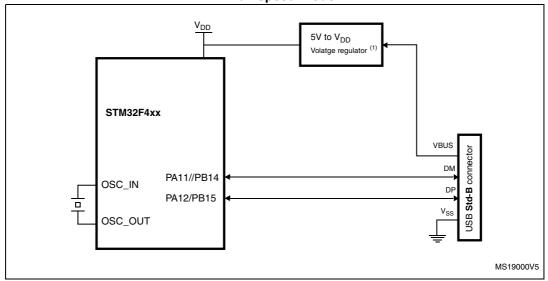
For a list of available options (speed, package, etc.) or for further information on any aspect

of this device, please contact your nearest ST sales office.

Appendix A Application block diagrams

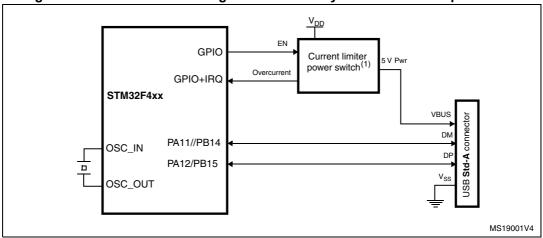
A.1 USB OTG full speed (FS) interface solutions

Figure 93. USB controller configured as peripheral-only and used in Full speed mode



- 1. External voltage regulator only needed when building a $V_{\mbox{\footnotesize{BUS}}}$ powered device.
- 2. The same application can be developed using the OTG HS in FS mode to achieve enhanced performance thanks to the large Rx/Tx FIFO and to a dedicated DMA controller.

Figure 94. USB controller configured as host-only and used in full speed mode



- The current limiter is required only if the application has to support a V_{BUS} powered device. A basic power switch can be used if 5 V are available on the application board.
- 2. The same application can be developed using the OTG HS in FS mode to achieve enhanced performance thanks to the large Rx/Tx FIFO and to a dedicated DMA controller.

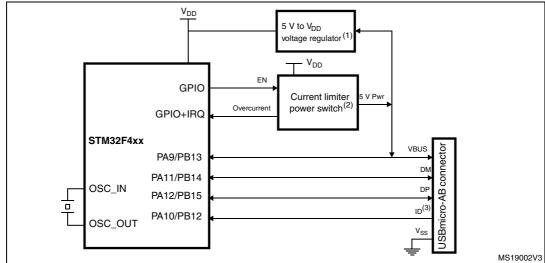


Figure 95. USB controller configured in dual mode and used in full speed mode

- 1. External voltage regulator only needed when building a $V_{\mbox{\scriptsize BUS}}$ powered device.
- 2. The current limiter is required only if the application has to support a V_{BUS} powered device. A basic power switch can be used if 5 V are available on the application board.
- 3. The ID pin is required in dual role only.
- 4. The same application can be developed using the OTG HS in FS mode to achieve enhanced performance thanks to the large Rx/Tx FIFO and to a dedicated DMA controller.

A.2 USB OTG high speed (HS) interface solutions

STM32F4xx · DM not connected USB HS OTG Ctr DF ULPI_CLK DM ULPI_D[7:0] $ID^{(2)}$ USB ULPI_DIR connector ULPI V_{BUS} ULPI_STP V_{SS} ULPI_NXT High speed OTG PHY 24 or 26 MHz XT⁽¹⁾

Figure 96. USB controller configured as peripheral, host, or dual-mode and used in high speed mode

MCO1 or MCO2

MS19005V2

It is possible to use MCO1 or MCO2 to save a crystal. It is however not mandatory to clock the STM32F41xxx with a 24 or 26 MHz crystal when using USB HS. The above figure only shows an example of a possible connection.

^{2.} The ID pin is required in dual role only.

A.3 Ethernet interface solutions

STM32 MII_TX_CLK MII_TX_EN MII_TXD[3:0] MCU Ethernet Ethernet MAC 10/100 PHY 10/100 MII_CRS MII MII_COL = 15 pins HCLK(1)_ MII_RX_CLK MII + MDC MII_RXD[3:0] MII_RX_DV = 17 pins IEEE1588 PTP Timer MII_RX_ER input trigge Timestamp comparator TIM2 MDIO MDC PPS_OUT(2) **HCLK** PLL **XTAL** osc 25 MHz PHY_CLK 25 MHz MCO1/MCO2 XT1 MS19968V1

Figure 97. MII mode using a 25 MHz crystal

- 1. f_{HCLK} must be greater than 25 MHz.
- 2. Pulse per second when using IEEE1588 PTP optional signal.

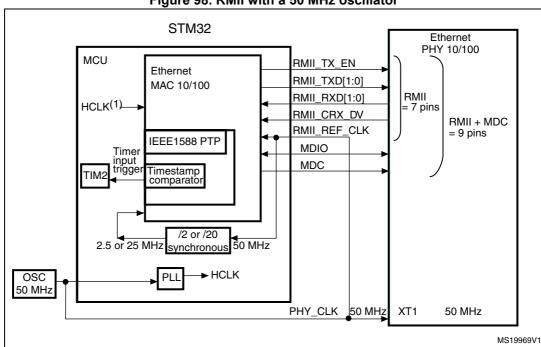


Figure 98. RMII with a 50 MHz oscillator

1. f_{HCLK} must be greater than 25 MHz.

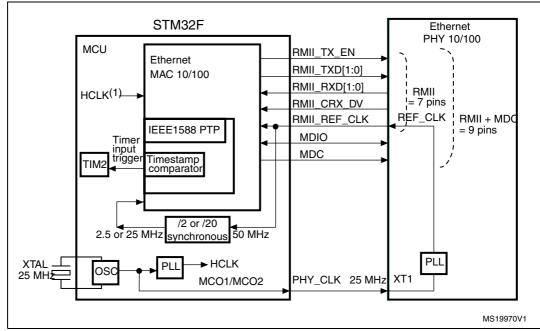


Figure 99. RMII with a 25 MHz crystal and PHY with PLL

- 1. f_{HCLK} must be greater than 25 MHz.
- 2. The 25 MHz (PHY_CLK) must be derived directly from the HSE oscillator, before the PLL block.

8 Revision history

Table 100. Document revision history

Date	Revision	Changes
15-Sep-2011	1	Initial release.
24-Jan-2012	2	Added WLCSP90 package on cover page. Renamed USART4 and USART5 into UART4 and UART5, respectively. Updated number of USB OTG HS and FS in Table 2: STM32F415xx and STM32F417xx: features and peripheral counts. Updated Figure 3: Compatible board design between STM32F10xx/STM32F2/STM32F41xxx for LQFP144 package and Figure 4: Compatible board design between STM32F2 and STM32F41xxx for LQFP176 and BGA176 packages, and removed note 1 and 2. Updated Section 2.2.9: Flexible static memory controller (FSMC). Modified I/Os used to reprogram the Flash memory for CAN2 and USB OTG FS in Section 2.2.13: Boot modes. Updated note in Section 2.2.14: Power supply schemes. PDR_ON no more available on LQFP100 package. Updated Section 2.2.16: Voltage regulator. Updated condition to obtain a minimum supply voltage of 1.7 V in the whole document. Renamed USART4/5 to UART4/5 and added LIN and IrDA feature for UART4 and UART5 in Table 5: USART feature comparison. Removed support of I2C for OTG PHY in Section 2.2.30: Universal serial bus on-the-go full-speed (OTG_FS). Added Table 6: Legend/abbreviations used in the pinout table. Table 7: STM32F41xxx pin and ball definitions: replaced V _{SS_3} , V _{SS_4} , and V _{SS_8} by V _{SS} ; reformatted Table 7: STM32F41xxx pin and ball definitions to better highlight I/O structure, and alternate functions versus additional functions; signal corresponding to LQFP100 pin 99 changed from PDR_ON to V _{SS} ; EVENTOUT added in the list of alternate functions for PD11 and PD12, respectively; PH10 alternate function for PF10; FSMC_CLE and FSMC_ALE added as alternate function for PF10; FSMC_CLE and FSMC_ALE added as alternate function TIM15_CH1_ETR renamed TIM5_CH1; updated PA4 and PA5 I/O structure to TTa. Removed OTG_HS_SCL, OTG_HS_SDA, OTG_FS_INTN in Table 7: STM32F41xxx pin and ball definitions and Table 9: Alternate function mapping. Changed TCM data RAM to CCM data RAM in Figure 18: STM32F41xxx memory map. Added I _{VDD} and I _{VSS} maximum values in Table 12: Current characteristics. Added Note 1

Table 100. Document revision history (continued)

Date	Revision	Changes
24-Jan-2012	2 (continued)	Added V ₁₂ in Table 19: Embedded reset and power control block characteristics. Updated Table 21: Typical and maximum current consumption in Run mode, code with data processing running from Flash memory (ART accelerator disabled) and Table 20: Typical and maximum current consumption in Run mode, code with data processing running from Flash memory (ART accelerator enabled) or RAM. Added Figure , Figure 25, Figure 26, and Figure 27. Updated Table 22: Typical and maximum current consumption in Sleep mode and removed Note 1. Updated Table 23: Typical and maximum current consumptions in Stop mode and Table 24: Typical and maximum current consumptions in Standby mode, Table 25: Typical and maximum current consumptions in VBAT mode, and Table 27: Switching output I/O current consumption. Section: On-chip peripheral current consumption: modified conditions, and updated Table 28: Peripheral current consumption and Note 2. Changed f _{HSE_ext} to 50 MHz and t _{r(HSE)} /t _{f(HSE)} maximum value in Table 30: High-speed external user clock characteristics. Added C _{in(LSE)} in Table 31: Low-speed external user clock characteristics. Updated maximum PLL input clock frequency, removed related note, and deleted jitter for MCO for RMII Ethernet typical value in Table 36: Main PLL characteristics. Updated maximum PLL12S input clock frequency and removed related note in Table 37: PLL12S (audio PLL) characteristics. Updated Section: Flash memory to specify that the devices are shipped to customers with the Flash memory erased. Updated Table 39: Flash memory characteristics, and added t _{ME} in Table 40: Flash memory programming. Updated Table 43: EMS characteristics, and Table 44: EMI characteristics. Updated Table 56: I2S dynamic characteristics Updated Table 43: EMS characteristics of TIMx connected to the APB1 domain and Table 52: Characteristics of TIMx connected to the APB1 domain and Table 53: Characteristics of TIMx connected to the APB2 domain. Updated Table 65: Dynamic characteristics.

Table 100. Document revision history (continued)

Date	Revision	Changes
24-Jan-2012	2 (continued)	Updated Table 61: USB HS clock timing parameters Updated Table 67: ADC characteristics. Updated Table 68: ADC accuracy at fADC = 30 MHz. Updated Note 1 in Table 74: DAC characteristics. Section 5.3.26: FSMC characteristics: updated Table 75 to Table 86, changed C _L value to 30 pF, and modified FSMC configuration for asynchronous timings and waveforms. Updated Figure 59: Synchronous multiplexed PSRAM write timings. Updated Table 98: Package thermal characteristics. Appendix A.1: USB OTG full speed (FS) interface solutions: modified Figure 93: USB controller configured as peripheral-only and used in Full speed mode added Note 2, updated Figure 94: USB controller configured as host-only and used in full speed mode and added Note 2, changed Figure 95: USB controller configured in dual mode and used in full speed mode and added Note 3. Appendix A.2: USB OTG high speed (HS) interface solutions: removed figures USB OTG HS device-only connection in FS mode and USB OTG HS host-only connection in FS mode, and updated Figure 96: USB controller configured as peripheral, host, or dual-mode and used in high speed mode and added Note 2. Added Appendix A.3: Ethernet interface solutions.

Table 100. Document revision history (continued)

Date	Revision	Changes
Date 31-May-2012		, ,
		Nested vectored interrupt controller (NVIC)
		Updated case of Regulator ON/internal reset ON, Regulator
		'
		Added Note 1 below Figure 16: STM32F41xxx UFBGA176 ballout.
		Added Note 1 below Figure 17: STM32F41xxx WLCSP90 ballout.
31-May-2012	3	·
		Removed I2S2_WS on PB6/AF5 in Table 9: Alternate function
		Replaced JTRST by NJTRST, removed ETH_RMII_TX_CLK, and
		Added Table 10: STM32F41x register boundary addresses.
		Updated Figure 18: STM32F41xxx memory map.
		Updated V _{DDA} and V _{REF+} decoupling capacitor in <i>Figure 21: Power supply scheme</i> .
		Added power dissipation maximum value for WLCSP90 in <i>Table 14: General operating conditions</i> .
		Updated V _{POR/PDR} in <i>Table 19: Embedded reset and power control block characteristics</i> .
		Updated notes in Table 21: Typical and maximum current consumption in Run mode, code with data processing running from Flash memory (ART accelerator disabled), Table 20: Typical and maximum current consumption in Run mode, code with data processing running from
		Flash memory (ART accelerator enabled) or RAM, and Table 22: Typical and maximum current consumption in Sleep mode.
		Updated maximum current consumption at T _A = 25 °n <i>Table 23: Typical and maximum current consumptions in Stop mode.</i>

Table 100. Document revision history (continued)

Date	Revision	Changes
31-May-2012	3 (continued)	Removed f _{HSE_ext} typical value in <i>Table 30: High-speed external user clock characteristics</i> . Updated <i>Table 32: HSE 4-26 MHz oscillator characteristics</i> and <i>Table 33: LSE oscillator characteristics</i> (fLSE = 32.768 kHz). Added f _{PLL48_OUT} maximum value in <i>Table 36: Main PLL characteristics</i> . Modified equation 1 and 2 in <i>Section 5.3.11: PLL spread spectrum clock generation</i> (SSCG) characteristics. Updated <i>Table 39: Flash memory characteristics, Table 40: Flash memory programming,</i> and <i>Table 41: Flash memory programming with VPP</i> . Updated <i>Section : Output driving current. Table 56: I²C characteristics: Note 4</i> updated and applied to t _{h(SDA)} in Fast mode, and removed note 4 related to t _{h(SDA)} minimum value. Updated <i>Table 67: ADC characteristics.</i> Updated note concerning ADC accuracy vs. negative injection current below <i>Table 68: ADC accuracy at fADC = 30 MHz.</i> Added WLCSP90 thermal resistance in <i>Table 98: Package thermal characteristics.</i> Updated <i>Table 90: WLCSP90 - 4.223 x 3.969 mm, 0.400 mm pitch wafer level chip scale package mechanical data.</i> Updated <i>Figure 87: UFBGA176+25 - 201-ball, 10 x 10 mm, 0.65 mm pitch, ultra fine pitch ball grid array package outline and <i>Table 95: UFBGA176+25 - 201-ball, 10 x 10 mm, 0.65 mm pitch, ultra fine pitch ball grid array package outline and Table 95: UFBGA176+25 - 201-ball, 10 x 10 x 0.65 mm pitch, ultra thin fine pitch ball grid array mechanical data. Added Figure 91: LQFP176 - 176-pin, 24 x 24 mm low profile quad flat recommended footprint. Removed 256 and 768 Kbyte Flash memory density from <i>Table 99: Ordering information scheme</i>.</i></i>

Table 100. Document revision history (continued)

Date	Revision	Changes
		Modified Note 1 below Table 2: STM32F415xx and STM32F417xx:
		features and peripheral counts.
		Updated <i>Figure 4</i> title.
		Updated Note 3 below Figure 21: Power supply scheme.
		Changed simplex mode into half-duplex mode in Section 2.2.25: Inter- integrated sound (I2S).
		Replaced DAC1_OUT and DAC2_OUT by DAC_OUT1 and DAC_OUT2, respectively.
		Updated pin 36 signal in Figure 15: STM32F41xxx LQFP176 pinout.
		Changed pin number from F8 to D4 for PA13 pin in <i>Table 7:</i>
		STM32F41xxx pin and ball definitions.
		Replaced TIM2_CH1/TIM2_ETR by TIM2_CH1_ETR for PA0 and PA5 pins in <i>Table 9: Alternate function mapping</i> .
		Changed system memory into System memory + OTP in <i>Figure 18: STM32F41xxx memory map</i> .
		Added Note 1 below Table 16: VCAP_1/VCAP_2 operating conditions.
		Updated I _{DDA} description in <i>Table 74: DAC characteristics</i> .
		Removed PA9/PB13 connection to VBUS in Figure 93: USB controller configured as peripheral-only and used in Full speed mode and Figure 94: USB controller configured as host-only and used in full speed mode.
		Updated SPI throughput on front page and Section 2.2.24: Serial
		peripheral interface (SPI)
04-Jun-2013	4	Updated operating voltages in <i>Table 2: STM32F415xx and STM32F417xx: features and peripheral counts</i>
		Updated note in Section 2.2.14: Power supply schemes
		Updated Section 2.2.15: Power supply supervisor
		Updated "Regulator ON" paragraph in Section 2.2.16: Voltage regulator
		Removed note in Section 2.2.19: Low-power modes
		Corrected wrong reference manual in Section 2.2.28: Ethernet MAC interface with dedicated DMA and IEEE 1588 support
		Updated Table 15: Limitations depending on the operating power supply range
		Updated Table 24: Typical and maximum current consumptions in Standby mode
		Updated Table 25: Typical and maximum current consumptions in VBAT mode
		Updated Table 37: PLLI2S (audio PLL) characteristics
		Updated Table 44: EMI characteristics
		Updated Table 49: Output voltage characteristics
		Updated Table 51: NRST pin characteristics
		Updated Table 55: SPI dynamic characteristics
		Updated Table 56: I2S dynamic characteristics
		Deleted Table 59
		Updated Table 62: ULPI timing
		Updated Figure 46: Ethernet SMI timing diagram

Table 100. Document revision history (continued)

Data	Revision	Changes
Date	Revision	Changes
04-Jun-2013	4 (continued)	Updated Figure 87: UFBGA176+25 - 201-ball, 10 x 10 mm, 0.65 mm pitch, ultra fine pitch ball grid array package outline Updated Table 95: UFBGA176+25 - 201-ball, 10 x 10 × 0.65 mm pitch, ultra thin fine pitch ball grid array mechanical data Updated Figure 5: STM32F41xxx block diagram Updated Section 2: Description Updated fotnote (3) in Table 2: STM32F415xx and STM32F417xx: features and peripheral counts Updated Figure 3: Compatible board design between STM32F10xx/STM32F2/STM32F41xxx for LQFP144 package Updated Figure 4: Compatible board design between STM32F2 and STM32F41xxx for LQFP176 and BGA176 packages Updated Section 2.2.16: Power supply schemes Updated Section 2.2.16: Voltage regulator, including figures. Updated Section 2.2.16: Voltage regulator, including footnote (2). Updated Table 14: General operating conditions, including footnote (3). Updated Table 15: Limitations depending on the operating power supply range, including footnote (3). Updated footnote (1) in Table 67: ADC characteristics. Updated footnote (2) in Table 68: ADC accuracy at fADC = 30 MHz. Updated footnote (1) in Table 74: DAC characteristics. Updated Figure 9: Regulator OFF. Updated Figure 9: Regulator ON/OFF and internal reset ON/OFF availability. Updated footnote (2) of Figure 21: Power supply scheme. Replaced respectively "12S3S_WS" by "12S3_WS", "12S3S_CK" by "12S3_CK" by "12S3_CK" and "FSMC_BLN1" by "FSMC_BL1" in Table 9: Alternate function mapping. Added "EVENTOUT" as alternate function "AF15" for pin PC13, PC14, PC15, PH0, PH1, Pl8 in Table 9: Alternate function mapping Replaced "DCMI_12" by "DCMI_D12" in Table 7: STM32F41xxx pin and ball definitions. Removed the following sentence from Section : 12C interface characteristics: "Unless otherwise specified, the parameters given in Table 56 are derived from tests performed under the ambient temperature, fpCLK1 frequency and VDD supply voltage conditions summarized in Table 14.". In

Table 100. Document revision history (continued)

Date	Revision	Changes
Date 04-Jun-2013		Updated Figure 6: Multi-AHB matrix. Updated Figure 7: Power supply supervisor interconnection with internal reset OFF Changed 1.2 V to V ₁₂ in Section : Regulator OFF Updated LQFP176 pin 48. Updated Section 1: Introduction. Updated Section 2: Description. Updated operating voltage in Table 2: STM32F415xx and STM32F417xx: features and peripheral counts. Updated Note 1. Updated Section 2.2.15: Power supply supervisor. Updated Section 2.2.16: Voltage regulator. Updated Figure 9: Regulator OFF. Updated Table 3: Regulator ON/OFF and internal reset ON/OFF availability. Updated Section 2.2.19: Low-power modes. Updated Section 2.2.20: VBAT operation. Updated Section 2.2.22: Inter-integrated circuit interface (I²C) . Updated pin 48 in Figure 15: STM32F41xxx LQFP176 pinout. Updated Table 6: Legend/abbreviations used in the pinout table. Updated Table 14: General operating conditions. Updated Table 15: Limitations depending on the operating power supply range. Updated Section 5.3.7: Wakeup time from low-power mode.
		Updated Table 3: Regulator ON/OFF and internal reset ON/OFF availability. Updated Section 2.2.19: Low-power modes. Updated Section 2.2.20: VBAT operation.
		Updated pin 48 in Figure 15: STM32F41xxx LQFP176 pinout. Updated Table 6: Legend/abbreviations used in the pinout table. Updated Table 7: STM32F41xxx pin and ball definitions.
04-Jun-2013	13 4 (continued)	Updated Table 15: Limitations depending on the operating power supply range. Updated Section 5.3.7: Wakeup time from low-power mode.
		Updated Table 34: HSI oscillator characteristics. Updated Section 5.3.15: I/O current injection characteristics. Updated Table 48: I/O static characteristics. Updated Table 51: NRST pin characteristics.
		Updated <i>Table 56: I²C characteristics</i> . Updated <i>Figure 39: I²C bus AC waveforms and measurement circuit</i> . Updated <i>Section 5.3.19: Communications interfaces</i> .
		Updated Table 67: ADC characteristics. Added Table 70: Temperature sensor calibration values. Added Table 73: Internal reference voltage calibration values. Updated Section 5.3.26: FSMC characteristics.
		Updated Section 5.3.28: SD/SDIO MMC card host interface (SDIO) characteristics. Updated Table 23: Typical and maximum current consumptions in Stop
		mode. Updated Section : SPI interface characteristics included Table 55. Updated Section : I2S interface characteristics included Table 56.
		Updated Table 64: Dynamic characteristics: Eternity MAC signals for SMI. Updated Table 66: Dynamic characteristics: Ethernet MAC signals for MII.

Table 100. Document revision history (continued)

Updated Table 64: Dynamic characteristics: Eternity MAC signals for SMI. Updated Table 66: Dynamic characteristics: Ethernet MAC signals for MII. Updated Table 79: Synchronous multiplexed NOR/PSRAM read timings. Updated Table 80: Synchronous multiplexed PSRAM write timings. Updated Table 81: Synchronous non-multiplexed NOR/PSRAM read timings. Updated Table 82: Synchronous non-multiplexed PSRAM write timings. Updated Table 82: Synchronous non-multiplexed PSRAM write timings. Updated Section 5.3.27: Camera interface (DCMI) timing specifications including Table 87: DCMI characteristics and addition of Figure 72: DCMI timing diagram. Updated Section 5.3.28: SD/SDIO MMC card host interface (SDIO) characteristics including Table 88.	Date	Revision	Changes
		4	Updated Table 64: Dynamic characteristics: Eternity MAC signals for SMI. Updated Table 66: Dynamic characteristics: Ethernet MAC signals for MII. Updated Table 79: Synchronous multiplexed NOR/PSRAM read timings. Updated Table 80: Synchronous multiplexed PSRAM write timings. Updated Table 81: Synchronous non-multiplexed NOR/PSRAM read timings. Updated Table 82: Synchronous non-multiplexed PSRAM write timings. Updated Table 82: Synchronous non-multiplexed PSRAM write timings. Updated Section 5.3.27: Camera interface (DCMI) timing specifications including Table 87: DCMI characteristics and addition of Figure 72: DCMI timing diagram. Updated Section 5.3.28: SD/SDIO MMC card host interface (SDIO)

Table 100. Document revision history (continued)

Table 100. Document revision history (continued)			
Date	Kevision	Gnanges	
Date	Revision 5	Replace Cortex-M4F by Cortex-M4 with FPU throughout the document. Updated Section: Regulator OFF and Table 3: Regulator ON/OFF and Internal reset ON/OFF availability for LQFP176. Updated Figure 15: STM32F41xxx LQFP176 pinout and Table 7: STM32F41xxx pin and ball definitions. Updated Figure 16: Multi-AHB matrix. Added note 1 below Figure 12: STM32F41xxx LQFP64 pinout, Figure 13: STM32F41xxx LQFP100 pinout, Figure 14: STM32F41xxx LQFP144 pinout and Figure 15: STM32F41xxx LQFP176 pinout. Updated IyDD and IySS in Table 12: Current characteristics. Updated PLS[2:0]=101 (falling edge) configuration in Table 19: Embedded reset and power control block characteristics. Added Section: Additional current consumption. Updated Section: On-chip peripheral current consumption. Updated Table 29: Low-power mode wakeup timings. Updated Table 32: HSE 4-26 MHz oscillator characteristics and Table 33: LSE oscillator characteristics (fLSE = 32.768 kHz). Changed condition related to VESD(CDM) in Table 45: ESD absolute maximum ratings. Updated Table 47: I/O current injection susceptibility, Table 48: I/O static characteristics, Table 49: Output voltage characteristics conditions, Table 50: I/O AC characteristics and Figure 37: I/O AC characteristics definition. Updated Section: I2C interface characteristics. Remove note 3 in Table 69: Temperature sensor characteristics. Updated Figure 72: DCMI timing diagram. Modified Figure 75: WLCSP90 - 4.223 x 3.969 mm, 0.400 mm pitch wafer level chip scale package mechanical data. Added Figure 76: WLCSP90 - 4.223 x 3.969 mm, 0.400 mm pitch wafer level chip scale package mechanical data. Added Figure 76: WLCSP90 - 4.223 x 3.969 mm, 0.400 mm pitch wafer level chip scale package mechanical data. Added Figure 78: LQFP64 - 64-pin 10 x 10 mm low-profile quad flat package mechanical data. Updated Figure 78: LQFP64 - 64-pin, 10 x 10 mm low-profile quad flat package mechanical data. Added Figure 88: UFBGA176+25 - 201-ball, 10 x 10 mm, 0.65 mm pitch, ultra fine pitch ball grid array mechanic	



Table 100. Document revision history (continued)

Date	Revision	Changes
22-Oct-2015	6	In the whole document, updated notes related to values guaranteed by design or by characterization. Updated <i>Table 34: HSI oscillator characteristics</i> . Changed f _{VCO_OUT} minimum value and VCO freq to 100 MHz in <i>Table 36: Main PLL characteristics</i> and <i>Table 37: PLLI2S (audio PLL) characteristics</i> . Updated <i>Figure 39: SPI timing diagram - slave mode and CPHA = 0</i> . Updated <i>Figure 53: 12-bit buffered /non-buffered DAC</i> . Removed note 1 related to better performance using a restricted V _{DD} range in <i>Table 68: ADC accuracy at fADC = 30 MHz</i> . Upated <i>Figure 84: LQFP144 - 144-pin, 20 x 20 mm low-profile quad flat package outline</i> . Updated <i>Figure 87: UFBGA176+25 - 201-ball, 10 x 10 mm, 0.65 mm pitch, ultra fine pitch ball grid array package outline</i> and <i>Table 95: UFBGA176+25 - 201-ball, 10 x 10 x 0.65 mm pitch, ultra thin fine pitch ball grid array mechanical data</i> .

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